# **SIEMENS**

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#### Introduction

#### SDA 9288X – The One-Chip Picture-In-Picture System

With SDA 9288X the semiconductor department for the first time presents a highly integrated one-chip solution for pictures insertion. The whole PIP circuitry consisting of analog-to-digital converter (ADC), picture memory, image processing, matrix and digital-to-analog converter (DAC) was packaged into a 32-pin P-DSO case. Compared to PIPplus the improved signal quality and various important features such as POP (picture outside picture) or multi-PIP extend the range of application considerably. Less space requirements on the PCB as well as reduced peripheral circuit complexity lower costs whereby the picture quality is increased at the same time.

# **SIEMENS**

# Picture-In-Picture Processor 1-Chip

**SDA 9288X** 

#### **Preliminary Data**

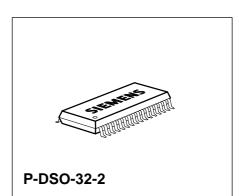
#### 1 Features

#### Multistandard Applications

625 lines / 50 Hz; 525 lines / 60 Hz standard (inset and parent channel)
Scan conversion systems as flickerfree display systems (parent channel)
HDTV (parent channel)



Clamping, AD-conversion, filtering, field memory, RGB-matrix, DA-conversion and clock generation integrated on one chip



#### • 2 Picture Sizes

1/9 or 1/16 of normal size

### • High Resolution Display

13.5 MHz / 27 MHz display clock frequency

212 luminance and 53 chrominance pixels per inset line for picture size 1/9

6 Bit amplitude resolution for each incoming signal component

Field and frame mode display

Horizontal and vertical filtering

Special antialias filtering for the luminance signal

# • 16:9 Compatibility

Operation in 4:3 and 16:9 sets

4:3 inset-signals on 16:9-displays or v.v. with picture size 1/9 and 1/16, respectively

# Analog Inputs

Y, +(B-Y), +(R-Y) or Y, -(B-Y), -(R-Y)

# Analog Outputs

Y, +(B-Y), +(R-Y) or Y, -(B-Y), -(R-Y) or RGB

3 RGB-matrices: EBU, NTSC (Japan); NTSC (USA)

Туре	Ordering Code	Package		
SDA 9288X	Q67101-H5083	P-DSO-32-2 (SMD)		

**SDA 9288X** 

### • Free Programmable Position of Inset Picture

Steps of 1 pixel and 1 line All PIP and POP positions are possible

#### Programmable Framing

4096 frame colors Variable frame width

- Freeze picture
- I<sup>2</sup>C Bus Control
- Threefold PIP / POP Facility

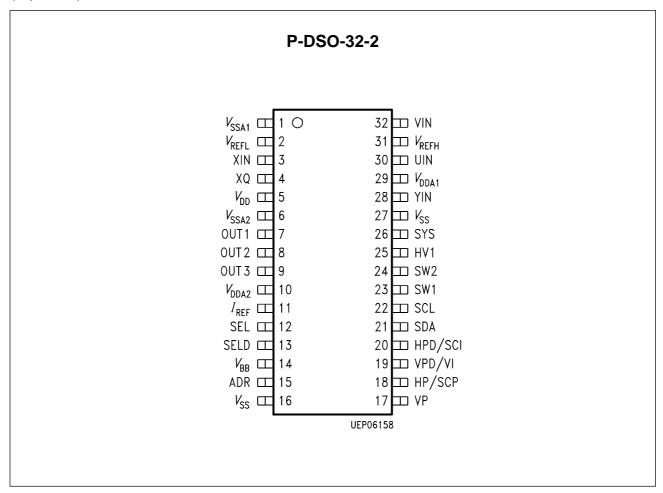
Three different I<sup>2</sup>C-addresses (pin-programmable) Tri-state outputs

- Numerical PLL Circuit for High Stability Clock Generation
- No Necessity of PAL / SECAM Delay Lines (using suitable color decoders i.e. TDA 8310)
- P-DSO-32-2 Package / 350 mil (SMD)
- 5 V Supply Voltage



### **Pin Configuration**

(top view)



# **Pin Definitions and Function**

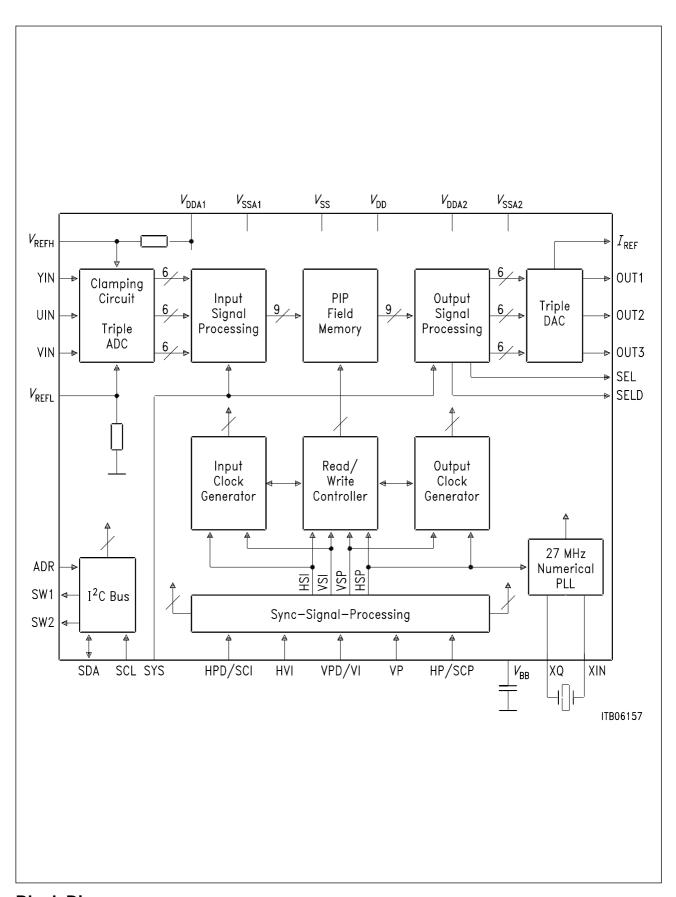
Pin No.	Symbol	Function	Description
1	$V_{\sf SSA1}$	S	Analog voltage supply ( $V_{ m SS}$ ) for ADC
2	$V_{REFL}$	I	Lower reference voltage for AD-converters
3	XIN	I	Quartz oscillator (input) or quartz clock (from another PIP IC) or line locked clock (27 MHz, from a digital parent channel)
4	XQ	Q	Quartz oscillator (output)
5	$V_{DD}$	S	Digital voltage supply $(V_{ m DD})$
6	$V_{\sf SSA2}$	S	Analog voltage supply $(V_{\rm SS})$ for DAC and PLL
7	OUT1	QA	Analog output: chrominance signal +(R-Y) or –(R-Y) or R
8	OUT2	Q A	Analog output: luminance signal Y or G
9	OUT3	Q A	Analog output: chrominance signal +(B-Y) or –(B-Y) or B
10	$V_{DDA2}$	S	Analog voltage supply $(V_{ m DD})$ for DAL and PLL
11	$I_{REF}$	Q A	Reference current for DA-converters
12	SEL	Q	Single frequency fast PIP switching output (Tristate)
13	SELD	Q	Double frequency fast PIP switching output (Tristate)
14	$V_{BB}$	S	Capacitor connection for smoothing internally generated substrate bias
15	ADR	I <sup>3</sup> L	I <sup>2</sup> C Bus address control
16, 27	$V_{ t SS}$	S	Digital voltage supply ( $V_{ m SS}$ )
17	VP	I	Multifrequency vertical sync for parent channel
18	HP/SCP	I	Multifrequency horizontal sync for parent channel
19	VPD/VI	I	Double frequency vertical sync for parent channel or vertical sync input for inset channel

# Pin Definitions and Function (cont'd)

Pin No.	Symbol	Function	Description
20	HPD/SCI	I	Double frequency horizontal sync for parent channel or horizontal input for inset channel
21	SDA	I/Q	I <sup>2</sup> C Bus data
22	SCL	1	I <sup>2</sup> C Bus clock
23	SW1	Q <sup>3</sup> L	I <sup>2</sup> C Bus controlled output 1
24	SW2	Q <sup>3</sup> L	I <sup>2</sup> C Bus controlled output 2
25	HVI	I <sub>3</sub> L	Special 3 level horizontal and vertical sync signal for inset channel
26	SYS	I <sup>3</sup> L	Input for standard depending internal switching (Low (L) = PAL, Mid (M) = NTSC, High (H) = SECAM)
28	YIN	ΙA	Analog input: luminance signal Y
29	$V_{DDA1}$	S	Analog voltage supply $(V_{ t DD})$ for ADC
30	UIN	ΙA	Analog input: chrominance signal +(B-Y) or -(B-Y)
31	$V_{REFH}$	1	Upper reference voltage for AD-converters
32	VIN	ΙA	Analog input: chrominance signal +(R-Y) or –(R-Y)

I = Input A = Analog  $^{3}L = 3 level$ 

Q = Output TTL = Digital (TTL) S = Supply voltage



# **Block Diagram**

#### 2 System Description

#### 2.1 AD-Conversion, Inset Synchronization

The inset video signal is fed to the SDA 9288X as analog luminance and chrominance components. The polarity of the chrominance signals is programmable. After clamping the video components are AD-converted with an amplitude resolution of 6 bit. The conversion is done using a 13.5 MHz clock for the luminance signal and a 3.375 MHz clock for the chrominance signals.

For inset synchronization it is possible to feed either a special 3 level signal via pin HVI (detection of horizontal and vertical pulses) or separate signals via pins SCI for horizontal and VI for vertical synchronization. SCI is the horizontal synchron signal of the inset channel. If the burst gate pulse of the sandcastle is used it must be adapted to TTL compatible levels by a simple external circuit. Centering of the displayed picture area is possible by a programmable delay for the horizontal synchronization signal (HSIDEL).

The inset horizontal synchronization signals are sampled with 27 MHz. This 27 MHz clock and the AD-converter clocks are derived from the parent horizontal synchronization pulse (see PLL description) or from the quartz frequency converted by a factor of 4/3.

Delay differences between luminance and chrominance signals at the input of the IC caused by chroma decoding are compensated by a programmable luminance delay line (YDEL) of about – 290 ns ... 740 ns (at decimation input; see application information).

By analyzing the synchronization pulses the line standard of the inset signal source is detected and interference noise on the vertical sync signal is removed. For applications with fixed line standard (only 625 lines or 525 lines) the automatic detection can be switched off.

The phase of the vertical sync pulse is programmable (VSIDEL; VSPDEL). By this way a correct detection of the field number is possible, an important condition for frame mode display. Remark: The adjustment of VSIDEL is influenced by HSIDEL (see waveform), vertical synchronization via pin HVI causes an additional internal delay for the vertical sync pulse of about  $16 \, \mu s$ .

# 2.2 Input Signal Processing

This stage performs the decimation of the inset signal by horizontal and vertical filtering and subsampling. A special antialias filter improves the frequency response of the luminance channel. It is optimized for the use of the horizontal decimation factor 3:1.

A window signal, derived from the sync pulses and the detected line standard, defines the part of the active video area used for decimation. For HSIDEL = 0 the decimation window is opened about 104 clock periods (13.5 MHz) after the horizontal synchronization pulse. For the 625 lines standard the 36<sup>th</sup> video line is the first decimated line, for the 525 lines standard decimation starts in the 26<sup>th</sup> video line.

The following decimation filters are implemented:

Horizontal Decimation	Component	Filter
3:1	luminance	$1 + z^{-1} + z^{-2}$
3:1	chrominance	$1 + 2 \times z^{-1} + z^{-2}$
4:1	luminance	$1 + z^{-1} + z^{-2} + z^{-3}$
4:1	chrominance	$1 + z^{-1} + z^{-2} + z^{-3}$

Vertical Decimation	Component	Filter
3:1	luminance	1 + z <sup>-L</sup> + z <sup>-2L</sup>
3:1	chrominance	$1 + 2 \times z^{-L} + z^{-2L}$
4:1	luminance	1 + z <sup>-L</sup> + z <sup>-2L</sup> + z <sup>-3L</sup>
4:1	chrominance	1 + z <sup>-L</sup> + z <sup>-2L</sup> + z <sup>-3L</sup>

$$z = e^{j\omega T}$$
,  $T = 1/13.5 \text{ MHz}$  for luminance  $T = 1/3.375 \text{ MHz}$  for chrominance

L = samples per line for luminance respectively chrominance

The realized chrominance filtering allows omitting the color decoder delay line for PAL and SECAM demodulation if the color decoder supplies the same output voltages independent of the kind of operation. In case of SECAM signals an amplification of the chrominance signals by a factor of 2 is necessary because just every second line a signal is present. This chrominance amplification is programmable via pin SYS or I<sup>2</sup>C Bus (AMSEC).

The horizontal and vertical decimation factors are free programmable (DECHOR, DECVER). Using different decimations horizontal and vertical 16:9 applications become realizable:

DECHOR = 1, DECVER = 0: picture size 1/9 for 4:3 inset signals on 16:9 displays.

DECHOR = 0, DECVER = 1: picture size 1/16 for 16:9 inset signals on 4:3 displays.

#### 2.3 PIP Field Memory

The onchip memory stores one decimated field of the inset picture. Its capacity is 169812 bits. The picture size depends on the horizontal and vertical decimation factors.

Horizontal Decimation	PIP Pixels per Line				
	Υ	(B-Y)	(R-Y)		
3:1	212	53	53		
4:1	160	40	40		

Vertical Decimation	Line Standard	PIP Lines
3:1	625	88
3:1	525	76
4:1	625	66
4:1	525	57

In field mode display just every second inset field is written into the memory, in frame mode display the memory is continuously written. Data are written with the lower inset clock frequency depending on the horizontal decimation factor (4.5 MHz or 3.375 MHz). Normally the read frequency is 13.5 MHz and 27 MHz for scan conversion systems. For progressive scan conversion systems and HDTV displays a mode is available (LINEDBL). Every line of the inset picture is read twice.

Memory writing can be stopped by program (FREEZE), a freeze picture display results (one field).

Having no scan conversion and the same line numbers in inset and parent channel (625 lines or 525 lines both) frame mode display is possible. The result is a higher vertical and time resolution because of displaying every incoming field. For this purpose the standards are internally analyzed and activating of frame mode display is blocked automatically when the described restrictions are not fulfilled.

As in the inset channel a field number detection is carried out for the parent channel. Depending on the phase between inset and parent signals a correction of the display raster for the read out data is performed by omitting or inserting lines when the read address counter outruns the write address counter.

The display position of the inset picture is free programmable (POSHOR, POSVER). The first possible picture position (without frame) is 54 clock periods (13.5 MHz or 27 MHz) after the horizontal and 4 lines after the vertical synchronization pulses. Starting at this position the picture can be moved over the whole display area. Even POPpositions (Picture Outside Picture) at 16:9 applications are possible.

Having different line standards in inset and parent channels we have a so called mixed mode display. It causes deformations in the aspect ratio of the inset picture. A special mixed mode display is available for the picture size 1/9 (MIXDIS):

- parent channel 625 lines, inset channel 525 lines: The inset picture is shifted down by 6 lines. By performing this shifting the centers of the inset pictures have the same position for both line standards.
- parent channel 525 lines, inset channel 625 lines: The inset picture gets a reduced line number of 76. The first 6 and the last 6 lines are omitted. This way the inset picture size is the same as for 525 lines inset signals. The display shows the center part of the original picture.

Synchronization of memory reading with the parent channel is achieved by processing the parent horizontal and vertical synchronization signals in the same way as described for the inset channel. The synchronization signals are fed to the IC at pin HP/SCP for horizontal synchronization and pin VP for vertical synchronization. In the same way as described for the inset channel the burst gate of the sandcastle signal can be used for horizontal synchronization. In scan conversion systems also the inputs HPD/SCI and VPDNI are available if the input HVI is activated for inset synchronization.

#### 2.4 Output Signal Processing

At the memory output the chrominance components are demultiplexed and linearly interpolated to the luminance sample rate.

Different output formats are available: luminance signal Y with inverted or non inverted chrominance signals (B-Y), (R-Y) or RGB. For the RGB-conversion 3 matrices are integrated:

Standard	B-Y	R-Y	G-Y	B-Y	R-Y	G-Y
EBU	1	0.588	0.345	0°	90°	237°
NTSC (Japan)	1	0.783	0.31	0°	95°	240°
NTSC (USA)	1	1.013	0.305	O°	104°	252°

Matrix selection is done by pin SYS or I<sup>2</sup>C Bus. The matrices are designed for the following input voltages (100 % white, 75 % color saturation):

Component	Input Voltage (without sync) in % of Full Scale Input Range of ADC
Y	75
B-Y	100
R-Y	100

#### **Matrix-Equations**

#### **EBU**

$$\begin{bmatrix} R \\ G \\ B \end{bmatrix} = \begin{bmatrix} 1 & 0 & 1 \\ 0 & 0.78125 & 1 \\ -0.1875 & -0.406251 \end{bmatrix} \begin{bmatrix} B - Y \\ R - Y \\ Y \end{bmatrix}$$

#### NTSC (Japan)

$$\begin{bmatrix} R \\ G \\ B \end{bmatrix} = \begin{bmatrix} 1 & 0 & 1 \\ -0.0625 & 1.09375 & 1 \\ -0.15625 & -0.375 & 1 \end{bmatrix} \begin{bmatrix} B - Y \\ R - Y \\ Y \end{bmatrix}$$

## NTSC (USA1)

$$\begin{bmatrix} R \\ G \\ B \end{bmatrix} = \begin{bmatrix} 1 & 0 & 1 \\ -0.25 & 1.375 & 1 \\ -0.09375 -0.406251 \end{bmatrix} \begin{bmatrix} B - Y \\ R - Y \\ Y \end{bmatrix}$$

A colored frame is added to the inset picture. 4096 frame colors are programmable, 4 bits for each component Y, (B-Y), (R-Y). The horizontal and vertical width of the frame are independently programmable. Width = 0 means display without frame.

For controlling an external switch (for example a RGB processor) a select signal is supplied. Pin SEL is active in normal 13.5 MHz reading mode, pin SELD is active using 27 MHz. The phases of these signals are programmable for adaption to different external output signal processing.

#### 2.5 DA-Conversion

The SDA 9288X includes three 6 bit DA-converters. Each converter supplies a current through an external resistor that is connected between VSSA and OUT1, OUT2, OUT3 respectively. The current is controlled by a digital control circuit. Each command DACONST or PIPON starts the adjustment cycle.

#### 2.6 PLL

A numerical PLL circuit supplies a clock of about 27 MHz with high stability. The generated clock is locked to the parent horizontal synchronization pulse. Its frequency depends linearly on the frequency of the sync signal and the quartz frequency. The recommended quartz frequencies are listed under "Recommended Operation Conditions". Using up to three SDA 9288X ICs in one application only a single quartz is necessary. Four time constants are programmable via I<sup>2</sup>C Bus. If the PLL is switched off an external 27 MHz parent line locked clock can be fed to the IC.

The inset clock generation is possible in two ways:

- 1. Synchron with the parent horizontal synchronization pulse (Bit CLISW = 0)
- 2. Synchron with the quartz frequency (Bit CLISW = 1;  $f_{\text{cli}} = 4/3 \times f_{\text{quartz}}$ ). In this mode the aspect ratio is independent on the parent sync frequency but depends on the used resonator type. It is only possible to use one of the two modes.

**Note:** Before setting bit D3 of Subaddress 00 (READ27) noise reduction of the VSP pulse must be switched off (D5 of subaddress 08 = 1).



#### 2.7 I<sup>2</sup>C Bus

#### 2.7.1 I2C-Bus Addresses

Three different I<sup>2</sup>C addresses are programmable via pin ADR.

Pin ADR	Address (bin.)	Address (hex.)
Low level ( $V_{ m SS}$ or $V_{ m SSA}$ )	11010110	D6
Mid level (open)	11011100	DC
High level ( $V_{\rm DD}$ or $V_{\rm DDA}$ )	11011110	DE

#### 2.7.2 I<sup>2</sup>C-Bus Receiver Format

S	Address	А	Subaddress	Α	Data byte	А	***	Α	Р
---	---------	---	------------	---	-----------	---	-----	---	---

S: Start condition

A: Acknowledge

P: Stop condition

Only write operation is possible. An automatically address increment function is implemented.

### 2.7.3 I<sup>2</sup>C-Bus Commands

Sub- Addr				Data	Bytes			
Hex.	D7	D6	D5	D4	D3	D2	D1	D0
00	0	SYSACT	FREEZE	PLLOFF	READ27	LINEDBL	FRAME	PIPON
01	0	SELDEL3	SELDEL2	SELDEL1	SELDEL0	MIXDIS	POSHOR9	POSHOR8
02	POSHOR7	POSHOR6	POSHOR5	POSHOR4	POSHOR3	POSHOR2	POSHOR1	POSHOR0
03	POSVER7	POSVER6	POSVER5	POSVER4	POSVER3	POSVER2	POSVER1	POSVER0
04	0	SW21	SW20	SW11	SW10	YDEL2	YDEL1	YDEL0
05	DECVER	DECHOR	INSVHI	CHRINS	PMOD1	PMOD0	IMOD1	IMOD0
06	0	0	0	CLISW	HSIDEL3	HSIDEL2	HSIDEL1	HSIDEL0
07	AMSEC	0	VSIISQ	VSIDEL4	VSIDEL3	VSIDEL2	VSIDEL1	VSIDEL0
08	PARSYND	0	VSPISQ	VSPDEL4	VSPDEL3	VSPDEL2	VSPDEL1	VSPDEL0
09	CON3	CON2	CON1	CON0	FRY5	FRY4	FRY3	FRY2
0A	FRV5	FRV4	FRV3	FRV2	FRU5	FRU4	FRU3	FRU2
0B	0	0	SELDOWN	FRWIDV1	FRWIDV0	FRWIDH2	FRWIDH1	FRWIDH0
0C	0	0	0	MAT2	MAT1	мато	CHRPIP	OUTFOR
0D	DACONST	PLLTC1	PLLTC2	0	0	0	0	DACONDE

After switching on the IC the data bytes of all registers are set to "0", the bit PLLOFF is set to "1".

Bit	Name	Function					
Subaddress	s 00						
D0	PIPON	0: PIP insertion off 1: PIP insertion on					
D1	FRAME	field display     frame display (under special restrictions)					
D2	LINEDBL	O: each line of the PIP memory is read once (normal operation)  1: each line of the PIP memory is read twice (line doubling for progressive scan conversion systems in parent channel)					
D3	READ27	O: PIP display with single read frequency (13.5 MHz) 1: PIP display with double read frequency (27 MHz) (see note page 18)					
D4	PLLOFF	internal PLL on     internal PLL off (external clock generation)					
D5	FREEZE	0: live picture 1: freeze picture					
D6	SYSACT	pin SYS inactive: selection of decimation amplification and RGB-matrix is done via I <sup>2</sup> C Bus     pin SYS active: selection of decimation amplification and RGB-matrix is done via pin SYS					
Subaddress	s 01						
D1 D0	POSHOR	2 MSBs of POSHOR (see also subaddress 02).					
D2	MIXDIS	O: PIP picture height depends just upon inset line standard, position upon POSHOR  1: modified PIP picture height and position for different inset and parent line standards (mixed display mode)					
D6 D3	SELDEL	Delay of output signal SELECT at pins SEL respectively SELD (– 8 + 7 periods of read frequency clock, programmable in 2's complement code). SELDEL = 0: SELECT signal has the same phase as the PIP picture signal referenced to the IC output.					

Bit	Name	Function						
Subaddres	ss 02							
D7 D0	POSHOR	Horizontal position of PIP picture (raster: 1 pixel).  Remark: the 2 MSBs of POSHOR are located at subaddress 01.  Warning: It is not allowed to adjust positions < 2 and > 740!						
Subaddres	ss 03							
D7 D0	POSVER	Vertical position of PIP picture (raster: 1 line).  Warning: It is not allowed to adjust positions > 220 (50 Hz) or > 182 (60 Hz)!						
Subaddres	ss 04							
D2 D0	YDEL	Delay of luminance input signal: 000: minimum delay 111: maximum delay; see application informations						
D4 D3	SW1	Direct control of output pin SW1 (3 levels):  00: low level  01: mid level  10: high level  11: high level						
D6 D5	SW2	Direct control of output pin SW2 (3 levels):  00: low level  01: mid level  10: high level  11: high level						

Bit	Name	Function						
Subaddres	s 05							
D1 D0	IMOD	00: automatic detection of line standard (inset signal) 01: fixed adjustment: 625 lines* 10: fixed adjustment: 525 lines* 11: freeze last automatically detected line standard						
D3 D2	PMOD	<ul> <li>00: automatic detection of line standard (parent signal)</li> <li>01: fixed adjustment: 625 lines*</li> <li>10: fixed adjustment: 525 lines*</li> <li>11: freeze last automatically detected line standard</li> </ul>						
D4	CHRINS	0: chrominance input signals +(B-Y), +(R-Y) 1: inverted chrominance input signals -(B-Y), -(R-Y)						
D5	INSHVI	inset synchronization signals via pins HPD/SCI and VPD/VI     inset synchronization signals via pin HVI (3-I. sandcastle signal)						
D6	DECHOR	0: horizontal decimation 3 to 1 1: horizontal decimation 4 to 1						
D7	DECVER	0: vertical decimation 3 to 1 1: vertical decimation 4 to 1						

<sup>\*</sup> Fixed adjustments for IMOD and PMOD result in undefined working conditions when signal standards are used which are different from the programmed values.

#### Subaddress 06

D3 D0	HSIDEL	Delay of horizontal synchronization pulse (inset signal). Raster: 6 clock periods (13.5 MHz). Warning: Adjustment of HSIDEL will influence the adjustment of VSIDEL (subaddr. 07); see waveforms!
D4	CLISW	inset clock synchronized with parent clock     inset clock synchronized with quartz frequency     Note: Only one of the two modes can be used.     Switching back from 1 to 0 is not possible!

Bit	Name	Function							
Subaddres	s 07								
D4 D0	VSIDEL	Delay of vertical synchronization pulse (inset signal) in steps of 2.37 $\mu s$ . Warning: Correct adjustment value is influenced by the adjustment of HSIDEL (subaddr. 06); see waveforms!							
D5	VSIISQ	Noise reduction of the VSI pulse (set to "0" under normal conditions).							
D7	AMSEC	unity amplification of decimation filters (normal mode)     amplification by a factor of 2 (SECAM signals without delay line in the chroma decoder)							
Subaddres	s 08								
D4 D0	VSPDEL	Delay of vertical synchronization pulse (parent signal) in steps of 2.37 $\mu s$ / 1.68 $\mu s$ (50/100 Hz).							
D5	VSPISQ	Noise reduction of the VSP pulse (should be set to "0" unde normal conditions).							
D7	PARSYND	<ul> <li>parent synchronization signals for double frequency read via pins HP/SCP and VP</li> <li>parent synchronization signals for double frequency read via pins HPD/SCI and VPD/VI (INSHVI = 1 required)</li> </ul>							
Subaddres	s 09								
D3 D0	FRY	Luminance component of frame color (4 MSBs of 6 bit).							
D7 D4	CON	contrast adjustment of PIP picture; steps and adjustment range depending on the external output resistors. Proposed value see at Characteristics.							
Subaddres	s 0A								
D3 D0	FRU	Chrominance component (B-Y) of frame color (4 MSBs of 6 bit).							
D7 D4	FRV	Chrominance component (R-Y) of frame color (4 MSBs of 6 bit).							

Bit	Name	Function						
Subaddress	s 0B							
D2 D0	FRWIDH	Horizontal width of PIP frame (0 7 pixels).						
D4 D3	FRWIDV	Vertical width of PIP frame (0 3 lines).						
D5	SELDOWN	open source output at pins SEL, SELD     TTL output at pins SEL, SELD						
Subaddress	s 0C							
D0	OUTFOR	0: format of output signals: Y, (B-Y), (R-Y) 1: format of output signals: R G B						
D1	CHRPIP	0: chrominance output signals: +(B-Y), +(R-Y) 1: inverted chrominance output signals: -(B-Y), -(R-Y)						
D2	MAT0	0: EBU RGB-matrix 1: NTSC RGB-matrix						
D3	MAT1	preselection of NTSC RGB-matrix (USA)     preselection of NTSC RGB-matrix (Japan)						
D4	MAT2	matrix section by bit MAT0     automatic matrix selection depending on inset line standard						
Subaddress	s 0D							
D0	DACONDE	Set to "0".						
D5 D6	PLLTC2 PLLTC1	Time constant of internal PLL:  00: medium damping, low resonance frequency  01: medium damping, high resonance frequency  10: high damping, low resonance frequency  11: high damping, high resonance frequency						
D7	DACONST	Changing from 0 to 1 starts automatic adjustment of OUT1 3 output current if D0 = 0.						

## 3 Absolute Maximum Ratings

Parameter	Symbol	Limi	t Values	Unit	Remark	
		min.	max.			
Ambient temperature		0	70	°C		
Storage temperature		- 55	125	°C		
Junction temperature			125	°C		
Soldering temperature			260	°C	duration < 10 s	
Input voltage		- 0.5	$V_{\rm DD}$ + 0.5	V	analog inputs (YIN, UIN, VIN, IREF)	
		<b>– 1</b>	7	V	all other pins	
Output voltage		- 0.5	$V_{\rm DD}$ + 0.5	V	pins OUT1, OUT2, OUT3, XQ, SW1, SW2	
		<b>–</b> 1	7	V	all other pins	
Supply voltages		<b>–</b> 1	7	V		
Supply voltage differentials		- 0.25	0.25	V		
Total power dissipation			900	mW		
Latch-up protection		- 100	100	mA	except pins OUT1, OUT2, OUT3, IREF, XQ, YIN, UIN, VIN	

All voltages listed are referenced to ground (0 V,  $V_{\rm SS}$ ) except where noted.

Absolute Maximum Ratings are those values beyond which damage to the device may occur. Functional operation under these conditions or at any other condition beyond those indicated in the operational sections of this specification is not implied.

# 4 Recommended Operating Conditions

Parameter	Symbol	Limit Values			Unit	Remark
		min.	typ.	max.		
Supply voltages	$V_{DDxx}$	4.75	5	5.5	V	
Ambient temperature	$T_{A}$	0	25	70	°C	
All TTL Inputs		1				
Low-level input voltage	$V_{IL}$	<b>- 1</b>		0.8	V	
High-level input voltage	$V_{IH}$	2.0		6	V	
All Three Level Inputs (3-L	.) see Figu	ure			1	
High-level input voltage	$V_{IH}$	3.5		6	V	
Low-level input voltage	$V_{IL}$	<b>– 1</b>		0.8	V	
Medium-level voltage	$V_{IM}$	open in	put, <b>see</b>	Charac	teristic	s
All 3-L Outputs see Figure		1				
High-level output current	$I_{OH}$	- 500		0	μΑ	
Low-level output current	$I_{OL}$	0		1.6	mA	
Inset Horizontal Sync TTL (All values are referred to		-			$(V_{\sf IM})$ aı	nd max( $V_{ extsf{ iny L}}$ ))
Horizontal frequency		14.53		16.72	kHz	
Signal rise time				100	ns	noisefree L/M to H transition
Signal high time		100			ns	
Signal medium or low time		900			ns	

# **Recommended Operating Conditions** (cont'd)

Parameter	Symbol	Li	imit Valu	ues	Unit	Remark
		min.	typ.	max.		
Inset Vertical Sync TTL an (All values are referred to	-				$V_{\scriptscriptstyle IM}$ ) ar	nd max( $V_{ extsf{ iny L}}$ ).
Signal medium or high time		17			μs	necessary for vertical sync detection
Signal low time		200			ns	
Parent Horizontal Sync TT (All values are referred to	-		•		nax( $V_{\shortparallel}$	_))
Sync frequency in single frequency display mode		14.53		16.72	kHz	quartz frequency 20.480 MHz
		15		17.19	kHz	quartz frequency 21.090 MHz
Sync frequency in double frequency display mode		29.06		33.44	kHz	quartz frequency 20.480 MHz
		30		34.375		quartz frequency 21.090 MHz
Signal rise time				100	ns	noisefree transition
Signal high time		100			ns	
Signal low time		900			ns	

# **Recommended Operating Conditions** (cont'd)

Parameter	Symbol	Limit Values			Unit	Remark	
		min.	typ.	max.			
Parent Vertical Sync TTL I (All values are referred to	•		ng min( $V$	′ <sub>เн</sub> ) and r	$max(V_{\shortparallel}$	.))	
Signal high time		200			ns		
Signal low time		200			ns		
Quartz / Ceramic Resonate	or		•	•	•		
Recommended frequency		20.25	20.48	21.3	MHz	21.09 MHz for MUSE	
Series resistance				10	Ω	$C_1, C_2 \le 33 \text{ pF}$	
				20	Ω	$C_1, C_2 \le 22 \text{ pF}$	
				30	Ω	$C_1, C_2 \le 15 \text{ pF}$	
				40	Ω	$C_1, C_2 \le 10 \text{ pF}$	
Optional TTL Clock Input:	XIN (All v	alues a	re referr	ed to mi	$in(V_IH)$	and max( $V_{\text{IL}}$ ))	
Clock input cycle time		35		40	ns	external line	
Clock input rise time				5	ns	locked 27 MHz clock	
Clock input fall time				5	ns	(I <sup>2</sup> C: internal PLL off)	
Clock input low time		10			ns		
Clock input high time		10			ns		

Recommended	Operating	Conditions	(cont'd)
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Parameter	Symbol	Li	mit Valu	es	Unit	Remark
		min.	typ.	max.		

Fast I<sup>2</sup>C Bus (All values are referred to  $\min(V_{\rm IH})$  and  $\max(V_{\rm IL})$ )
This specification of the bus lines need not to be identical with the I/O stages specification because of optional series resistors between bus lines and I/O pins.

SCL clock frequency	$f_{ t SCL}$	0	400	kHz	
Inactive time before start of transmission	$t_{BUF}$	1.3		μs	
Set-up time start condition	t <sub>SU; STA</sub>	0.6		μs	
Hold time start condition	$t_{HD;\;STA}$	0.6		μs	
SCL low time	$t_{LOW}$	1.3		μs	
SCL high time	$t_{HIGH}$	0.6		μs	
Set-up time DATA	$t_{ m SU;DAT}$	100		μs	
Hold time DATA	$t_{HD;\;DAT}$	0	0.9	μs	
SDA/SCL rise/fall times	$t_{R},t_{F}$	20 + \$	300	ns	$$ = 0.1 C_b / pF$
Set-up time stop condition	$t_{ m SU;STO}$	0.6		μs	
Capacitive load/bus line	$C_{b}$		400	pF	

#### I<sup>2</sup>C Bus Inputs/Output: SDA, SCL

High-level input voltage	$V_{IH}$	3		V <sub>DD</sub> + 0.5	V	also for SDA/ SCL input
Low-level input voltage	$V_{IL}$	- 0.5		1.5	V	stages
Spike duration at inputs		0	0	50	ns	
Low-level output current	$I_{OL}$			6	mA	

# **Recommended Operating Conditions** (cont'd)

Parameter	Symbol	L	Limit Values			Remark
		min.	typ.	max.		
Analog To Digital Conver	ters (6 bit)		-		•	
Input coupling capacitors		10	100		nF	necessary for
Y, U, V source resistance				1	kΩ	proper clamping
Reference voltage low	$V_{REFL}$	0.5	1.0	1.5	V	min and max
Reference voltage high	$V_{REFH}$	1.5	2.0	2.5	V	values only with optional external resistors, see also Characteristics
Reference voltage difference	$V_{REFH} - V_{REFL}$	0.5	1.0	2.0	V	
Digital To Analog Conver	ters (6 bit)					l
Full range output voltage	$V_{OFR}$		1	2	V	peak-to-peak
Reference resistance	R <sub>REF1</sub>	4.2	5.1	6.3	kΩ	bits CON = 0000 no contrast adjustment used
	$R_{REF2}$	6.0	6.8	7.5	kΩ	contrast adjustment via I <sup>2</sup> C Bus

### 5 Characteristics

# (Assuming Recommended Operating Conditions)

**Note:** The listed characteristics are ensured over the operating range of the integrated circuit unless restricted to nominal operating conditions. (All voltages refer to  $V_{\rm SS}$ ).

Parameter	Symbol	Limit	Values	Unit	Remark
		min.	max.		
Average total supply current	$I_{DDtot}$		160	mA	$I_{\text{DDtot}} = I_{\text{DD}} + I_{\text{DDA1}} + I_{\text{DDA2}}$ <b>Note:</b> The maxima do
Average digital supply current	$I_{DD}$		120	mA	not necessarily coincide
Average analog supply current	$I_{DDA1}$		40	mA	
Average analog supply current	$I_{DDA2}$		20	mA	
All Digital Inputs (TTL, I <sup>2</sup> C)	)		'		
Input capacitance	$C_{I}$		7	pF	not tested
Input leakage current		- 10	10	μА	including leakage current of SDA output stage
All Three Level Inputs (3-L	) see Figu	ure			
Input capacitance	$C_{I}$		7	pF	not tested
Medium-level open input voltage	$V_{IM}$	2.1	2.5	V	$/I_{IN}/\leq$ 1 $\mu$ A, $V_{DD}=5$ V
Differential input resistance	$R_{IN}$	8	14	kΩ	not tested

# Characteristics (cont'd)

Parameter	Symbol	Limit Values		Unit	Remark
		min.	max.		
SEL, SELD					
High-level output voltage	$V_{QH}$	2.4	$V_{DD}$	V	$I_{QH} = -200 \mu A$
High-level output voltage	$V_{QH}$	1.5	$V_{DD}$	V	$I_{\rm QH} = -4.5 \; {\rm mA}$
Low-level output voltage	$V_{QL}$		0.4	V	$I_{\rm QL}$ = 1.6 mA, only valid if bit SELDOWN = 1
Leakage current		- 10		μΑ	$V_{\rm Q}$ = 0 V $V_{\rm DD}$
Output capacitance			7	pF	not tested
All 3-L Outputs				1	
High-level output voltage	$V_{QH}$		4 3.9	V	$I_{\rm QH} = -\ 100\ \mu {\rm A}$ $I_{\rm QH} = -\ 500\ \mu {\rm A}$
Low-level output voltage	$V_{QL}$		0.4	V	$I_{\rm QL} = {\sf max}$
Medium-level output leakage current	$I_{QM}$	<b>– 1</b>	1	μΑ	tristate
Output capacitance			7	pF	not tested
I <sup>2</sup> C Inputs: SDA/SCL	·			1	
Schmitt trigger hysteresis	$V_{hys}$	0.2		V	not tested
I <sup>2</sup> C Input / Output: SDA (r	eferenced	to SCL;	open d	rain o	utput)
Low-level output voltage	$V_{QL}$		0.4	V	$I_{\rm QL}$ = 3 mA
Low-level output voltage	$V_{QL}$		0.6	V	$I_{\rm QL}$ = max
Output fall time from $\min(V_{\text{IH}})$ to $\max(V_{\text{IL}})$	$t_{\sf QF}$	$20 + 0.1 \times C_b/pF$	250	ns	10 pF $\leq C_{\rm b} \leq$ 400 pF

# Characteristics (cont'd)

Parameter	Symbol	Limit Values		Unit	Remark			
		min.	max.					
Analog To Digital Converters (6 bit)								
Y, U, V input leakage current		- 100	100	nA				
Y, U, V input capacitance			7	pF	not tested			
Input clamping error		<b>–</b> 1	1	LSB	settled state			
Input clamping current	$ I_{CLP} $	50	150	μΑ	for large devitations			
Reference voltage difference	$V_{REFH}$ – $V_{REFL}$	0.98	1.02	V	$V_{\text{DDA}} = 5 \text{ V},$ $(V_{\text{REFH}} - V_{\text{REFL}} \cong V_{\text{DDA1}}/5)$			
D.C. differential nonlinearity		- 0.5	0.5	LSB	$V_{REFH} - V_{REFL} = nom$			

# **Digital To Analog Converters (6 bit): Current Source Outputs OUT1, OUT2, OUT3** I<sup>2</sup>C: Contrast bits set to zero unless otherwise noted

D.C. differential nonlinearity	DNLE	- 0.5	0.5	LSB	$R_{REF}$ = 5.1 k $\Omega$
Full range output current	$I_{Q}$	tbf	tbf	mA	$V_{\mathrm{DDA}}$ = nom, $T_{\mathrm{A}}$ = nom, $R_{\mathrm{REF}}$ = 5.1 k $\Omega$ , $R_{\mathrm{L}}$ = 560 $\Omega$ , after adjustment
Output voltage $(U_{\rm Q} \sim 1.6 \times V_{\rm DDA} \times R_{\rm L}/R_{\rm REF})$	$U_{Q}$	tbf	tbf	V	$V_{\rm DDA}$ = nom, $T_{\rm A}$ = nom, $R_{\rm REF}$ = 5.1 k $\Omega$ , $R_{\rm L}$ = 560 $\Omega$ after adjustment
Tracking		- 3	3	%	$V_{ m DDA}$ = nom, $T_{ m A}$ = nom, $R_{ m REF}$ = 5.1 k $\Omega$ , $R_{ m L}$ = 560 $\Omega$

# Characteristics (cont'd)

Parameter	Symbol	Limit Values		Unit	Remark
		min.	max.		
Contrast increase		30		%	$V_{\rm DDA}$ = nom, $T_{\rm A}$ = nom, $R_{\rm REF}$ = 6.8 k $\Omega$ , $R_{\rm L}$ = 560 $\Omega$ contrast bits change from 0000 to 1111 for typical values see diagram
Supply voltage dependence of DAC output current					for typical values see diagram
Temperature dependence of DAC output current					for typical values see diagram
Dependence of DAC output current on external reference resistor					for typical values see diagram

### 6 Diagrams

### 6.1 Output Current of DA Converters

Nominal values:  $V_{\rm DDA}$  = 5 V;  $R_{\rm REF}$  = 5.1 k $\Omega$ ; T = 25  $^{\circ}{\rm C}$ 

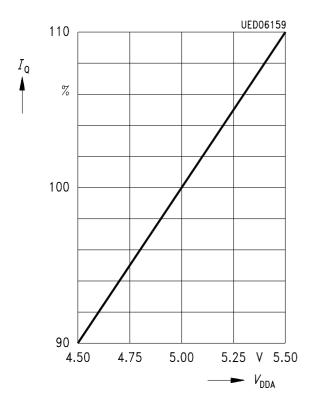
Measurements after adjustment via bit d7 of I2C Bus address 0D for each step.

Remark: The output currents are controlled in digital way, so inaccuracy of 1LSB (ca.

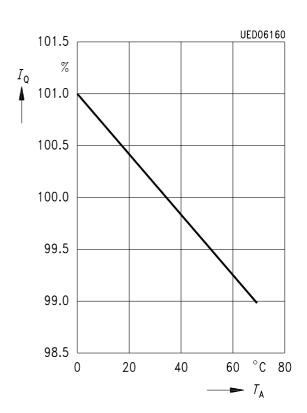
2 %) is always possible.

# **SIEMENS**

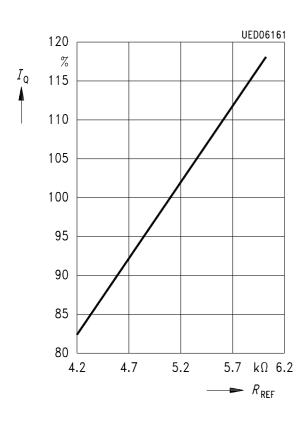
# Output current = $f(V_{DDA})$



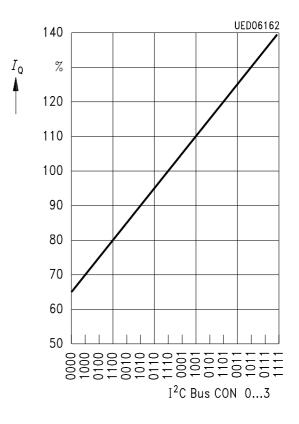
# Output current = $f(T_A)$



Output current =  $f(R_{REF})$ 



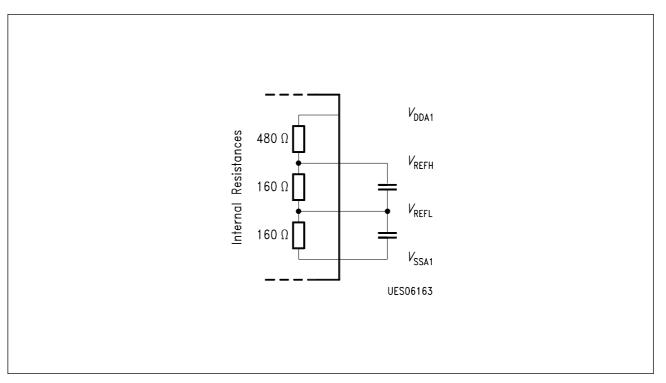
Current = f (CON 0 ... 3),  $R_{\rm REF}$  = 6.8 k $\Omega$ 



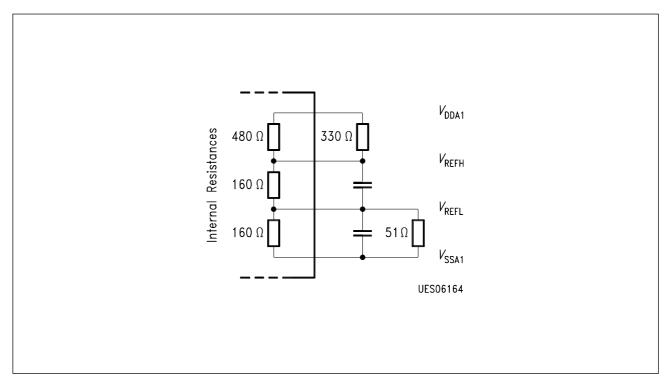


# 7 Application Information

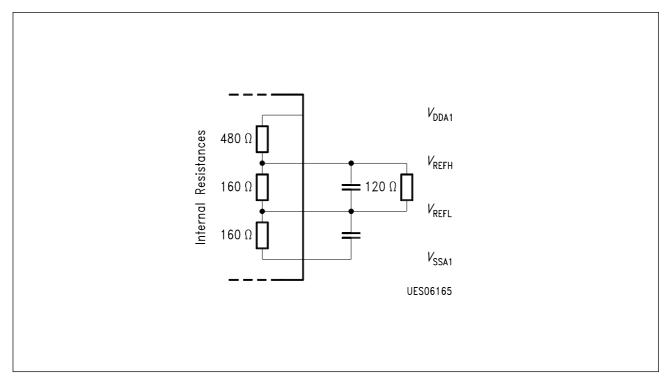
# 7.1 Reference Voltage Generation for ADC



Signal Input Range 1 Vpp at Y, U, V

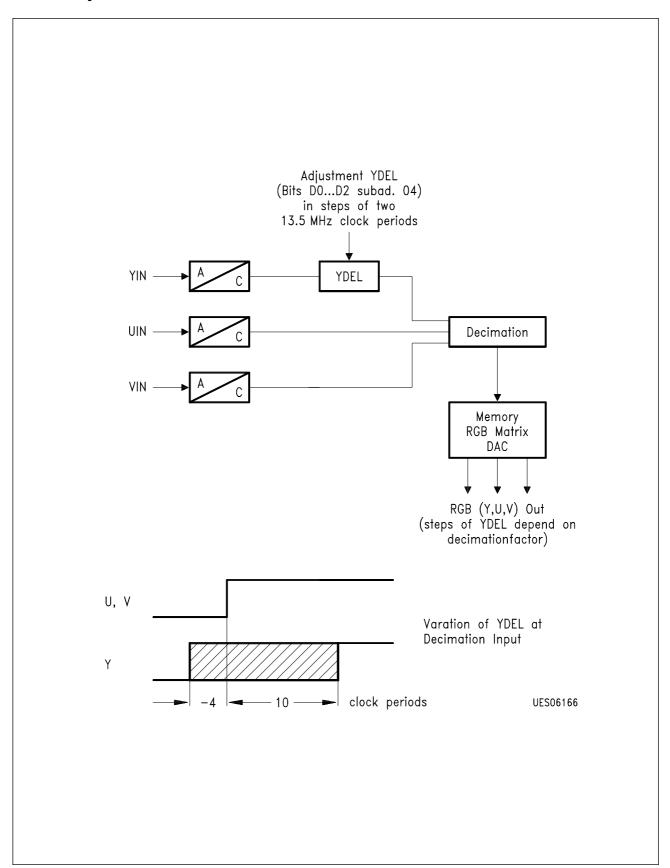


Signal Input Range 2 Vpp at Y, U, V

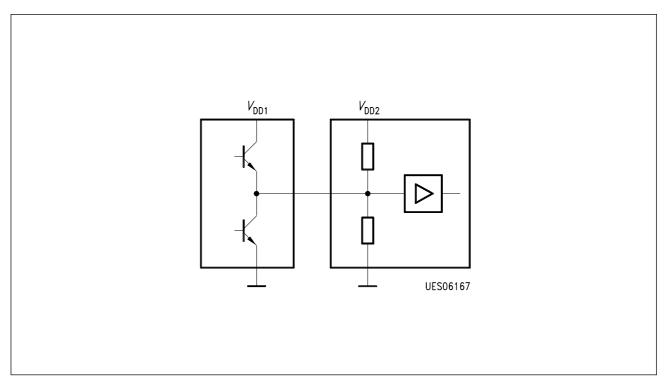


Signal Input Range 0.5 Vpp at Y, U, V

# 7.2 Adjustment of YDEL



# 7.3 Three Level Interface (3-L)

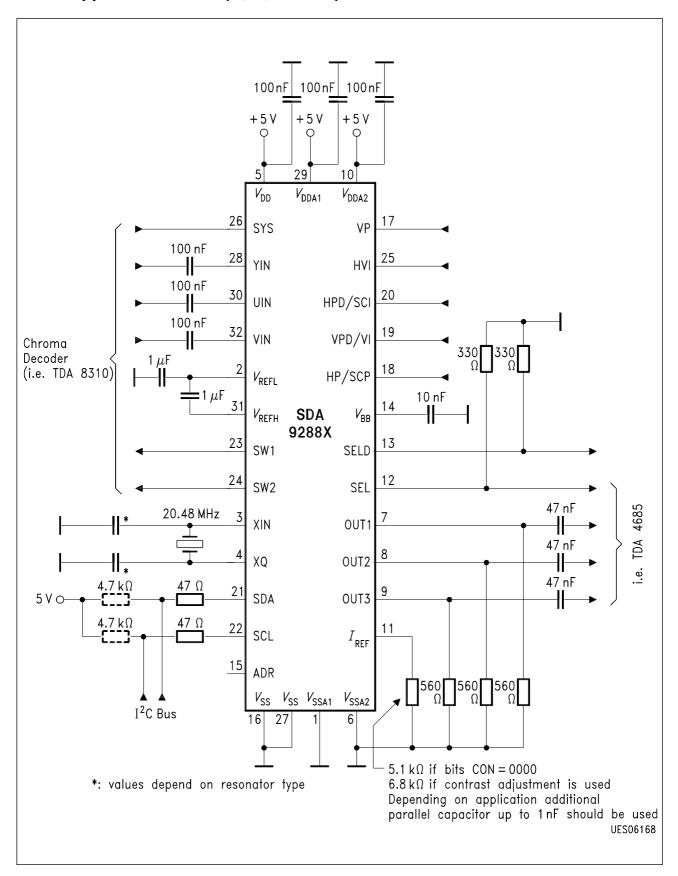


**High Level (H):** upper transistor on, lower transistor off

**Medium Level (M):** both transistors off (interface voltage determined by input stage)

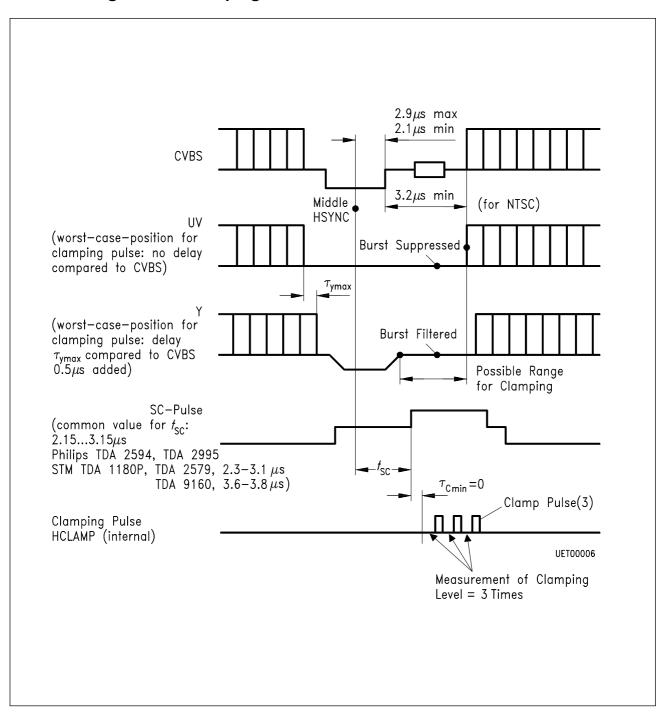
**Low Level (L):** upper transistor off, lower transistor on

### 7.4 Application Circuit (R, G, B-Mode)



#### 8 Waveforms

#### 8.1 Timing of ADC Clamping

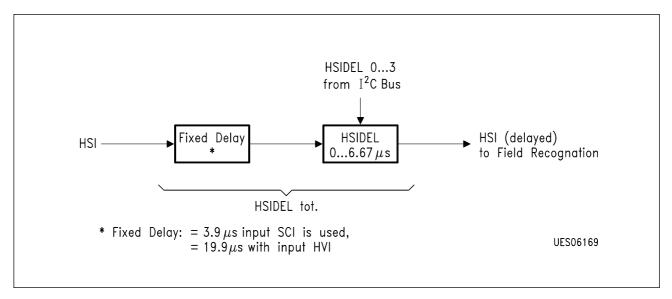


The values of  $t_m$  and  $t_c$  depend on the value of subaddress 06 (HSIDEL, CLISW). If it is 00:

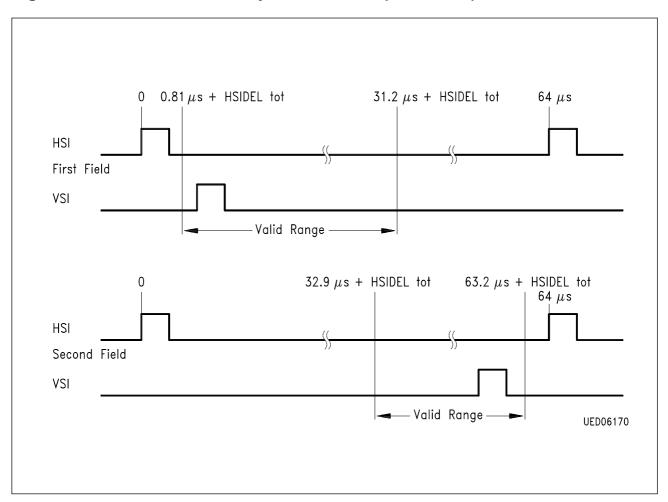
 $t_{\rm m}({\rm min})$  = 166 ns  $\pm$  6 % for the Y output and 240 ns  $\pm$  6 % for U, V;  $t_{\rm c}({\rm min})$  = 296 ns  $\pm$  6 % for the Y output and 444 ns  $\pm$  6 % for U, V;

To get the maximum values 444 ns for each step of HSIDEL adjustment must be added.

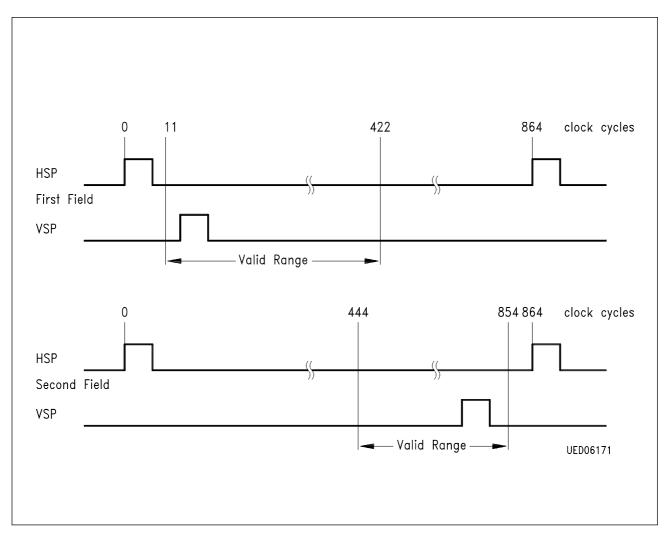
#### 8.2 Phase Relation of Sync Pulses at Frame Mode



#### Signal Flow of the Horizontal Synchronization (Insert Part)



Allowed Phase Relation of the Horizontal/Vertical Sync Pulses (Insert Channel) if VSIDEL 0:4 = 0000

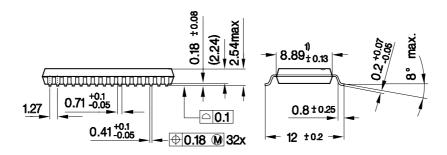


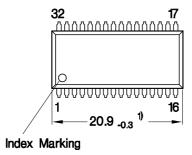
Allowed Phase Relation of the Horizontal/Vertical Sync Pulses (parent channel) if  $VSPDEL\ 0:4 = 0000$ 

#### 9 Package Outlines

### Plastic Package, P-DSO-32-2 (SMD)

(Plastic Dual Small Outline Package)





1) Does not include plastic or metal protrusion of 0.25 max. per side

#### **Sorts of Packing**

Package outlines for tubes, trays etc. are contained in our Data Book "Package Information"

SMD = Surface Mounted Device

Dimensions in mm