

## N-CHANNEL 30V - 0.0075Ω - 60A - DPAK STripFET™ III POWER MOSFET

PRELIMINARY DATA

TYPE	V <sub>DSS</sub>	R <sub>DS(on)</sub>	I <sub>D</sub>		
STD60NH03L	30 V	< 0.009Ω	60 A		
TVDICAL $P_{-n}(an) = 0.00750 \otimes 10.17$					

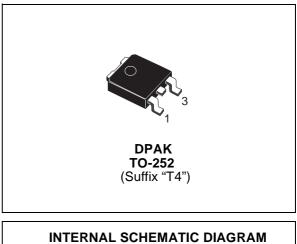
- TYPICAL R<sub>DS</sub>(on) = 0.0075Ω @ 10 V
  TYPICAL R<sub>DS</sub>(on) = 0.009Ω @ 5 V
- R<sub>DS(ON)</sub> \* Q<sub>g</sub> INDUSTRY's BENCHMARK
- CONDUCTION LOSSES REDUCED
- SWITCHING LOSSES REDUCED
- LOW THRESHOLD DEVICE
- SURFACE-MOUNTING DPAK (TO-252) POWER PACKAGE IN TAPE & REEL (SUFFIX "T4")

#### DESCRIPTION

The **STD60NH03L** utilizes the latest advanced design rules of ST's proprietary STripFET<sup>™</sup> technology. This is suitable for the most demanding DC-DC converter application where high efficiency is to be achieved.

#### **APPLICATIONS**

 SPECIFICALLY DESIGNED AND OPTIMISED FOR HIGH EFFICIENCY DC/DC CONVERTERS



#### **ORDERING INFORMATION**

SALES TYPE	MARKING	PACKAGE	PACKAGING
STD60NH03LT4	D60NH03L	DPAK	TAPE & REEL

Symbol	Parameter	Value	Unit
V <sub>DS</sub>	Drain-source Voltage (V <sub>GS</sub> = 0)	30	V
V <sub>DGR</sub>	Drain-gate Voltage ( $R_{GS}$ = 20 k $\Omega$ )	30	V
V <sub>GS</sub>	Gate- source Voltage	± 20	V
I <sub>D</sub>	Drain Current (continuous) at T <sub>C</sub> = 25°C	60	А
ID	Drain Current (continuous) at T <sub>C</sub> = 100°C	43	Α
I <sub>DM</sub> (1)	Drain Current (pulsed)	240	Α
P <sub>TOT</sub>	Total Dissipation at $T_C = 25^{\circ}C$	70	W
	Derating Factor	0.47	W/°C
E <sub>AS</sub> (2)	Single Pulse Avalanche Energy	300	mJ
T <sub>stg</sub>	Storage Temperature	55 to 175	°C
Tj	Max. Operating Junction Temperature	-55 to 175	°C

#### **ABSOLUTE MAXIMUM RATINGS**

#### THERMAL DATA

Rthj-case	Thermal Resistance Junction-case Max	2.14	°C/W
Rthj-amb	Thermal Resistance Junction-ambient Max	100	°C/W
TI	Maximum Lead Temperature for Soldering Purpose	275	°C

# **ELECTRICAL CHARACTERISTICS** (TCASE = 25 °C UNLESS OTHERWISE SPECIFIED) ON/OFF

Symbol	Parameter Test Conditions		Min.	Тур.	Max.	Unit
V <sub>(BR)DSS</sub>	Drain-source Breakdown Voltage	$I_D = 250 \ \mu A, \ V_{GS} = 0$	30			V
I <sub>DSS</sub>	Zero Gate Voltage Drain Current (V <sub>GS</sub> = 0)	V <sub>DS</sub> = Max Rating V <sub>DS</sub> = Max Rating, T <sub>C</sub> = 125 °C			1 10	μΑ μΑ
I <sub>GSS</sub>	Gate-body Leakage Current (V <sub>DS</sub> = 0)	V <sub>GS</sub> = ± 20 V			±100	nA
V <sub>GS(th)</sub>	Gate Threshold Voltage	$V_{DS} = V_{GS}, I_D = 250 \mu A$	1			
R <sub>DS(on)</sub>	Static Drain-source On Resistance	$V_{GS} = 10 \text{ V}, I_D = 30 \text{ A}$ $V_{GS} = 5 \text{ V}, I_D = 30 \text{ A}$		0.0075 0.009	0.009 0.017	Ω Ω

#### ELECTRICAL CHARACTERISTICS (CONTINUED)

#### DYNAMIC

Symbol	Parameter Test Conditions Min.		Min.	Тур.	Max.	Unit
g <sub>fs</sub> (3)	Forward Transconductance	$V_{DS} = 15 \text{ V}$ , $I_{D} = 30 \text{ A}$		TBD		S
Ciss	Input Capacitance	$V_{DS} = 10V, f = 1 \text{ MHz}, V_{GS} = 0$		2200		pF
Coss	Output Capacitance			380		pF
C <sub>rss</sub>	Reverse Transfer Capacitance			49		pF
R <sub>G</sub>	Gate Input Resistance	f=1 MHz Gate DC Bias = 0 Test Signal Level = 20mV Open Drain		1.5		Ω

#### SWITCHING ON

Symbol	Parameter	Test Conditions	Min.	Тур.	Max.	Unit
t <sub>d(on)</sub>	Turn-on Delay Time	V <sub>DD</sub> = 15 V, I <sub>D</sub> = 30 A		21		ns
t <sub>r</sub>	Rise Time	$R_G = 4.7\Omega V_{GS} = 5 V$ (see test circuit, Figure 3)		95		ns
Q <sub>g</sub> Q <sub>gs</sub> Q <sub>gd</sub>	Total Gate Charge Gate-Source Charge Gate-Drain Charge	$V_{DD} = 15 \text{ V}, I_D = 60 \text{ A},$ $V_{GS} = 5 \text{ V}$		15.7 8.3 3.4	21	nC nC nC
Q <sub>gls</sub> (4)	Third-Quadrant Gate Charge	$V_{DS} < 0 V, V_{GS} = 5 V, I_D = 60 A$		15		nC

#### SWITCHING OFF

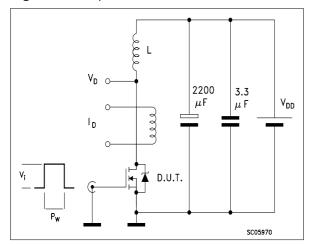
Symbol	Parameter	Test Conditions	Min.	Тур.	Max.	Unit
t <sub>d(off)</sub> t <sub>f</sub>	Turn-off-Delay Time Fall Time	V <sub>DD</sub> = 15 V, I <sub>D</sub> = 30 A, R <sub>G</sub> = 4.7 Ω, V <sub>GS</sub> = 5 V		19 15		ns ns
		(see test circuit, Figure 3)				

#### SOURCE DRAIN DIODE

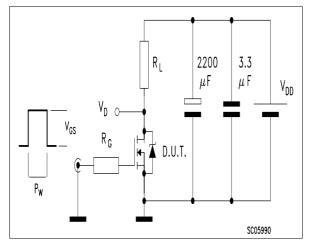
Symbol	Parameter Test Conditions Min. Typ.		Max.	Unit	
I <sub>SD</sub>	Source-drain Current			60	А
I <sub>SDM</sub> (1)	Source-drain Current (pulsed)			240	А
V <sub>SD</sub> (3)	Forward On Voltage	I <sub>SD</sub> = 30 A, V <sub>GS</sub> = 0		1.3	V
t <sub>rr</sub> Q <sub>rr</sub> I <sub>RRM</sub>	Reverse Recovery Time Reverse Recovery Charge Reverse Recovery Current	$I_{SD} = 60 \text{ A}, \text{ di/dt} = 100 \text{A/} \mu \text{s},$ $V_{DD} = 20 \text{ V}, \text{ T}_{\text{j}} = 150 ^{\circ}\text{C}$ (see test circuit, Figure 5)	32 51 3.2		ns nC A

Pulse width limited by safe operating area
 Starting T<sub>j</sub> = 25°C, I<sub>D</sub> = 30A, V<sub>DD</sub> = 20V
 Pulsed: Pulse duration = 300 μs, duty cycle 1.5 %.
 Gate charge for Syncronous Operation. See Appendix A

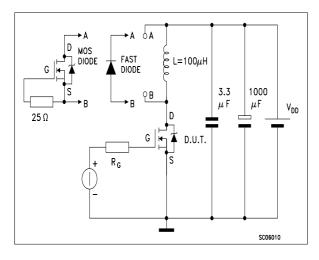
Fig. 1: Unclamped Inductive Load Test Circuit



**Fig. 3:** Switching Times Test Circuit For Resistive Load



**Fig. 5:** Test Circuit For Inductive Load Switching And Diode Recovery Times



#### Fig. 2: Unclamped Inductive Waveform

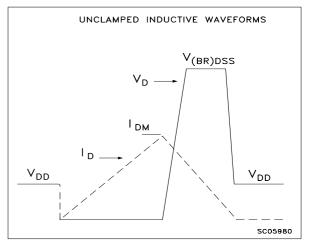
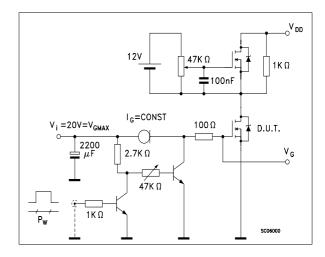
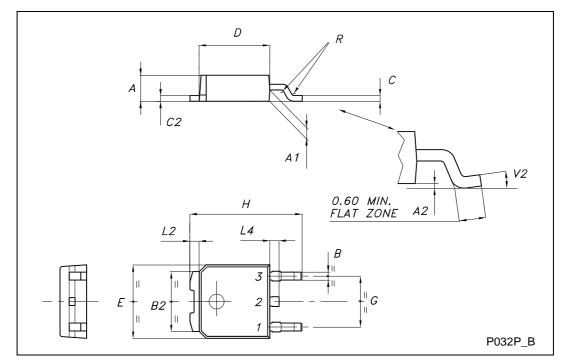


Fig. 4: Gate Charge test Circuit



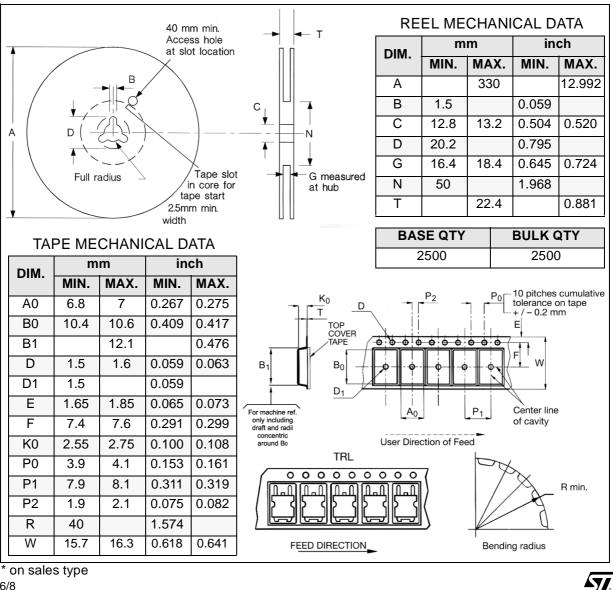
DIM.		mm			inch	
	MIN.	TYP.	MAX.	MIN.	TYP.	MAX.
А	2.20		2.40	0.087		0.094
A1	0.90		1.10	0.035		0.043
A2	0.03		0.23	0.001		0.009
В	0.64		0.90	0.025		0.035
B2	5.20		5.40	0.204		0.213
С	0.45		0.60	0.018		0.024
C2	0.48		0.60	0.019		0.024
D	6.00		6.20	0.236		0.244
E	6.40		6.60	0.252		0.260
G	4.40		4.60	0.173		0.181
Н	9.35		10.10	0.368		0.398
L2		0.8			0.031	
L4	0.60		1.00	0.024		0.039

### TO-252 (DPAK) MECHANICAL DATA



#### **TUBE SHIPMENT (no suffix)\* DPAK FOOTPRINT** 6.7 3.0 0.6 (±0.1) 1.8 1.6 **BASE QTY** 75 2.3 .3 (±0.2) **BULK QTY** 6.7 2.3 3000 5 Tube length 532 (±0.5) All dimensions 1.6 are in millimeters All dimensions are in millimeters 6 (±0.1)

### TAPE AND REEL SHIPMENT (suffix "T4")\*



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#### **Appendix A: Buck Converter Power Losses Estimation**

#### DESCRIPTION

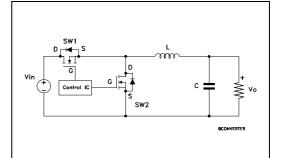
The power losses associated with the FETs in a Synchronous Buck converter can be estimated using the equations shown in the table below. The formulas give a good approximation, for the sake of performance comparison, of how different pairs of devices affect the converter efficiency. However a very important parameter, the working temperature, is not considered. The real device behavior is really dependent on how the heat generated inside the devices is removed to allow for a safer working junction temperature.

#### The low side (SW2) device requires:

- Very low RDS(on) to reduce conduction losses
- Small  $\mathsf{Q}_{\mathsf{gls}}$  to reduce the gate charge losses
- Small  $C_{\mbox{\scriptsize oss}}$  to reduce losses due to output capaci tance
- Small  $\ensuremath{\mathsf{Q}_{\text{rr}}}$  to reduce losses on SW1 during its turn-on
- The C<sub>gd</sub>/C<sub>gs</sub> ratio lower than V<sub>th</sub>/V<sub>GG</sub> ratio especially with low drain to source voltage to avoid the cross conduction phenomenon

#### The high side (SW1) device requires:

- Small  $\rm R_g$  and  $\rm L_s$  to allow higher gate current peak and to limit the voltage feedback on the gate
- Small  $\mathbf{Q}_{g}$  to have a faster commutation and to reduce gate charge losses
- Low R<sub>DS(on)</sub> to reduce the conduction losses



		High Side Switch (SW1)	Low Side Switch (SW2)
Pconduction		$R_{DS(on)SW1}^* I_L^2 * \delta$	$R_{DS(on)SW2} * I_L^2 * (1 - \delta)$
Pswitching		$V_{in} * (Q_{gsth(SW1)} + Q_{gd(SW1)}) * f * \frac{I_L}{I_g}$	Zero Voltage Switching
P <sub>diode</sub>	Recovery	Not Applicable	${}^{1}V_{in} \ast Q_{rr(SW2)} \ast f$
	Conduction	Not Applicable	$V_{\!_{f(SW2)}}*I_{L}*t_{deadtime}*f$
Pgate(Q	)	$Q_{g(SW1)} {}^{*}V_{gg} {}^{*}f$	$Q_{gls(SW2)} * V_{gg} * f$
P <sub>Qoss</sub>		$\frac{\underline{V_{in}}*Q_{css(SW1)}*f}{2}$	$\frac{\frac{V_{in}*Q_{oss(SW2)}*f}{2}}{2}$

Parameter	Meaning
δ	Duty-Cycle
Q <sub>gsth</sub>	Post Threshold Gate Charge
Q <sub>gls</sub>	Third Quadrant Gate Charge
Pconduction	On State Losses
Pswitching	On-off Transition Losses
Pdiode	Conduction and Reverse Recovery Diode Losses
Pdiode	Gate Drive Losses
P <sub>Qoss</sub>	Output Capacitance Losses

1 Dissipated by SW1 during turn-on

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