

FEATURES

Functional Complete DAS
Single Package
Monolithic
High Impedance Differential Inputs
Guaranteed Low 1 nA Input Bias Current
Guaranteed 80 dB Common-Mode Rejection
External Selectable Bandpass Filter Frequencies
Software Programmable Gain Selection
12-Bit A/D Converter with On-Chip Reference
Serial Communication Interface
Data Sampling 40,000 Samples/Second
 ± 5 V Supplies
Low 175 mW (typ) Power Consumption
Small 28-Terminal Surface Mount Package (PLCC)

APPLICATION

Small Signal Data Acquisition
ECG Signal Data Acquisition
Patient Monitoring

GENERAL DESCRIPTION

The AD79015 is a complete data acquisition system for very small signals (i.e., biomedical ECG) with a data sampling rate of minimum 40,000 samples/sec. It provides high accuracy, high stability, and functional completeness in a single 28-pin package.

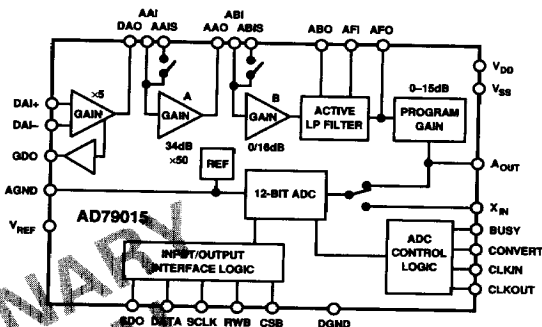
It includes a high performance instrumentation amplifier at the front-end, bandpass filter, and an accurate 12-bit ADC with on-chip reference.

An on-chip clock circuit is provided, which may be used with a crystal for stand-alone operation. Alternatively, the clock input may be driven from an external CMOS-compatible clock source such as a microprocessor clock.

The AD79015 serial interface is compatible with many microprocessors and digital signal processors such as the ADSP-2100, TMS32020, μ PD7720, and DSP-56000. It can also be used with general purpose serial to parallel converters such as shift registers.

The AD79015 is fabricated in Analog Devices' linear compatible CMOS process (LC²MOS), an advanced, all ion-implanted process that combines fast CMOS logic and linear, bipolar circuits on a single chip, thus achieving excellent linear performance while still retaining low CMOS power levels.

FUNCTIONAL BLOCK DIAGRAM



PRODUCT HIGHLIGHTS

1. Single chip, complete small signal DAS. It includes a high performance differential front end amplifier, programmable gain stages, externally controlled high and low corner frequencies, and a 12-bit AD converter with on-chip reference.
2. Input amplifier has extremely low input bias current of <1 nA over full temperature range. Typical input bias current at ambient temperature is 20 pA.
3. On-chip guard driver to minimize external components.
4. Software programmable gain setting with a gain range of 0 dB to 31 dB.
5. On-chip clock oscillator to minimize external components.
6. A serial interface is provided to make it easy to use AD79015 in applications where full isolation from the mains power is required.
7. Serial interface supports multichannel applications with minimal external components.
8. LC²MOS circuitry gives low power drain (175 mW typ) from +5 V, and -5 V supplies.

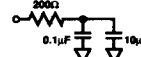
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AD79015—SPECIFICATIONS ($V_{DD} = +5\text{ V} \pm 5\%$, and $V_{SS} = -5\text{ V} \pm 5\%$. All specifications T_{MIN} to T_{MAX} unless otherwise noted.)

Parameter	All Grades	Units	Test Condition/Comments
INPUT AMPLIFIER			
GAIN			
Gain Range	14 5	dB V/V	Over Full Temperature Range
Gain Error	± 0.1	dB max	
VOLTAGE OFFSET			
Input Offset Voltage	10	mV max	
INPUT CURRENT			
Input Bias Current	1	nA max	Over Full Temperature Range Typical at +25°C
Typical Input Bias	20	pA	
INPUT			
Input Resistance	10^9	Ω min	
Input Capacitance	6	pF typ	
Differential Input Voltage Range			
Gain = 5 (DC Coupled)	± 0.5	V max	
Common-Mode Input Voltage	± 0.5	V max	
Common-Mode Rejection Ratio	80	dB min	
NOISE			
Voltage Noise (RTI)	2 4	$\mu\text{V p-p typ}$ $\mu\text{V p-p typ}$	Bandwidth 1 Hz–10 Hz @ +25°C Bandwidth 0.1 Hz–100 Hz @ +25°C Assume Gaussian Noise $-V_{p-p} = 6.6 \times V_{rms}$, 1% Probability of Error
GUARD DRIVER			
Capacitive Load	100	pF max	
Resistive Load	2	k Ω min	
AMPLIFIER A			
Gain	34 50 0.1	dB V/V dB max	
Gain Accuracy	1.2	%	
Input Offset Voltage	2	mV max	
Input Bias Current	5	nA typ	
AMPLIFIER B/LOW PASS FILTER			
Gain			
Low	0 1.0	dB V/V	of Absolute Value, over Full Temperature Range
High	16 6.3	dB V/V	
Gain Error	0.2 2.4	dB max %	
Input Offset	2	mV max	
Resistors in Network	5	% max	
PROGRAMMABLE GAIN AMPLIFIER			
Gain			
Minimum Gain	0 1.0	dB V/V	
Maximum Gain	15 5.6	dB V/V	
Gain Step Size	1 12.2	dB %	
Gain Accuracy	0.1 1.0	dB max V/V	
Input Offset Voltage	2	mV max	

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Parameter	All Grades	Units	Test Condition/Comments
CONVERTER (Core Cell Is the AD7870)			
DYNAMIC PERFORMANCE			
Signal-to-Noise Ratio (SNR)	68	dB	min, $V_{IN} = 1$ kHz Sine, $f_{SAMPLE} = 10$ kHz
Total Harmonic Distortion	-80	dB	max, $V_{IN} = 1$ kHz Sine, $f_{SAMPLE} = 10$ kHz
No Missed Codes	Guaranteed		
Track/Hold Acquisition Time	2	μs	max
Conversion Time	13.25	μs	@ 4 MHz Clock Frequency
DC ACCURACY			
Resolution	12	Bits	
Relative Accuracy	± 1	LSB	Typical
DNL	± 1	LSB	Typical
Bipolar Zero Offset	± 1	%	@ $V_{IN} = 0$ V
Full-Scale Error	± 1	%	@ $V_{IN} = -3$ V or $V_{IN} = +3$ V, Relative to Reference
ANALOG INPUT			
Input Voltage Range	± 3	V	
Input Current	± 550	μA	max
REFERENCE OUTPUT			
Reference Voltage @ +25°C	2.98/3.02	V min/V max	
Reference Tempco	± 40	ppm/V	AC Decoupling Required
DIGITAL INTERFACE			
INPUTS			
Logic "1" Voltage	+2.0	V min	$V_{IN} = 0$ to V_{DD}
Logic "0" Voltage	+0.4	V max	
Input Current	± 10	μA max	
Input Capacitance	10	pF max	
OUTPUTS			
Logic "1" Voltage	+2.0	V min	$I_{SOURCE} = 200 \mu A$ $I_{SINK} = 1.6$ mA
Logic "0" Voltage	+0.4	V max	
Floating State Leakage	± 10	μA max	
Floating State Capacitance	15	pF max	
POWER REQUIREMENTS			
V_{DD}	+5	V nominal	$\pm 5\%$ for Specified Performance
V_{SS}	-5	V nominal	$\pm 5\%$ for Specified Performance
I_{DD}	25	mA max	
I_{SS}	25	mA max	
Power Dissipation	175	mW typ	
TEMPERATURE RANGE (T_{MIN} to T_{MAX})			
	0 to +70	°C	



PRELIMINARY
TECHNICAL
DATA

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ABSOLUTE MAXIMUM RATINGS*(T_A = +25°C unless otherwise noted) V_{DD} to DGND -0.3 V to +7 V V_{SS} to DGND +0.3 V to -7 VAGND to DGND -0.3 V to $V_{DD} + 0.3$ VAnalog Input Voltage to AGND V_{SS} to V_{DD} Digital Input Voltage to DGND -0.3 V to $V_{DD} + 0.3$ VDigital Output Voltage to DGND -0.3 V to $V_{DD} + 0.3$ V

Operating Temperature Range 0°C to +70°C

Storage Temperature -65°C to +150°C

Power Dissipation (Any Package) to +75°C 450 mW

Derates above +75°C by 6 mW/°C

*Stress above those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. This is a stress rating only and functional operation of the device at these or any other condition above those indicated in the operational section of this specification is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability. Only one Absolute Maximum Rating may be applied at any one time.

CAUTION

ESD (electrostatic discharge) sensitive device. The digital control inputs are diode protected; however, permanent damage may occur on unconnected devices subject to high energy electrostatic fields. Unused devices must be stored in conductive foam or shunts. The protective foam should be discharged to the destination socket before devices are removed.

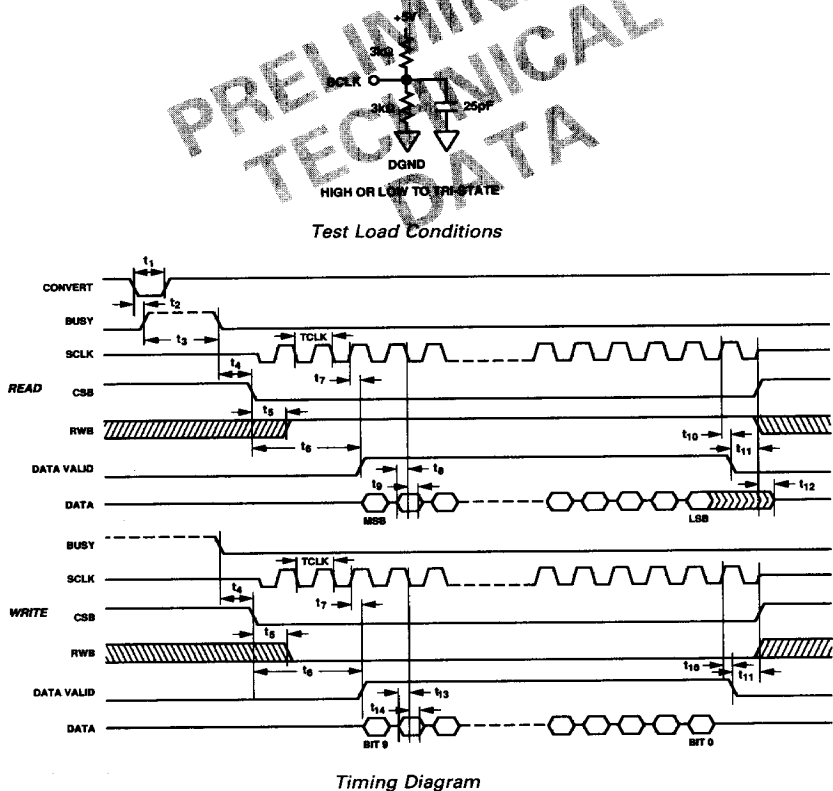


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TIMING CHARACTERISTICS ($V_{DD} = +5\text{ V}$, $V_{SS} = -5\text{ V}$, Ambient Temperature $+25^{\circ}\text{C}$)

Parameter	Limit	Units	Conditions/Comments
TIMING			
t_{CLK}	10	μs max	Clock Cycle Time 100 kHz Clock
	250	ns min	Clock Cycle Time 4 MHz Clock
t_1	t_{CLK}	ns min	ADC Start Convert Pulse Width
t_2	80	ns max	CONVERT \downarrow to BUSY \uparrow
t_3	$53^* t_{CLK}$	ns max	ADC Busy Period
t_4	0	ns min	BUSY \downarrow to CSB \downarrow
t_5	$2.5 t_{CLK}$	ns max	CSB \downarrow to RWB
t_6	$2.5 t_{CLK}$	ns min	CSB \downarrow to DATA VALID \uparrow
	$4.5 t_{CLK}$	ns max	CSB \downarrow to DATA VALID \uparrow
t_7	30	ns max	SCLK \uparrow to DATA VALID \uparrow
t_8	30	ns min	Data at Output before SCLK \downarrow
t_9	50	ns min	Data at Output after SCLK \downarrow
t_{10}	30	ns max	SCLK \uparrow to DATA VALID \downarrow
t_{11}	0	ns min	DATA VALID \downarrow to CSB/RWB \uparrow
t_{12}	50	ns max	CSB \uparrow to Data and SCLK Float (See diagram below.)
t_{13}	0	ns min	Data Setup Time before SCLK \downarrow
t_{14}	$0.5 t_{CLK}$	ns min	Data Hold Time after SCLK

*The Internal Logic is dynamic so must be continuously clocked at 100 kHz minimum.



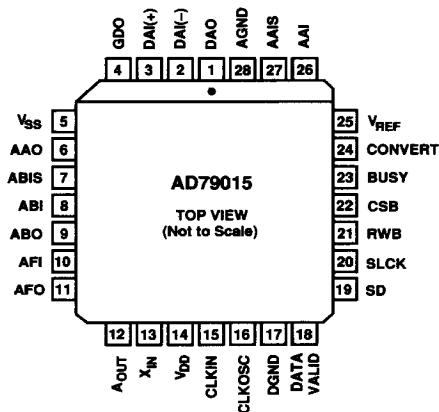
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PIN FUNCTION DESCRIPTION

Pin	Mnemonic	Description
1	DAO	Differential Amplifier Out
2	DAI(-)	Differential Amplifier In (-)
3	DAI(+)	Differential Amplifier In (+)
4	GDO	Guard Drive Output
5	V _{SS}	Negative Supply, -5 V
6	AAO	Amplifier A Out
7	ABIS	Amplifier B In (Switched)
8	ABI	Amplifier B In
9	ABO	Amplifier B Out
10	AFI	Active Filter In (-)
11	AFO	Active Filter Out
12	A _{OUT}	Analog Out
13	X _{IN}	External ADC Input
14	V _{DD}	Positive Supply, +5 V
15	CLKIN	Clock Input Pin. An external TTL compatible clock may be applied to this pin
16	CLKOSC	Clock Oscillator Pin
17	DGND	Digital Ground
18	DATA VALID	This pin signals valid data in/out during SCLK low-high transition
19	SDO	Serial Data In/Out. This pin is in tristate when CSB is high
20	SCLK	Serial Clock Output. This pin is in tristate when CSB is high
21	RWB	Read/Write Select
22	CSB	Chip Select
23	BUSY	Converter Busy
24	CONVERT	Start Conversion
25	V _{REF}	Voltage Reference Out
26	AAI	Amplifier A In
27	AAIS	Amplifier A In (Switched)
28	AGND	Analog Ground

PIN CONFIGURATION

PLCC



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AD79015

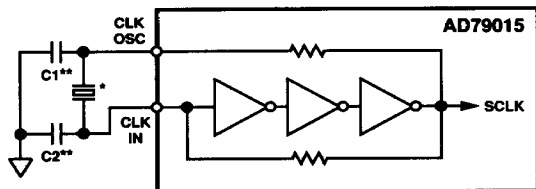
CIRCUIT INFORMATION

SIGNAL LEVELS

For an input gain of 5, the maximum Input Signal for Full-Scale ADC Output is ± 10 mV.

INTERNAL CLOCK OSCILLATOR

Figure 1 shows the AD79015 internal clock circuit. A crystal or ceramic resonator may be connected as in Figure 1 to provide a clock oscillator for the internal timing. Alternatively, the crystal/resonator may be omitted and an external CMOS-compatible clock source connected to CLKIN. The mark/space ratio of the external clock must be in the range of 45/55 and 55/45. An inverted CLKIN signal will appear at the SCLK output pin.



NOTES:

* 4MHz CRYSTAL/CERAMIC RESONATOR

** C1 AND C2 CAPACITANCE VALUES DEPEND ON CRYSTAL/CERAMIC RESONATOR MANUFACTURER. TYPICAL VALUES ARE FROM 30pF TO 100pF.

Figure 1. AD79015 Internal Clock Circuit

ACTIVE LOW-PASS FILTER

The internal active filter is implemented with a 2nd order negative feedback configuration.

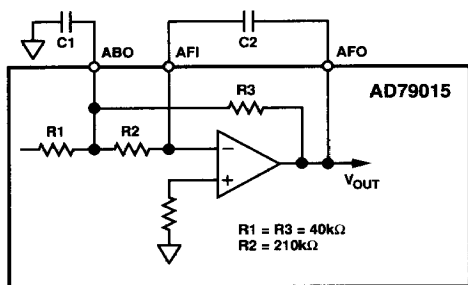


Figure 2. Low-Pass Filter

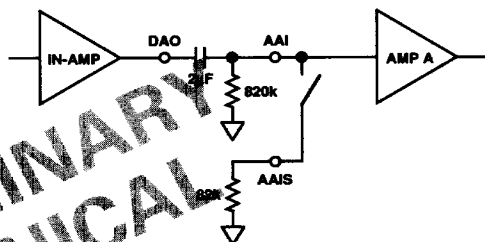
The filter cutoff frequency and filter damping factor are determined by selecting the appropriate values of C_1 and C_2 . The resistor value for R_1 and R_3 is $40\text{ k}\Omega$ and R_2 is $210\text{ k}\Omega$.

$$d = \sqrt{\frac{C_2}{C_1}} \left[\sqrt{\frac{R_2}{R_3}} + \sqrt{\frac{R_3}{R_2}} + \sqrt{\frac{R_2 \times R_3}{R_1}} \right] \text{ gain} = \frac{R_3}{R_1}$$

$$\text{Bandpass ripple } \frac{e_{OUT}}{e_{IN}} = -20 \log_{10} \left[\frac{d \sqrt{4 - d^2}}{2} \right]$$

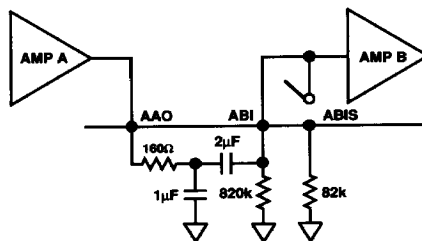
HIGH-PASS FILTER (Example Only)

This external high-pass filter can be implemented between the input gain stage and Amplifier A.



BAND PASS FILTER (Example Only)

This external band pass filter can be implemented between Amplifier A and Amplifier B.



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PROGRAMMING THE AD79015

The function of the part is set by writing a ten-bit word to the control register on chip using the serial interface. The timing for the write operation is provided in the timing diagrams.

The order and function of the bits in the control register is as follows:

- Bit 9 A "1" sets external input to the ADC.
- Bit 8 Internal use only. Must be set to a "1."
- Bit 7 Internal use only. Must be set to a "1."
- Bit 3–Bit 6 A 4-bit binary code to set the gain of the programmable gain block between 0 dB and 15 dB in steps of 1 dB.
"0000" is 0 dB and "1111" is 15 dB.
- Bit 2 A "0" sets the gain of 3rd stage to 0 dB.
A "1" sets the gain of 3rd stage to 16 dB.
- Bit 1 A "1" closes the internal switches at the inputs to the 2nd and 3rd stage amplifiers.
- Bit 0 Internal use only. Must be set to a "0."

Valid data available only after the first read from and write to the interface register

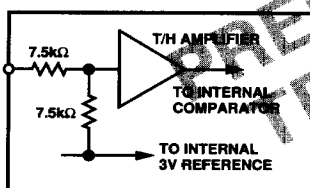


Figure 3. ADC Analog Input

ANALOG INPUT

Figure 3 shows the ADC analog input. The analog input range is 3 V into an input resistance of typically 15 kΩ. The designed code transition occurs midway between successive integer LSB values (i.e., 1/2 LSB, 3/2 LSBs, 5/2 LSBs . . . FS–3/2 LSBs). The output code is binary with 1 LSB = FS/4096 = 6 V/4096 = 1.46 mV. The ideal input/output transfer function is shown in Figure 4.

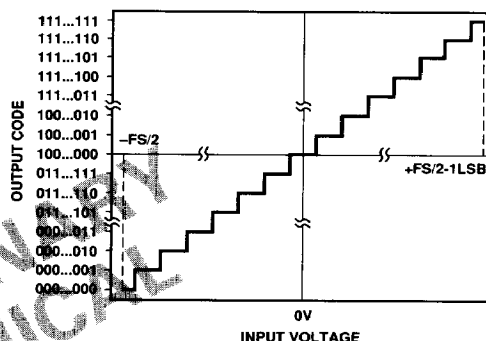


Figure 4. Bipolar Input/Output Transfer Function