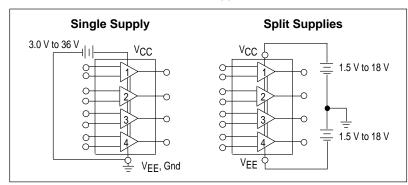


# **Quad Low Power Operational Amplifiers**

The MC3403 is a low cost, quad operational amplifier with true differential inputs. The device has electrical characteristics similar to the popular MC1741C. However, the MC3403 has several distinct advantages over standard operational amplifier types in single supply applications. The quad amplifier can operate at supply voltages as low as 3.0 V or as high as 36 V with quiescent currents about one third of those associated with the MC1741C (on a per amplifier basis). The common mode input range includes the negative supply, thereby eliminating the necessity for external biasing components in many applications. The output voltage range also includes the negative power supply voltage.

- Short Circuit Protected Outputs
- Class AB Output Stage for Minimal Crossover Distortion
- True Differential Input Stage
- Single Supply Operation: 3.0 V to 36 V
   Split Supply Operation: ±1.5 V to ±18 V
   Low Input Bias Currents: 500 nA Max
- Four Amplifiers Per Package
- Internally Compensated
- Similar Performance to Popular MC1741C
- Industry Standard Pinouts
- ESD Diodes Added for Increased Ruggedness



#### **MAXIMUM RATINGS**

Rating	Symbol	Value	Unit
Power Supply Voltages			Vdc
Single Supply	VCC	36	
Split Supplies	V <sub>CC</sub> , V <sub>EE</sub>	±18	
Input Differential Voltage Range (Note 1)	V <sub>IDR</sub>	±36	Vdc
Input Common Mode Voltage Range (Notes 1, 2)	VICR	±18	Vdc
Storage Temperature Range	T <sub>stg</sub>	-55 to +125	ç
Operating Ambient Temperature Range	TA		°C
MC3303		-40 to +85	
MC3403		0 to +70	
Junction Temperature	TJ	150	°C

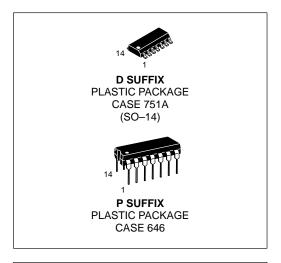
NOTES: 1. Split power supplies.

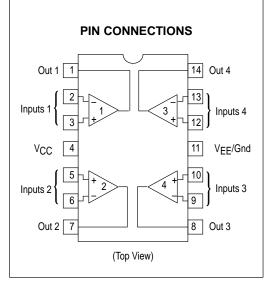
2. For supply voltages less than  $\pm 18$  V, the absolute maximum input voltage is equal to the supply voltage.

### MC3403 MC3303

## QUAD DIFFERENTIAL INPUT OPERATIONAL AMPLIFIERS

SEMICONDUCTOR TECHNICAL DATA





#### **ORDERING INFORMATION**

Device	Operating Temperature Range	Package
MC3303D MC3303P	$T_A = -40^{\circ} \text{ to } +85^{\circ}\text{C}$	SO-14 Plastic DIP
MC3403D MC3403P	$T_A = 0^\circ \text{ to } +70^\circ \text{C}$	SO-14 Plastic DIP

**ELECTRICAL CHARACTERISTICS** ( $V_{CC}$  = +15 V,  $V_{EE}$  = -15 V for MC3403;  $V_{CC}$  = +14 V,  $V_{EE}$  = Gnd for MC3303

 $T_A = 25^{\circ}C$ , unless otherwise noted.)

		MC3403		MC3303				
Characteristic	Symbol	Min	Тур	Max	Min	Тур	Max	Unit
Input Offset Voltage TA = Thigh to Tlow (Note 1)	V <sub>IO</sub>	- -	2.0 –	10 12	- -	2.0	8.0 10	mV
Input Offset Current TA = Thigh to Tlow	liO	_ _	30 -	50 200	- -	30 -	75 250	nA
Large Signal Open Loop Voltage Gain $V_O = \pm 10 \text{ V}, R_L = 2.0 \text{ k}\Omega$ $T_A = T_{high}$ to $T_{low}$	AVOL	20 15	200 –	- -	20 15	200 –	1 -	V/mV
Input Bias Current TA = T <sub>high</sub> to T <sub>low</sub>	I <sub>IB</sub>	_ _	–200 –	-500 -800	- -	-200 -	-500 -1000	nA
Output Impedance f = 20 Hz	z <sub>o</sub>	_	75	_	_	75	ı	Ω
Input Impedance f = 20 Hz	zį	0.3	1.0	_	0.3	1.0	_	MΩ
Output Voltage Range $\begin{array}{l} R_L=10 \ k\Omega \\ R_L=2.0 \ k\Omega \\ R_L=2.0 \ k\Omega, \ T_A=T_{high} \ to \ T_{low} \\ \\ \hline \\ Input Common Mode Voltage Range \\ \end{array}$	Vo	±12 ±10 ±10	±13.5 ±13 - +13 V	- - -	12 10 10 +12 V	12.5 12 - +12.5 V	- - -	\ \ \ \ \ \ \ \ \ \ \ \ \ \ \ \ \ \ \
		-VEE	-VEE		-VEE	-VEE		
Common Mode Rejection R <sub>S</sub> ≤ 10 k Ω	CMR	70	90	_	70	90	-	dB
Power Supply Current (V <sub>O</sub> = 0) R <sub>L</sub> = ∞	ICC, IEE	_	2.8	7.0	_	2.8	7.0	mA
Individual Output Short–Circuit Current (Note 2)	Isc	±10	±20	±45	±10	±30	±45	mA
Positive Power Supply Rejection Ratio	PSRR+	_	30	150	_	30	150	μV/V
Negative Power Supply Rejection Ratio	PSRR-	_	30	150	_	30	150	μV/V
Average Temperature Coefficient of Input Offset Current $T_A = T_{high}$ to $T_{low}$	ΔΙ <sub>ΙΟ</sub> /ΔΤ	_	50	_	-	50	ı	pA/°C
Average Temperature Coefficient of Input Offset Voltage TA = Thigh to Tlow	ΔV <sub>IO</sub> /ΔΤ	_	10	_	_	10	ı	μV/°C
Power Bandwidth AV = 1, RL = 10 k $\Omega$ , VO = 20 V(p-p), THD = 5%	BWp	_	9.0	-	-	9.0	-	kHz
Small–Signal Bandwidth $A_V = 1$ , $R_L = 10 \text{ k}\Omega$ , $V_O = 50 \text{ mV}$	BW	-	1.0	_	_	1.0	-	MHz
Slew Rate A <sub>V</sub> = 1, $V_i = -10 \text{ V}$ to +10 V	SR	_	0.6	-	-	0.6	-	V/µs
Rise Time A <sub>V</sub> = 1, R <sub>L</sub> = 10 k $\Omega$ , V <sub>O</sub> = 50 mV	tTLH	_	0.35	_	-	0.35	_	μs
Fall Time A <sub>V</sub> = 1, R <sub>L</sub> = 10 k $\Omega$ , V <sub>O</sub> = 50 mV	tTLH	_	0.35	_	-	0.35	-	μs
Overshoot A <sub>V</sub> = 1, R <sub>L</sub> = 10 k $\Omega$ , V <sub>O</sub> = 50 mV	os	_	20	_	_	20	_	%
Phase Margin A <sub>V</sub> = 1, R <sub>L</sub> = 2.0 k $\Omega$ , V <sub>O</sub> = 200 pF	φm	_	60	_	_	60	_	Degrees
Crossover Distortion (Vin = 30 mVpp, Vout= 2.0 Vpp, f = 10 kHz)	_	_	1.0	-	-	1.0	-	%

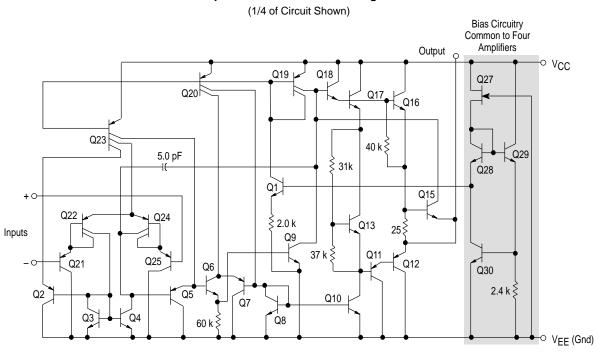
NOTES: 1. T<sub>high</sub> = +70°C for MC3403, +85°C for MC3303 T<sub>low</sub> = 0°C for MC3403, -40°C for MC3303 2. Not to exceed maximum package power dissipation.

 $\textbf{ELECTRICAL CHARACTERISTICS} \ \ (\text{V}_{CC} = 5.0 \text{ V}, \text{ V}_{EE} = \text{Gnd}, \text{ T}_{A} = 25^{\circ}\text{C}, \text{ unless otherwise noted.})$ 

		MC3403			MC3303			
Characteristic	Symbol	Min	Тур	Max	Min	Тур	Max	Unit
Input Offset Voltage	VIO	-	2.0	10	-	_	10	mV
Input Offset Current	ΙΙΟ	-	30	50	-	_	75	nA
Input Bias Current	I <sub>IB</sub>	_	-200	-500	_	_	-500	nA
Large Signal Open Loop Voltage Gain $R_L=2.0~\text{k}\Omega$	AVOL	10	200	-	10	200	-	V/mV
Power Supply Rejection Ratio	PSRR	-	-	150	-	_	150	μV/V
Output Voltage Range (Note 3) $R_L = 10 \text{ k}\Omega, V_{CC} = 5.0 \text{ V}$ $R_L = 10 \text{ k}\Omega, 5.0 \leq V_{CC} \leq 30 \text{ V}$	VOR	3.3 V <sub>CC</sub> -2.0	3.5 V <sub>CC</sub> -1.7	- -	3.3 V <sub>CC</sub> -2.0	3.5 V <sub>CC</sub> -1.7	- -	Vpp
Power Supply Current	Icc	-	2.5	7.0	_	2.5	7.0	mA
Channel Separation f = 1.0 kHz to 20 kHz (Input Referenced)	CS	-	-120	-	-	-120	-	dB

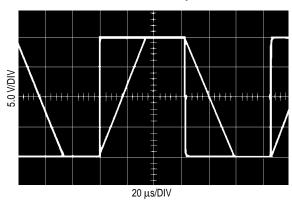
 $\textbf{NOTES:}\ 3.$  Output will swing to ground with a 10  $k\Omega$  pull down resistor.

#### **Representative Schematic Diagram**



#### CIRCUIT DESCRIPTION

#### **Inverter Pulse Response**



The MC3403/3303 is made using four internally compensated, two-stage operational amplifiers. The first stage of each consists of differential input device Q24 and Q22 with input buffer transistors Q25 and Q21 and the differential to single ended converter Q3 and Q4. The first

stage performs not only the first stage gain function but also performs the level shifting and transconductance reduction functions. By reducing the transconductance, a smaller compensation capacitor (only 5.0 pF) can be employed, thus saving chip area. The transconductance reduction is accomplished by splitting the collectors of Q24 and Q22. Another feature of this input stage is that the input common mode range can include the negative supply or ground, in single supply operation, without saturating either the input devices or the differential to single—ended converter. The second stage consists of a standard current source load amplifier stage.

The output stage is unique because it allows the output to swing to ground in single supply operation and yet does not exhibit any crossover distortion in split supply operation. This is possible because Class AB operation is utilized.

Each amplifier is biased from an internal voltage regulator which has a low temperature coefficient, thus giving each amplifier good temperature characteristics as well as excellent power supply rejection.

Figure 1. Sine Wave Response

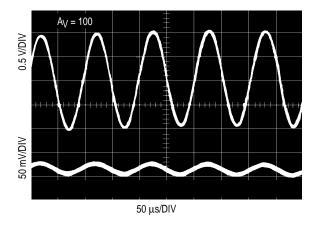


Figure 2. Open Loop Frequency Response

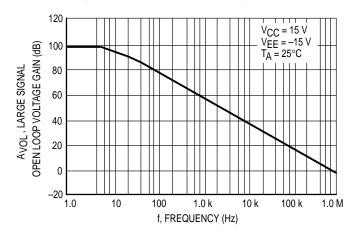


Figure 3. Power Bandwidth

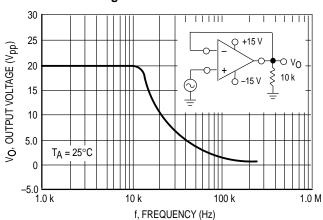


Figure 4. Output Swing versus Supply Voltage

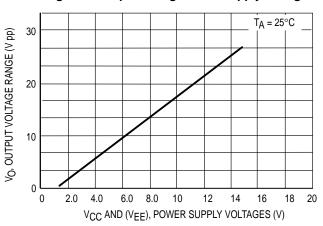


Figure 5. Input Bias Current versus Temperature

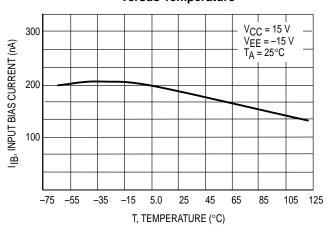


Figure 6. Input Bias Current versus Supply Voltage

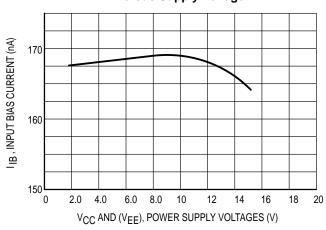


Figure 7. Voltage Reference

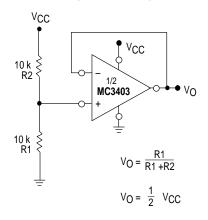


Figure 8. Wien Bridge Oscillator

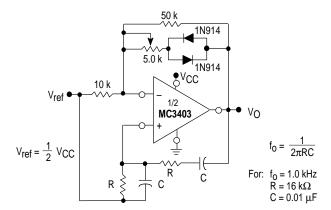


Figure 9. High Impedance Differential Amplifier

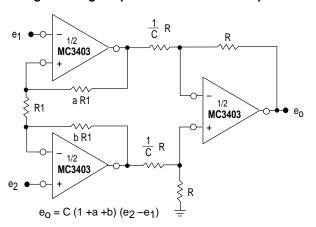


Figure 10. Comparator with Hysteresis

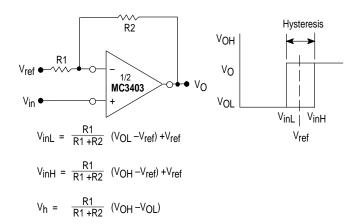


Figure 11. Bi-Quad Filter

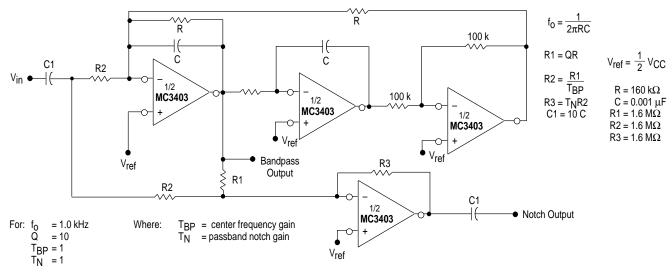


Figure 12. Function Generator

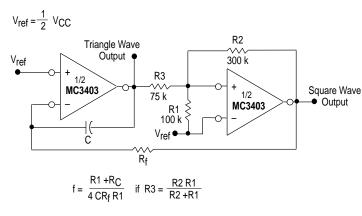
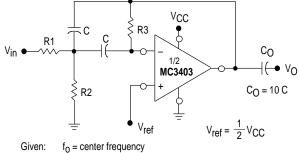


Figure 13. Multiple Feedback Bandpass Filter



Given:  $f_0$  = center frequency  $A(f_0)$  = gain at center frequency

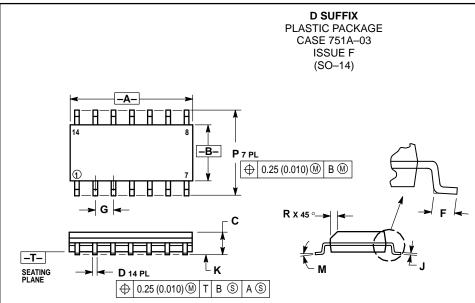
Choose value fo, C

Then: 
$$R3 = \frac{Q}{\pi f_0 C}$$
  $R1 = \frac{R3}{2 A(f_0)}$   $R2 = \frac{R1 R5}{4Q^2 R1 - R5}$ 

For less than 10% error from operational amplifier  $\frac{O_0 f_0}{BW} < 0.1$  where  $f_0$  and BW are expressed in Hz.

If source impedance varies, filter may be preceded with voltage follower buffer to stabilize filter parameters.

#### **OUTLINE DIMENSIONS**

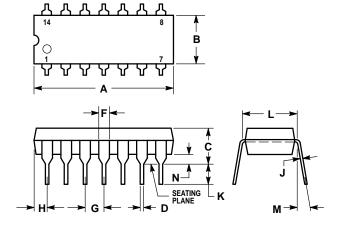


#### NOTES:

- DIMENSIONING AND TOLERANCING PER ANSI Y14.5M, 1982.
- 2. CONTROLLING DIMENSION: MILLIMETER.
  3. DIMENSIONS A AND B DO NOT INCLUDE
- DIMENSIONS A AND B DO NOT INCLUDE MOLD PROTRUSION.
  MAXIMUM MOLD PROTRUSION 0.15 (0.006) PER SIDE.
  DIMENSION D DOES NOT INCLUDE DAMBAR PROTRUSION. ALLOWABLE DAMBAR PROTRUSION SHALL BE 0.127 (0.005) TOTAL IN EXCESS OF THE D DIMENSION AT MAXIMUM MATERIAL CONDITION. MAXIMUM MATERIAL CONDITION.

	MILLIN	METERS	INCHES		
DIM	MIN	MAX	MIN MAX		
Α	8.55	8.75	0.337	0.344	
В	3.80	4.00	0.150	0.157	
С	1.35	1.75	0.054	0.068	
D	0.35	0.49	0.014	0.019	
F	0.40	1.25	0.016	0.049	
G	1.27	BSC	0.050 BSC		
J	0.19	0.25	0.008	0.009	
K	0.10	0.25	0.004	0.009	
M	0°	7°	0 °	7°	
Р	5.80	6.20	0.228	0.244	
R	0.25	0.50	0.010	0.019	

#### **P SUFFIX** PLASTIC PACKAGE CASE 646-06 ISSUE L



- LEADS WITHIN 0.13 (0.005) RADIUS OF TRUE
  POSITION AT SEATING PLANE AT MAXIMUM
  MATERIAL CONDITION.
   DIMENSION L TO CENTER OF LEADS WHEN
- FORMED PARALLEL.
  3. DIMENSION B DOES NOT INCLUDE MOLD FLASH.
  4. ROUNDED CORNERS OPTIONAL.

	INC	HES	MILLIN	IETERS	
DIM	MIN	MAX	MIN	MAX	
Α	0.715	0.770	18.16	19.56	
В	0.240	0.260	6.10	6.60	
С	0.145	0.185	3.69	4.69	
D	0.015	0.021	0.38	0.53	
F	0.040	0.070	1.02	1.78	
G	0.100	) BSC	2.54 BSC		
Н	0.052	0.095	1.32	2.41	
J	0.008	0.015	0.20	0.38	
K	0.115	0.135	2.92	3.43	
L	0.300	BSC	7.62 BSC		
M	0°	10°	0°	10°	
N	0.015	0.039	0.39	1.01	

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MC3403/D