

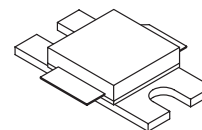
The RF MOSFET Line  
**RF Power Field Effect Transistors**  
N-Channel Enhancement-Mode Lateral MOSFETs

**MRF19030**  
**MRF19030S**

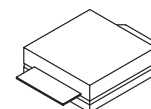
Designed for class AB PCN and PCS base station applications from 1.8 to 2.0 GHz. Suitable for FM, TDMA, CDMA and multicarrier amplifier applications.

**2.0 GHz, 30 W, 26 V**  
**LATERAL N-CHANNEL**  
**RF POWER MOSFETs**

- CDMA Performance @ 1990 MHz, 26 Volts  
IS-97 CDMA Pilot, Sync, Paging, Traffic Codes 8 Thru 13  
885 kHz — -47 dBc @ 30 kHz BW  
1.25 MHz — -55 dBc @ 12.5 kHz BW  
2.25 MHz — -55 dBc @ 1 MHz BW  
Output Power — 4.5 Watts (Avg.)  
Power Gain — 13.5 dB  
Efficiency — 17%
- Internally Matched, Controlled Q, for Ease of Use
- High Gain, High Efficiency and High Linearity
- Integrated ESD Protection
- Ease of Design for Gain and Insertion Phase Flatness
- Capable of Handling 10:1 VSWR, @ 26 Vdc, 1.93 GHz, 30 Watts CW Output Power
- Excellent Thermal Stability
- Characterized with Series Equivalent Large-Signal Impedance Parameters



**CASE 465E-02, STYLE 1**  
**(MRF19030)**



**CASE 465F-01, STYLE 1**  
**(MRF19030S)**

**MAXIMUM RATINGS**

| Rating   | Symbol    | Value        | Unit                         |
|--|-----------|--------------|------------------------------|
| Drain-Source Voltage   | $V_{DSS}$ | 65           | Vdc                          |
| Gate-Source Voltage  | $V_{GS}$  | +15, -0.5    | Vdc                          |
| Total Device Dissipation @ $T_C = 25^\circ\text{C}$<br>Derate above $25^\circ\text{C}$ | $P_D$     | 83.3<br>0.48 | Watts<br>W/ $^\circ\text{C}$ |
| Storage Temperature Range  | $T_{stg}$ | -65 to +200  | $^\circ\text{C}$             |
| Operating Junction Temperature   | $T_J$     | 200          | $^\circ\text{C}$             |

**ESD PROTECTION CHARACTERISTICS**

| Test Conditions  | Class        |
|------------------|--------------|
| Human Body Model | 2 (Typical)  |
| Machine Model    | M3 (Typical) |

**THERMAL CHARACTERISTICS**

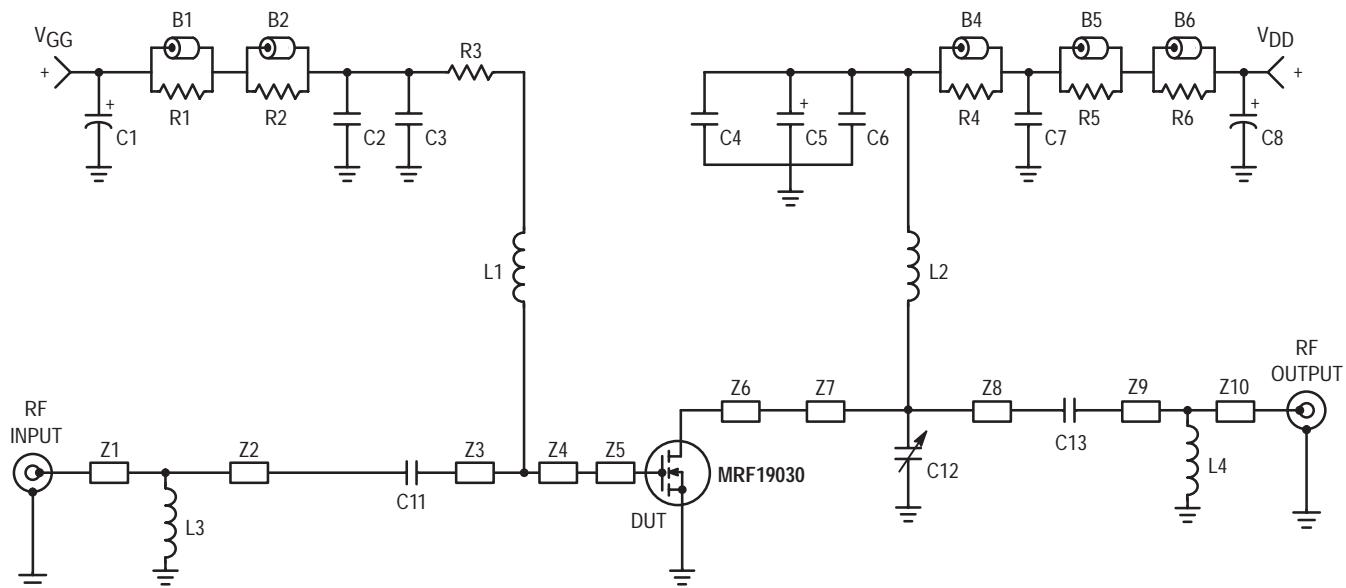
| Characteristic                       | Symbol          | Max | Unit                      |
|--------------------------------------|-----------------|-----|---------------------------|
| Thermal Resistance, Junction to Case | $R_{\theta JC}$ | 2.1 | $^\circ\text{C}/\text{W}$ |

NOTE – **CAUTION** – MOS devices are susceptible to damage from electrostatic charge. Reasonable precautions in handling and packaging MOS devices should be observed.

**ELECTRICAL CHARACTERISTICS** ( $T_C = 25^\circ\text{C}$  unless otherwise noted)

| Characteristic  | Symbol        | Min  | Typ  | Max | Unit            |
|---|---------------|--|------|-----|-----------------|
| <b>OFF CHARACTERISTICS</b>  |               |  |      |     |                 |
| Drain–Source Breakdown Voltage<br>( $V_{GS} = 0\text{ Vdc}$ , $I_D = 20\ \mu\text{A}$ )   | $V_{(BR)DSS}$ | 65   | —    | —   | Vdc             |
| Zero Gate Voltage Drain Current<br>( $V_{DS} = 28\text{ Vdc}$ , $V_{GS} = 0$ )  | $I_{DSS}$     | —  | —    | 1   | $\mu\text{Adc}$ |
| Gate–Source Leakage Current<br>( $V_{GS} = 5\text{ Vdc}$ , $V_{DS} = 0$ )   | $I_{GSS}$     | —  | —    | 1   | $\mu\text{Adc}$ |
| <b>ON CHARACTERISTICS</b>   |               |  |      |     |                 |
| Gate Threshold Voltage<br>( $V_{DS} = 10\text{ Vdc}$ , $I_D = 100\ \mu\text{Adc}$ )   | $V_{GS(th)}$  | 2  | 3    | 4   | Vdc             |
| Gate Quiescent Voltage<br>( $V_{DS} = 28\text{ Vdc}$ , $I_D = 300\text{ mA}$ )  | $V_{GS(Q)}$   | 2  | 3.3  | 4.5 | Vdc             |
| Drain–Source On–Voltage<br>( $V_{GS} = 10\text{ Vdc}$ , $I_D = 1\text{ Adc}$ )  | $V_{DS(on)}$  | —  | 0.29 | 0.4 | Vdc             |
| Forward Transconductance<br>( $V_{DS} = 10\text{ Vdc}$ , $I_D = 1\text{ Adc}$ )   | $g_{fs}$      | —  | 2    | —   | S               |
| <b>DYNAMIC CHARACTERISTICS</b>  |               |  |      |     |                 |
| Input Capacitance (Including Input Matching Capacitor in Package) (1)<br>( $V_{DS} = 26\text{ Vdc}$ , $V_{GS} = 0$ , $f = 1\text{ MHz}$ )   | $C_{iss}$     | —  | 98.5 | —   | pF              |
| Output Capacitance (1)<br>( $V_{DS} = 26\text{ Vdc}$ , $V_{GS} = 0$ , $f = 1\text{ MHz}$ )  | $C_{oss}$     | —  | 37   | —   | pF              |
| Reverse Transfer Capacitance<br>( $V_{DS} = 26\text{ Vdc}$ , $V_{GS} = 0$ , $f = 1\text{ MHz}$ )  | $C_{rss}$     | —  | 1.3  | —   | pF              |
| <b>FUNCTIONAL TESTS</b> (In Motorola Test Fixture)  |               |  |      |     |                 |
| Two–Tone Common–Source Amplifier Power Gain<br>( $V_{DD} = 26\text{ Vdc}$ , $P_{out} = 30\text{ W PEP}$ , $I_{DQ} = 300\text{ mA}$ ,<br>$f_1 = 1960.0\text{ MHz}$ , $f_2 = 1960.1\text{ MHz}$ )   | $G_{ps}$      | —  | 13   | —   | dB              |
| Two–Tone Drain Efficiency<br>( $V_{DD} = 26\text{ Vdc}$ , $P_{out} = 30\text{ W PEP}$ , $I_{DQ} = 300\text{ mA}$ ,<br>$f_1 = 1960.0\text{ MHz}$ , $f_2 = 1960.1\text{ MHz}$ )   | $\eta$        | —  | 36   | —   | %               |
| 3rd Order Intermodulation Distortion<br>( $V_{DD} = 26\text{ Vdc}$ , $P_{out} = 30\text{ W PEP}$ , $I_{DQ} = 300\text{ mA}$ ,<br>$f_1 = 1960.0\text{ MHz}$ , $f_2 = 1960.1\text{ MHz}$ )  | IMD           | —  | –31  | —   | dBc             |
| Input Return Loss<br>( $V_{DD} = 26\text{ Vdc}$ , $P_{out} = 30\text{ W PEP}$ , $I_{DQ} = 300\text{ mA}$ ,<br>$f_1 = 1960.0\text{ MHz}$ , $f_2 = 1960.1\text{ MHz}$ )   | IRL           | —  | –13  | —   | dB              |
| Two–Tone Common–Source Amplifier Power Gain<br>( $V_{DD} = 26\text{ Vdc}$ , $P_{out} = 30\text{ W PEP}$ , $I_{DQ} = 300\text{ mA}$ , $f_1 = 1930.0\text{ MHz}$ ,<br>$f_2 = 1930.1\text{ MHz}$ and $f_1 = 1990.0\text{ MHz}$ , $f_2 = 1990.1\text{ MHz}$ ) | $G_{ps}$      | 12   | 13   | —   | dB              |
| Two–Tone Drain Efficiency<br>( $V_{DD} = 26\text{ Vdc}$ , $P_{out} = 30\text{ W PEP}$ , $I_{DQ} = 300\text{ mA}$ , $f_1 = 1930.0\text{ MHz}$ ,<br>$f_2 = 1930.1\text{ MHz}$ and $f_1 = 1990.0\text{ MHz}$ , $f_2 = 1990.1\text{ MHz}$ )                   | $\eta$        | 33   | 36   | —   | %               |
| 3rd Order Intermodulation Distortion<br>( $V_{DD} = 26\text{ Vdc}$ , $P_{out} = 30\text{ W PEP}$ , $I_{DQ} = 300\text{ mA}$ , $f_1 = 1930.0\text{ MHz}$ ,<br>$f_2 = 1930.1\text{ MHz}$ and $f_1 = 1990.0\text{ MHz}$ , $f_2 = 1990.1\text{ MHz}$ )        | IMD           | —  | –31  | –28 | dBc             |
| Input Return Loss<br>( $V_{DD} = 26\text{ Vdc}$ , $P_{out} = 30\text{ W PEP}$ , $I_{DQ} = 300\text{ mA}$ , $f_1 = 1930.0\text{ MHz}$ ,<br>$f_2 = 1930.1\text{ MHz}$ and $f_1 = 1990.0\text{ MHz}$ , $f_2 = 1990.1\text{ MHz}$ )                           | IRL           | —  | –13  | –9  | dB              |
| Output Mismatch Stress<br>( $V_{DD} = 26\text{ Vdc}$ , $P_{out} = 30\text{ W CW}$ , $I_{DQ} = 300\text{ mA}$ ,<br>$f = 1930\text{ MHz}$ , $V_{SWR} = 10:1$ , All Phase Angles at Frequency of Tests)  | $\Psi$        | No Degradation In Output Power Before and After Test |      |     |                 |

(1) Part is internally matched both on input and output.



|          |   |       |                            |
|----------|---|-------|----------------------------|
| B1 – B6  | Ferrite Bead, Fair Rite #2743019447                               | Z1    | 0.595" x 0.080" Microstrip |
| C1, C8   | 470 $\mu$ F, 63 V, Electrolytic Capacitor, Panasonic #ECEV1HV100R | Z2    | 0.600" x 0.080" Microstrip |
| C2, C7   | 0.10 $\mu$ F, RF Chip Capacitor, B Case, Kemet                    | Z3    | 0.480" x 0.080" Microstrip |
| C3       | 5.1 pF, RF Chip Capacitor, B Case, ATC                            | Z4    | 0.280" x 0.325" Microstrip |
| C4       | 5.1 pF, RF Chip Capacitor, B Case, ATC                            | Z5    | 0.200" x 0.510" Microstrip |
| C5       | 22 $\mu$ F, 35 V, Tantalum Surface Mount Chip Capacitor, Sprague  | Z6    | 0.200" x 0.510" Microstrip |
| C6       | 91 pF, RF Chip Capacitor, B Case, ATC                             | Z7    | 0.280" x 0.325" Microstrip |
| C11, C13 | 10 pF, RF Chip Capacitor, B Case, ATC                             | Z8    | 0.480" x 0.080" Microstrip |
| C12      | 0.4 – 2.5 pF, Variable Capacitor, Johanson Gigatrim               | Z9    | 0.530" x 0.080" Microstrip |
| L1 – L4  | 8.0 nH Inductors, 3 Turn, Coilcraft                               | Z10   | 0.720" x 0.080" Microstrip |
| R1, R2   | 12 $\Omega$ , Fixed Film Chip Resistor, 0.08" x 0.13"             | Board | 0.030" Glass Teflon® Arlon |
| R3       | 3.75 $\Omega$ , Fixed Film Chip Resistor, 0.08" x 0.13"           |       | GX-0300-55-22, 2 oz. Cu    |
| R4 – R6  | 10 $\Omega$ , Fixed Film Chip Resistor, 0.08" x 0.13"             |       |                            |

Figure 1. MRF19030 Schematic

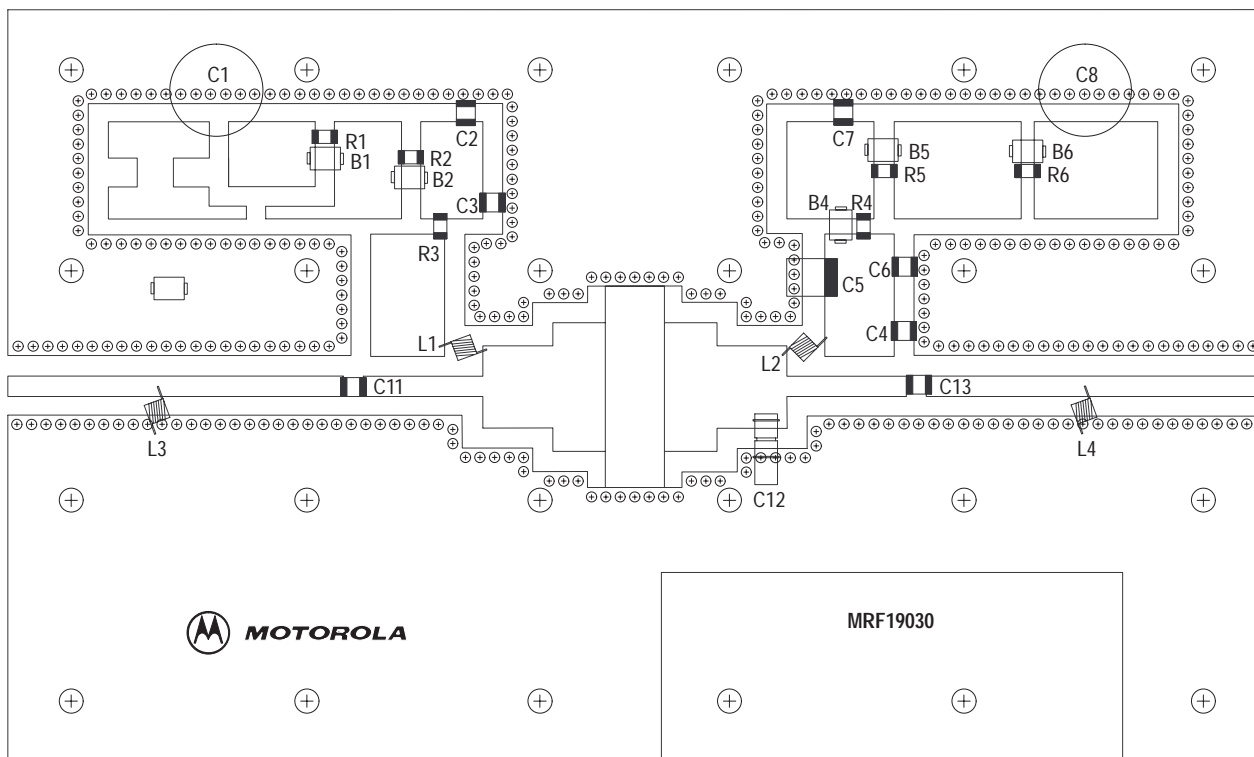
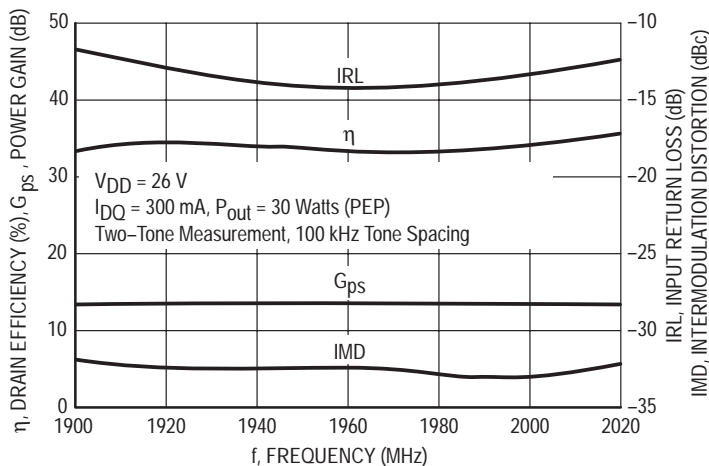
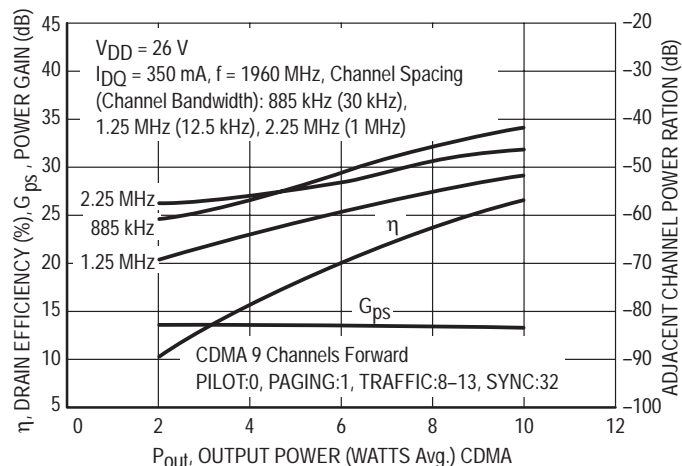


Figure 2. MRF19030 Populated PC Board Layout Diagram

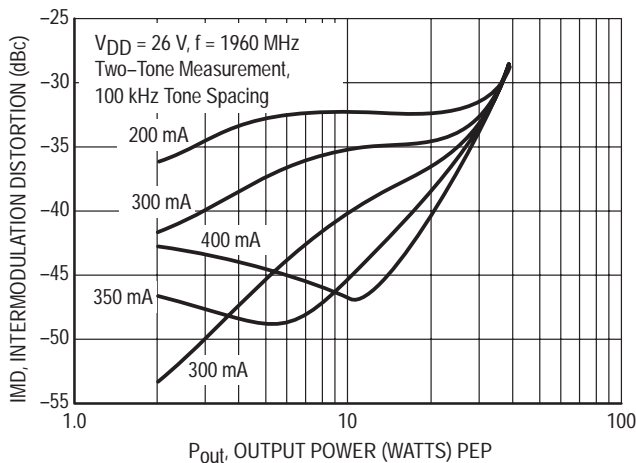
## TYPICAL CHARACTERISTICS



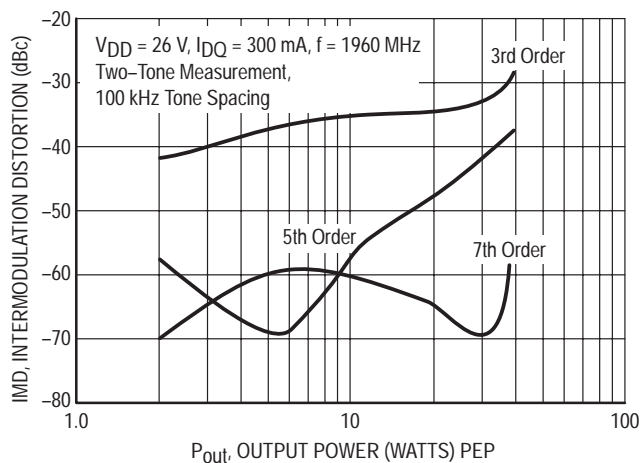
**Figure 3. Class AB Broadband Circuit Performance**



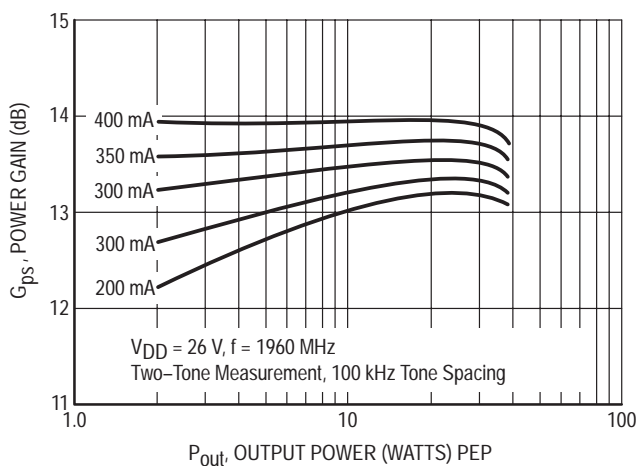
**Figure 4. CDMA ACPR, Power Gain and Drain Efficiency versus Output Power**



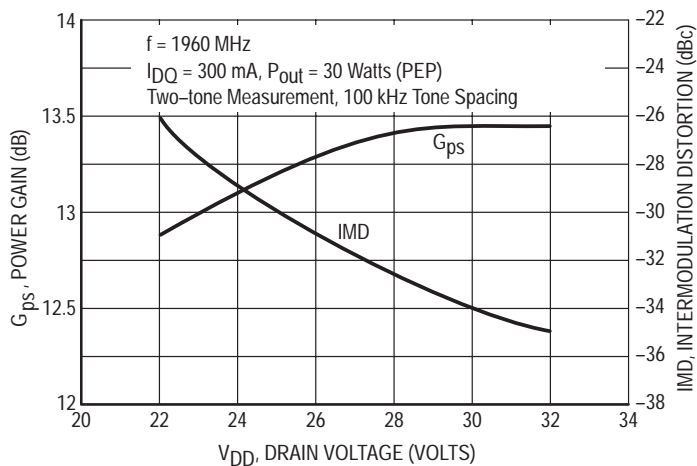
**Figure 5. Intermodulation Distortion versus Output Power**



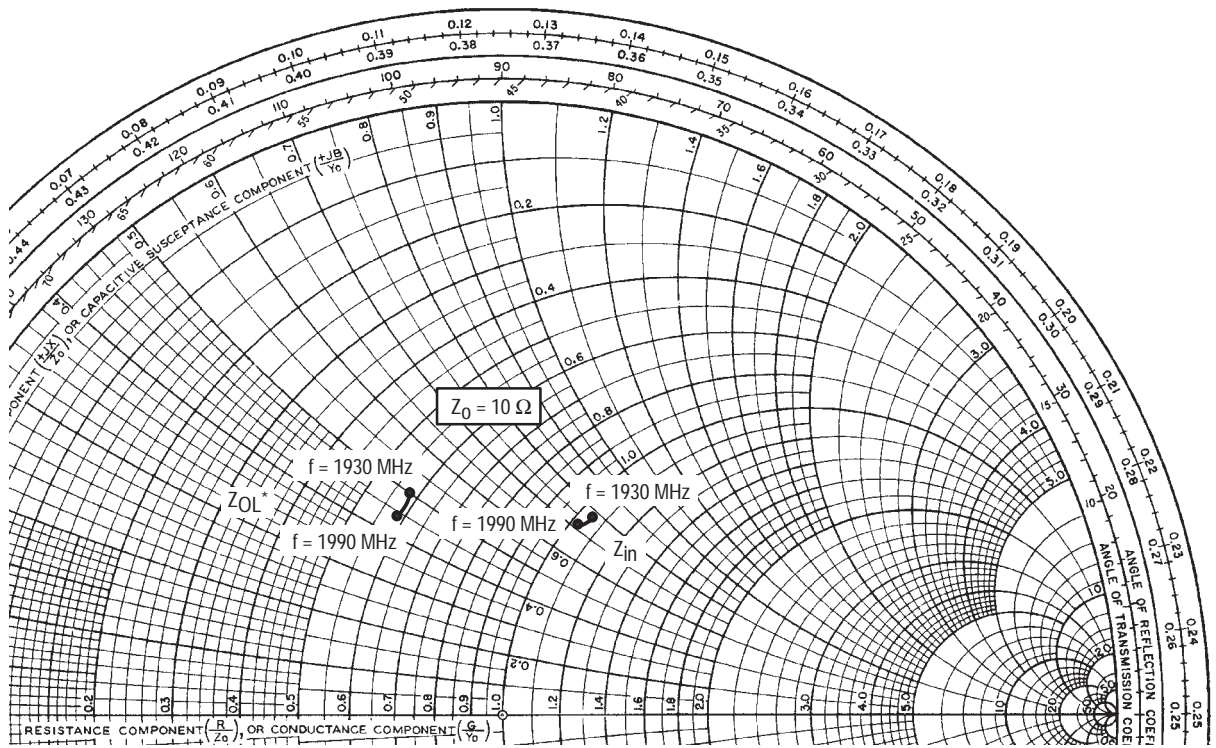
**Figure 6. Intermodulation Distortion Products versus Output Power**



**Figure 7. Power Gain versus Output Power**



**Figure 8. Power Gain and Intermodulation Distortion versus Supply Voltage**



$V_{DD} = 26\text{ V}$ ,  $I_{DQ} = 300\text{ mA}$ ,  $P_{out} = 30\text{ Watts (PEP)}$

| f<br>MHz | $Z_{in}$<br>$\Omega$ | $Z_{OL}^*$<br>$\Omega$ |
|----------|----------------------|------------------------|
| 1930     | $10.57 + j7.69$      | $5.81 + j5.01$         |
| 1960     | $10.54 + j7.43$      | $5.84 + j4.67$         |
| 1990     | $10.47 + j7.21$      | $5.84 + j4.35$         |

$Z_{in}$  = Complex conjugate of source impedance.

$Z_{OL}^*$  = Complex conjugate of the optimum load impedance at a given output power, voltage, IMD, bias current and frequency.

Note:  $Z_{OL}^*$  was chosen based on tradeoffs between gain, output power, drain efficiency and intermodulation distortion.

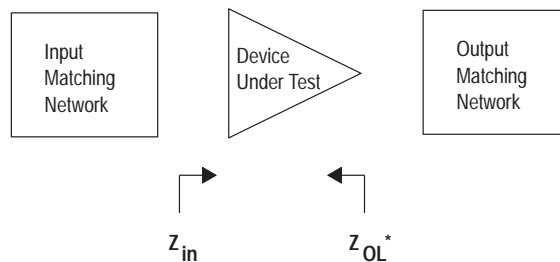
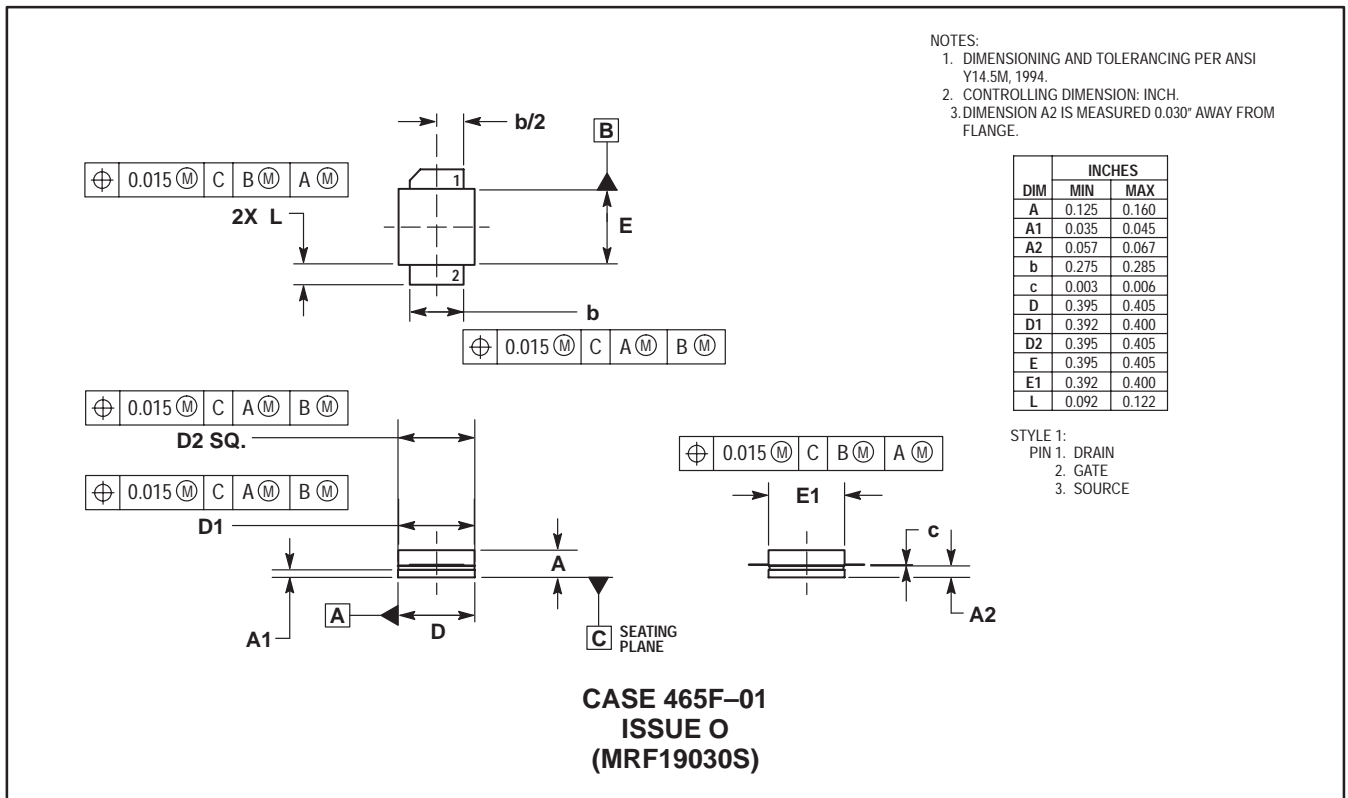
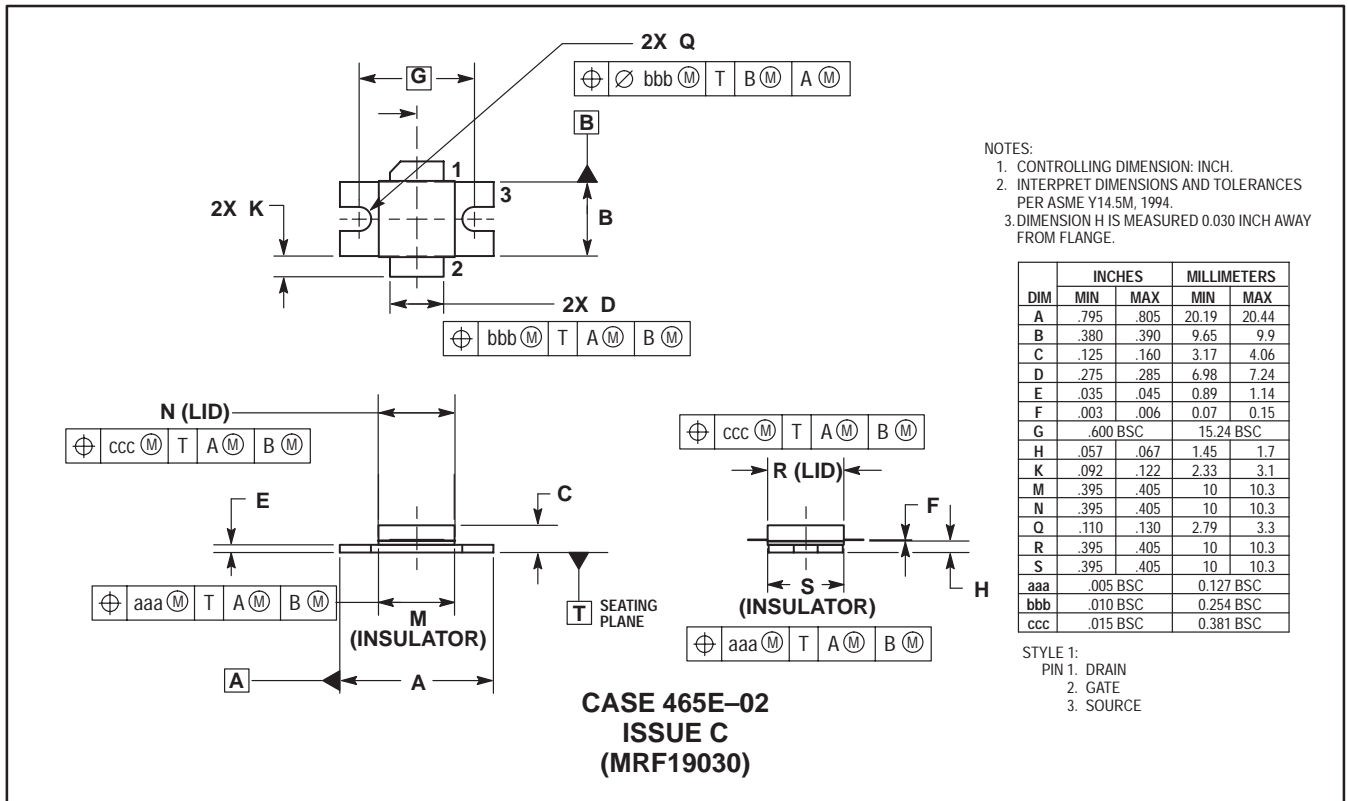



Figure 9. Series Equivalent Input and Output Impedance

# NOTES

## PACKAGE DIMENSIONS



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