

DUAL RETRIGGERABLE MONOSTABLE MULTIVIBRATOR

- HIGH SPEED $t_{PD} = 28 \text{ ns (TYP)}$ at $V_{CC} = 5V$
- LOW POWER DISSIPATION STANDBY STATE I_{CC} = 4 μ A (MAX.) at T_A = 25°C ACTIVE STATE ICC = 200 µA (TYP) at VCC = 5V
- HIGH NOISE IMMUNITY $V_{NIH} = V_{NII} = 28\% V_{CC} (MIN.)$
- OUTPUT DRIVE CAPABILITY 10 LSTTL LOADS
- SYMMETRICAL OUTPUT IMPEDANCE $|I_{OH}| = |I_{OL}| = 4 \text{ mA (MIN.)}$
- BALANCED PROPAGATION DELAYS tplh = tphi
- **WIDE OPERATING VOLTAGE RANGE** V_{CC} (OPR) = 2V to 6V
- WIDE OUTPUT PULSE WIDTH RANGE $t_{WOUT} = 120$ ns ~ 60s over at $V_{CC} = 4.5$ V
- PIN AND FUNCTION COMPATIBLE WITH 54/74LS123

DESCRIPTION

The M54/74HC123 is a high speed CMOS MONO-STABLE multivibrator fabricated with silicon gate C2MOS technology. It achieves the high speed operation similar to equivalent LSTTL while maintaining the CMOS low power dissipation. There are two trigger inputs, A INPUT (negative edge) and 8 INPUT (positive edge). These inputs are valid for rising/falling signals, (trt-l sec).

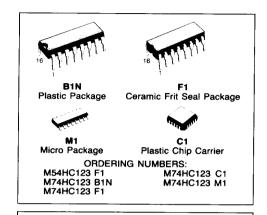
The device may also be triggered by using the CLR input (positive-edge) because of the Schmitt-trigger input; after triggering the output maintains the MO-NOSTABLE state for the time period determined by the external resistor Rx and capacitor Cx. Taking CLR low breaks this MONOSTABLE STATE. If the next trigger pulse occurs during the MONO-STABLE period it makes the MONOSTABLE period longer. Limit for values of Cx and Rx:

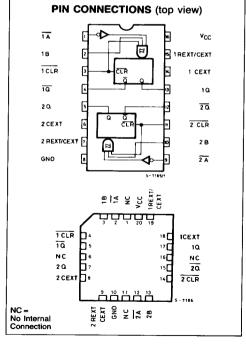
Cx: NO LIMIT

Rx: $V_{CC} = 2.0V 5K\Omega$ to $1M\Omega$ $V_{CC} = 3.0V 1K\Omega \text{ to } 1M\Omega$

voltage

All inputs are equipped with protection circuits against static discharge and transient excess





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1/8

TRUTH TABLE

	INPUTS		ОПТ	PUTS	NOTE
Ā	В	CL	Q	ā	NOIE
7_	Н	н	-T-	T	OUTPUT ENABLE
х	L	Н	L	н	INHIBIT
Н	х	Н	L	н	INHIBIT
L	<u>_</u>	Н	7.	7.	OUTPUT ENABLE
L	Н		7.	T	OUTPUT ENABLE
х	x	L	L	Н	INHIBIT

ABSOLUTE MAXIMUM RATINGS

Symbol	Parameter	Value	Unit
V _{CC}	Supply Voltage	- 0.5 to 7	V
V _I	DC Input Voltage	-0.5 to V _{CC} +0.5	V
v _o	DC Output Voltage	-0.5 to V _{CC} +0.5	٧
lik	DC Input Diode Current	± 20	mA
lok	DC Output Diode Current	± 20	mA
lo	DC Output Source Sink Current Per Output Pin	± 25	mA
I _{CC} or I _{GND}	DC V _{CC} or Ground Current	± 50	mA
PD	Power Dissipation	500 (*)	mW
T _{stq}	Storage Temperature	-65 to 150	°C

Absolute Maximum Ratings are those values beyond which damage to the device may occur. Functional operation under these condition is not implied.

RECOMMENDED OPERATING CONDITIONS

Symbol	Parameter	Value	Unit
Vcc	Supply Voltage	2 to 6	V
V _I	Input Voltage	0 to V _{CC}	V
Vo	Output Voltage	0 to V _{CC}	V
TA	Operating Temperature 74HC Series 54HC Series	40 to 85 55 to 125	°C
t _r , t _f	Input Rise and Fall Time (CLR only)	V _{CC}	ns
Сх	External Capacitor	NO LIMITATION	
Rx	External Resistor	V _{CC} {3 V 5K to 1M 1K to 1M	Ω

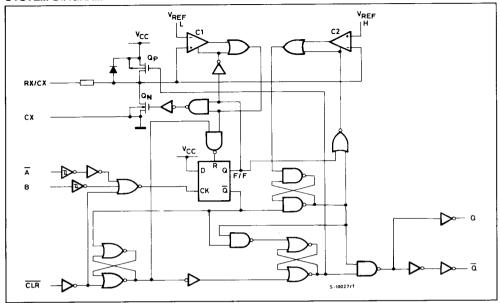
2/8

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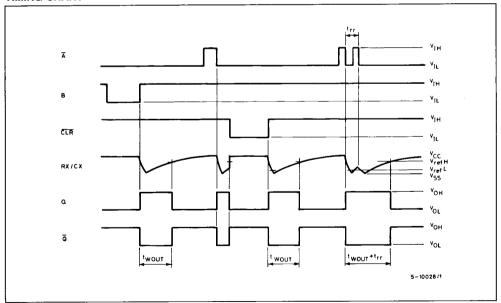
^{(*) 500} mW:

65°C derate to 300 mW by 10 mW/°C: 65°C to 85°C

SYSTEM DIAGRAM



TIMING CHART

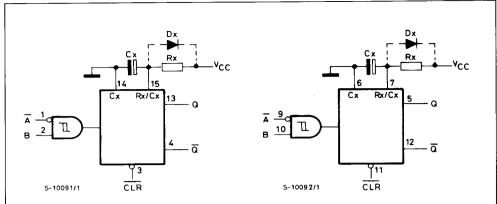


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3/8

195

BLOCK DIAGRAM



Note

- (1) Cx, Rx, Dx are external components.
- (2) Dx is a clamping diode

The external capacitor is charged to V_{CC} in the stand-by state, i.e. no trigger. When the supply voltage is turned off Cx is discharged mainly through an internal parasitic diode (see figures). If Cx is sufficiently large and V_{CC} decreases rapidy, there will be some possibility of damaging the I.C. with a surge current or latch-up. If the voltage supply filter capacitor is large enough and V_{CC} decrease slowly, the surge current is automatically limited and damage the I.C. is avoided. The maximum forward current of the parasitic diode is approximately 20 mA. In cases where Cx is large the time taken for the supply voltage to fall to 0.4 V_{CC} can be calculated as follows:

$$t_{f} \ge (V_{CC} - 0.7) \cdot Cx/20mA$$

In cases where $t_{\rm f}$ is too short an external clamping diode is required to protect the I.C. from the surge current.

FUNCTIONAL DESCRIPTION

Stand-by state

The external capacitor, Cx, is fully charged to V_{CC} in the stand-by state. Hence, before triggering, transistor Qp and Qn (connected to the Rx/Cx node) are both turned off. The two comparators that control the timing and the two reference voltage sources stop operating. The total supply current is therefore only leakage current.

Trigger operation

Triggering occurs when:

1 st) A is "low" and B has a falling edge;

2 nd) B is "high" and A has a rising edge;

3 rd) A is low and B is high and C1 has a rising edge.

After the multivibrator has been retriggered comparator C1 and C2 start operating and Qn is turned on. Cx then discharges through Qn. The voltage at the node R/C external falls.

When it reaches V_{REFL} the output of comparator C1 becomes low. This in turn resets the flip-flop and Qn is turned off.

At this point C1 stops functioning but C2 continues to operate.

The voltage at R/C external begins to rise with a time constant set by the external components Rx, Cx.

Triggering the multivibrator causes Q to go high after internal delay due to the flip-flop and the gate. Q remains high until the voltage at R/C external rises again to V_{REFH}. At this point C 2 output goes low and O goes low. C 2 stops operating. That means that after triggering when the voltage R/C external returns to V_{REFH} the multivibrator has returned to its MONOSTABLE STATE. In the case where Rx • Cx are large enough and the discharge time of the capacitor and the delay time in the I.C. can be ignored, the width of the output pulse tw (out) is as follows:

t_{W(OUT)} = 0.46 Cx • Rx

4/8

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196

FUNCTIONAL DESCRIPTION (Continued)

Re-trigger operation

When a second trigger pulse follows the first its effect will depend on the state of the multivibrator. If the capacitor Cx is being charged the voltage level of R/C external falls to Vrefl again and Q remains high i.e. the retrigger pulse arrives in a time shorter than the period Rx • Cx seconds, the capacitor charging time constant. If the second trigger pulse is very close to the initial trigger pulse it is ineffective; i.e., the second trigger must arrive in the capacitor discharge cycle to be ineffective.

Hence the minimum time for a second trigger to be effective depends on V_{CC} and Cx.

Reset operation

CL is normally high. If CL is low, the trigger is not effective because Q output goes low and trigger control flip-flop is reset.

Also transistor Op is turned on and Cx is charged quicky to V_{CC} . This means if CL input goes low, the IC becomes waiting state both in operating and non operating state.

DC SPECIFICATIONS

Symbol	Parameter	v _{cc}	Test Condition			_A = 25° C and 7			s5°C HC		125°C HC	Unit
					Min.	Тур.	Max.	Min.	Max.	Min.	Max.	
V _{IH}	High Level Input Voltage	2.0 4.5 6.0			1.5 3.15 4.2	_ _ _	-	1.5 3.15 4.2	-	1.5 3.15 4.2	_ _	٧
VIL	Low Level Input Voltage	2.0 4.5 6.0			_ _ _	_ _ _	0.5 1.35 1.8		0.5 1.35 1.8		0.5 1.35 1.8	v
Voн	High Level Output Voltage (Q, Q Output)	2.0 4.5 6.0 4.5 6.0	V _I = V _{IH} = V _{IL}	$I_{OH} = -20\mu A$ $I_{OH} = -4mA$ $I_{OH} = -5.2mA$	1.9 4.4 5.9 4.18 3.68	2.0 4.5 6.0 4.31 5.80	_ 	1.9 4.4 5.9 4.13 5.63	1	1.9 4.4 5.9 4.10 5.60	_ _ _ _	٧
V _{OL}	Low Level Output Voltage (Q, Q Output)	2.0 4.5 6.0 4.5 6.0	V _I = V _{IH} = V _{IL}	I _{OL} = 20 μA I _{OL} = 4 mA I _{OL} = 5.2 mA	- - -	0.0 0.0 0.0 0.17 0.18	0.1 0.1 0.1 0.26 0.26	- - - -	0.1 0.1 0.1 0.33 0.33		0.1 0.1 0.1 0.40 0.40	٧
IIN	Input Leakage Current	6.0	V _I = V	CC or GND		_	±0.1	_	±1.0	1	±1.0	μΑ
l ^{IN}	R/C Terminal Off-State Current	6.0	V _I = V _C	V _I =V _{CC} or GND		_	±0.5	_	±5.0	-	±10	μΑ
lcc	Quiescent Supply Current	6.0	V _I = V _C	V _I =V _{CC} or GND		_	4	_	40	_	80	μΑ
lcc'	Active State (1) Supply Current	2.0 4.5 6.0	Pir	or GND ns 2, 14 N=V _{CC/2}		40 0.1 0.2	120 0.3 0.6	1 -	160 0.4 0.8	-	200 0.5 1.0	μA mA mA

(1): Per Circuit

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5/8

AC ELECTRICAL CHARACTERISTICS ($V_{CC} = 5V$, $T_A = 25$ °C, $C_L = 15pF$, Input $t_r = t_f = 6ns$)

		54HC and 74HC						
Symbol tTLH tTHL tPLH tPHL	Parameter	Min.	Тур.	Max.	Unit			
	Output Transition Time		4	8	ns			
	Propagation Delay Time (A, B - Q, Q)		27	41	ns			
t _{PLH}	Propagation Delay Time (CLEAR, TRIGGER - Q, Q)		29	45	ns			
t _{PLH}	Propagation Delay Time (CLEAR - Q, Q)		21	33	ns			

AC ELECTRICAL CHARACTERISTICS ($C_L = 50 pF$, Input $t_f = t_f = 6 ns$)

Symbol	Parameter	V _{CC} Test Condition	Test Condition	T _A = 25°C 54HC and 74HC			- 40 to 85°C 74HC		– 55 to 125°C 54HC		Unit
				Min.	Тур.	Max.	Min.	Max.	Min.	Max.	
t _{TLH} t _{THL}	Output Transition Time	2.0 4.5 6.0			30 8 7	75 15 13	_ _ _	95 19 16	<u>-</u>	110 22 19	ns
t _{PLH}	Propagation Delay Time (Ā, B - Q, Q)	2.0 4.5 6.0		=	124 31 26	240 48 41	_ _ _	300 60 51	_ _ _	360 72 61	ns
t _{PLH}	Propagation Delay Time (CLR TRIG - Q, Q)	2.0 4.5 6.0		=	136 34 29	265 53 45	_ 	335 66 56	=	400 80 68	ns
t _{PLH} t _{PHL}	Propagation Delay Time (CLR - Q, Q)	2.0 4.5 6.0		=	100 25 21	195 39 33	_ _ _	245 49 42	_ 	295 59 50	ns
t _{rr}		4.5 6.0	Cx = 100pF Rx = 1KΩ	_	70 60	_	=	_	_	_	ns
	Time	4.5 6.0	$Cx = 0.01 \mu F$ $Rx = 1 K\Omega$	_	1.0 0.9	=		_	_	=	μS
Δtwout	Output Pulse Width Error Between Circuits in Same Package			_	±1	_	_	_	_	_	%
twout (Min)	Output Pulse Width	4.5	Cx = 0 Rx = 1kΩ	_	118	_	_	_	_		ns
twout	Output Pulse Width	4.5	Cx = 100pF Rx = 10kΩ	_	1.0		_	_	_	_	μS
		4.5	$Cx = 0.1\mu$ $Rx = 100k\Omega$	_	4.7	_	_	_	_	_	ms

6/8

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AC ELECTRICAL CHARACTERISTICS (Continued)

Symbol	Parameter	V _{CC} Test Condition	T _A = 25°C 54HC and 74HC			– 40 to 85°C 74HC		– 55 to 125°C 54HC		Unit	
				Min.	Тур.	Max.	Min.	Max.	Min.	Max.	
t _{W(H)}	Minimum Pulse Width (Trigger)	2.0 4.5 6.0		=	40 10 9	100 20 17	_ 	125 25 21	_ _ _	150 30 26	ns
t _{W(L)}	Minimum Clear Pulse Width	2.0 4.5 6.0		=	30 8 7	75 15 13	_ _ _	95 19 16		110 22 19	ns
C _{IN}	Input Capacitance			_	5	10	_	10		10	рF
C _{PD} (*)	Power Dissipation Capacitance			_	113	_	_	_	_	_	рF

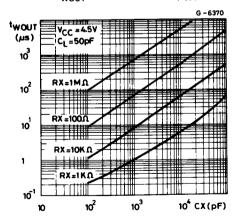
Note (*) C_{PD} is defined as the value the IC's of internal equivalent capacitance which is calculated from the operating current consumption without load. (Refer to Test Circuit)

Average operating current can be obtained by the equation hereunder $I_{CC(opr.)} = C_{PD} \cdot V_{CC} \cdot f_{IN} + I_{CC}'$ Duty/100 + $I_{CC}/2$ (per monostable)

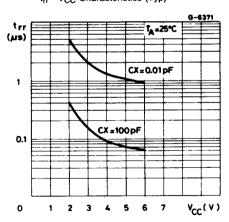
(I_{CC}': Active Supply Current)

(Duty: %)





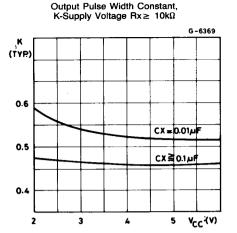
t_{rr} - V_{CC} Characteristics (Typ)



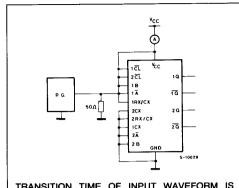
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7/8

199

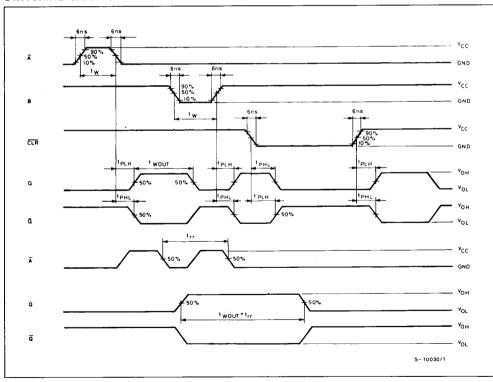


TEST CIRCUIT ICC (Opr)



TRANSITION TIME OF INPUT WAVEFORM IS THE SAME AS THAT IN SASE OF SWITCHING CHARACTERISTICS TESTS.

SWITCHING CHARACTERISTICS TEST WAVEFORM



8/8

200

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WIGGROULECTROMICS