



MONITOR AMPLIFIER

■ PROGRAMMABLE GAIN IN STEPS OF 6 dB

■ ON/OFF POSITION

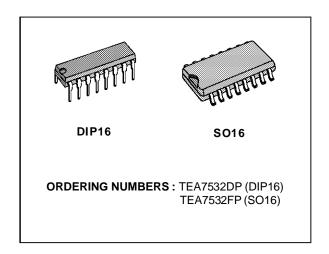
■ LOW VOLTAGE

■ POWER: 100 mW AT 5 V

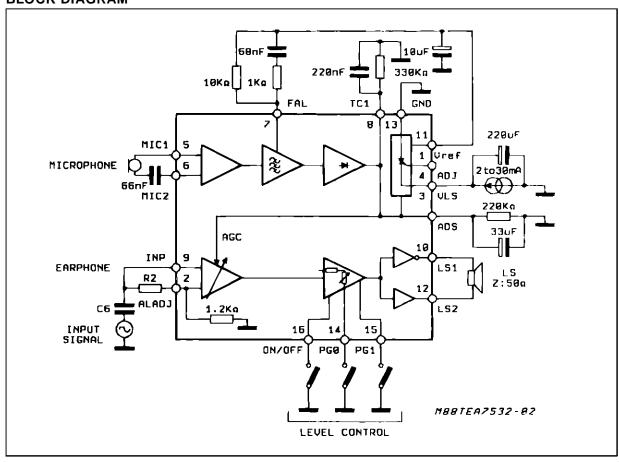
DESCRIPTION

This 16 pins IC is designed for monitor (loudspeaker) telephone set and provides:

- a) Signal amplification for monitoring (loudspeaker)
- b) Antiacoustic feedback (antilarsen)
- c) Antidistortion by automatic gain adaptation
- d) Antilarsen adjustment (full duplex)

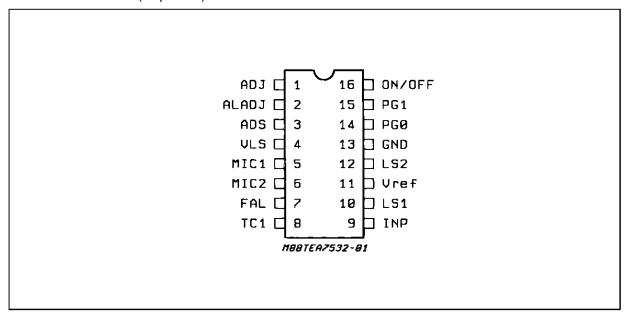


BLOCK DIAGRAM



July 1993

PIN CONNECTION (Top view)



PIN DESCRIPTION

N°	Symbol	Description					
1	ADJ	Adjust Internal Reference V _{LS}					
2	ALADJ	Antilarsen Adjustement					
3	ADS	Antidistortion					
4	V _{LS}	Supply					
5	MIC1	Microphone Input					
6	MIC2	Microphone input					
7	FAL	Antilarsen Filter					
8	TC1	Antilarsen Time Constant					
9	INP	Input Signal					
10	LS1	Output Loudspeaker 1					
11	V_{REF}	Internal Resistance					
12	LS2	Output Loudspeaker 2					
13	GND	Ground					
14 15 16	PG0 PG1 ON/OFF	Inputs Program Level to Loudspeaker					

ABSOLUTE MAXIMUM RATINGS

Symbol	Parameter	Value	Unit
T _{op}	Temperature Range	−5 to + 45	°C
V _{LS}	Supply Voltage	6	V
I _{LS}	Supply Current for T > 300ms for T ≤ 300ms	90 150	mA mA
VL	Voltage Level (pins, PG0, PG1, on/off)	0.6 > to V _S + 0.6	V



ELECTRICAL CHARACTERISTICS (T_{amb} = 25°C, I_{LS} = 30mA unless otherwise specified)

Symbol	Parameter	Test Conditions		Min.	Тур.	Max.	Unit	
V _{LS1}	V _{LS} Supply	I _{LS} = 2mA (fig. 7) I _{LS} = 30mA (fig. 7)		2.6	3.0 3.15	3.4 3.4	V	
V_{LSM}	V _{LS} Maximum	$I_{pin 1} = 50\mu$	A (fig. 7 ; So	= closed)			5.5	V
V_{ADJ}	Voltage Pin 1	$I_{LS} = 2mA to$	o 30mA (fig.	7)	1.1	1.25	1.4	V
G	Loudspeaker Amplifier Gain = $\frac{V_{10} - V_{12}}{V_9}$	ON/OFF	PG0	PG1				
G000 G001 G010 G011 G100		GND GND GND GND VLS	GND GND V _{LS} V _{LS} X	GND V _{LS} GND V _{LS} X	12 18 24 30	14 20 26 32 – 30	16 22 28 34 – 20	88888888888888888888888888888888888888
THD	Distortion	f = 300Hz to 2kHz, $V_{10} - V_{12} = 0.8V_{RMS}$, G = G011, (fig. 8)					2	%
G2	[V(10) - V(12)]/V2	$P_{G0} = P_{G1} =$	$=$ V_{LS} , $V_8 = 0$.8V (fig. 8)	30	32	34	dB
Z _{MICIN}	Microphone Input	Symetrical at (pins 5-6) Asymetrical at (pin 6) fig. 9			4.5		kΩ	
Z _{INPIN}	Earphone Input	(fig. 9)	(fig. 9)		2.2	2.8	3.4	kΩ
Z _{IN2}	Antilarsen Adjustment Input				1	1.2	1.45	kΩ
Voffs	Ouput Offset [V ₍₁₀₎ – V ₍₁₂₎]	G011; (fig. 8)			- 50		50	mV
I _{ON/OFF} I _{PG0} I _{PG1}	Input Current ON State	V _{PG1} = 0V ; (fig. 8)			- 10 - 10 - 10	- 5 - 5 - 5		μΑ μΑ μΑ
I _{ON/OFF} I _{PG0} I _{PG1}	Input Current OFF State	$V_{PG1} = V_{LS}$; (fig. 8)				1 1 1	μΑ μΑ μΑ
VIL ON/OFF VIL PG0 VIL PG1	Input Voltage ON State						0.45 0.45 0.45	V V V
VIH ON/OFF VIH PG0 VIH PG1	Input Voltage OFF State				1.5 1.5 1.5			V V V
Gміс	Microphone Gain = $V_{(7)}/[V_{(5)} - V_{(6)}]$	V _{MIC} = 10m	V _{DMC} f = 2k	Hz (fig. 10)	22.5	23.5	24.5	dB
Vg	Voltage Pin 8	$V_{MIC} = 10 \text{mV}_{RMS}, \text{ f} = 2 \text{kHz (fig. 10)}$		0.48	0.67	0.75	V	
G _{ATT}	Loudspeaker Attenuated Gain = $[V_{(10)} - V_{(12)}]/V_{(9)}$	G011 ; V ₈ = 0.6V ; (fig. 10) G011 ; V ₈ = 0.4V ; (fig. 10)			20	- 30 30	- 20	dB dB

FUNCTIONAL DESCRIPTION

TEA7532 performs the following functions:

The circuit amplifies the incoming signal and feeds it to the loudspeaker. PG0 and PG1 inputs are used to set the loudspeaker gain in a range of 32dB to 14dB in 6dB steps.

The TEA7532 inputs (PG0, PG1, ON/OFF) permit the loudspeaker to be cut-off thus ensuring privacy of communication.

- The antilarsen (antiacoustic feedback) system is incorporated.
- The maximum power available on a 50Ω impedance loudspeaker is 25mW at 3 volts and 100mW at 5V.

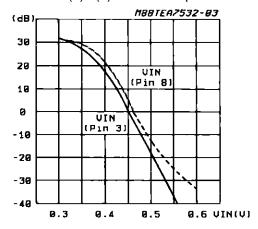
Limit values for external components:

R3 min = 5 k Ω (R3 adjust VLS), R7 max = 390 k Ω , R6 min = R7/35

R max between pin 5 and $6 = 10k\Omega + C \min = 10nF$.



Figure 1 : Loudspeaker Gain Versus Voltage on Pin (3) - (8) with Pin 2 Open.



(*) ATTENUATION = $\frac{\text{Zin2 (1.2 K)}}{\text{Zin2 (1.2 K)} + \text{R2 (E X T)}}$

Figure 3 : AC Output Voltage Versus Amplifier Gain.

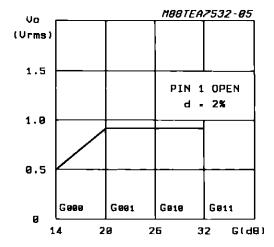


Figure 5: Distortion Versus Output Power.

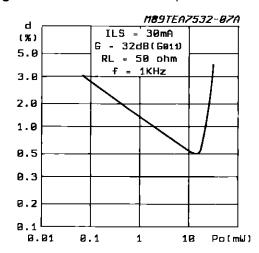
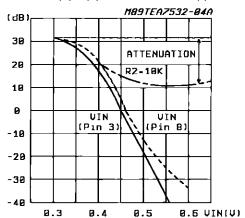


Figure 2: Loudspeaker Gain Versus Voltage on Pin (3) - (8) and Versus R₂. (*)



 $R2 = 10 \text{ K} \Rightarrow ATT = 20 \text{ dB}$

R2 = 3 K \Rightarrow ATT ≈ 10 dB

Figure 4 : Power Available on Loudspeaker Versus V_{LS} Typical Curve.

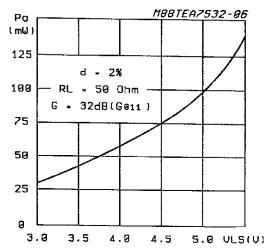
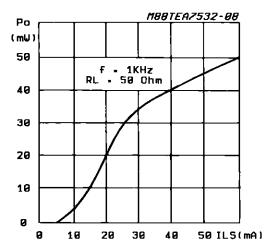


Figure 6: Output Power Versus Supply Current.



TEST CIRCUITS

Figure 7: Shuntvoltage Regulator/ Reference Voltage at Pin 1.

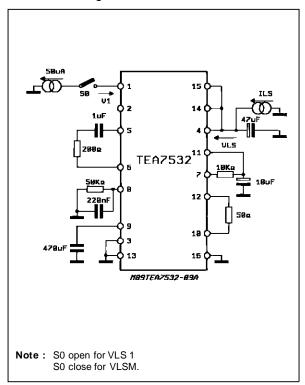


Figure 9: Impedance ZMIC, ZINP and Zin2.

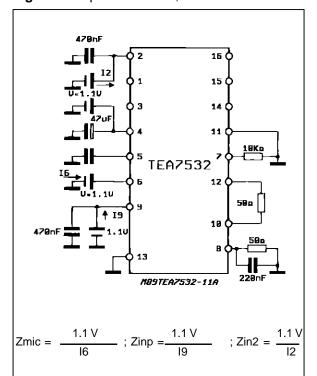


Figure 8 : Loudspeaker Amplifier : Gain/Distor tion/Output Offset.

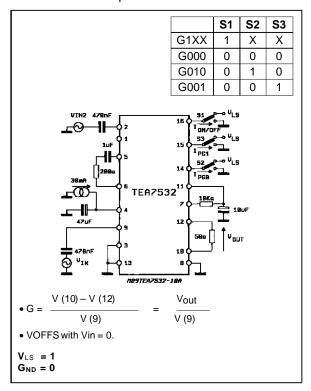
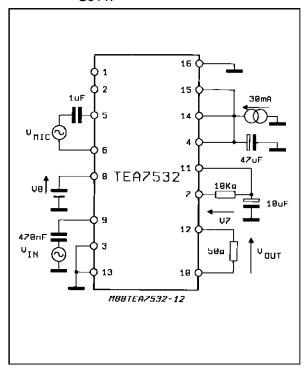


Figure 10 : Antiacoustic Feedback System at G011.



CIRCUIT DESCRIPTION

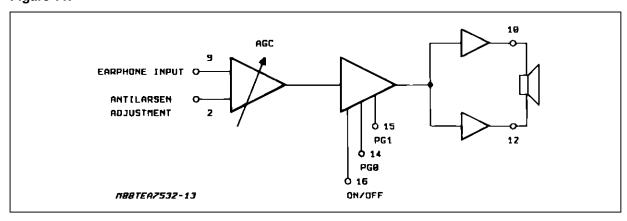
TEA7532 is a 16 pin DIL integrated circuit providing the following facilities:

Loudspeaker amplifier

 Antiacoustic feed-back system (antilarsen system)

1.1. LOUDSPEAKER AMPLIFIER

Figure 11.



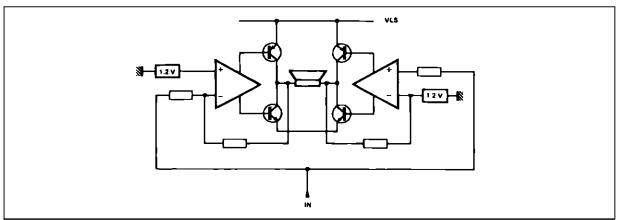
The amplifier is divided into 3 main sections.

- Automatic Gain Control (AGC)
- Preamplifier
- Push-pull amplifier (bridge structure)
- a) The AGC section is used for the antilarsen and antidistortion system.
- When used in a telephone set to avoid larsen effect the AGC automatically decreases loud-speaker amplifier gain.
- When the required output level exceeds the capabilities of the available current, the AGC

decreases the loudspeaker amplifier gain to avoid distortion.

- b) The preamplifier permits step control of amplifier gain in steps of 6 dB, using pins PG0 and PG1, which may be controlled using switches or by a microprocessor. The amplifier may be muted using the ON/OFF control signal (pin 16).
- c) The output amplifier uses a double push-pull configuration (H bridge) to get maximum dynamic range under limited supply conditions.

Figure 12.



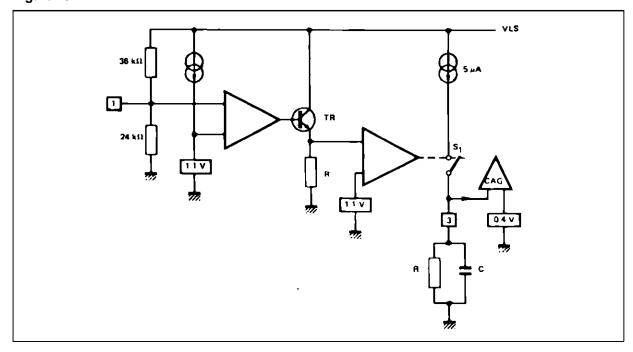
Amplifier dc supply.

In transmission mode, the supply voltage is controlled by the internal shunt DC regulator. For this reason, the TEA7532 should be supplied from a current source

Figure 13.

(see: supply considerations).

An antidistortion system is embodied which provides AGC control to avoid loudspeaker distortion under current-limited conditions.



Circuit action.

When the supply voltage is insufficient, the voltage at pin 1, falls below the reference voltage 1.1 V, resulting in transistor (TR) being switched off, resulting in zero current flow in resistor R. This state enables the gain control system. Under these conditions, the shunt DC supply will switch at a rate determined by the time constant of the RC network on pin 3.

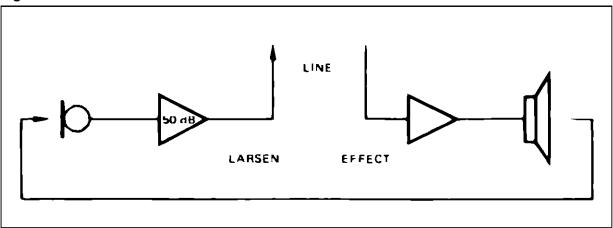
This switching action accomodates normal speech

characteristics under low supply conditions.

1.2. ANTIACOUSTIC FEED-BACK SYSTEM (ANTILARSEN SYSTEM)

The purpose of this system is to control AGC action in order to avoid acoustic feed-back between the loudspeaker and the microphone, when used in a telephone set.

Figure 14.

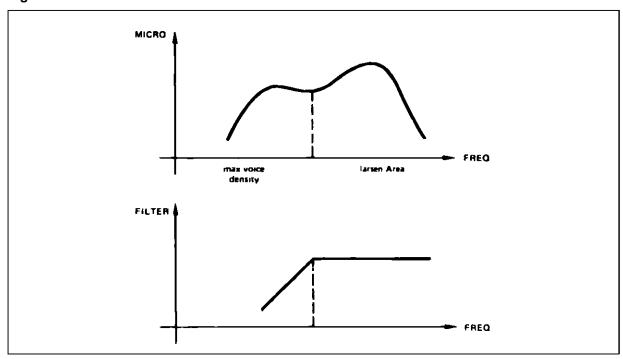


Principle of operation.

When examining the spectral density of the voice area and the larsen area, it can be seen that the

dominant features of each exist in different frequency bands.

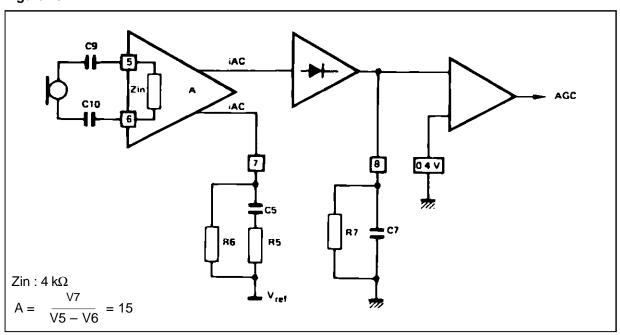
Figure 15.



To extract the larsen component, the microphone signal is first filtered by a second order filter (formed

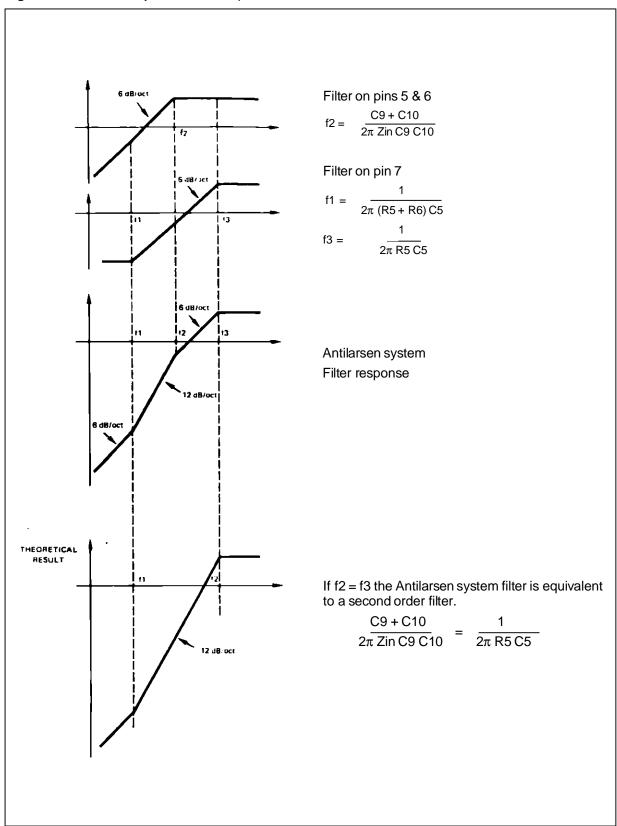
by two first order filters), then amplified and rectified in order to produce the AGC control signal.

Figure 16.



The first filter is generated by the capacitors on pins 5 and 6; the second filter by the R-C network on pin 7.

Figure 17: Antilarsen System Filter Response.

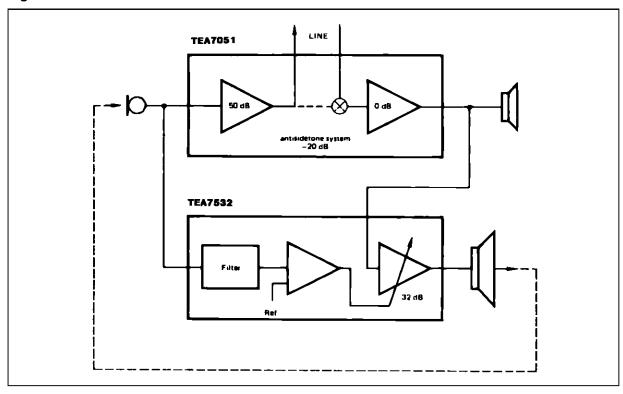


A complete telephone set has two antilarsen systems:

- one in the transmission circuit (for example : TEA7051) antisidetone network;
- one in the loudspeaker amplifier (for example : TEA7532).

Together these form a high efficiency antilarsen system.

Figure 18.



PIN FUNCTIONS

PIN 1: ADJUST VLS

This pin is used to adjust the IC supply voltage.

PIN 2: ANTILARSEN ADJUSTMENT

The AC signal at this pin is amplified to the loudspeaker without AGC attenuation. PIN 3: AUTOMATIC GAIN CONTROL FILTER

The antidistortion system response is adjusted by the R-C network on this pin.

The AGC will be switched ON when the level on pin 3 is greater than the reference voltage (0.4 V), the RC-network charges (current source ON) or discharges (current source OFF) according to the supply voltage.



THEORETICAL VOLTAGE ON PIN 3

Figure 19:

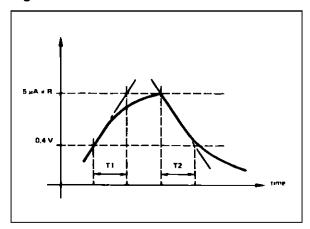
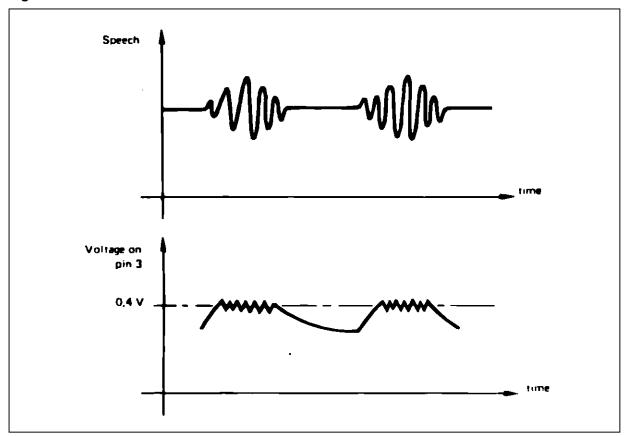


Figure 20.

- The value of R affects the system time constant and the charge/discharge duty cycle.
- The value of C only affects the system time constant.
- R should be greater or equal than $150 \, k\Omega$ for correct AGC operation.



PIN 4: CIRCUIT SUPPLY VOLTAGE With pin 1 open circuit, VLS is internally stabilized at 2.8 V.

When the TEA7532 is under AGC control, the voltage on this pin varies slightly (due to AGC action).

PIN 5/6: MICROPHONE INPUTS

These are used for antilarsen control.

PIN 7: ANTILARSEN FILTER 1

The second filter of the antilarsen system (1 st filter: pins 5-6) is formed by the RC network R5C5. In order to obtain a second order filter for the antilarsen system, the cut-off frequency defined at this pin, should be the same as that chosen for the first filter.

For correct TEA7532 operation R6 and R5 should be fixed at 10 $k\Omega$ and 1 $k\Omega$ respectively.

PIN 8: ANTILARSEN FILTER 2

The gain and the response of the antilarsen system can be adjusted respectively by the resistor and the capacitor on this pin, according to the acoustic characteristics of the telephone set.

The value of the resistor should not exceed 390 k Ω . When the voltage on this pin exceeds the threshold voltage of 0.4 V, the AGC system is enabled.

PIN 9: EARPHONE INPUT

Input for loudspeaker signal.

PIN 10-12: LOUDSPEAKER OUTPUTS

Maximum output voltage : $V_{pp} = 2 V_{LS} - 2.5 V$ (with a gain of 32 dB).

Maximum output current : depending of the supply current.

Two loudspeaker connection methods are possible, using the amplifier in either "H" mode or "B" mode.

Note: It is advisable to connect a 47 nF capacitor in parallel with the loudspeaker (between pins 10 and 12).

- "H" Mode

This is for low voltage working, but at a higher supply current. The highest output power is available in this mode, due to the 5.5 V maximum supply voltage restriction, imposed by the TEA7532.

Loudspeakerimpedance recommended value: 50Ω . Maximum gain available between earphone input and loudspeaker output: 32 dB.

Figure 21.

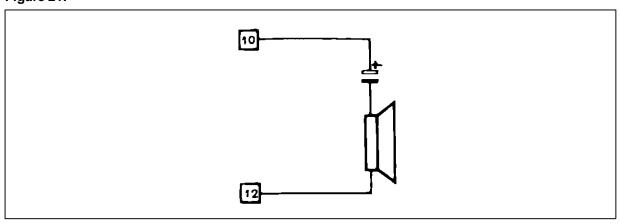
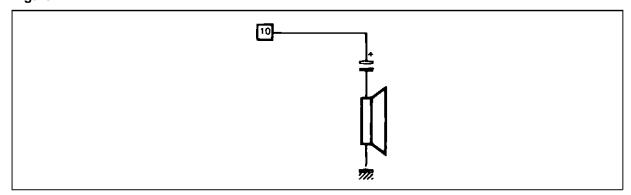


Figure 22.





- "B" Mode

This allows higher voltage operation, but at a lower supply current.

Loudspeakerimpedancerecommended value : 25Ω . Maximum gain available between earphone input and loudspeaker output : 32 - 6 = 26 dB.

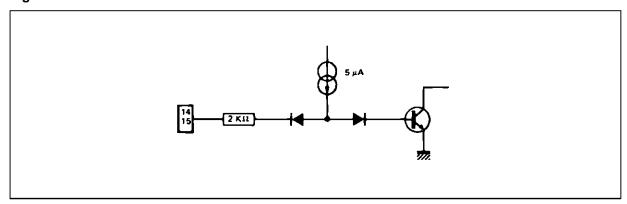
PIN 11: Vref: INTERNAL REFERENCE

Output which provides an internally regulated reference voltage.

Vref = 1.1 V typical

MAXIMUM AVAILABLE CURRENT : 5 μA

Figure 23.



PIN 13: GROUND

PIN 14-15: GAIN ADJUSTMENT INPUTS

These pins are used to adjust the loudspeaker amplifier gain. Four steps of 6 dB/step are available (pin open circuit = high level).

PIN 16: LOUDSPEAKER MUTING.

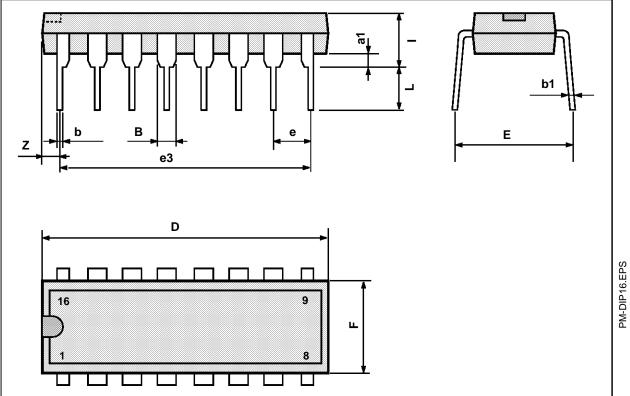
This pin is used to mute the loudspeaker. Pin opencircuit: high level = loudspeaker muted.

Pin low level: louspeaker enabled (see connection of pins 14 and 15).

PG0	PG1				
1	1	Gmax			
1	0	Gmax - <0> 6 dB			
0	1	Gmax - 12 dB			
0	0	Gmax - 18 dB			

DIP16 PACKAGE MECHANICAL DATA

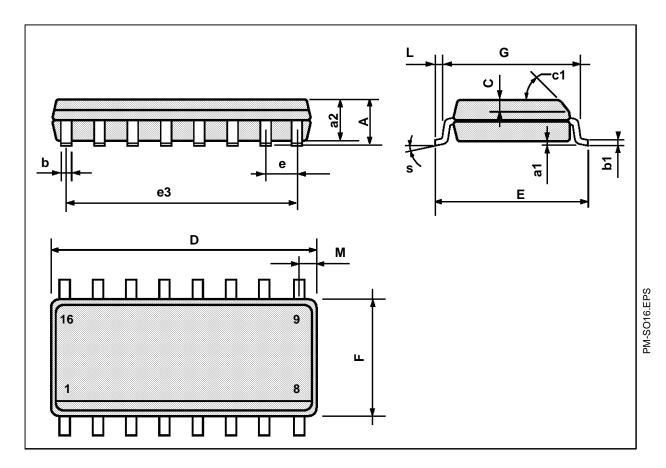
DIM.	mm			inch			
5	Min.	Тур.	Max.	Min.	Тур.	Max.	
a1	0.51			0.020			
В	0.77		1.65	0.030		0.065	
b		0.5			0.020		
b1		0.25			0.010		
D			20			0.787	DIP16 TRI
E		8.5			0.335		1
е		2.54			0.100		
e3		17.78			0.700		
F			7.1			0.280	
i			5.1			0.201	
L		3.3			0.130		
Z			1.27			0.050	



SO16 PACKAGE MECHANICAL DATA

Dimensions		Millimeters			Inches		
Dimensions	Min.	Тур.	Max.	Min.	Тур.	Max.	
Α			1.75			0.069	
a1	0.1		0.2	0.004		0.008	
a2			1.6			0.063	
b	0.35		0.46	0.014		0.018	
b1	0.19		0.25	0.007		0.010	
С		0.5			0.020		
c1		•	45°	(typ.)		•	
D	9.8		10	0.386		0.394	
E	5.8		6.2	0.228		0.244	
е		1.27			0.050		
e3		8.89			0.350		
F	3.8		4.0	0.150		0.157	
G	4.6		5.3	0.181		0.209	
L	0.5		1.27	0.020		0.050	
М			0.62			0.024	
S	8° (max.)						





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