

# PIC16F610/16HV610 PIC16F616/16HV616 Data Sheet

14-Pin, Flash-Based 8-Bit

**CMOS** Microcontrollers

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# MICROCHIP PIC16F610/616/16HV610/616

### 14-Pin Flash-Based, 8-Bit CMOS Microcontrollers

### High-Performance RISC CPU:

- Only 35 Instructions to Learn:
  - All single-cycle instructions except branches
- Operating Speed:
  - DC 20 MHz oscillator/clock input
  - DC 200 ns instruction cycle
- Interrupt Capability
- 8-Level Deep Hardware Stack
- Direct, Indirect and Relative Addressing modes

### **Special Microcontroller Features:**

- Precision Internal Oscillator:
- Factory calibrated to ±1%, typical
- User selectable frequency: 4 MHz or 8 MHz
- Power-Saving Sleep mode
- Voltage Range:
  - PIC16F610/616: 2.0V to 5.5V
  - PIC16HV610/616: 2.0V to user defined maximum (see note)
- Industrial and Extended Temperature Range
- Power-on Reset (POR)
- Power-up Timer (PWRT) and Oscillator Start-up Timer (OST)
- Brown-out Reset (BOR)
- Watchdog Timer (WDT) with Independent Oscillator for Reliable Operation
- Multiplexed Master Clear with Pull-up/Input Pin
- Programmable Code Protection
- High Endurance Flash:
  - 100,000 write Flash endurance
  - Flash retention: > 40 years

### Low-Power Features:

Standby Current:

Note:

- 50 nA @ 2.0V, typical
- Operating Current:
  - 20 μA @ 32 kHz, 2.0V, typical

cannot exceed 5V.

Voltage across internal shunt regulator

- 220  $\mu\text{A}$  @ 4 MHz, 2.0V, typical
- Watchdog Timer Current:
  - 1 μA @ 2.0V, typical

### **Peripheral Features:**

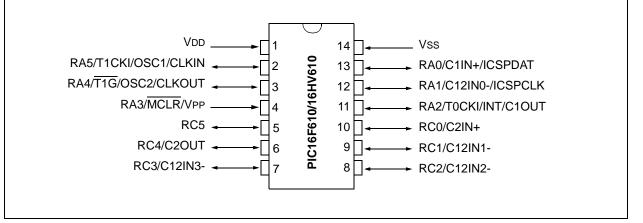
- Shunt Voltage Regulator (PIC16HV610/616 only):
  - 5 volt regulation
  - 4 mA to 50 mA shunt range
- 11 I/O Pins and 1 Input Only
  - High current source/sink for direct LED drive
  - Interrupt-on-Change pins
  - Individually programmable weak pull-ups
- Analog Comparator module with:
  - Two analog comparators
  - Programmable on-chip voltage reference (CVREF) module (% of VDD)
  - Fixed Voltage Reference
  - Comparator inputs and outputs externally accessible
  - SR Latch
  - Built-In Hysteresis (user selectable)
- Timer0: 8-Bit Timer/Counter with 8-Bit Programmable Prescaler
- Enhanced Timer1:
  - 16-bit timer/counter with prescaler
  - External Timer1 Gate (count enable)
  - Option to use OSC1 and OSC2 in LP mode as Timer1 oscillator if INTOSC mode selected
  - Timer1 oscillator
- In-Circuit Serial Programming<sup>™</sup> (ICSP<sup>™</sup>) via Two Pins

### PIC16F616/16HV616 only:

- A/D Converter:
  - 10-bit resolution
  - 8 external input channels
  - 2 internal reference channels
- Timer2: 8-Bit Timer/Counter with 8-Bit Period Register, Prescaler and Postscaler
- Enhanced Capture, Compare, PWM module:
  - 16-bit Capture, max. resolution 12.5 ns
  - 16-bit Compare, max. resolution 200 ns
  - 10-bit PWM with 1, 2 or 4 output channels, programmable "dead time", max. frequency 20 kHz

	Program Memory	Data Memory		10-bit A/D		Timers	Voltage Range	
Device	Flash (words)	SRAM (bytes)	I/O	(ch)	Comparators	8/16-bit		
PIC16F610	1024	64	11	—	2	1/1	2.0-5.5V	
PIC16HV610	1024	64	11	—	2	1/1	2.0-user defined	
PIC16F616	2048	128	11	8	2	2/1	2.0-5.5V	
PIC16HV616	2048	128	11	8	2	2/1	2.0-user defined	

### PIC16F610/16HV610 14-Pin Diagram (PDIP, SOIC, TSSOP)

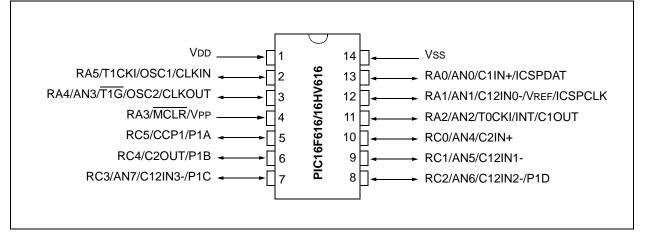


### TABLE 1: PIC16F610/16HV610 14-PIN SUMMARY

I/O	Pin	Comparators	Timer Interrupts		Pull-ups	Basic
RA0	13	C1IN+	_	IOC	Y	ICSPDAT
RA1	12	C12IN0-	_	IOC	Y	ICSPCLK
RA2	11	C1OUT	T0CKI	INT/IOC	Y	—
RA3 <sup>(1)</sup>	4	_	_	IOC	Y <sup>(2)</sup>	MCLR/Vpp
RA4	3	—	T1G	IOC	Y	OSC2/CLKOUT
RA5	2	—	T1CKI	IOC	Y	OSC1/CLKIN
RC0	10	C2IN+	_	—	—	—
RC1	9	C12IN1-		—	_	—
RC2	8	C12IN2-	—	—	—	—
RC3	7	C12IN3-		_	—	—
RC4	6	C2OUT	-	—	—	—
RC5	5	_		_	_	_
	1	_	_	_		Vdd
	14				—	Vss

Note 1: Input only.

### PIC16F616/16HV616 14-Pin Diagram (PDIP, SOIC, TSSOP)



### TABLE 2: PIC16F616/16HV616 14-PIN SUMMARY

I/O	Pin	Analog	Comparators	Timer	ССР	Interrupts	Pull-ups	Basic
RA0	13	AN0	C1IN+	_	—	IOC	Y	ICSPDAT
RA1	12	AN1/VREF	C12IN0-	—	—	IOC	Y	ICSPCLK
RA2	11	AN2	C1OUT	T0CKI	—	INT/IOC	Y	—
RA3 <sup>(1)</sup>	4	_	—	_	_	IOC	Y(2)	MCLR/Vpp
RA4	3	AN3	—	T1G	—	IOC	Y	OSC2/CLKOUT
RA5	2		—	T1CKI	—	IOC	Y	OSC1/CLKIN
RC0	10	AN4	C2IN+		—	_	—	—
RC1	9	AN5	C12IN1-	_	—		—	—
RC2	8	AN6	C12IN2-	_	P1D	_	—	—
RC3	7	AN7	C12IN3-	—	P1C		—	—
RC4	6	_	C2OUT	_	P1B	_	—	—
RC5	5	_	—		CCP1/P1A	_	—	—
	1	—	—		—	—	—	Vdd
	14	_	_		_	_	_	Vss

Note 1: Input only.

### PIC16F610/16HV610 16-Pin Diagram (QFN)

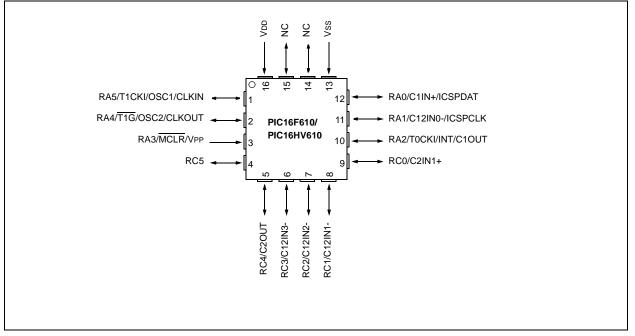


TABLE 3:	PIC16F610/16HV610	<b>16-PIN SUMMARY</b>

I/O	Pin	Comparators	Timers Interrupts		Pull-ups	Basic
RA0	12	C1IN+	_	IOC	Y	ICSPDAT
RA1	11	C12IN0-	_	IOC	Y	ICSPCLK
RA2	10	C1OUT	T0CKI	INT/IOC	Y	—
RA3 <sup>(1)</sup>	3	_	_	IOC	Y(2)	MCLR/Vpp
RA4	2	—	T1G	IOC	Y	OSC2/CLKOUT
RA5	1	—	T1CKI	IOC	Y	OSC1/CLKIN
RC0	9	C2IN+	_	—	—	—
RC1	8	C12IN1-		—	—	—
RC2	7	C12IN2-	_	—	—	—
RC3	6	C12IN3-		—	—	—
RC4	5	C2OUT	_	—	—	—
RC5	4	—	_	—	—	—
	16		_			Vdd
_	13	_			—	Vss

Note 1: Input only.

### PIC16F616/16HV616 16-Pin Diagram (QFN)

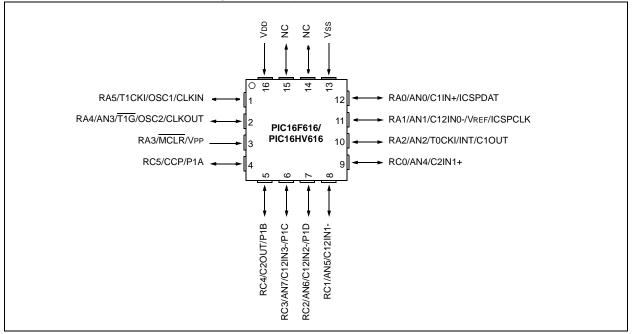


TABLE 4:	PIC16F616/16HV616	<b>16-PIN SUMMARY</b>

I/O	Pin	Analog	Comparators	Timers	ССР	Interrupts	Pull-ups	Basic
RA0	12	AN0	C1IN+	—	—	IOC	Y	ICSPDAT
RA1	11	AN1/VREF	C12IN0-	_	_	IOC	Y	ICSPCLK
RA2	10	AN2	C1OUT	T0CKI		INT/IOC	Y	—
RA3 <sup>(1)</sup>	3	_	_	_	_	IOC	Y(2)	MCLR/Vpp
RA4	2	AN3	—	T1G	—	IOC	Y	OSC2/CLKOUT
RA5	1		—	T1CKI	-	IOC	Y	OSC1/CLKIN
RC0	9	AN4	C2IN+	—	—	—	—	—
RC1	8	AN5	C12IN1-	_		—	—	—
RC2	7	AN6	C12IN2-	—	P1D	—	—	—
RC3	6	AN7	C12IN3-	—	P1C	—	—	—
RC4	5	_	C2OUT	_	P1B	—	—	—
RC5	4		—	_	CCP1/P1A	—	—	—
	16	_	—	_			—	Vdd
_	13	_		_	_	_	_	Vss

Note 1: Input only.

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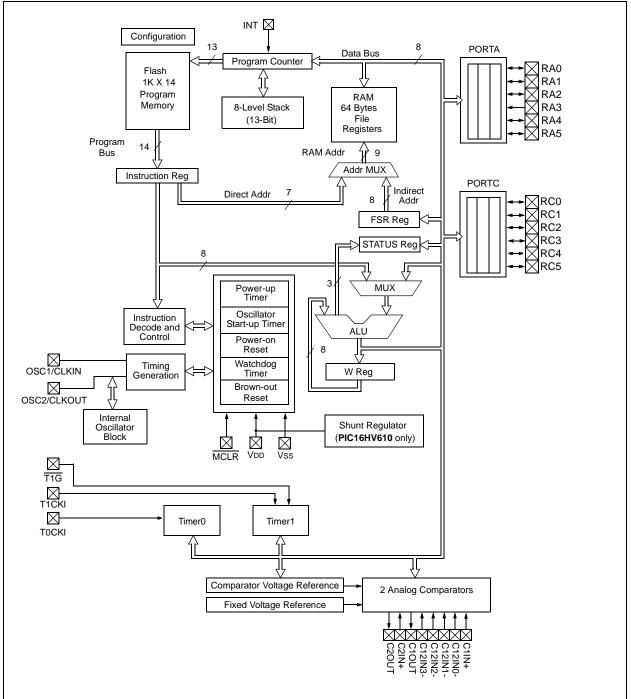
### 1.0 DEVICE OVERVIEW

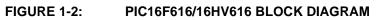
The PIC16F610/616/16HV610/616 is covered by this data sheet. It is available in 14-pin PDIP, SOIC, TSSOP and 16-pin QFN packages.

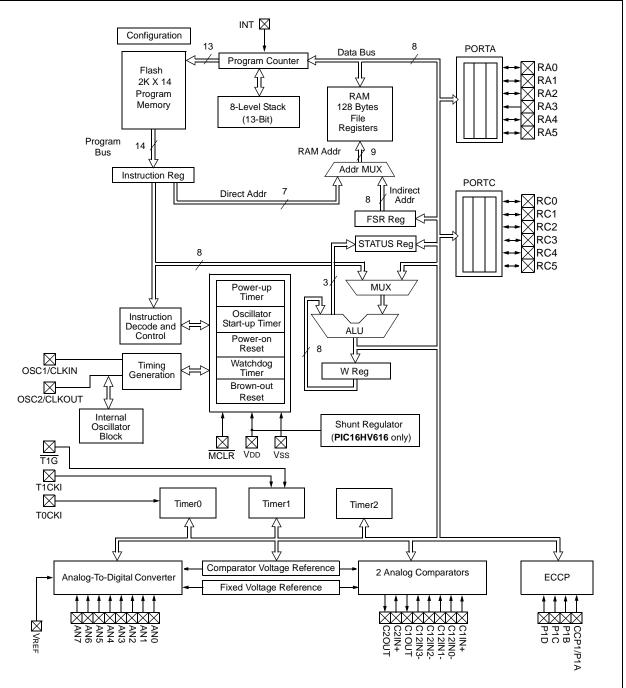
Block Diagrams and pinout descriptions of the devices are as follows:

- PIC16F610/16HV610 (Figure 1-1, Table 1-1)
- PIC16F616/16HV616 (Figure 1-2, Table 1-2)

### FIGURE 1-1: PIC16F610/16HV610 BLOCK DIAGRAM







Name	Function	Input Type	Output Type	Description
RA0/C1IN+/ICSPDAT	RA0	TTL	CMOS	PORTA I/O with prog. pull-up and interrupt-on-change
	C1IN+	AN	_	Comparator C1 non-inverting input
	ICSPDAT	ST	CMOS	Serial Programming Data I/O
RA1/C12IN0-/ICSPCLK	RA1	TTL	CMOS	PORTA I/O with prog. pull-up and interrupt-on-change
	C12IN0-	AN	_	Comparators C1 and C2 inverting input
	ICSPCLK	ST	_	Serial Programming Clock
RA2/T0CKI/INT/C1OUT	RA2	ST	CMOS	PORTA I/O with prog. pull-up and interrupt-on-change
	TOCKI	ST	_	Timer0 clock input
	INT	ST	_	External Interrupt
	C1OUT	_	CMOS	Comparator C1 output
RA3/MCLR/Vpp	RA3	TTL	_	PORTA input with interrupt-on-change
	MCLR	ST	_	Master Clear w/internal pull-up
	Vpp	HV	—	Programming voltage
RA4/T1G/OSC2/CLKOUT	RA4	TTL	CMOS	PORTA I/O with prog. pull-up and interrupt-on-change
	T1G	ST	_	Timer1 gate (count enable)
	OSC2	_	XTAL	Crystal/Resonator
	CLKOUT	_	CMOS	Fosc/4 output
RA5/T1CKI/OSC1/CLKIN	RA5	TTL	CMOS	PORTA I/O with prog. pull-up and interrupt-on-change
	T1CKI	ST	_	Timer1 clock input
	OSC1	XTAL	_	Crystal/Resonator
	CLKIN	ST	_	External clock input/RC oscillator connection
RC0/C2IN+	RC0	TTL	CMOS	PORTC I/O
	C2IN+	AN	_	Comparator C2 non-inverting input
RC1/C12IN1-	RC1	TTL	CMOS	PORTC I/O
	C12IN1-	AN	—	Comparators C1 and C2 inverting input
RC2/C12IN2-	RC2	TTL	CMOS	PORTC I/O
	C12IN2-	AN	_	Comparators C1 and C2 inverting input
RC3/C12IN3-	RC3	TTL	CMOS	PORTC I/O
	C12IN3-	AN	—	Comparators C1 and C2 inverting input
RC4/C2OUT	RC4	TTL	CMOS	PORTC I/O
	C2OUT	—	CMOS	Comparator C2 output
RC5	RC5	TTL	CMOS	PORTC I/O
Vdd	Vdd	Power	_	Positive supply
Vss	Vss	Power	_	Ground reference

TABLE 1-1:	PIC16F610/16HV610 PINOUT DESCRIPTION

Legend:

 AN = Analog input or output
 CMOS = CMOS compatible input or output
 HV = High Voltage

 ST = Schmitt Trigger input with CMOS levels
 TTL = TTL compatible input
 XTAL = Crystal

#### **TABLE 1-2:** PIC16F616/16HV616 PINOUT DESCRIPTION

Name	Function	Input Type	Output Type	Description
RA0/AN0/C1IN+/ICSPDAT	RA0	TTL	CMOS	PORTA I/O with prog. pull-up and interrupt-on-change
	AN0	AN	_	A/D Channel 0 input
	C1IN+	AN	_	Comparator C1 non-inverting input
	ICSPDAT	ST	CMOS	Serial Programming Data I/O
RA1/AN1/C12IN0-/VREF/ICSPCLK	RA1	TTL	CMOS	PORTA I/O with prog. pull-up and interrupt-on-change
	AN1	AN	_	A/D Channel 1 input
	C12IN0-	AN	_	Comparators C1 and C2 inverting input
	VREF	AN	_	External Voltage Reference for A/D
	ICSPCLK	ST	_	Serial Programming Clock
RA2/AN2/T0CKI/INT/C1OUT	RA2	ST	CMOS	PORTA I/O with prog. pull-up and interrupt-on-change
	AN2	AN	_	A/D Channel 2 input
	TOCKI	ST	—	Timer0 clock input
	INT	ST	—	External Interrupt
	C1OUT	_	CMOS	Comparator C1 output
RA3/MCLR/Vpp	RA3	TTL	_	PORTA input with interrupt-on-change
	MCLR	ST	_	Master Clear w/internal pull-up
	VPP	HV	_	Programming voltage
RA4/AN3/T1G/OSC2/CLKOUT	RA4	TTL	CMOS	PORTA I/O with prog. pull-up and interrupt-on-change
(A4/AN3/110/0302/02/02/001	AN3	AN		A/D Channel 3 input
	TIG			
		ST —		Timer1 gate (count enable)
	OSC2		XTAL	Crystal/Resonator
	CLKOUT		CMOS	Fosc/4 output
RA5/T1CKI/OSC1/CLKIN	RA5	TTL	CMOS	PORTA I/O with prog. pull-up and interrupt-on-change
	T1CKI	ST		Timer1 clock input
	OSC1	XTAL	_	Crystal/Resonator
	CLKIN	ST	-	External clock input/RC oscillator connection
RC0/AN4/C2IN+	RC0	TTL	CMOS	PORTC I/O
	AN4	AN	_	A/D Channel 4 input
	C2IN+	AN	—	Comparator C2 non-inverting input
RC1/AN5/C12IN1-	RC1	TTL	CMOS	PORTC I/O
	AN5	AN		A/D Channel 5 input
	C12IN1-	AN	—	Comparators C1 and C2 inverting input
RC2/AN6/C12IN2-/P1D	RC2	TTL	CMOS	PORTC I/O
	AN6	AN	_	A/D Channel 6 input
	C12IN2-	AN	—	Comparators C1 and C2 inverting input
	P1D	—	CMOS	PWM output
RC3/AN7/C12IN3-/P1C	RC3	TTL	CMOS	PORTC I/O
	AN7	AN	—	A/D Channel 7 input
	C12IN3-	AN	—	Comparators C1 and C2 inverting input
	P1C	—	CMOS	PWM output
RC4/C2OUT/P1B	RC4	TTL	CMOS	PORTC I/O
	C2OUT	—	CMOS	Comparator C2 output
	P1B	—	CMOS	PWM output
RC5/CCP1/P1A	RC5	TTL	CMOS	PORTC I/O
	CCP1	ST	CMOS	Capture input/Compare output
	P1A	—	CMOS	PWM output
/dd	Vdd	Power		Positive supply
Vss	Vss	Power	_	Ground reference

Legend:

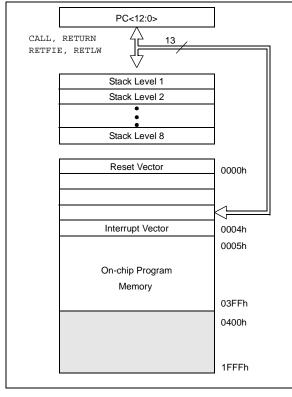
AN = Analog input or outputCMOS = CMOS compatible input or outputHV = High VoltageST = Schmitt Trigger input with CMOS levelsTTL = TTL compatible inputXTAL = Crystal

### 2.0 MEMORY ORGANIZATION

### 2.1 Program Memory Organization

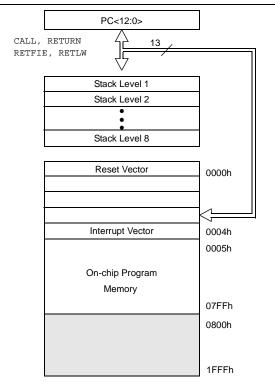
The PIC16F610/616/16HV610/616 has a 13-bit program counter capable of addressing an 8K x 14 program memory space. Only the first 1K x 14 (0000h-3FF) for the PIC16F610/16HV610 and the first 2K x 14 (0000h-07FFh) for the PIC16F616/16HV616 is physically implemented. Accessing a location above these boundaries will cause a wraparound within the first 1K x 14 space (PIC16F610/16HV610) and 2K x 14 space (PIC16F616/16HV616). The Reset vector is at 0000h and the interrupt vector is at 0004h (see Figure 2-1).

### FIGURE 2-1: PROGRAM MEMORY MAP AND STACK FOR THE PIC16F610/16HV610



### FIGURE 2-2:

#### PROGRAM MEMORY MAP AND STACK FOR THE PIC16F616/16HV616



### 2.2 Data Memory Organization

The data memory (see Figure 2-4) is partitioned into two banks, which contain the General Purpose Registers (GPR) and the Special Function Registers (SFR). The Special Function Registers are located in the first 32 locations of each bank. PIC16F610/16HV610 Register locations 40h-7Fh in Bank 0 are General Purpose Registers, implemented as static RAM. PIC16F616/16HV616 Register locations 20h-7Fh in Bank 0 and A0h-BFh in Bank 1 are General Purpose Registers, implemented as static RAM. Register locations F0h-FFh in Bank 1 point to addresses 70h-7Fh in Bank 0. All other RAM is unimplemented and returns '0' when read. The RP0 bit of the STATUS register is the bank select bit.

<u>RP0</u>

- $0 \rightarrow \text{Bank 0 is selected}$
- $1 \rightarrow \text{Bank 1 is selected}$

Note:	The IRP and RP1 bits of the STATUS			
	register are reserved and should always be			
	maintained as '0's.			

### 2.2.1 GENERAL PURPOSE REGISTER FILE

The register file is organized as  $64 \times 8$  in the PIC16F610/16HV610 and  $128 \times 8$  in the PIC16F616/16HV616. Each register is accessed, either directly or indirectly, through the File Select Register (FSR) (see Section 2.4 "Indirect Addressing, INDF and FSR Registers").

### 2.2.2 SPECIAL FUNCTION REGISTERS

The Special Function Registers are registers used by the CPU and peripheral functions for controlling the desired operation of the device (see Table 2-1). These registers are static RAM.

The special registers can be classified into two sets: core and peripheral. The Special Function Registers associated with the "core" are described in this section. Those related to the operation of the peripheral features are described in the section of that peripheral feature.

### FIGURE 2-3: DATA MEMORY MAP OF THE PIC16F610/16HV610

	File Address		File Address			
Indirect Addr.(1)	00h	Indirect Addr. <sup>(1)</sup>	80h			
TMR0	01h	OPTION REG	81h			
PCL	02h	PCL	82h			
STATUS	03h	STATUS	83h			
FSR	03h 04h	FSR	84h			
PORTA	05h	TRISA	85h			
1 OKIN	06h	INIOA	86h			
PORTC	07h	TRISC	87h			
Tokto	08h	11100	88h			
	09h		89h			
PCLATH	0Ah	PCLATH	8Ah			
INTCON		INTCON				
PIR1	0Bh	PIE1	8Bh			
FIRI	0Ch	FIEI	8Ch			
TMD4	0Dh	DOON	8Dh			
TMR1L	0Eh	PCON	8Eh			
TMR1H	0Fh	000711115	8Fh			
T1CON	10h	OSCTUNE	90h			
	11h	ANSEL	91h			
	12h		92h			
	13h		93h			
	14h		94h			
	15h	WPUA	95h			
	16h	IOCA	96h			
	17h		97h			
	18h		98h			
VRCON	19h	SRCON0	99h			
CM1CON0	1Ah	SRCON1	9Ah			
CM2CON0	1Bh		9Bh			
CM2CON1	1Ch		9Ch			
	1Dh		9Dh			
	1Eh		9Eh			
	1Fh		9Fh			
	20h		A0h			
	3Fh					
	40h					
General Purpose Registers						
64 Bytes	6Fh					
Accesses 70h-7Fh	70h 7Fh	Accesses 70h-7Fh	F0h FFh			
Bank 0		Bank 1				
Unimplemented data memory locations, read as '0'. <b>Note 1:</b> Not a physical register.						

### FIGURE 2-4:

### DATA MEMORY MAP OF THE PIC16F616/16HV616

	File Address	Ą	File ddres:			
Indirect Addr.(1)	00h	Indirect Addr.(1)	80h			
TMR0	01h	OPTION_REG	81h			
PCL	02h	PCL	82h			
STATUS	03h	STATUS	83h			
FSR	04h	FSR	84h			
PORTA	05h	TRISA	85h			
	06h		86h			
PORTC	07h	TRISC	87h			
	08h		88h			
	09h		89h			
PCLATH	0Ah	PCLATH	8Ah			
INTCON	0Bh	INTCON	8Bh			
PIR1	0Ch	PIE1	8Ch			
	0Dh		8Dh			
TMR1L	0Eh	PCON	8Eh			
TMR1H	0Eh	1001	8Fh			
T1CON	10h	OSCTUNE	90h			
TMR2		ANSEL	901 91h			
T2CON	11h 12h	PR2	91h 92h			
CCPR1L		F INZ				
CCPR1H	13h		93h			
	14h	\A/DLIA	94h			
CCP1CON PWM1CON	15h	IOCA	95h			
	16h	IUCA	96h			
ECCPAS	17h		97h			
VECON	18h		98h			
VRCON	19h	SRCON0	99h			
CM1CON0	1Ah	SRCON1	9Ah			
CM2CON0	1Bh		9Bh			
CM2CON1	1Ch		9Ch			
	1Dh	10050	9Dh			
ADRESH	1Eh	ADRESL	9Eh			
ADCON0	1Fh	ADCON1 General	9Fh			
	20h	Purpose	A0h			
		Registers				
General		32 Bytes	BFh			
Purpose			C0h			
Registers						
96 Bytes						
	7Fh	Accesses 70h-7Fh	F0h FFh			
Bank 0	_ // //	Bank 1				
Unimplemented data memory locations, read as '0'. Note 1: Not a physical register.						

Addr	Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Value on POR, BOR	Page
Bank 0											
00h	INDF	Addressing	this location	uses conten	ts of FSR to a	address data	memory (no	t a physical r	egister)	xxxx xxxx	24, 116
01h	TMR0	Timer0 Mod	dule's Registe	er						xxxx xxxx	45, 116
02h	PCL	Program Co	punter's (PC)	Least Signif	icant Byte					0000 0000	24, 116
03h	STATUS	IRP <sup>(1)</sup>	RP1 <sup>(1)</sup>	RP0	TO	PD	Z	DC	С	0001 1xxx	18, 116
04h	FSR	Indirect Dat	a Memory Ad	ddress Pointe	er					xxxx xxxx	24, 116
05h	PORTA	—	—	RA5	RA4	RA3	RA2	RA1	RA0	x0 x000	33, 116
06h	—	Unimpleme	nted							_	_
07h	PORTC	_	_	RC5	RC4	RC3	RC2	RC1	RC0	xx 00xx	42, 116
08h	—	Unimpleme	nted							_	_
09h	—	Unimpleme	nted							_	_
0Ah	PCLATH	—	—	_	Write	Buffer for up	oper 5 bits of	Program Co	unter	0 0000	24, 116
0Bh	INTCON	GIE	PEIE	T0IE	INTE	RAIE	T0IF	INTF	RAIF	0000 0000	20, 116
0Ch	PIR1	_	ADIF <sup>(2)</sup>	CCP1IF <sup>(2)</sup>	C2IF	C1IF	_	TMR2IF <sup>(2)</sup>	TMR1IF	-000 0-00	22, 116
0Dh	_	Unimpleme	Unimplemented								_
0Eh	TMR1L	Holding Re	Holding Register for the Least Significant Byte of the 16-bit TMR1 Register								49, 116
0Fh	TMR1H	Holding Re	gister for the	Most Signific	ant Byte of t	ne 16-bit TMI	R1 Register			xxxx xxxx	49, 116
10h	T1CON	T1GINV	TMR1GE	T1CKPS1	T1CKPS0	T1OSCEN	T1SYNC	TMR1CS	TMR10N	0000 0000	52, 116
11h	TMR2 <sup>(2)</sup>	Timer2 Mod	dule Register							0000 0000	55, 116
12h	T2CON <sup>(2)</sup>	—	TOUTPS3	TOUTPS2	TOUTPS1	TOUTPS0	TMR2ON	T2CKPS1	T2CKPS0	-000 0000	56, 116
13h	CCPR1L <sup>(2)</sup>	Capture/Co	mpare/PWM	Register 1 L	ow Byte					xxxx xxxx	86, 116
14h	CCPR1H <sup>(2)</sup>	Capture/Co	mpare/PWM	Register 1 H	ligh Byte					XXXX XXXX	86, 116
15h	CCP1CON <sup>(2)</sup>	P1M1	P1M0	DC1B1	DC1B0	CCP1M3	CCP1M2	CCP1M1	CCP1M0	0000 0000	85, 116
16h	PWM1CON <sup>(2)</sup>	PRSEN	PDC6	PDC5	PDC4	PDC3	PDC2	PDC1	PDC0	0000 0000	85, 116
17h	ECCPAS <sup>(2)</sup>	ECCPASE	ECCPAS2	ECCPAS1	ECCPAS0	PSSAC1	PSSAC0	PSSBD1	PSSBD0	0000 0000	102, 116
18h	—	Unimpleme	nted							_	_
19h	VRCON	C1VREN	C2VREN	VRR	FVREN	VR3	VR2	VR1	VR0	0000 0000	72, 116
1Ah	CM1CON0	C1ON	C1OUT	C1OE	C1POL	_	C1R	C1CH1	C1CH0	0000 -000	62, 116
1Bh	CM2CON0	C2ON	C2OUT	C2OE	C2POL	_	C2R	C2CH1	C2CH0	0000 -000	63, 116
1Ch	CM2CON1	MC10UT	MC2OUT	_	T1ACS	C1HYS	C2HYS	T1GSS	C2SYNC	00-0 0010	65, 116
1Dh	—	Unimpleme	nted							—	—
1Eh	ADRESH <sup>(2,3)</sup>	Most Signifi	icant 8 bits of	the left shift	ed A/D result	or 2 bits of r	ight shifted re	esult		xxxx xxxx	80, 116
1Fh	ADCON0(2)	ADFM	VCFG	CHS3	CHS2	CHS1	CHS0	GO/DONE	ADON	0000 0000	78, 116

#### **TABLE 2-1:** PIC16F610/616/16HV610/616 SPECIAL FUNCTION REGISTERS SUMMARY BANK 0

- = Unimplemented locations read as '0', u = unchanged, x = unknown, q = value depends on condition, shaded = unimplemented IRP and RP1 bits are reserved, always maintain these bits clear. Legend:

Note 1:

2: PIC16F616/16HV616 only.

3: Read-only register.

Addr	Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Value on POR, BOR	Page
Bank 1											
80h	INDF	Addressing	this location	uses content	s of FSR to	address data	a memory (no	ot a physical	register)	xxxx xxxx	24, 116
81h	OPTION_REG	RAPU	INTEDG	TOCS	T0SE	PSA	PS2	PS1	PS0	1111 1111	19, 116
82h	PCL	•		Least Signifi	cant Byte					0000 0000	24, 116
83h	STATUS	IRP <sup>(1)</sup>	RP1 <sup>(1)</sup>	RP0	TO	PD	Z	DC	С	0001 1xxx	18, 116
84h	FSR	Indirect Dat	a Memory Ad	ddress Pointe	er					xxxx xxxx	24, 116
85h	TRISA	-	-	TRISA5	TRISA4	TRISA3	TRISA2	TRISA1	TRISA0	11 1111	33, 116
86h	_	Unimpleme	nted							—	—
87h	TRISC			TRISC5	TRISC4	TRISC3	TRISC2	TRISC1	TRISC0	11 1111	42, 116
88h	_	Unimpleme	nted							_	—
89h	_	Unimpleme	nted							_	—
8Ah	PCLATH			_	Write	e Buffer for u	pper 5 bits of	f Program Co	ounter	0 0000	24, 116
8Bh	INTCON	GIE	PEIE	T0IE	INTE	RAIE	T0IF	INTF	RAIF	0000 0000	20, 116
8Ch	PIE1	_	ADIE <sup>(3)</sup>	CCP1IE <sup>(3)</sup>	C2IE	C1IE	_	TMR2IE <sup>(3)</sup>	TMR1IE	-000 0-00	21, 116
8Dh		Unimpleme	nted							_	_
8Eh	PCON	_	_	_	_	—	—	POR	BOR	dd	23, 116
8Fh		Unimpleme	nted							_	_
90h	OSCTUNE	_	_	_	TUN4	TUN3	TUN2	TUN1	TUN0	0 0000	31, 117
91h	ANSEL	ANS7	ANS6	ANS5	ANS4	ANS3 <sup>(3)</sup>	ANS2 <sup>(3)</sup>	ANS1	ANS0	1111 1111	34, 117
92h	PR2 <sup>(3)</sup>	Timer2 Mod	lule Period R	egister						1111 1111	55, 117
93h		Unimpleme	nted							_	_
94h		Unimpleme	nted							_	_
95h	WPUA	_	_	WPUA5	WPUA4	_	WPUA2	WPUA1	WPUA0	11 -111	35, 117
96h	IOCA	_	_	IOCA5	IOCA4	IOCA3	IOCA2	IOCA1	IOCA0	00 0000	35, 117
97h		Unimpleme	nted							_	_
98h		Unimpleme	nted							_	_
99h	SRCON0	SR1	SR0	C1SEN	C2REN	PULSS	PULSR	_	SRCLKEN	0000 00-0	69, 117
9Ah	SRCON1	SRCS1	SRCS0	_	_	_	_	_	—	00	69, 117
9Bh	—	Unimpleme	nted							—	
9Ch	_	Unimpleme	nted							—	
9Dh	_	Unimpleme	nted							—	—
9Eh	ADRESL <sup>(3,4)</sup>	Least Signif	icant 2 bits o	f the left shift	ed result or	8 bits of the	right shifted i	result		xxxx xxxx	80, 117
9Fh	ADCON1 <sup>(3)</sup>	_	ADCS2	ADCS1	ADCS0	—	_	_	—	-000	79, 117

#### **TABLE 2-2:** PIC16F610/616/16HV610/616 SPECIAL FUNCTION REGISTERS SUMMARY BANK 1

 – = Unimplemented locations read as '0', u = unchanged, x = unknown, q = value depends on condition, shaded = unimplemented IRP and RP1 bits are reserved, always maintain these bits clear.
 RA3 pull-up is enabled when MCLRE is '1' in the Configuration Word register. Legend:

Note 1:

2:

PIC16F616/16HV616 only. 3:

4: Read-only Register.

### 2.2.2.1 STATUS Register

The STATUS register, shown in Register 2-1, contains:

- the arithmetic status of the ALU
- · the Reset status

-n = Value at POR

• the bank select bits for data memory (RAM)

The STATUS register can be the destination for any instruction, like any other register. If the STATUS register is the destination for an instruction that affects the Z, DC or C bits, then the write to these three bits is disabled. These bits are set or cleared according to the device logic. Furthermore, the TO and PD bits are not writable. Therefore, the result of an instruction with the STATUS register as destination may be different than intended.

For example, CLRF STATUS, will clear the upper three bits and set the Z bit. This leaves the STATUS register as `000u uluu' (where u = unchanged).

It is recommended, therefore, that only BCF, BSF, SWAPF and MOVWF instructions are used to alter the STATUS register, because these instructions do not affect any Status bits. For other instructions not affecting any Status bits, see the **Section 13.0 "Instruction Set Summary"**.

- Note 1: Bits IRP and RP1 of the STATUS register are not used by the PIC16F610/616/16HV610/616 and should be maintained as clear. Use of these bits is not recommended, since this may affect upward compatibility with future products.
  - 2: The <u>C</u> and <u>DC</u> bits operate as a Borrow and Digit Borrow out bit, respectively, in subtraction. See the SUBLW and SUBWF instructions for examples.

x = Bit is unknown

### REGISTER 2-1: STATUS: STATUS REGISTER

'1' = Bit is set

Reserved	Reserved	R/W-0	R-1	R-1	R/W-x	R/W-x	R/W-x	
IRP	RP1	RP0	то	PD	Z	DC	С	
bit 7	•	•	•				bit 0	
Legend:	Legend:							
R = Readable bit W = Writable bit U = Unimplemented bit, read as '0'								

'0' = Bit is cleared

bit 7	IRP: This bit is reserved and should be maintained as '0'
bit 6	<b>RP1:</b> This bit is reserved and should be maintained as '0'
bit 5	<b>RP0:</b> Register Bank Select bit (used for direct addressing) 1 = Bank 1 (80h – FFh) 0 = Bank 0 (00h – 7Fh)
bit 4	TO: Time-out bit 1 = After power-up, CLRWDT instruction or SLEEP instruction 0 = A WDT time-out occurred
bit 3	<b>PD</b> : Power-down bit 1 = After power-up or by the CLRWDT instruction 0 = By execution of the SLEEP instruction
bit 2	<ul> <li>Z: Zero bit</li> <li>1 = The result of an arithmetic or logic operation is zero</li> <li>0 = The result of an arithmetic or logic operation is not zero</li> </ul>
bit 1	<b>DC:</b> Digit Carry/Borrow bit (ADDWF, ADDLW, SUBLW, SUBWF instructions), For Borrow, the polarity is reversed. 1 = A carry-out from the 4th low-order bit of the result occurred 0 = No carry-out from the 4th low-order bit of the result
bit 0	<b>C:</b> Carry/Borrow bit <sup>(1)</sup> (ADDWF, ADDLW, SUBLW, SUBWF instructions) 1 = A carry-out from the Most Significant bit of the result occurred 0 = No carry-out from the Most Significant bit of the result occurred
Note 1:	For Borrow, the polarity is reversed. A subtraction is executed by adding the two's complement of the second operand. For rotate (RRF, RLF) instructions, this bit is loaded with either the high-order or low-order bit of the source register.

### 2.2.2.2 OPTION Register

The OPTION register is a readable and writable register, which contains various control bits to configure:

- Timer0/WDT prescaler
- External RA2/INT interrupt
- Timer0
- Weak pull-ups on PORTA

Note: To achieve a 1:1 prescaler assignment for Timer0, assign the prescaler to the WDT by setting PSA bit to '1' of the OPTION register. See Section 5.1.3 "Software Programmable Prescaler".

### REGISTER 2-2: OPTION\_REG: OPTION REGISTER

R/W-1	R/W-1	R/W-1	R/W-1	R/W-1	R/W-1	R/W-1	R/W-1
RAPU	INTEDG	T0CS	TOSE	PSA	PS2	PS1	PS0
bit 7							bit 0

Legend:				
R = Readable bit	= Readable bit W = Writable bit U = Unimplemented bit, read as '0'			
-n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown	

bit 7	<b>RAPU:</b> PORTA Pull-up Enable bit 1 = PORTA pull-ups are disabled 0 = PORTA pull-ups are enabled by individual PORT latch values
bit 6	INTEDG: Interrupt Edge Select bit
	<ul><li>1 = Interrupt on rising edge of RA2/INT pin</li><li>0 = Interrupt on falling edge of RA2/INT pin</li></ul>
bit 5	TOCS: Timer0 Clock Source Select bit
	1 = Transition on RA2/T0CKI pin 0 = Internal instruction cycle clock (Fosc/4)
bit 4	T0SE: Timer0 Source Edge Select bit
	<ul><li>1 = Increment on high-to-low transition on RA2/T0CKI pin</li><li>0 = Increment on low-to-high transition on RA2/T0CKI pin</li></ul>
bit 3	PSA: Prescaler Assignment bit
	<ul><li>1 = Prescaler is assigned to the WDT</li><li>0 = Prescaler is assigned to the Timer0 module</li></ul>
bit 2-0	PS<2:0>: Prescaler Rate Select bits
	BIT VALUE TIMER0 RATE WDT RATE

000	1:2	1:1
001	1:4	1:2
010	1:8	1:4
011	1:16	1:8
100	1:32	1:16
101	1:64	1:32
110	1 : 128	1:64
111	1 : 256	1 : 128

### 2.2.2.3 INTCON Register

The INTCON register is a readable and writable register, which contains the various enable and flag bits for TMR0 register overflow, PORTA change and external RA2/INT pin interrupts.

Note: Interrupt flag bits are set when an interrupt condition occurs, regardless of the state of its corresponding enable bit or the global enable bit, GIE of the INTCON register. User software should ensure the appropriate interrupt flag bits are clear prior to enabling an interrupt.

### REGISTER 2-3: INTCON: INTERRUPT CONTROL REGISTER

| R/W-0 |
|-------|-------|-------|-------|-------|-------|-------|-------|
| GIE   | PEIE  | TOIE  | INTE  | RAIE  | T0IF  | INTF  | RAIF  |
| bit 7 |       |       |       |       |       |       | bit 0 |

Legend:				
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'		
-n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown	

bit 7	<b>GIE:</b> Global Interrupt Enable bit 1 = Enables all unmasked interrupts 0 = Disables all interrupts
bit 6	<b>PEIE:</b> Peripheral Interrupt Enable bit 1 = Enables all unmasked peripheral interrupts 0 = Disables all peripheral interrupts
bit 5	<b>T0IE:</b> Timer0 Overflow Interrupt Enable bit 1 = Enables the Timer0 interrupt 0 = Disables the Timer0 interrupt
bit 4	INTE: RA2/INT External Interrupt Enable bit 1 = Enables the RA2/INT external interrupt 0 = Disables the RA2/INT external interrupt
bit 3	<b>RAIE:</b> PORTA Change Interrupt Enable bit <sup>(1)</sup> 1 = Enables the PORTA change interrupt 0 = Disables the PORTA change interrupt
bit 2	<b>T0IF:</b> Timer0 Overflow Interrupt Flag bit <sup>(2)</sup> 1 = Timer0 register has overflowed (must be cleared in software) 0 = Timer0 register did not overflow
bit 1	INTF: RA2/INT External Interrupt Flag bit 1 = The RA2/INT external interrupt occurred (must be cleared in software) 0 = The RA2/INT external interrupt did not occur
bit 0	<b>RAIF:</b> PORTA Change Interrupt Flag bit 1 = When at least one of the PORTA <5:0> pins changed state (must be cleared in software) 0 = None of the PORTA <5:0> pins have changed state

Note 1: IOCA register must also be enabled.

2: T0IF bit is set when TMR0 rolls over. TMR0 is unchanged on Reset and should be initialized before clearing T0IF bit.

### 2.2.2.4 PIE1 Register

The PIE1 register contains the peripheral interrupt enable bits, as shown in Register 2-4.

Note: Bit PEIE of the INTCON register must be set to enable any peripheral interrupt.

### REGISTER 2-4: PIE1: PERIPHERAL INTERRUPT ENABLE REGISTER 1

U-0	R/W-0	R/W-0	R/W-0	R/W-0	U-0	R/W-0	R/W-0
—	ADIE <sup>(1)</sup>	CCP1IE <sup>(1)</sup>	C2IE	C1IE	—	TMR2IE <sup>(1)</sup>	TMR1IE
bit 7							bit 0

Legend:				
R = Read	able bit	W = Writable bit	U = Unimplemented bit,	read as '0'
-n = Value	e at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown
L:4 7				
bit 7	•	mented: Read as '0'	(1)	
bit 6		D Converter (ADC) Interrupt	Enable bit <sup>(1)</sup>	
		les the ADC interrupt bles the ADC interrupt		
bit 5	CCP1IE:	CCP1 Interrupt Enable bit <sup>(1)</sup>	1	
		les the CCP1 interrupt bles the CCP1 interrupt		
bit 4	<b>C2IE:</b> Co	mparator C2 Interrupt Enable	e bit	
	1 = Enab	les the Comparator C2 interr	rupt	
	0 = Disat	oles the Comparator C2 inter	rupt	
bit 3	<b>C1IE:</b> Co	mparator C1 Interrupt Enabl	e bit	
	1 = Enab	les the Comparator C1 interr	rupt	
	0 = Disat	oles the Comparator C1 inter	rupt	
bit 2	Unimple	mented: Read as '0'		
bit 1	TMR2IE:	Timer2 to PR2 Match Interru	upt Enable bit <sup>(1)</sup>	
	1 = Enab	les the Timer2 to PR2 match	n interrupt	
	0 = Disat	oles the Timer2 to PR2 matcl	h interrupt	
bit 0	TMR1IE:	Timer1 Overflow Interrupt E	nable bit	
	1 = Enab	les the Timer1 overflow inter	rupt	
	0 = Disat	oles the Timer1 overflow inte	rrupt	
Note 1:	PIC16F616/16	6HV616 only. PIC16F610/16	HV610 unimplemented, read a	as '0'.

### 2.2.2.5 PIR1 Register

The PIR1 register contains the peripheral interrupt flag bits, as shown in Register 2-5.

Note: Interrupt flag bits are set when an interrupt condition occurs, regardless of the state of its corresponding enable bit or the global enable bit, GIE of the INTCON register. User software should ensure the appropriate interrupt flag bits are clear prior to enabling an interrupt.

### REGISTER 2-5: PIR1: PERIPHERAL INTERRUPT REQUEST REGISTER 1

U-0	R/W-0	R/W-0	R/W-0	R/W-0	U-0	R/W-0	R/W-0
_	ADIF <sup>(1)</sup>	CCP1IF <sup>(1)</sup>	C2IF	C1IF	—	TMR2IF <sup>(1)</sup>	TMR1IF
bit 7							bit 0

Legend:				
R = Readable bit W = Writable bit		U = Unimplemented bit, read as '0'		
-n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown	

bit 7	Unimplemented: Read as '0'
bit 6	ADIF: A/D Interrupt Flag bit <sup>(1)</sup>
	1 = A/D conversion complete
	0 = A/D conversion has not completed or has not been started
bit 5	CCP1IF: CCP1 Interrupt Flag bit <sup>(1)</sup>
	Capture mode:
	1 = A TMR1 register capture occurred (must be cleared in software)
	0 = No TMR1 register capture occurred Compare mode:
	1 = A TMR1 register compare match occurred (must be cleared in software)
	0 = No TMR1 register compare match occurred
	<u>PWM mode</u> :
	Unused in this mode
bit 4	C2IF: Comparator C2 Interrupt Flag bit
	1 = Comparator C2 output has changed (must be cleared in software)
	0 = Comparator C2 output has not changed
bit 3	C1IF: Comparator C1 Interrupt Flag bit
	<ul> <li>1 = Comparator C1 output has changed (must be cleared in software)</li> <li>0 = Comparator C1 output has not changed</li> </ul>
<b>h</b> it 0	· · · · ·
bit 2	Unimplemented: Read as '0'
bit 1	TMR2IF: Timer2 to PR2 Match Interrupt Flag bit <sup>(1)</sup>
	<ul> <li>1 = Timer2 to PR2 match occurred (must be cleared in software)</li> <li>0 = Timer2 to PR2 match has not occurred</li> </ul>
bit 0	TMR1IF: Timer1 Overflow Interrupt Flag bit
	1 = Timer1 register overflowed (must be cleared in software)
	0 = Timer1 has not overflowed
Note 1:	PIC16F616/16HV616 only. PIC16F610/16HV610 unimplemented, read as '0'.

### 2.2.2.6 PCON Register

The Power Control (PCON) register (see Table 12-2) contains flag bits to differentiate between a:

- Power-on Reset (POR)
- Brown-out Reset (BOR)
- Watchdog Timer Reset (WDT)
- External MCLR Reset

The PCON register also controls the software enable of the  $\overline{\text{BOR}}.$ 

The PCON register bits are shown in Register 2-6.

### REGISTER 2-6: PCON: POWER CONTROL REGISTER

U-0	U-0	U-0	U-0	U-0	U-0	R/W-0	R/W-0 <sup>(1)</sup>	
—	—	—	—	—	—	POR	BOR	
bit 7	bit 7 bit 0							

Legend:							
R = Reada	ole bit	W = Writable bit	U = Unimplemented bit,	read as '0'			
-n = Value a	at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown			
bit 7-2	Unimplem	nented: Read as '0'					
bit 1	POR: Pow	er-on Reset Status bit					
	1 = No Power-on Reset occurred						
	0 = A Power-on Reset occurred (must be set in software after a Power-on Reset occurs)						
bit 0	BOR: Brow	wn-out Reset Status bit					
	1 = No Bro	own-out Reset occurred					

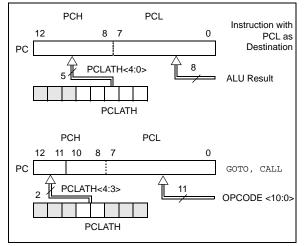
0 = A Brown-out Reset occurred (must be set in software after a Brown-out Reset occurs)

**Note 1:** Reads as '0' if Brown-out Reset is disabled.

### 2.3 PCL and PCLATH

The Program Counter (PC) is 13 bits wide. The low byte comes from the PCL register, which is a readable and writable register. The high byte (PC<12:8>) is not directly readable or writable and comes from PCLATH. On any Reset, the PC is cleared. Figure 2-5 shows the two situations for the loading of the PC. The upper example in Figure 2-5 shows how the PC is loaded on a write to PCL (PCLATH<4:0>  $\rightarrow$  PCH). The lower example in Figure 2-5 shows how the PC is loaded during a CALL or GOTO instruction (PCLATH<4:3>  $\rightarrow$  PCH).

#### FIGURE 2-5: LOADING OF PC IN DIFFERENT SITUATIONS



### 2.3.1 MODIFYING PCL

Executing any instruction with the PCL register as the destination simultaneously causes the Program Counter PC<12:8> bits (PCH) to be replaced by the contents of the PCLATH register. This allows the entire contents of the program counter to be changed by writing the desired upper 5 bits to the PCLATH register. When the lower 8 bits are written to the PCL register, all 13 bits of the program counter will change to the values contained in the PCLATH register.

A computed GOTO is accomplished by adding an offset to the program counter (ADDWF PCL). Care should be exercised when jumping into a look-up table or program branch table (computed GOTO) by modifying the PCL register. Assuming that PCLATH is set to the table start address, if the table length is greater than 255 instructions or if the lower 8 bits of the memory address rolls over from 0xFF to 0x00 in the middle of the table, then PCLATH must be incremented for each address rollover that occurs between the table beginning and the target location within the table.

For more information refer to Application Note AN556, "Implementing a Table Read" (DS00556).

### 2.3.2 STACK

The PIC16F610/616/16HV610/616 Family has an 8-level x 13-bit wide hardware stack (see Figure 2-1). The stack space is not part of either program or data space and the Stack Pointer is not readable or writable. The PC is PUSHed onto the stack when a CALL instruction is executed or an interrupt causes a branch. The stack is POPed in the event of a RETURN, RETLW or a RETFIE instruction execution. PCLATH is not affected by a PUSH or POP operation.

The stack operates as a circular buffer. This means that after the stack has been PUSHed eight times, the ninth push overwrites the value that was stored from the first push. The tenth push overwrites the second push (and so on).

- Note 1: There are no Status bits to indicate stack overflow or stack underflow conditions.
  - 2: There are no instructions/mnemonics called PUSH or POP. These are actions that occur from the execution of the CALL, RETURN, RETLW and RETFIE instructions or the vectoring to an interrupt address.

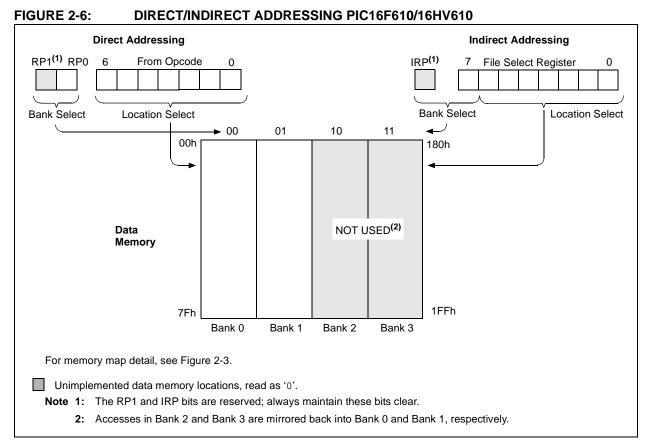
### 2.4 Indirect Addressing, INDF and FSR Registers

The INDF register is not a physical register. Addressing the INDF register will cause indirect addressing.

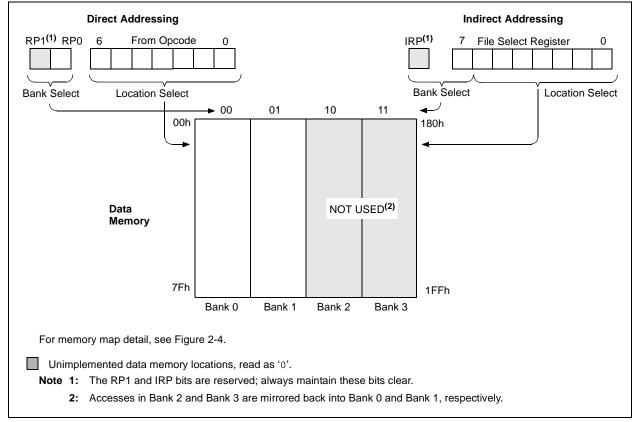
Indirect addressing is possible by using the INDF register. Any instruction using the INDF register actually accesses data pointed to by the File Select Register (FSR). Reading INDF itself indirectly will produce 00h. Writing to the INDF register indirectly results in a no operation (although Status bits may be affected). An effective 9-bit address is obtained by concatenating the 8-bit FSR and the IRP bit of the STATUS register, as shown in Figure 2-7.

A simple program to clear RAM location 40h-4Fh using indirect addressing is shown in Example 2-1.

	MOVLW	0x40	;initialize pointer
	MOVWF	FSR	;to RAM
NEXT	CLRF	INDF	;clear INDF register
	INCF	FSR, F	;inc pointer
	BTFSS	FSR,4	;all done?
	GOTO	NEXT	;no clear next
CONTI	CONTINUE		;yes continue
1			







NOTES:

### 3.0 OSCILLATOR MODULE

### 3.1 Overview

The Oscillator module has a wide variety of clock sources and selection features that allow it to be used in a wide range of applications while maximizing performance and minimizing power consumption. Figure 3-1 illustrates a block diagram of the Oscillator module.

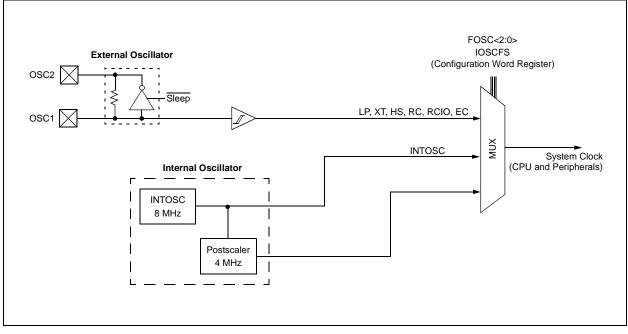
Clock sources can be configured from external oscillators, quartz crystal resonators, ceramic resonators and Resistor-Capacitor (RC) circuits. In addition, the system clock source can be configured with a choice of two selectable speeds: internal or external system clock source.

The Oscillator module can be configured in one of eight clock modes.

- 1. EC External clock with I/O on OSC2/CLKOUT.
- 2. LP 32 kHz Low-Power Crystal mode.
- 3. XT Medium Gain Crystal or Ceramic Resonator Oscillator mode.
- 4. HS High Gain Crystal or Ceramic Resonator mode.
- 5. RC External Resistor-Capacitor (RC) with Fosc/4 output on OSC2/CLKOUT.
- 6. RCIO External Resistor-Capacitor (RC) with I/O on OSC2/CLKOUT.
- 7. INTOSC Internal oscillator with Fosc/4 output on OSC2 and I/O on OSC1/CLKIN.
- 8. INTOSCIO Internal oscillator with I/O on OSC1/CLKIN and OSC2/CLKOUT.

Clock Source modes are configured by the FOSC<2:0> bits in the Configuration Word register (CONFIG). The Internal Oscillator module provides a selectable system clock mode of either 4 MHz (Postscaler) or 8 MHz (INTOSC).





### 3.2 Clock Source Modes

Clock Source modes can be classified as external or internal.

- External Clock modes rely on external circuitry for the clock source. Examples are: Oscillator modules (EC mode), quartz crystal resonators or ceramic resonators (LP, XT and HS modes) and Resistor-Capacitor (RC) mode circuits.
- Internal clock sources are contained internally within the Oscillator module. The Oscillator module has two selectable clock frequencies: 4 MHz and 8 MHz

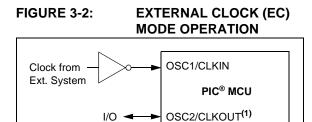
The system clock can be selected between external or internal clock sources via the FOSC<2:0> bits of the Configuration Word register.

### 3.3 External Clock Modes

### 3.3.1 EC MODE

The External Clock (EC) mode allows an externally generated logic level as the system clock source. When operating in this mode, an external clock source is connected to the OSC1 input and the OSC2 is available for general purpose I/O. Figure 3-2 shows the pin connections for EC mode.

The Oscillator Start-up Timer (OST) is disabled when EC mode is selected. Therefore, there is no delay in operation after a Power-on Reset (POR) or wake-up from Sleep. Because the PIC<sup>®</sup> MCU design is fully static, stopping the external clock input will have the effect of halting the device while leaving all data intact. Upon restarting the external clock, the device will resume operation as if no time had elapsed.



Note 1: Alternate pin functions are listed in the Section 1.0 "Device Overview".

### TABLE 3-1: OSCILLATOR DELAY EXAMPLES

Switch From Switch To		Frequency	Oscillator Delay
Sleep/POR	INTOSC	4 MHz to 8 MHz	Oscillator Warm-Up Delay (TWARM)
Sleep/POR	EC, RC	DC – 20 MHz	2 Instruction Cycles
Sleep/POR	LP, XT, HS	32 kHz to 20 MHz	1024 Clock Cycles (OST)

### 3.3.2 OSCILLATOR START-UP TIMER (OST)

If the Oscillator module is configured for LP, XT or HS modes, the Oscillator Start-up Timer (OST) counts 1024 oscillations from OSC1. This occurs following a Power-on Reset (POR) and when the Power-up Timer (PWRT) has expired (if configured), or a wake-up from Sleep. During this time, the program counter does not increment and program execution is suspended. The OST ensures that the oscillator circuit, using a quartz crystal resonator or ceramic resonator, has started and is providing a stable system clock to the Oscillator module. When switching between clock sources, a delay is required to allow the new clock to stabilize. These oscillator delays are shown in Table 3-1.

### 3.3.3 LP, XT, HS MODES

The LP, XT and HS modes support the use of quartz crystal resonators or ceramic resonators connected to OSC1 and OSC2 (Figure 3-3). The mode selects a low, medium or high gain setting of the internal inverter-amplifier to support various resonator types and speed.

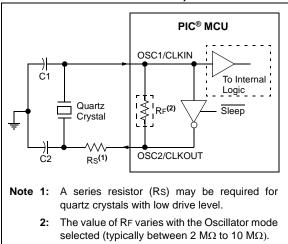
**LP** Oscillator mode selects the lowest gain setting of the internal inverter-amplifier. LP mode current consumption is the least of the three modes. This mode is designed to drive only 32.768 kHz tuning-fork type crystals (watch crystals).

**XT** Oscillator mode selects the intermediate gain setting of the internal inverter-amplifier. XT mode current consumption is the medium of the three modes. This mode is best suited to drive resonators with a medium drive level specification.

**HS** Oscillator mode selects the highest gain setting of the internal inverter-amplifier. HS mode current consumption is the highest of the three modes. This mode is best suited for resonators that require a high drive setting.

Figure 3-3 and Figure 3-4 show typical circuits for quartz crystal and ceramic resonators, respectively.

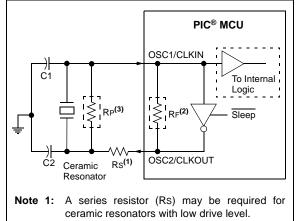
### FIGURE 3-3: QUARTZ CRYSTAL OPERATION (LP, XT OR HS MODE)



- Note 1: Quartz crystal characteristics vary according to type, package and manufacturer. The user should consult the manufacturer data sheets for specifications and recommended application.
  - **2:** Always verify oscillator performance over the VDD and temperature range that is expected for the application.
  - **3:** For oscillator design assistance, reference the following Microchip Applications Notes:
    - AN826, "Crystal Oscillator Basics and Crystal Selection for rfPIC<sup>®</sup> and PIC<sup>®</sup> Devices" (DS00826)
    - AN849, "Basic PIC<sup>®</sup> Oscillator Design" (DS00849)
    - AN943, "Practical PIC<sup>®</sup> Oscillator Analysis and Design" (DS00943)
    - AN949, "Making Your Oscillator Work" (DS00949)



### CERAMIC RESONATOR OPERATION (XT OR HS MODE)



- **2:** The value of RF varies with the Oscillator mode selected (typically between 2 MΩ to 10 MΩ).
- **3:** An additional parallel feedback resistor (RP) may be required for proper ceramic resonator operation.

### 3.3.4 EXTERNAL RC MODES

The external Resistor-Capacitor (RC) modes support the use of an external RC circuit. This allows the designer maximum flexibility in frequency choice while keeping costs to a minimum when clock accuracy is not required. There are two modes: RC and RCIO.

In RC mode, the RC circuit connects to OSC1. OSC2/ CLKOUT outputs the RC oscillator frequency divided by 4. This signal may be used to provide a clock for external circuitry, synchronization, calibration, test or other application requirements. Figure 3-5 shows the external RC mode connections.

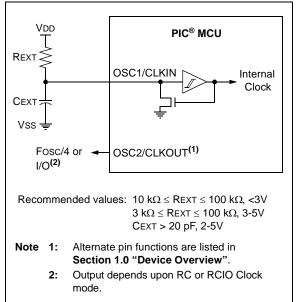


FIGURE 3-5: EXTERNAL RC MODES

In RCIO mode, the RC circuit is connected to OSC1. OSC2 becomes an additional general purpose I/O pin.

The RC oscillator frequency is a function of the supply voltage, the resistor (REXT) and capacitor (CEXT) values and the operating temperature. Other factors affecting the oscillator frequency are:

- threshold voltage variation
- component tolerances
- packaging variations in capacitance

The user also needs to take into account variation due to tolerance of external RC components used.

### 3.4 Internal Clock Modes

The Oscillator module provides a selectable system clock source of either 4 MHz or 8 MHz. The selectable frequency is configured through the IOSCFS bit of the Configuration Word.

The frequency of the internal oscillator can be can be user-adjusted via software using the OSCTUNE register.

### 3.4.1 INTOSC AND INTOSCIO MODES

The INTOSC and INTOSCIO modes configure the internal oscillators as the system clock source when the device is programmed using the oscillator selection or the FOSC<2:0> bits in the Configuration Word register (CONFIG). See Section 12.0 "Special Features of the CPU" for more information.

In **INTOSC** mode, OSC1/CLKIN is available for general purpose I/O. OSC2/CLKOUT outputs the selected internal oscillator frequency divided by 4. The CLKOUT signal may be used to provide a clock for external circuitry, synchronization, calibration, test or other application requirements.

In **INTOSCIO** mode, OSC1/CLKIN and OSC2/CLKOUT are available for general purpose I/O.

#### 3.4.1.1 OSCTUNE Register

The oscillator is factory calibrated but can be adjusted in software by writing to the OSCTUNE register (Register 3-1). The default value of the OSCTUNE register is '0'. The value is a 5-bit two's complement number.

When the OSCTUNE register is modified, the frequency will begin shifting to the new frequency. Code execution continues during this shift. There is no indication that the shift has occurred.

### REGISTER 3-1: OSCTUNE: OSCILLATOR TUNING REGISTER

U-0	U-0	U-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0
—	—	—	TUN4	TUN3	TUN2	TUN1	TUN0
bit 7							bit 0

Legend:			
R = Readable bit	W = Writable bit	U = Unimplemented bit	t, read as '0'
-n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown

### bit 7-5 Unimplemented: Read as '0'

bit 4-0

TUN<4:0>: Frequency Tuning bits
01111 = Maximum frequency
01110 =
•
•
•
00001 =
00000 = Oscillator module is running at the manufacturer calibrated frequency.
11111 =
•
•
•
10000 = Minimum frequency

### TABLE 3-2: SUMMARY OF REGISTERS ASSOCIATED WITH CLOCK SOURCES

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Value on POR, BOR	Value on all other Resets <sup>(1)</sup>
CONFIG <sup>(2)</sup>	IOSCFS	CP	MCLRE	PWRTE	WDTE	FOSC2	FOSC1	FOSC0	_	_
OSCTUNE	—	—	—	TUN4	TUN3	TUN2	TUN1	TUN0	0 0000	u uuuu

**Legend:** x = unknown, u = unchanged, - = unimplemented locations read as '0'. Shaded cells are not used by oscillators.

**Note 1:** Other (non Power-up) Resets include MCLR Reset and Watchdog Timer Reset during normal operation.

2: See Configuration Word register (Register 12-1) for operation of all register bits.

NOTES:

### 4.0 I/O PORTS

There are as many as eleven general purpose I/O pins and an input pin available. Depending on which peripherals are enabled, some or all of the pins may not be available as general purpose I/O. In general, when a peripheral is enabled, the associated pin may not be used as a general purpose I/O pin.

### 4.1 PORTA and the TRISA Registers

PORTA is a 6-bit wide, bidirectional port. The corresponding data direction register is TRISA (Register 4-2). Setting a TRISA bit (= 1) will make the corresponding PORTA pin an input (i.e., disable the output driver). Clearing a TRISA bit (= 0) will make the corresponding PORTA pin an output (i.e., enables output driver and puts the contents of the output latch on the selected pin). The exception is RA3, which is input only and its TRIS bit will always read as '1'. Example 4-1 shows how to initialize PORTA.

Reading the PORTA register (Register 4-1) reads the status of the pins, whereas writing to it will write to the PORT latch. All write operations are read-modify-write operations. Therefore, a write to a port implies that the

### REGISTER 4-1: PORTA: PORTA REGISTER

port pins are read, this value is modified and then written to the PORT data latch. RA3 reads '0' when MCLRE = 1.

The TRISA register controls the direction of the PORTA pins, even when they are being used as analog inputs. The user must ensure the bits in the TRISA register are maintained set when using them as analog inputs. I/O pins configured as analog input always read '0'.

Note:	The ANSEL register must be initialized to					
	configure an analog channel as a digital					
	input. Pins configured as analog inputs will					
	read '0' and cannot generate an interrupt.					

#### EXAMPLE 4-1: INITIALIZING PORTA

BCF	STATUS, RPO	;Bank 0
CLRF	PORTA	;Init PORTA
BSF	STATUS, RPO	;Bank 1
CLRF	ANSEL	;digital I/O
MOVLW	0Ch	;Set RA<3:2> as inputs
MOVWF	TRISA	;and set RA<5:4,1:0>
		;as outputs
BCF	STATUS, RPO	;Bank 0

U-0	U-0	R/W-x	R/W-0	R-x	R/W-0	R/W-0	R/W-0
—	—	RA5	RA4	RA3	RA2	RA1	RA0
bit 7							bit 0

Legend:				
R = Readabl	le bit	W = Writable bit	U = Unimplemented bit, r	ead as '0'
-n = Value at POR		'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown
= Value at	t POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown
bit 7-6	Unimple	mented: Read as '0'		
bit 5-0	RA<5:0>	: PORTA I/O Pin bit		
		TA nin is > VIH		

1 = PORTA pin is > VIH

0 = PORTA pin is < VIL

### REGISTER 4-2: TRISA: PORTA TRI-STATE REGISTER

U-0	U-0	R/W-1	R/W-1	R-1	R/W-1	R/W-1	R/W-1
—	—	TRISA5	TRISA4	TRISA3	TRISA2	TRISA1	TRISA0
bit 7							bit 0

Legend:					
R = Readable bit	W = Writable bit	U = Unimplemented bit, r	U = Unimplemented bit, read as '0'		
-n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown		

bit 7-6 Unimplemented: Read as '0'

TRISA<5:0>: PORTA Tri-State Control bit

1 = PORTA pin configured as an input (tri-stated)

0 = PORTA pin configured as an output

Note 1: TRISA<3> always reads '1'.

2: TRISA<5:4> always reads '1' in XT, HS and LP Oscillator modes.

bit 5-0

### 4.2 Additional Pin Functions

Every PORTA pin on the PIC16F610/616/16HV610/ 616 has an interrupt-on-change option and a weak pullup option. The next three sections describe these functions.

### 4.2.1 ANSEL REGISTER

The ANSEL register is used to configure the Input mode of an I/O pin to analog. Setting the appropriate ANSEL bit high will cause all digital reads on the pin to be read as '0' and allow analog functions on the pin to operate correctly.

The state of the ANSEL bits has no affect on digital output functions. A pin with TRIS clear and ANSEL set will still operate as a digital output, but the Input mode will be analog. This can cause unexpected behavior when executing read-modify-write instructions on the affected port.

### 4.2.2 WEAK PULL-UPS

Each of the PORTA pins, except RA3, has an individually configurable internal weak pull-up. Control bits WPUAx enable or disable each pull-up. Refer to Register 4-4. Each weak pull-up is automatically turned off when the port pin is configured as an output. The pull-ups are disabled on a Power-on Reset by the RAPU bit of the OPTION register). A weak pull-up is automatically enabled for RA3 when configured as MCLR and disabled when RA3 is an input. There is no software control of the MCLR pull-up.

### 4.2.3 INTERRUPT-ON-CHANGE

Each PORTA pin is individually configurable as an interrupt-on-change pin. Control bits IOCAx enable or disable the interrupt function for each pin. Refer to Register 4-5. The interrupt-on-change is disabled on a Power-on Reset.

For enabled interrupt-on-change pins, the values are compared with the old value latched on the last read of PORTA. The 'mismatch' outputs of the last read are OR'd together to set the PORTA Change Interrupt Flag bit (RAIF) in the INTCON register (Register 2-3).

This interrupt can wake the device from Sleep. The user, in the Interrupt Service Routine, clears the interrupt by:

- a) Any read or write of PORTA. This will end the mismatch condition, then,
- b) Clear the flag bit RAIF.

A mismatch condition will continue to set flag bit RAIF. Reading PORTA will end the mismatch condition and allow flag bit RAIF to be cleared. The <u>latch</u> holding the last read value is not affected by a MCLR nor BOR Reset. After these resets, the RAIF flag will continue to be set if a mismatch is present.

Note: If a change on the I/O pin should occur when any PORTA operation is being executed, then the RAIF interrupt flag may not get set.

### REGISTER 4-3: ANSEL: ANALOG SELECT REGISTER

R/W-1	R/W-1	R/W-1	R/W-1	R/W-1	R/W-1	R/W-1	R/W-1
ANS7	ANS6	ANS5	ANS4	ANS3 <sup>(2)</sup>	ANS2 <sup>(2)</sup>	ANS1	ANS0
bit 7			•				bit 0
Legend:							
<b>Legend:</b> R = Readable	bit	W = Writable	bit	U = Unimplen	nented bit, reac	l as '0'	

bit 7-0 ANS<7:0>: Analog Select bits

Analog select between analog or digital function on pins AN<7:0>, respectively.

1 =Analog input. Pin is assigned as analog input<sup>(1)</sup>.

0 = Digital I/O. Pin is assigned to port or special function.

- **Note 1:** Setting a pin to an analog input automatically disables the digital input circuitry, weak pull-ups, and interrupt-on-change if available. The corresponding TRIS bit must be set to Input mode in order to allow external control of the voltage on the pin.
  - 2: PIC16F616/HV616.

U-0	U-0	R/W-1	R/W-1	U-0	R/W-1	R/W-1	R/W-1	
_	—	WPUA5	WPUA4	—	WPUA2	WPUA1	WPUA0	
bit 7							bit 0	
Legend:								
R = Readable	bit	W = Writable	bit	U = Unimplemented bit, read as '0'				
-n = Value at I	POR	'1' = Bit is set		'0' = Bit is cleared		x = Bit is unknown		
bit 7-6 bit 5-4	-							

#### **REGISTER 4-4:** WPUA: WEAK PULL-UP PORTA REGISTER

bit 3 Unimplemented: Read as '0' bit 2-0

- WPUA<2:0>: Weak Pull-up Control bits
  - 1 =Pull-up enabled
  - 0 = Pull-up disabled

**Note 1:** Global RAPU must be enabled for individual pull-ups to be enabled.

- 2: The weak pull-up device is automatically disabled if the pin is in Output mode (TRISA = 0).
- 3: The RA3 pull-up is enabled when configured as MCLR and disabled as an input in the Configuration Word.
- 4: WPUA<5:4> always reads '1' in XT, HS and LP Oscillator modes.

#### **REGISTER 4-5: IOCA: INTERRUPT-ON-CHANGE PORTA REGISTER**

U-0	U-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0
—		IOCA5	IOCA4	IOCA3	IOCA2	IOCA1	IOCA0
bit 7							bit 0

Legend:			
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'	
-n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown

bit 7-6 Unimplemented: Read as '0'

bit 5-0 IOCA<5:0>: Interrupt-on-change PORTA Control bit

1 = Interrupt-on-change enabled

0 = Interrupt-on-change disabled

Note 1: Global Interrupt Enable (GIE) must be enabled for individual interrupts to be recognized.

2: IOCA<5:4> always reads '1' in XT, HS and LP Oscillator modes.

### 4.2.4 PIN DESCRIPTIONS AND DIAGRAMS

Each PORTA pin is multiplexed with other functions. The pins and their combined functions are briefly described here. For specific information about individual functions such as the Comparator or the ADC, refer to the appropriate section in this data sheet.

### 4.2.4.1 RA0/AN0<sup>(1)</sup>/C1IN+/ICSPDAT

Figure 4-1 shows the diagram for this pin. The RA0 pin is configurable to function as one of the following:

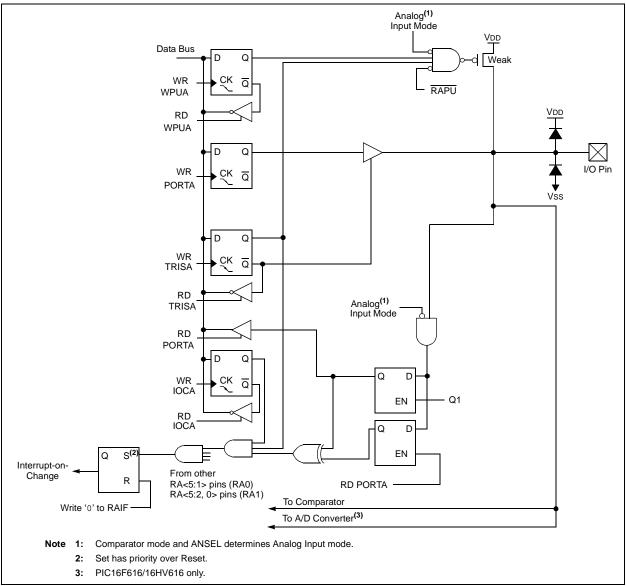
- a general purpose I/O
- an analog input for the ADC<sup>(1)</sup>
- an analog non-inverting input to the comparator
- In-Circuit Serial Programming data

### 4.2.4.2 RA1/AN1<sup>(1)</sup>/C12IN0-/VREF<sup>(1)</sup>/ ICSPCLK

Figure 4-1 shows the diagram for this pin. The RA1 pin is configurable to function as one of the following:

- a general purpose I/O
- an analog input for the ADC<sup>(1)</sup>
- an analog inverting input to the comparator
- a voltage reference input for the ADC<sup>(1)</sup>
- In-Circuit Serial Programming clock

Note 1: PIC16F616/16HV616 only.



### FIGURE 4-1: BLOCK DIAGRAM OF RA<1:0>

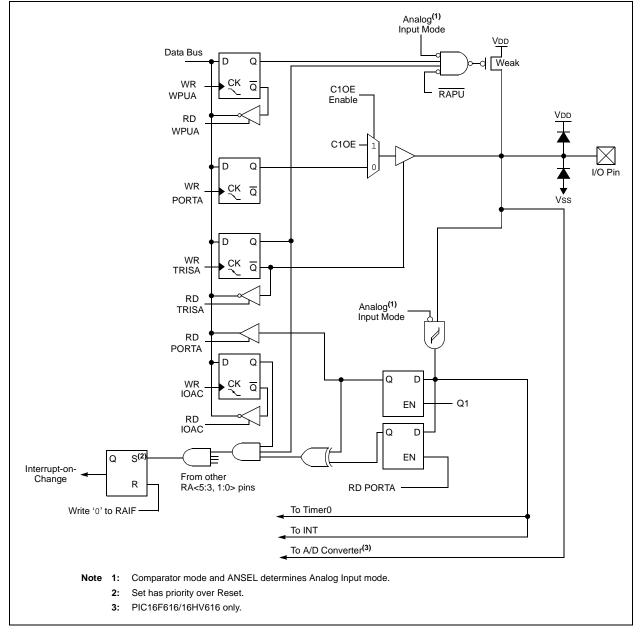
### 4.2.4.3 RA2/AN2<sup>(1)</sup>/T0CKI/INT/C1OUT

Figure 4-2 shows the diagram for this pin. The RA2 pin is configurable to function as one of the following:

- a general purpose I/O
- an analog input for the ADC<sup>(1)</sup>
- the clock input for TMR0
- an external edge triggered interrupt
- a digital output from Comparator C1



#### FIGURE 4-2: BLOCK DIAGRAM OF RA2

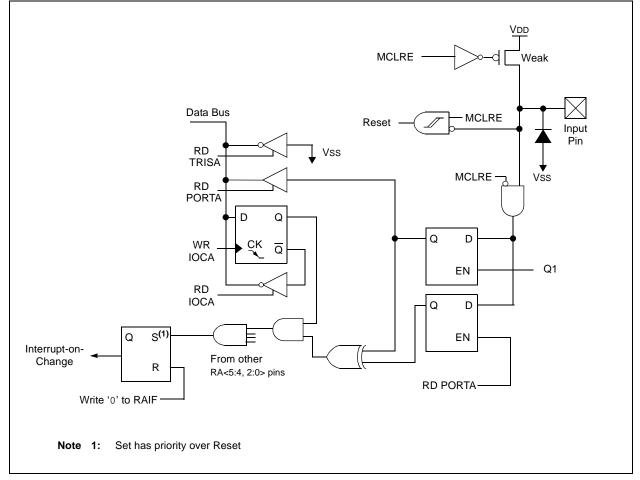


#### 4.2.4.4 RA3/MCLR/VPP

Figure 4-3 shows the diagram for this pin. The RA3 pin is configurable to function as one of the following:

- a general purpose input
- as Master Clear Reset with weak pull-up
- High Voltage Programming voltage input



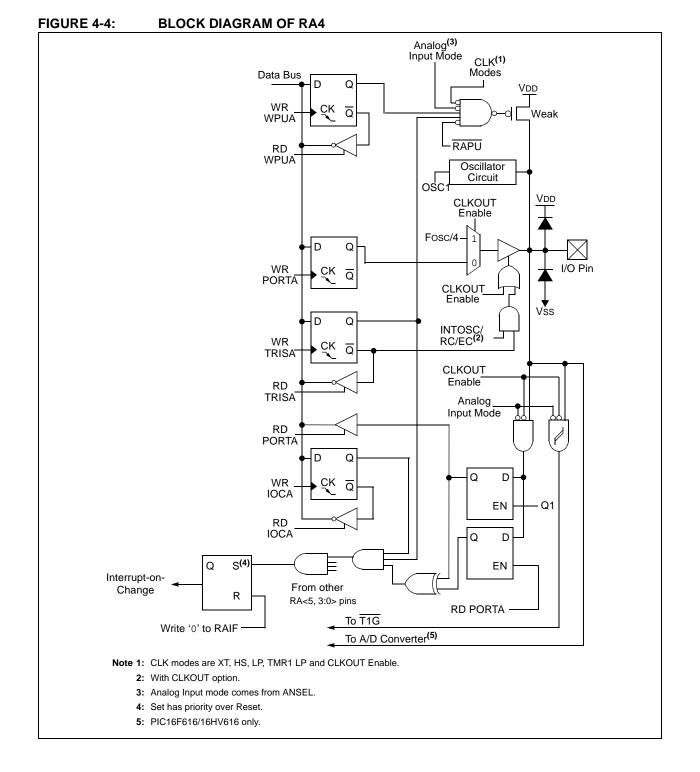


#### 4.2.4.5 RA4/AN3<sup>(1)</sup>/T1G/OSC2/CLKOUT

Figure 4-4 shows the diagram for this pin. The RA4 pin is configurable to function as one of the following:

- a general purpose I/O
- an analog input for the ADC<sup>(1)</sup>

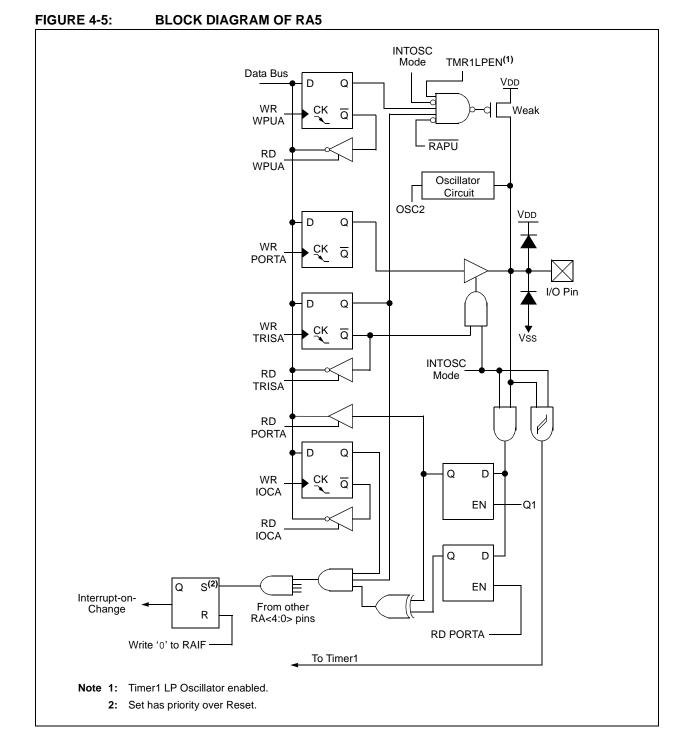
- a Timer1 gate (count enable)
- a crystal/resonator connection
- · a clock output
  - Note 1: PIC16F616/16HV616 only.



#### 4.2.4.6 RA5/T1CKI/OSC1/CLKIN

Figure 4-5 shows the diagram for this pin. The RA5 pin is configurable to function as one of the following:

- a general purpose I/O
- a Timer1 clock input
- a crystal/resonator connection
- a clock input



Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Value on POR, BOR	Value on all other Resets
ANSEL	ANS7	ANS6	ANS5	ANS4	ANS3 <sup>(1)</sup>	ANS2 <sup>(1)</sup>	ANS1	ANS0	1111 1111	1111 1111
CM1CON0	C10N	C10UT	C10E	C1POL	—	C1R	C1CH1	C1CH0	0000 -000	0000 -000
CM2CON0	C2ON	C2OUT	C2OE	C2POL	—	C2R	C2CH1	C2CH0	0000 -000	0000 -000
INTCON	GIE	PEIE	TOIE	INTE	RAIE	T0IF	INTF	RAIF	0000 0000	0000 0000
IOCA		—	IOCA5	IOCA4	IOCA3	IOCA2	IOCA1	IOCA0	00 0000	00 0000
OPTION_REG	RAPU	INTEDG	TOCS	TOSE	PSA	PS2	PS1	PS0	1111 1111	1111 1111
PORTA	_	_	RA5	RA4	RA3	RA2	RA1	RA0	x0 x000	u0 u000
TRISA	_	—	TRISA5	TRISA4	TRISA3	TRISA2	TRISA1	TRISA0	11 1111	11 1111
WPUA	_	_	WPUA5	WPUA4	_	WPUA2	WPUA1	WPUA0	11 -111	11 -111

#### TABLE 4-1: SUMMARY OF REGISTERS ASSOCIATED WITH PORTA

 $\label{eq:Legend: Legend: Legend: u = unchanged, - = unimplemented locations read as `0'. Shaded cells are not used by PORTA.$ 

Note 1: For PIC16F616/HV616 only.

#### 4.3 PORTC and the TRISC Registers

PORTC is a general purpose I/O port consisting of 6 bidirectional pins. The pins can be configured for either digital I/O or analog input to A/D Converter (ADC) or Comparator. For specific information about individual functions such as the Enhanced CCP or the ADC, refer to the appropriate section in this data sheet.

Note:	The ANSEL register must be initialized to
	configure an analog channel as a digital
	input. Pins configured as analog inputs will
	read '0' and cannot generate an interrupt.

#### EXAMPLE 4-2: INITIALIZING PORTC

BCF	STATUS, RPO	;Bank 0
CLRF	PORTC	;Init PORTC
BSF	STATUS, RPO	;Bank 1
CLRF	ANSEL	;digital I/O
MOVLW	0Ch	;Set RC<3:2> as inputs
MOVWF	TRISC	;and set RC<5:4,1:0>
		;as outputs
BCF	STATUS, RPO	;Bank 0

#### REGISTER 4-6: PORTC: PORTC REGISTER

U-0	U-0	R/W-x	R/W-x	R/W-0	R/W-0	R/W-x	R/W-x
—	—	RC5	RC4	RC3	RC2	RC1	RC0
bit 7							bit 0

#### Legend:

Legenu.			
R = Readable bit	W = Writable bit	U = Unimplemented bit,	read as '0'
-n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown

bit 5-0 RC<5:0>: PORTC I/O Pin bit

1 = PORTC pin is > VIH

0 = PORTC pin is < VIL

#### REGISTER 4-7: TRISC: PORTC TRI-STATE REGISTER

U-0	U-0	R/W-1	R/W-1	R/W-1	R/W-1	R/W-1	R/W-1
—	—	TRISC5	TRISC4	TRISC3	TRISC2	TRISC1	TRISC0
bit 7							bit 0

Legend:			
R = Readable bit	W = Writable bit	U = Unimplemented bit, read	l as '0'
-n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown

bit 7-6 Unimplemented: Read as '0'

TRISC<5:0>: PORTC Tri-State Control bit

1 = PORTC pin configured as an input (tri-stated)

0 = PORTC pin configured as an output

bit 5-0

#### 4.3.1 RC0/AN4<sup>(1)</sup>/C2IN+

The RC0 is configurable to function as one of the following:  $\label{eq:configurable}$ 

- a general purpose I/O
- an analog input for the ADC<sup>(1)</sup>
- an analog non-inverting input to Comparator C2

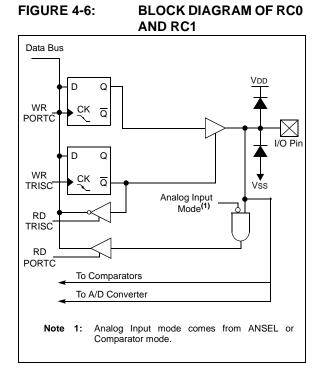
#### 4.3.2 RC1/AN5<sup>(1)</sup>/C12IN1-

The RC1 is configurable to function as one of the following:  $\label{eq:configurable}$ 

- a general purpose I/O
- an analog input for the ADC<sup>(1)</sup>

an analog inverting input to the comparator

Note 1: PIC16F616/16HV616 only.



#### 4.3.3 RC2/AN6<sup>(1)</sup>/C12IN2-/P1D<sup>(1)</sup>

The RC2 is configurable to function as one of the following:

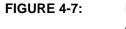
- a general purpose I/O
- an analog input for the ADC<sup>(1)</sup>
- an analog input to Comparators C1 and C2
- a digital output from the Enhanced CCP<sup>(1)</sup>

#### 4.3.4 RC3/AN7<sup>(1)</sup>/C12IN3-/P1C<sup>(1)</sup>

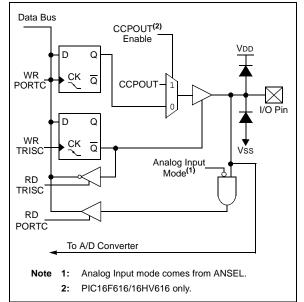
The RC3 is configurable to function as one of the following:

- a general purpose I/O
- an analog input for the ADC<sup>(1)</sup>
- an analog inverting input to Comparators C1 and C2
- a digital output from the Enhanced CCP<sup>(1)</sup>

Note 1: PIC16F616/16HV616 only.



#### 7: BLOCK DIAGRAM OF RC2 AND RC3



#### 4.3.5 RC4/C2OUT/P1B<sup>(1)</sup>

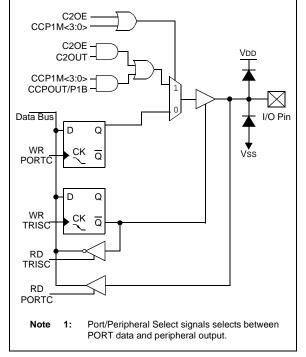
The RC4 is configurable to function as one of the following:

- a general purpose I/O
- a digital output from Comparator C2
- a digital output from the Enhanced CCP<sup>(1)</sup>

Note 1: PIC16F616/16HV616 only.

2: Enabling both C2OUT and P1B will cause a conflict on RC4 and create unpredictable results. Therefore, if C2OUT is enabled, the ECCP can not be used in Half-Bridge or Full-Bridge mode and vice-versa.





#### 4.3.6 RC5/CCP1<sup>(1)</sup>/P1A<sup>(1)</sup>

The RC5 is configurable to function as one of the following:  $\label{eq:configurable}$ 

- a general purpose I/O
- a digital input/output for the Enhanced CCP<sup>(1)</sup>
  - Note 1: PIC16F616/16HV616 only.

FIGURE 4-9: BLO

#### BLOCK DIAGRAM OF RC5 PIN

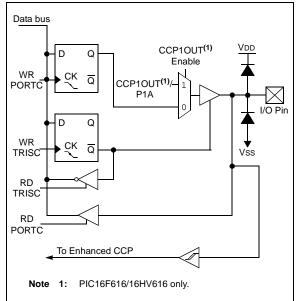


TABLE 4-2:	SUMMARY OF REGISTERS ASSOCIATED WITH PORTC
IADEE + 2.	

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Value on POR, BOR	Value on all other Resets
ANSEL	ANS7	ANS6	ANS5	ANS4	ANS3 <sup>(1)</sup>	ANS2 <sup>(1)</sup>	ANS1	ANS0	1111 1111	1111 1111
CCP1CON <sup>(1)</sup>	P1M1	P1M0	DC1B1	DC1B0	CCP1M3	CCP1M2	CCP1M1	CCP1M0	0000 0000	0000 0000
CM1CON0	C10N	C1OUT	C10E	C1POL	—	C1R	C1CH1	C1CH0	0000 -000	0000 -000
CM2CON0	C2ON	C2OUT	C2OE	C2POL	_	C2R	C2CH1	C2CH0	0000 -000	0000 -000
PORTC		_	RC5	RC4	RC3	RC2	RC1	RC0	xx 00xx	uu 00uu
TRISC	_	—	TRISC5	TRISC4	TRISC3	TRISC2	TRISC1	TRISC0	11 1111	11 1111

Legend: x = unknown, u = unchanged, - = unimplemented locations read as '0'. Shaded cells are not used by PORTC.

Note 1: PIC16F616/HV616 only.

#### 5.0 TIMER0 MODULE

The Timer0 module is an 8-bit timer/counter with the following features:

- 8-bit timer/counter register (TMR0)
- 8-bit prescaler (shared with Watchdog Timer)
- · Programmable internal or external clock source
- Programmable external clock edge selection
- Interrupt on overflow

Figure 5-1 is a block diagram of the Timer0 module.

#### 5.1 Timer0 Operation

When used as a timer, the Timer0 module can be used as either an 8-bit timer or an 8-bit counter.

#### 5.1.1 8-BIT TIMER MODE

When used as a timer, the Timer0 module will increment every instruction cycle (without prescaler). Timer mode is selected by clearing the T0CS bit of the OPTION register to '0'.

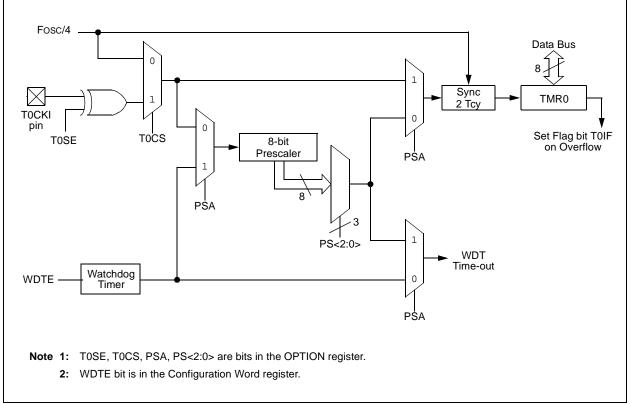
When TMR0 is written, the increment is inhibited for two instruction cycles immediately following the write.

Note: The value written to the TMR0 register can be adjusted, in order to account for the two instruction cycle delay when TMR0 is written.

#### 5.1.2 8-BIT COUNTER MODE

When used as a counter, the Timer0 module will increment on every rising or falling edge of the T0CKI pin. The incrementing edge is determined by the T0SE bit of the OPTION register. Counter mode is selected by setting the T0CS bit of the OPTION register to '1'.





### 5.1.3 SOFTWARE PROGRAMMABLE PRESCALER

A single software programmable prescaler is available for use with either Timer0 or the Watchdog Timer (WDT), but not both simultaneously. The prescaler assignment is controlled by the PSA bit of the OPTION register. To assign the prescaler to Timer0, the PSA bit must be cleared to a '0'.

There are 8 prescaler options for the Timer0 module ranging from 1:2 to 1:256. The prescale values are selectable via the PS<2:0> bits of the OPTION register. In order to have a 1:1 prescaler value for the Timer0 module, the prescaler must be assigned to the WDT module.

The prescaler is not readable or writable. When assigned to the Timer0 module, all instructions writing to the TMR0 register will clear the prescaler.

When the prescaler is assigned to WDT, a CLRWDT instruction will clear the prescaler along with the WDT.

#### 5.1.3.1 Switching Prescaler Between Timer0 and WDT Modules

As a result of having the prescaler assigned to either Timer0 or the WDT, it is possible to generate an unintended device Reset when switching prescaler values. When changing the prescaler assignment from Timer0 to the WDT module, the instruction sequence shown in Example 5-1 must be executed.

### EXAMPLE 5-1: CHANGING PRESCALER (TIMER0 $\rightarrow$ WDT)

BANKSEL	TMR0	;
CLRWDT		;Clear WDT
CLRF	TMR0	;Clear TMR0 and
		;prescaler
BANKSEI	OPTION_REG	;
BSF	OPTION_REG,PSA	;Select WDT
CLRWDT	;	
	;	
MOVLW	b'11111000'	;Mask prescaler
ANDWF	OPTION_REG,W	;bits
IORLW	b'00000101'	;Set WDT prescaler
MOVWF	OPTION_REG	;to 1:32

When changing the prescaler assignment from the WDT to the Timer0 module, the following instruction sequence must be executed (see Example 5-2).

EXAMPLE 5-2:	CHANGING PRESCALER
	(WDT $\rightarrow$ TIMER0)

CLRWDT		;Clear WDT and ;prescaler
BANKSEL	OPTION_REG	;
MOVLW	b'11110000'	;Mask TMR0 select and
ANDWF		prescaler bits
IORLW	b'00000011'	;Set prescale to 1:16
MOVWF	OPTION_REG	;

#### 5.1.4 TIMER0 INTERRUPT

Timer0 will generate an interrupt when the TMR0 register overflows from FFh to 00h. The T0IF interrupt flag bit of the INTCON register is set every time the TMR0 register overflows, regardless of whether or not the Timer0 interrupt is enabled. The T0IF bit must be cleared in software. The Timer0 interrupt enable is the T0IE bit of the INTCON register.

Note:	The Timer0 interrupt cannot wake the
	processor from Sleep since the timer is frozen during Sleep.
	nozen uunny Sleep.

### 5.1.5 USING TIMER0 WITH AN EXTERNAL CLOCK

When Timer0 is in Counter mode, the synchronization of the T0CKI input and the Timer0 register is accomplished by sampling the prescaler output on the Q2 and Q4 cycles of the internal phase clocks. Therefore, the high and low periods of the external clock source must meet the timing requirements as shown in **Section 15.0 "Electrical Specifications"**.

R/W-1	R/W-1	R/W-1	R/W-1	R/W-1	R/W-1	R/W-1	R/W-1			
RAPU	INTEDG	TOCS	TOSE	PSA	PS2	PS1	PS0			
bit 7							bit			
Legend:										
R = Readat	ala hit	W = Writable	hit	LI – Unimplor	mented bit, rea	ud ac '0'				
				-						
-n = Value a	at POR	'1' = Bit is se	[	'0' = Bit is cle	ared	x = Bit is unki	nown			
bit 7		TA Pull-up Ena	hle hit							
		oull-ups are dis								
				idual PORT late	h values					
bit 6	•	errupt Edge Se	•							
		<ol> <li>I = Interrupt on rising edge of INT pin</li> <li>Interrupt on falling edge of INT pin</li> </ol>								
bit 5	-	) Clock Source	-							
	1 = Transitio	1 = Transition on TOCKI pin								
		0 = Internal instruction cycle clock (Fosc/4)								
bit 4	TOSE: TMRC	0SE: TMR0 Source Edge Select bit								
		1 = Increment on high-to-low transition on T0CKI pin								
		0 = Increment on low-to-high transition on TOCKI pin								
bit 3	PSA: Presca	ler Assignmen	t bit							
	1 = Prescale	1 = Prescaler is assigned to the WDT								
		r is assigned to		nodule						
bit 2-0	<b>PS&lt;2:0&gt;:</b> Pr	escaler Rate S	elect bits							
	BIT	VALUE TMR0 R	ATE WDT RA	TE						
		000 1:2	1:1							
		001 1:4								
		010 1:8	1:4							
		011 1:1								
		100 1:3 101 1:6								
		110 1:12								
		111 1:2								

#### **REGISTER 5-1: OPTION\_REG: OPTION REGISTER**

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Valu POR,		Valu all o Res	ther
TMR0	Timer0 N	/lodules R	egister						xxxx	xxxx	uuuu	uuuu
INTCON	GIE	PEIE	TOIE	INTE	RAIE	TOIF	INTF	RAIF	0000	0000	0000	0000
OPTION_REG	RAPU	INTEDG	TOCS	TOSE	PSA	PS2	PS1	PS0	1111	1111	1111	1111
TRISA	_	—	TRISA5	TRISA4	TRISA3	TRISA2	TRISA1	TRISA0	11	1111	11	1111

**Legend:** – = Unimplemented locations, read as '0', u = unchanged, x = unknown. Shaded cells are not used by the Timer0 module.

NOTES:

### 6.0 TIMER1 MODULE WITH GATE CONTROL

The Timer1 module is a 16-bit timer/counter with the following features:

- 16-bit timer/counter register pair (TMR1H:TMR1L)
- Programmable internal or external clock source
- 3-bit prescaler
- · Optional LP oscillator
- Synchronous or asynchronous operation
- Timer1 gate (count enable) via comparator or  $\overline{\text{T1G}}$  pin
- Interrupt on overflow
- Wake-up on overflow (external clock, Asynchronous mode only)
- Time base for the Capture/Compare function
- Special Event Trigger (with ECCP)
- Comparator output synchronization to Timer1 clock

Figure 6-1 is a block diagram of the Timer1 module.

#### 6.1 Timer1 Operation

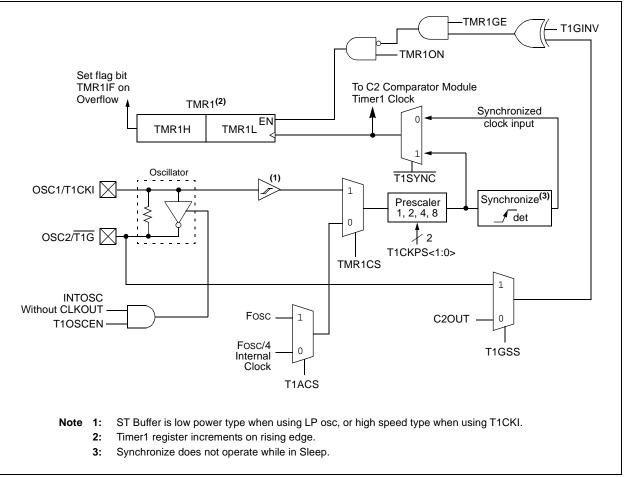
The Timer1 module is a 16-bit incrementing counter which is accessed through the TMR1H:TMR1L register pair. Writes to TMR1H or TMR1L directly update the counter.

When used with an internal clock source, the module is a timer. When used with an external clock source, the module can be used as either a timer or counter.

#### 6.2 Clock Source Selection

The TMR1CS bit of the T1CON register is used to select the clock source. When TMR1CS = 0, the clock source is FOSC/4. When TMR1CS = 1, the clock source is supplied externally.

Clock Source	TMR1CS	T1ACS
Fosc/4	0	0
Fosc	0	1
T1CKI pin	1	x



#### FIGURE 6-1: TIMER1 BLOCK DIAGRAM

#### 6.2.1 INTERNAL CLOCK SOURCE

When the internal clock source is selected the TMR1H:TMR1L register pair will increment on multiples of TCY as determined by the Timer1 prescaler.

#### 6.2.2 EXTERNAL CLOCK SOURCE

When the external clock source is selected, the Timer1 module may work as a timer or a counter.

When counting, Timer1 is incremented on the rising edge of the external clock input T1CKI. In addition, the Counter mode clock can be synchronized to the microcontroller system clock or run asynchronously.

If an external clock oscillator is needed (and the microcontroller is using the INTOSC without CLKOUT), Timer1 can use the LP oscillator as a clock source.

Note:	In Counter mode, a falling edge must be
	registered by the counter prior to the first
	incrementing rising edge.

#### 6.3 Timer1 Prescaler

Timer1 has four prescaler options allowing 1, 2, 4 or 8 divisions of the clock input. The T1CKPS bits of the T1CON register control the prescale counter. The prescale counter is not directly readable or writable; however, the prescaler counter is cleared upon a write to TMR1H or TMR1L.

#### 6.4 Timer1 Oscillator

A low-power 32.768 kHz crystal oscillator is built-in between pins OSC1 (input) and OSC2 (output). The oscillator is enabled by setting the T1OSCEN control bit of the T1CON register. The oscillator will continue to run during Sleep.

The Timer1 oscillator is shared with the system LP oscillator. Thus, Timer1 can use this mode only when the primary system clock is derived from the internal oscillator or when the oscillator is in the LP Oscillator mode. The user must provide a software time delay to ensure proper oscillator start-up.

TRISA5 and TRISA4 bits are set when the Timer1 oscillator is enabled. RA5 and RA4 bits read as '0' and TRISA5 and TRISA4 bits read as '1'.

Note: The oscillator requires a start-up and stabilization time before use. Thus, T1OSCEN should be set and a suitable delay observed prior to enabling Timer1.

#### 6.5 Timer1 Operation in Asynchronous Counter Mode

If control bit T1SYNC of the T1CON register is set, the external clock input is not synchronized. The timer continues to increment asynchronous to the internal phase clocks. The timer will continue to run during Sleep and can generate an interrupt on overflow, which will wake-up the processor. However, special precautions in software are needed to read/write the timer (see Section 6.5.1 "Reading and Writing Timer1 in Asynchronous Counter Mode").

- Note: When switching from synchronous to asynchronous operation, it is possible to skip an increment. When switching from asynchronous to synchronous operation, it is possible to produce an additional increment.
- Note: In asynchronous counter mode or when using the internal oscillator and T1ACS=1, Timer1 can not be used as a time base for the capture or compare modes of the ECCP module (for PIC16F616/HV616 only).

#### 6.5.1 READING AND WRITING TIMER1 IN ASYNCHRONOUS COUNTER MODE

Reading TMR1H or TMR1L while the timer is running from an external asynchronous clock will ensure a valid read (taken care of in hardware). However, the user should keep in mind that reading the 16-bit timer in two 8-bit values itself, poses certain problems, since the timer may overflow between the reads.

For writes, it is recommended that the user simply stop the timer and write the desired values. A write contention may occur by writing to the timer registers, while the register is incrementing. This may produce an unpredictable value in the TMR1H:TMR1L register pair.

#### 6.6 Timer1 Gate

Timer1 gate source is software configurable to be the T1G pin or the output of Comparator C2. This allows the device to directly time external events using T1G or analog events using Comparator C2. See the CM2CON1 register (Register 8-3) for selecting the Timer1 gate source. This feature can simplify the software for a Delta-Sigma A/D converter and many

other applications. For more information on Delta-Sigma A/D converters, see the Microchip web site (www.microchip.com).

Note:	TMR1GE bit of the T1CON register must
	be set to use either $\overline{T1G}$ or C2OUT as the
	Timer1 gate source. See the CM2CON1
	register (Register 8-3) for more informa-
	tion on selecting the Timer1 gate source.

Timer1 gate can be inverted using the T1GINV bit of the T1CON register, whether it originates from the T1G pin or Comparator C2 output. This configures Timer1 to measure either the active-high or active-low time between events.

#### 6.7 Timer1 Interrupt

The Timer1 register pair (TMR1H:TMR1L) increments to FFFFh and rolls over to 0000h. When Timer1 rolls over, the Timer1 interrupt flag bit of the PIR1 register is set. To enable the interrupt on rollover, you must set these bits:

- TMR1IE bit of the PIE1 register
- PEIE bit of the INTCON register
- GIE bit of the INTCON register
- T1SYNC bit of the T1CON register
- TMR1CS bit of the T1CON register
- T1OSCEN bit of the T1CON register (can be set)

The interrupt is cleared by clearing the TMR1IF bit in the Interrupt Service Routine.

Note: The TMR1H:TTMR1L register pair and the TMR1IF bit should be cleared before enabling interrupts.

#### 6.8 Timer1 Operation During Sleep

Timer1 can only operate during Sleep when setup in Asynchronous Counter mode. In this mode, an external crystal or clock source can be used to increment the counter. To set up the timer to wake the device:

- TMR1ON bit of the T1CON register must be set
- TMR1IE bit of the PIE1 register must be set
- PEIE bit of the INTCON register must be set

The device will wake-up on an overflow and execute the next instruction. If the GIE bit of the INTCON register is set, the device will call the Interrupt Service Routine (0004h).

#### 6.9 ECCP Capture/Compare Time Base (PIC16F616/16HV616 Only)

The ECCP module uses the TMR1H:TMR1L register pair as the time base when operating in Capture or Compare mode.

In Capture mode, the value in the TMR1H:TMR1L register pair is copied into the CCPR1H:CCPR1L register pair on a configured event.

In Compare mode, an event is triggered when the value CCPR1H:CCPR1L register pair matches the value in the TMR1H:TMR1L register pair. This event can be a Special Event Trigger.

For more information, see Section 10.0 "Enhanced Capture/Compare/PWM (With Auto-Shutdown and Dead Band) Module (PIC16F616/16HV616 Only)".

#### 6.10 ECCP Special Event Trigger (PIC16F616/16HV616 Only)

When the ECCP is configured to trigger a special event, the trigger will clear the TMR1H:TMR1L register pair. This special event does not cause a Timer1 interrupt. The ECCP module may still be configured to generate a ECCP interrupt.

In this mode of operation, the CCPR1H:CCPR1L register pair effectively becomes the period register for Timer1.

Timer1 should be synchronized to the Fosc to utilize the Special Event Trigger. Asynchronous operation of Timer1 can cause a Special Event Trigger to be missed.

In the event that a write to TMR1H or TMR1L coincides with a Special Event Trigger from the ECCP, the write will take precedence.

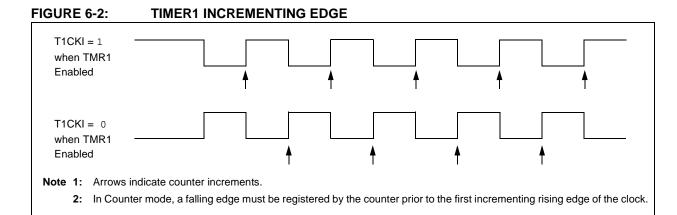
For more information, see **Section 10.2.4** "**Special Event Trigger**".

#### 6.11 Comparator Synchronization

The same clock used to increment Timer1 can also be used to synchronize the comparator output. This feature is enabled in the Comparator module.

When using the comparator for Timer1 gate, the comparator output should be synchronized to Timer1. This ensures Timer1 does not miss an increment if the comparator changes.

For more information, see Section 8.8.2 "Synchronizing Comparator C2 Output to Timer1".



#### 6.12 Timer1 Control Register

The Timer1 Control register (T1CON), shown in Register 6-1, is used to control Timer1 and select the various features of the Timer1 module.

#### REGISTER 6-1: T1CON: TIMER1 CONTROL REGISTER

R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0
T1GINV <sup>(1)</sup>	TMR1GE <sup>(2)</sup>	T1CKPS1	T1CKPS0	T1OSCEN	T1SYNC	TMR1CS	TMR10N
bit 7							bit 0

Legend:			
R = Readable bit	W = Writable bit	U = Unimplemented bit,	, read as '0'
-n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown

bit 7	<b>T1GINV:</b> Timer1 Gate Invert bit <sup>(1)</sup> 1 = Timer1 gate is active-high (Timer1 counts when gate is high) 0 = Timer1 gate is active-low (Timer1 counts when gate is low)
bit 6	TMR1GE: Timer1 Gate Enable bit <sup>(2)</sup> If TMR1ON = 0:         This bit is ignored         If TMR1ON = 1:         1 = Timer1 counting is controlled by the Timer1 Gate function         0 = Timer1 is always counting
bit 5-4	<b>T1CKPS&lt;1:0&gt;:</b> Timer1 Input Clock Prescale Select bits 11 = 1:8 Prescale Value 10 = 1:4 Prescale Value 01 = 1:2 Prescale Value 00 = 1:1 Prescale Value
bit 3	T1OSCEN: LP Oscillator Enable Control bit <u>If INTOSC without CLKOUT oscillator is active:</u> 1 = LP oscillator is enabled for Timer1 clock 0 = LP oscillator is off <u>Else:</u> This bit is ignored

#### REGISTER 6-1: T1CON: TIMER1 CONTROL REGISTER (CONTINUED)

bit 2	T1SYNC: Timer1 External Clock Input Synchronization Control bit
	<u>TMR1CS = 1:</u>
	1 = Do not synchronize external clock input
	0 = Synchronize external clock input
	<u>TMR1CS = 0:</u>
	This bit is ignored. Timer1 uses the internal clock
bit 1	TMR1CS: Timer1 Clock Source Select bit
	1 = External clock from T1CKI pin (on the rising edge) 0 = Internal clock <u>If TMR1ACS = 0:</u> FOSC/4 <u>If TMR1ACS = 1:</u> FOSC
bit 0	<b>TMR1ON:</b> Timer1 On bit 1 = Enables Timer1 0 = Stops Timer1

Note 1: T1GINV bit inverts the Timer1 gate logic, regardless of source.
 2: TMR1GE bit must be set to use either T1G pin or C2OUT, as selected by the T1GSS bit of the CM2CON1 register, as a Timer1 gate source.

#### TABLE 6-1: SUMMARY OF REGISTERS ASSOCIATED WITH TIMER1

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Value on POR, BOR	Value on all other Resets
CM2CON0	C2ON	C2OUT	C2OE	C2POL	_	C2R	C2CH1	C2CH0	0000 -000	0000 -000
CM2CON1	MC1OUT	MC2OUT	_	T1ACS	C1HYS	C2HYS	T1GSS	C2SYNC	00-0 0010	00-0 0010
INTCON	GIE	PEIE	TOIE	INTE	RAIE	T0IF	INTF	RAIF	0000 0000	0000 0000
PIE1	_	ADIE <sup>(1)</sup>	CCP1IE <sup>(1)</sup>	C2IE	C1IE	_	TMR2IE <sup>(1)</sup>	TMR1IE	-000 0-00	-000 0-00
PIR1	—	ADIF <sup>(1)</sup>	CCP1IF <sup>(1)</sup>	C2IF	C1IF	_	TMR2IF <sup>(1)</sup>	TMR1IF	-000 0-00	-000 0-00
TMR1H	1H Holding Register for the Most Significant Byte of the 16-bit TMR1 Register									uuuu uuuu
TMR1L	Holding Register for the Least Significant Byte of the 16-bit TMR1 Register									uuuu uuuu
T1CON	T1GINV	TMR1GE	T1CKPS1	T1CKPS0	T1OSCEN	T1SYNC	TMR1CS	TMR10N	0000 0000	uuuu uuuu

**Legend:** x = unknown, u = unchanged, - = unimplemented, read as '0'. Shaded cells are not used by the Timer1 module.

Note 1: PIC16F616/16HV616 only.

#### 7.0 TIMER2 MODULE (PIC16F616/16HV616 ONLY)

The Timer2 module is an 8-bit timer with the following features:

- 8-bit timer register (TMR2)
- 8-bit period register (PR2)
- Interrupt on TMR2 match with PR2
- Software programmable prescaler (1:1, 1:4, 1:16)
- Software programmable postscaler (1:1 to 1:16)

See Figure 7-1 for a block diagram of Timer2.

#### 7.1 Timer2 Operation

The clock input to the Timer2 module is the system instruction clock (Fosc/4). The clock is fed into the Timer2 prescaler, which has prescale options of 1:1, 1:4 or 1:16. The output of the prescaler is then used to increment the TMR2 register.

The values of TMR2 and PR2 are constantly compared to determine when they match. TMR2 will increment from 00h until it matches the value in PR2. When a match occurs, two things happen:

- TMR2 is reset to 00h on the next increment cycle.
- The Timer2 postscaler is incremented

The match output of the Timer2/PR2 comparator is then fed into the Timer2 postscaler. The postscaler has postscale options of 1:1 to 1:16 inclusive. The output of the Timer2 postscaler is used to set the TMR2IF interrupt flag bit in the PIR1 register.

FIGURE 7-1:	<b>TIMER2 BLOCK DIAGRAM</b>

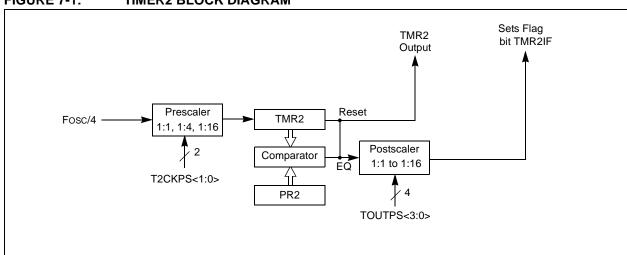
The TMR2 and PR2 registers are both fully readable and writable. On any Reset, the TMR2 register is set to 00h and the PR2 register is set to FFh.

Timer2 is turned on by setting the TMR2ON bit in the T2CON register to a '1'. Timer2 is turned off by setting the TMR2ON bit to a '0'.

The Timer2 prescaler is controlled by the T2CKPS bits in the T2CON register. The Timer2 postscaler is controlled by the TOUTPS bits in the T2CON register. The prescaler and postscaler counters are cleared when:

- A write to TMR2 occurs.
- A write to T2CON occurs.
- Any device Reset occurs (Power-on Reset, MCLR Reset, Watchdog Timer Reset, or Brown-out Reset).

Note: TMR2 is not cleared when T2CON is written.



U-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0						
_	TOUTPS3	TOUTPS2	TOUTPS1	TOUTPS0	TMR2ON	T2CKPS1	T2CKPS0						
oit 7	·					·	bit						
Legend:													
R = Readable	bit	W = Writable	bit	U = Unimplen	nented bit, rea	d as '0'							
-n = Value at I	POR	'1' = Bit is set		'0' = Bit is clea	ared	x = Bit is unki	nown						
bit 7	Unimplemen	ted: Read as '	0'										
bit 6-3	TOUTPS<3:0	)>: Timer2 Out	out Postscaler	Select bits									
	0000 = 1:1 P	ostscaler											
	0001 = 1:2 P												
	0010 = 1:3 P												
	0011 = 1:4 P												
		0100 = 1:5 Postscaler 0101 = 1:6 Postscaler											
		0101 = 1.7 Postscaler											
	0110 = 1.7 P												
	1000 = 1.9 P												
	1001 = 1:10												
	1010 = 1:11 I	Postscaler											
	1011 <b>= 1:12</b>	Postscaler											
	1100 = 1:13 Postscaler												
	1101 = 1:14												
	1110 = 1:15 Postscaler												
	1111 = 1:16												
bit 2	TMR2ON: Tir												
	1 = Timer2 is												
	0 = Timer2 is	s off											
bit 1-0	T2CKPS<1:0	>: Timer2 Cloc	k Prescale Se	lect bits									
	00 = Presca	ler is 1											
	01 = Presca												
	1x = Presca	ler is 16											
			· , , , , , , , , , , , , , , , , , , ,	DO DECICTE									

#### REGISTER 7-1: T2CON: TIMER2 CONTROL REGISTER

#### TABLE 7-1: SUMMARY OF ASSOCIATED TIMER2 REGISTERS

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Value on POR, BOR	Value on all other Resets
INTCON	GIE	PEIE	T0IE	INTE	RAIE	T0IF	INTF	RAIF	0000 0000	0000 0000
PIE1	_	ADIE <sup>(1)</sup>	CCP1IE <sup>(1)</sup>	C2IE	C1IE	_	TMR2IE <sup>(1)</sup>	TMR1IE	-000 0-00	-000 0-00
PIR1	_	ADIF <sup>(1)</sup>	CCP1IF <sup>(1)</sup>	C2IF	C1IF	_	TMR2IF <sup>(1)</sup>	TMR1IF	-000 0-00	-000 0-00
PR2 <sup>(1)</sup>	Timer2 M	Iodule Perio	d Register						1111 1111	1111 1111
TMR2 <sup>(1)</sup>	Holding	Register for t	0000 0000	0000 0000						
T2CON <sup>(1)</sup>	- TOUTPS3 TOUTPS2 TOUTPS1 TOUTPS0 TMR2ON T2CKPS1 T2CKPS0							-000 0000	-000 0000	
Lonondi									<u> </u>	

**Legend:** x = unknown, u = unchanged, - = unimplemented read as '0'. Shaded cells are not used for Timer2 module. **Note 1:** PIC16F616/16HV616 only.

#### 8.0 COMPARATOR MODULE

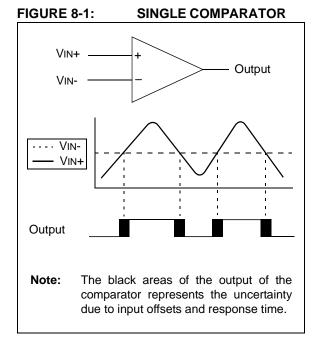
Comparators are used to interface analog circuits to a digital circuit by comparing two analog voltages and providing a digital indication of their relative magnitudes. The comparators are very useful mixed signal building blocks because they provide analog functionality independent of the device. The Analog Comparator module includes the following features:

- Independent comparator control
- Programmable input selection
- · Comparator output is available internally/externally
- Programmable output polarity
- Interrupt-on-change
- · Wake-up from Sleep
- PWM shutdown
- Timer1 gate (count enable)
- Output synchronization to Timer1 clock input
- SR Latch
- Programmable and fixed voltage reference
- User-enable Comparator Hysteresis

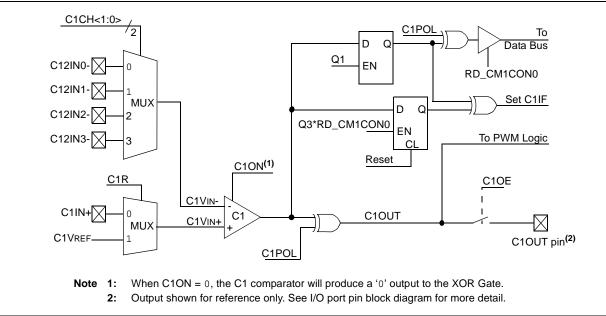
Note: Only Comparator C2 can be linked to Timer1.

#### 8.1 Comparator Overview

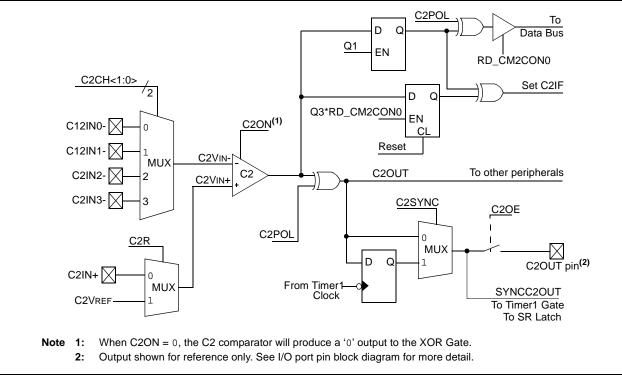
A single comparator is shown in Figure 8-1 along with the relationship between the analog input levels and the digital output. When the analog voltage at VIN+ is less than the analog voltage at VIN-, the output of the comparator is a digital low level. When the analog voltage at VIN+ is greater than the analog voltage at VIN-, the output of the comparator is a digital high level.











#### 8.2 Comparator Control

Each comparator has a separate control and Configuration register: CM1CON0 for Comparator C1 and CM2CON0 for Comparator C2. In addition, Comparator C2 has a second control register, CM2CON1, for controlling the interaction with Timer1 and simultaneous reading of both comparator outputs.

The CM1CON0 and CM2CON0 registers (see Registers 8-1 and 8-2, respectively) contain the control and Status bits for the following:

- Enable
- Input selection
- Reference selection
- Output selection
- Output polarity

#### 8.2.1 COMPARATOR ENABLE

Setting the CxON bit of the CMxCON0 register enables the comparator for operation. Clearing the CxON bit disables the comparator for minimum current consumption.

#### 8.2.2 COMPARATOR INPUT SELECTION

The CxCH<1:0> bits of the CMxCON0 register direct one of four analog input pins to the comparator inverting input.

Note:	To use CxIN+ and CxIN- pins as analog
	inputs, the appropriate bits must be set in
	the ANSEL register and the corresponding
	TRIS bits must also be set to disable the
	output drivers.

### 8.2.3 COMPARATOR REFERENCE SELECTION

Setting the CxR bit of the CMxCON0 register directs an internal voltage reference or an analog input pin to the non-inverting input of the comparator. See **Section 8.11 "Comparator Voltage Reference"** for more information on the internal voltage reference module.

#### 8.2.4 COMPARATOR OUTPUT SELECTION

The output of the comparator can be monitored by reading either the CxOUT bit of the CMxCON0 register or the MCxOUT bit of the CM2CON1 register. In order to make the output available for an external connection, the following conditions must be true:

- CxOE bit of the CMxCON0 register must be set
- · Corresponding TRIS bit must be cleared
- CxON bit of the CMxCON0 register must be set.
  - Note 1: The CxOE bit overrides the PORT data latch. Setting the CxON has no impact on the port override.
    - 2: The internal output of the comparator is latched with each instruction cycle. Unless otherwise specified, external outputs are not latched.

#### 8.2.5 COMPARATOR OUTPUT POLARITY

Inverting the output of the comparator is functionally equivalent to swapping the comparator inputs. The polarity of the comparator output can be inverted by setting the CxPOL bit of the CMxCON0 register. Clearing the CxPOL bit results in a non-inverted output.

Table 8-1 shows the output state versus input conditions, including polarity control.

#### TABLE 8-1: COMPARATOR OUTPUT STATE VS. INPUT CONDITIONS

Input Condition	CxPOL	CxOUT		
CxVIN - > CxVIN +	0	0		
CxVIN- < CxVIN+	0	1		
CxVIN - > CxVIN +	1	1		
CxVIN- < CxVIN+	1	0		

#### 8.3 Comparator Response Time

The comparator output is indeterminate for a period of time after the change of an input source or the selection of a new reference voltage. This period is referred to as the response time. The response time of the comparator differs from the settling time of the voltage reference. Therefore, both of these times must be considered when determining the total response time to a comparator input change. See the Comparator and Voltage Reference Specifications in **Section 15.0 "Electrical Specifications"** for more details.

#### 8.4 Comparator Interrupt Operation

The comparator interrupt flag can be set whenever there is a change in the output value of the comparator. Changes are recognized by means of a mismatch circuit which consists of two latches and an exclusive-or gate (see Figure 8-2 and Figure 8-3). One latch is updated with the comparator output level when the CMxCON0 register is read. This latch retains the value until the next read of the CMxCON0 register or the occurrence of a Reset. The other latch of the mismatch circuit is updated on every Q1 system clock. A mismatch condition will occur when a comparator output change is clocked through the second latch on the Q1 clock cycle. At this point the two mismatch latches have opposite output levels which is detected by the exclusive-or gate and fed to the interrupt circuitry. The mismatch condition persists until either the CMxCON0 register is read or the comparator output returns to the previous state.

- Note 1: A write operation to the CMxCON0 register will also clear the mismatch condition because all writes include a read operation at the beginning of the write cycle.
  - **2:** Comparator interrupts will operate correctly regardless of the state of CxOE.

The comparator interrupt is set by the mismatch edge and not the mismatch level. This means that the interrupt flag can be reset without the additional step of reading or writing the CMxCON0 register to clear the mismatch registers. When the mismatch registers are cleared, an interrupt will occur upon the comparator's return to the previous state, otherwise no interrupt will be generated.

Software will need to maintain information about the status of the comparator output, as read from the CMxCON0 register, or CM2CON1 register, to determine the actual change that has occurred.

The CxIF bit of the PIR1 register is the comparator interrupt flag. This bit must be reset in software by clearing it to '0'. Since it is also possible to write a '1' to this register, an interrupt can be generated.

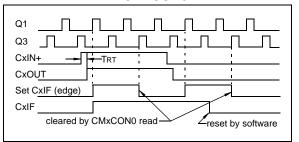
The CxIE bit of the PIE1 register and the PEIE and GIE bits of the INTCON register must all be set to enable comparator interrupts. If any of these bits are cleared, the interrupt is not enabled, although the CxIF bit of the PIR1 register will still be set if an interrupt condition occurs.

# FIGURE 8-4: COMPARATOR INTERRUPT TIMING W/O CMxCON0 READ

### FIGURE 8-5:

COMPARATOR INTERRUPT TIMING WITH CMxCON0 READ

reset by software



- Note 1: If a change in the CMxCON0 register (CxOUT) should occur when a read operation is being executed (start of the Q2 cycle), then the CxIF of the PIR1 register interrupt flag may not get set.
  - 2: When either comparator is first enabled, bias circuitry in the comparator module may cause an invalid output from the comparator until the bias circuitry is stable. Allow about 1  $\mu$ s for bias settling then clear the mismatch condition and interrupt flags before enabling comparator interrupts.

#### 8.5 Operation During Sleep

The comparator, if enabled before entering Sleep mode, remains active during Sleep. The additional current consumed by the comparator is shown separately in **Section 15.0** "**Electrical Specifications**". If the comparator is not used to wake the device, power consumption can be minimized while in Sleep mode by turning off the comparator. Each comparator is turned off by clearing the CxON bit of the CMxCON0 register.

A change to the comparator output can wake-up the device from Sleep. To enable the comparator to wake the device from Sleep, the CxIE bit of the PIE1 register and the PEIE bit of the INTCON register must be set. The instruction following the Sleep instruction always executes following a wake from Sleep. If the GIE bit of the INTCON register is also set, the device will then execute the interrupt service routine.

#### 8.6 Effects of a Reset

A device Reset forces the CMxCON0 and CM2CON1 registers to their Reset states. This forces both comparators and the voltage references to their OFF states.

R/W-0	R-0	R/W-0	R/W-0	U-0	R/W-0	R/W-0	R/W-0
C10N	C1OUT	C10E	C1POL	_	C1R	C1CH1	C1CH0
bit 7							bit
Legend:							
R = Readab	le bit	W = Writable	bit	U = Unimplei	mented bit, rea	ad as '0'	
-n = Value at POR		(1) = Bit is set		'0' = Bit is cle		x = Bit is unk	nown
bit 7	C1ON: Com	parator C1 Ena	ble bit				
		ator C1 is enab ator C1 is disab					
bit 6	C1OUT: Con	nparator C1 Ou	ıtput bit				
	C1OUT = 0 $C1OUT = 1$ $If C1POL = 0$ $C1OUT = 1$	<u>- (inverted pola</u> when C1VIN+ > when C1VIN+ < (non-inverted when C1VIN+ > when C1VIN+ >	C1VIN- C1VIN- polarity): C1VIN-				
bit 5		parator C1 Out	-				
	1 = C1OUT i	s present on th s internal only		1)			
bit 4	C1POL: Con	nparator C1 Ou	utput Polarity S	Select bit			
		ogic is inverted					
bit 3	Unimplemer	nted: Read as	'0'				
bit 2	C1R: Compa	arator C1 Refer	ence Select bi	t (non-inverting	input)		
		connects to C1 connects to C1					
bit 1-0	C1CH<1:0>:	Comparator C	1 Channel Sel	ect bit			
		- pin of C1 con - pin of C1 con	nects to C1VIN	٨-			

#### REGISTER 8-1: CM1CON0: COMPARATOR 1 CONTROL REGISTER 0

**Note 1:** Comparator output requires the following three conditions: C1OE = 1, C1ON = 1 and corresponding port TRIS bit = 0.

R/W-0	R-0	R/W-0	R/W-0	U-0	R/W-0	R/W-0	R/W-0
C2ON	C2OUT	C2OE	C2POL	—	C2R	C2CH1	C2CH0
bit 7							bit C
Legend:							
R = Readab	ole bit	W = Writable	bit	U = Unimple	emented bit, rea	ad as '0'	
-n = Value a	It POR	'1' = Bit is set		ʻ0' = Bit is cl		x = Bit is unk	nown
bit 7	C2ON: Com	parator C2 Ena	ble bit				
	1 = Compara	ator C2 is enabl ator C2 is disabl	ed				
bit 6	C2OUT: Cor	mparator C2 Ou	tput bit				
	C2OUT = 0 C2OUT = 1 <u>If C2POL = 0</u> C2OUT = 1	1_(inverted polar when C2VIN+ > when C2VIN+ < 0_(non-inverted when C2VIN+ > when C2VIN+ <	C2VIN- C2VIN- polarity): C2VIN-				
bit 5	C2OE: Com	parator C2 Outp	out Enable bit				
		is present on C2 is internal only	2OUT pin <sup>(1)</sup>				
bit 4	C2POL: Cor	mparator C2 Ou	tput Polarity S	Select bit			
		logic is inverted logic is not inve					
bit 3	Unimpleme	nted: Read as '	0'				
bit 2	C2R: Compa	arator C2 Refere	ence Select bi	ts (non-invertir	ng input)		
	-	connects to C2 connects to C2					
bit 1-0	C2CH<1:0>	Comparator C	2 Channel Sel	lect bits			
		pin of C2 conne					
		pin of C2 conne pin of C2 conne					
		pin of C2 conne					
Note 1: C	Comparator outp	ut requires the f	ollowing three	conditions: C2	20E = 1, C20N	N = 1 and corres	sponding port

#### REGISTER 8-2: CM2CON0: COMPARATOR 2 CONTROL REGISTER 0

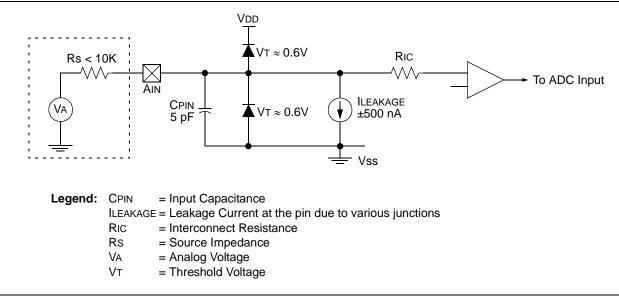
**Note 1:** Comparator output requires the following three conditions: C2OE = 1, C2ON = 1 and corresponding port TRIS bit = 0.

#### 8.7 Comparator Analog Input Connection Considerations

A simplified circuit for an analog input is shown in Figure 8-6. Since the analog input pins share their connection with a digital input, they have reverse biased ESD protection diodes to VDD and Vss. The analog input, therefore, must be between Vss and VDD. If the input voltage deviates from this range by more than 0.6V in either direction, one of the diodes is forward biased and a latch-up may occur.

A maximum source impedance of  $10 \text{ k}\Omega$  is recommended for the analog sources. Also, any external component connected to an analog input pin, such as a capacitor or a Zener diode, should have very little leakage current to minimize inaccuracies introduced.

- Note 1: When reading a PORT register, all pins configured as analog inputs will read as a '0'. Pins configured as digital inputs will convert as an analog input, according to the input specification.
  - 2: Analog levels on any pin defined as a digital input, may cause the input buffer to consume more current than is specified.



#### FIGURE 8-6: ANALOG INPUT MODEL

#### 8.8 Additional Comparator Features

There are three additional comparator features:

- Timer1 count enable (gate)
- Synchronizing output with Timer1
- Simultaneous read of comparator outputs

#### 8.8.1 COMPARATOR C2 GATING TIMER1

This feature can be used to time the duration or interval of analog events. Clearing the T1GSS bit of the CM2CON1 register will enable Timer1 to increment based on the output of Comparator C2. This requires that Timer1 is on and gating is enabled. See **Section 6.0 "Timer1 Module with Gate Control"** for details.

It is recommended to synchronize the comparator with Timer1 by setting the C2SYNC bit when the comparator is used as the Timer1 gate source. This ensures Timer1 does not miss an increment if the comparator changes during an increment.

#### 8.8.2 SYNCHRONIZING COMPARATOR C2 OUTPUT TO TIMER1

The Comparator C2 output can be synchronized with Timer1 by setting the C2SYNC bit of the CM2CON1 register. When enabled, the C2 output is latched on the falling edge of the Timer1 clock source. If a prescaler is used with Timer1, the comparator output is latched after the prescaling function. To prevent a race condition, the comparator output is latched on the falling edge of the Timer1 clock source and Timer1 increments on the rising edge of its clock source. See the Comparator Block Diagram (Figure 8-3) and the Timer1 Block Diagram (Figure 6-1) for more information.

#### 8.8.3 SIMULTANEOUS COMPARATOR OUTPUT READ

The MC1OUT and MC2OUT bits of the CM2CON1 register are mirror copies of both comparator outputs. The ability to read both outputs simultaneously from a single register eliminates the timing skew of reading separate registers.

Note 1: Obtaining the status of C1OUT or C2OUT by reading CM2CON1 does not affect the comparator interrupt mismatch registers.

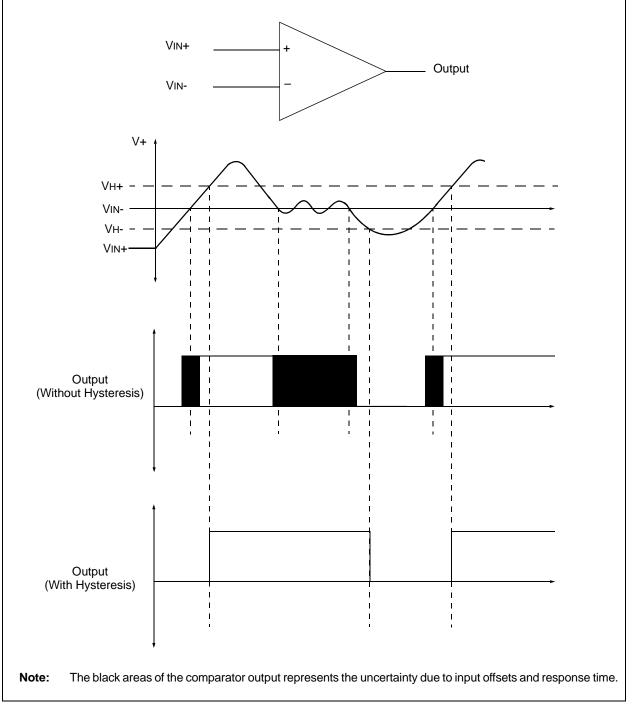
#### REGISTER 8-3: CM2CON1: COMPARATOR 2 CONTROL REGISTER 1

R-0	R-0	U-0	R/W-0	R/W-0	R/W-0	R/W-1	R/W-0
MC10UT	MC2OUT	—	T1ACS	C1HYS	C2HYS	T1GSS	C2SYNC
bit 7					• •		bit (
Legend:							
R = Readable	bit	W = Writable bit		U = Unimpleme	ented bit, read as	s 'O'	
-n = Value at F	POR	'1' = Bit is set		'0' = Bit is clear	red	x = Bit is unkno	own
bit 7	MC1OUT: Mirro	or Copy of C1OU	T bit				
bit 6		or Copy of C2OU					
bit 5	Unimplemente	19					
bit 4	1 = Timer1 clo	1 Alternate Clock ck source is the s ck source is the ir	ystem clock (Fo	,			
bit 3	1 = Comparato	arator C1 Hystere or C1 Hysteresis e or C1 Hysteresis c	enabled				
bit 2	1 = Comparato	arator C2 Hystere or C2 Hysteresis e or C2 Hysteresis c	enabled				
bit 1	1 = Timer1 gate	1 Gate Source Se e source is T1G e source is SYNC					
bit 0	1 = C2 Output	nparator C2 Outpu is synchronous to is asynchronous					

#### 8.9 Comparator Hysteresis

Each comparator has built-in hysteresis that is user enabled by setting the C1HYS or C2HYS bits of the CM2CON1 register. The hysteresis feature can help filter noise and reduce multiple comparator output transitions when the output is changing state. Figure 8-9 shows the relationship between the analog input levels and digital output of a comparator with and without hysteresis. The output of the comparator changes from a low state to a high state only when the analog voltage at VIN+ rises above the upper hysteresis threshold (VH+). The output of the comparator changes from a high state to a low state only when the analog voltage at VIN+ falls below the lower hysteresis threshold (VH-).

FIGURE 8-7: COMPARATOR HYSTERESIS



#### **TABLE 8-2:** SUMMARY OF REGISTERS ASSOCIATED WITH THE COMPARATOR AND **VOLTAGE REFERENCE MODULES**

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Value on POR, BOR	Value on all other Resets
ANSEL	ANS7	ANS6	ANS5	ANS4	ANS3 <sup>(1)</sup>	ANS2 <sup>(1)</sup>	ANS1	ANS0	1111 1111	1111 1111
CM1CON0	C1ON	C1OUT	C1OE	C1POL	C1SP	C1R	C1CH1	C1CH0	0000 0000	0000 0000
CM2CON0	C2ON	C2OUT	C2OE	C2POL	C2SP	C2R	C2CH1	C2CH0	0000 0000	0000 0000
CM2CON1	MC10UT	MC2OUT	_	T1ACS	C1HYS	C2HYS	T1GSS	C2SYNC	00-0 0010	00-0 0010
INTCON	GIE	PEIE	TOIE	INTE	RAIE	T0IF	INTF	RAIF	0000 000x	0000 000x
PIE1	_	ADIE <sup>(1)</sup>	CCP1IE <sup>(1)</sup>	C2IE	C1IE	_	TMR2IE <sup>(1)</sup>	TMR1IE	-000 0-00	-000 0-00
PIR1	—	ADIF <sup>(1)</sup>	CCP1IF <sup>(1)</sup>	C2IF	C1IF	_	TMR2IF <sup>(1)</sup>	TMR1IF	-000 0-00	-000 0-00
PORTA	—	_	RA5	RA4	RA3	RA2	RA1	RA0	x0 x000	x0 x000
PORTC	—	_	RC5	RC4	RC3	RC2	RC1	RC0	xx 00xx	uu 00uu
SRCON0	SR1	SR0	C1SEN	C2REN	PULSS	PULSR	_	SRCLKEN	0000 00-0	0000 00-0
SRCON1	SRCS1	SRCS0	_	_	_	—	_	_	00	00
TRISA	_	_	TRISA5	TRISA4	TRISA3	TRISA2	TRISA1	TRISA0	11 1111	11 1111
TRISC			TRISC5	TRISC4	TRISC3	TRISC2	TRISC1	TRISC0	1111 1111	1111 1111
VRCON	C1VREN	C2VREN	VRR	FVREN	VR3	VR2	VR1	VR0	0000 0000	0000 0000

 Legend:
 x = unknown, u = unchanged, - = unimplemented, read as '0'. Shaded cells are not used for comparator.

 Note
 1:
 PIC16F616/16HV616 only.

#### 8.10 Comparator SR Latch

The SR latch module provides additional control of the comparator outputs. The module consists of a single SR latch and output multiplexers. The SR latch can be set, reset or toggled by the comparator outputs. The SR latch may also be set or reset, independent of comparator output, by control bits in the SRCON0 control register. The SR latch output multiplexers select whether the latch outputs or the comparator outputs are directed to the I/O port logic for eventual output to a pin.

The SR latch also has a variable clock, which is connected to the set input of the latch. The SRCLKEN bit of SRCON0 enables the SR latch set clock. The clock will periodically pulse the set input of the latch. Control over the frequency of the SR latch set clock is provided by the SRCS<1:0> bits of SRCON1 register.

#### 8.10.1 LATCH OPERATION

The latch is a Set-Reset latch that does not depend on a clock source. Each of the Set and Reset inputs are active-high. Each latch input is connected to a comparator output and a software controlled pulse generator. The latch can be set by C1OUT or the PULSS bit of the SRCON0 register. The latch can be reset by C2OUT or the PULSR bit of the SRCON0 register. The latch is reset-dominant, therefore, if both Set and Reset

inputs are high the latch will go to the Reset state. Both the PULSS and PULSR bits are self resetting which means that a single write to either of the bits is all that is necessary to complete a latch Set or Reset operation.

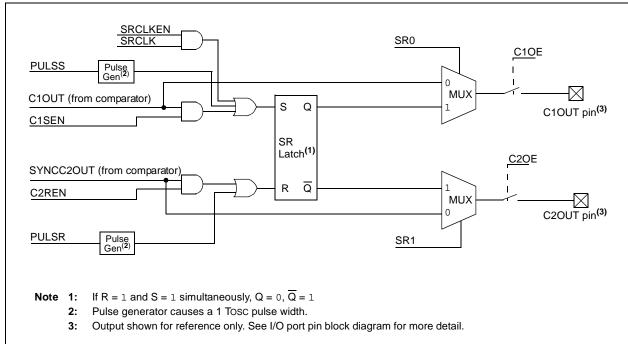
#### 8.10.2 LATCH OUTPUT

The SR<1:0> bits of the SRCON0 register control the latch output multiplexers and determine four possible output configurations. In these four configurations, the CxOUT I/O port logic is connected to:

- C1OUT and C2OUT
- C1OUT and SR latch Q
- C2OUT and SR latch Q
- SR latch Q and Q

After any Reset, the default output configuration is the unlatched C1OUT and C2OUT mode. This maintains compatibility with devices that do not have the SR latch feature.

The applicable TRIS bits of the corresponding ports must be cleared to enable the port pin output drivers. Additionally, the CxOE comparator output enable bits of the CMxCON0 registers must be set in order to make the comparator or latch outputs available on the output pins. The latch configuration enable states are completely independent of the enable states for the comparators.



#### FIGURE 8-8: SR LATCH SIMPLIFIED BLOCK DIAGRAM

R/W-0	R/W-0	R/W-0	R/W-0	R/S-0	R/S-0	U-0	R/W-0
SR1 <sup>(2)</sup>	SR0 <sup>(2)</sup>	C1SEN	C2REN	PULSS	PULSR	—	SRCLKEN
bit 7							bit 0
Legend:				S = Bit is set o	nlv -		
R = Readable t	oit	W = Writable bi	t		ented bit, read a	s 'O'	
-n = Value at P		'1' = Bit is set	·	'0' = Bit is clea		x = Bit is unkr	
	OR				ieu		101111
bit 7	1 = C2OUT	h Configuration bi pin is the latch $\overline{Q}$ pin is the C2 con	output				
bit 6	1 = C1OUT	n Configuration bi pin is the latch Q pin is the C1 Cor	output				
bit 5		et Enable bit arator output sets arator output has		latch			
bit 4	1 = C2  comparation	eset Enable bit arator output rese arator output has		latch			
bit 3	1 = Triggers	the SET Input of pulse generator to trigger pulse gen	set SR latch. E		reset by hardwa	are.	
bit 2	1 = Triggers	e the Reset Input o pulse generator to trigger pulse gen	reset SR latch		ely reset by hard	ware.	
bit 1	Unimplement	ed: Read as '0'					
bit 0	1 = Set input	R Latch Set Clock of SR latch is pul of SR latch is not	sed with SRCL				

#### REGISTER 8-4: SRCON0: SR LATCH CONTROL 0 REGISTER

the pin), regardless of the SR latch operation.2: To enable an SR Latch output to the pin, the appropriate CxOE, and TRIS bits must be properly configured.

#### REGISTER 8-5: SRCON1: SR LATCH CONTROL 1 REGISTER

R/W-0	R/W-0	U-0	U-0	U-0	U-0	U-0	U-0
SRCS1	SRCS0	—	—	—	—	—	—
bit 7							bit 0

Legend:		S = Bit is set only -	S = Bit is set only -		
R = Readable bit	W = Writable bit	U = Unimplemented bit, r	U = Unimplemented bit, read as '0'		
-n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown		

bit 7-6	SRCS<1:0>: SR Latch Clock Prescale bits			
	00 = Fosc/16			
	01 = Fosc/32			
	10 = Fosc/64			
	11 = Fosc/128			

bit 5-0 Unimplemented: Read as '0'

#### 8.11 Comparator Voltage Reference

The comparator voltage reference module provides an internally generated voltage reference for the comparators. The following features are available:

- Independent from Comparator operation
- Two 16-level voltage ranges
- Output clamped to Vss
- Ratiometric with VDD
- Fixed Reference (0.6V)

The VRCON register (Register 8-6) controls the voltage reference module shown in Figure 8-9.

#### 8.11.1 INDEPENDENT OPERATION

The comparator voltage reference is independent of the comparator configuration. Setting the FVREN bit of the VRCON register will enable the voltage reference.

#### 8.11.2 OUTPUT VOLTAGE SELECTION

The CVREF voltage reference has 2 ranges with 16 voltage levels in each range. Range selection is controlled by the VRR bit of the VRCON register. The 16 levels are set with the VR<3:0> bits of the VRCON register.

The CVREF output voltage is determined by the following equations:

#### EQUATION 8-1: CVREF OUTPUT VOLTAGE

VRR = 1 (low range):  $CVREF = (VR < 3:0 > /24) \times VDD$  VRR = 0 (high range):  $CVREF = (VDD/4) + (VR < 3:0 > \times VDD/32)$ 

The full range of Vss to VDD cannot be realized due to the construction of the module. See Figure 8-9.

#### 8.11.3 OUTPUT CLAMPED TO Vss

The fixed voltage reference output voltage can be set to Vss with no power consumption by clearing the FVREN bit of the VRCON register (FVREN = 0). This allows the comparator to detect a zero-crossing while not consuming additional module current.

#### 8.11.4 OUTPUT RATIOMETRIC TO VDD

The comparator voltage reference is VDD derived and therefore, the CVREF output changes with fluctuations in VDD. The tested absolute accuracy of the Comparator Voltage Reference can be found in **Section 15.0 "Electrical Specifications"**.

#### 8.11.5 FIXED VOLTAGE REFERENCE

The fixed voltage reference is independent of VDD, with a nominal output voltage of 0.6V. This reference can be enabled by setting the FVREN bit of the VRCON register to '1'. This reference is always enabled when the HFINTOSC oscillator is active.

#### 8.11.6 FIXED VOLTAGE REFERENCE STABILIZATION PERIOD

When the fixed voltage reference module is enabled, it will require some time for the reference and its amplifier circuits to stabilize. The user program must include a small delay routine to allow the module to settle. See the electrical specifications section for the minimum delay requirement.

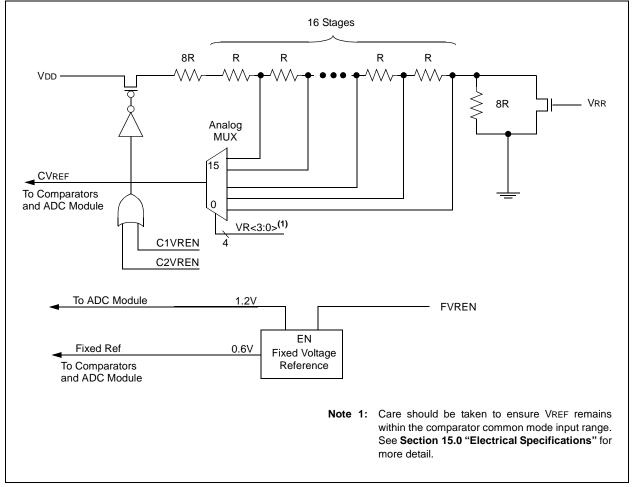
#### 8.11.7 VOLTAGE REFERENCE SELECTION

Multiplexers on the output of the voltage reference module enable selection of either the CVREF or fixed voltage reference for use by the comparators.

Setting the C1VREN bit of the VRCON register enables current to flow in the CVREF voltage divider and selects the CVREF voltage for use by C1. Clearing the C1VREN bit selects the fixed voltage for use by C1.

Setting the C2VREN bit of the VRCON register enables current to flow in the CVREF voltage divider and selects the CVREF voltage for use by C2. Clearing the C2VREN bit selects the fixed voltage for use by C2.

When both the C1VREN and C2VREN bits are cleared, current flow in the CVREF voltage divider is disabled minimizing the power drain of the voltage reference peripheral.



#### FIGURE 8-9: COMPARATOR VOLTAGE REFERENCE BLOCK DIAGRAM

R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0			
C1VREN	C2VREN	VRR	FVREN	VR3	VR2	VR1	VR0			
bit 7							bit 0			
Legend:										
R = Readable bit		W = Writable bit		U = Unimplemented bit, read as '0'						
-n = Value at POR		'1' = Bit is set		'0' = Bit is cleared		x = Bit is unknown				
bit 7	C1VREN: Comparator 1 Voltage Reference Enable bit									
	1 = CVREF circuit powered on and routed to C1VREF input of Comparator C1									
	0 = 0.6 Volt constant reference routed to C1VREF input of Comparator C1									
bit 6	C2VREN: Comparator 2 Voltage Reference Enable bit									
	1 = CVREF circuit powered on and routed to C2VREF input of Comparator C2									
	0 = 0.6 Volt constant reference routed to C2VREF input of Comparator C2									
bit 5	VRR: CVREF Range Selection bit									
	1 = Low range									
	0 = High rang		(							
bit 4	FVREN: Fixed Voltage Reference (0.6V) Enable bit									
	1 = Enabled 0 = Disabled									
bit 3-0	VR<3:0>: Comparator Voltage Reference CVREF Value Selection bits ( $0 \le VR < 3:0 > \le 15$ )									
	<u>When VRR = 1</u> : CVREF = (VR<3:0>/24) * VDD									
	<u>When VRR = 0</u> : CVREF = VDD/4 + (VR<3:0>/32) * VDD									

#### REGISTER 8-6: VRCON: VOLTAGE REFERENCE CONTROL REGISTER

## 9.0 ANALOG-TO-DIGITAL CONVERTER (ADC) MODULE (PIC16F616/16HV616 ONLY)

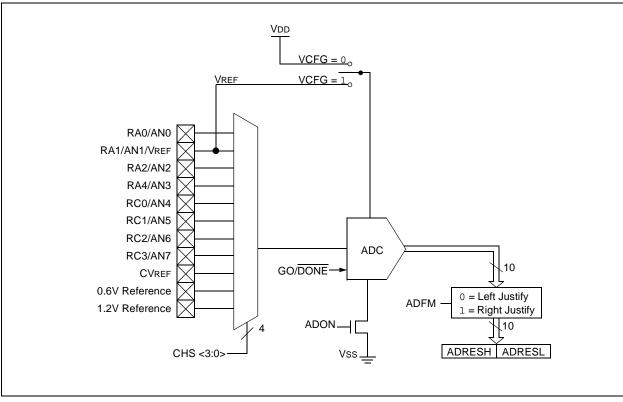
The Analog-to-Digital Converter (ADC) allows conversion of an analog input signal to a 10-bit binary representation of that signal. This device uses analog inputs, which are multiplexed into a single sample and hold circuit. The output of the sample and hold is connected to the input of the converter. The converter generates a 10-bit binary result via successive approximation and stores the conversion result into the ADC result registers (ADRESL and ADRESH).

The ADC voltage reference is software selectable to either VDD or a voltage applied to the external reference pins.

The ADC can generate an interrupt upon completion of a conversion. This interrupt can be used to wake-up the device from Sleep.

Figure 9-1 shows the block diagram of the ADC.

### FIGURE 9-1: ADC BLOCK DIAGRAM



Note: The ADRESL and ADRESH registers are read-only.

## 9.1 ADC Configuration

When configuring and using the ADC, the following functions must be considered:

- Port configuration
- · Channel selection
- ADC voltage reference selection
- ADC conversion clock source
- Interrupt control
- Results formatting

#### 9.1.1 PORT CONFIGURATION

The ADC can be used to convert both analog and digital signals. When converting analog signals, the I/O pin should be configured for analog by setting the associated TRIS and ANSEL bits. See the corresponding Port section for more information.

Note:	Analog voltages on any pin that is defined
	as a digital input may cause the input buf-
	fer to conduct excess current.

#### 9.1.2 CHANNEL SELECTION

The CHS bits of the ADCON0 register determine which channel is connected to the sample and hold circuit.

When changing channels, a delay is required before starting the next conversion. Refer to **Section 9.2 "ADC Operation"** for more information.

#### 9.1.3 ADC VOLTAGE REFERENCE

The VCFG bit of the ADCON0 register provides control of the positive voltage reference. The positive voltage reference can be either VDD or an external voltage source. The negative voltage reference is always connected to the ground reference.

### 9.1.4 CONVERSION CLOCK

The source of the conversion clock is software selectable via the ADCS bits of the ADCON1 register. There are seven possible clock options:

- Fosc/2
- Fosc/4
- Fosc/8
- Fosc/16
- Fosc/32
- Fosc/64
- FRC (dedicated internal oscillator)

The time to complete one bit conversion is defined as TAD. One full 10-bit conversion requires 11 TAD periods as shown in Figure 9-3.

For correct conversion, the appropriate TAD specification must be met. See A/D conversion requirements in **Section 15.0** "**Electrical Specifications**" for more information. Table 9-1 gives examples of appropriate ADC clock selections.

**Note:** Unless using the FRC, any changes in the system clock frequency will change the ADC clock frequency, which may adversely affect the ADC result.

ADC Clock	Period (TAD)		Device Frequency (Fosc)					
ADC Clock Source	ADCS<2:0>	20 MHz	8 MHz	4 MHz	1 MHz			
Fosc/2	000	100 ns <sup>(2)</sup>	250 ns <sup>(2)</sup>	500 ns <sup>(2)</sup>	2.0 μs			
Fosc/4	100	200 ns <sup>(2)</sup>	500 ns <sup>(2)</sup>	1.0 μs <b><sup>(2)</sup></b>	4.0 μs			
Fosc/8	001	400 ns <sup>(2)</sup>	1.0 μs <sup>(2)</sup>	2.0 μs	8.0 μs <b>(3)</b>			
Fosc/16	101	800 ns <sup>(2)</sup>	2.0 μs	4.0 μs	16.0 μs <b><sup>(3)</sup></b>			
Fosc/32	010	1.6 μs	4.0 μs	8.0 μs <sup>(3)</sup>	32.0 μs <b>(3)</b>			
Fosc/64	110	3.2 μs	8.0 μs <sup>(3)</sup>	16.0 μs <b>(3)</b>	64.0 μs <b>(3)</b>			
FRC	x11	2-6 μs <sup>(1,4)</sup>	2-6 μs <sup>(1,4)</sup>	2-6 μs <sup>(1,4)</sup>	2-6 μs <sup>(1,4)</sup>			

#### TABLE 9-1: ADC CLOCK PERIOD (TAD) Vs. DEVICE OPERATING FREQUENCIES (VDD $\geq$ 3.0V)

Legend: Shaded cells are outside of recommended range.

Note 1: The FRC source has a typical TAD time of 4  $\mu$ s for VDD > 3.0V.

2: These values violate the minimum required TAD time.

- **3:** For faster conversion times, the selection of another clock source is recommended.
- 4: When the device frequency is greater than 1 MHz, the FRC clock source is only recommended if the conversion will be performed during Sleep.

FIGURE 9	-2:	ANA	LOG-T	O-DIG	ITAL	CON	/ERSI			CLES				
		TAD TAD1 ↑ Conver ding Capa	b9 sion St	b8 arts	b7	b6	b5	b4	b3	b2	b1	TAD11 b0		
	l Set G	O/DONE	bit					GO bi ADIF	it is cle bit is s	ared, et,		0	] are loade analog ir	

### 9.1.5 INTERRUPTS

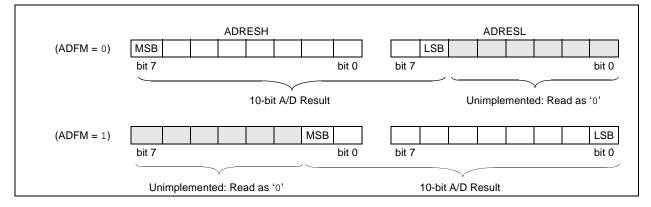
The ADC module allows for the ability to generate an interrupt upon completion of an analog-to-digital conversion. The ADC interrupt flag is the ADIF bit in the PIR1 register. The ADC interrupt enable is the ADIE bit in the PIE1 register. The ADIF bit must be cleared in software.

**Note:** The ADIF bit is set at the completion of every conversion, regardless of whether or not the ADC interrupt is enabled.

This interrupt can be generated while the device is operating or while in Sleep. If the device is in Sleep, the interrupt will wake-up the device. Upon waking from Sleep, the next instruction following the SLEEP instruction is always executed. If the user is attempting to wake-up from Sleep and resume in-line code execution, the global interrupt must be disabled. If the global interrupt is enabled, execution will switch to the interrupt service routine.

Please see **Section 9.1.5** "Interrupts" for more information.

#### FIGURE 9-3: 10-BIT A/D CONVERSION RESULT FORMAT



## 9.1.6 RESULT FORMATTING

The 10-bit A/D conversion result can be supplied in two formats, left justified or right justified. The ADFM bit of the ADCON0 register controls the output format.

Figure 9-4 shows the two output formats.

## 9.2 ADC Operation

#### 9.2.1 STARTING A CONVERSION

To enable the ADC module, the ADON bit of the ADCON0 register must be set to a '1'. Setting the GO/ DONE bit of the ADCON0 register to a '1' will start the analog-to-digital conversion.

Note:	The GO/DONE bit should not be set in the
	same instruction that turns on the ADC.
	Refer to Section 9.2.6 "A/D Conversion
	Procedure".

#### 9.2.2 COMPLETION OF A CONVERSION

When the conversion is complete, the ADC module will:

- Clear the GO/DONE bit
- Set the ADIF flag bit
- Update the ADRESH:ADRESL registers with new conversion result

#### 9.2.3 TERMINATING A CONVERSION

If a conversion must be terminated before completion, the GO/DONE bit can be cleared in software. The ADRESH:ADRESL registers will not be updated with the partially complete analog-to-digital conversion sample. Instead, the ADRESH:ADRESL register pair will retain the value of the previous conversion. Additionally, a 2 TAD delay is required before another acquisition can be initiated. Following this delay, an input acquisition is automatically started on the selected channel.

Note:	A device Reset forces all registers to their								
	Reset state. Thus, the ADC module is								
	turned off and any pending conversion is								
	terminated.								

#### 9.2.4 ADC OPERATION DURING SLEEP

The ADC module can operate during Sleep. This requires the ADC clock source to be set to the FRC option. When the FRC clock source is selected, the ADC waits one additional instruction before starting the conversion. This allows the SLEEP instruction to be executed, which can reduce system noise during the conversion. If the ADC interrupt is enabled, the device will wake-up from Sleep when the conversion completes. If the ADC interrupt is disabled, the ADC module is turned off after the conversion completes, although the ADON bit remains set.

When the ADC clock source is something other than FRC, a SLEEP instruction causes the present conversion to be aborted and the ADC module is turned off, although the ADON bit remains set.

#### 9.2.5 SPECIAL EVENT TRIGGER

The ECCP Special Event Trigger allows periodic ADC measurements without software intervention. When this trigger occurs, the GO/DONE bit is set by hardware and the Timer1 counter resets to zero.

Using the Special Event Trigger does not ensure proper ADC timing. It is the user's responsibility to ensure that the ADC timing requirements are met.

See Section 10.0 "Enhanced Capture/Compare/ PWM (With Auto-Shutdown and Dead Band) Module (PIC16F616/16HV616 Only)" for more information.

### 9.2.6 A/D CONVERSION PROCEDURE

This is an example procedure for using the ADC to perform an analog-to-digital conversion:

- 1. Configure Port:
  - Disable pin output driver (See TRIS register)
  - Configure pin as analog
- 2. Configure the ADC module:
  - Select ADC conversion clock
  - Configure voltage reference
  - Select ADC input channel
  - Select result format
  - Turn on ADC module
- 3. Configure ADC interrupt (optional):
  - Clear ADC interrupt flag
  - Enable ADC interrupt
  - Enable peripheral interrupt
  - Enable global interrupt<sup>(1)</sup>
- 4. Wait the required acquisition time<sup>(2)</sup>.
- 5. Start conversion by setting the GO/DONE bit.
- 6. Wait for ADC conversion to complete by one of the following:
  - Polling the GO/DONE bit
  - Waiting for the ADC interrupt (interrupts enabled)
- 7. Read ADC Result
- 8. Clear the ADC interrupt flag (required if interrupt is enabled).
  - **Note 1:** The global interrupt may be disabled if the user is attempting to wake-up from Sleep and resume in-line code execution.
    - 2: See Section 9.3 "A/D Acquisition Requirements".

#### EXAMPLE 9-1: A/D CONVERSION

;for poll ;and AN0 ;	ing, Vdd refe input. on start & po	gures the ADC erence, Frc clock lling for completion
, BANKSEL	ADCON1	;
MOVLW	B'01110000'	;ADC Frc clock
MOVWF	ADCON1	;
BANKSEL	TRISA	;
BSF	TRISA,0	;Set RA0 to input
BANKSEL	ANSEL	;
BSF	ANSEL,0	;Set RA0 to analog
BANKSEL	ADCON0	;
MOVLW	B'10000001'	;Right justify,
MOVWF	ADCON0	;Vdd Vref, AN0, On
CALL	SampleTime	;Acquisiton delay
BSF	ADCON0,GO	;Start conversion
BTFSC	ADCON0,GO	;Is conversion done?
GOTO	\$-1	;No, test again
BANKSEL	ADRESH	;
MOVF	ADRESH,W	;Read upper 2 bits
MOVWF	RESULTHI	;store in GPR space
BANKSEL	ADRESL	;
MOVF	ADRESL,W	;Read lower 8 bits
MOVWF	RESULTLO	;Store in GPR space

#### 9.2.7 ADC REGISTER DEFINITIONS

The following registers are used to control the operation of the ADC.

#### REGISTER 9-1: ADCON0: A/D CONTROL REGISTER 0

R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0
ADFM	VCFG	CHS3	CHS2	CHS1	CHS0	GO/DONE	ADON
oit 7							bit
Legend:						(-)	
R = Readabl		W = Writable bi	it	•	ented bit, read a		
-n = Value at	POR	'1' = Bit is set		'0' = Bit is clea	red	x = Bit is unkno	wn
bit 7	<b>ADFM:</b> A/D C 1 = Right justi 0 = Left justifio		Format Select	bit			
bit 6	<b>VCFG:</b> Voltag 1 = VREF pin 0 = VDD	e Reference bit					
bit 5-2	0000 = Char 0001 = Char 0010 = Char 0011 = Char 0100 = Char 0101 = Char 0110 = Char 0111 = Char 1000 = Rese 1001 = Rese 1010 = Rese 1011 = Rese 1100 = CVR 1101 = 0.6V 1110 = 1.2V	anel 01 (AN1) anel 02 (AN2) anel 03 (AN3) anel 04 (AN4) anel 05 (AN5) anel 05 (AN5) anel 06 (AN6) anel 07 (AN7) erved – do not use erved – do not use erved – do not use	ference <sup>(1)</sup>				
bit 1	1 = A/D conve This bit is	/D Conversion Sta ersion cycle in pro automatically clea ersion completed/r	gress. Setting t ared by hardwa				
bit 0	<b>ADON:</b> ADC I 1 = ADC is en 0 = ADC is dis		mes no operatii	ng current			

Note 1: When the CHS<3:0> bits change to select the 1.2V or 0.6V Fixed Voltage Reference, the reference output voltage will have a transient. If the Comparator module uses this VP6 reference voltage, the comparator output may momentarily change state due to the transient.

U-0	R/W-0	R/W-0	R/W-0	U-0	U-0	U-0	U-0		
_	ADCS2	ADCS1	ADCS0	—	—	_	—		
bit 7							bit		
Legend:									
R = Readable bit		W = Writable bi	t	U = Unimpleme	ented bit, read as	s '0'			
-n = Value at PO	R	'1' = Bit is set		'0' = Bit is clear	red	x = Bit is unkno	x = Bit is unknown		
bit 6-4	000 = Fosc/2 001 = Fosc/8 010 = Fosc/32	2 ock derived from a 6			00 kHz max)				
bit 3-0		+ ted: Read as '0'							

#### REGISTER 9-2: ADCON1: A/D CONTROL REGISTER 1

#### **REGISTER 9-3:** ADRESH: ADC RESULT REGISTER HIGH (ADRESH) ADFM = 0 (READ-ONLY)

R-x	R-x	R-x	R-x	R-x	R-x	R-x	R-x
ADRES9	ADRES8	ADRES7	ADRES6	ADRES5	ADRES4	ADRES3	ADRES2
bit 7							bit 0
Legend:							
R = Readable bit		W = Writable bi	t	U = Unimpleme	ented bit, read as	ʻ0'	

-n = Value at POR '1' = Bit is set '0' = Bit is cleared x = Bit is unknown

bit 7-0 ADRES<9:2>: ADC Result Register bits Upper 8 bits of 10-bit conversion result

#### **REGISTER 9-4:** ADRESL: ADC RESULT REGISTER LOW (ADRESL) ADFM = 0 (READ-ONLY)

R-x	R-x	U-0	U-0	U-0	U-0	U-0	U-0	
ADRES1	ADRES0	—	—	—	—	—	—	
bit 7							bit 0	
Legend:								
R = Readable b	it	W = Writable bi	t	U = Unimplemented bit, read as '0'				
-n = Value at PC	DR	'1' = Bit is set		'0' = Bit is cleared x = Bit is unknown				

bit 7-6	<b>ADRES&lt;1:0&gt;:</b> ADC Result Register bits Lower 2 bits of 10-bit conversion result
bit 5-0	Reserved: Do not use.

#### REGISTER 9-5: ADRESH: ADC RESULT REGISTER HIGH (ADRESH) ADFM = 1 (READ-ONLY)

U-0	U-0	U-0	U-0	U-0	U-0	R-x	R-x
—	—	—	—	—	_	ADRES9	ADRES8
bit 7							bit 0

Legend:						
R = Readable bit	able bit W = Writable bit U = Unimplemented bit, read as '0'					
-n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown			

bit 7-2 Reserved: Do not use.

bit 1-0 ADRES<9:8>: ADC Result Register bits Upper 2 bits of 10-bit conversion result

## REGISTER 9-6: ADRESL: ADC RESULT REGISTER LOW (ADRESL) ADFM = 1 (READ-ONLY)

| R-x    |
|--------|--------|--------|--------|--------|--------|--------|--------|
| ADRES7 | ADRES6 | ADRES5 | ADRES4 | ADRES3 | ADRES2 | ADRES1 | ADRES0 |
| bit 7  |        |        |        |        |        |        | bit 0  |

Legend:					
R = Readable bit	W = Writable bit	U = Unimplemented bit, r	U = Unimplemented bit, read as '0'		
-n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown		

bit 7-0 ADRES<7:0>: ADC Result Register bits Lower 8 bits of 10-bit conversion result

## 9.3 A/D Acquisition Requirements

For the ADC to meet its specified accuracy, the charge holding capacitor (CHOLD) must be allowed to fully charge to the input channel voltage level. The Analog Input model is shown in Figure 9-4. The source impedance (Rs) and the internal sampling switch (Rss) impedance directly affect the time required to charge the capacitor CHOLD. The sampling switch (Rss) impedance varies over the device voltage (VDD), see Figure 9-4. The maximum recommended impedance for analog sources is 10 k $\Omega$ . As the source impedance is decreased, the acquisition time may be decreased. After the analog input channel is selected (or changed), an A/D acquisition must be done before the conversion can be started. To calculate the minimum acquisition time, Equation 9-1 may be used. This equation assumes that 1/2 LSb error is used (1024 steps for the ADC). The 1/2 LSb error is the maximum error allowed for the ADC to meet its specified resolution.

#### EQUATION 9-1: ACQUISITION TIME EXAMPLE

Assumptions: Temperature =  $50^{\circ}C$  and external impedance of  $10k\Omega 5.0V VDD$  TACQ = Amplifier Settling Time + Hold Capacitor Charging Time + Temperature Coefficient = TAMP + TC + TCOFF $= 5\mu s + TC + [(Temperature - 25^{\circ}C)(0.05\mu s/^{\circ}C)]$ 

The value for TC can be approximated with the following equations:

$$V_{APPLIED}\left(1 - \frac{1}{2047}\right) = V_{CHOLD} \qquad ;[1] V_{CHOLD} \ charged \ to \ within \ 1/2 \ lsb$$

$$V_{APPLIED}\left(1 - e^{\frac{-Tc}{RC}}\right) = V_{CHOLD} \qquad ;[2] \ V_{CHOLD} \ charge \ response \ to \ V_{APPLIED} \ (1 - e^{\frac{-Tc}{RC}}) = V_{CHOLD} \qquad ;[2] \ V_{CHOLD} \ charge \ response \ to \ V_{APPLIED} \ (1 - e^{\frac{-Tc}{RC}}) = V_{CHOLD} \qquad ;[2] \ V_{CHOLD} \ charge \ response \ to \ V_{APPLIED} \ (1 - e^{\frac{-Tc}{RC}}) = V_{CHOLD} \qquad ;[2] \ V_{CHOLD} \ charge \ response \ to \ V_{APPLIED} \ (1 - e^{\frac{-Tc}{RC}}) = V_{CHOLD} \qquad ;[2] \ V_{CHOLD} \ charge \ response \ to \ V_{APPLIED} \ (1 - e^{\frac{-Tc}{RC}}) = V_{CHOLD} \qquad ;[2] \ V_{CHOLD} \ charge \ response \ to \ V_{APPLIED} \ (1 - e^{\frac{-Tc}{RC}}) = V_{CHOLD} \qquad ;[2] \ V_{CHOLD} \ charge \ response \ to \ V_{APPLIED} \ (1 - e^{\frac{-Tc}{RC}}) = V_{CHOLD} \qquad ;[2] \ V_{CHOLD} \ charge \ response \ to \ V_{APPLIED} \ (1 - e^{\frac{-Tc}{RC}}) = V_{CHOLD} \qquad ;[2] \ V_{CHOLD} \ charge \ response \ to \ V_{APPLIED} \ (1 - e^{\frac{-Tc}{RC}}) = V_{CHOLD} \qquad ;[2] \ V_{CHOLD} \ charge \ response \ to \ V_{APPLIED} \ (1 - e^{\frac{-Tc}{RC}}) = V_{CHOLD} \ (1 - e^{\frac{-Tc}{RC}}) = V$$

$$V_{APPLIED}\left(1 - e^{\frac{-1}{RC}}\right) = V_{APPLIED}\left(1 - \frac{1}{2047}\right) \quad \text{; combining [1] and [2]}$$

Solving for TC:

$$TC = -CHOLD(RIC + RSS + RS) \ln(1/2047)$$
  
=  $-10pF(1k\Omega + 7k\Omega + 10k\Omega) \ln(0.0004885)$   
=  $1.37\mu s$ 

Therefore:

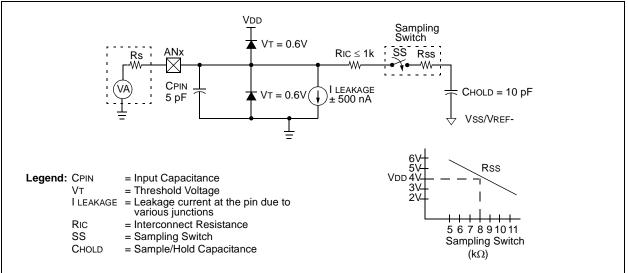
$$TACQ = 5\mu s + 1.37\mu s + [(50^{\circ}C - 25^{\circ}C)(0.05\mu s/^{\circ}C)]$$
  
= 7.67\mu s

**Note 1:** The reference voltage (VREF) has no effect on the equation, since it cancels itself out.

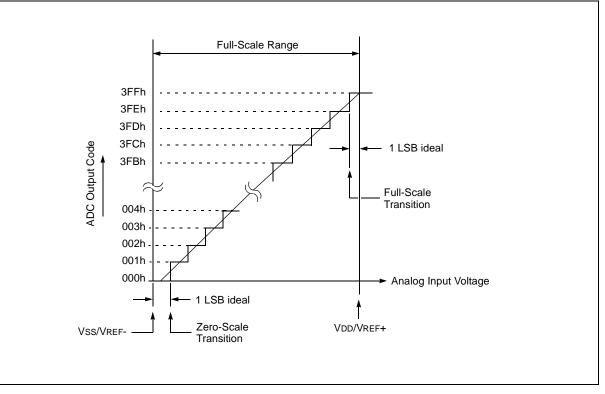
2: The charge holding capacitor (CHOLD) is not discharged after each conversion.

**3:** The maximum recommended impedance for analog sources is 10 k $\Omega$ . This is required to meet the pin leakage specification.

#### FIGURE 9-4: ANALOG INPUT MODEL







Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Value on POR, BOR	Value on all other Resets
ADCON0 <sup>(1)</sup>	ADFM	VCFG	CHS3	CHS2	CHS1	CHS0	GO/DONE	ADON	0000 0000	0000 0000
ADCON1 <sup>(1)</sup>	_	ADCS2	ADCS1	ADCS0	_	_	_	_	-000	-000
ANSEL	ANS	ANS6	ANS5	ANS4	ANS3 <sup>(1)</sup>	ANS2 <sup>(1)</sup>	ANS1	ANS0	1111 1111	1111 1111
ADRESH <sup>(1,2)</sup>	A/D Result	Register Hig	h Byte						xxxx xxxx	uuuu uuuu
ADRESL <sup>(1,2)</sup>	A/D Result	Register Lov	v Byte						xxxx xxxx	uuuu uuuu
INTCON	GIE	PEIE	TOIE	INTE	RAIE	T0IF	INTF	RAIF	0000 0000	0000 0000
PIE1	—	ADIE <sup>(1)</sup>	CCP1IE <sup>(1)</sup>	C2IE	C1IE	-	TMR2IE <sup>(1)</sup>	TMR1IE	-000 0-00	-000 0-00
PIR1	_	ADIF <sup>(1)</sup>	CCP1IF <sup>(1)</sup>	C2IF	C1IF	_	TMR2IF <sup>(1)</sup>	TMR1IF	-000 0-00	-000 0-00
PORTA	_	_	RA5	RA4	RA3	RA2	RA1	RA0	x0 x000	u0 u000
PORTC	_	_	RC5	RC4	RC3	RC2	RC1	RC0	xx 00xx	uu 00uu
TRISA	_		TRISA5	TRISA4	TRISA3	TRISA2	TRISA1	TRISA0	11 1111	11 1111
TRISC	_		TRISC5	TRISC4	TRISC3	TRISC2	TRISC1	TRISC0	11 1111	11 1111

#### TABLE 9-2:SUMMARY OF ASSOCIATED ADC REGISTERS

Legend: x = unknown, u = unchanged, - = unimplemented read as '0'. Shaded cells are not used for ADC module.

Note 1: PIC16F616/16HV616 only.

2: Read-only Register.

NOTES:

## 10.0 ENHANCED CAPTURE/ COMPARE/PWM (WITH AUTO-SHUTDOWN AND DEAD BAND) MODULE (PIC16F616/16HV616 ONLY)

The Enhanced Capture/Compare/PWM module is a peripheral which allows the user to time and control different events. In Capture mode, the peripheral allows the timing of the duration of an event. The Compare mode allows the user to trigger an external

event when a predetermined amount of time has expired. The PWM mode can generate a Pulse-Width Modulated signal of varying frequency and duty cycle.

Table 10-1 shows the timer resources required by the ECCP module.

#### TABLE 10-1: ECCP MODE – TIMER RESOURCES REQUIRED

ECCP Mode	Timer Resource
Capture	Timer1
Compare	Timer1
PWM	Timer2

## REGISTER 10-1: CCP1CON: ENHANCED CCP1 CONTROL REGISTER

R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0						
P1M1	P1M0	DC1B1	DC1B0	CCP1M3	CCP1M2	CCP1M1	CCP1M0						
bit 7							bit 0						
Legend:													
R = Readabl	e bit	W = Writable b	oit	U = Unimplen	nented bit, read	d as '0'							
-n = Value at	POR	'1' = Bit is set		'0' = Bit is clea		x = Bit is unkı	nown						
bit 7-6		WM Output Cor											
		<u>2&gt; = 00,01,10</u> signed as Capt		input P18 P10		ad as nort nins							
	If CCP1M<3	•	are/compare	input, i i b, i it	5, 1 10 assign								
		output; P1A mo	dulated; P1B,	P1C, P1D assi	gned as port p	ins							
	01 = Full-Bri	dge output forw	ard; P1D mod	lulated; P1A ac	tive; P1B, P1C	inactive							
		idge output; P1A					ed as port pins						
	11 = Full-Bri	dge output reve	rse; P1B mod	ulated; P1C ac	tive; P1A, P1D	inactive							
bit 5-4	DC1B<1:0>: PWM Duty Cycle Least Significant bits												
	Capture mode:												
	Unused.												
	Compare mode:												
	Unused. PWM mode:												
		e the two LSbs o	of the PWM du	uty cycle. The ei	ight MSbs are f	ound in CCPR	1L.						
bit 3-0		ECCP Mode Sel		, ,	0								
	0000 = Capt	ture/Compare/PW	M off (resets E	CCP module)									
	0001 = Unused (reserved)												
	0010 = Compare mode, toggle output on match (CCP1IF bit is set)												
	0011 = Unused (reserved) 0100 = Capture mode, every falling edge												
	0100 = Capture mode, every raining edge 0101 = Capture mode, every rising edge												
	0110 = Capture mode, every 4th rising edge												
		ture mode, every			0								
		pare mode, set or pare mode, clear											
		ipare mode, gene				et. CCP1 pin is u	naffected)						
		pare mode, trigg											
	conv	version, if the ADC	c module is ena	bled)									
		/ mode; P1A, P10	0,	,	0								
		/I mode; P1A, P10 /I mode; P1A, P10	-										
		/I mode; P1A, P10			0								

### 10.1 Capture Mode

In Capture mode, CCPR1H:CCPR1L captures the 16-bit value of the TMR1 register when an event occurs on pin CCP1. An event is defined as one of the following and is configured by the CCP1M<3:0> bits of the CCP1CON register:

- · Every falling edge
- Every rising edge
- Every 4th rising edge
- Every 16th rising edge

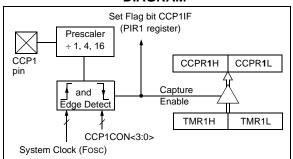
When a capture is made, the Interrupt Request Flag bit CCP1IF of the PIR1 register is set. The interrupt flag must be cleared in software. If another capture occurs before the value in the CCPR1H, CCPR1L register pair is read, the old captured value is overwritten by the new captured value (see Figure 10-1).

#### 10.1.1 CCP1 PIN CONFIGURATION

In Capture mode, the CCP1 pin should be configured as an input by setting the associated TRIS control bit.

Note:	If the CCP1 pin is configured as an output,								
	a write to the port can cause a capture								
	condition.								

#### FIGURE 10-1: CAPTURE MODE OPERATION BLOCK DIAGRAM



#### 10.1.2 TIMER1 MODE SELECTION

Timer1 must be running in Timer mode or Synchronized Counter mode for the CCP module to use the capture feature. In Asynchronous Counter mode, the capture operation may not work.

#### 10.1.3 SOFTWARE INTERRUPT

When the Capture mode is changed, a false capture interrupt may be generated. The user should keep the CCP1IE interrupt enable bit of the PIE1 register clear to avoid false interrupts. Additionally, the user should clear the CCP1IF interrupt flag bit of the PIR1 register following any change in operating mode.

#### 10.1.4 CCP PRESCALER

There are four prescaler settings specified by the CCP1M<3:0> bits of the CCP1CON register. Whenever the CCP module is turned off, or the CCP module is not in Capture mode, the prescaler counter is cleared. Any Reset will clear the prescaler counter.

Switching from one capture prescaler to another does not clear the prescaler and may generate a false interrupt. To avoid this unexpected operation, turn the module off by clearing the CCP1CON register before changing the prescaler (see Example 10-1).

#### EXAMPLE 10-1: CHANGING BETWEEN CAPTURE PRESCALERS

BANKSEI	CCP1CON	;Set Bank bits to point
		;to CCP1CON
CLRF	CCP1CON	;Turn CCP module off
MOVLW	NEW_CAPT_PS	;Load the W reg with
		; the new prescaler
		; move value and CCP ON
MOVWF	CCP1CON	;Load CCP1CON with this
		; value

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Value on POR, BOR	Value on all other Resets
CCP1CON <sup>(1)</sup>	P1M1	P1M0	DC1B1	DC1B0	CCP1M3	CCP1M2	CCP1M1	CCP1M0	0000 0000	0000 0000
CCPR1L <sup>(1)</sup>	Capture/Cor	mpare/PWM I	xxxx xxxx	uuuu uuuu						
CCPR1H <sup>(1)</sup>	Capture/Cor	mpare/PWM I	Register 1 Hig	gh Byte					xxxx xxxx	uuuu uuuu
INTCON	GIE	PEIE	T0IE	INTE	RAIE	T0IF	INTF	RAIF	0000 0000	0000 0000
PIE1	_	ADIE <sup>(1)</sup>	CCP1IE <sup>(1)</sup>	C2IE	C1IE		TMR2IE <sup>(1)</sup>	TMR1IE	-000 0-00	0000 0-00
PIR1	—	ADIF <sup>(1)</sup>	CCP1IF <sup>(1)</sup>	C2IF	C1IF	—	TMR2IF <sup>(1)</sup>	TMR1IF	-000 0-00	0000 0-00
T1CON	T1GINV	TMR1GE	T1CKPS1	T1CKPS0	T1OSCEN	T1SYNC	TMR1CS	TMR10N	0000 0000	uuuu uuuu
TMR1L	Holding Reg	ister for the L	east Significa	ant Byte of the	e 16-bit TMR	1 Register	•		xxxx xxxx	uuuu uuuu
TMR1H	Holding Reg	ister for the N	/lost Significa	nt Byte of the	e 16-bit TMR1	Register			xxxx xxxx	uuuu uuuu
TRISA	—	—	TRISA5	TRISA4	TRISA3	TRISA2	TRISA1	TRISA0	11 1111	11 1111
TRISC	_	_	TRISC5	TRISC4	TRISC3	TRISC2	TRISC1	TRISC0	11 1111	11 1111
						<u> </u>				

#### TABLE 10-2: SUMMARY OF REGISTERS ASSOCIATED WITH CAPTURE

Legend: -= Unimplemented locations, read as '0', u = unchanged, x = unknown. Shaded cells are not used by the Capture, Compare and PWM. Note 1: PIC16F616/16HV616 only.

## 10.2 Compare Mode

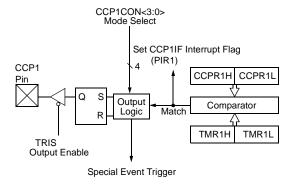
In Compare mode, the 16-bit CCPR1 register value is constantly compared against the TMR1 register pair value. When a match occurs, the CCP1 module may:

- Toggle the CCP1 output
- Set the CCP1 output
- Clear the CCP1 output
- Generate a Special Event Trigger
- Generate a Software Interrupt

The action on the pin is based on the value of the CCP1M<3:0> control bits of the CCP1CON register.

All Compare modes can generate an interrupt.

#### FIGURE 10-2: COMPARE MODE OPERATION BLOCK DIAGRAM



Special Event Trigger will:

- Clear TMR1H and TMR1L registers.
- NOT set interrupt flag bit TMR1IF of the PIR1 register.
- Set the GO/DONE bit to start the ADC conversion.

#### 10.2.1 CCP1 PIN CONFIGURATION

The user must configure the CCP1 pin as an output by clearing the associated TRIS bit.

Note:	Clearing the CCP1CON register will force								
	the CCP1 compare output latch to the								
	default low level. This is not the PORT I/O								
	data latch.								

#### 10.2.2 TIMER1 MODE SELECTION

In Compare mode, Timer1 must be running in either Timer mode or Synchronized Counter mode. The compare operation may not work in Asynchronous Counter mode.

#### 10.2.3 SOFTWARE INTERRUPT MODE

When Generate Software Interrupt mode is chosen (CCP1M<3:0> = 1010), the CCP1 module does not assert control of the CCP1 pin (see the CCP1CON register).

#### 10.2.4 SPECIAL EVENT TRIGGER

When Special Event Trigger mode is chosen (CCP1M<3:0> = 1011), the CCP1 module does the following:

- Resets Timer1
- Starts an ADC conversion if ADC is enabled

The CCP1 module does not assert control of the CCP1 pin in this mode (see the CCP1CON register).

The Special Event Trigger output of the CCP occurs immediately upon a match between the TMR1H, TMR1L register pair and the CCPR1H, CCPR1L register pair. The TMR1H, TMR1L register pair is not reset until the next rising edge of the Timer1 clock. This allows the CCPR1H, CCPR1L register pair to effectively provide a 16-bit programmable period register for Timer1.

- Note 1: The Special Event Trigger from the CCP module does not set interrupt flag bit TMR1IF of the PIR1 register.
  - 2: Removing the match condition by changing the contents of the CCPR1H and CCPR1L register pair, between the clock edge that generates the Special Event Trigger and the clock edge that generates the Timer1 Reset, will preclude the Reset from occurring.

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Value on POR, BOR	Value on all other Resets
CCP1CON <sup>(1)</sup>	P1M1	P1M0	DC1B1	DC1B0	CCP1M3	CCP1M2	CCP1M1	CCP1M0	0000 0000	0000 0000
CCPR1L <sup>(1)</sup>	Capture/Cor	mpare/PWM I	xxxx xxxx	uuuu uuuu						
CCPR1H <sup>(1)</sup>	Capture/Cor	mpare/PWM I	Register 1 Hig	gh Byte					xxxx xxxx	uuuu uuuu
INTCON	GIE	PEIE	T0IE	INTE	RAIE	T0IF	INTF	RAIF	0000 0000	0000 0000
PIE1	_	ADIE <sup>(1)</sup>	CCP1IE <sup>(1)</sup>	C2IE	C1IE		TMR2IE <sup>(1)</sup>	TMR1IE	-000 0-00	0000 0-00
PIR1	—	ADIF <sup>(1)</sup>	CCP1IF <sup>(1)</sup>	C2IF	C1IF	—	TMR2IF <sup>(1)</sup>	TMR1IF	-000 0-00	0000 0-00
T1CON	T1GINV	TMR1GE	T1CKPS1	T1CKPS0	T1OSCEN	T1SYNC	TMR1CS	TMR10N	0000 0000	uuuu uuuu
TMR1L	Holding Reg	ister for the L	east Significa	ant Byte of the	e 16-bit TMR	1 Register	•		xxxx xxxx	uuuu uuuu
TMR1H	Holding Reg	ister for the N	/lost Significa	nt Byte of the	e 16-bit TMR1	Register			xxxx xxxx	uuuu uuuu
TRISA	—	—	TRISA5	TRISA4	TRISA3	TRISA2	TRISA1	TRISA0	11 1111	11 1111
TRISC	_	_	TRISC5	TRISC4	TRISC3	TRISC2	TRISC1	TRISC0	11 1111	11 1111
						<u> </u>				

#### TABLE 10-3: SUMMARY OF REGISTERS ASSOCIATED WITH COMPARE

Legend: -= Unimplemented locations, read as '0', u = unchanged, x = unknown. Shaded cells are not used by the Capture, Compare and PWM. Note 1: PIC16F616/16HV616 only.

### 10.3 PWM Mode

The PWM mode generates a Pulse-Width Modulated signal on the CCP1 pin. The duty cycle, period and resolution are determined by the following registers:

- PR2
- T2CON
- CCPR1L
- CCP1CON

In Pulse-Width Modulation (PWM) mode, the CCP module produces up to a 10-bit resolution PWM output on the CCP1 pin. Since the CCP1 pin is multiplexed with the PORT data latch, the TRIS for that pin must be cleared to make the CCP1 pin an output.

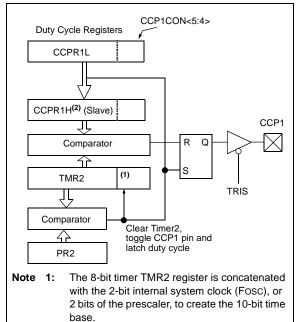
Note:	Clearing	the	CCP1CON	register	will
	relinquish	CCP	1 control of the	ne CCP1	pin.

Figure 10-3 shows a simplified block diagram of PWM operation.

Figure 10-4 shows a typical waveform of the PWM signal.

For a step-by-step procedure on how to set up the CCP module for PWM operation, see **Section 10.3.7** "Setup for PWM Operation".

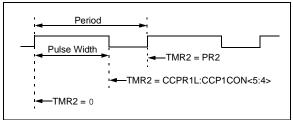
FIGURE 10-3: SIMPLIFIED PWM BLOCK DIAGRAM



2: In PWM mode, CCPR1H is a read-only register.

The PWM output (Figure 10-4) has a time base (period) and a time that the output stays high (duty cycle).

FIGURE 10-4: CCP PWM OUTPUT



#### 10.3.1 PWM PERIOD

The PWM period is specified by writing to the PR2 register of Timer2. The PWM period can be calculated using the formula of Equation 10-1.

#### EQUATION 10-1: PWM PERIOD

$$PWM Period = [(PR2) + 1] \bullet 4 \bullet Tosc \bullet$$
$$(TMR2 Prescale Value)$$

When TMR2 is equal to PR2, the following three events occur on the next increment cycle:

- TMR2 is cleared
- The CCP1 pin is set. (Exception: If the PWM duty cycle = 0%, the pin will not be set.)
- The PWM duty cycle is latched from CCPR1L into CCPR1H.

Note:	The Timer2 postscaler (see Section 7.1
	"Timer2 Operation") is not used in the
	determination of the PWM frequency.

#### 10.3.2 PWM DUTY CYCLE

The PWM duty cycle is specified by writing a 10-bit value to multiple registers: CCPR1L register and CCP1<1:0> bits of the CCP1CON register. The CCPR1L contains the eight MSbs and the CCP1<1:0> bits of the CCP1CON register contain the two LSbs. CCPR1L and CCP1<1:0> bits of the CCP1CON register can be written to at any time. The duty cycle value is not latched into CCPR1H until after the period completes (i.e., a match between PR2 and TMR2 registers occurs). While using the PWM, the CCPR1H register is read-only.

Equation 10-2 is used to calculate the PWM pulse width.

Equation 10-3 is used to calculate the PWM duty cycle ratio.

#### EQUATION 10-2: PULSE WIDTH

 $Pulse Width = (CCPR1L:CCP1CON < 5:4>) \bullet$ 

TOSC • (TMR2 Prescale Value)

## EQUATION 10-3: DUTY CYCLE RATIO

$$Duty Cycle Ratio = \frac{(CCPR1L:CCP1CON < 5:4>)}{4(PR2 + 1)}$$

The CCPR1H register and a 2-bit internal latch are used to double buffer the PWM duty cycle. This double buffering is essential for glitchless PWM operation.

The 8-bit timer TMR2 register is concatenated with either the 2-bit internal system clock (Fosc), or 2 bits of the prescaler, to create the 10-bit time base. The system clock is used if the Timer2 prescaler is set to 1:1.

When the 10-bit time base matches the CCPR1H and 2-bit latch, then the CCP1 pin is cleared (see Figure 10-3).

#### 10.3.3 PWM RESOLUTION

The resolution determines the number of available duty cycles for a given period. For example, a 10-bit resolution will result in 1024 discrete duty cycles, whereas an 8-bit resolution will result in 256 discrete duty cycles.

The maximum PWM resolution is 10 bits when PR2 is 255. The resolution is a function of the PR2 register value as shown by Equation 10-4.

#### EQUATION 10-4: PWM RESOLUTION

Resolution = 
$$\frac{\log[4(PR2 + 1)]}{\log(2)}$$
 bits

Note: If the pulse width value is greater than the period the assigned PWM pin(s) will remain unchanged.

#### TABLE 10-4: EXAMPLE PWM FREQUENCIES AND RESOLUTIONS (Fosc = 20 MHz)

PWM Frequency	1.22 kHz	4.88 kHz	19.53 kHz	78.12 kHz	156.3 kHz	208.3 kHz
Timer Prescale (1, 4, 16)	16	4	1	1	1	1
PR2 Value	0xFF	0xFF	0xFF	0x3F	0x1F	0x17
Maximum Resolution (bits)	10	10	10	8	7	6.6

#### TABLE 10-5: EXAMPLE PWM FREQUENCIES AND RESOLUTIONS (Fosc = 8 MHz)

PWM Frequency	1.22 kHz	4.90 kHz	19.61 kHz	76.92 kHz	153.85 kHz	200.0 kHz
Timer Prescale (1, 4, 16)	16	4	1	1	1	1
PR2 Value	0x65	0x65	0x65	0x19	0x0C	0x09
Maximum Resolution (bits)	8	8	8	6	5	5

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### 10.3.4 OPERATION IN SLEEP MODE

In Sleep mode, the TMR2 register will not increment and the state of the module will not change. If the CCP1 pin is driving a value, it will continue to drive that value. When the device wakes up, TMR2 will continue from its previous state.

#### 10.3.5 CHANGES IN SYSTEM CLOCK FREQUENCY

The PWM frequency is derived from the system clock frequency. Any changes in the system clock frequency will result in changes to the PWM frequency. See **Section 3.0** "**Oscillator Module**" for additional details.

#### 10.3.6 EFFECTS OF RESET

Any Reset will force all ports to Input mode and the CCP registers to their Reset states.

#### 10.3.7 SETUP FOR PWM OPERATION

The following steps should be taken when configuring the CCP module for PWM operation:

- 1. Configure the PWM pin (CCP1) as an input by setting the associated TRIS bit.
- 2. Set the PWM period by loading the PR2 register.
- Configure the CCP module for the PWM mode by loading the CCP1CON register with the appropriate values.
- 4. Set the PWM duty cycle by loading the CCPR1L register and CCP1 bits of the CCP1CON register.
- 5. Configure and start Timer2:
  - Clear the TMR2IF interrupt flag bit of the PIR1 register.
  - Set the Timer2 prescale value by loading the T2CKPS bits of the T2CON register.
  - Enable Timer2 by setting the TMR2ON bit of the T2CON register.
- 6. Enable PWM output after a new PWM cycle has started:
  - Wait until Timer2 overflows (TMR2IF bit of the PIR1 register is set).
  - Enable the CCP1 pin output by clearing the associated TRIS bit.

## 10.4 PWM (Enhanced Mode)

The Enhanced PWM Mode can generate a PWM signal on up to four different output pins with up to 10-bits of resolution. It can do this through four different PWM Output modes:

- Single PWM
- Half-Bridge PWM
- Full-Bridge PWM, Forward mode
- Full-Bridge PWM, Reverse mode

To select an Enhanced PWM mode, the P1M bits of the CCP1CON register must be set appropriately.

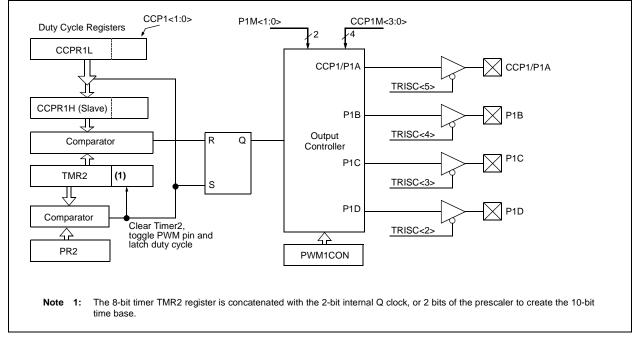
The PWM outputs are multiplexed with I/O pins and are designated P1A, P1B, P1C and P1D. The polarity of the PWM pins is configurable and is selected by setting the CCP1M bits in the CCP1CON register appropriately.

Table 10-6 shows the pin assignments for each Enhanced PWM mode.

Figure 10-5 shows an example of a simplified block diagram of the Enhanced PWM module.

**Note:** To prevent the generation of an incomplete waveform when the PWM is first enabled, the ECCP module waits until the start of a new PWM period before generating a PWM signal.

#### FIGURE 10-5: EXAMPLE SIMPLIFIED BLOCK DIAGRAM OF THE ENHANCED PWM MODE



Note 1: The TRIS register value for each PWM output must be configured appropriately.

- 2: Clearing the CCP1CON register will relinquish ECCP control of all PWM output pins.
- 3: Any pin not used by an Enhanced PWM mode is available for alternate pin functions

#### TABLE 10-6: EXAMPLE PIN ASSIGNMENTS FOR VARIOUS PWM ENHANCED MODES

ECCP Mode	P1M	CCP1/P1A	P1B	P1C	P1D
Single	00	Yes	No	No	No
Half-Bridge	10	Yes	Yes	No	No
Full-Bridge, Forward	01	Yes	Yes	Yes	Yes
Full-Bridge, Reverse	11	Yes	Yes	Yes	Yes

## FIGURE 10-6: EXAMPLE PWM (ENHANCED MODE) OUTPUT RELATIONSHIPS (ACTIVE-HIGH STATE)

P1M<		Signal	0	Width	─► Period ───	
00	(Single Output)	P1A Modulated		Delay <sup>(1)</sup>	Delaý <sup>(1)</sup>	İ
		P1A Modulated	— I			
10	(Half-Bridge)	P1B Modulated	'			
		P1A Active			, 	
(Full-Bridge, <sup>01</sup> Forward)	P1B Inactive			   	1 1 1	
	Forward)	P1C Inactive	i			
		P1D Modulated	=ļ			
		P1A Inactive			 	1 1 1
11	(Full-Bridge,	P1B Modulated				
	Reverse)	P1C Active –	;			
		P1D Inactive				
Relat	ionships:	c * (PR2 + 1) * (TMR2 Pr			-	'

Note 1: Dead-band delay is programmed using the PWM1CON register (Section 10.4.6 "Programmable Dead-Band Delay mode").

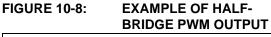
P1M<	1:0>	Signal	0 Pulse Width	<b>—</b>	PR2+1
		_		Period ——	<b>→</b>
00	(Single Output)	P1A Modulated			1
		P1A Modulated	Delay <sup>(1)</sup>	Delay <sup>(1)</sup>	
10	(Half-Bridge)	P1B Modulated			;
		P1A Active	1 1 		
01	(Full-Bridge, Forward)	P1B Inactive	· · · · · · · · · · · · · · · · · · ·	I	I  
		P1C Inactive	 	1	1
		P1D Modulated			
		P1A Inactive		   	 
11	(Full-Bridge, Reverse)	P1B Modulated		I	
	,	P1C Active			
		P1D Inactive			   
Relat	• Pulse Width = To	c * (PR2 + 1) * (TMR2 Presca >sc * (CCPR1L<7:0>:CCP1C( : * (PWM1CON<6:0>)		cale Value)	

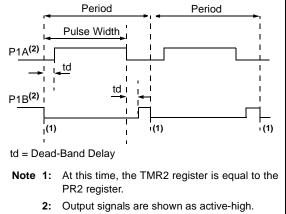
#### FIGURE 10-7: EXAMPLE ENHANCED PWM OUTPUT RELATIONSHIPS (ACTIVE-LOW STATE)

#### 10.4.1 HALF-BRIDGE MODE

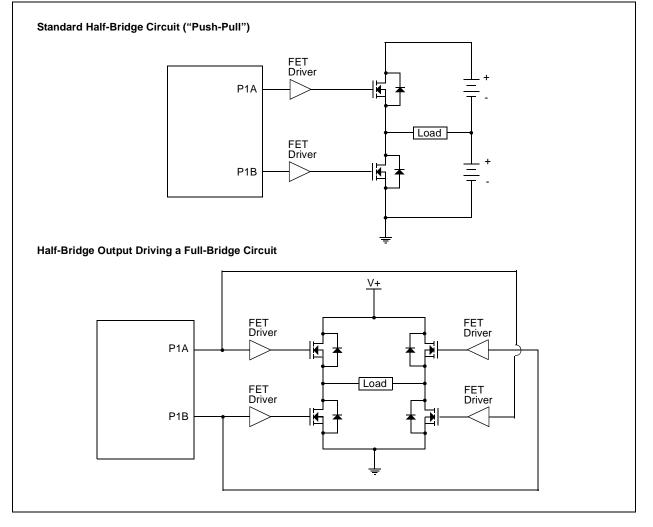
In Half-Bridge mode, two pins are used as outputs to drive push-pull loads. The PWM output signal is output on the CCP1/P1A pin, while the complementary PWM output signal is output on the P1B pin (see Figure 10-8). This mode can be used for half-bridge applications, as shown in Figure 10-9, or for full-bridge applications, where four power switches are being modulated with two PWM signals.

In Half-Bridge mode, the programmable dead-band delay can be used to prevent shoot-through current in halfbridge power devices. The value of the PDC<6:0> bits of the PWM1CON register sets the number of instruction cycles before the output is driven active. If the value is greater than the duty cycle, the corresponding output remains inactive during the entire cycle. See **10.4.6 "Programmable Dead-Band Delay mode"** for more details of the dead-band delay operations. Since the P1A and P1B outputs are multiplexed with the PORT data latches, the associated TRIS bits must be cleared to configure P1A and P1B as outputs.





## FIGURE 10-9: EXAMPLE OF HALF-BRIDGE APPLICATIONS



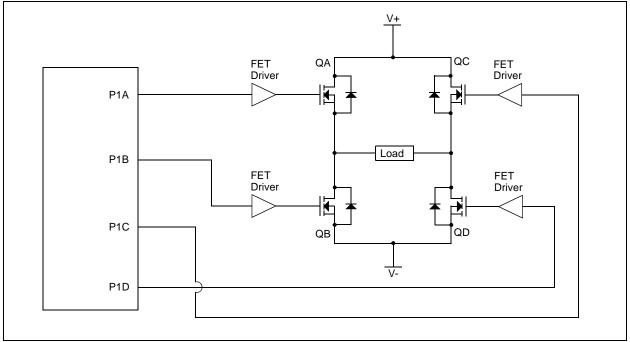
#### 10.4.2 FULL-BRIDGE MODE

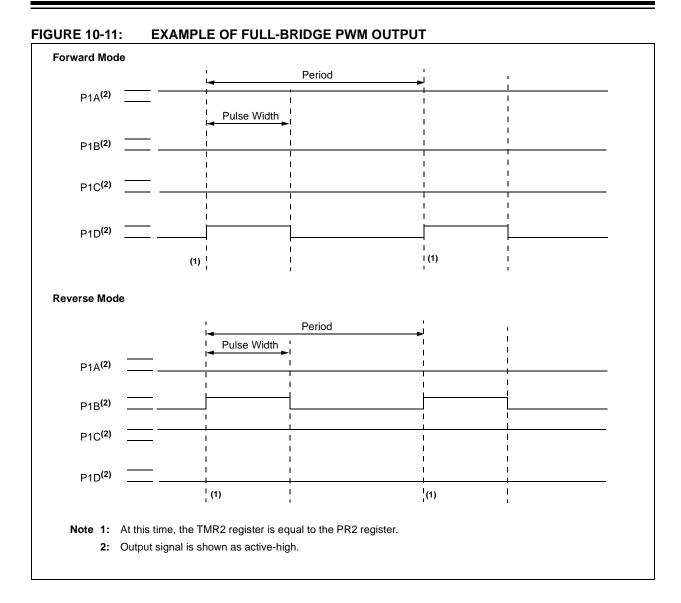
In Full-Bridge mode, all four pins are used as outputs. An example of full-bridge application is shown in Figure 10-10.

In the Forward mode, pin CCP1/P1A is driven to its active state, pin P1D is modulated, while P1B and P1C will be driven to their inactive state as shown in Figure 10-11.

In the Reverse mode, P1C is driven to its active state, pin P1B is modulated, while P1A and P1D will be driven to their inactive state as shown Figure 10-11. P1A, P1B, P1C and P1D outputs are multiplexed with the PORT data latches. The associated TRIS bits must be cleared to configure the P1A, P1B, P1C and P1D pins as outputs.

## FIGURE 10-10: EXAMPLE OF FULL-BRIDGE APPLICATION





## 10.4.2.1 Direction Change in Full-Bridge Mode

In the Full-Bridge mode, the P1M1 bit in the CCP1CON register allows users to control the forward/reverse direction. When the application firmware changes this direction control bit, the module will change to the new direction on the next PWM cycle.

A direction change is initiated in software by changing the P1M1 bit of the CCP1CON register. The following sequence occurs four Timer2 cycles prior to the end of the current PWM period:

- The modulated outputs (P1B and P1D) are placed in their inactive state.
- The associated unmodulated outputs (P1A and P1C) are switched to drive in the opposite direction.
- PWM modulation resumes at the beginning of the next period.

See Figure 10-12 for an illustration of this sequence.

The Full-Bridge mode does not provide dead-band delay. As one output is modulated at a time, dead-band delay is generally not required. There is a situation where dead-band delay is required. This situation occurs when both of the following conditions are true:

- 1. The direction of the PWM output changes when the duty cycle of the output is at or near 100%.
- 2. The turn off time of the power switch, including the power device and driver circuit, is greater than the turn on time.

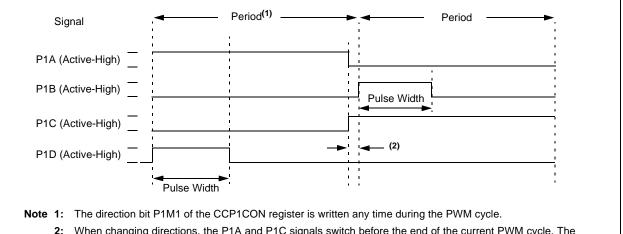
Figure 10-13 shows an example of the PWM direction changing from forward to reverse, at a near 100% duty cycle. In this example, at time t1, the output P1A and P1D become inactive, while output P1C becomes active. Since the turn off time of the power devices is longer than the turn on time, a shoot-through current will flow through power devices QC and QD (see Figure 10-10) for the duration of 't'. The same phenomenon will occur to power devices QA and QB for PWM direction change from reverse to forward.

If changing PWM direction at high duty cycle is required for an application, two possible solutions for eliminating the shoot-through current are:

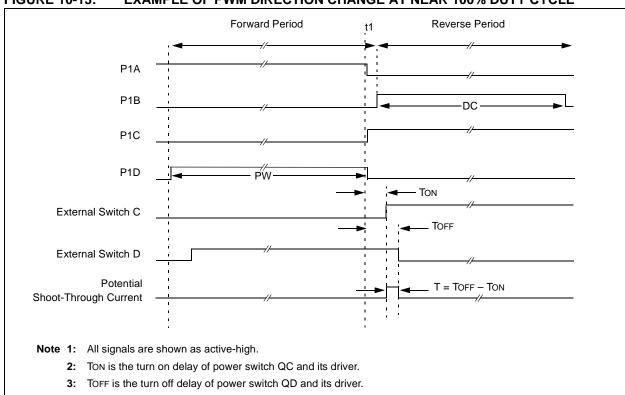
- 1. Reduce PWM duty cycle for one PWM period before changing directions.
- 2. Use switch drivers that can drive the switches off faster than they can drive them on.

Other options to prevent shoot-through current may exist.

#### FIGURE 10-12: EXAMPLE OF PWM DIRECTION CHANGE



2: When changing directions, the P1A and P1C signals switch before the end of the current PWM cycle. The modulated P1B and P1D signals are inactive at this time. The length of this time is four Timer2 counts.



### FIGURE 10-13: EXAMPLE OF PWM DIRECTION CHANGE AT NEAR 100% DUTY CYCLE

#### 10.4.3 START-UP CONSIDERATIONS

When any PWM mode is used, the application hardware must use the proper external pull-up and/or pull-down resistors on the PWM output pins.

Note: When the microcontroller is released from Reset, all of the I/O pins are in the highimpedance state. The external circuits must keep the power switch devices in the OFF state until the microcontroller drives the I/O pins with the proper signal levels or activates the PWM output(s).

The CCP1M<1:0> bits of the CCP1CON register allow the user to choose whether the PWM output signals are active-high or active-low for each pair of PWM output pins (P1A/P1C and P1B/P1D). The PWM output polarities must be selected before the PWM pins are configured as outputs. Changing the polarity configuration while the PWM pins are configured as outputs is not recommended since it may result in damage to the application circuits.

The P1A, P1B, P1C and P1D output latches may not be in the proper states when the PWM module is initialized. Enabling the PWM pins for output at the same time as the Enhanced PWM modes may cause damage to the application circuit. The Enhanced PWM modes must be enabled in the proper Output mode and complete a full PWM cycle before configuring the PWM pins as outputs. The completion of a full PWM cycle is indicated by the TMR2IF bit of the PIR1 register being set as the second PWM period begins.

#### 10.4.4 ENHANCED PWM AUTO-SHUTDOWN MODE

The PWM mode supports an Auto-Shutdown mode that will disable the PWM outputs when an external shutdown event occurs. Auto-Shutdown mode places the PWM output pins into a predetermined state. This mode is used to help prevent the PWM from damaging the application.

The auto-shutdown sources are selected using the ECCPASx bits of the ECCPAS register. A shutdown event may be generated by:

- A logic '0' on the INT pin
- Comparator C1
- Comparator C2
- · Setting the ECCPASE bit in firmware

A shutdown condition is indicated by the ECCPASE (Auto-Shutdown Event Status) bit of the ECCPAS register. If the bit is a '0', the PWM pins are operating normally. If the bit is a '1', the PWM outputs are in the shutdown state.

When a shutdown event occurs, two things happen:

The ECCPASE bit is set to '1'. The ECCPASE will remain set until cleared in firmware or an auto-restart occurs (see Section 10.4.5 "Auto-Restart Mode").

The enabled PWM pins are asynchronously placed in their shutdown states. The PWM output pins are grouped into pairs [P1A/P1C] and [P1B/P1D]. The state of each pin pair is determined by the PSSAC and PSSBD bits of the ECCPAS register. Each pin pair may be placed into one of three states:

- Drive logic '1'
- Drive logic '0'
- Tri-state (high-impedance)

#### REGISTER 10-2: ECCPAS: ENHANCED CAPTURE/COMPARE/PWM AUTO-SHUTDOWN CONTROL REGISTER

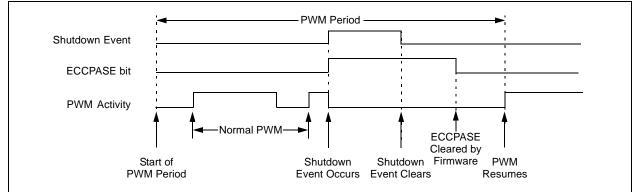
ECCPASE         ECCPAS2         ECCPAS1         ECCPAS0         PSSAC1         PSSBD1         PSSBD0           bit 7         bit 0         bit 0         bit 0         bit 0         bit 0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0
bit 7 bit 0	ECCPASE	ECCPAS2	ECCPAS1	ECCPAS0	PSSAC1	PSSAC0	PSSBD1	PSSBD0
	bit 7							bit 0

Legend:			
R = Readable bit	W = Writable bit	U = Unimplemented bit, rea	d as '0'
-n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown

<ul> <li>1 = A shutdown event has occurred; ECCP outputs are in shutd 0 = ECCP outputs are operating</li> <li>bit 6-4 ECCPAS&lt;2:0&gt;: ECCP Auto-shutdown Source Select bits</li> <li>000 = Auto-Shutdown is disabled</li> <li>001 = Comparator C1 output high</li> </ul>	
000 = Auto-Shutdown is disabled	iown state
010 = Comparator C2 output high <sup>(1)</sup> 011 = Either Comparators output is high 100 = VIL on INT pin 101 = VIL on INT pin or Comparator C1 output high 110 = VIL on INT pin or Comparator C2 output high	
111 = VIL on INT pin or either Comparators output is high	
bit 3-2 <b>PSSACn:</b> Pins P1A and P1C Shutdown State Control bits 00 = Drive pins P1A and P1C to '0' 01 = Drive pins P1A and P1C to '1' 1x = Pins P1A and P1C tri-state	
bit 1-0 <b>PSSBDn:</b> Pins P1B and P1D Shutdown State Control bits 00 = Drive pins P1B and P1D to '0' 01 = Drive pins P1B and P1D to '1' 1x = Pins P1B and P1D tri-state	

- Note 1: The auto-shutdown condition is a levelbased signal, not an edge-based signal. As long as the level is present, the autoshutdown will persist.
  - 2: Writing to the ECCPASE bit is disabled while an auto-shutdown condition persists.
  - 3: Once the auto-shutdown condition has been removed and the PWM restarted (either through firmware or auto-restart), the PWM signal will always restart at the beginning of the next PWM period.

#### FIGURE 10-14: PWM AUTO-SHUTDOWN WITH FIRMWARE RESTART (PRSEN = 0)

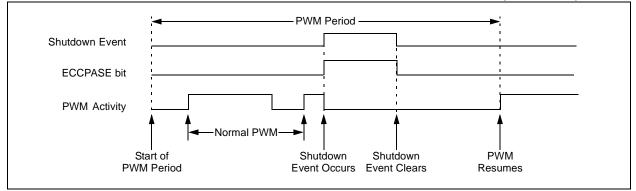


#### 10.4.5 AUTO-RESTART MODE

The Enhanced PWM can be configured to automatically restart the PWM signal once the auto-shutdown condition has been removed. Auto-restart is enabled by setting the PRSEN bit in the PWM1CON register.

If auto-restart is enabled, the ECCPASE bit will remain set as long as the auto-shutdown condition is active. When the auto-shutdown condition is removed, the ECCPASE bit will be cleared via hardware and normal operation will resume.

#### FIGURE 10-15: PWM AUTO-SHUTDOWN WITH AUTO-RESTART ENABLED (PRSEN = 1)

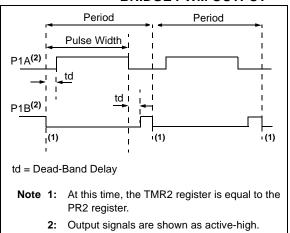


#### 10.4.6 PROGRAMMABLE DEAD-BAND DELAY MODE

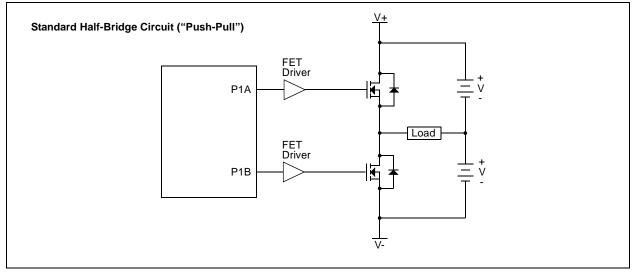
In half-bridge applications where all power switches are modulated at the PWM frequency, the power switches normally require more time to turn off than to turn on. If both the upper and lower power switches are switched at the same time (one turned on, and the other turned off), both switches may be on for a short period of time until one switch completely turns off. During this brief interval, a very high current (*shoot-through current*) will flow through both power switches, shorting the bridge supply. To avoid this potentially destructive shootthrough current from flowing during switching, turning on either of the power switches is normally delayed to allow the other switch to completely turn off.

In Half-Bridge mode, a digitally programmable deadband delay is available to avoid shoot-through current from destroying the bridge power switches. The delay occurs at the signal transition from the non-active state to the active state. See Figure 10-16 for illustration. The lower seven bits of the associated PWM1CON register (Register 10-3) sets the delay period in terms of microcontroller instruction cycles (TcY or 4 Tosc).

#### FIGURE 10-16: EXAMPLE OF HALF-BRIDGE PWM OUTPUT



## FIGURE 10-17: EXAMPLE OF HALF-BRIDGE APPLICATIONS



R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0
PRSEN	PDC6	PDC5	PDC4	PDC3	PDC2	PDC1	PDC0
bit 7							bit 0
l egend.							

#### REGISTER 10-3: PWM1CON: ENHANCED PWM CONTROL REGISTER

Legend:			
R = Readable bit	W = Writable bit	U = Unimplemented bit, rea	d as '0'
-n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown

#### bit 7 PRSEN: PWM Restart Enable bit

1 = Upon auto-shutdown, the ECCPASE bit clears automatically once the shutdown event goes away; the PWM restarts automatically

0 = Upon auto-shutdown, ECCPASE must be cleared in software to restart the PWM

bit 6-0 PDC<6:0>: PWM Delay Count bits

PDCn = Number of Fosc/4 (4 \* Tosc) cycles between the scheduled time when a PWM signal **should** transition active and the **actual** time it transitions active

#### TABLE 10-7: SUMMARY OF REGISTERS ASSOCIATED WITH PWM

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Value on POR, BOR	Value on all other Resets
CCP1CON <sup>(1)</sup>	P1M1	P1M0	DC1B1	DC1B0	CCP1M3	CCP1M2	CCP1M1	CCP1M0	0000 0000	0000 0000
CCPR1L <sup>(1)</sup>	1L <sup>(1)</sup> Capture/Compare/PWM Register 1 Low Byte									uuuu uuuu
CCPR1H <sup>(1)</sup>	H <sup>(1)</sup> Capture/Compare/PWM Register 1 High Byte								XXXX XXXX	uuuu uuuu
CM1CON0	C1ON	C1OUT	C1OE	C1POL	_	C1R	C1CH1	C1CH0	0000 -000	0000 -000
CM2CON0	C2ON	C2OUT	C2OE	C2POL	_	C2R	C2CH1	C2CH0	0000 -000	0000 -000
CM2CON1	MC1OUT	MC2OUT	_	T1ACS	C1HYS	C2HYS	T1GSS	C2SYNC	00-0 0010	00-0 0010
ECCPAS(1)	ECCPASE	ECCPAS2	ECCPAS1	ECCPAS0	PSSAC1	PSSAC0	PSSBD1	PSSBD0	0000 0000	0000 0000
INTCON	GIE	PEIE	T0IE	INTE	RAIE	T0IF	INTF	RAIF	0000 0000	0000 0000
PIE1	_	ADIE <sup>(1)</sup>	CCP1IE <sup>(1)</sup>	C2IE	C1IE		TMR2IE <sup>(1)</sup>	TMR1IE	-000 0-00	0000 0-00
PIR1	_	ADIF <sup>(1)</sup>	CCP1IF <sup>(1)</sup>	C2IF	C1IF		TMR2IF <sup>(1)</sup>	TMR1IF	-000 0-00	0000 0-00
PWM1CON <sup>(1)</sup>	PRSEN	PDC6	PDC5	PDC4	PDC3	PDC2	PDC1	PDC0	0000 0000	0000 0000
T2CON <sup>(1)</sup>	_	TOUTPS3	TOUTPS2	TOUTPS1	TOUTPS0	TMR2ON	T2CKPS1	T2CKPS0	-000 0000	-000 0000
TMR2 <sup>(1)</sup>	) Timer2 Module Register									0000 0000
TRISA	_	—	TRISA5	TRISA4	TRISA3	TRISA2	TRISA1	TRISA0	11 1111	11 1111
TRISC	_	_	TRISC5	TRISC4	TRISC3	TRISC2	TRISC1	TRISC0	11 1111	11 1111

Legend: -= Unimplemented locations, read as '0', u = unchanged, x = unknown. Shaded cells are not used by the Capture, Compare and PWM. Note 1: PIC16F616/16HV616 only.

NOTES:

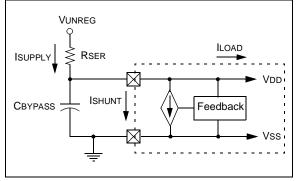
## 11.0 VOLTAGE REGULATOR

The PIC16HV610/16HV616 include a permanent internal 5 volt (nominal) shunt regulator in parallel with the VDD pin. This eliminates the need for an external voltage regulator in systems sourced by an unregulated supply. All external devices connected directly to the VDD pin will share the regulated supply voltage and contribute to the total VDD supply current (ILOAD).

## 11.1 Regulator Operation

A shunt regulator generates a specific supply voltage by creating a voltage drop across a pass resistor RSER. The voltage at the VDD pin of the microcontroller is monitored and compared to an internal voltage reference. The current through the resistor is then adjusted, based on the result of the comparison, to produce a voltage drop equal to the difference between the supply voltage VUNREG and the VDD of the microcontroller. See Figure 11-1 for voltage regulator schematic.





An external current limiting resistor, RSER, located between the unregulated supply, VUNREG, and the VDD pin, drops the difference in voltage between VUNREG and VDD. RSER must be between RMAX and RMIN as defined by Equation 11-1.

### EQUATION 11-1: RSER LIMITING RESISTOR

$$RMAX = \frac{(VUMIN - 5V)}{1.05 \cdot (4 MA + ILOAD)}$$

$$RMIN = \frac{(VUMAX - 5V)}{0.95 \cdot (50 \text{ MA})}$$

Where:

RMAX = maximum value of RSER (ohms)

RMIN = minimum value of RSER (ohms)

VUMIN = minimum value of VUNREG

VUMAX = maximum value of VUNREG

VDD = regulated voltage (5V nominal)

- ILOAD = maximum expected load current in mA including I/O pin currents and external circuits connected to VDD.
- 1.05 = compensation for +5% tolerance of RSER
- 0.95 = compensation for -5% tolerance of RSER

## 11.2 Regulator Considerations

The supply voltage VUNREG and load current are not constant. Therefore, the current range of the regulator is limited. Selecting a value for RSER must take these three factors into consideration.

Since the regulator uses the band gap voltage as the regulated voltage reference, this voltage reference is permanently enabled in the PIC16HV610/16HV616 devices.

NOTES:

## 12.0 SPECIAL FEATURES OF THE CPU

The PIC16F610/616/16HV610/616 has a host of features intended to maximize system reliability, minimize cost through elimination of external components, provide power-saving features and offer code protection.

These features are:

- Reset
  - Power-on Reset (POR)
  - Power-up Timer (PWRT)
  - Oscillator Start-up Timer (OST)
  - Brown-out Reset (BOR)
- Interrupts
- Watchdog Timer (WDT)
- Oscillator selection
- Sleep
- Code protection
- ID Locations
- In-Circuit Serial Programming<sup>™</sup>

The PIC16F610/616/16HV610/616 has two timers that offer necessary delays on power-up. One is the Oscillator Start-up Timer (OST), intended to keep the chip in Reset until the crystal oscillator is stable. The other is the Power-up Timer (PWRT), which provides a fixed delay of 64 ms (nominal) on power-up only, designed to keep the part in Reset while the power supply stabilizes. There is also circuitry to reset the device if a brown-out occurs, which can use the Powerup Timer to provide at least a 64 ms Reset. With these three functions-on-chip, most applications need no external Reset circuitry.

The Sleep mode is designed to offer a very low-current Power-Down mode. The user can wake-up from Sleep through:

- External Reset
- Watchdog Timer Wake-up
- An interrupt

Several oscillator options are also made available to allow the part to fit the application. The INTOSC option saves system cost while the LP crystal option saves power. A set of Configuration bits are used to select various options (see Register 12-1).

## 12.1 Configuration Bits

The Configuration bits can be programmed (read as '0'), or left unprogrammed (read as '1') to select various device configurations as shown in Register 12-1. These bits are mapped in program memory location 2007h.

**Note:** Address 2007h is beyond the user program memory space. It belongs to the special configuration memory space (2000h-3FFFh), which can be accessed only during programming. See the *Memory Programming Specification* (DS41284) for more information.

#### REGISTER 12-1: CONFIG: CONFIGURATION WORD REGISTER

_	—		_	_		BOREN1 <sup>(1)</sup>	BOREN0 <sup>(1)</sup>				
bit 15							bit 8				
IOSCFS	CP <sup>(2)</sup>	MCLRE <sup>(3)</sup>	PWRTE	WDTE	FOSC2	FOSC1	FOSC0				
bit 7	1		I	1			bit				
Legend:											
R = Readable	bit	W = Writable	bit	P = Program	mable'	U = Unimplem	nented bit,				
						read as '0'					
-n = Value at I	POR	'1' = Bit is set		'0' = Bit is cle	eared	x = Bit is unkr	IOWN				
	11		. ,								
bit 15-10	-	ented: Read as '1 D>: Brown-out Re		. h:4-(1)							
bit 9-8			eset Selection	DITS							
		<ul><li>11 = BOR enabled</li><li>10 = BOR enabled during operation and disabled in Sleep</li></ul>									
	0x = BOR d	x = BOR disabled									
bit 7	IOSCFS: Int	ternal Oscillator I	Frequency Se	elect bit							
	1 = 8 MHz										
	0 = 4  MHz	(2)									
bit 6		Protection bit <sup>(2)</sup>									
	<ul> <li>1 = Program memory code protection is disabled</li> <li>0 = Program memory code protection is enabled</li> </ul>										
bit 5	-	CLR Pin Function									
		in function is $\overline{MC}$		LR internally ti	ed to VDD						
bit 4	PWRTE: Po	wer-up Timer Er	nable bit								
	1 = PWRT d										
	0 = PWRT e										
bit 3		chdog Timer Ena	able bit								
	1 = WDT enabled 0 = WDT disabled										
bit 2-0	FOSC<2:0>	: Oscillator Sele	ction bits								
	111 = RC o	oscillator: CLKO	UT function or	n RA4/OSC2/C	CLKOUT pin, RO	C on RA5/OSC	1/CLKIN				
	<ul> <li>110 = RCIO oscillator: I/O function on RA4/OSC2/CLKOUT pin, RC on RA5/OSC1/CLKIN</li> <li>101 = INTOSC oscillator: CLKOUT function on RA4/OSC2/CLKOUT pin, I/O function on</li> </ul>										
	RA5/OSC1/CLKIN 100 = INTOSCIO oscillator: I/O function on RA4/OSC2/CLKOUT pin, I/O function on										
	RA5/OSC1/CLKIN 011 = EC: I/O function on RA4/OSC2/CLKOUT pin, CLKIN on RA5/OSC1/CLKIN										
	010 = HS o	oscillator: High-s	peed crystal/r	esonator on R	A4/OSC2/CLKC	OUT and RA5/C					
		oscillator: Crystal									
	000 = LP0	ISCHAIOF FOW-DO	wei uivsiai Ol		JLINUUT AHU KA	10/030 I/ULKII	N				

3: When MCLR is asserted in INTOSC or RC mode, the internal clock oscillator is disabled.

## 12.2 Calibration Bits

The 8 MHz internal oscillator is factory calibrated. These calibration values are stored in fuses located in the Calibration Word (2008h). The Calibration Word is not erased when using the specified bulk erase sequence in the *Memory Programming Specification* (DS41284) and thus, does not require reprogramming.

### 12.3 Reset

The PIC16F610/616/16HV610/616 differentiates between various kinds of Reset:

- a) Power-on Reset (POR)
- b) WDT Reset during normal operation
- c) WDT Reset during Sleep
- d) MCLR Reset during normal operation
- e) MCLR Reset during Sleep
- f) Brown-out Reset (BOR)

Some registers are not affected in any Reset condition; their status is unknown on POR and unchanged in any other Reset. Most other registers are reset to a "Reset state" on:

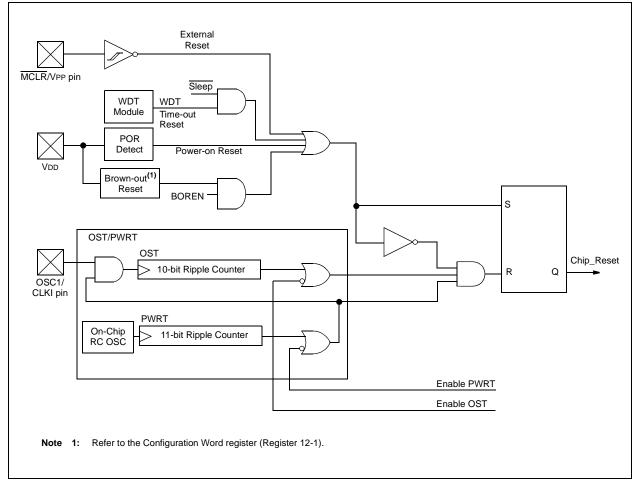
- Power-on Reset
- MCLR Reset
- MCLR Reset during Sleep
- WDT Reset
- Brown-out Reset (BOR)

WDT wake-up does not cause register resets in the same manner as a WDT Reset since wake-up is viewed as the resumption of normal operation. TO and  $\overline{PD}$  bits are set or cleared differently in different Reset situations, as indicated in Table 12-2. Software can use these bits to determine the nature of the Reset. See Table 12-4 for a full description of Reset states of all registers.

A simplified block diagram of the On-Chip Reset Circuit is shown in Figure 12-1.

The MCLR Reset path has a noise filter to detect and ignore small pulses. See **Section 15.0** "**Electrical Specifications**" for pulse-width specifications.

#### FIGURE 12-1: SIMPLIFIED BLOCK DIAGRAM OF ON-CHIP RESET CIRCUIT



### 12.3.1 POWER-ON RESET (POR)

The on-chip POR circuit holds the chip in Reset until VDD has reached a high enough level for proper operation. To take advantage of the POR, simply connect the MCLR pin through a resistor to VDD. This will eliminate external RC components usually needed to create Power-on Reset. A maximum rise time for VDD is required. See **Section 15.0** "**Electrical Specifications**" for details. If the BOR is enabled, the maximum rise time specification does not apply. The BOR circuitry will keep the device in Reset until VDD reaches VBOR (see **Section 12.3.4** "**Brown-out Reset (BOR)**").

Note: The POR circuit does not produce an internal Reset when VDD declines. To reenable the POR, VDD must reach Vss for a minimum of 100 μs.

When the device starts normal operation (exits the Reset condition), device operating parameters (i.e., voltage, frequency, temperature, etc.) must be met to ensure proper operation. If these conditions are not met, the device must be held in Reset until the operating conditions are met.

For additional information, refer to Application Note AN607, *"Power-up Trouble Shooting"* (DS00607).

### 12.3.2 MCLR

PIC16F610/616/16HV610/616 has a noise filter in the MCLR Reset path. The filter will detect and ignore small pulses.

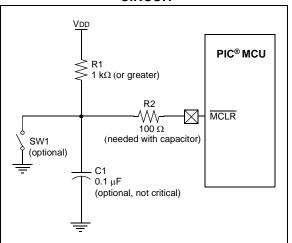
It should be noted that a WDT Reset does not drive  $\overline{\text{MCLR}}$  pin low.

Voltages applied to the MCLR pin that exceed its specification can result in both MCLR Resets and excessive current beyond the device specification during the ESD event. For this reason, Microchip recommends that the MCLR pin no longer be tied directly to VDD. The use of an RC network, as shown in Figure 12-2, is suggested.

An internal  $\overline{\text{MCLR}}$  option is enabled by clearing the  $\overline{\text{MCLRE}}$  bit in the Configuration Word register. When  $\overline{\text{MCLRE}} = 0$ , the Reset signal to the chip is generated internally. When the  $\overline{\text{MCLRE}} = 1$ , the RA3/ $\overline{\text{MCLR}}$  pin becomes an external Reset input. In this mode, the RA3/ $\overline{\text{MCLR}}$  pin has a weak pull-up to VDD.

#### FIGURE 12-2:

#### 



### 12.3.3 POWER-UP TIMER (PWRT)

The Power-up Timer provides a fixed 64 ms (nominal) time-out on power-up only, from POR or Brown-out Reset. The Power-up Timer operates from an internal RC oscillator. For more information, see **Section 3.4** "Internal Clock Modes". The chip is kept in Reset as long as PWRT is active. The PWRT delay allows the VDD to rise to an acceptable level. A Configuration bit, PWRTE, can disable (if set) or enable (if cleared or programmed) the Power-up Timer. The Power-up Timer should be enabled when Brown-out Reset is enabled, although it is not required.

The Power-up Timer delay will vary from chip-to-chip due to:

- VDD variation
- Temperature variation
- · Process variation

See DC parameters for details (Section 15.0 "Electrical Specifications").

Note:	Voltage spikes below Vss at the MCLR
	pin, inducing currents greater than 80 mA,
	may cause latch-up. Thus, a series resis-
	tor of 50-100 $\Omega$ should be used when
	applying a "low" level to the MCLR pin,
	rather than pulling this pin directly to Vss.

### 12.3.4 BROWN-OUT RESET (BOR)

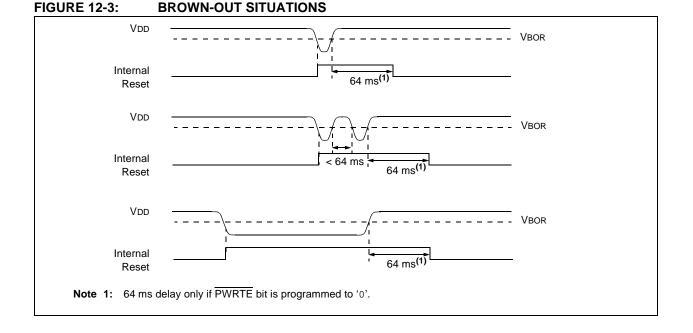
The BOREN0 and BOREN1 bits in the Configuration Word register select one of three BOR modes. Selecting BOREN<1:0> = 10, the BOR is automatically disabled in Sleep to conserve power and enabled on wake-up. See Register 12-1 for the Configuration Word definition.

A brown-out occurs when VDD falls below VBOR for greater than parameter TBOR (see **Section 15.0** "**Electrical Specifications**"). The brown-out condition will reset the device. This will occur regardless of VDD slew rate. A Brown-out Reset may not occur if VDD falls below VBOR for less than parameter TBOR.

On any Reset (Power-on, Brown-out Reset, Watchdog timer, etc.), the chip will remain in Reset until VDD rises above VBOR (see Figure 12-3). If enabled, the Power-up Timer will be invoked by the Reset and keep the chip in Reset an additional 64 ms.

Note: The Power-up Timer is enabled by the <u>PWRTE</u> bit in the Configuration Word register.

If VDD drops below VBOR while the Power-up Timer is running, the chip will go back into a Brown-out Reset and the Power-up Timer will be re-initialized. Once VDD rises above VBOR, the Power-up Timer will execute a 64 ms Reset.



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#### 12.3.5 TIME-OUT SEQUENCE

On power-up, the time-out sequence is as follows:

- PWRT time-out is invoked after POR has expired.
- OST is activated after the PWRT time-out has expired.

The total time-out will vary based on oscillator configuration and PWRTE bit status. For example, in EC mode with PWRTE bit erased (PWRT disabled), there will be no time-out at all. Figure 12-4, Figure 12-5 and Figure 12-6 depict time-out sequences.

Since the time-outs occur from the POR pulse, if  $\overline{\text{MCLR}}$  is kept low long enough, the time-outs will expire. Then, bringing  $\overline{\text{MCLR}}$  high will begin execution immediately (see Figure 12-5). This is useful for testing purposes or to synchronize more than one PIC16F610/616/ 16HV610/616 device operating in parallel.

Table 12-5 shows the Reset conditions for some special registers, while Table 12-4 shows the Reset conditions for all the registers.

#### 12.3.6 POWER CONTROL (PCON) REGISTER

The Power Control register PCON (address 8Eh) has two Status bits to indicate what type of Reset occurred last.

Bit 0 is  $\overrightarrow{BOR}$  (Brown-out).  $\overrightarrow{BOR}$  is unknown on Poweron Reset. It must then be set by the user and checked on subsequent Resets to see if  $\overrightarrow{BOR} = 0$ , indicating that a Brown-out has occurred. The  $\overrightarrow{BOR}$  Status bit is a "don't care" and is not necessarily predictable if the brown-out circuit is disabled (BOREN<1:0> = 00 in the Configuration Word register).

Bit 1 is POR (Power-on Reset). It is a '0' on Power-on Reset and unaffected otherwise. The user must write a '1' to this bit following a Power-on Reset. On a subsequent Reset, if POR is '0', it will indicate that a Power-on Reset has occurred (i.e., VDD may have gone too low).

For more information, see **Section 12.3.4** "**Brown-out Reset (BOR)**".

Oscillator Configuration	Powe	er-up	Brown-o	Wake-up from	
Oscillator Conliguration	<b>PWRTE</b> = 0	<b>PWRTE</b> = 1	<b>PWRTE</b> = 0	<b>PWRTE</b> = 1	Sleep
XT, HS, LP	TPWRT + 1024 • Tosc	1024 • Tosc	TPWRT + 1024 • Tosc	1024 • Tosc	1024 • Tosc
RC, EC, INTOSC	TPWRT		TPWRT		—

#### TABLE 12-1: TIME-OUT IN VARIOUS SITUATIONS

## TABLE 12-2: STATUS/PCON BITS AND THEIR SIGNIFICANCE

POR	BOR	то	PD	Condition
0	х	1	1	Power-on Reset
u	0	1	1	Brown-out Reset
u	u	0	u	WDT Reset
u	u	0	0	WDT Wake-up
u	u	u	u	MCLR Reset during normal operation
u	u	1	0	MCLR Reset during Sleep

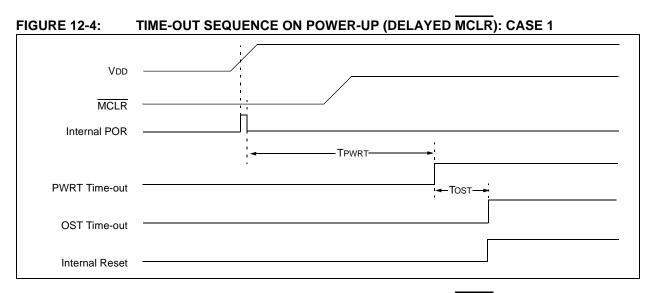
**Legend:** u = unchanged, x = unknown

#### TABLE 12-3: SUMMARY OF REGISTERS ASSOCIATED WITH BROWN-OUT RESET

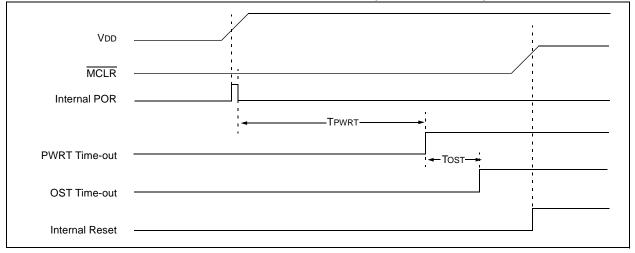
Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Value on POR, BOR	Value on all other Resets <sup>(1)</sup>
PCON	_		_	_			POR	BOR	dd	uu
STATUS	IRP	RP1	RP0	TO	PD	Z	DC	С	0001 1xxx	000q quuu

**Legend:** u = unchanged, x = unknown, - = unimplemented bit, reads as '0', q = value depends on condition. Shaded cells are not used by BOR.

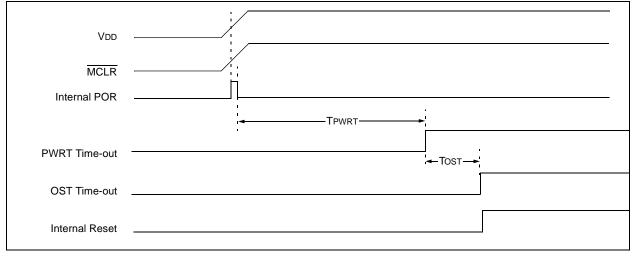
Note 1: Other (non Power-up) Resets include MCLR Reset and Watchdog Timer Reset during normal operation.











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Register	Address	Power-on Reset	MCLR Reset WDT Reset Brown-out Reset <sup>(1)</sup>	Wake-up from Sleep through Interrupt Wake-up from Sleep through WDT Time-out
W	_	xxxx xxxx	uuuu uuuu	uuuu uuuu
INDF	00h/80h	xxxx xxxx	XXXX XXXX	uuuu uuuu
TMR0	01h	xxxx xxxx	uuuu uuuu	uuuu uuuu
PCL	02h/82h	0000 0000	0000 0000	PC + 1 <sup>(3)</sup>
STATUS	03h/83h	0001 1xxx	000q quuu <b>(4)</b>	uuuq quuu <sup>(4)</sup>
FSR	04h/84h	xxxx xxxx	uuuu uuuu	uuuu uuuu
PORTA	05h	x0 x000	u0 u000	uu uuuu
PORTC	07h	xx xx00	uu 00uu	uu uuuu
PCLATH	0Ah/8Ah	0 0000	0 0000	u uuuu
INTCON	0Bh/8Bh	0000 0000	0000 0000	uuuu uuuu <sup>(2)</sup>
PIR1	0Ch	-000 0-00	-000 0-00	-uuu u-uu <b>(2)</b>
TMR1L	0Eh	xxxx xxxx	uuuu uuuu	uuuu uuuu
TMR1H	0Fh	xxxx xxxx	սսսս սսսս	uuuu uuuu
T1CON	10h	0000 0000	uuuu uuuu	-uuu uuuu
TMR2 <sup>(6)</sup>	11h	0000 0000	0000 0000	uuuu uuuu
T2CON <sup>(6)</sup>	12h	-000 0000	-000 0000	-uuu uuuu
CCPR1L <sup>(6)</sup>	13h	xxxx xxxx	uuuu uuuu	uuuu uuuu
CCPR1H <sup>(6)</sup>	14h	xxxx xxxx	uuuu uuuu	uuuu uuuu
CCP1CON <sup>(6)</sup>	15h	0000 0000	0000 0000	uuuu uuuu
PWM1CON <sup>(6)</sup>	16h	0000 0000	0000 0000	uuuu uuuu
ECCPAS <sup>(6)</sup>	17h	0000 0000	0000 0000	uuuu uuuu
VRCON	19h	0000 0000	0000 0000	uuuu uuuu
CM1CON0	1Ah	0000 -000	0000 -000	uuuu –uuu
CM2CON0	1Bh	0000 -000	0000 -000	uuuu -uuu
CM2CON1	1Ch	00-0 0000	00-0 0000	uu-u uuuu
ADRESH <sup>(6)</sup>	1Eh	xxxx xxxx	uuuu uuuu	uuuu uuuu
ADCON0 <sup>(6)</sup>	1Fh	0000 0000	0000 0000	uuuu uuuu
OPTION_REG	81h	1111 1111	1111 1111	uuuu uuuu
TRISA	85h	11 1111	11 1111	uu uuuu
TRISC	87h	11 1111	11 1111	uu uuuu
PIE1	8Ch	-000 0-00	-000 0-00	-uuu u-uu
PCON	8Eh	0x		uu

**Legend:** u = unchanged, x = unknown, - = unimplemented bit, reads as '0', q = value depends on condition.

**Note 1:** If VDD goes too low, Power-on Reset will be activated and registers will be affected differently.

2: One or more bits in INTCON and/or PIR1 will be affected (to cause wake-up).

**3:** When the wake-up is due to an interrupt and the GIE bit is set, the PC is loaded with the interrupt vector (0004h).

- 4: See Table 12-5 for Reset value for specific condition.
- **5**: If Reset was due to brown-out, then bit 0 = 0. All other Resets will cause bit 0 = u.
- 6: PIC16F616/16HV616 only.
- **7:** ANSEL <3:2> For PIC16F616/HV616 only.

Register	Address	Power-on Reset	MCLR Reset WDT Reset (Continued) Brown-out Reset <sup>(1)</sup>	Wake-up from Sleep through Interrupt Wake-up from Sleep through WDT Time-out (Continued)
OSCTUNE	90h	0 0000	u uuuu	u uuuu
ANSEL <sup>(7)</sup>	91h	1111 1111	1111 1111	uuuu uuuu
PR2 <sup>(6)</sup>	92h	1111 1111	1111 1111	1111 1111
WPUA	95h	11 -111	11 -111	uu -uuu
IOCA	96h	00 0000	00 0000	uu uuuu
SRCON0	99h	0000 00-0	0000 00-0	uuuu uu-u
SRCON1	9Ah	00	00	uu
ADRESL <sup>(6)</sup>	9Eh	xxxx xxxx	uuuu uuuu	uuuu uuuu
ADCON1 <sup>(6)</sup>	9Fh	-000	-000	-uuu

## TABLE 12-4: INITIALIZATION CONDITION FOR REGISTERS (CONTINUED)

**Legend:** u = unchanged, x = unknown, - = unimplemented bit, reads as '0', q = value depends on condition.

- Note 1: If VDD goes too low, Power-on Reset will be activated and registers will be affected differently.
  - 2: One or more bits in INTCON and/or PIR1 will be affected (to cause wake-up).
  - **3:** When the wake-up is due to an interrupt and the GIE bit is set, the PC is loaded with the interrupt vector (0004h).
  - 4: See Table 12-5 for Reset value for specific condition.
  - **5:** If Reset was due to brown-out, then bit 0 = 0. All other Resets will cause bit 0 = u.
  - 6: PIC16F616/16HV616 only.
  - 7: ANSEL <3:2> For PIC16F616/HV616 only.

#### TABLE 12-5: INITIALIZATION CONDITION FOR SPECIAL REGISTERS

Condition	Program Counter	Status Register	PCON Register
Power-on Reset	000h	0001 1xxx	0x
MCLR Reset during normal operation	000h	000u uuuu	uu
MCLR Reset during Sleep	000h	0001 Ouuu	uu
WDT Reset	000h	0000 uuuu	uu
WDT Wake-up	PC + 1	uuu0 Ouuu	uu
Brown-out Reset	000h	0001 luuu	10
Interrupt Wake-up from Sleep	PC + 1 <sup>(1)</sup>	uuul 0uuu	uu

**Legend:** u = unchanged, x = unknown, - = unimplemented bit, reads as '0'.

**Note 1:** When the wake-up is due to an interrupt and Global Interrupt Enable bit, GIE, is set, the PC is loaded with the interrupt vector (0004h) after execution of PC + 1.

## 12.4 Interrupts

The PIC16F610/616/16HV610/616 has multiple sources of interrupt:

- External Interrupt RA2/INT
- Timer0 Overflow Interrupt
- PORTA Change Interrupts
- 2 Comparator Interrupts
- A/D Interrupt (PIC16F616/16HV616 only)
- Timer1 Overflow Interrupt
- Timer2 Match Interrupt (PIC16F616/16HV616 only)
- Enhanced CCP Interrupt (PIC16F616/16HV616 only)

The Interrupt Control register (INTCON) and Peripheral Interrupt Request Register 1 (PIR1) record individual interrupt requests in flag bits. The INTCON register also has individual and global interrupt enable bits.

The Global Interrupt Enable bit, GIE of the INTCON register, enables (if set) all unmasked interrupts, or disables (if cleared) all interrupts. Individual interrupts can be disabled through their corresponding enable bits in the INTCON register and PIE1 register. GIE is cleared on Reset.

When an interrupt is serviced, the following actions occur automatically:

- The GIE is cleared to disable any further interrupt.
- The return address is pushed onto the stack.
- The PC is loaded with 0004h.

The Return from Interrupt instruction, RETFIE, exits the interrupt routine, as well as sets the GIE bit, which re-enables unmasked interrupts.

The following interrupt flags are contained in the INT-CON register:

- INT Pin Interrupt
- PORTA Change Interrupt
- Timer0 Overflow Interrupt

The peripheral interrupt flags are contained in the special register, PIR1. The corresponding interrupt enable bit is contained in special register, PIE1.

The following interrupt flags are contained in the PIR1 register:

- A/D Interrupt
- 2 Comparator Interrupts
- Timer1 Overflow Interrupt
- Timer2 Match Interrupt
- Enhanced CCP Interrupt

For external interrupt events, such as the INT pin or PORTA change interrupt, the interrupt latency will be three or four instruction cycles. The exact latency depends upon when the interrupt event occurs (see Figure 12-8). The latency is the same for one or twocycle instructions. Once in the Interrupt Service Routine, the source(s) of the interrupt can be determined by polling the interrupt flag bits. The interrupt flag bit(s) must be cleared in software before re-enabling interrupts to avoid multiple interrupt requests.

- Note 1: Individual interrupt flag bits are set, regardless of the status of their corresponding mask bit or the GIE bit.
  - 2: When an instruction that clears the GIE bit is executed, any interrupts that were pending for execution in the next cycle are ignored. The interrupts, which were ignored, are still pending to be serviced when the GIE bit is set again.

For additional information on Timer1, Timer2, comparators, ADC, Enhanced CCP modules, refer to the respective peripheral section.

#### 12.4.1 RA2/INT INTERRUPT

The external interrupt on the RA2/INT pin is edgetriggered; either on the rising edge if the INTEDG bit of the OPTION register is set, or the falling edge, if the INTEDG bit is clear. When a valid edge appears on the RA2/INT pin, the INTF bit of the INTCON register is set. This interrupt can be disabled by clearing the INTE control bit of the INTCON register. The INTF bit must be cleared by software in the Interrupt Service Routine before re-enabling this interrupt. The RA2/INT interrupt can wake-up the processor from Sleep, if the INTE bit was set prior to going into Sleep. See **Section 12.7** "**Power-Down Mode (Sleep)**" for details on Sleep and Figure 12-9 for timing of wake-up from Sleep through RA2/INT interrupt.

Note: The ANSEL register must be initialized to configure an analog channel as a digital input. Pins configured as analog inputs will read '0' and cannot generate an interrupt.

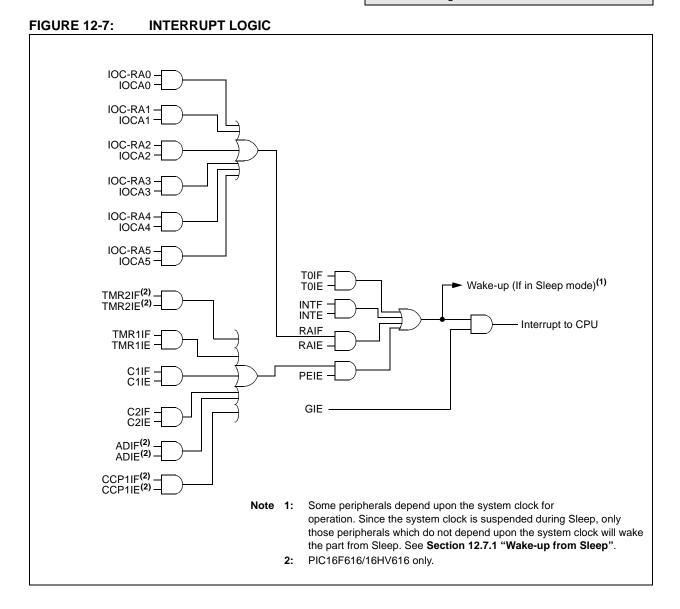
#### 12.4.2 TIMER0 INTERRUPT

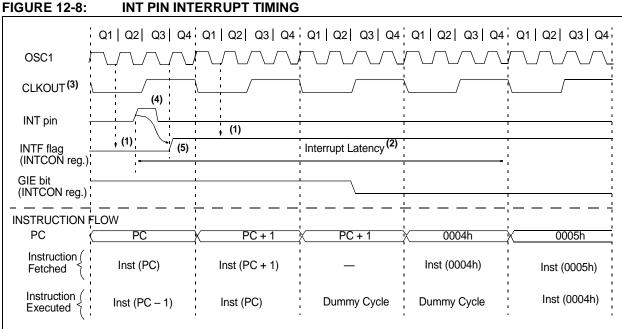
An overflow (FFh  $\rightarrow$  00h) in the TMR0 register will set the T0IF bit of the INTCON register. The interrupt can be enabled/disabled by setting/clearing T0IE bit of the INTCON register. See **Section 5.0 "Timer0 Module"** for operation of the Timer0 module.

#### 12.4.3 PORTA INTERRUPT-ON-CHANGE

An input change on PORTA sets the RAIF bit of the INTCON register. The interrupt can be enabled/ disabled by setting/clearing the RAIE bit of the INTCON register. Plus, individual pins can be configured through the IOCA register.

Note: If a change on the I/O pin should occur when any PORTA operation is being executed, then the RAIF interrupt flag may not get set.





Note 1: INTF flag is sampled here (every Q1).

- 2: Asynchronous interrupt latency = 3-4 TCY. Synchronous latency = 3 TCY, where TCY = instruction cycle time. Latency is the same whether Inst (PC) is a single cycle or a 2-cycle instruction.
- 3: CLKOUT is available only in INTOSC and RC Oscillator modes.
- 4: For minimum width of INT pulse, refer to AC specifications in Section 15.0 "Electrical Specifications".
- 5: INTF is enabled to be set any time during the Q4-Q1 cycles.

#### TABLE 12-6: SUMMARY OF REGISTERS ASSOCIATED WITH INTERRUPTS

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Value on POR, BOR	Value on all other Resets
INTCON	GIE	PEIE	T0IE	INTE	RAIE	T0IF	INTF	RAIF	0000 0000	0000 0000
IOCA	_	_	IOCA5	IOCA4	IOCA3	IOCA2	IOCA1	IOCA0	00 0000	00 0000
PIR1	—	ADIF <sup>(1)</sup>	CCP1IF <sup>(1)</sup>	C2IF	C1IF	_	TMR2IF <sup>(1)</sup>	TMR1IF	-000 0-00	-000 0-00
PIE1	_	ADIE <sup>(1)</sup>	CCP1IE <sup>(1)</sup>	C2IE	C1IE	-	TMR2IE <sup>(1)</sup>	TMR1IE	-000 0-00	-000 0-00

 $\label{eq:legend: Legend: x = unknown, u = unchanged, - = unimplemented read as `0', q = value depends upon condition. Shaded cells are not used by the interrupt module.$ 

**Note 1:** PIC16F616/16HV616 only.

## 12.5 Context Saving During Interrupts

During an interrupt, only the return PC value is saved on the stack. Typically, users may wish to save key registers during an interrupt (e.g., W and STATUS registers). This must be implemented in software.

Temporary holding registers W\_TEMP and STATUS\_TEMP should be placed in the last 16 bytes of GPR (see Figure 2-4). These 16 locations are common to all banks and do not require banking. This makes context save and restore operations simpler. The code shown in Example 12-1 can be used to:

- Store the W register
- Store the STATUS register
- Execute the ISR code
- Restore the Status (and Bank Select Bit register)
- Restore the W register

Note:	The PIC16F610/616/16HV610/616 does			
	not require saving the PCLATH. However,			
	if computed GOTO's are used in both the			
	ISR and the main code, the PCLATH must			
	be saved and restored in the ISR.			

#### EXAMPLE 12-1: SAVING STATUS AND W REGISTERS IN RAM

MOVWF SWAPF	W_TEMP STATUS,W	;Copy W to TEMP register ;Swap status to be saved into W ;Swaps are used because they do not affect the status bits
MOVWF	STATUS_TEMP	;Save status to bank zero STATUS_TEMP register
:(ISR)		;Insert user code here
SWAPF	STATUS_TEMP,W	;Swap STATUS_TEMP register into W ;(sets bank to original state)
MOVWF	STATUS	;Move W into STATUS register
SWAPF	W_TEMP,F	;Swap W_TEMP
SWAPF	W_TEMP,W	;Swap W_TEMP into W

## 12.6 Watchdog Timer (WDT)

The Watchdog Timer is a free running, on-chip RC oscillator, which requires no external components. This RC oscillator is separate from the external RC oscillator of the CLKIN pin and INTOSC. That means that the WDT will run, even if the clock on the OSC1 and OSC2 pins of the device has been stopped (for example, by execution of a SLEEP instruction). During normal operation, a WDT Time-out generates a device Reset. If the device is in Sleep mode, a WDT Time-out causes the device to wake-up and continue with normal operation. The WDT can be permanently disabled by programming the Configuration bit, WDTE, as clear (Section 12.1 "Configuration Bits").

### 12.6.1 WDT PERIOD

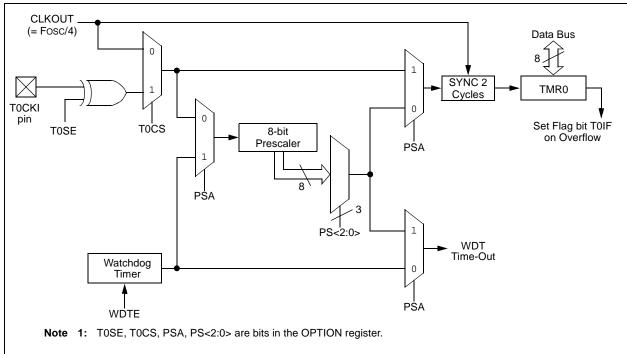
The WDT has a nominal time-out period of 18 ms (with no prescaler). The time-out periods vary with temperature, VDD and process variations from part to part (see Table 15-4, Parameter 31). If longer time-out periods are desired, a prescaler with a division ratio of up to 1:128 can be assigned to the WDT under software control by writing to the OPTION register. Thus, time-out periods up to 2.3 seconds can be realized.

The CLRWDT and SLEEP instructions clear the WDT and the prescaler, if assigned to the WDT, and prevent it from timing out and generating a device Reset.

The  $\overline{\text{TO}}$  bit in the STATUS register will be cleared upon a Watchdog Timer Time-out.

#### 12.6.2 WDT PROGRAMMING CONSIDERATIONS

It should also be taken in account that under worstcase conditions (i.e., VDD = Min., Temperature = Max., Max. WDT prescaler) it may take several seconds before a WDT Time-out occurs.



#### FIGURE 12-2: WATCHDOG TIMER BLOCK DIAGRAM

#### TABLE 12-7: WDT STATUS

Conditions	WDT		
WDTE = 0			
CLRWDT Command	Cleared		
Exit Sleep + System Clock = EXTRC, INTRC, EC			
Exit Sleep + System Clock = XT, HS, LP	Cleared until the end of OST		

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Value on POR, BOR	Value on all other Resets
OPTION_REG	RAPU	INTEDG	TOCS	T0SE	PSA	PS2	PS1	PS0	1111 1111	1111 1111
CONFIG <sup>(1)</sup>	IOSCFS	CP	MCLRE	PWRTE	WDTE	FOSC2	FOSC1	FOSC0	_	

#### TABLE 12-8: SUMMARY OF REGISTERS ASSOCIATED WITH WATCHDOG TIMER

Legend: Shaded cells are not used by the Watchdog Timer.

Note 1: See Register 12-1 for operation of all Configuration Word register bits.

## 12.7 Power-Down Mode (Sleep)

The Power-Down mode is entered by executing a  $\ensuremath{\mathtt{SLEEP}}$  instruction.

If the Watchdog Timer is enabled:

- WDT will be cleared but keeps running.
- PD bit in the STATUS register is cleared.
- TO bit is set.
- Oscillator driver is turned off.
- I/O ports maintain the status they had before SLEEP was executed (driving high, low or high-impedance).

For lowest current consumption in this mode, all I/O pins should be either at VDD or VSS, with no external circuitry drawing current from the I/O pin and the comparators and CVREF should be disabled. I/O pins that are highimpedance inputs should be pulled high or low externally to avoid switching currents caused by floating inputs. The TOCKI input should also be at VDD or VSS for lowest current consumption. The contribution from on-chip pullups on PORTA should be considered.

The  $\overline{\text{MCLR}}$  pin must be at a logic high level.

Note:	It should be noted that a Reset generated
	by a WDT time-out does not drive MCLR
	pin low.

#### 12.7.1 WAKE-UP FROM SLEEP

The device can wake-up from Sleep through one of the following events:

- 1. External Reset input on MCLR pin.
- 2. Watchdog Timer wake-up (if WDT was enabled).
- 3. Interrupt from RA2/INT pin, PORTA change or a peripheral interrupt.

The first event will cause a device Reset. The two latter events are considered a continuation of program execution. The TO and PD bits in the STATUS register can be used to determine the cause of device Reset. The PD bit, which is set on power-up, is cleared when Sleep is invoked. TO bit is cleared if WDT wake-up occurred.

The following peripheral interrupts can wake the device from Sleep:

- 1. Timer1 interrupt. Timer1 must be operating as an asynchronous counter.
- 2. ECCP Capture mode interrupt.
- 3. A/D conversion (when A/D clock source is RC).
- 4. Comparator output changes state.
- 5. Interrupt-on-change.
- 6. External Interrupt from INT pin.

Other peripherals cannot generate interrupts since during Sleep, no on-chip clocks are present.

When the SLEEP instruction is being executed, the next instruction (PC + 1) is prefetched. For the device to wake-up through an interrupt event, the corresponding interrupt enable bit must be set (enabled). Wake-up is regardless of the state of the GIE bit. If the GIE bit is clear (disabled), the device continues execution at the instruction after the SLEEP instruction. If the GIE bit is set (enabled), the device executes the instruction after the SLEEP instruction, then branches to the interrupt address (0004h). In cases where the execution of the instruction following SLEEP is not desirable, the user should have a NOP after the SLEEP instruction.

Note:	If the global interrupts are disabled (GIE is cleared) and any interrupt source has both its interrupt enable bit and the correspond-
	ing interrupt flag bits set, the device will immediately wake-up from Sleep.

The WDT is cleared when the device wakes up from Sleep, regardless of the source of wake-up.

#### 12.7.2 WAKE-UP USING INTERRUPTS

When global interrupts are disabled (GIE cleared) and any interrupt source has both its interrupt enable bit and interrupt flag bit set, one of the following will occur:

- If the interrupt occurs **before** the execution of a SLEEP instruction, the SLEEP instruction will complete as a NOP. Therefore, the WDT and WDT prescaler and postscaler (if enabled) will not be cleared, the TO bit will not be set and the PD bit will not be cleared.
- If the interrupt occurs during or after the execution of a SLEEP instruction, the device will Immediately wake-up from Sleep. The SLEEP instruction is executed. Therefore, the WDT and WDT prescaler and postscaler (if enabled) will be cleared, the TO bit will be set and the PD bit will be cleared.

Even if the flag bits were checked before executing a SLEEP instruction, it may be possible for flag bits to become set before the SLEEP instruction completes. To determine whether a SLEEP instruction executed, test the PD bit. If the PD bit is set, the SLEEP instruction was executed as a NOP.

To ensure that the WDT is cleared, a CLRWDT instruction should be executed before a SLEEP instruction. See Figure 12-9 for more details.

FIGURE 12-9:	WAKE-UP FROM S	SLEEP THRO	UGH INTER	RUPT		
; Q1 Q2 OSC1/¯ CLKOUT <sup>(4)</sup> ; INT pin	:  Q3   Q4 ; Q1   Q2   Q3   Q4 ; Q ////////////////////////////////////		Q1 Q2 Q3 Q4 	( Q1 Q2  Q3 Q4; /~_~/	Q1 Q2 Q3 Q4;( 	Q1 Q2 Q3 Q4 \/
INTF flag (INTCON reg.)			Interrupt Laten	<sub>Cy</sub> (3)		
GIE bit (INTCON reg.)		rocessor in Sleep				
Instruction Flow PC X F	PC { PC+1 }	PC + 2	X PC + 2	PC+2 X	0004h X	0005h
	c) = Sleep Inst(PC + 1)		Inst(PC + 2)		Inst(0004h)	Inst(0005h)
Instruction { Inst(	PC – 1) Sleep		Inst(PC + 1)	Dummy Cycle	Dummy Cycle	Inst(0004h)
2: Tost = 1	or LP Oscillator mode assumed 1024 Tosc (drawing not to scale ' assumed. In this case after w	e). This delay does				in-line.

4: CLKOUT is not available in XT, HS, LP or EC Oscillator modes, but shown here for timing reference.

## 12.8 Code Protection

If the code protection bit(s) have not been programmed, the on-chip program memory can be read out using  $ICSP^{TM}$  for verification purposes.

Note:	The entire Flash program memory will be				
	erased when the code protection is turned				
	off. See the Memory Programming				
	Specification (DS41284) for more				
	information.				

## 12.9 ID Locations

Four memory locations (2000h-2003h) are designated as ID locations where the user can store checksum or other code identification numbers. These locations are not accessible during normal execution but are readable and writable during Program/Verify mode. Only the Least Significant 7 bits of the ID locations are used.

#### 12.10 In-Circuit Serial Programming™

The PIC16F610/616/16HV610/616 microcontrollers can be serially programmed while in the end application circuit. This is simply done with five connections for:

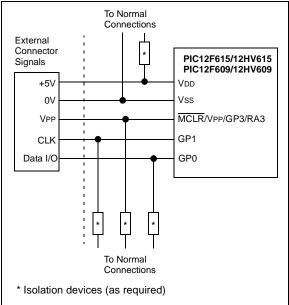
- clock
- data
- power
- ground
- programming voltage

This allows customers to manufacture boards with unprogrammed devices and then program the microcontroller just before shipping the product. This also allows the most recent firmware or a custom firmware to be programmed.

The device is placed into a Program/Verify mode by holding the RA0 and RA1 pins low, while raising the MCLR (VPP) pin from VIL to VIHH. See the *Memory Programming Specification* (DS41284) for more information. RA0 becomes the programming data and RA1 becomes the programming clock. Both RA0 and RA1 are Schmitt Trigger inputs in Program/Verify mode.

A typical In-Circuit Serial Programming connection is shown in Figure 12-10.

#### FIGURE 12-10: TYPICAL IN-CIRCUIT SERIAL PROGRAMMING™ CONNECTION



Note:	To erase the device VDD must be above				
	the Bulk Erase VDD minimum given in the				
	Memory Programming Specification				
	(DS41284)				

## 12.11 In-Circuit Debugger

Since in-circuit debugging requires access to three pins, MPLAB<sup>®</sup> ICD 2 development with an 14-pin device is not practical. A special 28-pin PIC16F610/616/ 16HV610/616 ICD device is used with MPLAB ICD 2 to provide separate clock, data and MCLR pins and frees all normally available pins to the user.

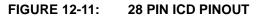
A special debugging adapter allows the ICD device to be used in place of a PIC16F610/616/16HV610/616 device. The debugging adapter is the only source of the ICD device.

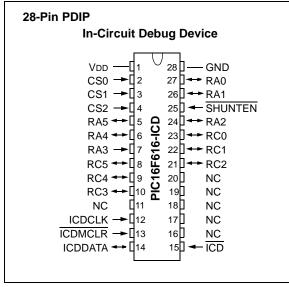
When the ICD pin on the PIC16F610/616/16HV610/ 616 ICD device is held low, the In-Circuit Debugger functionality is enabled. This function allows simple debugging functions when used with MPLAB ICD 2. When the microcontroller has this feature enabled, some of the resources are not available for general use. Table 12-9 shows which features are consumed by the background debugger.

TABLE 12-9: DEBUGGER RESOURCES
--------------------------------

Resource	Description
I/O pins	ICDCLK, ICDDATA
Stack	1 level
Program Memory	Address 0h must be NOP 700h-7FFh

For more information, see "*MPLAB*<sup>®</sup> *ICD 2 In-Circuit Debugger User's Guide*" (DS51331), available on Microchip's web site (www.microchip.com).





NOTES:

## 13.0 INSTRUCTION SET SUMMARY

The PIC16F610/616/16HV610/616 instruction set is highly orthogonal and is comprised of three basic categories:

- Byte-oriented operations
- Bit-oriented operations
- Literal and control operations

Each PIC16 instruction is a 14-bit word divided into an **opcode**, which specifies the instruction type and one or more **operands**, which further specify the operation of the instruction. The formats for each of the categories is presented in Figure 13-1, while the various opcode fields are summarized in Table 13-1.

Table 13-2 lists the instructions recognized by the MPASM<sup>TM</sup> assembler.

For **byte-oriented** instructions, 'f' represents a file register designator and 'd' represents a destination designator. The file register designator specifies which file register is to be used by the instruction.

The destination designator specifies where the result of the operation is to be placed. If 'd' is zero, the result is placed in the W register. If 'd' is one, the result is placed in the file register specified in the instruction.

For **bit-oriented** instructions, 'b' represents a bit field designator, which selects the bit affected by the operation, while 'f' represents the address of the file in which the bit is located.

For **literal and control** operations, 'k' represents an 8-bit or 11-bit constant, or literal value.

One instruction cycle consists of four oscillator periods; for an oscillator frequency of 4 MHz, this gives a normal instruction execution time of 1  $\mu$ s. All instructions are executed within a single instruction cycle, unless a conditional test is true, or the program counter is changed as a result of an instruction. When this occurs, the execution takes two instruction cycles, with the second cycle executed as a NOP.

All instruction examples use the format '0xhh' to represent a hexadecimal number, where 'h' signifies a hexadecimal digit.

## 13.1 Read-Modify-Write Operations

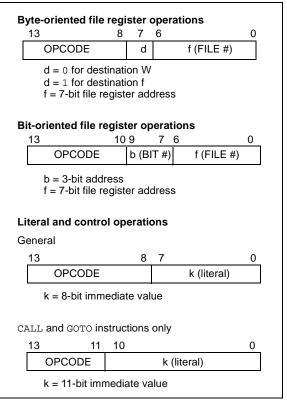
Any instruction that specifies a file register as part of the instruction performs a Read-Modify-Write (RMW) operation. The register is read, the data is modified, and the result is stored according to either the instruction or the destination designator 'd'. A read operation is performed on a register even if the instruction writes to that register.

For example, a CLRF PORTA instruction will read PORTA, clear all the data bits, then write the result back to PORTA. This example would have the unintended consequence of clearing the condition that set the RAIF flag.

#### TABLE 13-1: OPCODE FIELD DESCRIPTIONS

Field	Description
f	Register file address (0x00 to 0x7F)
W	Working register (accumulator)
b	Bit address within an 8-bit file register
k	Literal field, constant data or label
x	Don't care location (= $0$ or 1). The assembler will generate code with x = $0$ . It is the recommended form of use for compatibility with all Microchip software tools.
d	Destination select; $d = 0$ : store result in W, d = 1: store result in file register f. Default is $d = 1$ .
PC	Program Counter
TO	Time-out bit
С	Carry bit
DC	Digit carry bit
Z	Zero bit
PD	Power-down bit

#### FIGURE 13-1: GENERAL FORMAT FOR INSTRUCTIONS



Mnemonic, Operands		Description			14-Bit	Opcode	Status		
				MSb			LSb	Affected	Notes
	BYTE-ORIENTED FILE REGISTER OPERATIONS								
ADDWF	f, d	Add W and f	1	00	0111	dfff	ffff	C, DC, Z	1, 2
ANDWF	f, d	AND W with f	1	00	0101	dfff	ffff	Z	1, 2
CLRF	f	Clear f	1	00	0001	lfff	ffff	Z	2
CLRW	-	Clear W	1	00	0001	0xxx	xxxx	Z	
COMF	f, d	Complement f	1	00	1001	dfff	ffff	Z	1, 2
DECF	f, d	Decrement f	1	00	0011	dfff	ffff	Z	1, 2
DECFSZ	f, d	Decrement f, Skip if 0	1 <b>(2)</b>	00	1011	dfff	ffff		1, 2, 3
INCF	f, d	Increment f	1	00	1010	dfff	ffff	Z	1, 2
INCFSZ	f, d	Increment f, Skip if 0	1(2)	00	1111	dfff	ffff		1, 2, 3
IORWF	f, d	Inclusive OR W with f	1	00	0100	dfff	ffff	Z	1, 2
MOVF	f, d	Move f	1	00	1000	dfff	ffff	Z	1, 2
MOVWF	f	Move W to f	1	00	0000	lfff	ffff		,
NOP	_	No Operation	1	00	0000	0xx0	0000		
RLF	f, d	Rotate Left f through Carry	1	00	1101	dfff	ffff	С	1, 2
RRF	f, d	Rotate Right f through Carry	1	00	1100	dfff	ffff	C	1, 2
SUBWF	f, d	Subtract W from f	1	00	0010		ffff	C, DC, Z	1, 2
SWAPF	f, d	Swap nibbles in f	1	00	1110		ffff	-, -,	1, 2
XORWF	f, d	Exclusive OR W with f	1	00	0110	dfff	ffff	Z	1, 2
	BIT-ORIENTED FILE REGISTER OPERATIONS								
BCF	f, b	Bit Clear f	1	01	00bb	bfff	ffff		1, 2
BSF	f, b	Bit Set f	1	01		bfff			1, 2
BTFSC	f, b	Bit Test f, Skip if Clear	1 (2)	01		bfff			3
BTFSS	f, b	Bit Test f, Skip if Set	1 (2)	01	11bb	bfff	ffff		3
	., 2				1100	~~~~			•
ADDLW	k	Add literal and W	1	11	111x	kkkk	kkkk	C, DC, Z	
ANDLW	k	AND literal with W	1	11	1001		kkkk	Z	
CALL	k	Call Subroutine	2	10		kkkk		_	
CLRWDT	_	Clear Watchdog Timer	1	00	0000	0110	0100	TO, PD	
GOTO	k	Go to address	2	10		kkkk			
IORLW	k	Inclusive OR literal with W	1	11	1000	kkkk		z	
MOVLW	k	Move literal to W	1	11		kkkk			
RETFIE	_	Return from interrupt	2	00	0000	0000	1001		
RETLW	k	Return with literal in W	2	11		kkkk			
RETURN	_	Return from Subroutine	2	00	0000	0000	1000		
SLEEP	_	Go into Standby mode	1	00	0000	0110	0011	TO, PD	
SUBLW	k	Subtract W from literal	1	11			kkkk	C, DC, Z	
XORLW	k	Exclusive OR literal with W	1	11	1010	kkkk		Z U, DO, Z	
	iv.				T0T0	17171717	17171717	<u> </u>	

#### TABLE 13-2: PIC16F610/616/16HV610/616 INSTRUCTION SET

**Note 1:** When an I/O register is modified as a function of itself (e.g., MOVF PORTA, 1), the value used will be that value present on the pins themselves. For example, if the data latch is '1' for a pin configured as input and is driven low by an external device, the data will be written back with a '0'.

2: If this instruction is executed on the TMR0 register (and where applicable, d = 1), the prescaler will be cleared if assigned to the Timer0 module.

**3:** If the Program Counter (PC) is modified, or a conditional test is true, the instruction requires two cycles. The second cycle is executed as a NOP.

ADDLW	Add literal and W
Syntax:	[ <i>label</i> ] ADDLW k
Operands:	$0 \leq k \leq 255$
Operation:	$(W) + k \to (W)$
Status Affected:	C, DC, Z
Description:	The contents of the W register are added to the eight-bit literal 'k' and the result is placed in the W register.

13.2	Instruction	Descriptions
------	-------------	--------------

BCF	Bit Clear f
Syntax:	[ label ] BCF f,b
Operands:	$\begin{array}{l} 0 \leq f \leq 127 \\ 0 \leq b \leq 7 \end{array}$
Operation:	$0 \rightarrow (f < b >)$
Status Affected:	None
Description:	Bit 'b' in register 'f' is cleared.

ADDWF	Add W and f
Syntax:	[ label ] ADDWF f,d
Operands:	$\begin{array}{l} 0 \leq f \leq 127 \\ d \in [0,1] \end{array}$
Operation:	(W) + (f) $\rightarrow$ (destination)
Status Affected:	C, DC, Z
Description:	Add the contents of the W register with register 'f'. If 'd' is '0', the result is stored in the W register. If 'd' is '1', the result is stored back in register 'f'.

BSF	Bit Set f
Syntax:	[ <i>label</i> ] BSF f,b
Operands:	$\begin{array}{l} 0 \leq f \leq 127 \\ 0 \leq b \leq 7 \end{array}$
Operation:	$1 \rightarrow (f < b >)$
Status Affected:	None
Description:	Bit 'b' in register 'f' is set.

ANDLW	AND literal with W
Syntax:	[ <i>label</i> ] ANDLW k
Operands:	$0 \leq k \leq 255$
Operation:	(W) .AND. (k) $\rightarrow$ (W)
Status Affected:	Z
Description:	The contents of W register are AND'ed with the eight-bit literal 'k'. The result is placed in the W register.

BTFSC	Bit Test f, Skip if Clear
Syntax:	[ label ] BTFSC f,b
Operands:	$\begin{array}{l} 0 \leq f \leq 127 \\ 0 \leq b \leq 7 \end{array}$
Operation:	skip if (f <b>) = <math>0</math></b>
Status Affected:	None
Description:	If bit 'b' in register 'f' is '1', the next instruction is executed. If bit 'b' in register 'f' is '0', the next instruction is discarded, and a NOP is executed instead, making this a two-cycle instruction.

ANDWF	AND W with f
Syntax:	[ label ] ANDWF f,d
Operands:	$\begin{array}{l} 0 \leq f \leq 127 \\ d \in [0,1] \end{array}$
Operation:	(W) .AND. (f) $\rightarrow$ (destination)
Status Affected:	Z
Description:	AND the W register with register 'f'. If 'd' is '0', the result is stored in the W register. If 'd' is '1', the result is stored back in register 'f'.

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BTFSS	Bit Test f, Skip if Set
Syntax:	[ label ] BTFSS f,b
Operands:	$\begin{array}{l} 0 \leq f \leq 127 \\ 0 \leq b < 7 \end{array}$
Operation:	skip if (f <b>) = <math>1</math></b>
Status Affected:	None
Description:	If bit 'b' in register 'f' is '0', the next instruction is executed. If bit 'b' is '1', then the next instruction is discarded and a NOP is executed instead, making this a two-cycle instruction.

CLRWDT	Clear Watchdog Timer
Syntax:	[label] CLRWDT
Operands:	None
Operation:	$00h \rightarrow WDT$ $0 \rightarrow WDT \text{ prescaler,}$ $1 \rightarrow \overline{TO}$ $1 \rightarrow \overline{PD}$
Status Affected:	TO, PD
Description:	CLRWDT instruction resets the Watchdog Timer. It also resets the prescaler of the WDT. Status bits TO and PD are set.

CALL	Call Subroutine
Syntax:	[ <i>label</i> ] CALL k
Operands:	$0 \leq k \leq 2047$
Operation:	$\begin{array}{l} (PC)+1 \rightarrow TOS, \\ k \rightarrow PC < 10:0>, \\ (PCLATH < 4:3>) \rightarrow PC < 12:11> \end{array}$
Status Affected:	None
Description:	Call Subroutine. First, return address (PC + 1) is pushed onto the stack. The eleven-bit immediate address is loaded into PC bits <10:0>. The upper bits of the PC are loaded from PCLATH. CALL is a two-cycle instruction.

COMF	Complement f
Syntax:	[label] COMF f,d
Operands:	$\begin{array}{l} 0\leq f\leq 127\\ d\in [0,1] \end{array}$
Operation:	$(\overline{f}) \rightarrow (destination)$
Status Affected:	Z
Description:	The contents of register 'f' are complemented. If 'd' is '0', the result is stored in W. If 'd' is '1', the result is stored back in register 'f'.

CLRF	Clear f
Syntax:	[label]CLRF f
Operands:	$0 \le f \le 127$
Operation:	$\begin{array}{l} \text{O0h} \rightarrow (\text{f}) \\ 1 \rightarrow \text{Z} \end{array}$
Status Affected:	Z
Description:	The contents of register 'f' are cleared and the Z bit is set.

-	
Syntax:	[ label ] DECF f,d
Operands:	$\begin{array}{l} 0 \leq f \leq 127 \\ d \in [0,1] \end{array}$
Operation:	(f) - 1 $\rightarrow$ (destination)
Status Affected:	Z
Description:	Decrement register 'f'. If 'd' is '0', the result is stored in the W register. If 'd' is '1', the result is stored back in register 'f'.

Decrement f

DECF

CLRW	Clear W
Syntax:	[label] CLRW
Operands:	None
Operation:	$\begin{array}{l} \text{00h} \rightarrow (\text{W}) \\ 1 \rightarrow \text{Z} \end{array}$
Status Affected:	Z
Description:	W register is cleared. Zero bit (Z) is set.

DECFSZ	Decrement f, Skip if 0
Syntax:	[label] DECFSZ f,d
Operands:	$\begin{array}{l} 0 \leq f \leq 127 \\ d \in [0,1] \end{array}$
Operation:	(f) - 1 $\rightarrow$ (destination); skip if result = 0
Status Affected:	None
Description:	The contents of register 'f' are decremented. If 'd' is '0', the result is placed in the W register. If 'd' is '1', the result is placed back in register 'f'. If the result is '1', the next instruction is executed. If the result is '0', then a NOP is executed instead, making it a two-cycle instruction.

INCFSZ	Increment f, Skip if 0
Syntax:	[ <i>label</i> ] INCFSZ f,d
Operands:	$\begin{array}{l} 0 \leq f \leq 127 \\ d \in [0,1] \end{array}$
Operation:	(f) + 1 $\rightarrow$ (destination), skip if result = 0
Status Affected:	None
Description:	The contents of register 'f' are incremented. If 'd' is '0', the result is placed in the W register. If 'd' is '1', the result is placed back in register 'f'. If the result is '1', the next instruction is executed. If the result is '0', a NOP is executed instead, making it a two-cycle instruction.

GOTO	Unconditional Branch
Syntax:	[ <i>label</i> ] GOTO k
Operands:	$0 \leq k \leq 2047$
Operation:	$k \rightarrow PC<10:0>$ PCLATH<4:3> $\rightarrow$ PC<12:11>
Status Affected:	None
Description:	GOTO is an unconditional branch. The eleven-bit immediate value is loaded into PC bits <10:0>. The upper bits of PC are loaded from PCLATH<4:3>. GOTO is a two-cycle instruction.

IORLW	Inclusive OR literal with W
Syntax:	[ <i>label</i> ] IORLW k
Operands:	$0 \leq k \leq 255$
Operation:	(W) .OR. $k \rightarrow$ (W)
Status Affected:	Z
Description:	The contents of the W register are OR'ed with the eight-bit literal 'k'. The result is placed in the W register.

INCF	Increment f
Syntax:	[label] INCF f,d
Operands:	$\begin{array}{l} 0\leq f\leq 127\\ d\in \left[0,1\right] \end{array}$
Operation:	(f) + 1 $\rightarrow$ (destination)
Status Affected:	Z
Description:	The contents of register 'f' are incremented. If 'd' is '0', the result is placed in the W register. If 'd' is '1', the result is placed back in register 'f'.

IORWF	Inclusive OR W with f
Syntax:	[ <i>label</i> ] IORWF f,d
Operands:	$\begin{array}{l} 0 \leq f \leq 127 \\ d \in [0,1] \end{array}$
Operation:	(W) .OR. (f) $\rightarrow$ (destination)
Status Affected:	Z
Description:	Inclusive OR the W register with register 'f'. If 'd' is '0', the result is placed in the W register. If 'd' is '1', the result is placed back in register 'f'.

MOVF	Move f
Syntax:	[ <i>label</i> ] MOVF f,d
Operands:	$\begin{array}{l} 0 \leq f \leq 127 \\ d \in \left[0,1\right] \end{array}$
Operation:	(f) $\rightarrow$ (dest)
Status Affected:	Z
Description:	The contents of register 'f' is moved to a destination dependent upon the status of 'd'. If $d = 0$ , destination is W register. If $d = 1$ , the destination is file register 'f' itself. $d = 1$ is useful to test a file register since status flag Z is affected.
Words:	1
Cycles:	1
Example:	MOVF FSR, 0
	After Instruction W = value in FSR register Z = 1

MOVWF	Move W to f
Syntax:	[ label ] MOVWF f
Operands:	$0 \le f \le 127$
Operation:	$(W) \rightarrow (f)$
Status Affected:	None
Description:	Move data from W register to register 'f'.
Words:	1
Cycles:	1
Example:	MOVW OPTION F
	Before Instruction OPTION = 0xFF
	W = 0x4F After Instruction
	OPTION = 0x4F
	W = 0x4F

MOVLW	Move literal to W
Syntax:	[ <i>label</i> ] MOVLW k
Operands:	$0 \le k \le 255$
Operation:	$k \rightarrow (W)$
Status Affected:	None
Description:	The eight-bit literal 'k' is loaded into W register. The "don't cares" will assemble as '0's.
Words:	1
Cycles:	1
Example:	MOVLW 0x5A
	After Instruction W = 0x5A

NOP	No Operation
Syntax:	[label] NOP
Operands:	None
Operation:	No operation
Status Affected:	None
Description:	No operation.
Words:	1
Cycles:	1
Example:	NOP

RETFIE	Return from Interrupt		
Syntax:	[label] RETFIE		
Operands:	None		
Operation:	$\begin{array}{l} TOS \to PC, \\ 1 \to GIE \end{array}$		
Status Affected:	None		
Description:	Return from Interrupt. Stack is POPed and Top-of-Stack (TOS) is loaded in the PC. Interrupts are enabled by setting Global Interrupt Enable bit, GIE (INTCON<7>). This is a two-cycle instruction.		
Words:	1		
Cycles:	2		
Example:	RETFIE		
	After Interrupt PC = TOS GIE = 1		

RETLW	Return with literal in W		
Syntax:	[ <i>label</i> ] RETLW k		
Operands:	$0 \le k \le 255$		
Operation:	$k \rightarrow (W);$ TOS $\rightarrow PC$		
Status Affected:	None		
Description:	The W register is loaded with the eight-bit literal 'k'. The program counter is loaded from the top of the stack (the return address). This is a two-cycle instruction.		
Words:	1		
Cycles:	2		
Example:	CALL TABLE;W contains ;table offset ;value GOTO DONE		
TABLE	•		
DONE	• ADDWF PC ;W = offset RETLW k1 ;Begin table RETLW k2 ; • • RETLW kn ;End of table Before Instruction W = 0x07		
	After Instruction		
RETURN	W = value of k8 Return from Subroutine		
Syntax:	[label] RETURN		
Operands:	None		
Operation:	$TOS \rightarrow PC$		
Status Affected:	None		
Description:	Return from subroutine. The stack is POPed and the top of the stack (TOS) is loaded into the program counter. This is a two-cycle instruction.		

Rotate Left f through Carry		
[label] RLF f,d		
$\begin{array}{l} 0 \leq f \leq 127 \\ d \in [0,1] \end{array}$		
See description below		
С		
The contents of register 'f' are rotated one bit to the left through the Carry flag. If 'd' is '0', the result is placed in the W register. If 'd' is '1', the result is stored back in register 'f'. -C - Register f		
1		
1		
RLF REG1,0		
Before Instruction		
REG1 = 1110 0110		
C = 0		
C = 0 After Instruction		
C = 0		

SLEEP	Enter Sleep mode
Syntax:	[label] SLEEP
Operands:	None
Operation:	$\begin{array}{l} \text{00h} \rightarrow \text{WDT,} \\ 0 \rightarrow \text{WDT prescaler,} \\ 1 \rightarrow \overline{\text{TO}}, \\ 0 \rightarrow \text{PD} \end{array}$
Status Affected:	TO, PD
Description:	The power-down Status bit, $\overline{\text{PD}}$ is cleared. Time-out Status bit, $\overline{\text{TO}}$ is set. Watchdog Timer and its prescaler are cleared. The processor is put into Sleep mode with the oscillator stopped.

label] RRF f,d	
$\leq f \leq 127$ $\in [0,1]$	
See description below	
С	
C The contents of register 'f' are rotated one bit to the right through the Carry flag. If 'd' is '0', the result is placed in the W register. If 'd' is '1', the result is placed back in register 'f'.	

SUBLW	Subtract W from literal		
Syntax:	[ <i>label</i> ] SUBLW k		
Operands:	$0 \le k \le 255$		
Operation:	$k \text{ - } (W) \to (W)$		
Status Affected:	C, DC, Z		
Description:	The W register is subtracted (2's complement method) from the eight-bit literal 'k'. The result is placed in the W register.		
	Result	Condition	
	<b>C</b> 0	M > k	

Result	Condition
<b>C</b> = 0	W > k
<b>C</b> = 1	$W \leq k$
DC = 0	W<3:0> > k<3:0>
<b>DC</b> = 1	W<3:0> ≤ k<3:0>

SUBWF	Subtract W from f		
Syntax:	[ <i>label</i> ] SUBWF f,d		
Operands:	$\begin{array}{l} 0 \leq f \leq 127 \\ d  \in  [0,1] \end{array}$		
Operation:	(f) - (W) $\rightarrow$ (destination)		
Status Affected:	C, DC, Z		
Description:	Subtract (2's complement method) W register from register 'f'. If 'd' is '0', the result is stored in the W register. If 'd' is '1', the result is stored back in register 'f'.		
	<b>C</b> = 0	W > f	
	<b>C</b> = 1	$W \leq f$	
	DC = 0	W<3:0> > f<3:0>	

**DC** = 1

 $W < 3:0 > \le f < 3:0 >$ 

XORWF	Exclusive OR W with f	
Syntax:	[ <i>label</i> ] XORWF f,d	
Operands:	$\begin{array}{l} 0 \leq f \leq 127 \\ d \in [0,1] \end{array}$	
Operation:	(W) .XOR. (F.) $\rightarrow$ (destination)	
Status Affected:	Z	
Description:	Exclusive OR the contents of the W register with register 'f'. If 'd' is '0', the result is stored in the W register. If 'd' is '1', the result is stored back in register 'f'.	

SWAPF	Swap Nibbles in f
Syntax:	[ <i>label</i> ] SWAPF f,d
Operands:	$\begin{array}{l} 0 \leq f \leq 127 \\ d \in [0,1] \end{array}$
Operation:	$(f<3:0>) \rightarrow (destination<7:4>),$ $(f<7:4>) \rightarrow (destination<3:0>)$
Status Affected:	None
Description:	The upper and lower nibbles of register 'f' are exchanged. If 'd' is '0', the result is placed in the W register. If 'd' is '1', the result is placed in register 'f'.

XORLW	Exclusive OR literal with W	
Syntax:	[label] XORLW k	
Operands:	$0 \le k \le 255$	
Operation:	(W) .XOR. $k \rightarrow (W)$	
Status Affected:	Z	
Description:	The contents of the W register are XOR'ed with the eight-bit literal 'k'. The result is placed in the W register.	

NOTES:

## 14.0 DEVELOPMENT SUPPORT

The PIC<sup>®</sup> microcontrollers and dsPIC<sup>®</sup> digital signal controllers are supported with a full range of software and hardware development tools:

- Integrated Development Environment
- MPLAB<sup>®</sup> IDE Software
- Compilers/Assemblers/Linkers
  - MPLAB C Compiler for Various Device Families
  - HI-TECH C for Various Device Families
  - MPASM<sup>™</sup> Assembler
  - MPLINK<sup>™</sup> Object Linker/ MPLIB<sup>™</sup> Object Librarian
  - MPLAB Assembler/Linker/Librarian for Various Device Families
- Simulators
  - MPLAB SIM Software Simulator
- Emulators
  - MPLAB REAL ICE™ In-Circuit Emulator
- In-Circuit Debuggers
  - MPLAB ICD 3
  - PICkit<sup>™</sup> 3 Debug Express
- Device Programmers
  - PICkit<sup>™</sup> 2 Programmer
  - MPLAB PM3 Device Programmer
- Low-Cost Demonstration/Development Boards, Evaluation Kits, and Starter Kits

## 14.1 MPLAB Integrated Development Environment Software

The MPLAB IDE software brings an ease of software development previously unseen in the 8/16/32-bit microcontroller market. The MPLAB IDE is a Windows<sup>®</sup> operating system-based application that contains:

- A single graphical interface to all debugging tools
  - Simulator
  - Programmer (sold separately)
  - In-Circuit Emulator (sold separately)
  - In-Circuit Debugger (sold separately)
- · A full-featured editor with color-coded context
- A multiple project manager
- Customizable data windows with direct edit of contents
- High-level source code debugging
- Mouse over variable inspection
- Drag and drop variables from source to watch windows
- Extensive on-line help
- Integration of select third party tools, such as IAR C Compilers

The MPLAB IDE allows you to:

- Edit your source files (either C or assembly)
- One-touch compile or assemble, and download to emulator and simulator tools (automatically updates all project information)
- Debug using:
  - Source files (C or assembly)
  - Mixed C and assembly
  - Machine code

MPLAB IDE supports multiple debugging tools in a single development paradigm, from the cost-effective simulators, through low-cost in-circuit debuggers, to full-featured emulators. This eliminates the learning curve when upgrading to tools with increased flexibility and power.

## 14.2 MPLAB C Compilers for Various Device Families

The MPLAB C Compiler code development systems are complete ANSI C compilers for Microchip's PIC18, PIC24 and PIC32 families of microcontrollers and the dsPIC30 and dsPIC33 families of digital signal controllers. These compilers provide powerful integration capabilities, superior code optimization and ease of use.

For easy source level debugging, the compilers provide symbol information that is optimized to the MPLAB IDE debugger.

## 14.3 HI-TECH C for Various Device Families

The HI-TECH C Compiler code development systems are complete ANSI C compilers for Microchip's PIC family of microcontrollers and the dsPIC family of digital signal controllers. These compilers provide powerful integration capabilities, omniscient code generation and ease of use.

For easy source level debugging, the compilers provide symbol information that is optimized to the MPLAB IDE debugger.

The compilers include a macro assembler, linker, preprocessor, and one-step driver, and can run on multiple platforms.

## 14.4 MPASM Assembler

The MPASM Assembler is a full-featured, universal macro assembler for PIC10/12/16/18 MCUs.

The MPASM Assembler generates relocatable object files for the MPLINK Object Linker, Intel<sup>®</sup> standard HEX files, MAP files to detail memory usage and symbol reference, absolute LST files that contain source lines and generated machine code and COFF files for debugging.

The MPASM Assembler features include:

- Integration into MPLAB IDE projects
- User-defined macros to streamline assembly code
- Conditional assembly for multi-purpose source files
- Directives that allow complete control over the assembly process

### 14.5 MPLINK Object Linker/ MPLIB Object Librarian

The MPLINK Object Linker combines relocatable objects created by the MPASM Assembler and the MPLAB C18 C Compiler. It can link relocatable objects from precompiled libraries, using directives from a linker script.

The MPLIB Object Librarian manages the creation and modification of library files of precompiled code. When a routine from a library is called from a source file, only the modules that contain that routine will be linked in with the application. This allows large libraries to be used efficiently in many different applications.

The object linker/library features include:

- Efficient linking of single libraries instead of many smaller files
- Enhanced code maintainability by grouping related modules together
- Flexible creation of libraries with easy module listing, replacement, deletion and extraction

## 14.6 MPLAB Assembler, Linker and Librarian for Various Device Families

MPLAB Assembler produces relocatable machine code from symbolic assembly language for PIC24, PIC32 and dsPIC devices. MPLAB C Compiler uses the assembler to produce its object file. The assembler generates relocatable object files that can then be archived or linked with other relocatable object files and archives to create an executable file. Notable features of the assembler include:

- · Support for the entire device instruction set
- · Support for fixed-point and floating-point data
- Command line interface
- · Rich directive set
- Flexible macro language
- MPLAB IDE compatibility

## 14.7 MPLAB SIM Software Simulator

The MPLAB SIM Software Simulator allows code development in a PC-hosted environment by simulating the PIC MCUs and dsPIC<sup>®</sup> DSCs on an instruction level. On any given instruction, the data areas can be examined or modified and stimuli can be applied from a comprehensive stimulus controller. Registers can be logged to files for further run-time analysis. The trace buffer and logic analyzer display extend the power of the simulator to record and track program execution, actions on I/O, most peripherals and internal registers.

The MPLAB SIM Software Simulator fully supports symbolic debugging using the MPLAB C Compilers, and the MPASM and MPLAB Assemblers. The software simulator offers the flexibility to develop and debug code outside of the hardware laboratory environment, making it an excellent, economical software development tool.

#### 14.8 MPLAB REAL ICE In-Circuit Emulator System

MPLAB REAL ICE In-Circuit Emulator System is Microchip's next generation high-speed emulator for Microchip Flash DSC and MCU devices. It debugs and programs PIC<sup>®</sup> Flash MCUs and dsPIC<sup>®</sup> Flash DSCs with the easy-to-use, powerful graphical user interface of the MPLAB Integrated Development Environment (IDE), included with each kit.

The emulator is connected to the design engineer's PC using a high-speed USB 2.0 interface and is connected to the target with either a connector compatible with incircuit debugger systems (RJ11) or with the new high-speed, noise tolerant, Low-Voltage Differential Signal (LVDS) interconnection (CAT5).

The emulator is field upgradable through future firmware downloads in MPLAB IDE. In upcoming releases of MPLAB IDE, new devices will be supported, and new features will be added. MPLAB REAL ICE offers significant advantages over competitive emulators including low-cost, full-speed emulation, run-time variable watches, trace analysis, complex breakpoints, a ruggedized probe interface and long (up to three meters) interconnection cables.

#### 14.9 MPLAB ICD 3 In-Circuit Debugger System

MPLAB ICD 3 In-Circuit Debugger System is Microchip's most cost effective high-speed hardware debugger/programmer for Microchip Flash Digital Signal Controller (DSC) and microcontroller (MCU) devices. It debugs and programs PIC<sup>®</sup> Flash microcontrollers and dsPIC<sup>®</sup> DSCs with the powerful, yet easyto-use graphical user interface of MPLAB Integrated Development Environment (IDE).

The MPLAB ICD 3 In-Circuit Debugger probe is connected to the design engineer's PC using a high-speed USB 2.0 interface and is connected to the target with a connector compatible with the MPLAB ICD 2 or MPLAB REAL ICE systems (RJ-11). MPLAB ICD 3 supports all MPLAB ICD 2 headers.

## 14.10 PICkit 3 In-Circuit Debugger/ Programmer and PICkit 3 Debug Express

The MPLAB PICkit 3 allows debugging and programming of PIC<sup>®</sup> and dsPIC<sup>®</sup> Flash microcontrollers at a most affordable price point using the powerful graphical user interface of the MPLAB Integrated Development Environment (IDE). The MPLAB PICkit 3 is connected to the design engineer's PC using a full speed USB interface and can be connected to the target via an Microchip debug (RJ-11) connector (compatible with MPLAB ICD 3 and MPLAB REAL ICE). The connector uses two device I/O pins and the reset line to implement in-circuit debugging and In-Circuit Serial Programming<sup>™</sup>.

The PICkit 3 Debug Express include the PICkit 3, demo board and microcontroller, hookup cables and CDROM with user's guide, lessons, tutorial, compiler and MPLAB IDE software.

## 14.11 PICkit 2 Development Programmer/Debugger and PICkit 2 Debug Express

The PICkit<sup>™</sup> 2 Development Programmer/Debugger is a low-cost development tool with an easy to use interface for programming and debugging Microchip's Flash families of microcontrollers. The full featured Windows® programming interface supports baseline (PIC10F, PIC12F5xx, PIC16F5xx), midrange (PIC12F6xx, PIC16F), PIC18F, PIC24, dsPIC30, dsPIC33, and PIC32 families of 8-bit, 16-bit, and 32-bit microcontrollers, and many Microchip Serial EEPROM products. With Microchip's powerful MPLAB Integrated Development Environment (IDE) the PICkit<sup>™</sup> 2 enables in-circuit debugging on most PIC<sup>®</sup> microcontrollers. In-Circuit-Debugging runs, halts and single steps the program while the PIC microcontroller is embedded in the application. When halted at a breakpoint, the file registers can be examined and modified.

The PICkit 2 Debug Express include the PICkit 2, demo board and microcontroller, hookup cables and CDROM with user's guide, lessons, tutorial, compiler and MPLAB IDE software.

### 14.12 MPLAB PM3 Device Programmer

The MPLAB PM3 Device Programmer is a universal, CE compliant device programmer with programmable voltage verification at VDDMIN and VDDMAX for maximum reliability. It features a large LCD display (128 x 64) for menus and error messages and a modular, detachable socket assembly to support various package types. The ICSP™ cable assembly is included as a standard item. In Stand-Alone mode, the MPLAB PM3 Device Programmer can read, verify and program PIC devices without a PC connection. It can also set code protection in this mode. The MPLAB PM3 connects to the host PC via an RS-232 or USB cable. The MPLAB PM3 has high-speed communications and optimized algorithms for quick programming of large memory devices and incorporates an MMC card for file storage and data applications.

### 14.13 Demonstration/Development Boards, Evaluation Kits, and Starter Kits

A wide variety of demonstration, development and evaluation boards for various PIC MCUs and dsPIC DSCs allows quick application development on fully functional systems. Most boards include prototyping areas for adding custom circuitry and provide application firmware and source code for examination and modification.

The boards support a variety of features, including LEDs, temperature sensors, switches, speakers, RS-232 interfaces, LCD displays, potentiometers and additional EEPROM memory.

The demonstration and development boards can be used in teaching environments, for prototyping custom circuits and for learning about various microcontroller applications.

In addition to the PICDEM<sup>™</sup> and dsPICDEM<sup>™</sup> demonstration/development board series of circuits, Microchip has a line of evaluation kits and demonstration software for analog filter design, KEELOQ<sup>®</sup> security ICs, CAN, IrDA<sup>®</sup>, PowerSmart battery management, SEEVAL<sup>®</sup> evaluation system, Sigma-Delta ADC, flow rate sensing, plus many more.

Also available are starter kits that contain everything needed to experience the specified device. This usually includes a single application and debug capability, all on one board.

Check the Microchip web page (www.microchip.com) for the complete list of demonstration, development and evaluation kits.

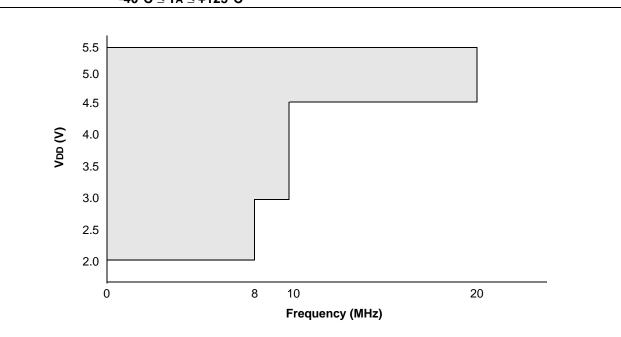
## 15.0 ELECTRICAL SPECIFICATIONS

## Absolute Maximum Ratings<sup>(†)</sup>

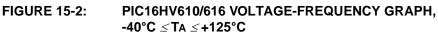
Ambient temperature under bias	40° to +125°C	
Storage temperature	5°C to +150°C	
Voltage on VDD with respect to Vss	-0.3V to +6.5V	
Voltage on MCLR with respect to Vss	).3V to +13.5V	
Voltage on all other pins with respect to Vss0.3V to	o (Vdd + 0.3V)	
Total power dissipation <sup>(1)</sup>	800 mW	
Maximum current out of Vss pin	95 mA	
Maximum current into Vod pin	95 mA	
Input clamp current, Iικ (VI < 0 or VI > VDD)	± 20 mA	
Output clamp current, loк (Vo < 0 or Vo >VDD)		
Maximum output current sunk by any I/O pin	25 mA	
Maximum output current sourced by any I/O pin	25 mA	
Maximum current sunk by PORTA and PORTC (combined)	90 mA	
Maximum current sourced PORTA and PORTC (combined)	90 mA	
<b>Note 1:</b> Power dissipation is calculated as follows: PDIS = VDD x {IDD $-\sum$ IDH} + $\sum$ {(VDD $-$ VOH) x IOL).	IOH} <b>+</b> ∑(VOI x	

† NOTICE: Stresses above those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. This is a stress rating only and functional operation of the device at those or any other conditions above those indicated in the operation listings of this specification is not implied. Exposure above maximum rating conditions for extended periods may affect device reliability.

FIGURE 15-1: PIC16F610/616 VOLTAGE-FREQUENCY GRAPH, -40°C  $\leq$  Ta  $\leq$  +125°C



**Note 1:** The shaded region indicates the permissible combinations of voltage and frequency.



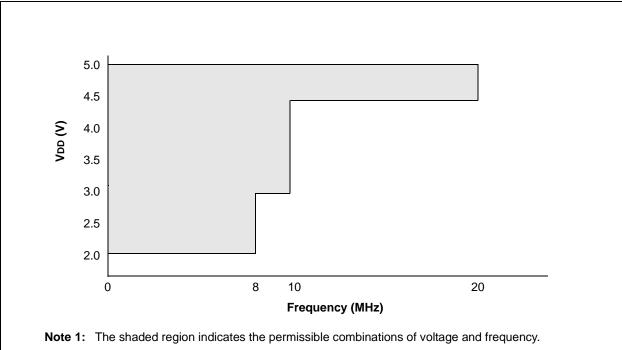


FIGURE 15-3: PIC16F610/616 FREQUENCY TOLERANCE GRAPH,

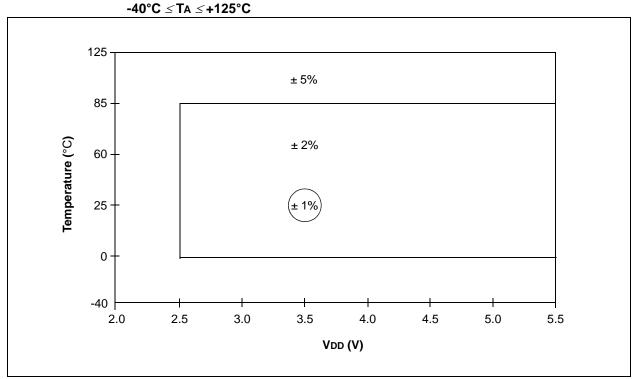
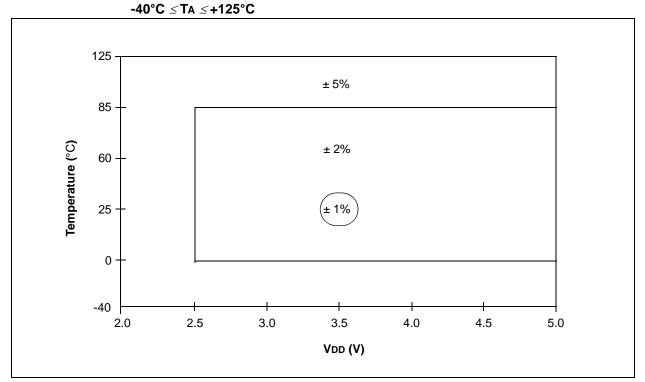


FIGURE 15-4: PIC16HV610/616 FREQUENCY TOLERANCE GRAPH,



## 15.1 DC Characteristics: PIC16F610/616/16HV610/616-I (Industrial) PIC16F610/616/16HV610/616-E (Extended)

DC CHA	DC CHARACTERISTICS			$\begin{array}{ll} \mbox{Standard Operating Conditions (unless otherwise stated)} \\ \mbox{Operating temperature} & -40^{\circ}\mbox{C} \leq T\mbox{A} \leq +85^{\circ}\mbox{C for industrial} \\ & -40^{\circ}\mbox{C} \leq T\mbox{A} \leq +125^{\circ}\mbox{C for extended} \end{array}$						
Param No.	Sym	Characteristic	Min	Тур†	Max	Units	Conditions			
	Vdd	Supply Voltage								
D001		PIC16F610/616	2.0	—	5.5	V	Fosc < = 4 MHz			
D001		PIC16HV610/616	2.0	—	(2)	V	Fosc < = 4 MHz			
D001B		PIC16F610/616	2.0	—	5.5	V	Fosc < = 8 MHz			
D001B		PIC16HV610/616	2.0	—	(2)	V	Fosc < = 8 MHz			
D001C		PIC16F610/616	3.0	—	5.5	V	Fosc < = 10 MHz			
D001C		PIC16HV610/616	3.0	—	(2)	V	Fosc < = 10 MHz			
D001D		PIC16F610/616	4.5	—	5.5	V	Fosc < = 20 MHz			
D001D		PIC16HV610/616	4.5	—	(2)	V	Fosc < = 20 MHz			
D002*	Vdr	RAM Data Retention Voltage <sup>(1)</sup>	1.5		—	V	Device in Sleep mode			
D003	VPOR	VDD Start Voltage to ensure internal Power-on Reset signal	—	Vss	—	V	See Section 12.3.1 "Power-on Reset (POR)" for details.			
D004*	Svdd	VDD Rise Rate to ensure internal Power-on Reset signal	0.05			V/ms	See Section 12.3.1 "Power-on Reset (POR)" for details.			

\* These parameters are characterized but not tested.

† Data in "Typ" column is at 5.0V, 25°C unless otherwise stated. These parameters are for design guidance only and are not tested.

**Note 1:** This is the limit to which VDD can be lowered in Sleep mode without losing RAM data.

2: User defined. Voltage across the shunt regulator should not exceed 5V.

## 15.2 DC Characteristics: PIC16F610/616-I (Industrial) PIC16F610/616-E (Extended)

DC CH	ARACTERISTICS		ard Oper ing temp		-40°C ⊴	s otherwise stated) 35°C for industrial 125°C for extended	
Param	Device Characteristics	Min	Тур†	Max	Units		Conditions
No.						VDD	Note
D010	Supply Current (IDD) <sup>(1, 2)</sup>	—	13	25	μA	2.0	Fosc = 32 kHz
	PIC16F610/616	_	19	29	μA	3.0	LP Oscillator mode
		—	32	51	μΑ	5.0	
D011*		—	135	225	μA	2.0	Fosc = 1 MHz
		_	185	285	μA	3.0	XT Oscillator mode
		_	300	405	μA	5.0	
D012		—	240	360	μA	2.0	Fosc = 4 MHz
		—	360	505	μA	3.0	XT Oscillator mode
		—	0.66	1.0	mA	5.0	
D013*		—	75	110	μΑ	2.0	Fosc = 1 MHz
		_	155	255	μA	3.0	EC Oscillator mode
		_	345	530	μA	5.0	
D014		_	185	255	μA	2.0	Fosc = 4 MHz
		—	325	475	μA	3.0	EC Oscillator mode
		_	0.665	1.0	mA	5.0	
D016*		_	245	340	μΑ	2.0	Fosc = 4 MHz
			360	485	μA	3.0	INTOSC mode
		—	0.620	0.845	mA	5.0	
D017			395	550	μΑ	2.0	Fosc = 8 MHz
			0.620	0.850	mA	3.0	INTOSC mode
		—	1.2	1.6	mA	5.0	
D018			175	235	μA	2.0	Fosc = 4 MHz
			285	390	μΑ	3.0	EXTRC mode <sup>(3)</sup>
		—	530	750	μΑ	5.0	
D019			2.2	3.1	mA	4.5	Fosc = 20 MHz
		—	2.8	3.35	mA	5.0	HS Oscillator mode

These parameters are characterized but not tested.

† Data in "Typ" column is at 5.0V, 25°C unless otherwise stated. These parameters are for design guidance only and are not tested.

**Note 1:** The test conditions for all IDD measurements in active operation mode are: OSC1 = external square wave, from rail-to-rail; all I/O pins tri-stated, pulled to VDD; MCLR = VDD; WDT disabled.

2: The supply current is mainly a function of the operating voltage and frequency. Other factors, such as I/O pin loading and switching rate, oscillator type, internal code execution pattern and temperature, also have an impact on the current consumption.

**3:** For RC oscillator configurations, current through REXT is not included. The current through the resistor can be extended by the formula IR = VDD/2REXT (mA) with REXT in k $\Omega$ .

## 15.3 DC Characteristics: PIC16HV610/616-I (Industrial) PIC16HV610/616-E (Extended)

DC CH	ARACTERISTICS		ard Operating temp		-40°C ⊴	≤ TA ≤ +8	<b>s otherwise stated)</b> 35°C for industrial 125°C for extended
Param	Device Characteristics	Min	Тур†	Max	Units		Conditions
No.	Derrice entanderricher		.,,,,,	max	onno	Vdd	Note
D010	Supply Current (IDD) <sup>(1, 2)</sup>		160	230	μΑ	2.0	Fosc = 32 kHz
	PIC16HV610/616	_	240	310	μΑ	3.0	LP Oscillator mode
		_	280	400	μΑ	4.5	
D011*		—	270	380	μΑ	2.0	Fosc = 1 MHz
		_	400	560	μΑ	3.0	XT Oscillator mode
		—	520	780	μΑ	4.5	1
D012		_	380	540	μΑ	2.0	Fosc = 4 MHz
		_	575	810	μΑ	3.0	XT Oscillator mode
		_	0.875	1.3	mA	4.5	
D013*		—	215	310	μΑ	2.0	Fosc = 1 MHz
		_	375	565	μΑ	3.0	EC Oscillator mode
		—	570	870	μΑ	4.5	
D014		-	330	475	μΑ	2.0	Fosc = 4 MHz
		_	550	800	μΑ	3.0	EC Oscillator mode
		—	0.85	1.2	mA	4.5	
D016*			310	435	μΑ	2.0	Fosc = 4 MHz
		_	500	700	μΑ	3.0	INTOSC mode
		—	0.74	1.1	mA	4.5	
D017		_	460	650	μΑ	2.0	Fosc = 8 MHz
		_	0.75	1.1	mA	3.0	INTOSC mode
		—	1.2	1.6	mA	4.5	
D018		_	320	465	μA	2.0	Fosc = 4 MHz
		_	510	750	μA	3.0	EXTRC mode <sup>(3)</sup>
		—	0.770	1.0	mA	4.5	
D019		—	2.5	3.4	mA	4.5	Fosc = 20 MHz HS Oscillator mode

\* These parameters are characterized but not tested.

† Data in "Typ" column is at 4.5V, 25°C unless otherwise stated. These parameters are for design guidance only and are not tested.

**Note 1:** The test conditions for all IDD measurements in active operation mode are: OSC1 = external square wave, from rail-to-rail; all I/O pins tri-stated, pulled to VDD; MCLR = VDD; WDT disabled.

2: The supply current is mainly a function of the operating voltage and frequency. Other factors, such as I/O pin loading and switching rate, oscillator type, internal code execution pattern and temperature, also have an impact on the current consumption.

**3:** For RC oscillator configurations, current through REXT is not included. The current through the resistor can be extended by the formula IR = VDD/2REXT (mA) with REXT in k $\Omega$ .

## 15.4 DC Characteristics: PIC16F610/616- I (Industrial)

	ARACTERISTICS		ard Oper ing temp				<b>s otherwise stated)</b> 35°C for industrial
Param	Device Characteristics	Min	Typ†	Мах	Units		Conditions
No.	Device Characteristics		וקעי	Max	Unito	VDD	Note
D020	Power-down Base Current(IPD) <sup>(2)</sup>	_	0.05	0.9	μA	2.0	WDT, BOR, Comparators, VREF and T1OSC disabled
		_	0.15	1.2	μA	3.0	
	PIC16F610/616	_	0.35	1.5	μA	5.0	
		—	150	500	nA	3.0	-40°C $\leq$ TA $\leq$ +25°C for industrial
D021		—	0.5	1.5	μA	2.0	WDT Current <sup>(1)</sup>
		_	2.5	4.0	μA	3.0	
		_	9.5	17	μA	5.0	-
D022		—	5.0	9	μA	3.0	BOR Current <sup>(1)</sup>
		_	6.0	12	μA	5.0	
D023		_	105	115	μA	2.0	Comparator Current <sup>(1)</sup> , both
		_	110	125	μA	3.0	comparators enabled
		_	116	140	μA	5.0	
D024		_	50	60	μA	2.0	Comparator Current <sup>(1)</sup> , single
		_	55	65	μΑ	3.0	comparator enabled
			60	75	μA	5.0	
D025			30	40	μΑ	2.0	CVREF Current <sup>(1)</sup> (high range)
		_	45	60	μA	3.0	
		—	75	105	μA	5.0	
D026*			39	50	μA	2.0	CVREF Current <sup>(1)</sup> (low range)
			59	80	μΑ	3.0	4
			98	130	μA	5.0	
D027			5.5	10	μA	2.0	T1OSC Current <sup>(1)</sup> , 32.768 kHz
			7.0	12	μA	3.0	4
			8.5	14	μA	5.0	
D028			0.2	1.6	μA	3.0	A/D Current <sup>(1)</sup> , no conversion in
		—	0.36	1.9	μΑ	5.0	progress.

\* These parameters are characterized but not tested.

† Data in "Typ" column is at 5.0V, 25°C unless otherwise stated. These parameters are for design guidance only and are not tested.

Note 1: The peripheral current is the sum of the base IDD or IPD and the additional current consumed when this peripheral is enabled. The peripheral ∆ current can be determined by subtracting the base IDD or IPD current from this limit. Max values should be used when calculating total current consumption.

2: The power-down current in Sleep mode does not depend on the oscillator type. Power-down current is measured with the part in Sleep mode, with all I/O pins in high-impedance state and tied to VDD.

### 15.5 DC Characteristics: PIC16F610/616-E (Extended)

DC CHA	RACTERISTICS		ard Operating temp				s otherwise stated) 125°C for extended	
Param	Device Characteristics	Min	Trunt	Max	Units		Conditions	
No.	Device Characteristics		Тур†	Max	Units	Vdd	Note	
D020E	Power-down Base		0.05	4.0	μA	2.0	WDT, BOR, Comparators, VREF and	
	Current (IPD) <sup>(2)</sup> PIC16F610/616		0.15	5.0	μΑ	3.0	T1OSC disabled	
	PIC10F010/010		0.35	8.5	μΑ	5.0		
D021E			0.5	5.0	μΑ	2.0	WDT Current <sup>(1)</sup>	
			2.5	8.0	μΑ	3.0		
			9.5	19	μΑ	5.0		
D022E			5.0	15	μΑ	3.0	BOR Current <sup>(1)</sup>	
			6.0	19	μΑ	5.0		
D023E			105	130	μΑ	2.0	Comparator Current <sup>(1)</sup> , both	
			110	140	μΑ	3.0	comparators enabled	
			116	150	μΑ	5.0		
D024E			50	70	μΑ	2.0	Comparator Current <sup>(1)</sup> , single	
			55	75	μΑ	3.0	comparator enabled	
			60	80	μΑ	5.0		
D025E			30	40	μΑ	2.0	CVREF Current <sup>(1)</sup> (high range)	
			45	60	μΑ	3.0		
			75	105	μΑ	5.0		
D026E*		_	39	50	μΑ	2.0	CVREF Current <sup>(1)</sup> (low range)	
			59	80	μΑ	3.0		
			98	130	μΑ	5.0		
D027E		—	5.5	16	μA	2.0	T1OSC Current <sup>(1)</sup> , 32.768 kHz	
			7.0	18	μΑ	3.0		
			8.5	22	μΑ	5.0		
D028E			0.2	6.5	μA	3.0	A/D Current <sup>(1)</sup> , no conversion in	
		_	0.36	10	μA	5.0	progress	

\* These parameters are characterized but not tested.

† Data in "Typ" column is at 5.0V, 25°C unless otherwise stated. These parameters are for design guidance only and are not tested.

Note 1: The peripheral current is the sum of the base IDD or IPD and the additional current consumed when this peripheral is enabled. The peripheral ∆ current can be determined by subtracting the base IDD or IPD current from this limit. Max values should be used when calculating total current consumption.

2: The power-down current in Sleep mode does not depend on the oscillator type. Power-down current is measured with the part in Sleep mode, with all I/O pins in high-impedance state and tied to VDD.

#### 15.6 DC Characteristics: PIC16HV610/616- I (Industrial)

DC CHA	RACTERISTICS		ard Operating temp				s otherwise stated) 35°C for industrial	
Param						Conditions		
No.	Device Characteristics	Min	Тур†	Max	Units	VDD	Note	
D020	Power-down Base Current(IPD) <sup>(2,3)</sup>	-	135	200	μΑ	2.0	WDT, BOR, Comparators, VREF and T1OSC disabled	
			210	280	μΑ	3.0		
	PIC16HV610/616		260	350	μΑ	4.5		
D021			135	200	μΑ	2.0	WDT Current <sup>(1)</sup>	
			210	285	μΑ	3.0		
			265	360	μΑ	4.5		
D022			215	285	μA	3.0	BOR Current <sup>(1)</sup>	
			265	360	μΑ	4.5		
D023			240	340	μA	2.0	Comparator Current <sup>(1)</sup> , both	
			320	420	μA	3.0	comparators enabled	
			370	500	μΑ	4.5		
D024			185	270	μA	2.0	Comparator Current <sup>(1)</sup> , single	
			265	350	μA	3.0	comparator enabled	
			320	430	μA	4.5		
D025			165	235	μA	2.0	CVREF Current <sup>(1)</sup> (high range)	
			255	330	μA	3.0		
			330	430	μA	4.5		
D026*		_	175	245	μA	2.0	CVREF Current <sup>(1)</sup> (low range)	
			275	350	μΑ	3.0		
		_	355	450	μΑ	4.5		
D027			140	205	μΑ	2.0	T1OSC Current <sup>(1)</sup> , 32.768 kHz	
			220	290	μΑ	3.0		
		—	270	360	μA	4.5		
D028			210	280	μΑ	3.0	A/D Current <sup>(1)</sup> , no conversion in	
		—	260	350	μA	4.5	progress	

\* These parameters are characterized but not tested.

† Data in "Typ" column is at 4.5V, 25°C unless otherwise stated. These parameters are for design guidance only and are not tested.

**Note 1:** The peripheral current is the sum of the base IDD or IPD and the additional current consumed when this peripheral is enabled. The peripheral ∆ current can be determined by subtracting the base IDD or IPD current from this limit. Max values should be used when calculating total current consumption.

2: The power-down current in Sleep mode does not depend on the oscillator type. Power-down current is measured with the part in Sleep mode, with all I/O pins in high-impedance state and tied to VDD.

3: Shunt regulator is always enabled and always draws operating current.

## 15.7 DC Characteristics: PIC16HV610/616-E (Extended)

DC CHA	RACTERISTICS		ard Oper ting temp				s otherwise stated) 125°C for extended	
Param	Device Characteristics	Min	Typ†	Max	Units	Conditions		
No.	Device Gilaracteristics	WIIN	וקעי	IVIAX		VDD	Note	
D020E	Power-down Base		135	200	μA	2.0	WDT, BOR, Comparators, VREF and	
		_	210	280	μΑ	3.0	T1OSC disabled	
	PIC16HV610/616		260	350	μΑ	4.5		
D021E			135	200	μΑ	2.0	WDT Current <sup>(1)</sup>	
			210	285	μΑ	3.0		
			265	360	μΑ	4.5		
D022E			215	285	μΑ	3.0	BOR Current <sup>(1)</sup>	
			265	360	μA	4.5		
D023E			240	360	μA	2.0	Comparator Current <sup>(1)</sup> , both	
			320	440	μA	3.0	comparators enabled	
			370	500	μA	4.5		
D024E			185	280	μA	2.0	Comparator Current <sup>(1)</sup> , single	
			265	360	μA	3.0	comparator enabled	
			320	430	μA	4.5		
D025E		—	165	235	μA	2.0	CVREF Current <sup>(1)</sup> (high range)	
		_	255	330	μA	3.0		
		_	330	430	μA	4.5		
D026E*		_	175	245	μA	2.0	CVREF Current <sup>(1)</sup> (low range)	
			275	350	μA	3.0	1	
			355	450	μA	4.5	1	
D027E		_	140	205	μA	2.0	T1OSC Current <sup>(1)</sup> , 32.768 kHz	
			220	290	μA	3.0	7	
			270	360	μΑ	4.5	1	
D028E		—	210	280	μA	3.0	A/D Current <sup>(1)</sup> , no conversion in	
		_	260	350	μA	4.5	progress	

These parameters are characterized but not tested.

† Data in "Typ" column is at 4.5V, 25°C unless otherwise stated. These parameters are for design guidance only and are not tested.

Note 1: The peripheral current is the sum of the base IDD or IPD and the additional current consumed when this peripheral is enabled. The peripheral ∆ current can be determined by subtracting the base IDD or IPD current from this limit. Max values should be used when calculating total current consumption.

2: The power-down current in Sleep mode does not depend on the oscillator type. Power-down current is measured with the part in Sleep mode, with all I/O pins in high-impedance state and tied to VDD.

3: Shunt regulator is always enabled and always draws operating current.

### 15.8 DC Characteristics: PIC16F610/616/16HV610/616- I (Industrial) PIC16F610/616/16HV610/616 - E (Extended)

DC CHA	RACTER	ISTICS	Standard Operat Operating temper		-40°Ċ ≤	TA ≤ +8	<b>herwise stated)</b> 5°C for industrial 25°C for extended
Param No.	Sym	Characteristic	Min	Тур†	Мах	Units	Conditions
	VIL	Input Low Voltage					
		I/O port:					
D030		with TTL buffer	Vss	—	0.8	V	$4.5V \leq V\text{DD} \leq 5.5V$
D030A			Vss	—	0.15 VDD	V	$2.0V \leq V\text{DD} \leq 4.5V$
D031		with Schmitt Trigger buffer	Vss	—	0.2 Vdd	V	$2.0V \leq V\text{DD} \leq 5.5V$
D032		MCLR, OSC1 (RC mode)	Vss	—	0.2 Vdd	V	
D033		OSC1 (XT and LP modes)	Vss	_	0.3	V	
D033A		OSC1 (HS mode)	Vss	—	0.3 Vdd	V	
	Viн	Input High Voltage					
D040		with TTL buffer	2.0		VDD	v	$4.5V \le VDD \le 5.5V$
D040A			0.25 Vpd + 0.8		VDD	v	$4.5V \le VDD \le 3.5V$ $2.0V \le VDD \le 4.5V$
D040/		with Schmitt Trigger buffer	0.8 Vpp	_	VDD	v	2.0V < VDD < 5.5V
D042		MCLR	0.8 VDD	_	VDD	v	2.00 2 000 2 0.00
D043		OSC1 (XT and LP modes)	1.6	_	VDD	v	
D043A		OSC1 (HS mode)	0.7 VDD	_	VDD	v	
D043B		OSC1 (RC mode)	0.9 VDD	_	VDD	v	(Note 1)
	lı∟	Input Leakage Current <sup>(2,3)</sup>				-	
D060		I/O ports	_	± 0.1	± 1	μA	$\label{eq:VSS} \begin{split} &V{\sf SS} \leq V{\sf PIN} \leq V{\sf DD}, \\ &P{\sf in} \mbox{ at high-impedance} \end{split}$
D061		RA3/MCLR <sup>(3,4)</sup>	_	± 0.7	± 5	μA	$VSS \leq VPIN \leq VDD$
D063		OSC1	_	± 0.1	± 5	μA	Vss $\leq$ VPIN $\leq$ VDD, XT, HS and LP oscillator configuration
D070*	IPUR	PORTA Weak Pull-up Current <sup>(5)</sup>	50	250	400	μA	VDD = 5.0V, VPIN = VSS
	Vol	Output Low Voltage	—	—	0.6	V	IOL = 7.0 mA, VDD = 4.5V, -40°C to +125°C
D080		I/O ports	_	_	0.6	V	IOL = 8.5 mA, VDD = 4.5V, -40°C to +85°C
	Vон	Output High Voltage	Vdd - 0.7	-	—	V	IOH = -2.5 mA, VDD = 4.5V, -40°C to +125°C
D090		I/O ports <sup>(2)</sup>	Vdd - 0.7	-	—	V	IOH = -3.0 mA, VDD = 4.5V, -40°C to +85°C

\* These parameters are characterized but not tested.

† Data in "Typ" column is at 5.0V, 25°C unless otherwise stated. These parameters are for design guidance only and are not tested.

Note 1: In RC oscillator configuration, the OSC1/CLKIN pin is a Schmitt Trigger input. It is not recommended to use an external clock in RC mode.

2: Negative current is defined as current sourced by the pin.

3: The leakage current on the MCLR pin is strongly dependent on the applied voltage level. The specified levels represent normal operating conditions. Higher leakage current may be measured at different input voltages.

4: This specification applies to RA3/MCLR configured as RA3 input with internal pull-up disabled.

5: This specification applies to all weak pull-up pins, including the weak pull-up on RA3/MCLR. When RA3/MCLR is configured as MCLR reset pin, the weak pull-up is always enabled.

### 15.9 DC Characteristics: PIC16F610/616/16HV610/616- I (Industrial) PIC16F610/616/16HV610/616 - E (Extended)

			Standard Operat Operating temper	•	-40°Ċ ≤	TA ≤ +8	herwise stated) 5°C for industrial 25°C for extended
Param No.	Sym	Characteristic	Min	Тур†	Max	Units	Conditions
D101*	COSC2	Capacitive Loading Specs on Output Pins OSC2 pin	_	_	15	pF	In XT, HS and LP modes when external clock is used to drive OSC1
D101A*	Сю	All I/O pins	_	_	50	pF	
		Program Flash Memory					
D130	Eр	Cell Endurance	10K	100K	_	E/W	$-40^{\circ}C \leq TA \leq +85^{\circ}C$
D130A	ED	Cell Endurance	1K	10K	_	E/W	$+85^{\circ}C \le TA \le +125^{\circ}C$
D131	Vpr	VDD for Read	Vmin	—	5.5	V	VMIN = Minimum operating voltage
D132	VPEW	VDD for Erase/Write	4.5	_	5.5	V	
D133	TPEW	Erase/Write cycle time	_	2	2.5	ms	
D134	Tretd	Characteristic Retention	40	-	—	Year	Provided no other specifications are violated

\* These parameters are characterized but not tested.

† Data in "Typ" column is at 5.0V, 25°C unless otherwise stated. These parameters are for design guidance only and are not tested.

Note 1: In RC oscillator configuration, the OSC1/CLKIN pin is a Schmitt Trigger input. It is not recommended to use an external clock in RC mode.

## 15.10 Thermal Considerations

Param No.	Sym	Characteristic	Тур	Units	Conditions
TH01	θJA	Thermal Resistance	70*	C/W	14-pin PDIP package
		Junction to Ambient	85.0*	C/W	14-pin SOIC package
			100*	C/W	14-pin TSSOP package
			37*	C/W	16-pin QFN 4x4mm package
TH02	θJC	Thermal Resistance	32.5*	C/W	14-pin PDIP package
		Junction to Case	31.0*	C/W	14-pin SOIC package
			31.7*	C/W	14-pin TSSOP package
			2.6*	C/W	16-pin QFN 4x4mm package
TH03	TDIE	Die Temperature	150*	С	
TH04	PD	Power Dissipation	—	W	PD = PINTERNAL + PI/O
TH05	PINTERNAL	Internal Power Dissipation	—	W	PINTERNAL = IDD x VDD (NOTE 1)
TH06	Pi/o	I/O Power Dissipation	—	W	$PI/O = \Sigma (IOL * VOL) + \Sigma (IOH * (VDD - VOH))$
TH07	Pder	Derated Power	-	W	Pder = PDmax (Tdie - Ta)/θja (NOTE 2)

These parameters are characterized but not tested.

Note 1: IDD is current to run the chip alone without driving any load on the output pins.

**2:** TA = Ambient Temperature.

## 15.11 Timing Parameter Symbology

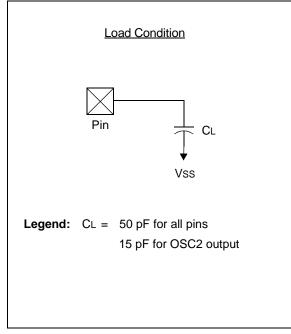
The timing parameter symbols have been created with one of the following formats:

#### 1. TppS2ppS

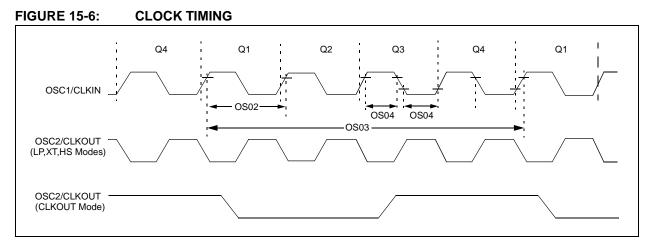
2. TppS

<u>z. rpp3</u>			
т			
F	Frequency	Т	Time
Lowerc	ase letters (pp) and their meanings:		
рр			
сс	CCP1	osc	OSC1
ck	CLKOUT	rd	RD
CS	CS	rw	RD or WR
di	SDI	sc	SCK
do	SDO	SS	SS
dt	Data in	tO	TOCKI
io	I/O Port	t1	T1CKI
mc	MCLR	wr	WR
Upperc	ase letters and their meanings:		
S			
F	Fall	Р	Period
н	High	R	Rise
I	Invalid (High-impedance)	V	Valid
L	Low	Z	High-impedance

#### FIGURE 15-5: LOAD CONDITIONS



## 15.12 AC Characteristics: PIC16F610/616/16HV610/616 (Industrial, Extended)



## TABLE 15-1: CLOCK OSCILLATOR TIMING REQUIREMENTS

	Standard Operating Conditions (unless otherwise stated)         Operating temperature $-40^{\circ}C \le TA \le +125^{\circ}C$									
Param No.	Sym	Characteristic	Min	Тур†	Max	Units	Conditions			
OS01	Fosc	External CLKIN Frequency <sup>(1)</sup>	DC	_	37	kHz	LP Oscillator mode			
			DC	_	4	MHz	XT Oscillator mode			
			DC	_	20	MHz	HS Oscillator mode			
			DC	_	20	MHz	EC Oscillator mode			
		Oscillator Frequency <sup>(1)</sup>	_	32.768	_	kHz	LP Oscillator mode			
			0.1	_	4	MHz	XT Oscillator mode			
			1		20	MHz	HS Oscillator mode			
			DC	_	4	MHz	RC Oscillator mode			
OS02	Tosc	External CLKIN Period <sup>(1)</sup>	27		×	μS	LP Oscillator mode			
			250	_	×	ns	XT Oscillator mode			
			50	_	×	ns	HS Oscillator mode			
			50	_	$\infty$	ns	EC Oscillator mode			
		Oscillator Period <sup>(1)</sup>	_	30.5	_	μS	LP Oscillator mode			
			250	_	10,000	ns	XT Oscillator mode			
			50	_	1,000	ns	HS Oscillator mode			
			250	_	—	ns	RC Oscillator mode			
OS03	TCY	Instruction Cycle Time <sup>(1)</sup>	200	Тсү	DC	ns	Tcy = 4/Fosc			
OS04*	TosH,	External CLKIN High,	2	_	_	μS	LP oscillator			
	TosL	External CLKIN Low	100	—	—	ns	XT oscillator			
			20	—	_	ns	HS oscillator			
OS05*	TosR,	External CLKIN Rise,	0	_	~	ns	LP oscillator			
	TosF	External CLKIN Fall	0	—	$\infty$	ns	XT oscillator			
			0	—	$\infty$	ns	HS oscillator			

\* These parameters are characterized but not tested.

† Data in "Typ" column is at 5V, 25°C unless otherwise stated. These parameters are for design guidance only and are not tested.

**Note 1:** Instruction cycle period (TCY) equals four times the input oscillator time base period. All specified values are based on characterization data for that particular oscillator type under standard operating conditions with the device executing code. Exceeding these specified limits may result in an unstable oscillator operation and/or higher than expected current consumption. All devices are tested to operate at "min" values with an external clock applied to OSC1 pin. When an external clock input is used, the "max" cycle time limit is "DC" (no clock) for all devices.

## TABLE 15-2: OSCILLATOR PARAMETERS

	<b>Operating</b> Temperatu	$\begin{array}{ll} \mbox{ Conditions (unless otherw)} & \mbox{ J Conditions (unless otherw)} &  J Con$	,					
Param No.	Sym	Characteristic	Freq. Tolerance	Min	Тур†	Max	Units	Conditions
OS06	TWARM	Internal Oscillator Switch when running <sup>(3)</sup>	—		_	2	Tosc	Slowest clock
OS07	INTosc	Internal Calibrated	±1%	3.96	4.0	4.04	MHz	$VDD = 3.5V, T_A = 25^{\circ}C$
		INTOSC Frequency <sup>(2)</sup> (4MHz)	±2%	3.92	4.0	4.08	MHz	$2.5V \le VDD \le 5.5V$ , $0^{\circ}C \le TA \le +85^{\circ}C$
			±5%	3.80	4.0	4.2	MHz	$2.0V \le VDD \le 5.5V$ , -40°C $\le$ TA $\le$ +85°C (Ind.), -40°C $\le$ TA $\le$ +125°C (Ext.)
OS08	INTosc	Internal Calibrated	±1%	7.92	8.0	8.08	MHz	VDD = 3.5V, T <sub>A</sub> = 25°C
		INTOSC Frequency <sup>(2)</sup> (8MHz)	±2%	7.84	8.0	8.16	MHz	$2.5V \le VDD \le 5.5V$ , $0^{\circ}C \le TA \le +85^{\circ}C$
			±5%	7.60	8.0	8.40	MHz	$2.0V \le VDD \le 5.5V$ , -40°C $\le$ TA $\le$ +85°C (Ind.), -40°C $\le$ TA $\le$ +125°C (Ext.)
OS10*	TIOSC ST	INTOSC Oscillator Wake-	—	5.5	12	24	μs	VDD = 2.0V, -40°C to +85°C
		up from Sleep	—	3.5	7	14	μs	VDD = 3.0V, -40°C to +85°C
		Start-up Time	—	3	6	11	μS	VDD = 5.0V, -40°C to +85°C

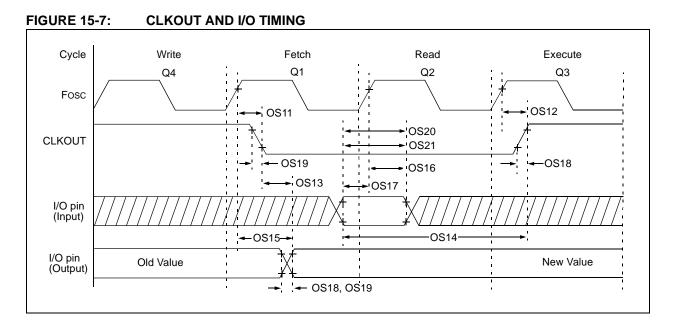
\* These parameters are characterized but not tested.

† Data in "Typ" column is at 5.0V, 25°C unless otherwise stated. These parameters are for design guidance only and are not tested.

**Note 1:** Instruction cycle period (TcY) equals four times the input oscillator time base period. All specified values are based on characterization data for that particular oscillator type under standard operating conditions with the device executing code. Exceeding these specified limits may result in an unstable oscillator operation and/or higher than expected current consumption. All devices are tested to operate at "min" values with an external clock applied to the OSC1 pin. When an external clock input is used, the "max" cycle time limit is "DC" (no clock) for all devices.

2: To ensure these oscillator frequency tolerances, VDD and VSS must be capacitively decoupled as close to the device as possible. 0.1 µF and 0.01 µF values in parallel are recommended.

3: By design.



### TABLE 15-3: CLKOUT AND I/O TIMING PARAMETERS

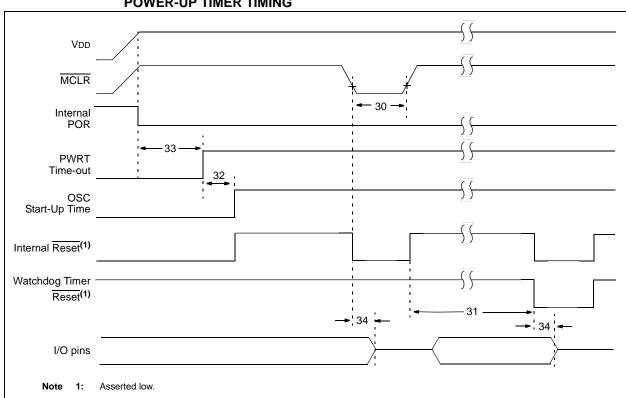
		Conditions (unless otherwise stated) re -40°C $\leq$ TA $\leq$ +125°C					
Param No.	Sym	Characteristic	Min	Тур†	Max	Units	Conditions
OS11	TosH2ckL	Fosc↑ to CLKOUT↓ <sup>(1)</sup>	—	—	70	ns	Vdd = 5.0V
OS12	TosH2cкH	Fosc↑ to CLKOUT↑ <sup>(1)</sup>	_	—	72	ns	VDD = 5.0V
OS13	TcĸL2ıoV	CLKOUT↓ to Port out valid <sup>(1)</sup>	—	_	20	ns	
OS14	ТюV2скН	Port input valid before CLKOUT <sup>(1)</sup>	Tosc + 200 ns	_	_	ns	
OS15	TosH2ıoV	Fosc↑ (Q1 cycle) to Port out valid	—	50	70*	ns	VDD = 5.0V
OS16	TosH2iol	Fosc↑ (Q2 cycle) to Port input invalid (I/O in hold time)	50	—	_	ns	VDD = 5.0V
OS17	TioV2osH	Port input valid to Fosc↑ (Q2 cycle) (I/O in setup time)	20	—	_	ns	
OS18	TIOR	Port output rise time <sup>(2)</sup>		15 40	72 32	ns	VDD = 2.0V VDD = 5.0V
OS19	TIOF	Port output fall time <sup>(2)</sup>		28 15	55 30	ns	VDD = 2.0V VDD = 5.0V
OS20*	Tinp	INT pin input high or low time	25	—	_	ns	
OS21*	Trap	PORTA interrupt-on-change new input level time	Тсү	—		ns	

\* These parameters are characterized but not tested.

† Data in "Typ" column is at 5.0V, 25°C unless otherwise stated.

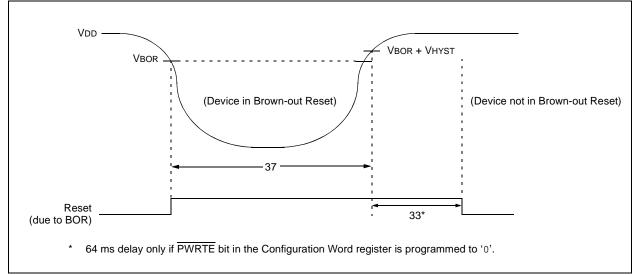
Note 1: Measurements are taken in RC mode where CLKOUT output is 4 x Tosc.

2: Includes OSC2 in CLKOUT mode.



## FIGURE 15-8: RESET, WATCHDOG TIMER, OSCILLATOR START-UP TIMER AND POWER-UP TIMER TIMING





## TABLE 15-4:RESET, WATCHDOG TIMER, OSCILLATOR START-UP TIMER, POWER-UP TIMER<br/>AND BROWN-OUT RESET PARAMETERS

	<b>d Opera</b> ng Tempe	ting Conditions (unless otherwise rature $-40^{\circ}C \le TA \le +125^{\circ}C$	se state	ed)			
Param No.	Sym	Characteristic	Min	Тур†	Max	Units	Conditions
30	ТмсL	MCLR Pulse Width (low)	2 5		—	μS μS	VDD = 5V, -40°C to +85°C VDD = 5V, -40°C to +125°C
31*	Twdt	Watchdog Timer Time-out Period (No Prescaler)	10 10	20 20	30 35	ms ms	VDD = 5V, -40°C to +85°C VDD = 5V, -40°C to +125°C
32	Tost	Oscillation Start-up Timer Period <sup>(1, 2)</sup>	_	1024	—	Tosc	(NOTE 3)
33*	TPWRT	Power-up Timer Period	40	65	140	ms	
34*	Tioz	I/O High-impedance from MCLR Low or Watchdog Timer Reset	—	_	2.0	μS	
35*	VBOR	Brown-out Reset Voltage	2.0	2.15	2.3	V	(NOTE 4)
36*	VHYST	Brown-out Reset Hysteresis	_	100	—	mV	
37*	TBOR	Brown-out Reset Minimum Detection Period	100	—	—	μS	$VDD \leq VBOR$

Legend: TBD = To Be Determined

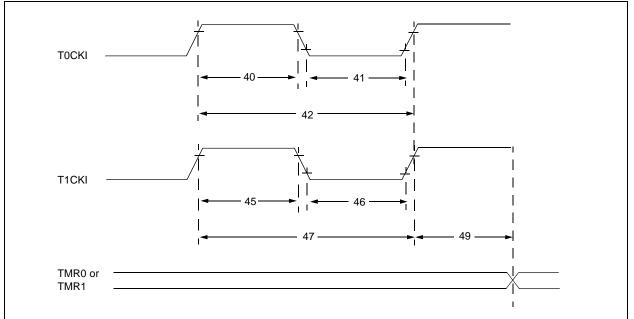
\* These parameters are characterized but not tested.

† Data in "Typ" column is at 5V, 25°C unless otherwise stated. These parameters are for design guidance only and are not tested.

- 2: By design.
- 3: Period of the slower clock.
- **4:** To ensure these voltage tolerances, VDD and VSS must be capacitivey decoupled as close to the device as possible. 0.1 μF and 0.01 μF values in parallel are recommended.

**Note 1:** Instruction cycle period (TCY) equals four times the input oscillator time base period. All specified values are based on characterization data for that particular oscillator type under standard operating conditions with the device executing code. Exceeding these specified limits may result in an unstable oscillator operation and/or higher than expected current consumption. All devices are tested to operate at "min" values with an external clock applied to the OSC1 pin. When an external clock input is used, the "max" cycle time limit is "DC" (no clock) for all devices.

### FIGURE 15-10: TIMER0 AND TIMER1 EXTERNAL CLOCK TIMINGS

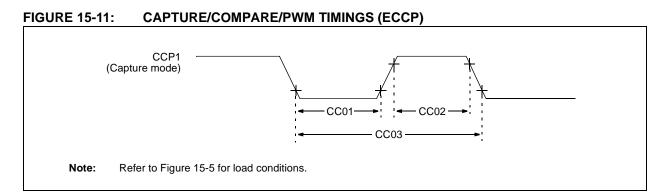


#### TABLE 15-5: TIMER0 AND TIMER1 EXTERNAL CLOCK REQUIREMENTS

Param No.	Sym		Characteristic		Min	Тур†	Max	Units	Conditions
40*	T⊤0H	T0CKI High F	Pulse Width	No Prescaler	0.5 Tcy + 20	—	_	ns	
			With Prescaler		10	—	_	ns	
41*	TT0L	T0CKI Low P	ulse Width	No Prescaler	0.5 TCY + 20	—	_	ns	
			With Prescaler		10	_	_	ns	
42*	Тт0Р	T0CKI Period	:		Greater of: 20 or <u>Tcy + 40</u> N	_	_	ns	N = prescale value (2, 4,, 256)
45*	TT1H	T1CKI High	Synchronous, I	No Prescaler	0.5 TCY + 20	—	—	ns	
		Time	Synchronous, with Prescaler		15	—	_	ns	
			Asynchronous		30	—	_	ns	
46*	TT1L	T1CKI Low	Synchronous, I	No Prescaler	0.5 TCY + 20	—	_	ns	
		Time	Synchronous, with Prescaler		15	_	_	ns	
			Asynchronous		30	—	_	ns	
47*	TT1P	T1CKI Input Period	Synchronous		Greater of: 30 or <u>Tcy + 40</u> N	—	—	ns	N = prescale value (1, 2, 4, 8)
			Asynchronous		60	—	—	ns	
48	F⊤1		ator Input Frequabled by setting		-	32.768	—	kHz	
49*	TCKEZTMR1	Delay from E Increment	xternal Clock E	dge to Timer	2 Tosc	—	7 Tosc	-	Timers in Sync mode

These parameters are characterized but not tested.

† Data in "Typ" column is at 5V, 25°C unless otherwise stated. These parameters are for design guidance only and are not tested.



## TABLE 15-6: CAPTURE/COMPARE/PWM REQUIREMENTS (ECCP)

	r <b>d Opera</b> ng Tempe	ting Conditions (unless erature $-40^{\circ}C \le TA \le +1$	•					
Param No.	Sym	Character	ristic	Min	Тур†	Max	Units	Conditions
CC01*	TccL	CCP1 Input Low Time	No Prescaler	0.5Tcy + 20	_	_	ns	
			With Prescaler	20		_	ns	
CC02*	TccH	CCP1 Input High Time	No Prescaler	0.5Tcy + 20			ns	
			With Prescaler	20			ns	
CC03*	TccP	CCP1 Input Period		<u>3Tcy + 40</u> N		_	ns	N = prescale value (1, 4 or 16)

These parameters are characterized but not tested.

† Data in "Typ" column is at 5V, 25°C unless otherwise stated. These parameters are for design guidance only and are not tested.

## TABLE 15-7: COMPARATOR SPECIFICATIONS

#### Standard Operating Conditions (unless otherwise stated)

Operating Temperature	$-40^{\circ}C \le TA \le +125^{\circ}C$

Operatin	ng Tempera	ature $-40^{\circ}\text{C} \le 1\text{A} \le +125^{\circ}\text{C}$						
Param No.	Sym	Characteristics		Min	Тур†	Мах	Units	Comments
CM01	Vos	Input Offset Voltage <sup>(2)</sup>		—	± 5.0	± 10	mV	
CM02	Vсм	Input Common Mode Voltage		0	_	Vdd - 1.5	V	
CM03*	CMRR	Common Mode Rejection Ratio		+55	—		dB	
CM04*	TRT	Response Time <sup>(1)</sup>	Falling	_	150	600	ns	
			Rising	_	200	1000	ns	
CM05*	TMC2COV	Comparator Mode Change to Output Valid		_	_	10	μS	
CM06*	VHYS	Input Hysteresis Voltage			45	60	mV	

\* These parameters are characterized but not tested.

† Data in "Typ" column is at 5V, 25°C unless otherwise stated. These parameters are for design guidance only and are not tested.

**Note 1:** Response time is measured with one comparator input at (VDD - 1.5)/2 - 100 mV to (VDD - 1.5)/2 + 20 mV. The other input is at (VDD -1.5)/2.

2: Input offset voltage is measured with one comparator input at (VDD - 1.5V)/2.

#### TABLE 15-8: COMPARATOR VOLTAGE REFERENCE (CVREF) SPECIFICATIONS

	d Operat	$\begin{array}{ll} \mbox{ing Conditions (unless otherwise} \\ \mbox{ature} & -40^{\circ}C \leq TA \leq +125^{\circ}C \end{array}$	stated)				
Param No.	Sym	Characteristics	Min	Тур†	Max	Units	Comments
CV01	Clsb	Step Size <sup>(2)</sup>		Vdd/24 Vdd/32		V V	Low Range (VRR = 1) High Range (VRR = 0)
CV02	CACC	Absolute Accuracy <sup>(3)</sup>			± 1/2 ± 1/2	LSb LSb	Low Range (VRR = 1) High Range (VRR = 0)
CV03	CR	Unit Resistor Value (R)	_	2k	_	Ω	
CV04	CST	Settling Time <sup>(1)</sup>	—	—	10	μS	

† Data in "Typ" column is at 5V, 25°C unless otherwise stated. These parameters are for design guidance only and are not tested.

**Note 1:** Settling time measured while VRR = 1 and VR<3:0> transitions from '0000' to '1111'.

- 2: See Section 8.11 "Comparator Voltage Reference" for more information.
- **3:** Absolute Accuracy when CVREF output is  $\leq$  (VDD-1.5).

#### TABLE 15-9:VOLTAGE REFERENCE SPECIFICATIONS

VR Voltage Reference Specifications			Standard Operating Conditions (unless otherwise stated)Operating temperature $-40^{\circ}C \le TA \le +125^{\circ}C$						
Param No.	Symbol	Characteristics	Min	Тур	Max	Units	Comments		
VR01	VP6out	VP6 voltage output	0.50	0.6	0.7	V			
VR02	V1P2out	V1P2 voltage output	1.05	1.20	1.35	V			
VR03*	TSTABLE	Settling Time	_	10		μS			

These parameters are characterized but not tested.

SHUNT	REGULAT	OR CHARACTERISTICS	$\begin{array}{llllllllllllllllllllllllllllllllllll$						
Param No.	Symbol	Characteristics	Min	Тур	Max	Units	Comments		
SR01	VSHUNT	Shunt Voltage	4.75	5	5.4	V			
SR02	ISHUNT	Shunt Current	4	—	50	mA			
SR03*	TSETTLE	Settling Time			150	ns	To 1% of final value		
SR04	CLOAD	Load Capacitance	0.01	-	10	μF	Bypass capacitor on VDD pin		
SR05	$\Delta$ ISNT	Regulator operating current	—	180	—	μΑ	Includes band gap reference current		

### TABLE 15-10: SHUNT REGULATOR SPECIFICATIONS (PIC16HV610/616 only)

These parameters are characterized but not tested.

\*

### TABLE 15-11: PIC16F616/16HV616 A/D CONVERTER (ADC) CHARACTERISTICS:

		$\begin{array}{ll} \mbox{rating Conditions (unless)} \\ \mbox{verature} & -40^{\circ}\mbox{C} \leq \mbox{TA} \leq -40^{\circ}\mbox{C} < -40^{\circ}\mbox{C} $		vise stated	1)		
Param No.	Sym	Characteristic	Min	Тур†	Мах	Units	Conditions
AD01	Nr	Resolution	_	_	10 bits	bit	
AD02	EIL	Integral Error	_	_	±1	LSb	VREF = 5.12V <sup>(5)</sup>
AD03	Edl	Differential Error	_	—	±1	LSb	No missing codes to 10 bits VREF = 5.12V <sup>(5)</sup>
AD04	EOFF	Offset Error	_	+1.5	+ 2.0	LSb	VREF = 5.12V <sup>(5)</sup>
AD07	Egn	Gain Error	_	_	±1	LSb	VREF = 5.12V <sup>(5)</sup>
AD06 AD06A	Vref	Reference Voltage <sup>(3)</sup>	2.2 2.5	_	— Vdd	V	Absolute minimum to ensure 1 LSb accuracy
AD07	VAIN	Full-Scale Range	Vss	_	Vref	V	
AD08	Zain	Recommended Impedance of Analog Voltage Source	—		10	kΩ	
AD09*	IREF	VREF Input Current <sup>(3)</sup>	10	—	1000	μΑ	During VAIN acquisition. Based on differential of VHOLD to VAIN.
			—	_	50	μΑ	During A/D conversion cycle.

\* These parameters are characterized but not tested.

† Data in "Typ" column is at 5.0V, 25°C unless otherwise stated. These parameters are for design guidance only and are not tested.

**Note 1:** Total Absolute Error includes integral, differential, offset and gain errors.

**2:** The A/D conversion result never decreases with an increase in the input voltage and has no missing codes.

3: ADC VREF is from external VREF or VDD pin, whichever is selected as reference input.

4: When ADC is off, it will not consume any current other than leakage current. The power-down current specification includes any such leakage from the ADC module.

**5:** VREF = 5V for PIC16HV616.

## TABLE 15-12: PIC16F616/16HV616 A/D CONVERSION REQUIREMENTS

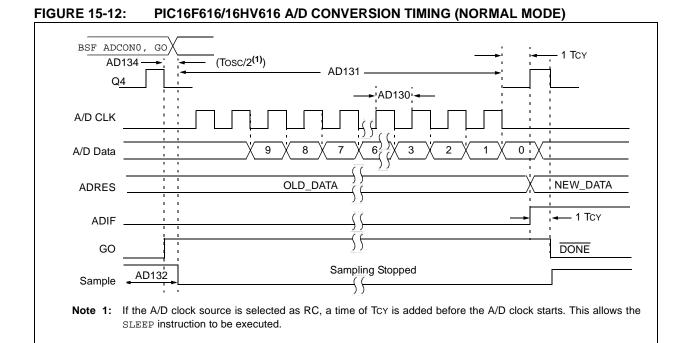
Param No.	Sym	Characteristic	Min	Тур†	Max	Units	Conditions
AD130*	TAD	A/D Clock Period	1.6	—	9.0	μS	Tosc-based, VREF $\geq$ 3.0V
			3.0	—	9.0	μS	Tosc-based, VREF full range
		A/D Internal RC Oscillator Period	3.0	6.0	9.0	μS	ADCS<1:0> = 11 (ADRC mode) At VDD = 2.5V
			1.6	4.0	6.0	μS	At VDD = 5.0V
AD131	TCNV	Conversion Time (not including Acquisition Time) <sup>(1)</sup>	_	11	_	Tad	Set GO/DONE bit to new data in A/D Result register
AD132*	TACQ	Acquisition Time		11.5		μS	
AD133*	TAMP	Amplifier Settling Time	_	—	5	μS	
AD134	Tgo	Q4 to A/D Clock Start	—	Tosc/2		—	
			_	Tosc/2 + Tcy		_	If the A/D clock source is selected as RC, a time of TcY is added before the A/D clock starts. This allows the SLEEP instruction to be executed.

\* These parameters are characterized but not tested.

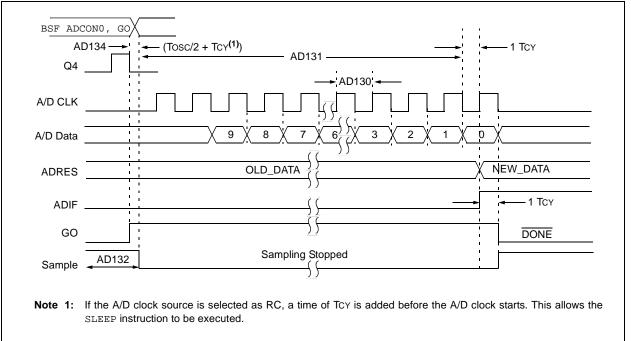
† Data in "Typ" column is at 5.0V, 25°C unless otherwise stated. These parameters are for design guidance only and are not tested.

Note 1: ADRESH and ADRESL registers may be read on the following TCY cycle.

2: See Section 9.3 "A/D Acquisition Requirements" for minimum conditions.







## 15.13 High Temperature Operation

This section outlines the specifications for <u>the</u> <u>PIC16F616 device operating in a temperature range</u> <u>between -40°C and 150°C.<sup>(4)</sup></u> The specifications between -40°C and 150°C<sup>(4)</sup> are identical to those shown in DS41302 and DS80329.

Note 1:	Writes	are	<u>not</u>	allowed	for	Flash			
	Program Memory above 125°C.								
2.		imina	enaci	fications ar	o inci	hasear			

- All AC timing specifications are increased by 30%. This derating factor will include parameters such as TPWRT.
- **3:** The temperature range indicator in the part number is "H" for -40°C to 150°C.<sup>(4)</sup>

Example: PIC16F616T-H/ST indicates the device is shipped in a tAPE and reel configuration, in the TSSOP package, and is rated for operation from -40°C to 150°C.<sup>(4)</sup>

4: AEC-Q100 reliability testing for devices intended to operate at 150°C is 1,000 hours. Any design in which the total operating time from 125°C to 150°C will be greater than 1,000 hours is not warranted without prior written approval from Microchip Technology Inc.

#### TABLE 15-13: ABSOLUTE MAXIMUM RATINGS

Parameter	Source/Sink	Value	Units
Max. Current: VDD	Source	20	mA
Max. Current: Vss	Sink	50	mA
Max. Current: PIN	Source	5	mA
Max. Current: PIN	Sink	10	mA
Pin Current: at Voн	Source	3	mA
Pin Current: at VoL	Sink	8.5	mA
Port Current: A and C	Source	20	mA
Port Current: A and C	Sink	50	mA
Maximum Junction Temperature		155	°C

**Note:** Stresses above those listed under *Absolute Maximum Ratings* may cause permanent damage to the device. This is a stress rating only, and functional operation of the device at those or any other conditions above those indicated in the operation listings of this specification is not implied. Exposure above maximum rating conditions for extended periods may affect device reliability.

Param Device		Unite	Min	Turn	Max	Condition			
No.	Characteristics	Units	IVIIII	Тур	Max	Vdd	Note		
D010			_	13	58	2.0			
	Supply Current (IDD)	μA	_	19	67	3.0	IDD LP OSC (32 kHz)		
			_	32	92	5.0			
D011				135	316	2.0			
		μA	_	185	400	3.0	IDD XT OSC (1 MHz)		
			_	300	537	5.0			
D012				240	495	2.0			
		μA	_	360	680	3.0	IDD XT OSC (4 MHz)		
		mA	_	0.660	1.20	5.0			
D013				75	158	2.0			
		μA	_	155	338	3.0	IDD EC OSC (1 MHz)		
			_	345	792	5.0			
D014		μA		185	357	2.0			
		μΛ	_	325	625	3.0	IDD EC OSC (4 MHz)		
		mA	_	0.665	1.30	5.0			
D016				245	476	2.0			
		μΑ	_	360	672	3.0	IDD INTOSC (4 MHz)		
			_	620	1.10	5.0			
D017		μΑ		395	757	2.0			
		mA	_	0.620	1.20	3.0	IDD INTOSC (8 MHz)		
			_	1.20	2.20	5.0			
D018				175	332	2.0			
		μA		285	518	3.0	IDD EXTRC (4 MHz)		
				530	972	5.0			
D019		mA		2.20	4.10	4.5	IDD HS OSC (20 MHz)		
				2.80	4.80	5.0			

## TABLE 15-14: DC CHARACTERISTICS FOR IDD SPECIFICATIONS FOR PIC16F616 – H (High Temp.)

Param	Device	1 Juniter		<b>T</b>			Condition	
No.	Characteristics	Units	Min	Тур	Max	Vdd	Note	
D020E			_	0.05	12	2.0		
	Power Down IPD	μA	_	0.15	13	3.0	IPD Base	
			_	0.35	14	5.0		
D021E			—	0.5	20	2.0		
		μΑ	_	2.5	25	3.0	WDT Current	
			_	9.5	36	5.0		
D022E		μA	—	5.0	28	3.0	BOR Current	
		μΛ	_	6.0	36	5.0		
D023E			_	105	195	2.0		
		μA	_	110	210	3.0	IPD Current (Both Comparators Enabled)	
		μΑ	_	116	220	5.0		
			_	50	105	2.0		
			_	55	110	3.0	IPD Current (One Comparator Enabled)	
				60	125	5.0		
D024E				30	58	2.0		
		μA	_	45	85	3.0	IPD (CVREF, High Range)	
				75	142	5.0		
D025E			_	39	76	2.0		
		μA		59	114	3.0	IPD (CVREF, Low Range)	
			_	98	190	5.0	<u> </u>	
D026E			_	5.5	30	2.0		
		μA	_	7.0	35	3.0	IPD (T1 OSC, 32 kHz)	
				8.5	45	5.0		
D027E		μA	—	0.2	12	3.0	IPD (A2D on, not converting)	
		μι	_	0.3	15	5.0		

#### TABLE 15-15: DC CHARACTERISTICS FOR IPD SPECIFICATIONS FOR PIC16F616 - H (High Temp.)

### TABLE 15-16: WATCHDOG TIMER SPECIFICATIONS FOR PIC16F616 – H (High Temp.)

Param No.	Sym	Characteristic	Units	Min	Тур	Max	Conditions
31	Twdt	Watchdog Timer Time-out Period (No Prescaler)	ms	6	20	70	150°C Temperature

#### TABLE 15-17: LEAKAGE CURRENT SPECIFICATIONS FOR PIC16F616 – H (High Temp.)

Param No.	Sym	Characteristic	Units	Min	Тур	Max	Conditions
D061	lı∟	Input Leakage Current <sup>(1)</sup> (GP3/RA3/MCLR)	μA	—	±0.5	±5.0	$Vss \le Vpin \le Vdd$
D062	lı∟	Input Leakage Current <sup>(2)</sup> (GP3/RA3/MCLR)	μA	50	250	400	VDD = 5.0V

**Note 1:** This specification applies when GP3/RA3/MCLR is configured as an input with the pull-up disabled. The leakage current for the GP3/RA3/MCLR pin is higher than for the standard I/O port pins.

2: This specification applies when GP3/RA3/MCLR is configured as the MCLR Reset pin function with the weak pull-up enabled.

TABLE 13-18. OSCILLATOR FARAMETERS FOR FICTOR 010 - In (ringh Temp.)								
Param No.	Sym	Characteristic	Frequency Tolerance	Units	Min	Тур	Max	Conditions
OS08	INTosc	Int. Calibrated INTOSC Freq. <sup>(1)</sup>	±10%	MHz	7.2	8.0		$\begin{array}{l} 2.0V \leq V \text{DD} \leq 5.5V \\ -40^{\circ}\text{C} \leq T \text{A} \leq 150^{\circ}\text{C} \end{array}$

## TABLE 15-18: OSCILLATOR PARAMETERS FOR PIC16F616 – H (High Temp.)

Note 1: To ensure these oscillator frequency tolerances, VDD and Vss must be capacitively decoupled as close to the device as possible. 0.1 μF and 0.01 μF values in parallel are recommended.

#### TABLE 15-19: COMPARATOR SPECIFICATIONS FOR PIC16F616 – H (High Temp.)

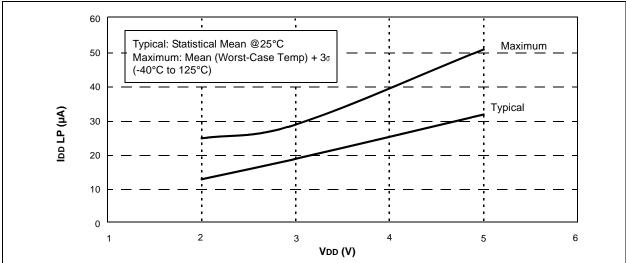
Param No.	Sym	Characteristic	Units	Min	Тур	Max	Conditions
CM01	Vos	Input Offset Voltage	mV		±5	±20	(VDD - 1.5)/2

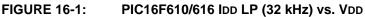
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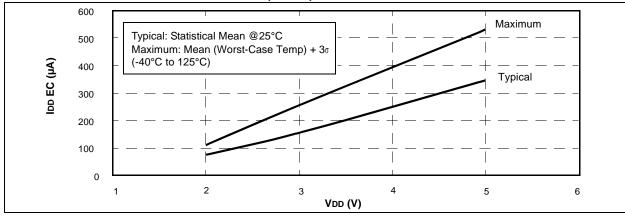
## 16.0 DC AND AC CHARACTERISTICS GRAPHS AND TABLES

**Note:** The graphs and tables provided following this note are a statistical summary based on a limited number of samples and are provided for informational purposes only. The performance characteristics listed herein are not tested or guaranteed. In some graphs or tables, the data presented may be outside the specified operating range (e.g., outside specified power supply range) and therefore, outside the warranted range.

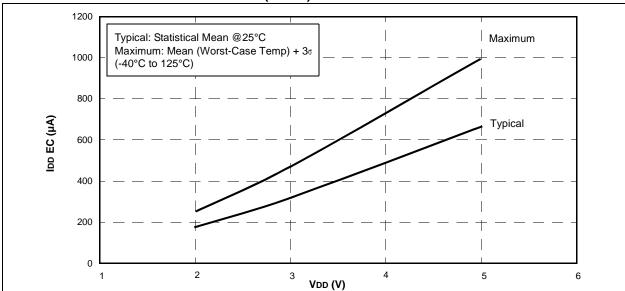
"Typical" represents the mean of the distribution at 25°C. "Maximum" or "minimum" represents (mean +  $3\sigma$ ) or (mean -  $3\sigma$ ) respectively, where s is a standard deviation, over each temperature range.

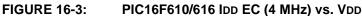


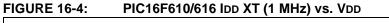


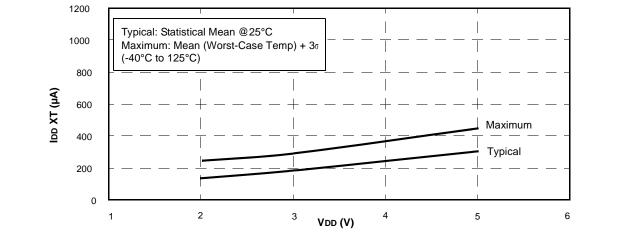


### FIGURE 16-2: PIC16F610/616 IDD EC (1 MHz) vs. VDD

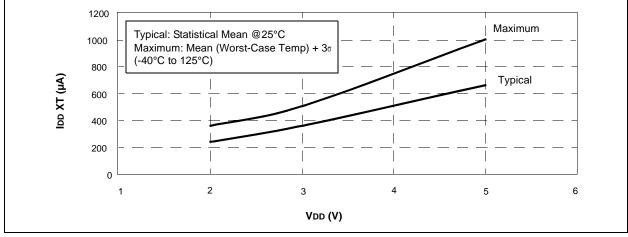


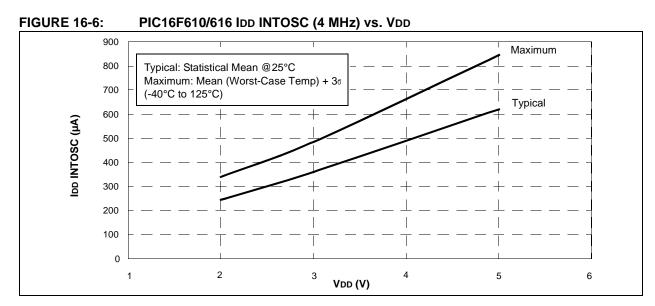


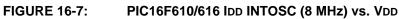


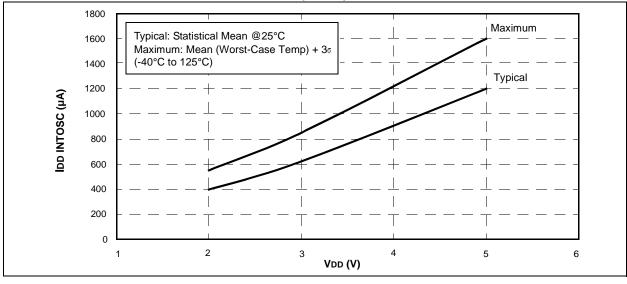


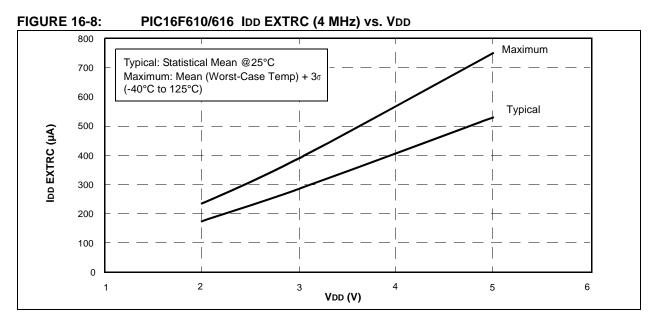




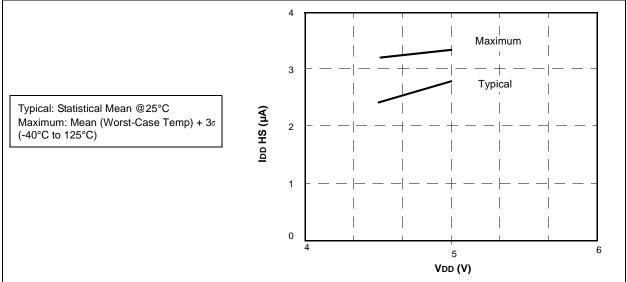


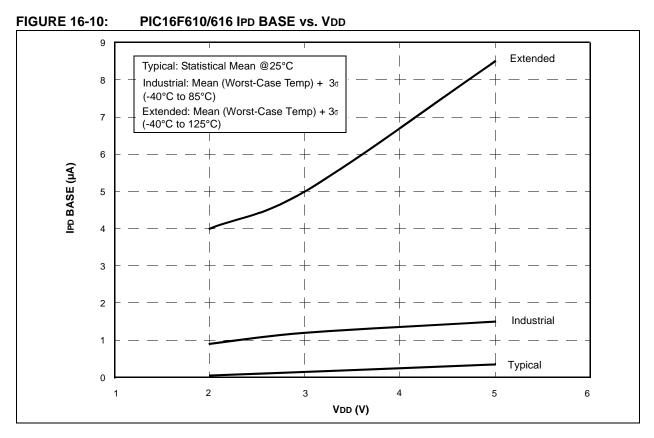




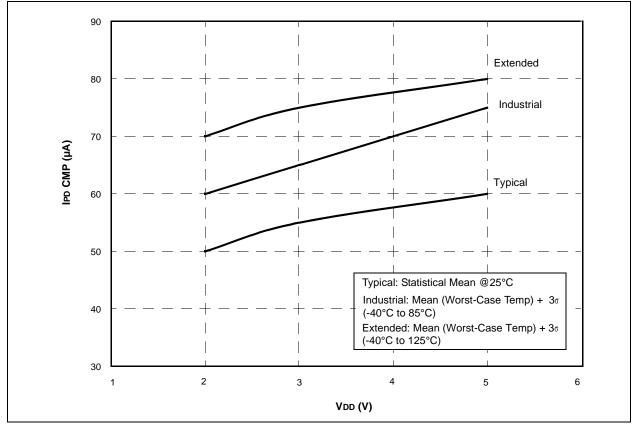




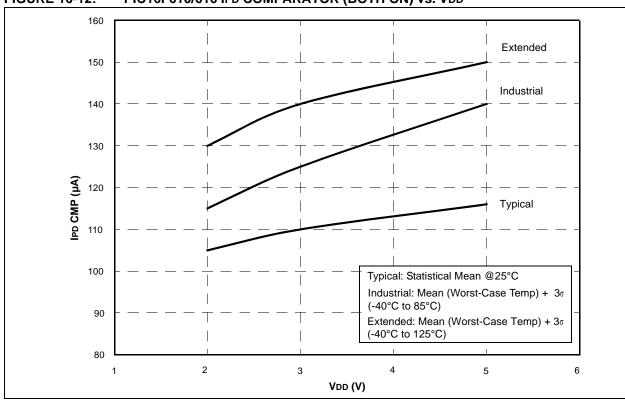






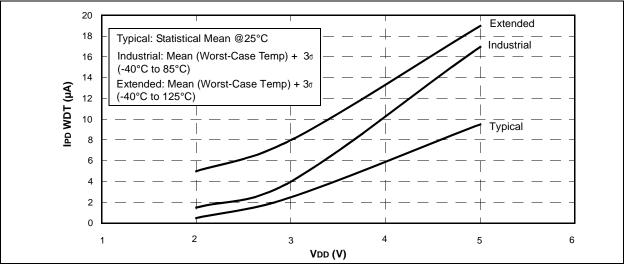


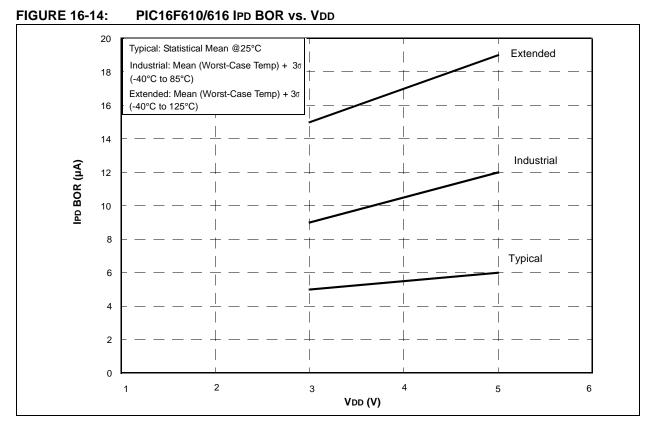
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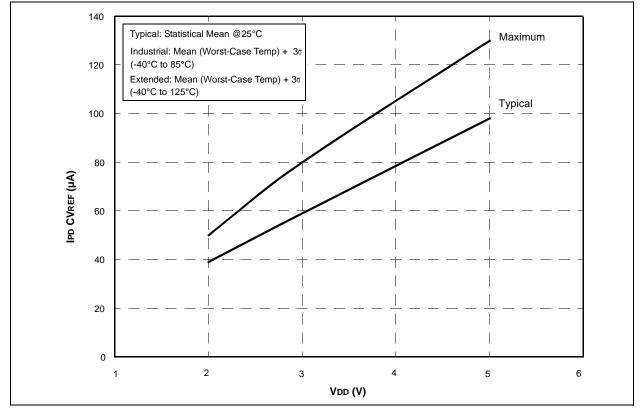
### FIGURE 16-12: PIC16F610/616 IPD COMPARATOR (BOTH ON) vs. VDD

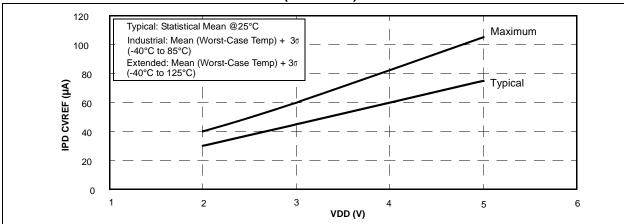






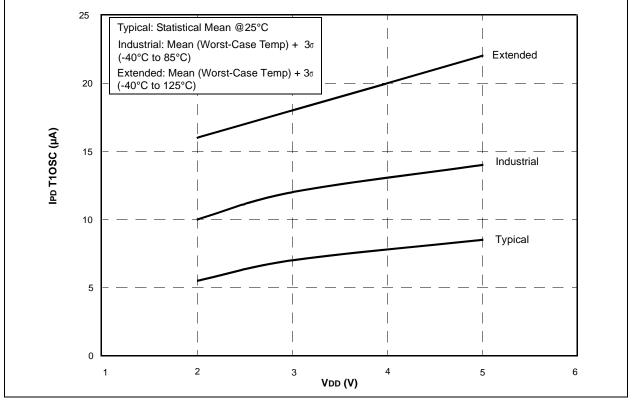


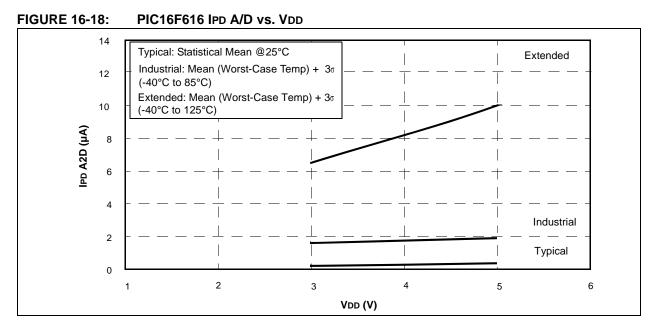




## FIGURE 16-16: PIC16F610/616 IPD CVREF (HI RANGE) vs. VDD

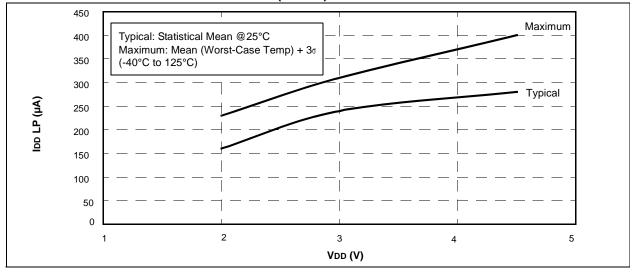


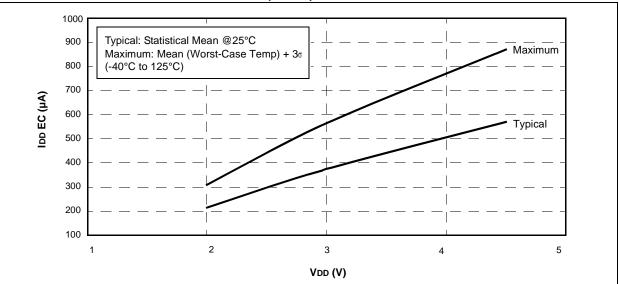






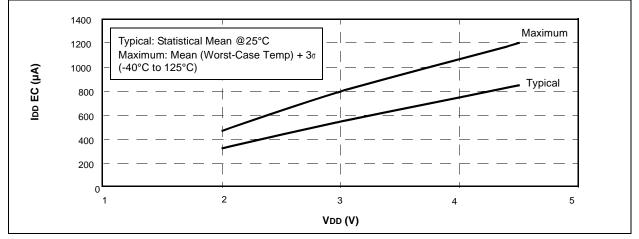
PIC16HV610/616 IDD LP (32 kHz) vs. VDD



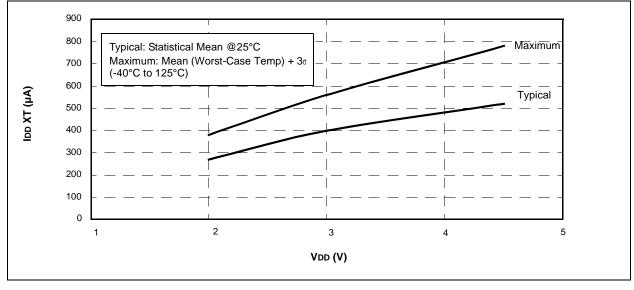


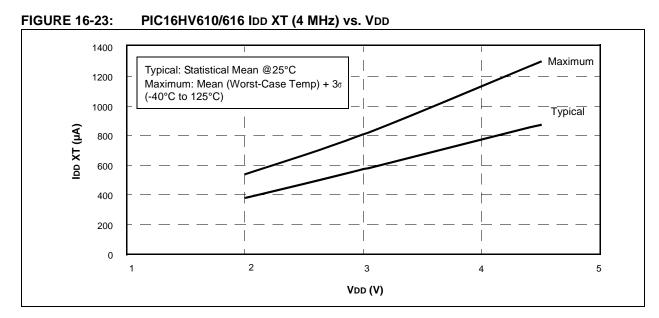
#### FIGURE 16-20: PIC16HV610/616 IDD EC (1 MHz) vs. VDD



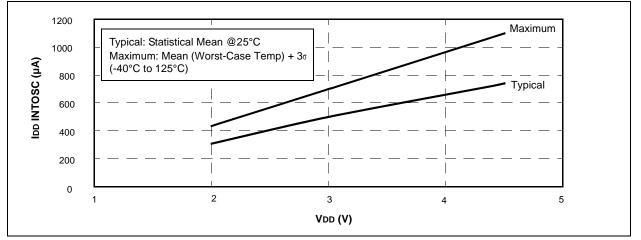




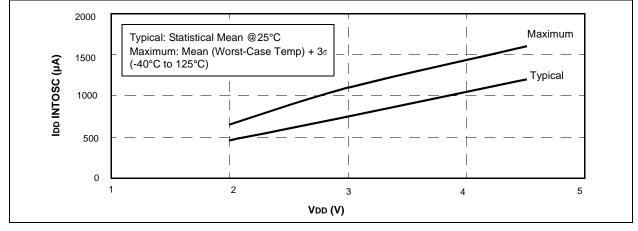


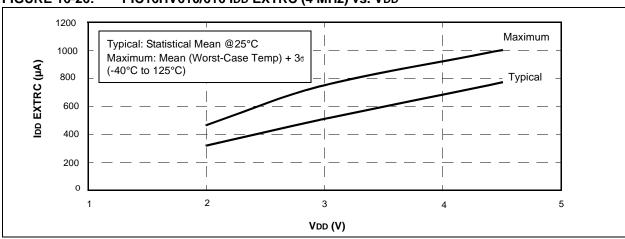






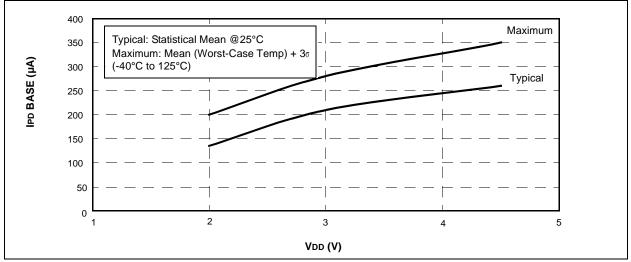


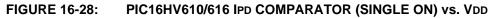


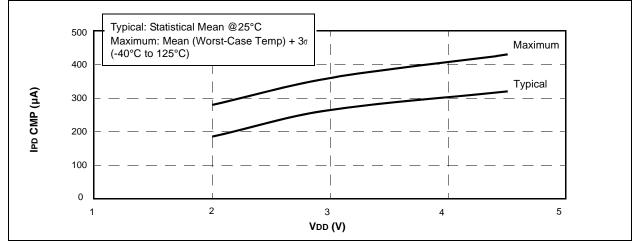


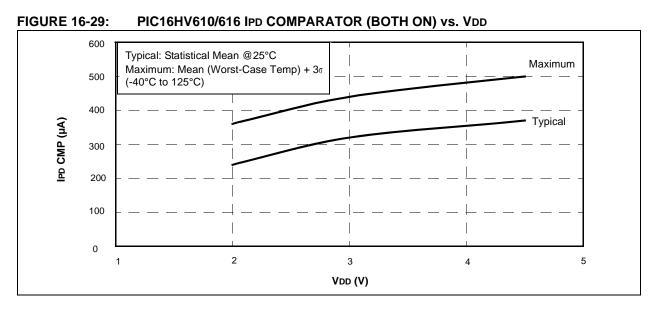
### FIGURE 16-26: PIC16HV610/616 IDD EXTRC (4 MHz) vs. VDD



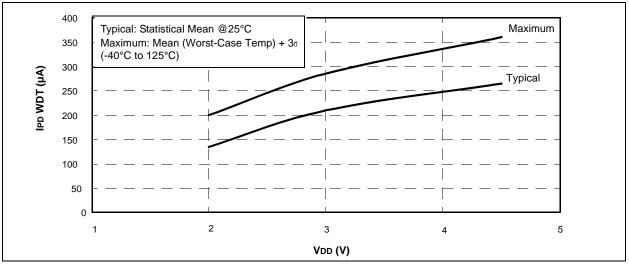




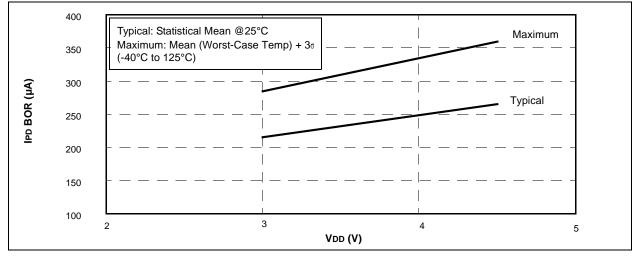






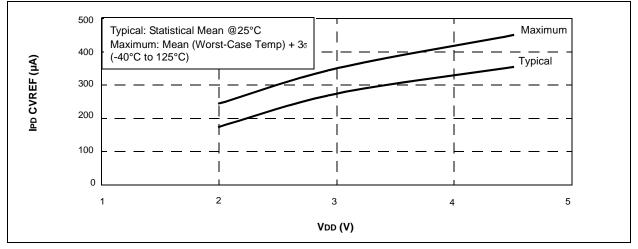


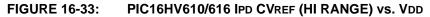


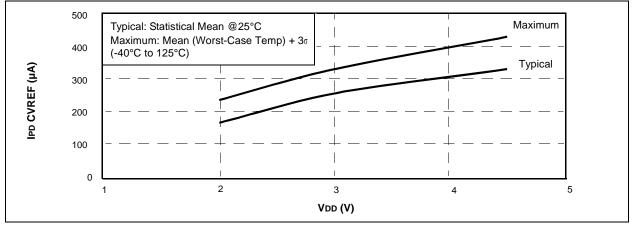


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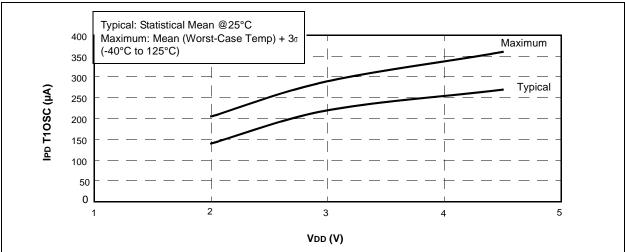
### FIGURE 16-32: PIC16HV610/616 IPD CVREF (LOW RANGE) vs. VDD

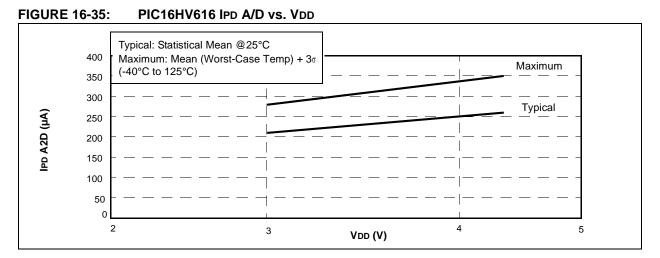




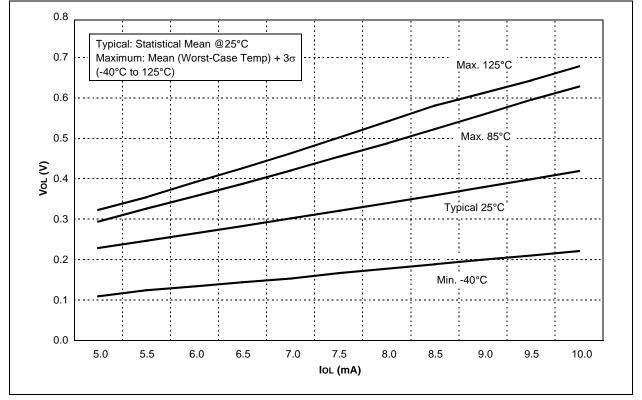


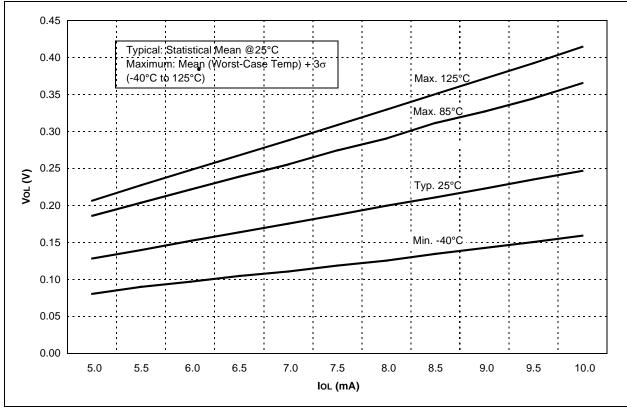






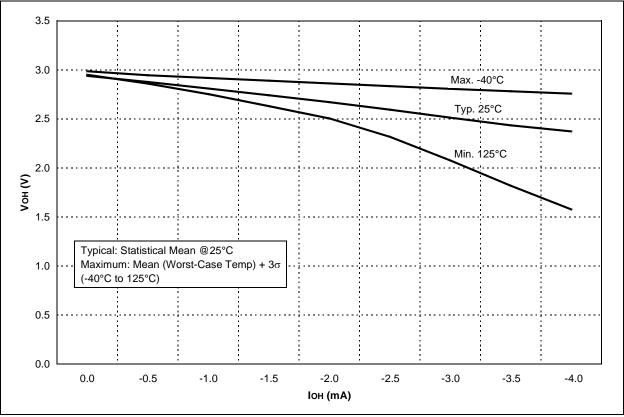


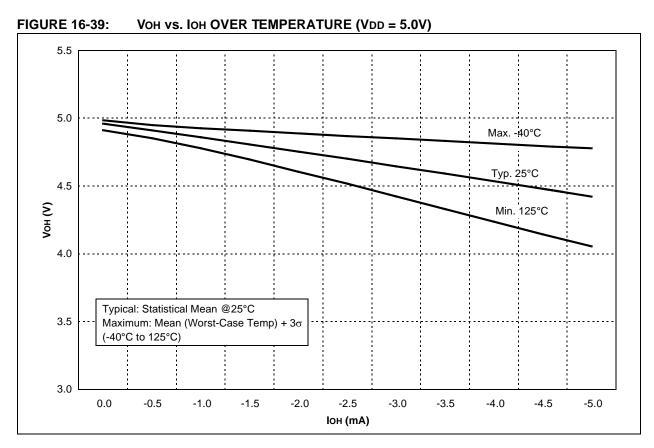




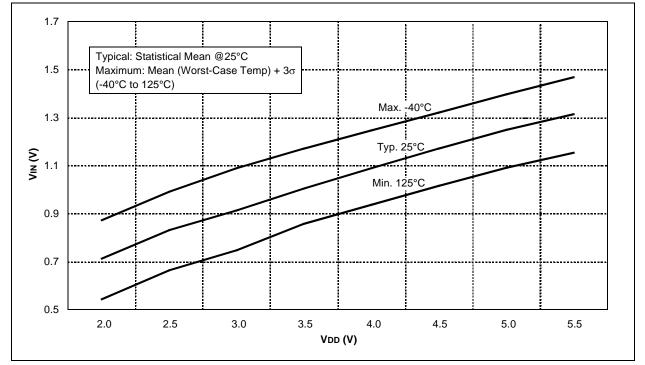


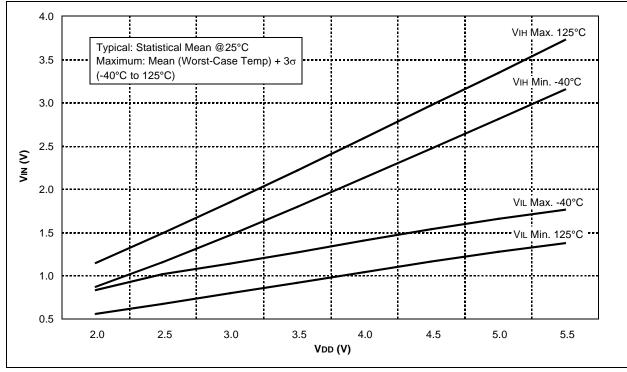






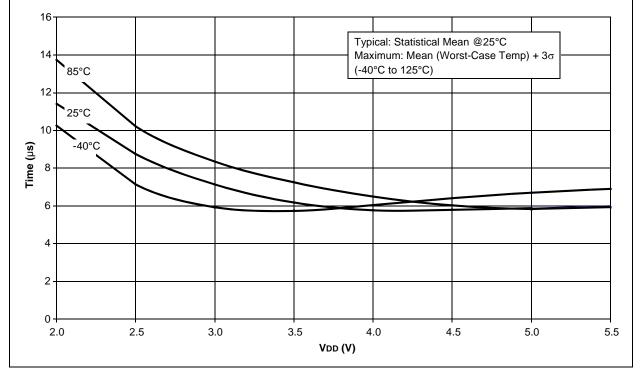


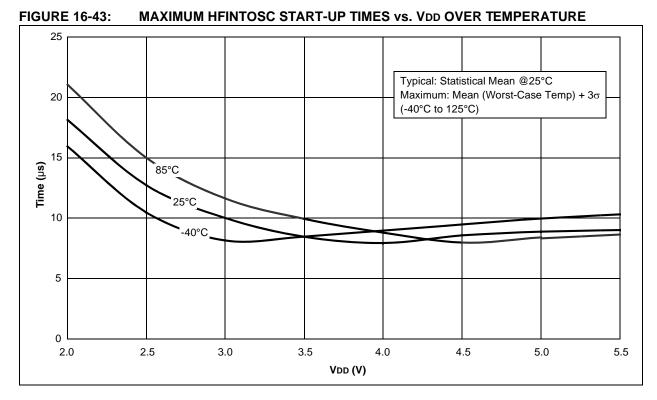




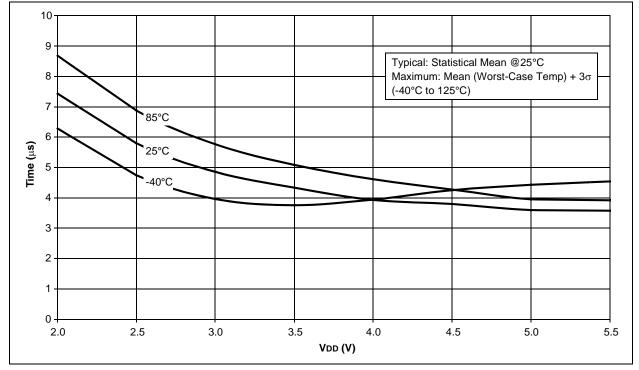
#### FIGURE 16-41: SCHMITT TRIGGER INPUT THRESHOLD VIN vs. VDD OVER TEMPERATURE











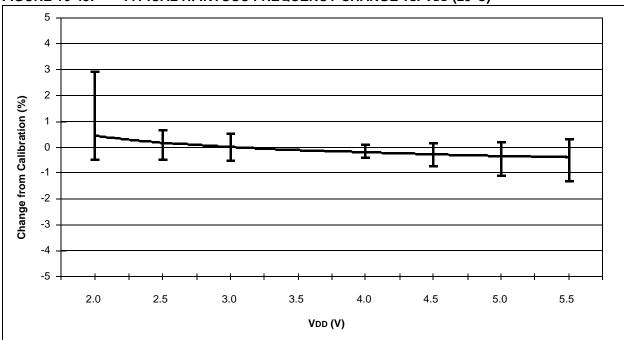
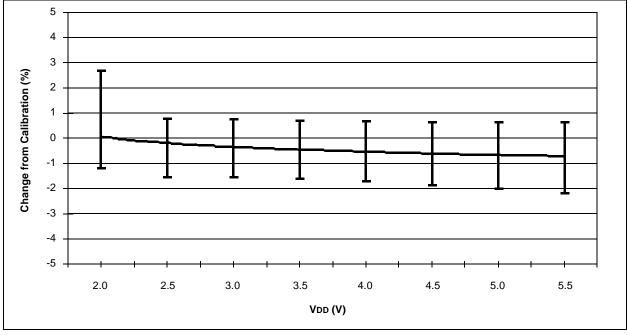


FIGURE 16-45: TYPICAL HFINTOSC FREQUENCY CHANGE vs. VDD (25°C)





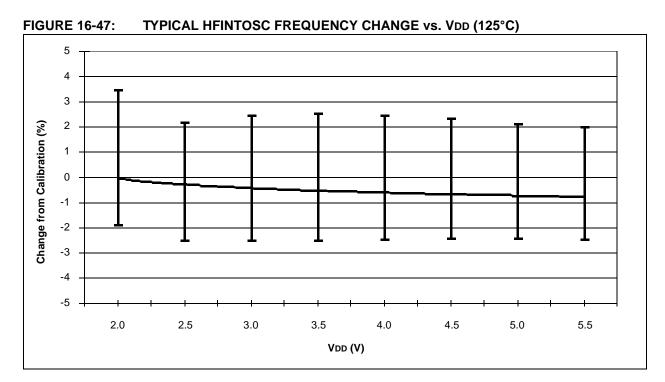
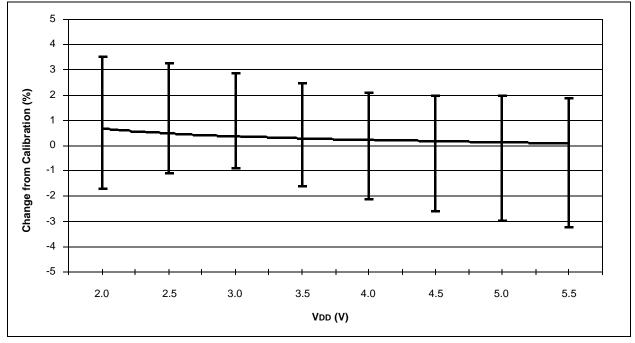
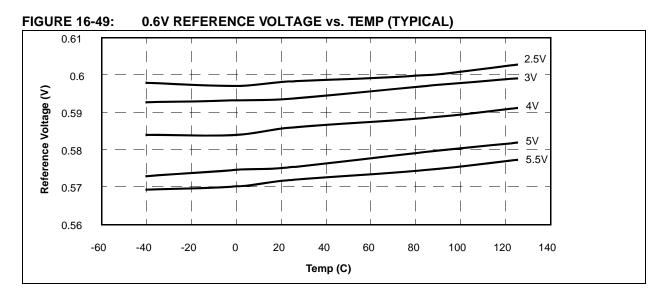
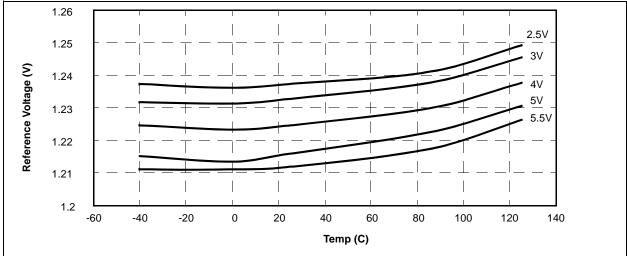


FIGURE 16-48: TYPICAL HFINTOSC FREQUENCY CHANGE vs. Vdd (-40°C)

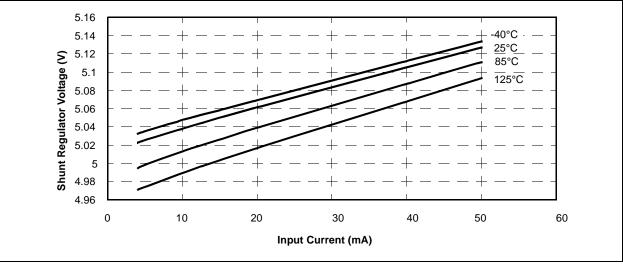


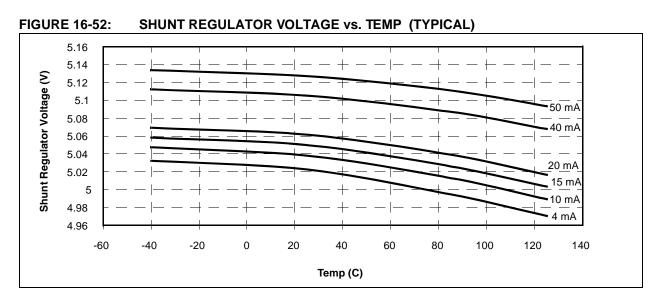




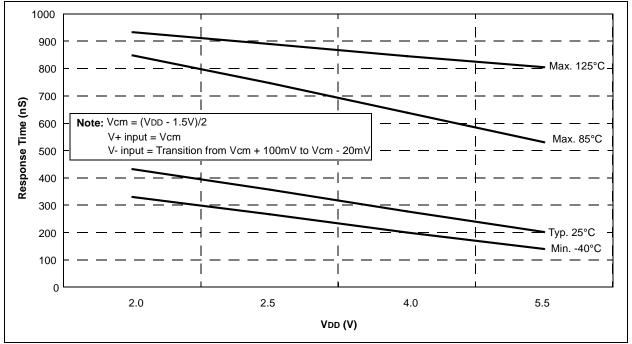


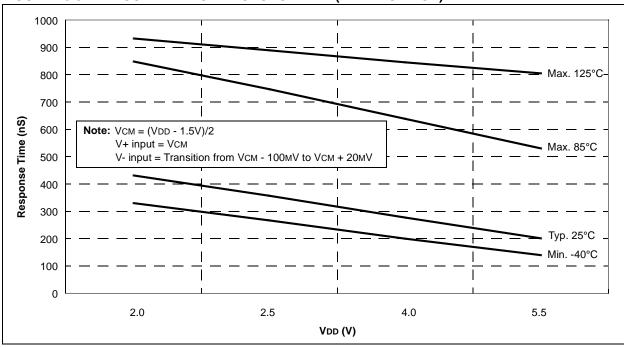






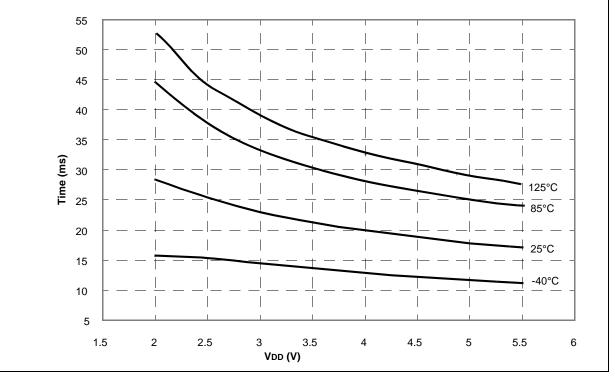






#### FIGURE 16-54: COMPARATOR RESPONSE TIME (FALLING EDGE)



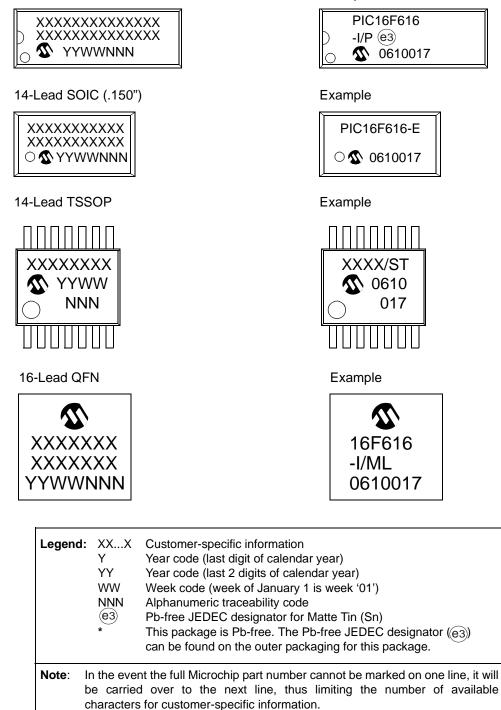


Example

## **17.0 PACKAGING INFORMATION**

### 17.1 Package Marking Information

#### 14-Lead PDIP



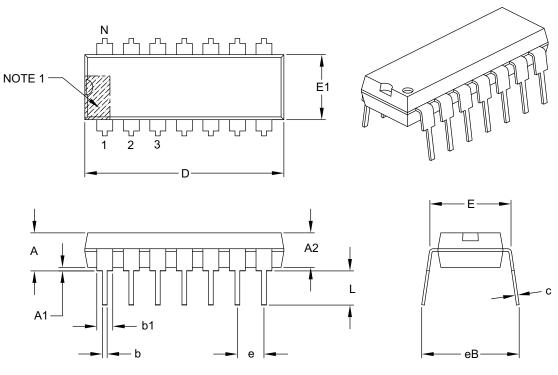
\* Standard PIC<sup>®</sup> device marking consists of Microchip part number, year code, week code, and traceability code. For PIC<sup>®</sup> device marking beyond this, certain price adders apply. Please check with your Microchip Sales Office. For QTP devices, any special marking adders are included in QTP price.

#### 17.2 Package Details

The following sections give the technical details of the packages.

### 14-Lead Plastic Dual In-Line (P) – 300 mil Body [PDIP]

**Note:** For the most current package drawings, please see the Microchip Packaging Specification located at http://www.microchip.com/packaging



	Units		INCHES	
Dimensio	on Limits	MIN	NOM	MAX
Number of Pins	Ν		14	
Pitch	е		.100 BSC	
Top to Seating Plane	А	-	-	.210
Molded Package Thickness	A2	.115	.130	.195
Base to Seating Plane	A1	.015	-	-
Shoulder to Shoulder Width	E	.290	.310	.325
Molded Package Width	E1	.240	.250	.280
Overall Length	D	.735	.750	.775
Tip to Seating Plane	L	.115	.130	.150
Lead Thickness	С	.008	.010	.015
Upper Lead Width	b1	.045	.060	.070
Lower Lead Width	b	.014	.018	.022
Overall Row Spacing §	eB	-	-	.430

#### Notes:

1. Pin 1 visual index feature may vary, but must be located with the hatched area.

2. § Significant Characteristic.

3. Dimensions D and E1 do not include mold flash or protrusions. Mold flash or protrusions shall not exceed .010" per side.

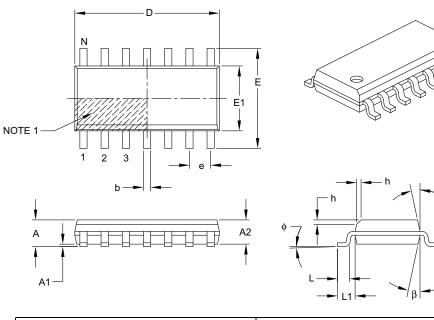
4. Dimensioning and tolerancing per ASME Y14.5M.

BSC: Basic Dimension. Theoretically exact value shown without tolerances.

Microchip Technology Drawing C04-005B

### 14-Lead Plastic Small Outline (SL) – Narrow, 3.90 mm Body [SOIC]

**Note:** For the most current package drawings, please see the Microchip Packaging Specification located at http://www.microchip.com/packaging



	Units		MILLIMETERS		
	Dimension Limits	MIN	NOM	MAX	
Number of Pins	N		14		
Pitch	е		1.27 BSC		
Overall Height	A	-	-	1.75	
Molded Package Thickness	A2	1.25	-	-	
Standoff §	A1	0.10	-	0.25	
Overall Width	E	6.00 BSC			
Molded Package Width	E1	3.90 BSC			
Overall Length	D	8.65 BSC			
Chamfer (optional)	h	0.25 – 0.50		0.50	
Foot Length	L	0.40 – 1.27		1.27	
Footprint	L1	1.04 REF			
Foot Angle	¢	0°	-	8°	
Lead Thickness	С	0.17	-	0.25	
Lead Width	b	0.31	-	0.51	
Mold Draft Angle Top	α	5°	-	15°	
Mold Draft Angle Bottom	β	5°	-	15°	

Notes:

1. Pin 1 visual index feature may vary, but must be located within the hatched area.

2. § Significant Characteristic.

3. Dimensions D and E1 do not include mold flash or protrusions. Mold flash or protrusions shall not exceed 0.15 mm per side.

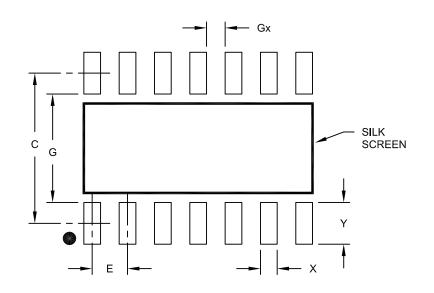
- 4. Dimensioning and tolerancing per ASME Y14.5M.
  - BSC: Basic Dimension. Theoretically exact value shown without tolerances.

REF: Reference Dimension, usually without tolerance, for information purposes only.

Microchip Technology Drawing C04-065B

14-Lead Plastic Small Outline (SL) - Narrow, 3.90 mm Body [SOIC]

**Note:** For the most current package drawings, please see the Microchip Packaging Specification located at http://www.microchip.com/packaging



## RECOMMENDED LAND PATTERN

	Units	N		S
Dimension	Limits	MIN	NOM	MAX
Contact Pitch	E		1.27 BSC	
Contact Pad Spacing	С		5.40	
Contact Pad Width	Х			0.60
Contact Pad Length	Y			1.50
Distance Between Pads	Gx	0.67		
Distance Between Pads	G	3.90		

Notes:

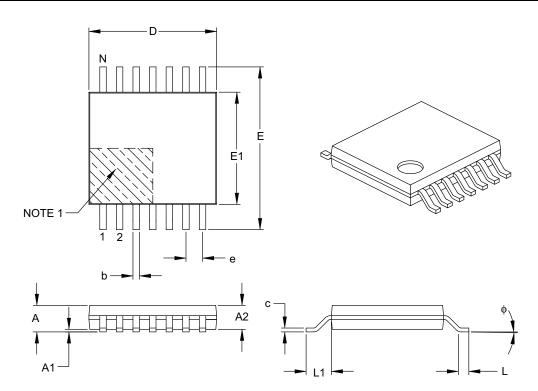
1. Dimensioning and tolerancing per ASME Y14.5M

BSC: Basic Dimension. Theoretically exact value shown without tolerances.

Microchip Technology Drawing No. C04-2065A

### 14-Lead Plastic Thin Shrink Small Outline (ST) – 4.4 mm Body [TSSOP]

**Note:** For the most current package drawings, please see the Microchip Packaging Specification located at http://www.microchip.com/packaging



	Units		MILLIMETERS		
Dimensio	n Limits	MIN	NOM	MAX	
Number of Pins	Ν		14		
Pitch	е		0.65 BSC		
Overall Height	Α	-	-	1.20	
Molded Package Thickness	A2	0.80	1.00	1.05	
Standoff	A1	0.05	-	0.15	
Overall Width	Е	6.40 BSC			
Molded Package Width	E1	4.30	4.40	4.50	
Molded Package Length	D	4.90	5.00	5.10	
Foot Length	L	0.45	0.60	0.75	
Footprint	L1	1.00 REF			
Foot Angle	φ	0°	-	8°	
Lead Thickness	С	0.09	-	0.20	
Lead Width	b	0.19	-	0.30	

#### Notes:

1. Pin 1 visual index feature may vary, but must be located within the hatched area.

2. Dimensions D and E1 do not include mold flash or protrusions. Mold flash or protrusions shall not exceed 0.15 mm per side.

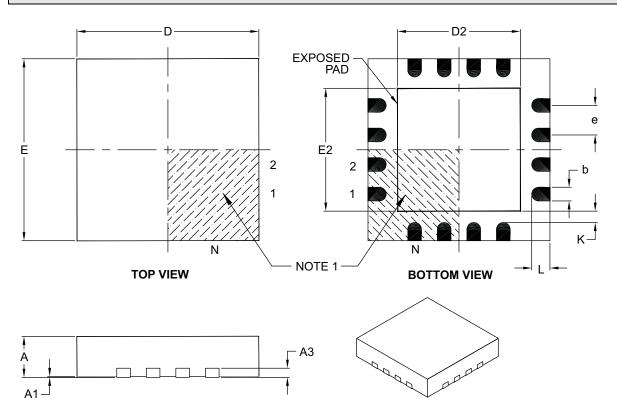
- 3. Dimensioning and tolerancing per ASME Y14.5M.
  - BSC: Basic Dimension. Theoretically exact value shown without tolerances.

REF: Reference Dimension, usually without tolerance, for information purposes only.

Microchip Technology Drawing C04-087B

#### 16-Lead Plastic Quad Flat, No Lead Package (ML) – 4x4x0.9 mm Body [QFN]

**Note:** For the most current package drawings, please see the Microchip Packaging Specification located at http://www.microchip.com/packaging



	Units MILLIMETERS		6	
Dimensio	n Limits	MIN	NOM	MAX
Number of Pins	Ν		16	
Pitch	е		0.65 BSC	
Overall Height	Α	0.80	0.90	1.00
Standoff	A1	0.00	0.02	0.05
Contact Thickness	A3		0.20 REF	
Overall Width	Е		4.00 BSC	
Exposed Pad Width	E2	2.50	2.65	2.80
Overall Length	D	4.00 BSC		
Exposed Pad Length	D2	2.50	2.65	2.80
Contact Width	b	0.25	0.30	0.35
Contact Length	L	0.30	0.40	0.50
Contact-to-Exposed Pad	К	0.20	-	-

#### Notes:

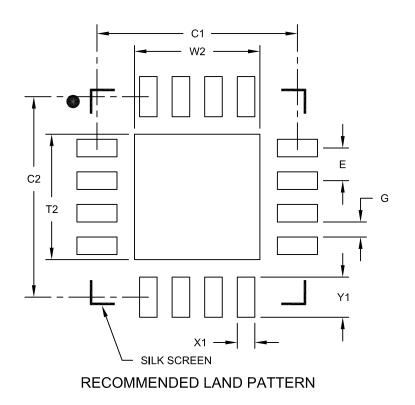
- 1. Pin 1 visual index feature may vary, but must be located within the hatched area.
- 2. Package is saw singulated.
- 3. Dimensioning and tolerancing per ASME Y14.5M.
  - BSC: Basic Dimension. Theoretically exact value shown without tolerances.

REF: Reference Dimension, usually without tolerance, for information purposes only.

Microchip Technology Drawing C04-127B

#### 16-Lead Plastic Quad Flat, No Lead Package (ML) - 4x4x0.9mm Body [QFN]

**Note:** For the most current package drawings, please see the Microchip Packaging Specification located at http://www.microchip.com/packaging



Units		MILLIMETERS		
Dimension Limits		MIN	NOM	MAX
Contact Pitch	E	E 0.65 BSC		
Optional Center Pad Width	W2			2.50
Optional Center Pad Length	T2			2.50
Contact Pad Spacing	C1		4.00	
Contact Pad Spacing	C2		4.00	
Contact Pad Width (X28)	X1			0.35
Contact Pad Length (X28)	Y1			0.80
Distance Between Pads	G	0.30		

Notes:

1. Dimensioning and tolerancing per ASME Y14.5M

BSC: Basic Dimension. Theoretically exact value shown without tolerances.

Microchip Technology Drawing No. C04-2127A

NOTES:

### APPENDIX A: DATA SHEET REVISION HISTORY

#### **Revision A**

This is a new data sheet.

#### Revision B (12/06)

Added PIC16F610/16HV610 parts. Replaced Package Drawings.

#### Revision C (03/2007)

Replaced Package Drawings (Rev. AM); Replaced Development Support Section; Revised Product ID System.

#### Revision D (06/2008)

Added Graphs; Revised 28-Pin ICD Pinout, Electrical Specifications Section; Package Details.

#### **Revision E (09/2009)**

Added section 15.13 (High Temperature Operation) to the Electrical Specifications Chapter; Other minor corrections.

#### **Revision F (11/2009)**

Updated Figure 16-52.

### APPENDIX B: MIGRATING FROM OTHER PIC<sup>®</sup> DEVICES

This discusses some of the issues in migrating from other  $\text{PIC}^{\textcircled{0}}$  devices to the <code>PIC16F6XX</code> Family of devices.

### B.1 PIC16F676 to PIC16F610/616/16HV610/616

#### TABLE B-1: FEATURE COMPARISON

Feature	PIC16F676	PIC16F610/16HV610	PIC16F616/16HV616
Max Operating Speed	20 MHz	20 MHz	20 MHz
Max Program Memory (Words)	1024	1024	2048
SRAM (bytes)	64	64	128
A/D Resolution	10-bit	None	10-bit
Timers (8/16-bit)	1/1	1/1	2/1
Oscillator Modes	8	8	8
Brown-out Reset	Y	Y	Y
Internal Pull-ups	RA0/1/2/4/5	RA0/1/2/4/5, MCLR	RA0/1/2/4/5, MCLR
Interrupt-on-change	RA0/1/2/3/4/5	RA0/1/2/3/4/5	RA0/1/2/3/4/5
Comparator	1	2	2
ECCP	N	Ν	Y
INTOSC Frequencies	4 MHz	4/8 MHz	4/8 MHz
Internal Shunt Regulator	Ν	Y (PIC16HV610)	Y (PIC16HV616)

**Note:** This device has been designed to perform to the parameters of its data sheet. It has been tested to an electrical specification designed to determine its conformance with these parameters. Due to process differences in the manufacture of this device, this device may have different performance characteristics than its earlier version. These differences may cause this device to perform differently in your application than the earlier version of this device.

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PART NO.	x <u>/xx xxx</u>	Examples:
Device	Temperature Package Pattern Range	<ul> <li>a) PIC16F610-E/P 301 = Extended Temp., PDIP package, 20 MHz, QTP pattern #301</li> <li>b) PIC16F616-E/P 301 = Extended Temp., PDIP package, 20 MHz, QTP pattern #301</li> <li>c) PIC16HV610-E/P 301 = Extended Temp.,</li> </ul>
Device:	PIC16F610/616/16HV610/616, PIC16F610/616/16HV610/ 616T <sup>(1)</sup>	<ul> <li>PDIP package, 20 MHz, QTP pattern #301</li> <li>d) PIC16HV616-E/P 301 = Extended Temp., PDIP package, 20 MHz, QTP pattern #301</li> <li>e) PIC16F610-I/SL = Industrial Temp., SOIC</li> </ul>
Temperature Range:	I = $-40^{\circ}$ C to $+85^{\circ}$ C (Industrial) E = $-40^{\circ}$ C to $+125^{\circ}$ C (Extended) H = $-40^{\circ}$ C to $+150^{\circ}$ C (High Temp.) <sup>(2)</sup>	<ul> <li>package, 20 MHz</li> <li>f) PIC16F616-I/SL = Industrial Temp., SOIC package, 20 MHz</li> <li>g) PIC16HV610-I/SL = Industrial Temp., SOIC package, 20 MHz</li> </ul>
Package:	ML = Quad Flat No Leads (QFN) P = Plastic DIP (PDIP) SL = 14-lead Small Outline (3.90 mm) (SOIC) ST = Thin Shrink Small Outline (4.4 mm) (TSSOP)	<ul> <li>h) PIC16HV616-I/SL = Industrial Temp., SOIC package, 20 MHz</li> <li>i) PIC16F610T-E/ST Tape and Reel, Extended Temp., TSSOP package, 20 MHz</li> <li>j) PIC16F616T-E/ST Tape and Reel, Extended</li> </ul>
Pattern:	QTP, SQTP or ROM Code; Special Requirements (blank otherwise)	<ul> <li>Temp., TSSOP package, 20 MHz</li> <li>PIC16HV610T-E/ST Tape and Reel, Extended Temp., TSSOP package, 20 MHz</li> <li>PIC16HV616T-E/ST Tape and Reel, Extended Temp., TSSOP package, 20 MHz</li> <li>m) PIC16F616 - H/SL = High Temp., SOIC package, 20 MHz.</li> </ul>
		Note 1:T = in tape and reel for TSSOP, SOIC and QFN packages only.2:High Temp. available for PIC16F616 only.



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