## MC14568B

## Phase Comparator and Programmable Counters

The MC14568B consists of a phase comparator, a divide-by-4, 16, 64 or 100 counter and a programmable divide-by-N 4-bit binary counter (all positive-edge triggered) constructed with MOS P -channel and N -channel enhancement mode devices (complementary MOS) in a monolithic structure.

The MC14568B has been designed for use in conjunction with a programmable divide-by-N counter for frequency synthesizers and phaselocked loop applications requiring low power dissipation and/or high noise immunity.

This device can be used with both counters cascaded and the output of the second counter connected to the phase comparator (CTL high), or used independently of the programmable divide-by-N counter, for example cascaded with a MC14569B, MC14522B or MC14526B (CTL low).

- Supply Voltage Range $=3.0$ to 18 V
- Capable of Driving Two Low-Power TTL Loads, One Low-Power Schottky TTL Load or Two HTL Loads Over the Rated Temperature Range.
- Chip Complexity: 549 FETs or 137 Equivalent Gates

MAXIMUM RATINGS* (Voltages referenced to $\mathrm{V}_{\mathrm{SS}}$ )

| Rating | Symbol | Value | Unit |
| :--- | :---: | :---: | :---: |
| DC Supply Voltage | $\mathrm{V}_{\mathrm{DD}}$ | -0.5 to +18 | Vdc |
| Input Voltage, All Inputs | $\mathrm{V}_{\text {in }}$ | -0.5 to $\mathrm{V}_{\mathrm{DD}}+0.5$ | Vdc |
| DC Input Current, per Pin | $\mathrm{I}_{\text {in }}$ | $\pm 10$ | mAdc |
| Power Dissipation, per Packaget | $\mathrm{P}_{\mathrm{D}}$ | 500 | mW |
| Operating Temperature Range | $\mathrm{T}_{\mathrm{A}}$ | -55 to +125 | ${ }^{\circ} \mathrm{C}$ |
| Storage Temperature Range | $\mathrm{T}_{\text {stg }}$ | -65 to +150 | ${ }^{\circ} \mathrm{C}$ |

* Maximum Ratings are those values beyond which damage to the device may occur. $\dagger$ Temperature Derating:

Plastic "P and D/DW" Packages: - $7.0 \mathrm{~mW} /{ }^{\circ} \mathrm{C}$ From $65^{\circ} \mathrm{C}$ To $125^{\circ} \mathrm{C}$
Ceramic "L" Packages: $-12 \mathrm{~mW} /{ }^{\circ} \mathrm{C}$ From $100^{\circ} \mathrm{C}$ To $125^{\circ} \mathrm{C}$


CTL HIGH


CTL LOW


REV 3
1/94

ELECTRICAL CHARACTERISTICS (Voltages Referenced to $\mathrm{V}_{\text {SS }}$ )

| Characteristic | Symbol | VDD <br> Vdc | $-55^{\circ} \mathrm{C}$ |  | $25^{\circ} \mathrm{C}$ |  |  | $125^{\circ} \mathrm{C}$ |  | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  |  | Min | Max | Min | Typ \# | Max | Min | Max |  |
| Output Voltage$\mathrm{V}_{\mathrm{in}}=\mathrm{V}_{\mathrm{DD}} \text { or } 0$ | V OL | $\begin{aligned} & 5.0 \\ & 10 \\ & 15 \end{aligned}$ | - | $\begin{aligned} & 0.05 \\ & 0.05 \\ & 0.05 \end{aligned}$ | - | $\begin{aligned} & 0 \\ & 0 \\ & 0 \end{aligned}$ | $\begin{aligned} & 0.05 \\ & 0.05 \\ & 0.05 \end{aligned}$ | - | $\begin{aligned} & 0.05 \\ & 0.05 \\ & 0.05 \end{aligned}$ | Vdc |
|  | $\mathrm{V}_{\mathrm{OH}}$ | $\begin{aligned} & 5.0 \\ & 10 \\ & 15 \end{aligned}$ | $\begin{gathered} 4.95 \\ 9.95 \\ 14.95 \end{gathered}$ | - | $\begin{gathered} 4.95 \\ 9.95 \\ 14.95 \end{gathered}$ | $\begin{aligned} & 5.0 \\ & 10 \\ & 15 \end{aligned}$ | - | $\begin{gathered} 4.95 \\ 9.95 \\ 14.95 \end{gathered}$ | - | Vdc |
| $\begin{aligned} & \text { Input Voltage\# } \ddagger \\ & \left(\mathrm{V}_{\mathrm{O}}=4.5 \text { or } 0.5 \mathrm{Vdc}\right) \\ & \left(\mathrm{V}_{\mathrm{O}}=9.0 \text { or } 1.0 \mathrm{Vdc}\right) \\ & \left(\mathrm{V}_{\mathrm{O}}=13.5 \text { or } 1.5 \mathrm{Vdc}\right) \\ & \\ & \left(\mathrm{V}_{\mathrm{O}}=0.5 \text { or } 4.5 \mathrm{Vdc}\right) \\ & \left(\mathrm{V}_{\mathrm{O}}=1.0 \text { or } 9.0 \mathrm{Vdc}\right) \\ & \left(\mathrm{V}_{\mathrm{O}}=1.5 \text { or } 13.5 \mathrm{Vdc}\right) \end{aligned}$ | $\mathrm{V}_{\mathrm{IL}}$ | $\begin{aligned} & 5.0 \\ & 10 \\ & 15 \\ & \hline \end{aligned}$ | - | $\begin{aligned} & 1.5 \\ & 3.0 \\ & 4.0 \end{aligned}$ | - | $\begin{aligned} & 2.25 \\ & 4.50 \\ & 6.75 \end{aligned}$ | $\begin{aligned} & 1.5 \\ & 3.0 \\ & 4.0 \end{aligned}$ | - | $\begin{aligned} & 1.5 \\ & 3.0 \\ & 4.0 \end{aligned}$ | Vdc |
|  | $\mathrm{V}_{\mathrm{IH}}$ | $\begin{aligned} & 5.0 \\ & 10 \\ & 15 \end{aligned}$ | $\begin{aligned} & 3.5 \\ & 7.0 \\ & 11 \end{aligned}$ | - | $\begin{gathered} 3.5 \\ 7.0 \\ 11 \end{gathered}$ | $\begin{aligned} & 2.75 \\ & 5.50 \\ & 8.25 \end{aligned}$ | - | $\begin{aligned} & 3.5 \\ & 7.0 \\ & 11 \end{aligned}$ | - | Vdc |
| Output Drive Current  <br> $(\mathrm{VOH}=2.5 \mathrm{Vdc})$ Source <br> $(\mathrm{VOH}=4.6 \mathrm{Vdc})$  <br> $(\mathrm{VOH}=9.5 \mathrm{Vdc})$  <br> $\left(\mathrm{V}_{\mathrm{OH}}=13.5 \mathrm{Vdc}\right)$  | ${ }^{\mathrm{I}} \mathrm{OH}$ | $\begin{aligned} & 5.0 \\ & 5.0 \\ & 10 \\ & 15 \end{aligned}$ | $\begin{gathered} -1.2 \\ -0.25 \\ -0.62 \\ -1.8 \end{gathered}$ | - | $\begin{aligned} & -1.0 \\ & -0.2 \\ & -0.5 \\ & -1.5 \end{aligned}$ | $\begin{gathered} -1.7 \\ -0.36 \\ -0.9 \\ -3.5 \end{gathered}$ | - | $\begin{gathered} -0.7 \\ -0.14 \\ -0.35 \\ -1.1 \end{gathered}$ | - | mAdc |
| $\begin{aligned} & (\mathrm{VOL}=0.4 \mathrm{Vdc}) \\ & (\mathrm{VOL}=0.5 \mathrm{Vdc}) \\ & \left(\mathrm{V}_{\mathrm{OL}}=1.5 \mathrm{Vdc}\right) \end{aligned}$ | ${ }^{\text {IOL}}$ | $\begin{aligned} & 5.0 \\ & 10 \\ & 15 \end{aligned}$ | $\begin{gathered} \hline 0.64 \\ 1.6 \\ 4.2 \end{gathered}$ | - | $\begin{gathered} 0.51 \\ 1.3 \\ 3.4 \end{gathered}$ | $\begin{gathered} 0.88 \\ 2.25 \\ 8.8 \end{gathered}$ | - | $\begin{gathered} 0.36 \\ 0.9 \\ 2.4 \end{gathered}$ | - | mAdc |
| Input Current | lin | 15 | - | $\pm 0.1$ | - | $\pm 0.00001$ | $\pm 0.1$ | - | $\pm 1.0$ | $\mu \mathrm{Adc}$ |
| Input Capacitance | $\mathrm{C}_{\text {in }}$ | - | - | - | - | 5.0 | 7.5 | - | - | pF |
| ```Quiescent Current (Per Package) Vin = 0 or V DD, lout = 0 \muA``` | IDD | $\begin{aligned} & \hline 5.0 \\ & 10 \\ & 15 \end{aligned}$ | - | $\begin{aligned} & 5.0 \\ & 10 \\ & 20 \end{aligned}$ | - | $\begin{aligned} & \hline 0.005 \\ & 0.010 \\ & 0.015 \end{aligned}$ | $\begin{aligned} & 5.0 \\ & 10 \\ & 20 \end{aligned}$ | - | $\begin{aligned} & 150 \\ & 300 \\ & 600 \end{aligned}$ | $\mu \mathrm{Adc}$ |
| Total Supply Current** $\dagger$ <br> (Dynamic plus Quiescent, Per Package) ( $C_{L}=50 \mathrm{pF}$ on all outputs, all buffers switching) | ${ }^{1} \mathrm{~T}$ | $\begin{aligned} & 5.0 \\ & 10 \\ & 15 \end{aligned}$ | $\begin{aligned} & \mathrm{I} T=(0.2 \mu \mathrm{~A} / \mathrm{kHz}) \mathrm{f}+\mathrm{IDD} \\ & \mathrm{IT}=(0.4 \mu \mathrm{~A} / \mathrm{kHz}) \mathrm{f}+\mathrm{IDD} \\ & \mathrm{I} \mathrm{~T}=(0.9 \mu \mathrm{~A} / \mathrm{kHz}) \mathrm{f}+\mathrm{IDD} \end{aligned}$ |  |  |  |  |  |  | $\mu \mathrm{Adc}$ |
| Three-State Leakage Current Pins 1, 13 | ${ }^{1} \mathrm{TL}$ | 15 | - | $\pm 0.1$ | - | $\pm 0.0001$ | $\pm 0.1$ | - | $\pm 3.0$ | $\mu \mathrm{Adc}$ |

\#Noise immunity for worst input combination.
Noise Margin for both "1" and " 0 " level $=1.0 \mathrm{~V}$ min $@ \mathrm{~V}_{\mathrm{DD}}=5.0 \mathrm{~V}$

$$
\begin{aligned}
& 2.0 \mathrm{~V} \min @ V_{D D}=10 \mathrm{~V} \\
& 2.5 \mathrm{~V} \text { min } @ \mathrm{~V}_{\mathrm{DD}}=15 \mathrm{~V}
\end{aligned}
$$

$\dagger$ To calculate total supply current at loads other than 50 pF :

$$
I_{T}\left(C_{L}\right)=I_{T}(50 \mathrm{pF})+1 \times 10^{-3}\left(C_{L}-50\right) V_{D D^{f}}
$$

where: $\mathrm{I}_{\mathrm{T}}$ is in $\mu \mathrm{A}$ (per package), $\mathrm{C}_{\mathrm{L}}$ in $\mathrm{pF}, \mathrm{V}_{\mathrm{DD}}$ in V , and f in kHz is input frequency.
** The formulas given are for the typical characteristics only at $25^{\circ} \mathrm{C}$.
$\ddagger$ Pin 15 is connected to $V_{S S}$ or $V_{D D}$ for input voltage test.
PIN ASSIGNMENT


SWITCHING CHARACTERISTICS $\left(C_{L}=50 \mathrm{pF}, \mathrm{T}_{\mathrm{A}}=25^{\circ} \mathrm{C}\right)$

| Characteristic | Symbol | $\overline{\mathrm{V}_{\mathrm{DD}}}$ | Min | Typ | Max | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Output Rise Time | tTLH | $\begin{aligned} & 5.0 \\ & 10 \\ & 15 \end{aligned}$ | - | $\begin{gathered} 180 \\ 90 \\ 65 \end{gathered}$ | $\begin{aligned} & 360 \\ & 180 \\ & 130 \end{aligned}$ | ns |
| Output Fall Time | ${ }_{\text {t }}$ HL | $\begin{aligned} & 5.0 \\ & 10 \\ & 15 \end{aligned}$ | - | $\begin{aligned} & 100 \\ & 50 \\ & 40 \end{aligned}$ | $\begin{aligned} & \hline 200 \\ & 100 \\ & 80 \end{aligned}$ | ns |
| Minimum Pulse Width, C1, Q1/C2, or PC in Input | tWH | $\begin{aligned} & 5.0 \\ & 10 \\ & 15 \end{aligned}$ | - | $\begin{aligned} & \hline 125 \\ & 60 \\ & 45 \end{aligned}$ | $\begin{aligned} & \hline 250 \\ & 120 \\ & 90 \end{aligned}$ | ns |
| Maximum Clock Rise and Fall Time, C1, Q1/C2, or PC in Input $^{\text {I }}$ | $\begin{aligned} & \hline \mathrm{t} \mathrm{TLH}, \\ & \mathrm{t} \text { THL } \end{aligned}$ | $\begin{aligned} & 5.0 \\ & 10 \\ & 15 \end{aligned}$ | $\begin{aligned} & \hline 15 \\ & 15 \\ & 15 \end{aligned}$ | - | - | $\mu \mathrm{s}$ |

PHASE COMPARATOR

| Input Resistance | $\mathrm{R}_{\text {in }}$ | 5.0 to 15 | - | $10^{6}$ | - | $\mathrm{M} \Omega$ |
| :--- | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Input Sensitivity, dc Coupled | - | 5.0 to 15 | See Input Voltage |  |  |  |
| Turn-Off Delay Time, | tPHL | 5.0 | - | 550 | 1100 | ns |
| PC $_{\text {out }}$ and LD Outputs |  | 10 | - | 195 | 390 |  |
|  |  | 15 | - | 120 | 240 |  |
| Turn-On Delay Time. | tPLH | 5.0 | - | 675 | 1350 | ns |
| PC $_{\text {out }}$ and LD Outputs |  | 10 | - | 300 | 600 |  |
|  |  | 15 | - | 190 | 380 |  |

DIVIDE-BY-4, 16, 64 OR 100 COUNTER (D1)

| Maximum Clock Pulse Frequency Division Ratio $=4,64$ or 100 | ${ }_{\mathrm{c}} \mathrm{l}$ | 5.0 10 15 | $\begin{aligned} & 3.0 \\ & 8.0 \\ & 10 \end{aligned}$ | $\begin{aligned} & 6.0 \\ & 16 \\ & 22 \end{aligned}$ | - | MHz |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Division Ratio $=16$ |  | $\begin{aligned} & \hline 5.0 \\ & 10 \\ & 15 \end{aligned}$ | $\begin{aligned} & 1.0 \\ & 3.0 \\ & 50 \end{aligned}$ | $\begin{aligned} & \hline 2.5 \\ & 6.3 \\ & 9.7 \end{aligned}$ | - |  |
| Propagation Delay Time, Q1/C2 Output Division Ratio $=4,64$ or 100 | $\begin{aligned} & \text { tPLH, } \\ & \text { tPHL } \end{aligned}$ | $\begin{aligned} & 5.0 \\ & 10 \\ & 15 \end{aligned}$ | - | $\begin{aligned} & 450 \\ & 190 \\ & 130 \end{aligned}$ | $\begin{aligned} & 900 \\ & 380 \\ & 260 \end{aligned}$ | ns |
| Division Ratio $=16$ |  | $\begin{aligned} & \hline 5.0 \\ & 10 \\ & 15 \end{aligned}$ | - | $\begin{aligned} & 720 \\ & 300 \\ & 200 \end{aligned}$ | $\begin{gathered} \hline 1440 \\ 600 \\ 400 \end{gathered}$ |  |

PROGRAMMABLE DIVIDE-BY-N 4-BIT COUNTER (D2)

| Maximum Clock Pulse Frequency <br> (Figure 3a) | $\mathrm{f}_{\mathrm{Cl}}$ | 5.0 | 1.2 | 1.8 | - | MHz |
| :--- | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  | 10 | 3.0 | 8.5 | - |  |
| Turn-On Delay Time, "0" Output |  | 15 | 4.0 | 12 | - |  |
| (Figure 3a) | tPLH | 5.0 | - | 450 | 900 | ns |
|  |  | 10 | - | 190 | 380 |  |
| Turn-Off Delay Time, "0" Output | 15 | - | 130 | 260 |  |  |
| (Figure 3a) |  | tPHL | 5.0 | - | 225 | 450 |
|  | 10 | - | 85 | 170 | ns |  |
|  |  | 15 | - | 60 | 150 |  |
| Minimum Preset Enable Pulse Width |  |  | - | 75 | 250 | ns |
|  |  | tWH(PE) | 5.0 | - | 40 | 100 |
|  | 10 | - | 30 | 75 |  |  |

## SWITCHING TIME TEST CIRCUITS AND WAVEFORMS



Figure 2. Counter D1

a.

Figure 3. Counter D2
b.


Typical Maximum Frequency Divider D1 Division ratios: 4, 64 or $100\left(C_{L}=50 \mathrm{pF}\right)$


Typical Maximum Frequency Divider D1 Division ratio: 16 ( $\left.C_{L}=50 \mathrm{pF}\right)$


Typical Maximum Frequency Divider D2 Division ratio: $2\left(C_{L}=50 \mathrm{pF}\right)$


## OPERATING CHARACTERISTICS

The MC14568B contains a phase comparator, a fixed divider $(\div 4, \div 16, \div 64, \div 100)$ and a programmable divide-by-N 4-bit counter.

## PHASE COMPARATOR

The phase comparator is a positive edge controlled logic circuit. It essentially consists of four flip-flops and an output pair of MOS transistors. Only one of its inputs ( $\mathrm{PC}_{\mathrm{in}}$, pin 14) is accessible externally. The second is connected to the output of one of the two counters D1 or D2 (see block diagram).

Duty cycles of both input signals (at A and B) need not be taken into consideration since the comparator responds to leading edges only.

If both input signals have identical frequencies but different phases, with signal A (pin 14) leading signal B (Ref.), the comparator output will be high for the time equal to the phased difference.

If signal A lags signal B, the output will be low for the same time. In between, the output will be in a three-state condition and the voltage on the capacitor of an RC filter normally connected at this point will have some intermediate value (see Figure 4). When used in a phase locked loop, this value will adjust the Voltage Controlled Oscillator frequency by reducing the phase difference between the reference signal and the divided VCO frequency to zero.


Figure 4. Phase Comparator Waveforms

If the input signals have different frequencies, the output signal will be high when signal $B$ has a lower frequency than signal A, and low otherwise.

Under the same conditions of frequency difference, the output will vary between $\mathrm{V}_{\mathrm{OH}}$ (or $\mathrm{V}_{\mathrm{OL}}$ ) and some intermediate value until the frequencies of both signals are equal and their phase difference equal to zero, i.e. until locked condition is obtained.

Capture and lock range will be determined by the VCO frequency range. The comparator is provided with a lock indicator output, which will stay at logic 1 in locked conditions.

The state diagram (Figure 5) depicts the internal state transitions. It assumes that only one transition on either signal occurs at any time. It shows that a change of the output state is always associated with a positive transition of either signal. For a negative transition, the output does not change state. A positive transition may not cause the output to change, this happens when the signals have different frequencies.

## DIVIDE BY 4, 16, 64 OR 100 COUNTER (D1)

This counter is able to work at an input frequency of 5 MHz for a $V_{D D}$ value of 10 volts over the standard temperature range when dividing by 4,64 and 100. Programming is accomplished by use of inputs F and G (pins 10 and 11) according to the truth table shown. Connecting the Control input (CTL, pin 15), to VDD allows cascading this counter with the programmable divide-by-N counter provided in the same package. Independent operation is obtained when the Control input is connected to VSS.

The different division ratios have been chosen to generate the reference frequencies corresponding to the channel spacings normally required in frequency synthesizer applications. For example. with the division ratio 100 and a 5 MHz crystal stabilized source a reference frequency of 50 kHz is supplied to the comparator. The lower division ratios permit operation with low frequency crystals.

| INPUT STATE |  |  |
| :---: | :---: | :---: |
| PC ${ }_{\text {Out }}$ | 0 | OUTPUT DISCONNECTED |

Figure 5. Phase Comparator State Diagram

If used in cascade with the programmable divide-by-N counter, practically all usual reference frequencies, or channel spacings of $25,20,12.5,10,6.25 \mathrm{kHz}$, etc. are easily achievable.

## PROGRAMMABLE DIVIDE-BY-N <br> 4-BIT COUNTER (D2)

This counter is programmable by using inputs DP0 ... DP3
(pins 7 ... 4). The Preset Enable input enables the parallel preset inputs DP0... DP3. The "0" output must be externally connected to the PE input for single stage applications. Since there is not a cascade feedback input, this counter, when cascaded, must be used as the most significant digit. Because of this, it can be cascaded with binary counters as well as with BCD counters (MC14569B, MC14522B, MC14526B).

## TYPICAL APPLICATIONS



Figure 6. Cascading MC14568B and MC14522B or MC14526B with MC14569B

(143.5 MHz)

Figure 7. Frequency Synthesizer with MC14568B and MC14569B Using a Mixer (Channel Spacing 10 kHz)


```
Divide ratio = 160N N + 16N2 + N3
Example:
fout = N (MHz) + N2 (x 100 kHz) + N N (x25 kHz)
Frequency range = 5 MHz
Channel spacing = 25 kHz
Reference frequency = 6.25 kHz
```

Figure 8. Frequency Synthesizer Using MC14568B, MC14569B and MC14522B (Without Mixer)


Figure 9. Typical 23-Channel CB Frequency Synthesizer for Double Conversion Transceivers


## D SUFFIX

PLASTIC SOIC PACKAGE
CASE 751B-05
ISSUE J


NOTES:

1. Dimensioning and tolerancing per ansi Y14.5M, 1982.
2. CONTROLLING DIMENSION: MLLIIMETER.
3. DIMENSIONS A AND B DO NOT INCLUDE MOLD PROTRUSION.
4. MAXIMUM MOLD PROTRUSION 0.15 (0.006) PERSIDE.
5. DIMENSION D DOES NOT INCLUDE DAMBAR PROTRUSION. ALLOWABLE DAMBAR PROTTUSION. ALLOLABELEDABAR
PRTRUSION SHAL LE $0.127(0.005)$ TOTAL
 MAXIMUM MATERIAL CONDITION.

|  | MILLIMETERS |  | INCHES |  |  |  |
| :---: | ---: | ---: | ---: | ---: | :---: | :---: |
| DIM | MIN | MAX | MIN | MAX |  |  |
| A | 9.80 | 10.00 | 0.386 | 0.393 |  |  |
| B | 3.80 | 4.00 | 0.150 | 0.157 |  |  |
| C | 1.35 | 1.75 | 0.054 | 0.068 |  |  |
| D | 0.35 | 0.49 | 0.014 | 0.019 |  |  |
| F | 0.40 | 1.25 | 0.016 | 0.049 |  |  |
| G | 1.27 |  | BSC | 0.050 |  | BSC |
| J | 0.19 | 0.25 | 0.008 | 0.009 |  |  |
| K | 0.10 | 0.25 | 0.004 | 0.009 |  |  |
| M | 0 | $7^{\circ}$ | $0^{\circ}$ | $7^{\circ}$ |  |  |
| P | 5.80 | 6.20 | 0.229 | 0.244 |  |  |
| R | 0.25 | 0.50 | 0.010 | 0.019 |  |  |


#### Abstract

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