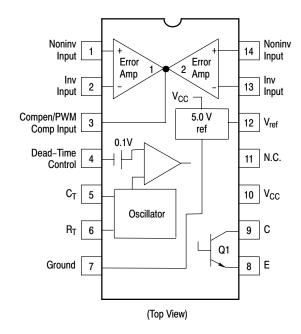
Fixed Frequency, PWM, Voltage Mode Single Ended Controllers

The MC34060A is a low cost fixed frequency, pulse width modulation control circuit designed primarily for single-ended SWITCHMODE $^{\text{TM}}$ power supply control.

The MC34060A is specified over the commercial operating temperature range of 0° to $+70^{\circ}$ C, and the MC33060A is specified over an automotive temperature range of -40° to $+85^{\circ}$ C.

- Complete Pulse Width Modulation Control Circuitry
- On-Chip Oscillator with Master or Slave Operation
- On–Chip Error Amplifiers
- On-Chip 5.0 V Reference, 1.5% Accuracy
- Adjustable Dead–Time Control
- Uncommitted Output Transistor Rated to 200 mA Source or Sink
- Undervoltage Lockout

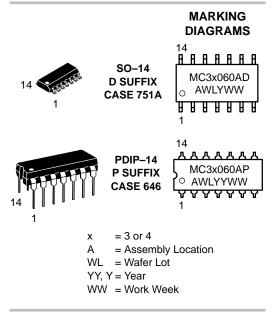
PIN CONNECTIONS



1



http://onsemi.com



ORDERING INFORMATION

See detailed ordering and shipping information in the package dimensions section on page 14 of this data sheet.

MAXIMUM RATINGS (Full operating ambient temperature range applies, unless otherwise noted.)

Rating	Symbol	Value	Unit
Power Supply Voltage	V _{CC}	42	V
Collector Output Voltage	V _C	42	V
Collector Output Current (Note 1)	I _C	500	mA
Amplifier Input Voltage Range	V _{in}	-0.3 to +42	V
Power Dissipation @ T _A ≤ 45°C	P _D	1000	mW
Operating Junction Temperature	TJ	125	°C
Storage Temperature Range	T _{stg}	-55 to +125	°C
Operating Ambient Temperature Range For MC34060A For MC33060A	T _A	0 to +70 -40 to +85	°C

THERMAL CHARACTERISTICS

Characteristics	Symbol	P Suffix Package	D Suffix Package	Unit
Thermal Resistance, Junction-to-Ambient	$R_{\theta JA}$	80	120	°C/W
Derating Ambient Temperature	T _A	45	45	°C

RECOMMENDED OPERATING CONDITIONS

Condition/Value	Symbol	Min	Тур	Max	Unit
Power Supply Voltage	V _{CC}	7.0	15	40	V
Collector Output Voltage	V _C	_	30	40	V
Collector Output Current	I _C	_	_	200	mA
Amplifier Input Voltage	V _{in}	-0.3	_	V _{CC} –2	V
Current Into Feedback Terminal	I _{fb}	-	-	0.3	mA
Reference Output Current	I _{ref}	-	_	10	mA
Timing Resistor	R _T	1.8	47	500	kΩ
Timing Capacitor	C _T	0.00047	0.001	10	μF
Oscillator Frequency	f _{osc}	1.0	25	200	kHz
PWM Input Voltage (Pins 3 and 4)	_	-0.3	_	5.3	V

^{1.} Maximum thermal limits must be observed.

ELECTRICAL CHARACTERISTICS (V_{CC} = 15 V, C_T = 0.01 μ F, R_T = 12 $k\Omega$, unless otherwise noted. For typical values T_A = 25°C, for min/max values T_A is the operating ambient temperature range that applies, unless otherwise noted.)

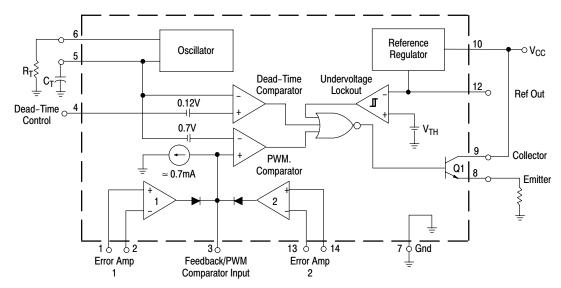
Characteristics	Symbol	Min	Тур	Max	Unit
REFERENCE SECTION					
Reference Voltage (I_O = 1.0 mA, T_A 25°C) T_A = T_{low} to T_{high} – MC34060A – MC33060A	V _{ref}	4.925 4.9 4.85	5.0 - -	5.075 5.1 5.1	V
Line Regulation ($V_{CC} = 7.0 \text{ V to } 40 \text{ V}, I_{O} = 10 \text{ mA}$)	Reg _{line}	-	2.0	25	mV
Load Regulation (I _O = 1.0 mA to 10 mA)	Reg _{load}	-	2.0	15	mV
Short Circuit Output Current (V _{ref} = 0 V)	I _{SC}	15	35	75	mA
OUTPUT SECTION	•	•		•	•
Collector Off–State Current (V _{CC} = 40 V, V _{CE} = 40 V)	I _{C(off)}	-	2.0	100	μΑ
Emitter Off–State Current (V _{CC} = 40 V, V _{CE} = 40 V, V _E = 0 V)	I _{E(off)}	-	_	-100	μΑ
Collector–Emitter Saturation Voltage (Note 2) Common–Emitter $(V_E = 0 \text{ V, } I_C = 200 \text{ mA})$ Emitter–Follower $(V_C = 15 \text{ V, } I_E = -200 \text{ mA})$	V _{sat(C)}	-	1.1	1.5 2.5	V
Output Voltage Rise Time (T _A = 25°C) Common–Emitter (See Figure 12) Emitter–Follower (See Figure 13)	t _r	- -	100 100	200 200	ns
Output Voltage Fall Time (T _A = 25°C) Common–Emitter (See Figure 12) Emitter–Follower (See Figure 13)	t _r	- -	40 40	100 100	ns
ERROR AMPLIFIER SECTION					
Input Offset Voltage (V _{O[Pin 3]} = 2.5 V)	V _{IO}	_	2.0	10	mV
Input Offset Current (V _{C[Pin 3]} = 2.5 V)	I _{IO}	-	5.0	250	nA
Input Bias Current (V _{O[Pin 3]} = 2.5 V)	I _{IB}	-	-0.1	-2.0	μΑ
Input Common Mode Voltage Range (V _{CC} = 40 V)	V _{ICR}	0 to V _{CC} –2.0	-	_	V
Inverting Input Voltage Range	V _{IR(INV)}	-0.3 to V _{CC} -2.0	-	_	V
Open–Loop Voltage Gain ($\Delta V_{\rm O}$ = 3.0 V, $V_{\rm O}$ = 0.5 V to 3.5 V, $R_{\rm L}$ = 2.0 k Ω)	A _{VOL}	70	95	_	dB
Unity–Gain Crossover Frequency (V $_{O}$ = 0.5 V to 3.5 V, R $_{L}$ = 2.0 k Ω)	f _c	-	600	_	kHz
Phase Margin at Unity–Gain (V_O = 0.5 V to 3.5 V, R_L = 2.0 k Ω)	φт	-	65	-	deg.
Common Mode Rejection Ratio (V _{CC} = 40 V, V _{in} = 0 V to 38 V))	CMRR	65	90	_	dB
Power Supply Rejection Ratio $(\Delta V_{CC}=33~V,~V_{O}=2.5~V,~R_{L}=2.0~k\Omega)$	PSRR	-	100	-	dB
Output Sink Current (V _{O[Pin 3]} = 0.7 V)	I ₀ -	0.3	0.7	-	mA
Output Source Current (V _{O[Pin 3]} = 3.5 V)	I _O +	-2.0	-4.0	-	mA

^{2.} Low duty cycle techniques are used during test to maintain junction temperature as close to ambient temperatures as possible.

T_{low} = -40°C for MC33060A = 0°C for MC34060A T_{high} = +85°C for MC33060A = +70°C for MC34060A

ELECTRICAL CHARACTERISTICS (continued) ($V_{CC} = 15 \text{ V}$, $C_T = 0.01 \text{ }\mu\text{F}$, $R_T = 12 \text{ }k\Omega$, unless otherwise noted. For typical values $T_A = 25^{\circ}\text{C}$, for min/max values T_A is the operating ambient temperature range that applies, unless otherwise noted.)

Characteristics	Symbol	Min	Тур	Max	Unit
PWM COMPARATOR SECTION (Test circuit Figure 11)					
Input Threshold Voltage (Zero Duty Cycle)	V _{TH}	-	3.5	4.5	V
Input Sink Current (V _[Pin 3] = 0.7 V)	II	0.3	0.7	_	mA
DEAD-TIME CONTROL SECTION (Test circuit Figure 11)					
Input Bias Current (Pin 4) (V _{in} = 0 V to 5.25 V)	I _{IB(DT)}	-	-1.0	-10	μА
Maximum Output Duty Cycle $(V_{in}=0~V,~C_T=0.01~\mu\text{F},~R_T=12~k\Omega)\\ (V_{in}=0~V,~C_T=0.001~\mu\text{F},~R_T=47~k\Omega)$	DC _{max}	90 -	96 92	100 –	%
Input Threshold Voltage (Pin 4) (Zero Duty Cycle) (Maximum Duty Cycle)	V _{TH}	_ 0	2.8 -	3.3 -	V
OSCILLATOR SECTION					
$ \begin{array}{l} \text{Frequency} \\ (C_T = 0.01 \ \mu\text{F}, \ R_T = 12 \ \text{k}\Omega, \ T_A = 25^{\circ}\text{C}) \\ T_A = T_{low} \ \text{to} \ T_{high} - \text{MC34060A} \\ \qquad - \ \text{MC33060A} \\ (C_T = 0.001 \ \mu\text{F}, \ R_T = 47 \ \text{k}\Omega) \end{array} $	f _{osc}	9.7 9.5 9.0	10.5 - - 25	11.3 11.5 11.5 –	kHz
Standard Deviation of Frequency* $(C_T = 0.001 \; \mu\text{F, R}_T = 47 \; k\Omega)$	of _{osc}	-	1.5	-	%
Frequency Change with Voltage (V _{CC} = 7.0 V to 40 V)	$\Delta f_{ m osc}(\Delta V)$	-	0.5	2.0	%
Frequency Change with Temperature $(\Delta T_A = T_{low} \text{ to } T_{high})$ $(C_T = 0.01 \ \mu\text{F}, R_T = 12 \ \text{k}\Omega)$	$\Delta f_{ m osc}(\Delta T)$	- -	4.0	- -	%
UNDERVOLTAGE LOCKOUT SECTION					
Turn–On Threshold (V _{CC} increasing, I _{ref} = 1.0 mA)	V_{th}	4.0	4.7	5.5	V
Hysteresis	V _H	50	150	300	mV
TOTAL DEVICE					
Standby Supply Current (Pin 6 at V_{ref} , all other inputs and outputs open) ($V_{CC} = 15 \text{ V}$) ($V_{CC} = 40 \text{ V}$)	I _{CC}	- -	5.5 7.0	10 15	mA
Average Supply Current (V _[Pin 4] = 2.0 V, C _T = 0.001 μ F, R _T = 47 k Ω). See Figure 11.	I _S	-	7.0	_	mA



This device contains 46 active transistors.

Figure 1. Block Diagram

Description

The MC34060A is a fixed–frequency pulse width modulation control circuit, incorporating the primary building blocks required for the control of a switching power supply (see Figure 1). An internal–linear sawtooth oscillator is frequency–programmable by two external components, R_T and C_T . The approximate oscillator frequency is determined by:

$$f_{\rm osc} \cong \frac{1.2}{R_{\rm T} \bullet C_{\rm T}}$$

For more information refer to Figure 3.

Output pulse width modulation is accomplished by comparison of the positive sawtooth waveform across capacitor C_T to either of two control signals. The output is enabled only during that portion of time when the sawtooth voltage is greater than the control signals. Therefore, an increase in control—signal amplitude causes a corresponding linear decrease of output pulse width. (Refer to the Timing Diagram shown in Figure 2.)

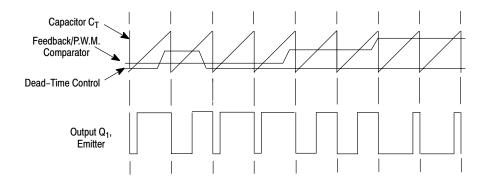


Figure 2. Timing Diagram

APPLICATIONS INFORMATION

The control signals are external inputs that can be fed into the dead–time control, the error amplifier inputs, or the feed–back input. The dead–time control comparator has an effective 120 mV input offset which limits the minimum output dead time to approximately the first 4% of the sawtooth–cycle time. This would result in a maximum duty cycle of 96%. Additional dead time may be imposed on the output by setting the dead time–control input to a fixed voltage, ranging between 0 V to 3.3 V.

The pulse width modulator comparator provides a means for the error amplifiers to adjust the output pulse width from the maximum percent on—time, established by the dead time control input, down to zero, as the voltage at the feedback pin varies from 0.5~V to 3.5~V. Both error amplifiers have a common mode input range from -0.3~V to $(V_{CC}-2.0~V)$, and may be used to sense power supply output voltage and current. The error–amplifier outputs are active high and are ORed together at the noninverting input of the pulse–width modulator comparator. With this configuration, the amplifier that demands minimum output on time, dominates control of the loop.

The MC34060A has an internal 5.0 V reference capable of sourcing up to 10 mA of load currents for external bias circuits. The reference has an internal accuracy of $\pm 5\%$ with a typical thermal drift of less than 50 mV over an operating temperature range of 0° to $+70^{\circ}$ C.

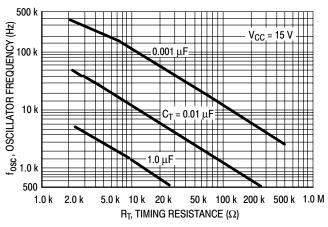


Figure 3. Oscillator Frequency versus Timing Resistance

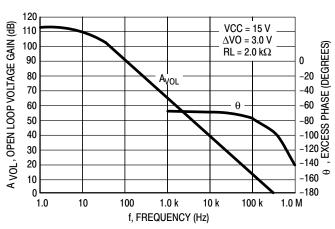


Figure 4. Open Loop Voltage Gain and Phase versus Frequency

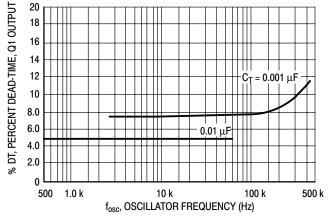


Figure 5. Percent Deadtime versus Oscillator Frequency

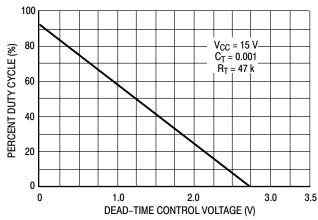


Figure 6. Percent Duty Cycle versus Dead-Time Control Voltage

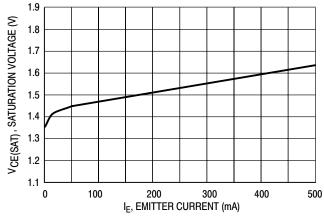


Figure 7. Emitter–Follower Configuration
Output Saturation Voltage versus
Emitter Current

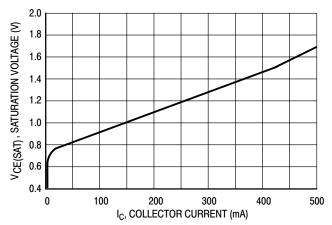


Figure 8. Common–Emitter Configuration
Output Saturation Voltage versus
Collector Current

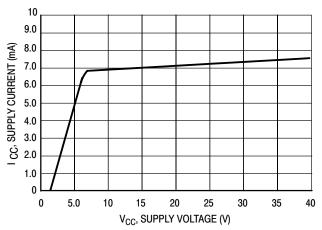


Figure 9. Standby Supply Current versus Supply Voltage

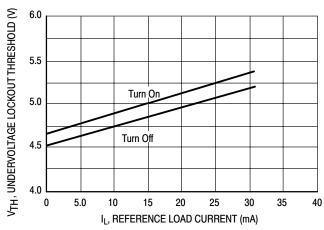


Figure 10. Undervoltage Lockout Thresholds versus Reference Load Current

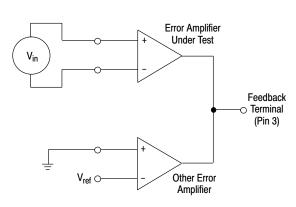


Figure 11. Error Amplifier Characteristics

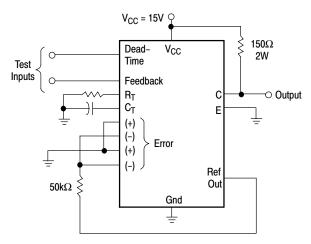


Figure 12. Deadtime and Feedback Control

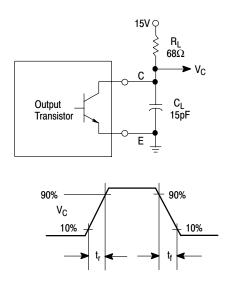


Figure 13. Common–Emitter Configuration and Waveform

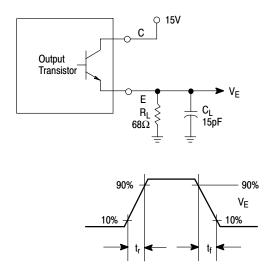


Figure 14. Emitter–Follower Configuration and Waveform

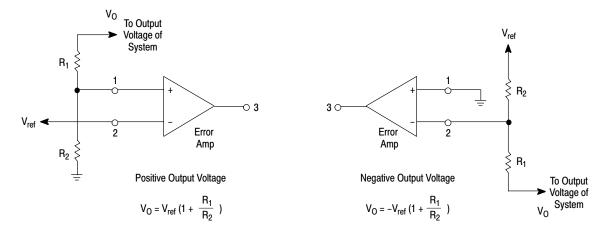


Figure 15. Error Amplifier Sensing Techniques

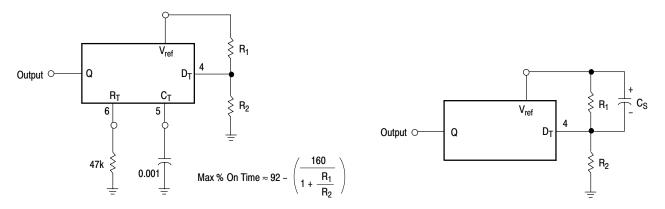


Figure 16. Deadtime Control Circuit

Figure 17. Soft-Start Circuit

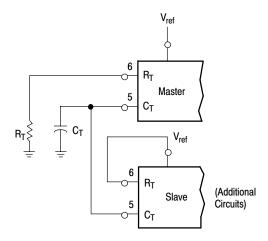
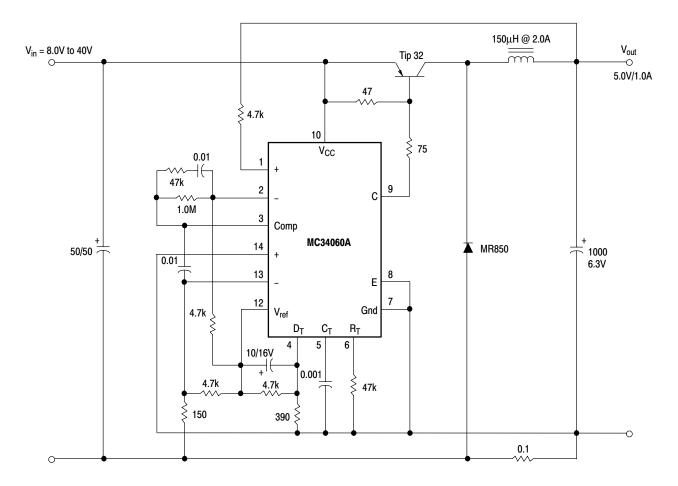
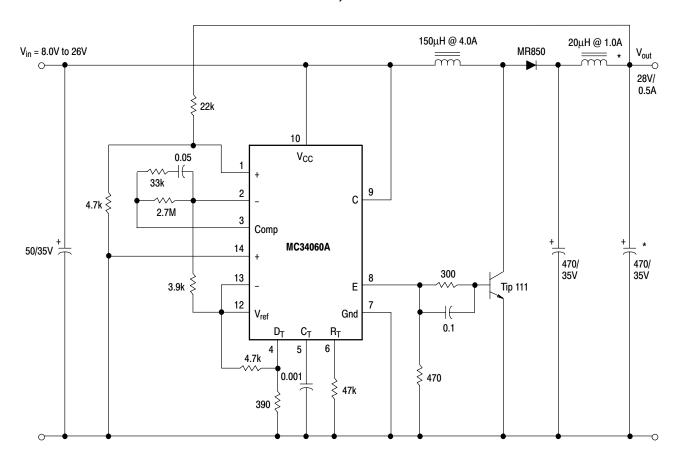


Figure 18. Slaving Two or More Control Circuits



Test	Conditions	Results
Line Regulation	$V_{in} = 8.0 \text{ V to } 40 \text{ V}, I_{O} = 1.0 \text{ A}$	25 mV 0.5%
Load Regulation	$V_{in} = 12 \text{ V}, I_{O} = 1.0 \text{ mA to } 1.0 \text{ A}$	3.0 mV 0.06%
Output Ripple	$V_{in} = 12 \text{ V}, I_{O} = 1.0 \text{ A}$	75 mV p–p P.A.R.D.
Short Circuit Current	$V_{in} = 12 \text{ V}, R_L = 0.1 \Omega$	1.6 A
Efficiency	V _{in} = 12 V, I _O = 1.0 A	73%

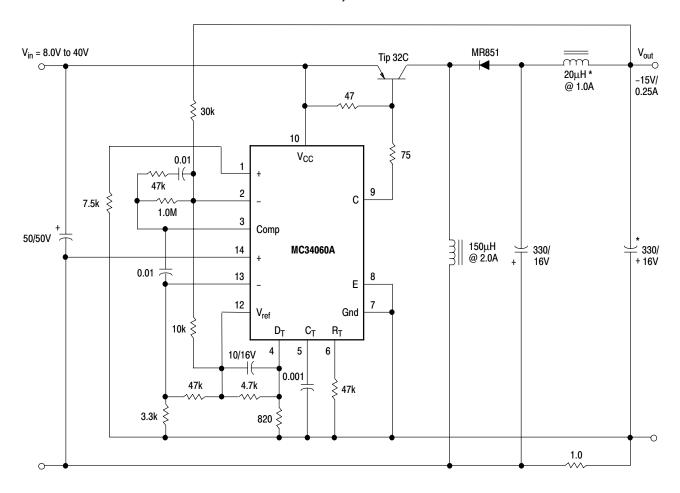
Figure 19. Step-Down Converter with Soft-Start and Output Current Limiting



Test	Conditions	Results
Line Regulation	$V_{in} = 8.0 \text{ V to } 26 \text{ V}, I_{O} = 0.5 \text{ A}$	40 mV 0.14%
Load Regulation	$V_{in} = 12 \text{ V}, I_{O} = 1.0 \text{ mA to } 0.5 \text{ A}$	5.0 mV 0.18%
Output Ripple	$V_{in} = 12 \text{ V}, I_{O} = 0.5 \text{ A}$	24 mV p-p P.A.R.D.
Efficiency	$V_{in} = 12 \text{ V}, I_{O} = 0.5 \text{ A}$	75%

^{*}Optional circuit to minimize output ripple

Figure 20. Step-Up Converter



Test	Conditions	Results	
Line Regulation	V_{in} = 8.0 V to 40 V, I_{O} = 250 mA	52 mV 0.35%	
Load Regulation	$V_{in} = 12 \text{ V}, I_{O} = 1.0 \text{ to } 250 \text{ mA}$	47 mV 0.32%	
Output Ripple	$V_{in} = 12 \text{ V}, I_{O} = 250 \text{ mA}$	10 mV p-p P.A.R.D.	
Short Circuit Current	V_{in} = 12 V, R_L = 0.1 Ω	330 mA	
Efficiency	$V_{in} = 12 \text{ V}, I_{O} = 250 \text{ mA}$	86%	

^{*}Optional circuit to minimize output ripple

Figure 21. Step-Up/Down Voltage Inverting Converter with Soft-Start and Current Limiting

5.0V/3.0A

1N5824

T₂

1N4003

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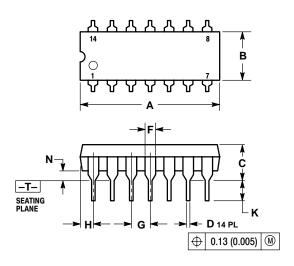
Figure 22. 33 W Off-Line Flyback Converter with Soft-Start and Primary Power Limiting

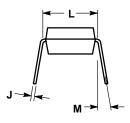
ORDERING INFORMATION

Device	Operating Temperature Range	Package	Shipping
MC34060AD		SO-14	55 Units/Rail
MC34060ADR2	$T_A = 0^\circ$ to $+70^\circ$ C	SO-14	2500 Tape & Reel
MC34060AP		PDIP-14	25 Units/Rail
MC33060AD		SO-14	55 Units/Rail
MC33060ADR2	T _A = -40° to +85°C	SO-14	2500 Tape & Reel
MC33060AP		PDIP-14	25 Units/Rail

PACKAGE DIMENSIONS

PDIP-14 **P SUFFIX** CASE 646-06 ISSUE M



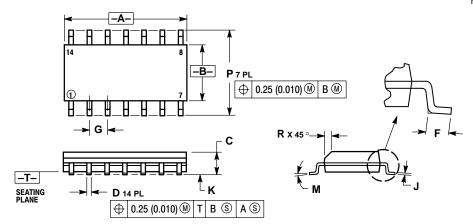


NOTES:

- 1. DIMENSIONING AND TOLERANCING PER ANSI Y14.5M, 1982. 2. CONTROLLING DIMENSION: INCH.
- DIMENSION L TO CENTER OF LEADS WHEN FORMED PARALLEL.
- 4. DIMENSION B DOES NOT INCLUDE MOLD FLASH.
 5. ROUNDED CORNERS OPTIONAL.

	INCHES		MILLIM	ETERS
DIM	MIN	MAX	MIN	MAX
Α	0.715	0.770	18.16	18.80
В	0.240	0.260	6.10	6.60
С	0.145	0.185	3.69	4.69
D	0.015	0.021	0.38	0.53
F	0.040	0.070	1.02	1.78
G	0.100 BSC		2.54	BSC
Н	0.052	0.095	1.32	2.41
J	0.008	0.015	0.20	0.38
K	0.115	0.135	2.92	3.43
L	0.290	0.310	7.37	7.87
M		10°		10°
N	0.015	0.039	0.38	1.01

SO-14 **D SUFFIX** CASE 751A-03 ISSUE F



- NOTES:
 1. DIMENSIONING AND TOLERANCING PER ANSI
- DIMENSIONING AND TOLERANCING PER ANS Y14.5M, 1982.
 CONTROLLING DIMENSION: MILLIMETER.
 DIMENSIONS A AND B DO NOT INCLUDE MOLD PROTRUSION.
 MAXIMUM MOLD PROTRUSION 0.15 (0.006) PER SIDE.
 DIMENSION D DOES NOT INCLUDE DAMBAR PROTRUSION. ALLOWABLE DAMBAR.
- PROTRUSION. ALLOWABLE DAMBAR
 PROTRUSION. ALLOWABLE DAMBAR
 PROTRUSION SHALL BE 0.127 (0.005) TOTAL
 IN EXCESS OF THE D DIMENSION AT
 MAXIMUM MATERIAL CONDITION.

	MILLIMETERS		INC	HES
DIM	MIN	MAX	MIN	MAX
Α	8.55	8.75	0.337	0.344
В	3.80	4.00	0.150	0.157
С	1.35	1.75	0.054	0.068
D	0.35	0.49	0.014	0.019
F	0.40	1.25	0.016	0.049
G	1.27 BSC		0.050	BSC
J	0.19	0.25	0.008	0.009
K	0.10	0.25	0.004	0.009
M	0°	7°	0°	7°
Р	5.80	6.20	0.228	0.244
R	0.25	0.50	0.010	0.019

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