

# High Speed Buffer Amplifiers

## ADLH0033G/ADLH0033CG

## **FEATURES**

2nd Source—Replaces All LH0033G Series Wide Bandwidth—dc to 100MHz High Slew Rate—1500V/ $\mu$ s Operates on Single or Dual Power Supplies Operation Guaranteed –55°C to +125°C (ADLH0033G) High  $10^{11}\,\Omega$  Input Impedance

## **APPLICATIONS**

High-Speed Line Drivers
Video Impedance Transformation
High-Speed A/D Input Buffers
Nuclear Instrumentation Amplifiers
Coaxial Cable Drive

### GENERAL DESCRIPTION

The ADLH0033G and ADLH0033CG are superhigh speed (1500V/ $\mu$ s slew rate) and high input impedance (10<sup>11</sup>  $\Omega$ ) buffer amplifiers, designed to replace all LH0033 series amplifiers in applications such as high-speed line drivers or as high impedance buffers for fast A/D converters and comparators.

The ADLH0033G is guaranteed over the temperature range of -55°C to +125°C, while the commercial grade ADLH0033CG is guaranteed over the range of -25°C to +85°C.

Guaranteed operation over temperature of the ADLH0033G is achieved by using specially selected junction FET's and the latest state-of-the-art laser trimming techniques. They are available in the industry standard 12 pin TO-8 metal can.

## **OPERATION WITHIN AN OP AMP LOOP**

When using the ADLH0033G/ADLH0033CG as a current booster or isolation buffer with op amps such as LH0032, 118, 741, etc., an isolation resistor of at least  $47\Omega$  must be

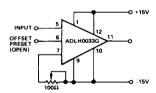
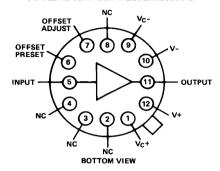


Figure 1. Offset Adjustment

## ADLH0033G/ADLH0033CG OUTLINE AND PIN DESIGNATIONS



used between the op amp's output and the input of the ADLH0033G.

### HEAT SINKING

To assure maximum output drive capability of the ADLH0033G/ADLH0033CG over temperature, heat sinks should be used. The cases are electrically isolated from the circuit and thus may be connected to system grounds.

### POWER SUPPLY BYPASSING

To prevent oscillation, power supply bypassing is recommended. Use low-inductance ceramic disc caps, keeping lead lengths as short as possible (1/4" to 1/2" max from device package), connected between ground plane and each supply lead. Use one or two  $0.1\mu F$  caps in parallel with a  $4.7\mu F$  tantalum for best results.

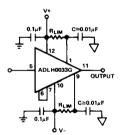


Figure 2. Short Circuit Protection Using Current Limiting Resistors (RLIM)

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## SPECIFICATIONS

ADLH0033G ADLH0033CG

ABSOLUTE MAXIMUM RATINGS

Supply Voltage (V+ - V-)

Maximum Power Dissipation (see curves) Maximum Junction Temperature

Input Voltage Continuous Output Current

Peak Output Current

Operating Temperature ADLH0033G ADLH0033CG

Storage Temperature Range Lead Temperature (Soldering, 10 sec.) 40V 1.5W

175°C Equal to Supplies

±100mA

±250mA

-55°C to +125°C -25°C to +85°C

-65°C to +150°C

w	Conditions	ADLH0033G			ADLH0033CG			]
Parameter		min	typ	max	min	typ	max	Units
DC ELECTRICAL CHARACT	TERISTICS <sup>1</sup> , <sup>2</sup>							
Input Bias Current	T <sub>C</sub> = 25°C		0.1	0.15		0.15		nA
	_			10			5	nΑ
Input Impedance	$R_L = 1k\Omega$	1010	1011		1010	1011	-	Ω
Voltage Gain	$V_{IN} = 1V \text{ rms, } f = 1kHz,$	0.96	0.98	1.0	0.96	0.98	1.0	V/V
	$R_L = 1k\Omega$ , $R_S = 100k\Omega$							
Output Offset Voltage	$R_S = 100k\Omega$ , $T_C = 25^{\circ}C$		5	10		12	20	mV
	$R_S = 100k\Omega$			15			25	mV
Output Offset Voltage TC	$R_S = 100k\Omega$	1	50	100		50	100	μV/°C
Output Impedance	$V_{IN} = 1V \text{ rms}, f = 1kHz$	i	6	10		6	10	Ω
	$R_S = 100k\Omega$ , $R_L = 1k\Omega$							
Output Voltage Swing	$R_L = 1k\Omega$	±12	±13		±12	±13		v
	$R_{L}^{-} = 100\Omega$ , $T_{C} = 25^{\circ}C$	±9			±9			v
Supply Current	$V_{1N} = 0V, V_S = \pm 15V$		20	25		21	25	mA
Power Consumption	$V_{IN} = 0V$ , $V_S = \pm 15V$		600	660		630	720	mW
AC ELECTRICAL CHARACT	ERISTICS ( $T_C = 25^{\circ}C$ , $V_S = \pm 15$	$V, R_S = 50\Omega,$	R <sub>L</sub> = 1kS	2)				
Slew Rate	V <sub>IN</sub> = ±10V	1000	1500		1000	1400		V/µs
Bandwidth	V <sub>IN</sub> = 1V rms	j i	100	l		100		MHz
Phase Nonlinearity	BW = 1 to $20MHz$		2			2		Degrees
Rise Time	$\Delta V_{IN} = 0.5V$	l l	2.9		5	3.2	5	ns
Propagation Delay	$\Delta V_{IN} = 0.5V$	1	1.2			1.5		ns
Harmonic Distortion	f>1kHz	1 1	<0.1	i		<0.1		%
МТВГ								
Meantime Between Failure	1.962×10 <sup>7</sup>	l	Ţ					hours
PACKAGE OPTION <sup>3</sup>			•			·		
TO-8 (H-12A)		I A1	ADLH0033G		ADLH0033CG			

## ORDERING INFORMATION

Model ADLH0033CG ADLH0033G

Temperature Range -25°C to +85°C -55°C to +125°C

NOTES <sup>1</sup> Unless otherwise specified, these specifications apply for +15V applied to pins 1 and 12, -15V applied to pins 9 and 10, and pin 6 connected to pin 7. <sup>2</sup> Unless otherwise noted, specifications apply over a temperature range, -55°C<T<sub>C</sub><+125°C for the ADLH0033C6, and -25°C<T<sub>C</sub><+85°C for the ADLH0033CG. Typical values shown  $= 28^{\circ}C = 28^{\circ}C$ are for T<sub>C</sub> = 25°C.

<sup>&</sup>lt;sup>3</sup>See Section 16 for package outline information.

Specifications subject to change without notice.

## Applying the ADLH0033G/ADLH0033CG

## LAYOUT CONSIDERATIONS

As is the case with any high-speed design, proper layout is critical to avoid the introduction of unnecessary errors due to high-frequency coupling, stray capacitance, and the like.

Large ground planes should be used whenever possible to provide a low resistance, low inductance circuit path, as well as shielding the effects of high-frequency coupling. Sockets should be avoided, as the increased inter-lead capacitance can degrade bandwidth. Input and output connections should be kept as short as practical.

## OFFSET ADJUSTMENT

The ADLH0033G/ADLH0033CG are factory trimmed for output voltage offsets well within the guaranteed limits, thereby eliminating the need to calibrate each device individually. To use this feature, simply connect Pin 6 (OFFSET PRESET) to Pin 7 (OFFSET ADJUST).

When it is desirable to eliminate any errors due to output offsets, the circuit of Figure 1 may be used to adjust these errors to zero.

## SHORT CIRCUIT PROTECTION

The circuit of Figure 2 is used to protect the ADLH0033G/ADLH0033CG from short circuits on the output. The value of  $R_{LIM}$  is determined by the following:

$$R_{LIM}\cong \frac{V^{+}}{I_{sc}} = \frac{V^{-}}{I_{sc}}$$

Where I<sub>sc</sub> = Output Current under short circuit conditions ≤100mA.

Note that output voltage swing will also be somewhat limited in this configuration; however, decoupling of Pins 1 and 9 through disc type capacitors to ground as shown in Figure 2 will restore full output swing for transient pulses.

## **OPERATION WITH ASYMMETRICAL SUPPLIES**

Since Symmetrical Power Supplies may not always be desirable or available, the ADLH0033G/ADLH0033CG is designed to operate on Asymmetrical Supplies. This causes an apparent output offset; however, this is because of the amplifier's gain of less than unity. To accurately predict the output voltage shift due to Asymmetrical Supplies, use the following formula:

$$A_{VO} \cong (1 - A_V) \frac{(V + - V -)}{2} = 0.005 (V + - V -)$$

Where A<sub>V</sub> = No Load Voltage Gain, typically 0.99 V+ = Positive Supply Voltage

V- = Negative Supply Voltage

Of course, these apparent offsets may be adjusted to zero by using the circuit shown in Figure 1, OFFSET ADJUSTMENT.

## CAPACITIVE LOADING

The ADLH0033G/ADLH0033CG have been designed to drive capacitive loads of several thousand picofarads (such as coaxial cable) without oscillation. In these applications, peak current resulting from (C X dv/dt) should be limited below the absolute maximum peak current rating of ±250mA.

Also, power dissipation due to driving capacitive loads plus standby power should be kept below the total power rating of 1.5W.

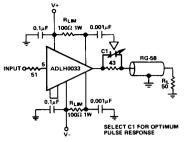


Figure 3. Coaxial Cable Drive

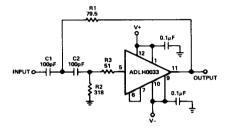


Figure 4. Wideband Two Pole High Pass Filter

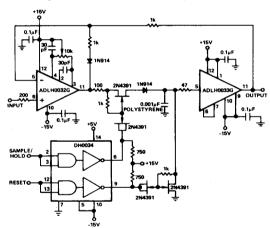


Figure 5. High Speed Peak Detector

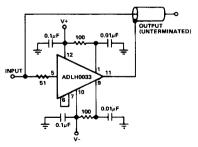


Figure 6. High Speed Shield/Line Driver

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## **Typical Performance Curves**

