

# 3.3V TIMEKEEPER CONTROLLER

PRELIMINARY DATA

- CONVERTS LOW POWER SRAM into NVRAMs
- YEAR 2000 COMPLIANT (4-Digit Year)
- USES SUPER CAPACITOR or LITHIUM BATTERY
- BATTERY LOW FLAG
- INTEGRATED REAL TIME CLOCK, POWER-FAIL CONTROL CIRCUIT
- AUTOMATIC POWER-FAIL CHIP DESELECT and WRITE PROTECTION
- WATCHDOG TIMER
- WRITE PROTECT VOLTAGES (V<sub>PFD</sub> = Power-fail Deselect Voltage):
  - M48T212A:  $2.7V \le V_{PFD} \le 3.0V$
- MICROPROCESSOR POWER-ON RESET
- PROGRAMMABLE ALARM OUTPUT ACTIVE in the BATTERY BACKED-UP MODE
- PACKAGING INCLUDES a 44-LEAD SOIC and SNAPHAT TOP (to be Ordered Separately)

# **DESCRIPTION**

The M48T212A is a self-contained device that includes a real time clock (RTC), programmable alarms, a watchdog timer, and two external chip enable outputs which provide control of up to four (two in parallel) external low-power static RAMs.

A built-in 32.768 kHz oscillator (extended crystal controlled) is used for the clock/calendar function. Access to all TIMEKEEPER functions and the external RAM is the same as conventional byte-wide SRAM. The 16 TIMEKEEPER Registers offer Century, Year, Month, Date, Day, Hour, Minute, Second, Control, Calibration, Alarm, Watchdog, and Flags. Externally attached static RAMs are controlled by the M48T212A via the  $\overline{\text{E1}}_{\text{CON}}$  and  $\overline{\text{E2}}_{\text{CON}}$  signals (see Table 4).

Automatic backup and write protection for  $\underline{an}$  external SRAM is provided through  $V_{OUT}$ ,  $\overline{E1}_{CON}$  and  $\overline{E2}_{CON}$  pins. (Users are urged to insure that voltage specifications, for both the controller chip and external SRAM chosen, are similar).

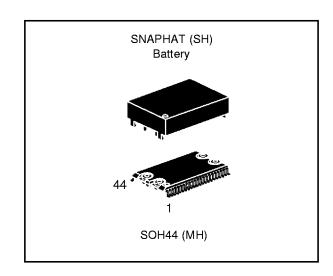
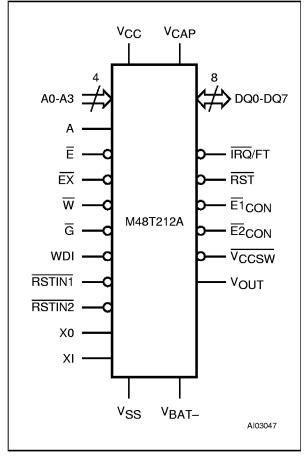
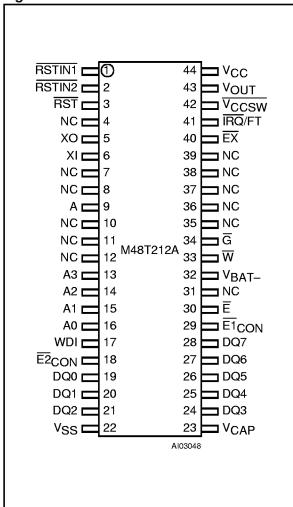


Figure 1. Logic Diagram



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Figure 2. SOIC Connections



**Table 1. Signal Names** 

A0-A3	Address Inputs
DQ0-DQ7	Data Inputs/Outputs
хо	Oscillator Output
ΧI	Oscillator Input
RSTIN1	Reset 1 Input
RSTIN2	Reset 2 Input
RST	Reset Output (Open Drain)
WDI	Watchdog Input
Α	Bank Select Input
Ē	Chip Enable Input
EX	External Chip Enable Input
G	Output Enable Input
₩	Write Enable Input
E1 <sub>CON</sub>	RAM Chip Enable 1 Output
E2 <sub>CON</sub>	RAM Chip Enable 2 Output
ĪRQ/FT	Int/Freq Test Output (Open Drain)
Vccsw	V <sub>CC</sub> Switch Output
V <sub>OUT</sub>	Supply Voltage Output
VCAP	Super Capacitor Input
V <sub>BAT</sub>	Battery Ground Pin (optional)
Vcc	Supply Voltage
V <sub>SS</sub>	Ground
NC	Not Connected internally

The lithium energy source (or super capacitor) used to permanently power the real time clock is also used to retain RAM data in the absence of  $V_{CC}$  power through the  $V_{OUT}$  pin.

The chip enable outputs to RAM ( $\overline{\text{E1}}_{\text{CON}}$ ) and  $\overline{\text{E2}}_{\text{CON}}$ ) are controlled during power transients to prevent data corruption. The date is automatically adjusted for months with less than 31 days and corrects for leap years. The internal watchdog timer provides programmable alarm windows.

The nine clock bytes (Fh - 9h and 1h) are not the actual clock counters, they are memory locations consisting of BiPORT<sup>TM</sup> read/write memory cells within the static RAM array. Clock circuitry updates the clock bytes with current information once per second. The information can be accessed by the user in the same manner as any other location in the static memory array.

Byte 8h is the clock control register. This byte controls user access to the clock information and also stores the clock calibration setting. Byte 7h con-

tains the watchdog timer setting. The watchdog timer can generate either a reset or an interrupt, depending on the state of the Watchdog Steering bit (WDS). Bytes 6h-2h include bits that, when programmed, provide for clock alarm functionality.

Alarms are activated when the register content matches the month, date, hours, minutes, and seconds of the clock registers. Byte 1h contains century information. Byte 0h contains additional flag information pertaining to the watchdog timer, alarm and battery status.

The M48T212A also has its own Power-Fail Detect circuit. This control circuitry constantly monitors the supply voltage for an out of tolerance condition. When  $V_{\rm CC}$  is out of tolerance, the circuit write protects the TIMEKEEPER register data and external SRAM, providing data security in the midst of unpredictable system operation. As  $V_{\rm CC}$  falls, the control circuitry automatically switches to the battery, maintaining data and clock operation until valid power is restored.

Table 2. Absolute Maximum Ratings (1)

Symbol	Parameter	Value	Unit
T <sub>A</sub>	Ambient Operating Temperature	0 to 70	°C
T <sub>STG</sub>	Storage Temperature (V <sub>CC</sub> Off, Oscillator Off) SNAPHAT SOIC	–40 to 85 –55 to 125	°C
T <sub>SLD</sub> <sup>(2)</sup>	Lead Solder Temperature for 10 sec	260	°C
V <sub>IO</sub>	Input or Output Voltages	–0.3 to 4.6	V
V <sub>CC</sub>	Supply Voltage	–0.3 to 4.6	٧
lo	Output Current	20	mA
P <sub>D</sub>	Power Dissipation	1	W

Note: 1. Stresses greater than those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. This is a stress rating only and functional operation of the device at these or any other conditions above those indicated in the operational section of this specification is not implied. Exposure to the absolute maximum rating conditions for extended periods of time may affect reliability

CAUTION: Negative undershoots below -0.3V are not allowed on any pin while in the Battery Back-up mode.

CAUTION: Do NOT wave solder SOIC to avoid damaging SNAPHAT sockets.

Table 3. Operating Modes (1)

Mode	Vcc	Ē	G	w	DQ7-DQ0	Power
Deselect		V <sub>IH</sub>	Х	Х	High-Z	Standby
Write	3.0V to 3.6V	V <sub>IL</sub>	Х	V <sub>IL</sub>	D <sub>IN</sub>	Active
Read	3.00 10 3.00	V <sub>IL</sub>	V <sub>IL</sub>	V <sub>IH</sub>	Dout	Active
Read		V <sub>IL</sub>	V <sub>IH</sub>	V <sub>IH</sub>	High-Z	Active
Deselect	$V_{SO}$ to $V_{PFD}$ (min) $^{(2)}$	Х	Х	Х	High-Z	CMOS Standby
Deselect			Х	High-Z	Battery Back-Up	

Note: 1.  $X = V_{IH}$  or  $V_{IL}$ .

# Table 4. Truth Table for SRAM Bank Select (1)

Mode	V <sub>CC</sub>	ĒΧ	Α	E1 <sub>CON</sub>	E2 <sub>CON</sub>	Power
Select		Low	Low	Low	High	Active
Jelect	3.0V to 3.6V	Low	High	High	Low	Active
Deselect		High	Х	High	High	Standby
Deselect	V <sub>SO</sub> to V <sub>PFD</sub> (min) <sup>(2)</sup>	Х	Х	High	High	CMOS Standby
Deselect	≤ V <sub>SO</sub> <sup>(2)</sup>	Х	Х	High	High	Battery Back-Up

Note: 1.  $X = V_{IH}$  or  $V_{IL}$ .

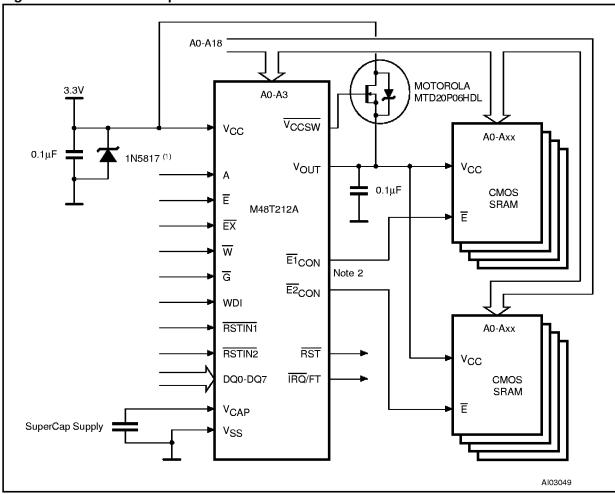
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<sup>2.</sup> Soldering temperature not to exceed 260°C for 10 seconds (total thermal budget not to exceed 150°C for longer than 30 seconds).

<sup>2.</sup> V<sub>SO</sub> = Battery Back-up Switchover Voltage. (See Table 7 for details).

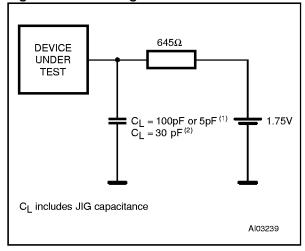
<sup>2.</sup> V<sub>SO</sub> = Battery Back-up Switchover Voltage. (See Table 7 for details).

Figure 3. Hardware Hookup



Note: 1. See description in Power Supply Decoupling and Undershoot Protection. 2. Traces connecting  $\overline{\text{E1}_{\text{CON}}}$  and  $\overline{\text{E2}_{\text{CON}}}$  to external SRAM should be as short as possible.

Figure 4. AC Testing Load Circuit (3,4)



**Table 5. AC Measurement Conditions** 

Input Rise and Fall Times	≤5ns
Input Pulse Voltages	0 to 3V
Input and Output Timing Ref. Voltages	1.5V

Note that Output Hi-Z is defined as the point where data is no longer driven.

Note: 1. DQ0-DQ7

2. E1<sub>CON</sub> and E2<sub>CON</sub>

3. Excluding open-drain output pins

# Table 6. Capacitance $^{(1)}$ (T<sub>A</sub> = 25 °C, f = 1 MHz)

[5	Symbol	Parameter	Test Condition	Min	Max	Unit
	CIN	Input Capacitance	V <sub>IN</sub> = 0V		10	pF
	C <sub>OUT</sub> <sup>(2)</sup>	Input/Output Capacitance	V <sub>OUT</sub> = 0V		10	pF

Note: 1. Sampled only, not 100% tested.

2. Outputs deselected.

# **Table 7. DC Characteristics**

 $(T_A = 0 \text{ to } 70^{\circ}\text{C}; V_{CC} = 3\text{V to } 3.6\text{V})$ 

Symbol	Parameter	Test Condition	Min	Тур	Max	Unit
I <sub>LI</sub> (1,2)	Input Leakage Current	$0V \le V_{IN} \le V_{CC}$			±1	μА
l <sub>LO</sub> <sup>(1)</sup>	Output Leakage Current	0V ≤ V <sub>OUT</sub> ≤ V <sub>CC</sub>			±1	μΑ
lcc	Supply Current	Outputs Open		4	10	mA
l <sub>CC1</sub>	Supply Current (Standby) TTL	Ē = V <sub>IH</sub>			3	mΑ
I <sub>CC2</sub>	Supply Current (Standby) CMOS	Ē = V <sub>CC</sub> −0.2			2	mA
IBAT	Battery Current OSC ON			575	800	nA
IBAI	Battery Current OSC OFF				100	nA
V <sub>IL</sub>	Input Low Voltage		-0.3		0.8	٧
V <sub>IH</sub>	Input High Voltage		2.0		V <sub>CC</sub> + 0.3	٧
Val	Output Low Voltage	I <sub>OL</sub> = 2.1mA			0.4	٧
V <sub>OL</sub>	Output Low Voltage (open drain) (3)	I <sub>OL</sub> = 10mA			0.4	٧
V <sub>OH</sub>	Output High Voltage	I <sub>OH</sub> = -1.0mA	2.4			٧
V <sub>OHB</sub> <sup>(4)</sup>	V <sub>OH</sub> Battery Back-up	I <sub>OUT2</sub> = -1.0μA	2.0		3.6	٧
I <sub>OUT1</sub> (5)	V <sub>OUT</sub> Current (Active)	V <sub>OUT1</sub> > V <sub>CC</sub> −0.3			70	mA
I <sub>OUT2</sub>	V <sub>OUT</sub> Current (Battery Back-up)	$V_{OUT2} > V_{BAT} - 0.3$			100	μΑ
V <sub>PFD</sub>	Power-fail Deselect Voltage		2.7	2.9	3.0	٧
V <sub>SO</sub>	Battery Back-up Switchover Voltage			V <sub>PFD</sub> –100mV		٧
V <sub>BAT</sub>	Battery Voltage			3.0		٧
V <sub>CAP</sub>	Capacitor Voltage (6)			V <sub>CC</sub>		٧

- Note: 1. Outputs deselected.

  2. RSTIN1 and RSTIN2 internally pulled-up to V<sub>CC</sub> through 100KΩ resistor. WDI internally pulled-down to V<sub>SS</sub> through 100KΩ resistor.

  3. For IRQ/FT & RST pins (Open Drain).

  - 4. Conditioned outputs (E1CON E2CON) can only sustain CMOS leakage currents in the battery back-up mode. Higher leakage currents will reduce battery life.
  - 5. External SRAM must match TIMEKEE PER Controller chip  $V_{CC}$  specification.
  - 6. When fully charged.

Figure 5. Power Down/Up AC Waveform

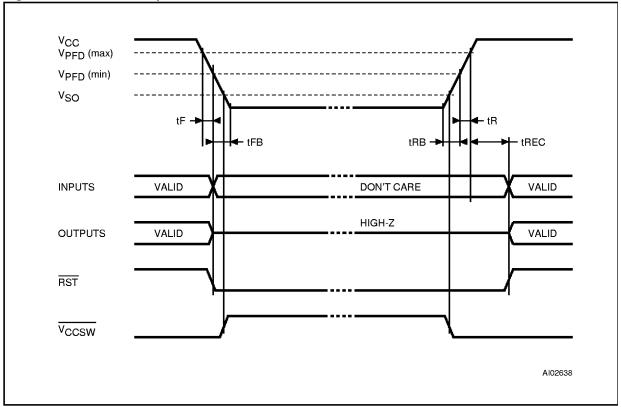


Table 8. Power Down/Up AC Characteristics ( $T_A = 0 \text{ to } 70^{\circ}\text{C}$ )

Symbol	Parameter	Min	Max	Unit
t <sub>F</sub>	V <sub>PFD</sub> (max) to V <sub>PFD</sub> (min) V <sub>CC</sub> Fall Time	300		μs
t <sub>FB</sub>	V <sub>PFD</sub> (min) to V <sub>SS</sub> V <sub>CC</sub> Fall Time	150		μs
t <sub>R</sub>	V <sub>PFD</sub> (min) to V <sub>PFD</sub> (max) V <sub>CC</sub> Rise Time	10		μs
t <sub>REC</sub>	V <sub>PFD</sub> (max) to RST High	40	200	ms
t <sub>RB</sub>	V <sub>SS</sub> to V <sub>PFD</sub> (min) V <sub>CC</sub> Rise Time	5		μs

Table 9. Chip Enable Control and Bank Select Characteristics ( $T_A=0\ to\ 70^{\circ}C)$ 

		M48T212A		
Symbol	Parameter	-85		Unit
		Min	Max	
t <sub>EXPD</sub>	EX to E1 <sub>CON</sub> or E2 <sub>CON</sub> (Low or High)		15	ns
t <sub>APD</sub>	A to $\overline{E1}_{CON}$ or $\overline{E2}_{CON}$ (Low or High)		15	ns

Figure 6. Chip Enable Control and Bank Select Timing

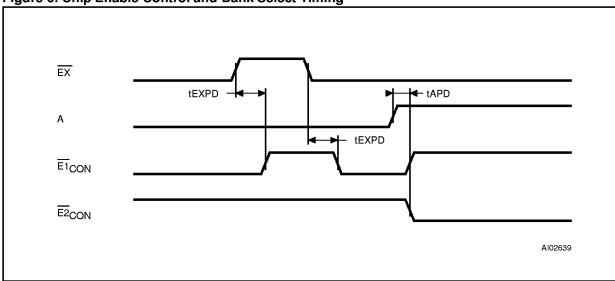


Table 10. Read Mode Characteristics ( $T_A = 0 \text{ to } 70^{\circ}C$ )

		M48T	212A		
Symbol	Parameter	-8	Unit		
		Min	Max		
t <sub>AVAV</sub>	Read Cycle Time	85		ns	
t <sub>AVQV</sub>	Address Valid to Output Valid		85	ns	
t <sub>ELQV</sub>	Chip Enable Low to Output Valid		85	ns	
<sup>t</sup> GLQV	Output Enable Low to Output Valid		35	ns	
t <sub>ELQX</sub> (1)	Chip Enable Low to Output Transition	5		ns	
t <sub>GLQX</sub> <sup>(1)</sup>	Output Enable Low to Output Transition	0		ns	
t <sub>EHQZ</sub> (1)	Chip Enable High to Output Hi-Z	25		ns	
t <sub>GHQZ</sub> <sup>(1)</sup>	Output Enable High to Output Hi-Z		25	ns	
t <sub>AXQX</sub>	Address Transition to Output Transition	5		ns	

Note: 1.  $C_L = 5pF$ 

#### **Address Decoding**

The M48T212A accommodates 4 address lines (A3-A0) which allow access to the sixteen bytes of the TIMEKEEPER clock registers. All TIMEKEEP-ER registers reside in the controller chip itself. All TIMEKEEPER registers are accessed by enabling E (Chip Enable).

#### **READ MODE**

The M48T212A executes a read cycle whenever  $\overline{W}$  (Write Enable) is high and  $\overline{E}$  (Chip Enable) is low. The unique address specified by the address inputs (A3-A0) defines which one of the on-chip TIMEKÈEPER registers is to be accessed. When the address presented to the M48T212A is in the range of 0h-Fh, one of the on-board TIMEKEEP-ER registers is accessed and valid data will be available to the eight data output drivers within tavov after the address input signal is stable, providing that the E and G access times are also satisfied. If they are not, then data access must be measured from the latter occurring signal ( $\overline{E}$  or  $\overline{G}$ ) and the limiting parameter is either  $t_{ELQV}$  for  $\overline{E}$  or t<sub>GLQV</sub> for  $\overline{G}$  rather than the address access time. When EX input is low, an external SRAM location will be selected.

Note: Care should be taken to avoid taking both E and EX low simultaneously to avoid bus conten-

**Table 11. Write Mode AC Characteristics**  $(T_A = 0 \text{ to } 70^{\circ}\text{C})$ 

		M48T	212A		
Symbol	Parameter	-8	-85		
		Min	Max		
t <sub>AVAV</sub>	Write Cycle Time	85		ns	
t <sub>AVWL</sub>	Address Valid to Write Enable Low	0		ns	
t <sub>AVEL</sub>	Address Valid to Chip Enable Low	0		ns	
t <sub>WLWH</sub>	Write Enable Pulse Width	55		ns	
<sup>†</sup> ELEH	Chip Enable Low to Chip Enable High 60				
t <sub>WHAX</sub>	t <sub>WHAX</sub> Write Enable High to Address Transition			ns	
<sup>†</sup> EHAX	Chip Enable High to Address Transition	0		ns	
t <sub>DVWH</sub>	Input Valid to Write Enable High	30		ns	
t <sub>DVEH</sub>	Input Valid to Chip Enable High	30		ns	
t <sub>WHDX</sub>	Write Enable High to Input Transition	0		ns	
t <sub>EHDX</sub>	Chip Enable High to Input Transition	0		ns	
t <sub>WLQZ</sub> (1,2)	Write Enable Low to Output High-Z		25	ns	
<sup>t</sup> AVWH	Address Valid to Write Enable High	65		ns	
t <sub>AVEH</sub>	Address Valid to Chip Enable High	65		ns	
t <sub>WHQX</sub> (1,2)	Write Enable High to Output Transition	5		ns	

Note: 1. C<sub>L</sub> = 5pF.

2. If E goes low simultaneously with W going low, the outputs remain in the high impedance state.

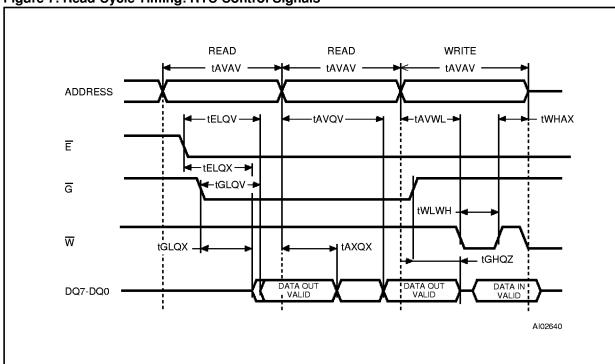
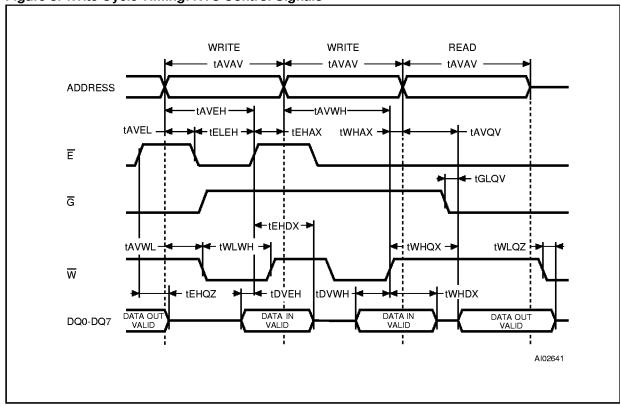


Figure 7. Read Cycle Timing: RTC Control Signals

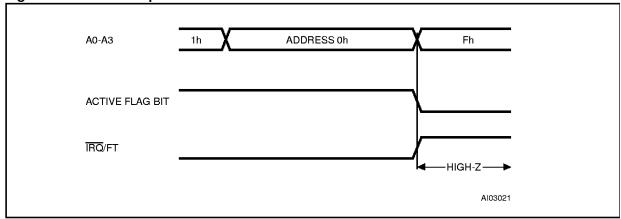




**Table 12. Alarm Repeat Modes** 

RPT5	RPT4	RPT3	RPT2	RPT1	Alarm Setting
1	1	1	1	1	Once per Second
1	1	1	1	0	Once per Minute
1	1	1	0	0	Once per Hour
1	1	0	0	0	Once per Day
1	0	0	0	0	Once per Month
0	0	0	0	0	Once per Year

Figure 9. Alarm Interrupt Reset Waveforms



#### WRITE MODE

The M48T212A is in the Write Mode whenever  $\overline{W}$  (Write Enable) and  $\overline{E}$  (Chip Enable) are in a low state after the address inputs are stable. The start of a write is referenced from the latter occurring falling edge of  $\overline{W}$  or  $\overline{E}$ . A write is terminated by the earlier rising edge of  $\overline{W}$  or  $\overline{E}$ . The addresses must be held valid throughout the cycle.  $\overline{E}$  or  $\overline{W}$  must return high for a minimum of  $t_{EHAX}$  from Chip Enable or  $t_{WHAX}$  from Write Enable prior to the initiation of another read or write cycle. Data-in must be valid  $t_{DVWH}$  prior to the end of write and remain valid for  $t_{WHDX}$  afterward.

 $\overline{G}$  should be kept high during write cycles to avoid bus contention; although, if the output bus has been activated by a low on  $\overline{E}$  and  $\overline{G}$  a low on  $\overline{W}$  will disable the outputs  $t_{WLQZ}$  after  $\overline{W}$  falls.

When  $\overline{E}$  is low during the write, one of the onboard TIMEKEEPER registers will be selected and data will be written into the device. When  $\overline{EX}$  is low (and  $\overline{E}$  is high) an external SRAM location is selected.

**Note:** Care should be taken to avoid taking both  $\overline{E}$  and  $\overline{EX}$  low simultaneously to avoid bus contention

#### **DATA RETENTION MODE**

With valid V<sub>CC</sub> applied, the M48T212A can be accessed as described above with read or write cycles. Should the supply voltage decay, the M48T212A will automatically deselect, write protecting itself (and any external SRAM) when V<sub>CC</sub> falls between V<sub>PFD</sub> (max) and V<sub>PFD</sub> (min). This is accomplished by internally inhibiting access to the clock registers via the  $\overline{\rm E}$  signal. At this time, the Reset pin ( $\overline{\rm RST}$ ) is driven active and will remain active until V<sub>CC</sub> returns to nominal levels.

External RAM access is inhibited in a similar manner by forcing  $\overline{E1}_{CON}$  and  $\overline{E2}_{CON}$  to a high level. This level is within 0.2V of the  $V_{BAT}$ .  $\overline{E1}_{CON}$  and  $\overline{E2}_{CON}$  will remain at this level as long as  $V_{CC}$  remains at an out-of tolerance condition.

When  $V_{CC}$  falls below the level of the battery ( $V_{BAT}$ ), power input is switched from the  $V_{CC}$  pin to the battery and the clock registers and external SRAM are maintained from the attached battery supply. All outputs become high impedance. The  $V_{OUT}$  pin is capable of supplying  $100\mu A$  of current to the attached memory with less than 0.3V drop under this condition. On power up, when  $V_{CC}$  returns to a nominal value, write protection continues for 200ms (max) by inhibiting  $\overline{E1}_{CON}$  or  $\overline{E2}_{CON}$ .

Table 13. TIMEKEEPER Register Map

Address								Function/Range			
Addicos	D7	D6	D5	D4	D3	D2	D1	DO	BCD Format		
Fh		10 Y	ears		Year			Year	00-99		
Eh	0	0	0	10M	10M Month			Month	01-12		
Dh	0	0	10 [	Date		Date: Day	of Month		Date	01-31	
Ch	0	FT	0	0	0 0 Day of Week			Day	01-7		
Bh	0	0	10 ⊢	ours Hours (24 Hour Format)			Hour	00-23			
Ah	0		10 Minutes	Minutes			Min	00-59			
9h	ST	1	0 Second	s	s Seconds			Sec	00-59		
8h	W	R	S		Calibration			Control			
7h	WDS	BMB4	вмвз	BMB2	BMB1 BMB0 RB1 RB0		Watchdog				
6h	AFE	0	ABE	Al 10M	Alarm Month			A Month	01-12		
5h	RPT4	RPT5	Al 10	Date	Alarm Date			A Date	01-31		
4h	RPT3	0	Al 10	Hour	r Alarm Hour			A Hour	00-23		
3h	RPT2	Alaı	m 10 Mini	utes	Alarm Minutes			A Min	00-59		
2h	RPT1	Alar	m 10 Seco	onds	Alarm Seconds			A Sec	00-59		
1h		1000	Year		100 Year			Century	00-99		
0h	WDF	AF	Υ	BL	YYYY			Flag	_		

Keys: S = Sign Bit

FT = Frequency Test Bit

R = Read Bit
W = Write Bit
ST = Stop Bit
0 = Must be set to zero
BL = Battery Low Flag

BMB0-BMB4 = Watchdog Multiplier Bits

AFE = Alarm Flag Enable Flag RB0-RB1 = Watchdog Resolution Bits WDS = Watchdog Steering Bit

ABE = Alarm in Battery Back-Up Mode Enable Bit

RPT1-RPT5 = Alarm Repeat Mode Bits

WDF = Watchdog flag AF = Alarm flag Y = '1' or '0'

The RST signal also remains active during this time (see Figure 5).

**Note:** Most low power SRAMs on the market today can be used with the M48T212A TIMEKEEP-ER Controller. There are, however some criteria which should be used in making the final choice of an SRAM to use. The SRAM must be designed in a way where the chip enable input disables all other inputs to the SRAM. This allows inputs to the M48T212A and SRAMs to be Don't Care once  $V_{CC}$  falls below  $V_{PFD}$ (min). The SRAM should also guarantee data retention down to  $V_{CC}$  = 2.0V. The chip enable access time must be sufficient to meet the system needs with the chip enable output propagation delays included.

If the SRAM includes a second chip enable pin  $(\overline{E2})$ , this pin should be tied to  $V_{OUT}$ .

If data retention lifetime is a critical parameter for the system, it is important to review the data retention current specifications for the particular SRAMs being evaluated. Most SRAMs specify a data retention current at 3.0V. Manufacturers generally specify a typical condition for room temperature along with a worst case condition (generally at elevated temperatures). The system level requirements will determine the choice of which value to use.

The data retention current value of the SRAMs can then be added to the I<sub>BAT</sub> value of the M48T212A to determine the total current requirements for data retention. The available battery capacity can then be divided by this current to determine the amount of data retention available.

For a further more detailed review of lifetime calculations, please see Application Note AN1012.

#### TIMEKEEPER REGISTERS

The M48T212A offers 16 internal registers which contain TIMEKEEPER, Alarm, Watchdog, Flag, and Control data. These registers are memory locations which contain external (user accessible) and internal copies of the data (usually referred to as BiPORT<sup>TM</sup> TIMEKEEPER cells).

The external copies are independent of internal functions except that they are updated periodically by the simultaneous transfer of the incremented internal copy. TIMEKEEPER and Alarm Registers store data in BCD. Control, Watchdog and Flags Registers store data in Binary Format.

#### **CLOCK OPERATIONS**

#### Reading the Clock

Updates to the TIMEKEEPER registers should be halted before clock data is read to prevent reading data in transition. Because the BiPORT TIME-KEEPER cells in the RAM array are only data registers, and not the actual clock counters, updating the registers can be halted without disturbing the clock itself.

Updating is halted when a '1' is written to the READ bit, D6 in the Control Register (8h). As long as a '1' remains in that position, updating is halted. After a halt is issued, the registers reflect the count; that is, the day, date, and time that were current at the moment the halt command was issued

All of the TIMEKEEPER registers are updated simultaneously. A halt will not interrupt an update in progress. Updating occurs 1 second after the READ bit is reset to a '0'.

# **Setting the Clock**

Bit D7 of the Control Register (8h) is the WRITE bit. Setting the WRITE bit to a '1', like the READ bit, halts updates to the TIMEKEEPER registers. The user can then load them with the correct day, date, and time data in 24 hour BCD format (see Table 13).

Resetting the WRITE bit to a '0' then transfers the values of all time registers (Fh-9h, 1h) to the actual TIMEKEEPER counters and allows normal operation to resume. After the WRITE bit is reset, the next clock update will occur one second later.

**Note:** Upon power-up following a power failure, the READ bit will automatically be set to a '1'. This will prevent the clock from updating the TIME-

KEEPER registers, and will allow the user to read the exact time of the power-down event.

Resetting the READ Bit to a '0' will allow the clock to update these registers with the current time. The WRITE Bit will be reset to a '0' upon powerup.

#### Stopping and Starting the Oscillator

The oscillator may be stopped at any time. If the device is going to spend a significant amount of time on the shelf, the oscillator can be turned off to minimize current drain on the battery. The STOP bit is located at Bit D7 within the Seconds Register (9h). Setting it to a '1' stops the oscillator. When reset to a '0', the M48T212A oscillator starts within one second

**Note:** It is not necessary to set the WRITE bit when setting or resetting the FREQUENCY TEST bit (FT) or the STOP bit (ST).

#### **SETTING ALARM CLOCK REGISTERS**

Address locations 6h-2h contain the alarm settings. The alarm can be configured to go off at a prescribed time on a specific month, date, hour, minute, or second or repeat every year, month, day, hour, minute, or second. It can also be programmed to go off while the M48T212A is in the battery back-up to serve as a system wake-up call. Bits RPT5-RPT1 put the alarm in the repeat mode of operation. Table 12 shows the possible configurations. Codes not listed in the table default to the once per second mode to quickly alert the user of an incorrect alarm setting.

When the clock information matches the alarm clock settings based on the match criteria defined by RPT5-RPT1, the AF (Alarm Flag) is set.

If AFE (Alarm Flag Enable) is also set, the alarm condition activates the IRQ/FT pin. The IRQ/FT output is cleared by a read to the Flags register as shown in Figure 9. A subsequent read of the Flags register will reset the Alarm Flag (D6; Register 0h). The IRQ/FT pin can also be activated in the battery back-up mode. The IRQ/FT will go low if an alarm occurs and both ABE (Alarm in Battery Back-up Mode Enable) and AFE are set. The ABE and AFE bits are reset during power-up, therefore an alarm generated during power-up will only set AF. The user can read the Flag Register at system boot-up to determine if an alarm was generated while the M48T212A was in the deselect mode during power-up. Figure 10 illustrates the back-up mode alarm timing.

*5*77

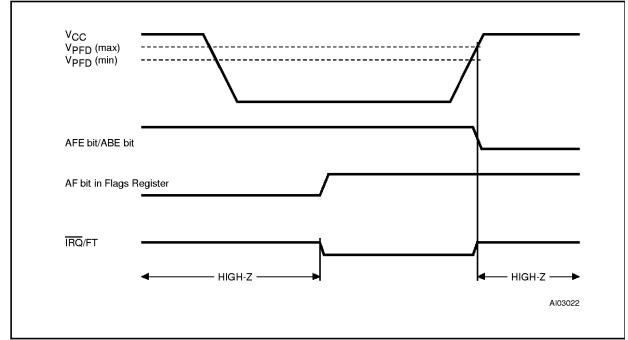


Figure 10. Back-Up Mode Alarm Waveforms

#### **WATCHDOG TIMER**

of-control microprocessor. The user programs the watchdog timer by setting the desired amount of time-out into the Watchdog Register, address 7h. Bits BMB4-BMB0 store a binary multiplier and the two lower order bits RB1-RB0 select the resolution, where 00=1/16 second, 01=1/4 second, 10=1 second, and 11=4 seconds. The amount of timeout is then determined to be the multiplication of the five bit multiplier value with the resolution. (For example: writing 00001110 in the Watchdog Register = 3\*1 or 3 seconds). If the processor does not reset the timer within the specified period, the M48T212A sets the WDF (Watchdog Flag) and generates a watchdog interrupt or a microproces-

The watchdog timer can be used to detect an out-

The most significant bit of the Watchdog Register is the Watchdog Steering Bit (WDS). When set to a '0', the watchdog will activate the  $\overline{\text{IRQ}}/\text{FT}$  pin when timed-out. When WDS is set to a '1', the watchdog will output a negative pulse on the RST pin for 40 to 200 ms. The Watchdog register and the FT bit will reset to a '0' at the end of a Watchdog time-out when the WDS bit is set to a '1'.

The watchdog timer can be reset by two methods:

- 1. a transition (high-to-low or low-to-high) can be applied to the Watchdog Input pin (WDI) or
- 2. the microprocessor can perform a write of the Watchdog Register.

The time-out period then starts over. The WDI pin should be tied to V<sub>SS</sub> if not used. The watchdog will be reset on each transition (edge) seen by the WDI pin. In the order to perform a software reset of the watchdog timer, the original time-out period can be written into the Watchdog Register, effectively restarting the count-down cycle.

Should the watchdog timer time-out, and the WDS bit is programmed to output an interrupt, a value of 00h needs to be written to the Watchdog Register in order to clear the  $\overline{\text{IRQ}}/\text{FT}$  pin. This will also disable the watchdog function until it is again programmed correctly. A read of the Flags Register will reset the Watchdog Flag (Bit D7; Register 0h).

The watchdog function is automatically disabled upon power-up and the Watchdog Register is cleared. If the watchdog function is set to output to the  $\overline{IRQ}/FT$  pin and the frequency test function is activated, the watchdog function prevails and the frequency test function is denied.

sor reset.

## **VCC SWITCH OUTPUT**

 $\overline{V_{CCSW}}$  output goes low when  $V_{OUT}$  switches to  $V_{CC}$  turning on a customer supplied P-Channel MOSFET (see Figure 3). The Motorola MTD20P06HDL is recommended. This MOSFET in turn connects  $V_{OUT}$  to a separate supply when the current requirement is greater than  $I_{OUT1}$  (see Table 7). This output may also be used simply to indicate the status of the internal battery switchover comparator, which controls the source ( $V_{CC}$  or battery) of the  $V_{OUT}$  output.

#### **POWER-ON RESET**

The M48T212A continuously monitors  $V_{CC}$ . When  $V_{CC}$  falls to the power fail detect trip point, the  $\overline{RST}$  pulls low (open drain) and remains low on powerup for 40 to 200ms after  $V_{CC}$  passes  $V_{PFD}$ . The  $\overline{RST}$  pin is an open drain output and an appropriate pull-up resistor to  $V_{CC}$  should be chosen to control rise time.

**Note**: If the  $\overline{RST}$  output is fed back into either of the  $\overline{RSTIN}$  inputs (for a microprocessor with a bidirectional reset) then a  $1k\Omega$  (max) pull-up resistor is recommended.

### Reset Inputs (RSTIN1 & RSTIN2)

The M48T212A provides two independent inputs which can generate an output reset. The duration and function of these resets is identical to a reset generated by a power cycle. Table 14 and Figure 12 illustrate the AC reset characteristics of this function. During the time RST is enabled (t<sub>R1HRH</sub> & t<sub>R2HRH</sub>), the Reset Inputs are ignored.

**Note:**  $\overline{RSTIN1}$  and  $\overline{RSTIN2}$  are each internally pulled up to  $V_{CC}$  through a 100K $\Omega$  resistor.

## Calibrating the Clock

The M48T212A is driven by a quartz controlled oscillator with a nominal frequency of 32,768 Hz. The devices are factory calibrated at 25°C and tested for accuracy. Clock accuracy will not exceed  $\pm 35$  ppm (parts per million) oscillator frequency error at 25°C, which equates to about  $\pm 1.53$  minutes per month. When the Calibration circuit is properly employed, accuracy improves to better than  $\pm 1/-2$  ppm at 25°C.

The oscillation rate of crystals changes with temperature. The M48T212A design employs periodic counter correction. The calibration circuit adds or subtracts counts from the oscillator divider circuit at the divide by 256 stage, as shown in Figure 11. The number of times pulses which are blanked (subtracted, negative calibration) or split (added, positive calibration) depends upon the value loaded into the five Calibration bits found in the Control Register. Adding counts speeds the clock up, subtracting counts slows the clock down.

The Calibration bits occupy the five lower order bits (D4-D0) in the Control Register 8h. These bits

can be set to represent any value between 0 and 31 in binary form. Bit D5 is a Sign bit; '1' indicates positive calibration, '0' indicates negative calibration. Calibration occurs within a 64 minute cycle. The first 62 minutes in the cycle may, once per minute, have one second either shortened by 128 or lengthened by 256 oscillator cycles.

If a binary '1' is loaded into the register, only the first 2 minutes in the 64 minute cycle will be modified; if a binary 6 is loaded, the first 12 will be affected, and so on.

Therefore, each calibration step has the effect of adding 512 or subtracting 256 oscillator cycles for every 125,829,120 actual oscillator cycles, that is +4.068 or -2.034 ppm of adjustment per calibration step in the calibration register. Assuming that the oscillator is running at exactly 32,768 Hz, each of the 31 increments in the Calibration byte would represent +10.7 or -5.35 seconds per month which corresponds to a total range of +5.5 or -2.75 minutes per month.

Two methods are available for ascertaining how much calibration a given M48T212A may require. The first involves setting the clock, letting it run for a month and comparing it to a known accurate reference and recording deviation over a fixed period of time. Calibration values, including the number of seconds lost or gained in a given period, can be found in Application Note AN934: TIMEKEEPER Calibration.

This allows the designer to give the end user the ability to calibrate the clock as the environment requires, even if the final product is packaged in a non-user serviceable enclosure. The designer could provide a simple utility that accesses the Calibration byte.

The second approach is better suited to a manufacturing environment, and involves the use of the  $\overline{IRQ}/FT$  pin. The pin will toggle at 512Hz, when the Stop bit (ST, D7 of 9h) is '0', the Frequency Test bit (FT, D6 of Ch) is '1', the Alarm Flag Enable bit (AFE, D7 of 6h) is '0', and the Watchdog Steering bit (WDS, D7 of 7h) is '1' or the Watchdog Register (7h = 0) is reset.

Any deviation from 512 Hz indicates the degree and direction of oscillator frequency shift at the test temperature. For example, a reading of 512.010124 Hz would indicate a +20 ppm oscillator frequency error, requiring a -10 (WR001010) to be loaded into the Calibration Byte for correction. Note that setting or changing the Calibration Byte does not affect the Frequency test output frequency.

The  $\overline{IRQ}/FT$  pin is an open drain output which requires a pull-up resistor to  $V_{CC}$  for proper operation. A 500-10k $\Omega$  resistor is recommended in order to control the rise time. The FT bit is cleared on power-up.

**Table 14. Reset AC Characteristics** 

 $(T_A = 0 \text{ to } 70^{\circ}\text{C}; V_{CC} = 3\text{V to } 3.6\text{V})$ 

Symbol	Parameter	Min	Max	Unit
t <sub>R1</sub> <sup>(1)</sup>	RSTIN1 Low to RSTIN1 High	200		ns
t <sub>R2</sub> (2)	RSTIN2 Low to RSTIN2 High	100		ms
t <sub>R1HRH</sub> (3)	RSTIN1 High to RST High	40	200	ms
t <sub>R2HRH</sub> (3)	RSTIN2 High to RST High	40	200	ms

Note: 1. Pulse width less than 50ns will result in no RESET (for noise immunity).

2. Pulse width less than 20ms will result in no RESET (for noise immunity).

3.  $C_L = 5pF$  (see Figure 4).

Table 15. Crystal Electrical Characteristics (Externally Supplied)

Symbol	Description	Min	Тур	Max	Unit
fo	Resonant Frequency		32,768		kHz
R <sub>S</sub>	Series Resistance		50	70	kΩ
CL	Load Capacitance		12.5		pF

Note: Load capacitors are integrated within the M48T212A. Circuit board layout considerations for the 32kHz crystal of minimum trace lengths and isolation from RF generating signals should be taken into account. ST recommends the KDS DT-38 Tuning Fork Type quartz crystal for all temperature operations. KDS can be contacted at 913-491-6825 or at <a href="http://www.kdsj.co.jp">http://www.kdsj.co.jp</a> for forther information on this crystal type.

#### **BATTERY LOW WARNING**

The M48T212A automatically performs battery voltage monitoring upon power-up and at factory-programmed time intervals of approximately 24 hours. The Battery Low (BL) bit, Bit D4 of Flags Register 0h, will be asserted if the battery voltage is found to be less than approximately 2.5V. The BL bit will remain asserted until completion of battery replacement and subsequent battery low monitoring tests, either during the next power-up sequence or the next scheduled 24-hour interval.

If a battery low is generated during a power-up sequence, this indicates that the battery is below approximately 2.5 volts and may not be able to maintain data integrity in the SRAM. Data should be considered suspect and verified as correct. A fresh battery should be installed.

If a battery low indication is generated during the 24-hour interval check, this indicates that the battery is near end of life. However, data is not compromised due to the fact that a nominal Vcc is supplied. In order to insure data integrity during

subsequent periods of battery back-up mode, the battery should be replaced. The battery should be replaced with  $V_{CC}$  powering the device to avoid data loss.

The M48T212A only monitors the battery when a nominal Vcc is applied to the device. Thus applications which require extensive durations in the battery back-up mode should be powered-up periodically (at least once every few months) in order for this technique to be beneficial.

Additionally, if a battery low is indicated, data integrity should be verified upon power-up via a checksum or other technique.

**Note**: Battery Low warning is only valid when using a 3V button cell battery. Use a super capacitor for back-up supply causes the BL flag to be invalid.

## **INITIAL POWER-ON DEFAULTS**

Upon application of power to the device, the following register bits are set to a '0' state: WDS, BMB0-BMB4, RB0-RB1, AFE, ABE, W and FT.

Figure 11. Calibration Waveform

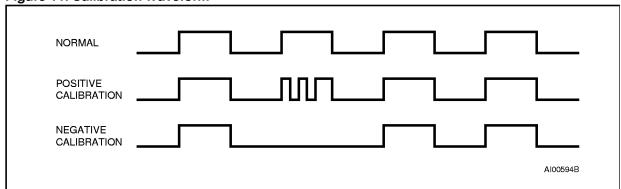
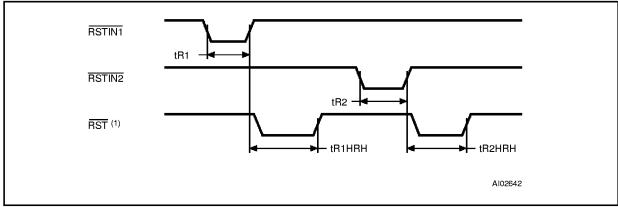
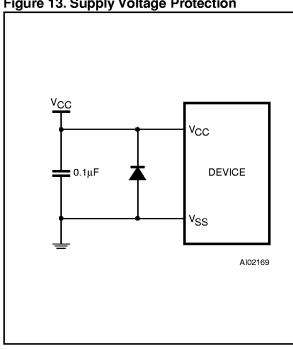


Figure 12. RSTIN1 & RSTIN2 Timing Waveforms



Note: 1. With pull-up resistor.

Figure 13. Supply Voltage Protection



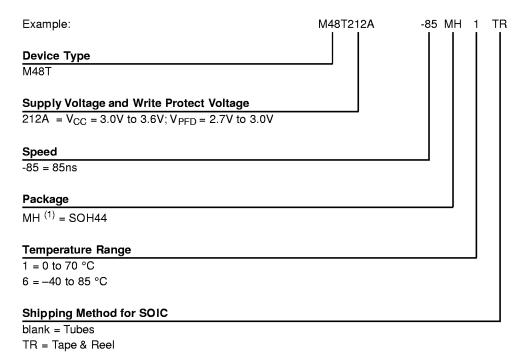
## **POWER SUPPLY DECOUPLING** AND UNDERSHOOT PROTECTION

Note: I<sub>CC</sub> transients, including those produced by output switching, can produce voltage fluctuations, resulting in spikes on the V<sub>CC</sub> bus. These transients can be reduced if capacitors are used to store energy, which stabilizes the V<sub>CC</sub> bus. The energy stored in the bypass capacitors will be released as low going spikes are generated or energy will be absorbed when overshoots occur.

A ceramic bypass capacitor value of 0.1µF is recommended in order to provide the needed filtering. In addition to transients that are caused by normal SRAM operation, power cycling can generate negative voltage spikes on V<sub>CC</sub> that drive it to values below V<sub>SS</sub> by as much as one volt. These negative spikes can cause data corruption in the SRAM while in battery backup mode.

To protect from these voltage spikes, ST recommends connecting a schottky diode from V<sub>CC</sub> to  $V_{\rm SS}$  (cathode connected to  $V_{\rm CC}$ , anode to  $V_{\rm SS}$ ). (Schottky diode 1N5817 is recommended for through hole and MBRS120T3 is recommended for surface mount).

# **Table 16. Ordering Information Scheme**



Note: 1. The SOIC package (SOH44) requires the battery package (SNAPHAT) which is ordered separately under the part number "M4Txx-BR12SH1" in plastic tube or "M4Txx-BR12SH1TR" in Tape & Reel form.

Caution: Do not place the SNAPHAT battery package "M4Txx-BR12SH1" in conductive foam since will drain the lithium button-cell battery.

For a list of available options (Speed, Package, etc...) or for further information on any aspect of this device, please contact the STMicroelectronics Sales Office nearest to you.

Table 17. SOH44 - 44 lead Plastic Small Outline, 4-socket battery, SNAPHAT, Package Mechanical Data

Symb		mm		inches			
	Тур	Min	Max	Тур	Min	Max	
А			3.05			0.120	
A1		0.05	0.36		0.002	0.014	
A2		2.34	2.69		0.092	0.106	
В		0.36	0.46		0.014	0.018	
С		0.15	0.32		0.006	0.012	
D		17.71	18.49		0.697	0.728	
E		8.23	8.89		0.324	0.350	
е	0.81	-	-	0.032	-	-	
eВ		3.20	3.61		0.126	0.142	
Н		11.51	12.70		0.453	0.500	
L		0.41	1.27		0.016	0.050	
α		0°	8°		0°	8°	
N		44			44		
СР			0.10			0.004	

Figure 14. SOH44 - 44 lead Plastic Small Outline, 4-socket battery, SNAPHAT, Package Outline

A

B

A

C

B

A

C

B

A

C

SOH-A

Drawing is not to scale.

Table 18. M4T28-BR12SH SNAPHAT Housing for 48 mAh Battery & Crystal, Package Mechanical Data

Symb		mm		inches			
	Тур	Min	Max	Тур	Min	Max	
Α			9.78			0.385	
A1		6.73	7.24		0.265	0.285	
<b>A</b> 2		6.48	6.99		0.255	0.275	
А3			0.38			0.015	
В		0.46	0.56		0.018	0.022	
D		21.21	21.84		0.835	0.860	
Е		14.22	14.99		0.560	0.590	
eA		15.55	15.95		0.612	0.628	
eB		3.20	3.61		0.126	0.142	
L		2.03	2.29		0.080	0.090	

Drawing is not to scale.

Table 19. M4T32-BR12SH SNAPHAT Housing for 120 mAh Battery & Crystal, Package Mechanical Data

Symb		mm		inches			
	Тур	Min	Max	Тур	Min	Max	
А			10.54			0.415	
A1		8.00	8.51		0.315	.0335	
A2		7.24	8.00		0.285	0.315	
А3			0.38			0.015	
В		0.46	0.56		0.018	0.022	
D		21.21	21.84		0.835	0.860	
E		17.27	18.03		0.680	.0710	
eA		15.55	15.95		0.612	0.628	
eB		3.20	3.61		0.126	0.142	
L		2.03	2.29		0.080	0.090	

Drawing is not to scale.