

# PIC18F87K22 Family Data Sheet

64/80-Pin, High-Performance, 1-Mbit Enhanced Flash Microcontrollers with 12-Bit A/D and nanoWatt XLP Technology

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# PIC18F87K22 FAMILY

### 64/80-Pin, High-Performance, 1-Mbit Enhanced Flash MCUs with 12-Bit A/D and nanoWatt XLP Technology

#### Low-Power Features:

- Power-Managed modes:
  - Run: CPU on, peripherals on
  - Idle: CPU off, peripherals on
  - Sleep: CPU off, peripherals off
- Two-Speed Oscillator Start-up
- Fail-Safe Clock Monitor
- Power-Saving Peripheral Module Disable (PMD)
- Ultra Low-Power Wake-up
- Fast Wake-up, 1 μs Typical
- · Low-Power WDT, 300 nA Typical
- Ultra Low 50 nA Input Leakage
- Run mode Currents Down to 5.5 μA, Typical
- Idle mode Currents Down to 1.7 μA Typical
- Sleep mode Currents Down to Very Low 20 nA, Typical
- RTCC Current Downs to Very Low 700 nA, Typical

#### **Special Microcontroller Features:**

- Operating Voltage Range: 1.8V to 5.5V
- On-Chip 3.3V Regulator
- · Operating Speed up to 64 MHz
- Up to 128 Kbytes On-Chip Flash Program Memory
- Data EEPROM of 1,024 Bytes
- 4K x 8 General Purpose Registers (SRAM)
- 10,000 Erase/Write Cycle Flash Program Memory, Minimum
- 1,000,000 Erase/write Cycle Data EEPROM Memory, Typical
- · Flash Retention: 40 Years, Minimum
- Three Internal Oscillators: LF-INTRC (31 kHz), MF-INTOSC (500 kHz) and HF-INTOSC (16 MHz)
- Self-Programmable under Software Control
- · Priority Levels for Interrupts
- 8 x 8 Single-Cycle Hardware Multiplier
- Extended Watchdog Timer (WDT):
  - Programmable period from 4 ms to 4,194s (about 70 minutes)
- In-Circuit Serial Programming<sup>™</sup> (ICSP<sup>™</sup>) via Two Pins
- In-Circuit Debug via Two Pins
- · Programmable:
  - BOR
  - LVD

	Program Memory Data Memory		12-Bit CCP/		MSSP			Е	tors	t, e	Bus	_				
Device	Flash (bytes)	# Single-Word Instructions	SRAM (bytes)	EEPROM (bytes)	I/O	A/D (ch)	ECCP (PWM)		SPI	Master I <sup>2</sup> C™	EUSAR	Comparators	Timers 8/16-Bit	External I	CTMU	RTCC
PIC18F65K22	32K	16,383	2K	1K	53	16	5/3	2	Y	Y	2	3	4/4	Ν	Y	Y
PIC18F66K22	64K	32,768	4K	1K	53	16	7/3	2	Y	Y	2	3	6/5	Ν	Y	Y
PIC18F67K22	128K	65,536	4K	1K	53	16	7/3	2	Y	Y	2	3	6/5	Ν	Y	Y
PIC18F85K22	32K	16,383	2K	1K	69	24	5/3	2	Υ	Y	2	3	4/4	Y	Y	Y
PIC18F86K22	64K	32,768	4K	1K	69	24	7/3	2	Y	Y	2	3	6/5	Y	Y	Y
PIC18F87K22	128K	65,536	4K	1K	69	24	7/3	2	Y	Y	2	3	6/5	Y	Y	Y

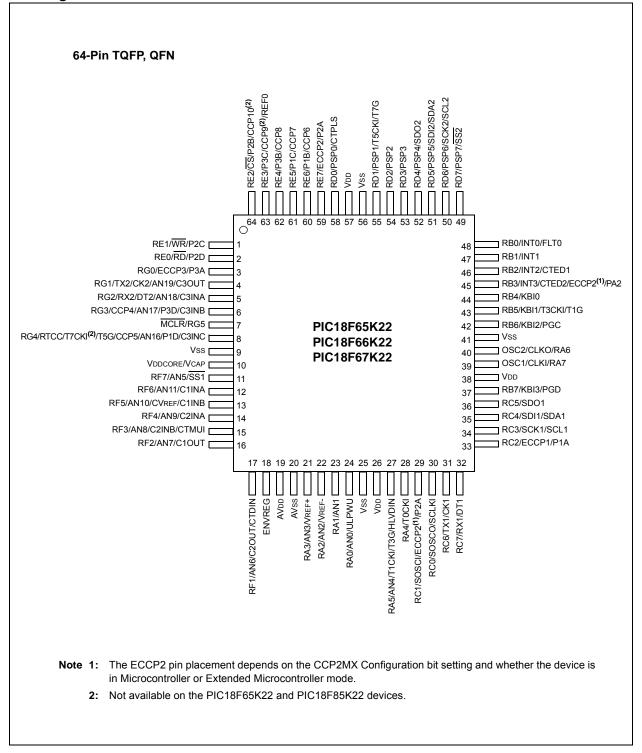
### **Peripheral Highlights:**

- Up to Ten CCP/ECCP modules:
  - Up to seven Capture/Compare/PWM (CCP) modules
  - Three Enhanced Capture/Compare/PWM (ECCP) modules
- Up to Eleven 8/16-Bit Timer/Counter modules:
  - Timer0 8/16-bit timer/counter with 8-bit programmable prescaler
  - Timer1,3 16-bit timer/counter
  - Timer2,4,6,8 8-bit timer/counter
  - Timer5,7 16-bit timer/counter for 64k and 128k parts
  - Timer10,12 8-bit timer/counter for 64k and 128k parts
- Three Analog Comparators
- Configurable Reference Clock Output
- Hardware Real-Time Clock and Calendar (RTCC) module with Clock, Calendar and Alarm Functions

- Charge Time Measurement Unit (CTMU):
  - Capacitance measurement for mTouch™ sensing solution
  - Time measurement with 1 ns typical resolution
  - Integrated temperature sensor
- High-Current Sink/Source 25 mA/25 mA (PORTB and PORTC)
- · Up to Four External Interrupts
- Two Master Synchronous Serial Port (MSSP) modules:
  - 3/4-wire SPI (supports all four SPI modes)
  - I<sup>2</sup>C<sup>™</sup> Master and Slave modes
- Two Enhanced Addressable USART modules:
  - LIN/J2602 support
  - Auto-Baud Detect (ABD)
- 12-Bit A/D Converter with up to 24 Channels:
  - Auto-acquisition and Sleep operationDifferential input mode of operation
- Integrated Voltage Reference

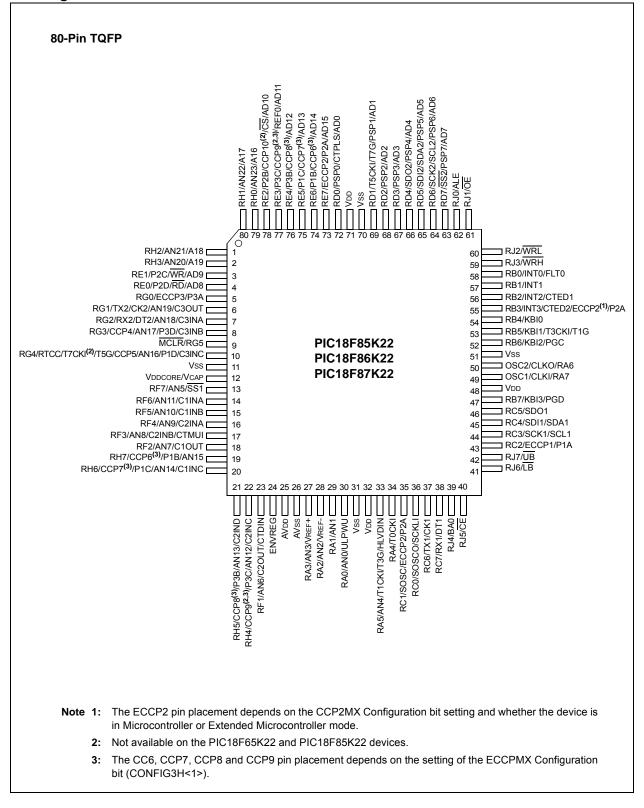
## PIC18F87K22 FAMILY

#### Pin Diagrams – PIC18F6XK22



## PIC18F87K22 FAMILY

#### Pin Diagrams – PIC18F8XK22



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### 1.0 DEVICE OVERVIEW

This document contains device-specific information for the following devices:

- PIC18F65K22 PIC18F85K22
- PIC18F66K22 PIC18F86K22
- PIC18F67K22 PIC18F87K22

This family combines the traditional advantages of all PIC18 microcontrollers – namely, high computational performance and a rich feature set – with an extremely competitive price point. These features make the PIC18F87K22 family a logical choice for many high-performance applications where price is a primary consideration.

### 1.1 Core Features

#### 1.1.1 nanoWatt TECHNOLOGY

All of the devices in the PIC18F87K22 family incorporate a range of features that can significantly reduce power consumption during operation. Key items include:

- Alternate Run Modes: By clocking the controller from the Timer1 source or the Internal RC oscillator, power consumption during code execution can be reduced.
- **Multiple Idle Modes:** The controller can also run with its CPU core disabled but the peripherals still active. In these states, power consumption can be reduced even further.
- **On-the-Fly Mode Switching:** The power-managed modes are invoked by user code during operation, allowing the user to incorporate power-saving ideas into their application's software design.
- nanoWatt XLP: An extra low-power Sleep, BOR, RTCC and Watchdog Timer

## 1.1.2 OSCILLATOR OPTIONS AND FEATURES

All of the devices in the PIC18F87K22 family offer different oscillator options, allowing users a range of choices in developing application hardware. These include:

- External Resistor/Capacitor (RC); RA6 available
- External Resistor/Capacitor with Clock Out (RCIO)
- Three External Clock modes:
  - External Clock (EC); RA6 available
  - External Clock with Clock Out (ECIO)
- External Crystal (XT, HS, LP)A Phase Lock Loop (PLL) frequency multiplier,
- available to the External Oscillator modes, which allows clock speeds of up to 64 MHz. PLL can also be used with the internal oscillator.

- An internal oscillator block that provides a 16 MHz clock (±2% accuracy) and an INTRC source (approximately 31 kHz, stable over temperature and VDD)
  - Operates as HF-INTOSC or MF-INTOSC when block selected for 16 MHz or 500 kHz
  - Frees the two oscillator pins for use as additional general purpose I/O

The internal oscillator block provides a stable reference source that gives the family additional features for robust operation:

- Fail-Safe Clock Monitor: This option constantly monitors the main clock source against a reference signal provided by the internal oscillator. If a clock failure occurs, the controller is switched to the internal oscillator, allowing for continued low-speed operation or a safe application shutdown.
- **Two-Speed Start-up:** This option allows the internal oscillator to serve as the clock source from Power-on Reset, or wake-up from Sleep mode, until the primary clock source is available.

#### 1.1.3 MEMORY OPTIONS

The PIC18F87K22 family provides ample room for application code, from 32 Kbytes to 128 Kbytes of code space. The Flash cells for program memory are rated to last up to 10,000 erase/write cycles. Data retention without refresh is conservatively estimated to be greater than 40 years.

The Flash program memory is readable and writable. During normal operation, the PIC18F87K22 family also provides plenty of room for dynamic application data with up to 3,862 bytes of data RAM.

#### 1.1.4 EXTERNAL MEMORY BUS

Should 128 Kbytes of memory be inadequate for an application, the 80-pin members of the PIC18F87K22 family have an External Memory Bus (EMB) enabling the controller's internal Program Counter to address a memory space of up to 2 Mbytes. This is a level of data access that few 8-bit devices can claim and enables:

- Using combinations of on-chip and external memory of up to 2 Mbytes
- Using external Flash memory for reprogrammable application code or large data tables
- Using external RAM devices for storing large amounts of variable data

#### 1.1.5 EXTENDED INSTRUCTION SET

The PIC18F87K22 family implements the optional extension to the PIC18 instruction set, adding eight new instructions and an Indexed Addressing mode. Enabled as a device configuration option, the extension has been specifically designed to optimize re-entrant application code originally developed in high-level languages, such as 'C'.

### 1.1.6 EASY MIGRATION

All devices share the same rich set of peripherals except that the devices with 32 Kbytes of program memory (PIC18F65K22 and PIC18F85K22) have two less CCPs and three less timers. This provides a smooth migration path within the device family as applications evolve and grow.

The consistent pinout scheme, used throughout the entire family, also aids in migrating to the next larger device. This is true when moving between the 64-pin members, between the 80-pin members, or even jumping from 64-pin to 80-pin devices.

All of the devices in the family share the same rich set of peripherals, except for those with 32 Kbytes of program memory (PIC18F65K22 and PIC18F85K22). Those devices have two less CCPs and three less timers.

The PIC18F87K22 family is also largely pin compatible with other PIC18 families, such as the PIC18F8720 and PIC18F8722 and the PIC18F85J11. This allows a new dimension to the evolution of applications, allowing developers to select different price points within Microchip's PIC18 portfolio, while maintaining a similar feature set.

#### 1.2 Other Special Features

- Communications: The PIC18F87K22 family incorporates a range of serial communication peripherals, including two Enhanced USARTs (EUSART) that support LIN/J2602, and two Master SSP modules, capable of both SPI and I<sup>2</sup>C<sup>™</sup> (Master and Slave) modes of operation.
- CCP Modules: PIC18F87K22 family devices incorporate up to seven Capture/Compare/PWM (CCP) modules. Up to six different time bases can be used to perform several different operations at once.
- ECCP Modules: The PIC18F87K22 family has three Enhanced CCP (ECCP) modules to maximize flexibility in control applications:
  - Up to eight different time bases for performing several different operations at once
  - Up to four PWM outputs for each module, for a total of 12 PWMs
  - Other beneficial features, such as polarity selection, programmable dead time, auto-shutdown and restart, and Half-Bridge and Full-Bridge Output modes

- 12-Bit A/D Converter: The PIC18F87K22 family has differential ADC. It incorporates programmable acquisition time, allowing for a channel to be selected and a conversion to be initiated without waiting for a sampling period, and thus, reducing code overhead.
- Charge Time Measurement Unit (CTMU): The CTMU is a flexible analog module that provides accurate differential time measurement between pulse sources, as well as asynchronous pulse generation.
- Together with other on-chip analog modules, the CTMU can precisely measure time, measure capacitance or relative changes in capacitance, or generate output pulses that are independent of the system clock.
- LP Watchdog Timer (WDT): This enhanced version incorporates a 22-bit prescaler, allowing an extended time-out range that is stable across operating voltage and temperature. See Section 31.0 "Electrical Characteristics" for time-out periods.
- Real-Time Clock and Calendar Module (RTCC): The RTCC module is intended for applications requiring that accurate time be maintained for extended periods of time with minimum to no intervention from the CPU.
- The module is a 100-year clock and calendar with automatic leap year detection. The range of the clock is from 00:00:00 (midnight) on January 1, 2000 to 23:59:59 on December 31, 2099.

#### 1.3 Details on Individual Family Members

Devices in the PIC18F87K22 family are available in 64-pin and 80-pin packages. Block diagrams for the two groups are shown in Figure 1-1 and Figure 1-2.

The devices are differentiated from each other in these ways:

- Flash Program Memory:
  - PIC18FX5K22 (PIC18F65K22 and PIC18F85K22) – 32 Kbytes
  - PIC18FX6K22 (PIC18F66K22 and PIC18F86K22) – 64 Kbytes
  - PIC18FX7K22 (PIC18F67K22 and PIC18F87K22) – 128 Kbytes
- Data RAM:
  - All devices except PIC18FX5K22 4 Kbytes
  - PIC18FX5K22 2 Kbytes
- I/O Ports:
  - PIC18F6XK22 (64-pin devices) 7 bidirectional ports
  - PIC18F8XK22 (80-pin devices) 9 bidirectional ports

- CCP modules:
  - PIC18FX5K22 (PIC18F65K22 and PIC18F85K22) – 5 CCP modules
  - PIC18FX6K22 and PIC18FX7K22 (PIC18F66K22, PIC18F86K22, PIC18F67K22, and PIC18F87K22) – 7 CCP modules
- · Timer modules:
  - PIC18FX5K22 (PIC18F65K22 and PIC18F85K22) – Four 8-bit timer/counters and four 16-bit timer/counters
  - PIC18FX6K22 and PIC18FX7K22 (PIC18F66K22, PIC18F86K22, PIC18F67K22, and PIC18F87K22) – Six 8-bit timer/counters and five 16-bit timer/counters
- A/D Channels:
  - PIC18F6XK22 (64-pin devices) 24 channels
  - PIC18F8XK22 (80-pin devices) 16 channels

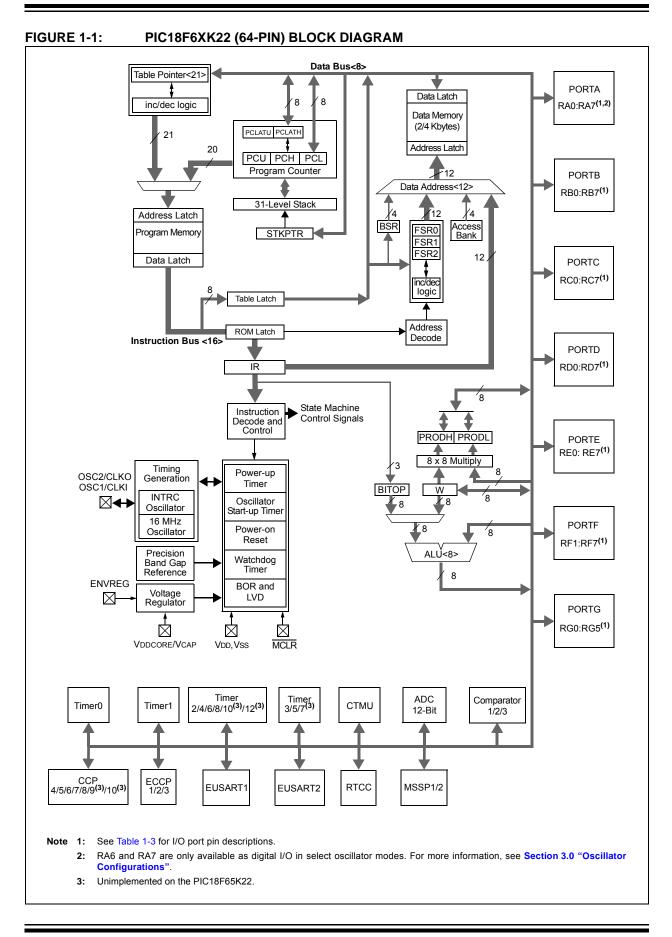
All other features for devices in this family are identical. These are summarized in Table 1-1 and Table 1-2.

The pinouts for all devices are listed in Table 1-3 and Table 1-4.

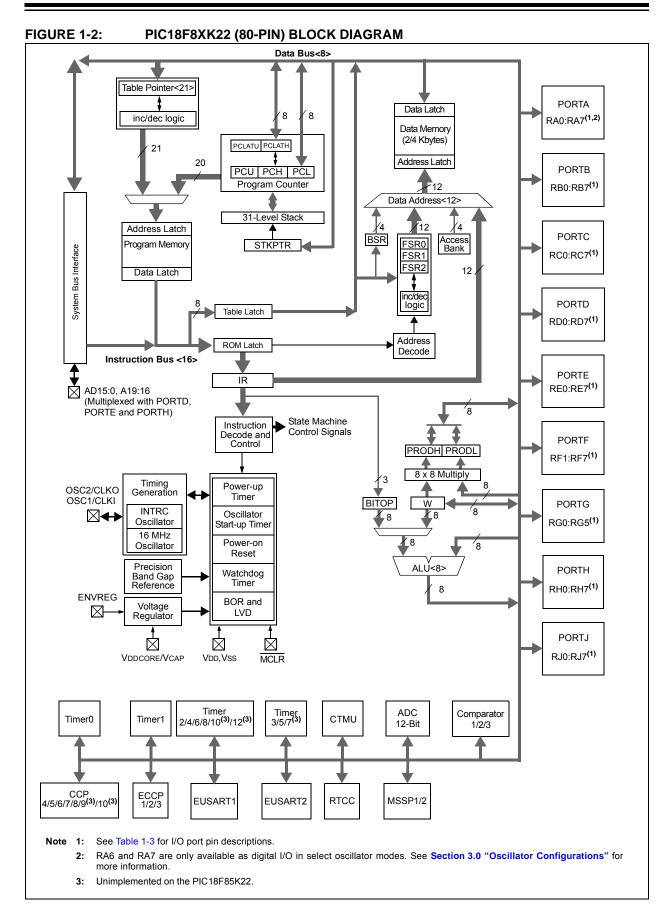
Features	PIC18F65K22	PIC18F66K22	PIC18F67K22				
Operating Frequency	DC – 64 MHz						
Program Memory (Bytes)	32K	64K	128K				
Program Memory (Instructions)	16,384	32,768	65,536				
Data Memory (Bytes)	2K	4K	4K				
Interrupt Sources	42	4	8				
I/O Ports		Ports A, B, C, D, E, F, G					
Parallel Communications	Parallel Slave Port (PSP)						
Timers	8 11						
Comparators	3						
СТМИ	Yes						
RTCC	Yes						
Capture/Compare/PWM (CCP) Modules	5	7	7				
Enhanced CCP (ECCP) Modules	3						
Serial Communications	Two MSSPs and two Enhanced USARTs (EUSART)						
12-Bit Analog-to-Digital Module	16 Input Channels						
Resets (and Delays)	POR, BOR, RESET Instruction, Stack Full, Stack Underflow, MCLR, WI (PWRT, OST)						
Instruction Set	75 Instructions, 83 with Extended Instruction Set Enabled						
Packages	64-Pin QFN, 64-Pin TQFP						

### TABLE 1-2: DEVICE FEATURES FOR THE PIC18F8XK22 (80-PIN DEVICES)

Features	PIC18F85K22	PIC18F86K22	PIC18F87K22				
Operating Frequency	DC – 64 MHz						
Brogram Momony (Buton)	32K	64K	128K				
Program Memory (Bytes)	(Up to 2 Mbytes with Extended Memory)						
Program Memory (Instructions)	16,384	32,768	65,536				
Data Memory (Bytes)	2K	4K	4K				
Interrupt Sources	42	4	48				
I/O Ports	Ports A, B, C, D, E, F, G, H, J						
Parallel Communications	Parallel Slave Port (PSP)						
Timers	8 11						
Comparators	3						
СТМИ	Yes						
RTCC	Yes						
Capture/Compare/PWM (CCP) Modules	5	7	7				
Enhanced CCP (ECCP) Modules	3						
Serial Communications	Two MSSPs and 2 Enhanced USARTs (EUSART)						
12-Bit Analog-to-Digital Module	24 Input Channels						
Resets (and Delays)	POR, BOR, RESET Instruction, Stack Full, Stack Underflow, MCLR, WDT (PWRT, OST)						
Instruction Set	75 Instructions, 83 with Extended Instruction Set Enabled						
Packages	80-Pin TQFP						



## PIC18F87K22 FAMILY



Din Nome	Pin Number	Pin	Buffer	B Market				
Pin Name	QFN/TQFP		Туре	Description				
	7			Master Clear (input) or programming voltage (input).				
MCLR/RG5								
MCLR		1	ST	This pin is an active-low Reset to the device.				
RG5		1	ST	General purpose, input only pin.				
OSC1/CLKI/RA7	39			Oscillator crystal or external clock input.				
OSC1		1	CMOS	Oscillator crystal input.				
CLKI		1	CMOS	External clock source input. Always associated				
				with pin function, OSC1. (See related OSC1/CLKI,				
				OSC2/CLKO pins.)				
RA7		I/O	TTL	General purpose I/O pin.				
OSC2/CLKO/RA6	40			Oscillator crystal or clock output.				
OSC2		0	—	Oscillator crystal output. Connects to crystal or				
				resonator in Crystal Oscillator mode.				
CLKO		0	—	In certain oscillator modes, OSC2 pin outputs CLKO,				
				which has 1/4 the frequency of OSC1 and denotes the				
				instruction cycle rate.				
RA6		I/O	TTL	General purpose I/O pin.				
Legend: TTL = TTL con	npatible input			CMOS = CMOS compatible input or output				
ST = Schmitt	Trigger input w	ith CN	10S levels	Analog = Analog input				

#### **TABLE 1-3**: PIC18F6XK22 PINOUT I/O DESCRIPTIONS

1 = Input

Ρ = Power

 $I^2C = I^2C^{TM}/SMBus$ 

**Note 1:** Default assignment for ECCP2 when the CCP2MX Configuration bit is set.

2: Alternate assignment for ECCP2 when the CCP2MX Configuration bit is cleared.

3: Not available on PIC18F65K22 and PIC18F85K22 devices.

4: The CC6, CCP7, CCP8 and CCP9 pin placement depends on the setting of the ECCPMX Configuration bit (CONFIG3H<1>).

0

OD

= Output

Pin Name	Pin Number	Pin	Buffer	Description			
	QFN/TQFP	Туре	Туре	Description			
				PORTA is a bidirectional I/O port.			
RA0/AN0/ULPWU RA0 AN0 ULPWU	24	I/O I I	TTL Analog Analog	Digital I/O. Analog Input 0. Ultra Low-Power Wake-up input.			
RA1/AN1 RA1 AN1	23	I/O I	TTL Analog	Digital I/O. Analog Input 1.			
RA2/AN2/VREF- RA2 AN2 VREF-	22	I/O I I	TTL Analog Analog	Digital I/O. Analog Input 2. A/D reference voltage (low) input.			
RA3/AN3/VREF+ RA3 AN3 VREF+	21	I/O I I	TTL Analog Analog	Digital I/O. Analog Input 3. A/D reference voltage (high) input.			
RA4/T0CKI RA4 T0CKI	28	I/O I	ST ST	Digital I/O. Timer0 external clock input.			
RA5/AN4/T1CKI/T3G/ HLVDIN RA5 AN4 T1CKI T3G HLVDIN	27	I/O I I I	TTL Analog ST ST Analog	Digital I/O. Analog Input 4. Timer1 clock input. Timer3 external clock gate input. High/Low-Voltage Detect input.			
RA6				See the OSC2/CLKO/RA6 pin.			
RA7				See the OSC1/CLKI/RA7 pin.			
Legend: TTL = TTL con ST = Schmitt I = Input P = Power $I^2C = I^2C^{TM}/SI$	Trigger input w	vith CM	1OS levels	CMOS = CMOS compatible input or output Analog = Analog input O = Output OD = Open-Drain (no P diode to VDD)			

**Note 1:** Default assignment for ECCP2 when the CCP2MX Configuration bit is set.

2: Alternate assignment for ECCP2 when the CCP2MX Configuration bit is cleared.

3: Not available on PIC18F65K22 and PIC18F85K22 devices.

Din Nama	Pin Number	Pin Type	Buffer Type	Description
Pin Name	QFN/TQFP			Description
				PORTB is a bidirectional I/O port. PORTB can be software programmed for internal weak pull-ups on all inputs.
RB0/INT0/FLTO RB0 INT0 FLT0	48	I/O I I	TTL ST ST	Digital I/O. External Interrupt 0. Enhanced PWM Fault input for ECCP1/2/3.
RB1/INT1 RB1 INT1	47	I/O I	TTL ST	Digital I/O. External Interrupt 1.
RB2/INT2/CTED1 RB2 INT2 CTED1	46	I/O I I	TTL ST ST	Digital I/O. External Interrupt 2. CTMU Edge 1 input.
RB3/INT3/CTED2/ ECCP2/P2A RB3 INT3 CTED2 ECCP2 P2A	45	I/O I I I/O O	TTL ST ST ST	Digital I/O. External Interrupt 3. CTMU Edge 2 input. Capture 2 input/Compare 2 output/PWM2. Enhanced PWM2 Output A.
RB4/KBI0 RB4 KBI0	44	I/O I	TTL TTL	Digital I/O. Interrupt-on-change pin.
RB5/KBI1/T3CKI/T1G RB5 KBI1 T3CKI T1G	43	I/O I I	TTL TTL ST ST	Digital I/O. Interrupt-on-change pin. Timer3 clock input. Timer1 external clock gate input.
RB6/KBI2/PGC RB6 KBI2 PGC	42	I/O I I/O	TTL TTL ST	Digital I/O. Interrupt-on-change pin. In-Circuit Debugger and ICSP™ programming clock pin.
RB7/KBI3/PGD RB7 KBI3 PGD	37	I/O I I/O	TTL TTL ST	Digital I/O. Interrupt-on-change pin. In-Circuit Debugger and ICSP programming data pin.
Legend: TTL = TTL co ST = Schmit I = Input P = Power $I^2C = I^2C M/S$	t Trigger input w	vith CM	10S levels	CMOS = CMOS compatible input or output Analog = Analog input O = Output OD = Open-Drain (no P diode to VDD)

I<sup>2</sup>C = I<sup>2</sup>C™/SMBus

**Note 1:** Default assignment for ECCP2 when the CCP2MX Configuration bit is set.

**2:** Alternate assignment for ECCP2 when the CCP2MX Configuration bit is cleared.

**3:** Not available on PIC18F65K22 and PIC18F85K22 devices.

Din Nome	Pin Number	Pin	Buffer	Description				
Pin Name	QFN/TQFP	Туре	Туре	Description				
				PORTC is a bidirectional I/O port.				
RC0/SOSCO/SCLKI RC0 SOSCO SCLKI	30	I/O O I	ST — ST	Digital I/O. SOSC oscillator output. Digital SOSC input.				
RC1/SOSCI/ECCP2/P2A RC1 SOSCI ECCP2 <sup>(1)</sup> P2A	29	I/O I I/O O	ST CMOS ST	Digital I/O. SOSC oscillator input. Capture 2 input/Compare 2 output/PWM2 output. Enhanced PWM2 Output A.				
RC2/ECCP1/P1A RC2 ECCP1 P1A	33	I/O I/O O	ST ST	Digital I/O. Capture 1 input/Compare 1 output/PWM1 output. Enhanced PWM1 Output A.				
RC3/SCK1/SCL1 RC3 SCK1 SCL1 <sup>(4)</sup>	34	I/O I/O I/O	ST ST I <sup>2</sup> C	Digital I/O. Synchronous serial clock input/output for SPI mode. Synchronous serial clock input/output for I <sup>2</sup> C™ mode.				
RC4/SDI1/SDA1 RC4 SDI1 SDA1 <sup>(4)</sup>	35	I/O I I/O	ST ST I <sup>2</sup> C	Digital I/O. SPI data in. I <sup>2</sup> C data I/O.				
RC5/SDO1 RC5 SDO1	36	I/O O	ST —	Digital I/O. SPI data out.				
RC6/TX1/CK1 RC6 TX1 CK1	31	I/O O I/O	ST — ST	Digital I/O. EUSART asynchronous transmit. EUSART synchronous clock (see related RX1/DT1).				
RC7/RX1/DT1 RC7 RX1 DT1	32	I/O I I/O	ST ST ST	Digital I/O. EUSART asynchronous receive. EUSART synchronous data (see related TX1/CK1).				
Legend:       TTL = TTL compatible input       CMOS       = CMOS compatible input or output         ST = Schmitt Trigger input with CMOS levels       Analog       = Analog input         I = Input       O       = Output								

P = Power

 $I^2C = I^2C^{TM}/SMBus$ 

Note 1: Default assignment for ECCP2 when the CCP2MX Configuration bit is set.

2: Alternate assignment for ECCP2 when the CCP2MX Configuration bit is cleared.

3: Not available on PIC18F65K22 and PIC18F85K22 devices.

**4:** The CC6, CCP7, CCP8 and CCP9 pin placement depends on the setting of the ECCPMX Configuration bit (CONFIG3H<1>).

OD

Dia Nama	Pin Number	Pin	Buffer Type				
Pin Name	QFN/TQFP	Туре		Description			
				PORTD is a bidirectional I/O port.			
RD0/PSP0/CTPLS RD0 PSP0 CTPLS	58	I/O I/O O	ST TTL	Digital I/O. Parallel Slave Port data. CTMU pulse generator output.			
RD1/PSP1/T5CKI/T7G RD1 PSP1 T5CKI T7G	55	I/O I/O I	ST TTL ST ST	Digital I/O. Parallel Slave Port. Timer5 clock input. Timer7 external clock gate input.			
RD2/PSP2 RD2 PSP2	54	I/O O	ST TTL	Digital I/O. Parallel Slave Port.			
RD3/PSP3 RD3 PSP3	53	I/O I/O	ST TTL	Digital I/O. Parallel Slave Port.			
RD4/PSP4/SDO2 RD4 PSP4 SDO2	52	I/O I/O O	ST TTL	Digital I/O. Parallel Slave Port. SPI data out.			
RD5/PSP5/SDI2/SDA2 RD5 PSP5 SDI2 SDA2	51	I/O I/O I I/O	ST TTL ST I <sup>2</sup> C	Digital I/O. Parallel Slave Port. SPI data in. I <sup>2</sup> C™ data I/O.			
RD6/PSP6/SCK2/SCL2 RD6 PSP6 SCK2 SCL2 <sup>(4)</sup>	50	I/O I/O I/O I/O	ST TTL ST I <sup>2</sup> C	Digital I/O. Parallel Slave Port. Synchronous serial clock. Synchronous serial clock I/O for I <sup>2</sup> C mode.			
RD7/PSP7/ <mark>SS2</mark> RD7 <u>PSP</u> 7 SS2	49	I/O I/O I	ST TTL TTL	Digital I/O. Parallel Slave Port. SPI slave select input.			
Legend:TTL = TTL compatible inputCMOS= CMOS compatible input or outputST = Schmitt Trigger input with CMOS levelsAnalog= Analog inputI = InputO= OutputP = PowerOD= Open-Drain (no P diode to VDD)							

TABLE 1-3: PIC18F6XK22 PINOUT I/O DESCRIPTIONS (CONTINUE
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 $I^2C = I^2C^{TM}/SMBus$ 

Note 1: Default assignment for ECCP2 when the CCP2MX Configuration bit is set.

**2:** Alternate assignment for ECCP2 when the CCP2MX Configuration bit is cleared.

3: Not available on PIC18F65K22 and PIC18F85K22 devices.

Din Nama	Pin Number	Pin	Buffer Type	Description	
Pin Name	QFN/TQFP	Туре		Description	
				PORTE is a bidirectional I/O port.	
RE0/RD/P2D RE0 RD P2D	2	I/O I O	ST TTL	Digital I/O. Parallel Slave Port read strobe. EECP2 PWM Output D.	
RE1/WR/P2C RE1 WR P2C	1	I/O I O	ST TTL —	Digital I/O. Parallel Slave Port write strobe. ECCP2 PWM Output C.	
RE2/CS/P2B/CCP10 RE2 CS P2B CCP10 <sup>(3)</sup>	64	I/O I 0 I/O	ST TTL — S/T	Digital I/O. Parallel Slave Port chip select. ECCP2 PWM Output B. Capture 10 input/Compare 10 output/PWM10 output.	
RE3/P3C/CCP9/REFO RE3 P3C CCP9 <sup>(3,4)</sup> REFO	63	I/O O I/O O	ST — S/T —	Digital I/O. ECCP3 PWM Output C. Capture 9 input/Compare 9 output/PWM9 output. Reference clock out.	
RE4/P3B/CCP8 RE4 P3B CCP8 <sup>(4)</sup>	62	I/O O I/O	ST — S/T	Digital I/O. ECCP3 PWM Output B. Capture 8 input/Compare 8 output/PWM8 output.	
RE5/P1C/CCP7 RE5 P1C CCP7 <sup>(4)</sup>	61	I/O O I/O	ST — S/T	Digital I/O. ECCP1 PWM Output C. Capture 7 input/Compare 7 output/PWM7 output.	
RE6/P1B/CCP6 RE6 P1B CCP6 <sup>(4</sup>	60	I/O O I/O	ST — S/T	Digital I/O. ECCP1 PWM Output B. Capture 6 input/Compare 6 output/PWM6 output.	
RE7/ECCP2/P2A RE7 ECCP2 <sup>(2)</sup> P2A	59	I/O I/O O	ST ST	Digital I/O. Capture 2 input/Compare 2 output/PWM2 output. ECCP2 PWM Output A.	
Legend: TTL = TTL con ST = Schmitt I = Input P = Power $I^2C = I^2C^{TM}/SI$	Trigger input w	∕ith C№	1OS levels	CMOS= CMOS compatible input or outputAnalog= Analog inputO= OutputOD= Open-Drain (no P diode to VDD)	

Note 1: Default assignment for ECCP2 when the CCP2MX Configuration bit is set.

**2:** Alternate assignment for ECCP2 when the CCP2MX Configuration bit is cleared.

3: Not available on PIC18F65K22 and PIC18F85K22 devices.

Din Nama	Pin Number	Pin Buffer	Buffer	Description
Pin Name	QFN/TQFP	Туре	ре Туре	Description
				PORTF is a bidirectional I/O port.
RF1/AN6/C2OUT/CTDIN RF1 AN6 C2OUT CTDIN	17	I/O   0 	ST Analog — ST	Digital I/O. Analog Input 6. Comparator 2 output. CTMU pulse delay input.
RF2/AN7/C1OUT RF2 AN7 C1OUT	16	I/O I O	ST Analog —	Digital I/O. Analog Input 7. Comparator 1 output.
RF3/AN8/C2INB/CTMUI RF3 AN8 C2INB CTMUI	15	I/O     0	ST Analog Analog —	Digital I/O. Analog Input 8. Comparator 2 Input B. CTMU pulse generator charger for the C2INB comparator input.
RF4/AN9/C2INA RF4 AN9 C2INA	14	I/O I I	ST Analog Analog	Digital I/O. Analog Input 9. Comparator 2 Input A.
RF5/AN10/CVREF/C1INB RF5 AN10 CVREF C1INB	13	I/O   0 	ST Analog Analog Analog	Digital I/O. Analog Input 10. Comparator reference voltage output. Comparator 1 Input B.
RF6/AN11/C1INA RF6 AN11 C1INA	12	I/O I I	ST Analog Analog	Digital I/O. Analog Input 11. Comparator 1 Input A.
RF7/AN5/ <u>SS1</u> RF7 <u>AN5</u> SS1	11	I/O O I	ST Analog TTL	Digital I/O. Analog Input 5. SPI1 slave select input.
I = Input P = Power I <sup>2</sup> C = I <sup>2</sup> C™/SI	Trigger input w MBus			CMOS = CMOS compatible input or output Analog = Analog input O = Output OD = Open-Drain (no P diode to VDD)

Note 1: Default assignment for ECCP2 when the CCP2MX Configuration bit is set.

2: Alternate assignment for ECCP2 when the CCP2MX Configuration bit is cleared.

3: Not available on PIC18F65K22 and PIC18F85K22 devices.

Din Nama	Pin Number	Pin	Buffer	Description
Pin Name	QFN/TQFP	Туре	Туре	
				PORTG is a bidirectional I/O port.
RG0/ECCP3/P3A RG0 ECCP3 P3A	3	I/O I/O O	ST ST	Digital I/O. Capture 3 input/Compare 3 output/PWM3 output. ECCP3 PWM Output A.
RG1/TX2/CK2/AN19/ C3OUT RG1 TX2 CK2 AN19 C3OUT	4	I∕O I∕O I∕O I O	ST — ST Analog —	Digital I/O. EUSART asynchronous transmit. EUSART synchronous clock (see related RX2/DT2). Analog Input 19. Comparator 3 output.
RG2/RX2/DT2/AN18/ C3INA RG2 RX2 DT2 AN18 C3INA	5	1/0    /0 	ST ST ST Analog Analog	Digital I/O. EUSART asynchronous receive. EUSART synchronous data (see related TX2/CK2). Analog Input 18. Comparator 3 Input A.
RG3/CCP4/AN17/P3D/ C3INB RG3 CCP4 AN17 P3D C3INB	6	I/O I/O I O I	ST S/T Analog — Analog	Digital I/O. Capture 4 input/Compare 4 output/PWM4 output. Analog Input 18. ECCP3 PWM Output D. Comparator 3 Input B.
RG4/RTCC/T7CKI/T5G/ CCP5/AN16/P1D/C3INC RG4 RTCC T7CKI <sup>(3)</sup> T5G CCP5 AN16 P1D C3INC	8	∕0 0 − 	ST ST ST ST Analog Analog	Digital I/O. RTCC output Timer7 clock input. Timer5 external clock gate input. Capture 5 input/Compare 5 output/PWM5 output. Analog Input 16. ECCP1 PWM Output D. Comparator 3 Input C.
RG5	7			See the MCLR/RG5 pin.
Legend: TTL = TTL con ST = Schmitt I = Input P = Power $\mu^{2}c = \mu^{2}cTM(c)$	Trigger input w	vith C№	IOS levels	CMOS= CMOS compatible input or outputAnalog= Analog inputO= OutputOD= Open-Drain (no P diode to VDD)

P = Power $I^2C = I^2C^{TM}/SMBus$ 

**Note 1:** Default assignment for ECCP2 when the CCP2MX Configuration bit is set.

2: Alternate assignment for ECCP2 when the CCP2MX Configuration bit is cleared.

**3:** Not available on PIC18F65K22 and PIC18F85K22 devices.

Pin Name	Pin Number	Pin	Buffer	Description
	QFN/TQFP	Туре	Туре	
Vss	9, 25, 41, 56	Р		Ground reference for logic and I/O pins.
Vdd	26, 38, 57	Р	_	Positive supply for logic and I/O pins.
AVss	20	Р	_	Ground reference for analog modules.
AVdd	19	Р	_	Positive supply for analog modules.
ENVREG	18	Ι	ST	Enable for on-chip voltage regulator.
Vddcore/Vcap Vddcore Vcap	10	Ρ		Core logic power or external filter capacitor connection. External filter capacitor connection (regulator enabled/disabled).
Legend: TTL = TTL com ST = Schmitt	npatible input Trigger input w	vith CN	10S levels	CMOS = CMOS compatible input or output Analog = Analog input

= Input

= Power Ρ

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 $I^2C = I^2C^{TM}/SMBus$ 

Note 1: Default assignment for ECCP2 when the CCP2MX Configuration bit is set.

2: Alternate assignment for ECCP2 when the CCP2MX Configuration bit is cleared.

3: Not available on PIC18F65K22 and PIC18F85K22 devices.

4: The CC6, CCP7, CCP8 and CCP9 pin placement depends on the setting of the ECCPMX Configuration bit (CONFIG3H<1>).

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OD

= Output

#### **TABLE 1-4:** PIC18F8XK22 PINOUT I/O DESCRIPTIONS

Dia Nama	Pin Number	Pin Buffer	Description		
Pin Name	TQFP	Туре	Туре	Description	
	9			Master Clear (input) or programming voltage (input).	
MCLR/RG5					
RG5		1	ST	This pin is an active-low Reset to the device.	
MCLR		I	ST	General purpose, input only pin.	
OSC1/CLKI/RA7	49			Oscillator crystal or external clock input.	
OSC1		1	CMOS	Oscillator crystal input.	
CLKI		1	CMOS	External clock source input. Always associated	
				with pin function, OSC1. (See related OSC1/CLKI,	
				OSC2/CLKO pins.)	
RA7		I/O	TTL	General purpose I/O pin.	
OSC2/CLKO/RA6	50			Oscillator crystal or clock output.	
OSC2		0		Oscillator crystal output. Connects to crystal or	
				resonator in Crystal Oscillator mode.	
CLKO		0	—	In certain oscillator modes, OSC2 pin outputs CLKO,	
				which has 1/4 the frequency of OSC1 and denotes the	
				instruction cycle rate.	
RA6		I/O	TTL	General purpose I/O pin.	
Legend: TTL = TTL comp	patible input			CMOS = CMOS compatible input or output	
ST = Schmitt Trigger input with CMOS levels Analog = Analog input					

- = Input 1
- Ρ = Power
- $I^2C = I^2C^{\text{TM}}/\text{SMBus}$

Note 1: Default assignment for ECCP2 when the CCP2MX Configuration bit is set.

- 2: Alternate assignment for ECCP2 when the CCP2MX Configuration bit is cleared.
- 3: Not available on PIC18F65K22 and PIC18F85K22 devices.
- 4: PSP is available only in Microcontroller mode.
- 5: The CC6, CCP7, CCP8 and CCP9 pin placement depends on the setting of the ECCPMX Configuration bit (CONFIG3H<1>).

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OD

= Output

Dia Maraa	Pin Number	Pin	Buffer	Description
Pin Name	TQFP	Туре	Туре	
				PORTA is a bidirectional I/O port.
RA0/AN0/ULPWU RA0 AN0 ULPWU	30	I/O I I	TTL Analog Analog	Digital I/O. Analog Input 0. Ultra Low-Power Wake-up input.
RA1/AN1 RA1 AN1	29	I/O I	TTL Analog	Digital I/O. Analog Input 1.
RA2/AN2/VREF- RA2 AN2 VREF-	28	I/O I I	TTL Analog Analog	Digital I/O. Analog Input 2. A/D reference voltage (low) input.
RA3/AN3/VREF+ RA3 AN3 VREF+	27	I/O I I	TTL Analog Analog	Digital I/O. Analog Input 3. A/D reference voltage (high) input.
RA4/T0CKI RA4 T0CKI	34	I/O I	ST ST	Digital I/O. Timer0 external clock input.
RA5/AN4/T1CKI/ T3G/HLVDIN RA5 AN4 T1CKI T3G HLVDIN	33	I/O     	TTL Analog ST ST Analog	Digital I/O. Analog Input 4. Timer1 clock input. Timer3 external clock gate input. High/Low-Voltage Detect input.
RA6				See the OSC2/CLKO/RA6 pin.
RA7				See the OSC1/CLKI/RA7 pin.
Legend: TTL = TTL comp ST = Schmitt T I = Input P = Power $I^2C = I^2C^{TM}/SM$	rigger input wit	h CMC	S levels	CMOS= CMOS compatible input or outputAnalog= Analog inputO= OutputOD= Open-Drain (no P diode to VDD)

**Note 1:** Default assignment for ECCP2 when the CCP2MX Configuration bit is set.

2: Alternate assignment for ECCP2 when the CCP2MX Configuration bit is cleared.

3: Not available on PIC18F65K22 and PIC18F85K22 devices.

4: PSP is available only in Microcontroller mode.

Pin Name	Pin Number	Pin	Buffer	Description
Fill Name	TQFP	Туре	Туре	Description
				PORTB is a bidirectional I/O port. PORTB can be software programmed for internal weak pull-ups on all inputs.
RB0/INT0/FLT0 RB0 INT0 FLT0	58	I/O I I	TTL ST ST	Digital I/O. External Interrupt 0. Enhanced PWM Fault input for ECCP1/2/3.
RB1/INT1 RB1 INT1	57	I/O I	TTL ST	Digital I/O. External Interrupt 1.
RB2/INT2/CTED1 RB2 INT2 CTED1	56	I/O   	TTL ST ST	Digital I/O. External Interrupt 2. CTMU Edge 1 input.
RB3/INT3/CTED2/ ECCP2/P2A RB3 INT3 CTED2 ECCP2 P2A	55	1/0    /0 0	TTL ST ST ST ST	Digital I/O. External Interrupt 3. CTMU Edge 2 input. Capture 2 input/Compare 2 output/PWM2 output. Enhanced PWM2 Output A.
RB4/KBI0 RB4 KBI0	54	I/O I	TTL TTL	Digital I/O. Interrupt-on-change pin.
RB5/KBI1/T3CKI/T1G RB5 KBI1 T3CKI T1G	53	I/O     	TTL TTL ST ST	Digital I/O. Interrupt-on-change pin. Timer3 clock input. Timer1 external clock gate input.
RB6/KBI2/PGC RB6 KBI2 PGC	52	I/O I I/O	TTL TTL ST	Digital I/O. Interrupt-on-change pin. In-Circuit Debugger and ICSP™ programming clock pin.
RB7/KBI3/PGD RB7 KBI3 PGD	47	I/O I I/O	TTL TTL ST	Digital I/O. Interrupt-on-change pin. In-Circuit Debugger and ICSP programming data pin.
Legend: TTL = TTL com ST = Schmitt I = Input P = Power	Trigger input wit	h CMC	)S levels	CMOS = CMOS compatible input or output Analog = Analog input O = Output OD = Open-Drain (no P diode to VDD)

I<sup>2</sup>C = I<sup>2</sup>C™/SMBus

Note 1: Default assignment for ECCP2 when the CCP2MX Configuration bit is set.

2: Alternate assignment for ECCP2 when the CCP2MX Configuration bit is cleared.

3: Not available on PIC18F65K22 and PIC18F85K22 devices.

4: PSP is available only in Microcontroller mode.

Din Nama	Pin Number	Pin	Buffer	Description
Pin Name	TQFP	Туре	Туре	Description
				PORTC is a bidirectional I/O port.
RC0/SOSCO/SCKLI RC0 SOSCO SCKLI	36	I/O O I	ST — ST	Digital I/O. SOSC oscillator output. Digital SOSC input.
RC1/SOSCI/ECCP2/P2A RC1 SOSCI ECCP2 <sup>(1)</sup> P2A	35	I/O I I/O O	ST CMOS ST —	Digital I/O. SOSC oscillator input. Capture 2 input/Compare 2 output/PWM2 output. Enhanced PWM2 Output A.
RC2/ECCP1/P1A RC2 ECCP1 P1A	43	I/O I/O O	ST ST	Digital I/O. Capture 1 input/Compare 1 output/PWM1 output. Enhanced PWM1 Output A.
RC3/SCK1/SCL1 RC3 SCK1 SCL1	44	I/O I/O I/O	ST ST I <sup>2</sup> C	Digital I/O. Synchronous serial clock input/output for SPI mode. Synchronous serial clock input/output for I <sup>2</sup> C™ mode.
RC4/SDI1/SDA1 RC4 SDI1 SDA1	45	I/O I I/O	ST ST I <sup>2</sup> C	Digital I/O. SPI data in. I <sup>2</sup> C data I/O.
RC5/SDO1 RC5 SDO1	46	I/O O	ST —	Digital I/O. SPI data out.
RC6/TX1/CK1 RC6 TX1 CK1	37	I/O O I/O	ST — ST	Digital I/O. EUSART asynchronous transmit. EUSART synchronous clock (see related RX1/DT1).
RC7/RX1/DT1 RC7 RX1 DT1	38	I/O I I/O	ST ST ST	Digital I/O. EUSART asynchronous receive. EUSART synchronous data (see related TX1/CK1).
Legend: TTL = TTL com ST = Schmitt 1		h CMC	)S levels	CMOS = CMOS compatible input or output Analog = Analog input

= Input Ρ = Power

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 $I^2C = I^2C^{TM}/SMBus$ 

= Output = Open-Drain (no P diode to VDD) OD

Note 1: Default assignment for ECCP2 when the CCP2MX Configuration bit is set.

2: Alternate assignment for ECCP2 when the CCP2MX Configuration bit is cleared.

3: Not available on PIC18F65K22 and PIC18F85K22 devices.

4: PSP is available only in Microcontroller mode.

5: The CC6, CCP7, CCP8 and CCP9 pin placement depends on the setting of the ECCPMX Configuration bit (CONFIG3H<1>).

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Pin Name	Pin Number	Pin	Buffer	Description
Fin Name	TQFP	Туре	Туре	Description
				PORTD is a bidirectional I/O port.
RD0/PSP0/CTPLS/AD0 RD0 PSP0 <sup>(4)</sup> CTPLS AD0	72	I/O I/O O I/O	ST TTL ST TTL	Digital I/O Parallel Slave Port data CTMU pulse generator output External Memory Address/Data 0
RD1/T5CKI/T7G/PSP1/AD1 RD1 T5CKI T7G PSP1 <sup>(4)</sup> AD1	69	I/O      /O  /O	ST ST ST TTL TTL	Digital I/O Timer5 clock input Timer7 external clock gate input Parallel Slave Port data External Memory Address/Data 1
RD2/PSP2/AD2 RD2 PSP2 <sup>(4)</sup> AD2	68	I/O I/O I/O	ST TTL TTL	Digital I/O. Parallel Slave Port data. External Memory Address/Data 2.
RD3/PSP3/AD3 RD3 PSP3 <sup>(4)</sup> AD3	67	I/O I/O I/O	ST TTL TTL	Digital I/O. Parallel Slave Port data. External Memory Address/Data 3.
RD4/SDO2/PSP4/AD4 RD4 SDO2 PSP4 <sup>(4)</sup> AD4	66	I/O O I/O I/O	ST — TTL TTL	Digital I/O. SPI data out. Parallel Slave Port data. External Memory Address/Data 4.
RD5/SDI2/SDA2/PSP5/ AD5	65			
RD5 SDI2 SDA2 PSP5 <sup>(4)</sup> AD5		I/O I I/O I/O I/O	ST ST I <sup>2</sup> C TTL TTL	Digital I/O. SPI data in. I <sup>2</sup> C™ data I/O. Parallel Slave Port data. External Memory Address/Data 5.
Legend: TTL = TTL comp ST = Schmitt T I = Input P = Power $I^2C$ = $I^2C$ TW/CM	rigger input wit	h CMC	)S levels	CMOS= CMOS compatible input or outputAnalog= Analog inputO= OutputOD= Open-Drain (no P diode to VDD)

 $I^2C = I^2C^{TM}/SMBus$ Note 1: Default assignment for ECCP2 when the CCP2MX Configuration bit is set.

2: Alternate assignment for ECCP2 when the CCP2MX Configuration bit is cleared.

3: Not available on PIC18F65K22 and PIC18F85K22 devices.

4: PSP is available only in Microcontroller mode.

= Open-Drain (no P diode to VDD)

Pin Name	Pin Number	Pin Buffer	Description	
Pin Name	TQFP	Туре	Туре	Description
RD6/SCK2/SCL2/PSP6/ AD6	64			
RD6 SCK2		1/O 1/O	ST ST	Digital I/O. Synchronous serial clock input/output for SPI mode.
SCL2 PSP6 <sup>(4)</sup>		1/0 1/0	I <sup>2</sup> C TTL	Synchronous serial clock input/output for I <sup>2</sup> C™ mode. Parallel Slave Port data.
AD6		I/O	TTL	External Memory Address/Data 6.
RD7/SS2/PSP7/AD7	63			
RD7		I/O	ST	Digital I/O.
SS2		I	TTL	SPI slave select input.
PSP7 <sup>(4)</sup>		I/O	TTL	Parallel Slave Port data.
AD7		I/O	TTL	External Memory Address/Data 7.
Legend: TTL = TTL com ST = Schmitt T I = Input	patible input rigger input wit	h CMC	)S levels	CMOS = CMOS compatible input or output Analog = Analog input O = Output

#### TABLE 1-4: PIC18F8XK22 PINOUT I/O DESCRIPTIONS (CONTINUED)

Note 1: Default assignment for ECCP2 when the CCP2MX Configuration bit is set.

2: Alternate assignment for ECCP2 when the CCP2MX Configuration bit is cleared.

3: Not available on PIC18F65K22 and PIC18F85K22 devices.

4: PSP is available only in Microcontroller mode.

= Power

I<sup>2</sup>C = I<sup>2</sup>C<sup>™</sup>/SMBus

Ρ

5: The CC6, CCP7, CCP8 and CCP9 pin placement depends on the setting of the ECCPMX Configuration bit (CONFIG3H<1>).

OD

Pin Name	Pin Number	Pin	Buffer	Description
	TQFP	Туре	Туре	
				PORTE is a bidirectional I/O port.
RE0/P2D/RD/AD8 RE0 P2D RD <sup>(4)</sup> AD8	4	I/O O I I/O	ST — TTL TTL	Digital I/O. ECCP2 PWM Output D. Parallel Slave Port read strobe. External Memory Address/Data 8.
RE1/P2C/WR/AD9 RE1 P2C WR <sup>(4)</sup> AD9	3	I/O O I I/O	ST — TTL TTL	Digital I/O. ECCP2 PWM Output C. Parallel Slave Port write strobe. External Memory Address/Data 9.
RE2/P2B/CCP10/ <del>C</del> S/ AD10 RE2 P2B <u>CCP10<sup>(3)</sup> CS<sup>(4)</sup> AD10</u>	78	1/0 1/0 1/0 1/0	ST ST ST TTL TTL	Digital I/O. ECCP2 PWM Output B. Capture 10 input/Compare 10 output/PWM10 output. Parallel Slave Port chip select. External Memory Address/Data 10.
RE3/P3C/CCP9/REFO AD11 RE3 P3C CCP9 <sup>(3,5)</sup> REFO AD11	77	I/O O I/O I/O	ST — S/T — TTL	Digital I/O. ECCP3 PWM Output C. Capture 9 input/Compare 9 output/PWM9 output. Reference clock out. External Memory Address/Data 11.
RE4/P3B/CCP8/AD12 RE4 P3B CCP8 <sup>(5)</sup> AD12	76	I/O O I/O I/O	ST — ST TTL	Digital I/O. ECCP4 PWM Output B. Capture 8 input/Compare 8 output/PWM8 output. External Memory Address/Data 12.
RE5/P1C/CCP7/AD13 RE5 P1C CCP7 <sup>(5)</sup> AD13	75	I/O O I/O I/O	ST — ST TTL	Digital I/O. ECCP1 PWM Output C. Capture 7 input/Compare 7 output/PWM7 output. External Memory Address/Data 13.
Legend: TTL = TTL com ST = Schmitt I = Input	npatible input Trigger input wit	h CMC	)S levels	CMOS = CMOS compatible input or output Analog = Analog input O = Output

P = Power I<sup>2</sup>C = I<sup>2</sup>C™/SMBus

Note 1: Default assignment for ECCP2 when the CCP2MX Configuration bit is set.

2: Alternate assignment for ECCP2 when the CCP2MX Configuration bit is cleared.

- 3: Not available on PIC18F65K22 and PIC18F85K22 devices.
- 4: PSP is available only in Microcontroller mode.

5: The CC6, CCP7, CCP8 and CCP9 pin placement depends on the setting of the ECCPMX Configuration bit (CONFIG3H<1>).

OD

Pin Number	Pin	Buffer	<b>D</b> esistential
TQFP			Description
74			
	I/O	ST	Digital I/O.
	0		ECCP1 PWM Output B.
	I/O	ST	Capture 6 input/Compare 6 output/PWM6 output.
	I/O	ST	External Memory Address/Data 14.
73			
	I/O	ST	Digital I/O.
	I/O	ST	Capture 2 input/Compare 2 output/PWM2 output.
	0	—	ECCP2 PWM Output A.
	I/O	ST	External Memory Address/Data 15.
patible input			CMOS = CMOS compatible input or output
Frigger input wit	h CMC	OS levels	Analog = Analog input
	TQFP 74 73 patible input	TQFP         Type           74         I/O           73         I/O           1/O         I/O           73         I/O           I/O         I/O           1/O         I/O           73         I/O           I/O         I/O           I/O         I/O           I/O         I/O           I/O         I/O           I/O         I/O           I/O         I/O           Patible input         I/O	TQFP         Fill         Butter           74         Type         Type           74         I/O         ST           0          I/O           I/O         ST         I/O           73         I/O         ST           0          I/O           1/O         ST         ST           0          I/O           1/O         ST         ST           0          I/O           I/O         ST         ST           0          I/O           I/O         ST         ST

P = Power

= Input

1

 $I^2C = I^2C^{TM}/SMBus$ 

Note 1: Default assignment for ECCP2 when the CCP2MX Configuration bit is set.

2: Alternate assignment for ECCP2 when the CCP2MX Configuration bit is cleared.

3: Not available on PIC18F65K22 and PIC18F85K22 devices.

4: PSP is available only in Microcontroller mode.

5: The CC6, CCP7, CCP8 and CCP9 pin placement depends on the setting of the ECCPMX Configuration bit (CONFIG3H<1>).

0

OD

= Output

Pin Name	Pin Number	Pin	Buffer	Description
	TQFP	Туре	Туре	Description
				PORTF is a bidirectional I/O port.
RF1/AN6/C2OUT/CTDIN RF1 AN6 C2OUT CTDIN	23	I/O I O I	ST Analog — ST	Digital I/O. Analog Input 6. Comparator 2 output. CTMU pulse delay input.
RF2/AN7/C1OUT RF2 AN7 C1OUT	18	I/O I O	ST Analog —	Digital I/O. Analog Input 7. Comparator 1 output.
RF3/AN8/C2INB/CTMUI RF3 AN8 C2INB CTMUI	17	I/O I I O	ST Analog Analog —	Digital I/O. Analog Input 8. Comparator 2 Input B. CTMU pulse generator charger for the C2INB comparator input.
RF4/AN9/C2INA RF4 AN9 C2INA	16	I/O I I	ST Analog Analog	Digital I/O. Analog Input 9. Comparator 2 Input A.
RF5/AN10/C1INB RF5 AN10 C1INB	15	I/O I I	ST Analog Analog	Digital I/O. Analog Input 10. Comparator 1 Input B.
RF6/AN11/C1INA RF6 AN11 C1INA	14	I/O I I	ST Analog Analog	Digital I/O. Analog Input 11. Comparator 1 Input A.
RF7/AN5/SS1 RF7 <u>AN5</u> SS1	13	I/O O I	ST Analog ST	Digital I/O. Analog Input 5. SPI slave select input.
Legend: TTL = TTL com ST = Schmitt T I = Input	patible input Trigger input wit	h CMC	)S levels	CMOS = CMOS compatible input or output Analog = Analog input O = Output

Note 1: Default assignment for ECCP2 when the CCP2MX Configuration bit is set.

2: Alternate assignment for ECCP2 when the CCP2MX Configuration bit is cleared.

3: Not available on PIC18F65K22 and PIC18F85K22 devices.

**4:** PSP is available only in Microcontroller mode.

= Power

 $I^2C = I^2C^{\text{TM}}/\text{SMBus}$ 

Р

5: The CC6, CCP7, CCP8 and CCP9 pin placement depends on the setting of the ECCPMX Configuration bit (CONFIG3H<1>).

OD

Pin Name	Pin Number	Pin	Buffer Type	Description
	TQFP	Туре		Description
				PORTG is a bidirectional I/O port.
RG0/ECCP3/P3A RG0 ECCP3 P3A	5	I/O I/O O	ST ST	Digital I/O. Capture 3 input/Compare 3 output/PWM3 output. ECCP3 PWM Output A.
RG1/TX2/CK2/AN19/ C3OUT RG1 TX2 CK2 AN19 C3OUT	6	1/0 1/0 1/0 0	ST — ST Analog —	Digital I/O. EUSART asynchronous transmit. EUSART synchronous clock (see related RX2/DT2). Analog Input 19. Comparator 3 output.
RG2/RX2/DT2/AN18/ C3INA RG2 RX2 DT2 AN18 C3INA	7	I/O    /O   	ST ST ST Analog Analog	Digital I/O. EUSART asynchronous receive. EUSART synchronous data (see related TX2/CK2). Analog Input 18. Comparator 3 Input A.
RG3/CCP4/AN17/P3D/ C3INB RG3 CCP4 AN17 P3D C3INB	8	1/0 1/0 	ST ST Analog Analog	Digital I/O. Capture 4 input/Compare 4 output/PWM4 output. Analog Input 17. ECCP3 PWM Output D. Comparator 3 Input B.
RG4/RTCC/T7CKI/T5G/ CCP5/AN16/P1D/C3INC RG4 RTCC T7CKI <sup>(3)</sup> T5G CCP5 AN16 P1D C3INC	10	1/0 0 1/0 - 0 -	ST ST ST ST Analog Analog	Digital I/O. RTCC output. Timer7 clock input. Timer5 external clock gate input. Capture 5 input/Compare 5 output/PWM5 output. Analog Input 16. ECCP1 PWM Output D. Comparator 3 Input C.
RG5	9			See the MCLR/RG5 pin.
Legend: TTL = TTL com		h CMC	)S levels	CMOS = CMOS compatible input or output Analog = Analog input O = Output OD = Open-Drain (no P diode to VDD)

P = Power $I^{2}C = I^{2}C^{TM}/SMBus$ 

OD = Open-Drain (no P diode to VDD)

**Note 1:** Default assignment for ECCP2 when the CCP2MX Configuration bit is set.

**2:** Alternate assignment for ECCP2 when the CCP2MX Configuration bit is cleared.

- **3:** Not available on PIC18F65K22 and PIC18F85K22 devices.
- 4: PSP is available only in Microcontroller mode.

## PIC18F87K22 FAMILY

Pin Name	Pin Number	Pin	Buffer	Description
	TQFP	Туре	Туре	
				PORTH is a bidirectional I/O port.
RH0/AN23/A16 RH0 AN23 A16	79	I/O I I/O	ST Analog TTL	Digital I/O. Analog Input 23. External Memory Address/Data 16.
RH1/AN22/A17 RH1 AN22 A17	80	I/O I I/O	ST Analog TTL	Digital I/O. Analog Input 22. External Address/Data 17.
RH2/AN21/A18 RH2 AN21 A18	1	I/O I I/O	ST Analog TTL	Digital I/O. Analog Input 21. External Address/Data 18.
RH3/AN20/A19 RH3 AN20 A19	2	I/O I I/O	ST Analog TTL	Digital I/O. Analog Input 20. External Address/Data 19.
RH4/CCP9/P3C/AN12/ C2INC RH4 CCP9 <sup>(3,5)</sup> P3C AN12 C2INC	22	1/0 1/0 	ST ST — Analog Analog	Digital I/O. Capture 9 input/Compare 9 output/PWM9 output. ECCP3 PWM Output C. Analog Input 12. Comparator 2 Input C.
RH5/CCP8/P3B/AN13/ C2IND RH5 CCP8 <sup>(5)</sup> P3B AN13 C2IND	21	I/O I/O I -	ST ST — Analog Analog	Digital I/O. Capture 8 input/Compare 8 output/PWM8 output. ECCP3 PWM Output B. Analog Input 13. Comparator 1 Input D.
RH6/CCP7/P1C/AN14/ C1INC RH6 CCP7 <sup>(5)</sup> P1C AN14 C1INC	20 Patible input	I/O I/O I	ST ST — Analog Analog	Digital I/O. Capture 7 input/Compare 7 output/PWM7 output. ECCP1 PWM Output C. Analog Input 14. Comparator 1 Input C. CMOS = CMOS compatible input or output

0

OD

5: The CC6, CCP7, CCP8 and CCP9 pin placement depends on the setting of the ECCPMX Configuration bit

= Output

= Open-Drain (no P diode to VDD)

#### TABLE 1-4: PIC18F8XK22 PINOUT I/O DESCRIPTIONS (CONTINUED)

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L

= Input

 $I^2C = I^2C^{TM}/SMBus$ 

(CONFIG3H<1>).

Note 1: Default assignment for ECCP2 when the CCP2MX Configuration bit is set.

3: Not available on PIC18F65K22 and PIC18F85K22 devices.

4: PSP is available only in Microcontroller mode.

2: Alternate assignment for ECCP2 when the CCP2MX Configuration bit is cleared.

P = Power

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= Open-Drain (no P diode to VDD)

#### TABLE 1-4: PIC18F8XK22 PINOUT I/O DESCRIPTIONS (CONTINUED)

Pin Name	Pin Number	Pin		Description
	TQFP	Туре		
RH7/CCP6/P1B/AN15	19			
RH7		I/O	ST	Digital I/O.
CCP6 <sup>(5)</sup>		I/O	ST	Capture 6 input/Compare 6 output/PWM6 output.
P1B		0	—	ECCP1 PWM Output B.
AN15		I	Analog	Analog Input 15.
Legend: TTL = TTL compatible input				CMOS = CMOS compatible input or output
ST = Schmitt Trigger input with CMOS levels			Analog = Analog input	
I = Input				O = Output

= Power Ρ

 $I^2C = I^2C^{TM}/SMBus$ 

Note 1: Default assignment for ECCP2 when the CCP2MX Configuration bit is set.

2: Alternate assignment for ECCP2 when the CCP2MX Configuration bit is cleared.

3: Not available on PIC18F65K22 and PIC18F85K22 devices.

4: PSP is available only in Microcontroller mode.

5: The CC6, CCP7, CCP8 and CCP9 pin placement depends on the setting of the ECCPMX Configuration bit (CONFIG3H<1>).

OD

Pin Name	Pin Number	Pin	Buffer	Description
Fill Name	TQFP	Туре Туре		Description
				PORTJ is a bidirectional I/O port.
RJ0/ALE RJ0 ALE	62	I/O O	ST —	Digital I/O. External memory address latch enable.
RJ1/OE RJ1 OE	61	I/O O	ST —	Digital I/O. External memory output enable.
RJ2/WRL RJ2 WRL	60	I/O O	ST —	Digital I/O. External memory write low control.
RJ3/WRH RJ3 WRH	59	I/O O	ST —	Digital I/O. External memory high control.
RJ4/BA0 RJ4 BA0	39	I/O O	ST —	Digital I/O. External Memory Byte Address 0 control
RJ5/CE RJ5 CE	40	I/O O	ST —	Digital I/O External memory chip enable control.
RJ6/LB RJ6 LB	41	I/O O	ST —	Digital I/O. External memory low byte control.
RJ7/UB RJ7 UB	42	I/O O	ST —	Digital I/O. External memory high byte control.
Vss	11, 31, 51, 70	Р	—	Ground reference for logic and I/O pins.
VDD	32, 48, 71	Р	—	Positive supply for logic and I/O pins.
AVss	26	Р	—	Ground reference for analog modules.
AVDD	25	P		Positive supply for analog modules.
ENVREG VDDCORE/VCAP	24 12	I	ST	Enable for on-chip voltage regulator. Core logic power or external filter capacitor connection.
Vddcore Vcap		Ρ	_	External filter capacitor connection (regulator enabled/disabled).
I = Input P = Power I <sup>2</sup> C = I <sup>2</sup> C™/SM	rigger input with Bus			CMOS = CMOS compatible input or output Analog = Analog input O = Output OD = Open-Drain (no P diode to VDD)

2: Alternate assignment for ECCP2 when the CCP2MX Configuration bit is cleared.

3: Not available on PIC18F65K22 and PIC18F85K22 devices.

4: PSP is available only in Microcontroller mode.

# 2.0 GUIDELINES FOR GETTING STARTED WITH PIC18FXXKXX MICROCONTROLLERS

# 2.1 Basic Connection Requirements

Getting started with the PIC18F87K22 family family of 8-bit microcontrollers requires attention to a minimal set of device pin connections before proceeding with development.

The following pins must always be connected:

- All VDD and Vss pins (see Section 2.2 "Power Supply Pins")
- All AVDD and AVss pins, regardless of whether or not the analog device features are used (see Section 2.2 "Power Supply Pins")
- MCLR pin
   (see Section 2.3 "Master Clear (MCLR) Pin")
- ENVREG (if implemented) and VCAP/VDDCORE pins (see Section 2.4 "Voltage Regulator Pins (ENVREG and VCAP/VDDCORE)")

These pins must also be connected if they are being used in the end application:

- PGC/PGD pins used for In-Circuit Serial Programming<sup>™</sup> (ICSP<sup>™</sup>) and debugging purposes (see Section 2.5 "ICSP Pins")
- OSCI and OSCO pins when an external oscillator source is used

(see Section 2.6 "External Oscillator Pins")

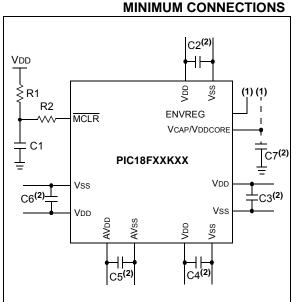
Additionally, the following pins may be required:

• VREF+/VREF- pins are used when external voltage reference for analog modules is implemented

**Note:** The AVDD and AVss pins must always be connected, regardless of whether any of the analog modules are being used.

The minimum mandatory connections are shown in Figure 2-1.

# FIGURE 2-1: RECOMMENDED



#### Key (all values are recommendations):

C1 through C6: 0.1  $\mu\text{F},$  20V ceramic R1: 10 k $\Omega$ 

R2: 100Ω to 470Ω

- Note 1: See Section 2.4 "Voltage Regulator Pins (ENVREG and VCAP/VDDCORE)" for explanation of ENVREG pin connections.
  - 2: The example shown is for a PIC18F device with five VDD/Vss and AVDD/AVss pairs. Other devices may have more or less pairs; adjust the number of decoupling capacitors appropriately.

# 2.2 Power Supply Pins

#### 2.2.1 DECOUPLING CAPACITORS

The use of decoupling capacitors on every pair of power supply pins, such as VDD, VSS, AVDD and AVSS, is required.

Consider the following criteria when using decoupling capacitors:

- Value and type of capacitor: A 0.1  $\mu$ F (100 nF), 10-20V capacitor is recommended. The capacitor should be a low-ESR device, with a resonance frequency in the range of 200 MHz and higher. Ceramic capacitors are recommended.
- Placement on the printed circuit board: The decoupling capacitors should be placed as close to the pins as possible. It is recommended to place the capacitors on the same side of the board as the device. If space is constricted, the capacitor can be placed on another layer on the PCB using a via; however, ensure that the trace length from the pin to the capacitor is no greater than 0.25 inch (6 mm).
- Handling high-frequency noise: If the board is experiencing high-frequency noise (upward of tens of MHz), add a second ceramic type capacitor in parallel to the above described decoupling capacitor. The value of the second capacitor can be in the range of 0.01  $\mu$ F to 0.001  $\mu$ F. Place this second capacitor next to each primary decoupling capacitor. In high-speed circuit designs, consider implementing a decade pair of capacitances as close to the power and ground pins as possible (e.g., 0.1  $\mu$ F in parallel with 0.001  $\mu$ F).
- Maximizing performance: On the board layout from the power supply circuit, run the power and return traces to the decoupling capacitors first, and then to the device pins. This ensures that the decoupling capacitors are first in the power chain. Equally important is to keep the trace length between the capacitor and the power pins to a minimum, thereby reducing PCB trace inductance.

### 2.2.2 TANK CAPACITORS

On boards with power traces running longer than six inches in length, it is suggested to use a tank capacitor for integrated circuits, including microcontrollers, to supply a local power source. The value of the tank capacitor should be determined based on the trace resistance that connects the power supply source to the device, and the maximum current drawn by the device in the application. In other words, select the tank capacitor so that it meets the acceptable voltage sag at the device. Typical values range from 4.7  $\mu F$  to 47  $\mu F$ .

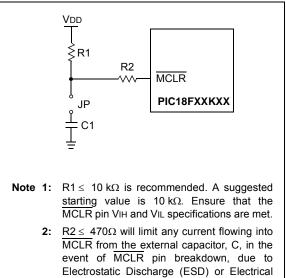
# 2.3 Master Clear (MCLR) Pin

The MCLR pin provides two specific device functions: Device Reset, and Device Programming and Debugging. If programming and debugging are not required in the end application, a direct connection to VDD may be all that is required. The addition of other components, to help increase the application's resistance to spurious Resets from voltage sags, may be beneficial. A typical configuration is shown in Figure 2-1. Other circuit designs may be implemented, depending on the application's requirements.

During programming and debugging, the resistance and capacitance that can be added to the pin must be considered. Device programmers and debuggers drive the MCLR pin. Consequently, specific voltage levels (VIH and VIL) and fast signal transitions must not be adversely affected. Therefore, specific values of R1 and C1 will need to be adjusted based on the application and PCB requirements. For example, it is recommended that the capacitor, C1, be isolated from the MCLR pin during programming and debugging operations by using a jumper (Figure 2-2). The jumper is replaced for normal run-time operations.

Any components associated with the  $\overline{\text{MCLR}}$  pin should be placed within 0.25 inch (6 mm) of the pin.

#### FIGURE 2-2: EXAMPLE OF MCLR PIN CONNECTIONS



Overstress (EOS). Ensure that the MCLR pin

VIH and VIL specifications are met.

# 2.4 Voltage Regulator Pins (ENVREG and VCAP/VDDCORE)

The on-chip voltage regulator enable pin, ENVREG, must always be connected directly to either a supply voltage or to ground. Tying ENVREG to VDD enables the regulator, while tying it to ground disables the regulator. Refer to **Section 28.3 "On-Chip Voltage Regulator**" for details on connecting and using the on-chip regulator.

When the regulator is enabled, a low-ESR (< 5 $\Omega$ ) capacitor is required on the VCAP/VDDCORE pin to stabilize the voltage regulator output voltage. The VCAP/VDDCORE pin must not be connected to VDD and must use a capacitor of 10  $\mu$ F connected to ground. The type can be ceramic or tantalum. Suitable examples of capacitors are shown in Table 2-1. Capacitors with equivalent specifications can be used.

Designers may use Figure 2-3 to evaluate ESR equivalence of candidate devices.

It is recommended that the trace length not exceed 0.25 inch (6 mm). Refer to **Section 31.0** "**Electrical Characteristics**" for additional information.

When the regulator is disabled, the VCAP/VDDCORE pin must be tied to a voltage supply at the VDDCORE level. Refer to **Section 31.0 "Electrical Characteristics"** for information on VDD and VDDCORE.

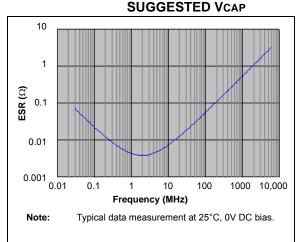
Some PIC18FXXKXX families, or some devices within a family, do not provide the option of enabling or disabling the on-chip voltage regulator:

- Some devices (with the name, PIC18LFXXKXX) permanently disable the voltage regulator. These devices do not have the ENVREG pin and require a 0.1  $\mu$ F capacitor on the VCAP/VDDCORE pin. The VDD level of these devices must comply with the "voltage regulator disabled" specification for Parameter D001, in Section 31.0 "Electrical Characteristics".
- Some devices permanently enable the voltage regulator.
   These devices also do not have the ENV/REC picture.

These devices also do not have the ENVREG pin. The 10  $\mu F$  capacitor is still required on the VCAP/VDDCORE pin.

FIGURE 2-3:

FREQUENCY vs. ESR PERFORMANCE FOR



# TABLE 2-1: SUITABLE CAPACITOR EQUIVALENTS

Make	Part #	Nominal Capacitance	Base Tolerance	Rated Voltage	Temp. Range
TDK	C3216X7R1C106K	10 µF	±10%	16V	-55 to 125°C
TDK	C3216X5R1C106K	10 µF	±10%	16V	-55 to 85°C
Panasonic	ECJ-3YX1C106K	10 µF	±10%	16V	-55 to 125°C
Panasonic	ECJ-4YB1C106K	10 µF	±10%	16V	-55 to 85°C
Murata	GRM32DR71C106KA01L	10 µF	±10%	16V	-55 to 125°C
Murata	GRM31CR61C106KC31L	10 µF	±10%	16V	-55 to 85°C

# 2.4.1 CONSIDERATIONS FOR CERAMIC CAPACITORS

In recent years, large value, low-voltage, surface-mount ceramic capacitors have become very cost effective in sizes up to a few tens of microfarad. The low-ESR, small physical size and other properties make ceramic capacitors very attractive in many types of applications.

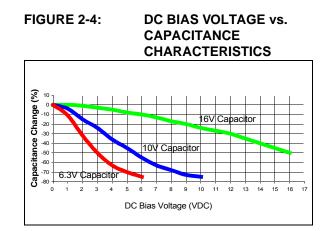
Ceramic capacitors are suitable for use with the internal voltage regulator of this microcontroller. However, some care is needed in selecting the capacitor to ensure that it maintains sufficient capacitance over the intended operating range of the application.

Typical low-cost, 10  $\mu$ F ceramic capacitors are available in X5R, X7R and Y5V dielectric ratings (other types are also available, but are less common). The initial tolerance specifications for these types of capacitors are often specified as ±10% to ±20% (X5R and X7R), or -20%/+80% (Y5V). However, the effective capacitance that these capacitors provide in an application circuit will also vary based on additional factors, such as the applied DC bias voltage and the temperature. The total in-circuit tolerance is, therefore, much wider than the initial tolerance specification.

The X5R and X7R capacitors typically exhibit satisfactory temperature stability (ex:  $\pm 15\%$  over a wide temperature range, but consult the manufacturer's data sheets for exact specifications). However, Y5V capacitors typically have extreme temperature tolerance specifications of  $\pm 22\%/-82\%$ . Due to the extreme temperature tolerance, a 10  $\mu$ F nominal rated Y5V type capacitor may not deliver enough total capacitance to meet minimum internal voltage regulator stability and transient response requirements. Therefore, Y5V capacitors are not recommended for use with the internal regulator if the application must operate over a wide temperature range.

In addition to temperature tolerance, the effective capacitance of large value ceramic capacitors can vary substantially, based on the amount of DC voltage applied to the capacitor. This effect can be very significant, but is often overlooked or is not always documented.

A typical DC bias voltage vs. capacitance graph for X7R type and Y5V type capacitors is shown in Figure 2-4.



When selecting a ceramic capacitor to be used with the internal voltage regulator, it is suggested to select a high-voltage rating, so that the operating voltage is a small percentage of the maximum rated capacitor voltage. For example, choose a ceramic capacitor rated at 16V for the 2.5V core voltage. Suggested capacitors are shown in Table 2-1.

# 2.5 ICSP Pins

The PGC and PGD pins are used for In-Circuit Serial Programming<sup>TM</sup> (ICSP<sup>TM</sup>) and debugging purposes. It is recommended to keep the trace length between the ICSP connector and the ICSP pins on the device as short as possible. If the ICSP connector is expected to experience an ESD event, a series resistor is recommended, with the value in the range of a few tens of ohms, not to exceed 100 $\Omega$ .

Pull-up resistors, series diodes and capacitors on the PGC and PGD pins are not recommended as they will interfere with the programmer/debugger communications to the device. If such discrete components are an application requirement, they should be removed from the circuit during programming and debugging. Alternatively, refer to the AC/DC characteristics and timing requirements information in the respective device Flash programming specification for information on capacitive loading limits, and pin input voltage high (VIH) and input low (VIL) requirements.

For device emulation, ensure that the "Communication Channel Select" (i.e., PGCx/PGDx pins), programmed into the device, matches the physical connections for the ICSP to the Microchip debugger/emulator tool.

For more information on available Microchip development tools connection requirements, refer to **Section 30.0 "Development Support"**.

# 2.6 External Oscillator Pins

Many microcontrollers have options for at least two oscillators: a high-frequency primary oscillator and a low-frequency secondary oscillator (refer to **Section 3.0 "Oscillator Configurations**" for details).

The oscillator circuit should be placed on the same side of the board as the device. Place the oscillator circuit close to the respective oscillator pins with no more than 0.5 inch (12 mm) between the circuit components and the pins. The load capacitors should be placed next to the oscillator itself, on the same side of the board.

Use a grounded copper pour around the oscillator circuit to isolate it from surrounding circuits. The grounded copper pour should be routed directly to the MCU ground. Do not run any signal traces or power traces inside the ground pour. Also, if using a two-sided board, avoid any traces on the other side of the board where the crystal is placed.

Layout suggestions are shown in Figure 2-4. In-line packages may be handled with a single-sided layout that completely encompasses the oscillator pins. With fine-pitch packages, it is not always possible to completely surround the pins and components. A suitable solution is to tie the broken guard sections to a mirrored ground layer. In all cases, the guard trace(s) must be returned to ground.

In planning the application's routing and I/O assignments, ensure that adjacent port pins, and other signals in close proximity to the oscillator, are benign (i.e., free of high frequencies, short rise and fall times, and other similar noise).

For additional information and design guidance on oscillator circuits, please refer to these Microchip Application Notes, available at the corporate web site (www.microchip.com):

- AN826, "Crystal Oscillator Basics and Crystal Selection for rfPIC<sup>™</sup> and PICmicro<sup>®</sup> Devices"
- AN849, "Basic PICmicro<sup>®</sup> Oscillator Design"
- AN943, "Practical PICmicro<sup>®</sup> Oscillator Analysis and Design"
- AN949, "Making Your Oscillator Work"

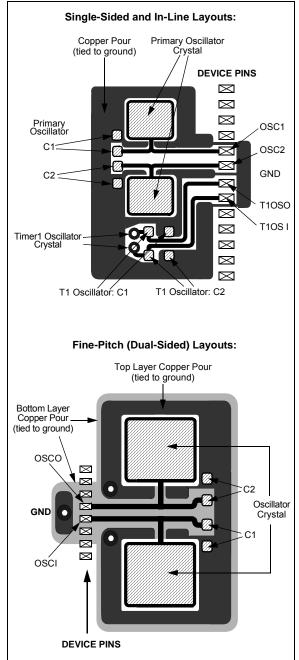
# 2.7 Unused I/Os

Unused I/O pins should be configured as outputs and driven to a logic low state. Alternatively, connect a 1 k $\Omega$  to 10 k $\Omega$  resistor to Vss on unused pins and drive the output to logic low.

#### FIGURE 2-5:

#### PLACEMENT OF THE OSCILLATOR CIRCUIT

SUGGESTED



NOTES:

# 3.0 OSCILLATOR CONFIGURATIONS

# 3.1 Oscillator Types

The PIC18F87K22 family of devices can be operated in the following oscillator modes:

- EC External clock, RA6 available
- ECIO External clock, clock out RA6 (Fosc/4 on RA6)
- HS High-Speed Crystal/Resonator
- XT Crystal/Resonator
- LP Low-Power Crystal
- RC External Resistor/Capacitor, RA6 available
- RCIO External Resistor/Capacitor, clock out RA6 (Fosc/4 on RA6)
- INTIO2 Internal Oscillator with I/O on RA6 and RA7
- INTIO1 Internal Oscillator with Fosc/4 output on RA6 and I/O on RA7

There is also an option for running the 4xPLL on any of the clock sources in the input frequency range of 4 to 16 MHz.

The PLL is enabled by setting the PLLCFG bit (CONFIG1H<4>) or the PLLEN bit (OSCTUNE<6>).

For the EC and HS mode, the PLLEN (software) or PLLCFG (CONFIG) bit can be used to enable the PLL.

For the INTIOx modes (HF-INTOSC):

- Only the PLLEN can enable the PLL (PLLCFG is ignored).
- When the oscillator is configured for the internal oscillator (FOSC<3:0> = 100x), the PLL can be enabled only when the HF-INTOSC frequency is 8 or 16 MHz.

When the RA6 and RA7 pins are not used for an oscillator function or CLKOUT function, they are available as general purpose I/Os. To optimize power consumption when using EC/HS/ XT/LP/RC as the primary oscillator, the frequency input range can be configured to yield an optimized power bias:

- Low-Power Bias External frequency less than 160 kHz
- Medium Power Bias External frequency between 160 kHz and 16 MHz
- High-Power Bias External frequency greater than 16 MHz

All of these modes are selected by the user by programming the FOSC<3:0> Configuration bits (CONFIG1H<3:0>). In addition, PIC18F87K22 family devices can switch between different clock sources, either under software control or, under certain conditions, automatically. This allows for additional power savings by managing device clock speed in real time without resetting the application. The clock sources for the PIC18F87K22 family of devices are shown in Figure 3-1.

For the HS and EC mode, there are additional power modes of operation – depending on the frequency of operation.

HS1 is the Medium Power mode with a frequency range of 4 MHz to 16 MHz. HS2 is the High-Power mode, where the oscillator frequency can go from 16 MHz to 25 MHz. HS1 and HS2 are achieved by setting the CONFIG1H<3:0> correctly. (For details, see Register 28-2 on page 406.)

EC mode has these modes of operation:

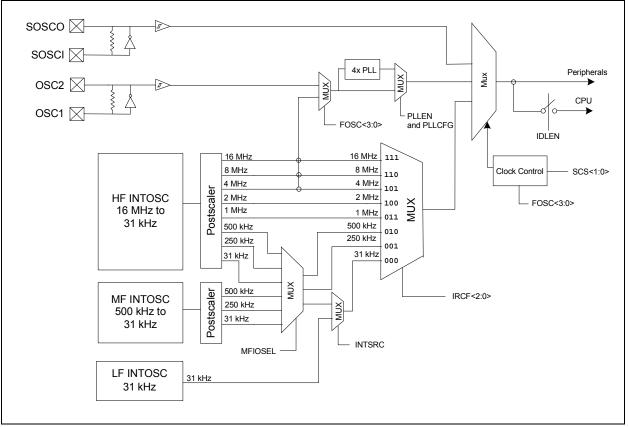
- EC1 For low power with a frequency range up to 160 kHz
- EC2 Medium power with a frequency range of 160 kHz to 16 MHz
- EC3 High power with a frequency range of 16 MHz to 64 MHz

EC1, EC2 and EC3 are achieved by setting the CONFIG1H<3:0> correctly. (For details, see Register 28-2 on page 406.)

Table 3-1 shows the HS and EC modes' frequencyrange and FOSC<3:0> settings.

Mode	Frequency Range	FOSC<3:0> Setting
EC1 (low power)		1101
(EC1 & EC1IO)	DC-160 kHz	1100
EC2 (medium power)	160 kHz-16 MHz	1011
(EC2 & EC2IO)		1010
EC3 (high power)	16 MHz-64 MHz	0101
(EC3 & EC3IO)		0100
HS1 (medium power)	4 MHz-16 MHz	0011
HS2 (high power)	16 MHz-25 MHz	0010
ХТ	100 kHz-4 MHz	0001
LP	31.25 kHz	0000
RC (External)	0-4 MHz	001x
ΙΝΤΙΟ	32 KHz-16 MHz	100x (and OSCCON, OSCCON2)





# 3.2 Control Registers

The OSCCON register (Register 3-1) controls the main aspects of the device clock's operation. It selects the oscillator type to be used, which of the power-managed modes to invoke and the output frequency of the INTOSC source. It also provides status on the oscillators.

The OSCTUNE register (Register 3-3) controls the tuning and operation of the internal oscillator block. It also implements the PLLEN bit which controls the operation of the Phase Locked Loop (PLL) (see Section 3.5.3 "PLL Frequency Multiplier").

## REGISTER 3-1: OSCCON: OSCILLATOR CONTROL REGISTER<sup>(1)</sup>

R/W-0	R/W-1	R/W-1	R/W-0	R <sup>(1)</sup>	R-0	R/W-0	R/W-0
IDLEN	IRCF2 <sup>(2)</sup>	IRCF1 <sup>(2)</sup>	IRCF0 <sup>(2)</sup>	OSTS	HFIOFS	SCS1 <sup>(4)</sup>	SCS0 <sup>(4)</sup>
bit 7							bit 0
l egend:							

Legena:			
R = Readable bit	W = Writable bit	U = Unimplemented bit, read	d as '0'
-n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown

bit 7	IDLEN: Idle Enable bit
	1 = Device enters an Idle mode when a SLEEP instruction is executed
	0 = Device enters Sleep mode when a SLEEP instruction is executed
bit 6-4	IRCF<2:0>: Internal Oscillator Frequency Select bits <sup>(2)</sup>
	111 = HF-INTOSC output frequency is used (16 MHz)
	110 = HF-INTOSC/2 output frequency is used (8 MHz, default)
	101 = HF-INTOSC/4 output frequency is used (4 MHz)
	100 = HF-INTOSC/8 output frequency is used (2 MHz)
	011 = HF-INTOSC/16 output frequency is used (1 MHz)
	If INTSRC = 0 and MFIOSEL = 0: <sup>(3,5)</sup>
	010 = HF-INTOSC/32 output frequency is used (500 kHz)
	001 = HF-INTOSC/64 output frequency is used (250 kHz)
	000 = LF-INTOSC output frequency is used (31.25 kHz)
	$\frac{\text{If INTSRC} = 1 \text{ and } \text{MFIOSEL} = 0}{(3,5)}$
	010 = HF-INTOSC/32 output frequency is used (500 kHz)
	001 = HF-INTOSC/64 output frequency is used (250 kHz)
	000 = HF-INTOSC/512 output frequency is used (31.25 kHz) If INTSRC = 0 and MFIOSEL = 1: <sup>(3,5)</sup>
	010 = MF-INTOSC output frequency is used (500 kHz)
	001 = MF-INTOSC output frequency is used (250 kHz)
	000 = LF-INTOSC output frequency is used (31.25 kHz)(6)
	If INTSRC = 1 and MFIOSEL = $1$ : <sup>(3,5)</sup>
	010 = MF-INTOSC output frequency is used (500 kHz)
	001 = MF-INTOSC/2 output frequency is used (250 kHz)
	000 = MF-INTOSC/16 output frequency is used (31.25 kHz)
bit 3	OSTS: Oscillator Start-up Timer Time-out Status bit <sup>(1)</sup>
Dit 0	1 = Oscillator Start-up Timer (OST) time-out has expired; primary oscillator is running, as defined by
	FOSC<3:0>
	0 = Oscillator Start-up Timer (OST) time-out is running; primary oscillator is not ready – device is
	running from internal oscillator (HF-INTOSC, MF-INTOSC or LF-INTOSC)
Note 1:	The Reset state depends on the state of the IESO Configuration bit (CONFIG1H<7>).
2:	Modifying these bits will cause an immediate clock frequency switch if the internal oscillator is providing
۷.	the device clocks.
3:	Source selected by the INTSRC bit (OSCTUNE<7>).
3. 4:	
	Modifying these bits will cause an immediate clock source switch.
5:	INTSRC = OSCTUNE<7> and MFIOSEL = OSCCON2<0>.

- 5: INTSRC = OSCTUNE<7> and MFIOSEL = OSCCON2<0>.
- **6:** Lowest power option for an internal source.

# **REGISTER 3-1:** OSCCON: OSCILLATOR CONTROL REGISTER<sup>(1)</sup> (CONTINUED)

- bit 2 **HFIOFS:** INTOSC Frequency Stable bit
  - 1 = HF-INTOSC oscillator frequency is stable
    - 0 = HF-INTOSC oscillator frequency is not stable
- bit 1-0 SCS<1:0>: System Clock Select bits<sup>(4)</sup>
  - 1x = Internal oscillator block (LF-INTOSC, MF-INTOSC or HF-INTOSC)
  - 01 = SOSC oscillator
  - 00 = Default primary oscillator (OSC1/OSC2 or HF-INTOSC with or without PLL; defined by the FOSC<3:0> Configuration bits, CONFIG1H<3:0>.)
- Note 1: The Reset state depends on the state of the IESO Configuration bit (CONFIG1H<7>).
  - 2: Modifying these bits will cause an immediate clock frequency switch if the internal oscillator is providing the device clocks.
  - 3: Source selected by the INTSRC bit (OSCTUNE<7>).
  - **4:** Modifying these bits will cause an immediate clock source switch.
  - **5:** INTSRC = OSCTUNE<7> and MFIOSEL = OSCCON2<0>.
  - 6: Lowest power option for an internal source.

## REGISTER 3-2: OSCCON2: OSCILLATOR CONTROL REGISTER 2

U-0	R-0	U-0	U-0	R/W-0	U-0	R-x	R/W-0
—	SOSCRUN		—	SOSCGO	—	MFIOFS	MFIOSEL
bit 7							bit 0

Legend:			
R = Readable bit	W = Writable bit	U = Unimplemented bit	, read as '0'
-n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown

bit 7	Unimplemented: Read as '0'
bit 6	SOSCRUN: SOSC Run Status bit
	<ul> <li>1 = System clock comes from a secondary SOSC</li> <li>0 = System clock comes from an oscillator other than SOSC</li> </ul>
bit 5-4	Unimplemented: Read as '0'
bit 3	SOSCGO: Oscillator Start Control bit
	<ul> <li>1 = Oscillator is running, even if no other sources are requesting it</li> <li>0 = Oscillator is shut off if no other sources are requesting it (When the SOSC is selected to run from a digital clock input, rather than an external crystal, this bit has no effect.)</li> </ul>
bit 2	Unimplemented: Read as '0'
bit 1	MFIOFS: MF-INTOSC Frequency Stable bit
	1 = MF-INTOSC is stable 0 = MF-INTOSC is not stable
bit 0	MFIOSEL: MF-INTOSC Select bit
	<ul> <li>1 = MF-INTOSC is used in place of HF-INTOSC frequencies of 500 kHz, 250 kHz and 31.25 kHz</li> <li>0 = MF-INTOSC is not used</li> </ul>

R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0
INTSRC	PLLEN	TUN5	TUN4	TUN3	TUN2	TUN1	TUN0
bit 7							bit 0
Legend:							
R = Readable	e bit	W = Writable	bit	U = Unimplem	nented bit, read	l as '0'	
-n = Value at	POR	'1' = Bit is set		'0' = Bit is clea	ared	x = Bit is unkr	iown
bit 7	1 = 31.25 k⊦	z device clock	derived from 1	y Source Select 6 MHz INTOSC : RC 31 kHz oscil	source (divide-		I, HF-INTOSC)
bit 6		quency Multiplie nabled				,	
bit 5-0	011111 = Ma 0000001 000000 = Ce 111111	aximum frequer • •	; fast RC oscill	Frequency Tunir	-	d frequency	

#### REGISTER 3-3: OSCTUNE: OSCILLATOR TUNING REGISTER

## 3.3 Clock Sources and Oscillator Switching

Essentially, PIC18F87K22 family devices have these independent clock sources:

- Primary oscillators
- Secondary oscillators
- Internal oscillator

The **primary oscillators** can be thought of as the main device oscillators. These are any external oscillators connected to the OSC1 and OSC2 pins, and include the External Crystal and Resonator modes and the External Clock modes. If selected by the FOSC<3:0> Configuration bits (CONFIG1H<3:0>), the internal oscillator block may be considered a primary oscillator. The internal oscillator block can be one of the following:

- 31 kHz LF-INTRC source
- 31 kHz to 500 kHz MF-INTOSC source
- · 31 kHz to 16 MHz HF-INTOSC source

The particular mode is defined by the FOSC Configuration bits. The details of these modes are covered in Section 3.5 "External Oscillator Modes".

The **secondary oscillators** are external clock sources that are not connected to the OSC1 or OSC2 pin. These sources may continue to operate, even after the controller is placed in a power-managed mode. PIC18F87K22 family devices offer the SOSC (Timer1/3/5/7) oscillator as a secondary oscillator source. This oscillator, in all power-managed modes, is often the time base for functions, such as a Real-Time Clock (RTC).

The SOSC can be enabled from any peripheral that requests it. There are eight ways the SOSC can be enabled: if the SOSC is selected as the source by any of the odd timers, which is done by each respective SOSCEN bit (TxCON<3>), if the SOSC is selected as the RTCC source by the RTCOSC Configuration bit (CONFIG3L<1>), if the SOSC is selected as the CPU clock source by the SCS bits (OSCCON<1:0>) or if the SOSCGO bit is set (OSCCON2<3>). The SOSCGO bit is used to warm up the SOSC so that it is ready before any peripheral requests it.

The secondary oscillator has three Run modes. The SOSCSEL<1:0> bits (CONFIG1L<4:3>) decide the SOSC mode of operation:

- 11 = High-power SOSC circuit
- 10 = Digital (SCLKI) mode
- 01 = Low-power SOSC circuit

If a secondary oscillator is not desired and digital I/O on port pins, RC0 and RC1, is needed, the SOSCSEL bits must be set to Digital mode.

In addition to being a primary clock source in some circumstances, the **internal oscillator** is available as a power-managed mode clock source. The LF-INTOSC source is also used as the clock source for several special features, such as the WDT and Fail-Safe Clock Monitor. The internal oscillator block is discussed in more detail in **Section 3.6** "Internal Oscillator **Block**".

The PIC18F87K22 family includes features that allow the device clock source to be switched from the main oscillator, chosen by device configuration, to one of the alternate clock sources. When an alternate clock source is enabled, various power-managed operating modes are available.

# 3.3.1 OSC1/OSC2 OSCILLATOR

The OSC1/OSC2 oscillator block is used to provide the oscillator modes and frequency ranges:

Mode	Design Operating Frequency
LP	31.25-100 kHz
ХТ	100 kHz to 4 MHz
HS	4 MHz to 25 MHz
EC	0 to 64 MHz (external clock)
EXTRC	0 to 4 MHz (external RC)

The crystal-based oscillators (XT, HS and LP) have a built-in start-up time. The operation of the EC and EXTRC clocks is immediate.

### 3.3.2 CLOCK SOURCE SELECTION

The System Clock Select bits, SCS<1:0> (OSCCON2<1:0>), select the clock source. The available clock sources are the primary clock defined by the FOSC<3:0> Configuration bits, the secondary clock (SOSC oscillator) and the internal oscillator. The clock source changes after one or more of the bits is written to, following a brief clock transition interval.

The OSTS (OSCCON<3>) and SOSCRUN (OSCCON<6>) bits indicate which clock source is currently providing the device clock. The OSTS bit indicates that the Oscillator Start-up Timer (OST) has timed out and the primary clock is providing the device clock in primary clock modes. The SOSCRUN bit indicates when the SOSC oscillator (from Timer1/3/5/7) is providing the device clock in secondary clock modes. In power-managed modes, only one of these bits will be set at any time. If neither of these bits is set, the INTRC is providing the clock, or the internal oscillator has just started and is not yet stable.

The IDLEN bit (OSCCON<7>) determines if the device goes into Sleep mode or one of the Idle modes when the SLEEP instruction is executed.

The use of the flag and control bits in the OSCCON register is discussed in more detail in **Section 4.0** "**Power-Managed Modes**".

- Note 1: The Timer1/3/5/7 oscillator must be enabled to select the secondary clock source. The Timerx oscillator is enabled by setting the SOSCEN bit in the Timerx Control register (TxCON<3>). If the Timerx oscillator is not enabled, then any attempt to select a secondary clock source when executing a SLEEP instruction will be ignored.
  - 2: It is recommended that the Timerx oscillator be operating and stable before executing the SLEEP instruction or a very long delay may occur while the Timerx oscillator starts.

#### 3.3.2.1 System Clock Selection and Device Resets

Since the SCS bits are cleared on all forms of Reset, this means the primary oscillator, defined by the FOSC<3:0> Configuration bits, is used as the primary clock source on device Resets. This could either be the internal oscillator block by itself, or one of the other primary clock source (HS, EC, XT, LP, External RC and PLL-Enabled modes).

In those cases when the internal oscillator block, without PLL, is the default clock on Reset, the Fast RC oscillator (INTOSC) will be used as the device clock source. It will initially start at 8 MHz; the postscaler selection that corresponds to the Reset value of the IRCF<2:0> bits ('110').

Regardless of which primary oscillator is selected, INTRC will always be enabled on device power-up. It serves as the clock source until the device has loaded its configuration values from memory. It is at this point that the FOSC Configuration bits are read and the oscillator selection of the operational mode is made.

Note that either the primary clock source or the internal oscillator will have two bit setting options for the possible values of the SCS<1:0> bits, at any given time.

# 3.3.3 OSCILLATOR TRANSITIONS

PIC18F87K22 family devices contain circuitry to prevent clock "glitches" when switching between clock sources. A short pause in the device clock occurs during the clock switch. The length of this pause is the sum of two cycles of the old clock source and three to four cycles of the new clock source. This formula assumes that the new clock source is stable.

Clock transitions are discussed in greater detail in Section 4.1.2 "Entering Power-Managed Modes".

# 3.4 RC Oscillator

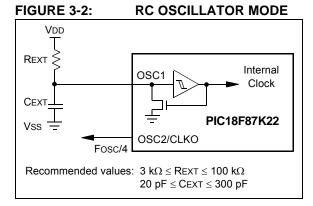
For timing-insensitive applications, the RC and RCIO Oscillator modes offer additional cost savings. The actual oscillator frequency is a function of several factors:

- Supply Voltage
- Values of the External Resistor (REXT) and Capacitor (CEXT)
- Operating Temperature

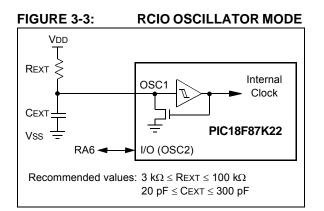
Given the same device, operating voltage and temperature, and component values, there will also be unit to unit frequency variations. These are due to factors, such as:

- Normal manufacturing variation
- Difference in lead frame capacitance between package types (especially for low CEXT values)
- Variations within the tolerance of limits of  $\ensuremath{\mathsf{REXT}}$  and  $\ensuremath{\mathsf{CEXT}}$

In the RC Oscillator mode, the oscillator frequency, divided by 4, is available on the OSC2 pin. This signal may be used for test purposes or to synchronize other logic. Figure 3-2 shows how the R/C combination is connected.



The RCIO Oscillator mode (Figure 3-3) functions like the RC mode, except that the OSC2 pin becomes an additional general purpose I/O pin. The I/O pin becomes bit 6 of PORTA (RA6).



# 3.5 External Oscillator Modes

# 3.5.1 CRYSTAL OSCILLATOR/CERAMIC RESONATORS (HS MODES)

In HS or HSPLL Oscillator modes, a crystal or ceramic resonator is connected to the OSC1 and OSC2 pins to establish oscillation. Figure 3-4 shows the pin connections.

The oscillator design requires the use of a crystal rated for parallel resonant operation.

**Note:** Use of a crystal rated for series resonant operation may give a frequency out of the crystal manufacturer's specifications.

# TABLE 3-2:CAPACITOR SELECTION FOR<br/>CERAMIC RESONATORS

Typical Capacitor Values Used:						
Mode	Freq.	OSC1	OSC2			
HS	8.0 MHz	27 pF	27 pF			

Capacitor values are for design guidance only.

22 pF

16.0 MHz

22 pF

Different capacitor values may be required to produce acceptable oscillator operation. The user should test the performance of the oscillator over the expected VDD and temperature range for the application. Refer to the following application notes for oscillator-specific information:

- AN588, "PIC<sup>®</sup> Microcontroller Oscillator Design Guide"
- AN826, "Crystal Oscillator Basics and Crystal Selection for rfPIC<sup>®</sup> and PIC<sup>®</sup> Devices"
- AN849, "Basic PIC<sup>®</sup> Oscillator Design"
- AN943, "Practical PIC<sup>®</sup> Oscillator Analysis and Design"
- AN949, "Making Your Oscillator Work"

See the notes following Table 3-3 for additional information.

# TABLE 3-3:CAPACITOR SELECTION FOR<br/>CRYSTAL OSCILLATOR

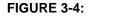
Osc Type	Crystal Freq.	Typical Capacitor Values Tested:		
	Fieq.	C1	C2	
HS	4 MHz	27 pF	27 pF	
	8 MHz	22 pF	22 pF	
	20 MHz	15 pF	15 pF	

#### Capacitor values are for design guidance only.

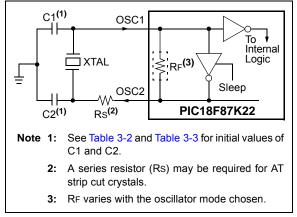
Different capacitor values may be required to produce acceptable oscillator operation. The user should test the performance of the oscillator over the expected VDD and temperature range for the application.

Refer to the Microchip application notes cited in Table 3-2 for oscillator-specific information. Also see the notes following this table for additional information.

- Note 1: Higher capacitance increases the stability of the oscillator but also increases the start-up time.
  - 2: Since each resonator/crystal has its own characteristics, the user should consult the resonator/crystal manufacturer for appropriate values of external components.
  - 3: Rs may be required to avoid overdriving crystals with a low drive level specification.
  - 4: Always verify oscillator performance over the VDD and temperature range that is expected for the application.



#### CRYSTAL/CERAMIC RESONATOR OPERATION (HS OR HSPLL CONFIGURATION)

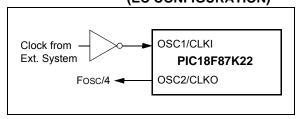


### 3.5.2 EXTERNAL CLOCK INPUT (EC MODES)

The EC and ECPLL Oscillator modes require an external clock source to be connected to the OSC1 pin. There is no oscillator start-up time required after a Power-on Reset or after an exit from Sleep mode.

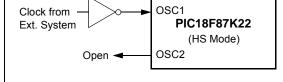
In the EC Oscillator mode, the oscillator frequency, divided by 4, is available on the OSC2 pin. This signal may be used for test purposes or to synchronize other logic. Figure 3-5 shows the pin connections for the EC Oscillator mode.

#### FIGURE 3-5: EXTERNAL CLOCK INPUT OPERATION (EC CONFIGURATION)



An external clock source may also be connected to the OSC1 pin in HS mode, as shown in Figure 3-6. In this configuration, the divide-by-4 output on OSC2 is not available. Current consumption in this configuration will be somewhat higher than EC mode, as the internal oscillator's feedback circuitry will be enabled (in EC mode, the feedback circuit is disabled).

# FIGURE 3-6: EXTERNAL CLOCK INPUT OPERATION (HS OSC CONFIGURATION)



# 3.5.3 PLL FREQUENCY MULTIPLIER

A Phase Locked Loop (PLL) circuit is provided as an option for users who want to use a lower frequency oscillator circuit, or to clock the device up to its highest rated frequency from a crystal oscillator. This may be useful for customers who are concerned with EMI due to high-frequency crystals, or users who require higher clock speeds from an internal oscillator.

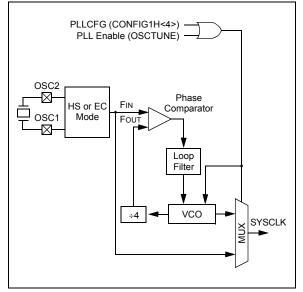
# 3.5.3.1 HSPLL and ECPLL Modes

The HSPLL and ECPLL modes provide the ability to selectively run the device at four times the external oscillating source to produce frequencies up to 64 MHz.

The PLL is enabled by setting the PLLEN bit (OSCTUNE<6>) or the PLLCFG bit (CONFIG1H<4>). The PLLEN bit provides a software control for the PLL, even if PLLCFG is set to '0'. The PLL is enabled only when the HS or EC oscillator frequency is within the 4 MHz to16 MHz input range.

This enables additional flexibility for controlling the application's clock speed in software. The PLLEN should be enabled in HS or EC Oscillator mode only if the input frequency is in the range of 4 MHz-16 MHz.

## FIGURE 3-7: PLL BLOCK DIAGRAM



### 3.5.3.2 PLL and HF-INTOSC

The PLL is available to the internal oscillator block when the internal oscillator block is configured as the primary clock source. In this configuration, the PLL is enabled in software and generates a clock output of up to 64 MHz.

The operation of INTOSC with the PLL is described in **Section 3.6.2 "INTPLL Modes**". Care should be taken that the PLL is enabled only if the HF-INTOSC postscaler is configured for 8 MHz or 16 MHz.

### 3.6 Internal Oscillator Block

The PIC18F87K22 family of devices includes an internal oscillator block which generates two different clock signals. Either clock can be used as the microcontroller's clock source, which may eliminate the need for an external oscillator circuit on the OSC1 and/or OSC2 pins.

The internal oscillator consists of three blocks, depending on the frequency of operation. They are HF-INTOSC, MF-INTOSC and LF-INTRC.

In HF-INTOSC mode, the internal oscillator can provide a frequency ranging from 31 kHz to 16 MHz, with the postscaler deciding the selected frequency (IRCF<2:0>).

The INTSRC bit (OSCTUNE<7>) and MFIOSEL bit (OSCCON2<0>) also decide which INTOSC provides the lower frequency (500 kHz to 31 KHz). For the HF-INTOSC to provide these frequencies, INTSRC = 1 and MFIOSEL = 0.

In HF-INTOSC, the postscaler (IRCF<2:0>) provides the frequency range of 31 kHz to 16 MHz. If HF-INTOSC is used with the PLL, the input frequency to the PLL should be 8 MHz or 16 MHz (IRCF<2:0> = 111 or 110).

For MF-INTOSC mode to provide a frequency range of 500 kHz to 31 kHz, INTSRC = 1 and MFIOSEL = 1. The postscaler (IRCF<2:0>), in this mode, provides the frequency range of 31 kHz to 500 kHz.

The LF-INTRC can provide only 31 kHz if INTSRC = 0.

The LF-INTRC provides 31 kHz and is enabled if it is selected as the device clock source. The mode is enabled automatically when any of the following are enabled:

- · Power-up Timer
- Fail-Safe Clock Monitor
- · Watchdog Timer
- Two-Speed Start-up

These features are discussed in greater detail in Section 28.0 "Special Features of the CPU".

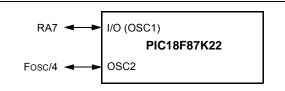
The clock source frequency (HF-INTOSC, MF-INTOSC or LF-INTRC direct) is selected by configuring the IRCF bits of the OSCCON register, as well the INTSRC and MFIOSEL bits. The default frequency on device Resets is 8 MHz.

#### 3.6.1 INTIO MODES

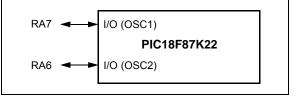
Using the internal oscillator as the clock source eliminates the need for up to two external oscillator pins, which can then be used for digital I/O. Two distinct oscillator configurations, which are determined by the FOSC Configuration bits, are available:

- In INTIO1 mode, the OSC2 pin (RA6) outputs Fosc/4, while OSC1 functions as RA7 (see Figure 3-8) for digital input and output.
- In INTIO2 mode, OSC1 functions as RA7 and OSC2 functions as RA6 (see Figure 3-9). Both are available as digital input and output ports.

#### FIGURE 3-8: INTIO1 OSCILLATOR MODE



#### FIGURE 3-9: INTIO2 OSCILLATOR MODE



#### 3.6.2 INTPLL MODES

The 4x Phase Lock Loop (PLL) can be used with the HF-INTOSC to produce faster device clock speeds than are normally possible with the internal oscillator sources. When enabled, the PLL produces a clock speed of 32 MHz or 64 MHz.

PLL operation is controlled through software. The control bit, PLLEN (OSCTUNE<6>), is used to enable or disable its operation. Additionally, the PLL will only function when the selected HF-INTOSC frequency is either 8 MHz or 16 MHz (OSCCON<6:4> = 111 or 110).

Like the INTIO modes, there are two distinct INTPLL modes available:

- In INTPLL1 mode, the OSC2 pin outputs Fosc/4, while OSC1 functions as RA7 for digital input and output. Externally, this is identical in appearance to INTIO1 (Figure 3-8).
- In INTPLL2 mode, OSC1 functions as RA7 and OSC2 functions as RA6, both for digital input and output. Externally, this is identical to INTIO2 (Figure 3-9).

#### 3.6.3 INTERNAL OSCILLATOR OUTPUT FREQUENCY AND TUNING

The internal oscillator block is calibrated at the factory to produce an INTOSC output frequency of 16 MHz. It can be adjusted in the user's application by writing to TUN<5:0> (OSCTUNE<5:0>) in the OSCTUNE register (Register 3-3).

When the OSCTUNE register is modified, the INTOSC (HF-INTOSC and MF-INTOSC) frequency will begin shifting to the new frequency. The oscillator will require some time to stabilize. Code execution continues during this shift and there is no indication that the shift has occurred.

The LF-INTOSC oscillator operates independently of the HF-INTOSC or the MF-INTOSC source. Any changes in the HF-INTOSC or the MF-INTOSC source, across voltage and temperature, are not necessarily reflected by changes in LF-INTOSC or vice versa. The frequency of LF-INTOSC is not affected by OSCTUNE.

## 3.6.4 INTOSC FREQUENCY DRIFT

The INTOSC frequency may drift as VDD or temperature changes, and can affect the controller operation in a variety of ways. It is possible to adjust the INTOSC frequency by modifying the value in the OSCTUNE register. Depending on the device, this may have no effect on the LF-INTOSC clock source frequency.

Tuning INTOSC requires knowing when to make the adjustment, in which direction it should be made, and in some cases, how large a change is needed. Three compensation techniques are shown here.

# 3.6.4.1 Compensating with the EUSART

An adjustment may be required when the EUSART begins to generate framing errors or receives data with errors while in Asynchronous mode. Framing errors indicate that the device clock frequency is too high. To adjust for this, decrement the value in OSCTUNE to reduce the clock frequency. On the other hand, errors in data may suggest that the clock speed is too low. To compensate, increment OSCTUNE to increase the clock frequency.

# 3.6.4.2 Compensating with the Timers

This technique compares device clock speed to some reference clock. Two timers may be used; one timer is clocked by the peripheral clock, while the other is clocked by a fixed reference source, such as the SOSC oscillator.

Both timers are cleared, but the timer clocked by the reference source generates interrupts. When an interrupt occurs, the internally clocked timer is read and both timers are cleared. If the internally clocked timer value is much greater than expected, then the internal oscillator block is running too fast. To adjust for this, decrement the OSCTUNE register.

# 3.6.4.3 Compensating with the CCP Module in Capture Mode

A CCP module can use free-running Timer1 (or Timer3), clocked by the internal oscillator block and an external event with a known period (i.e., AC power frequency). The time of the first event is captured in the CCPRxH:CCPRxL registers and is recorded for use later. When the second event causes a capture, the time of the first event is subtracted from the time of the second event. Since the period of the external event is known, the time difference between events can be calculated.

If the measured time is much greater than the calculated time, the internal oscillator block is running too fast. To compensate, decrement the OSCTUNE register. If the measured time is much less than the calculated time, the internal oscillator block is running too slow. To compensate, increment the OSCTUNE register.

# 3.7 Reference Clock Output

In addition to the Fosc/4 clock output, in certain oscillator modes, the device clock in the PIC18F87K22 family can also be configured to provide a reference clock output signal to a port pin. This feature is available in all oscillator configurations and allows the user to select a greater range of clock submultiples to drive external devices in the application.

This reference clock output is controlled by the REFOCON register (Register 3-4). Setting the ROON bit (REFOCON<7>) makes the clock signal available on the REFO (RE3) pin. The RODIV<3:0> bits enable the selection of 16 different clock divider options.

The ROSSLP and ROSEL bits (REFOCON<5:4>) control the availability of the reference output during Sleep mode. The ROSEL bit determines if the oscillator on OSC1 and OSC2, or the current system clock source, is used for the reference clock output. The ROSSLP bit determines if the reference source is available on RE3 when the device is in Sleep mode.

To use the reference clock output in Sleep mode, both the ROSSLP and ROSEL bits must be set. The device clock must also be configured for an EC or HS mode. If not, the oscillator on OSC1 and OSC2 will be powered down when the device enters Sleep mode. Clearing the ROSEL bit allows the reference output frequency to change as the system clock changes during any clock switches.

	U-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	
ROON	_	ROSSLP	ROSEL <sup>(1)</sup>	RODIV3	RODIV2	RODIV1	RODIV0	
bit 7							bit	
Legend:								
R = Readab	le bit	W = Writable	bit	U = Unimpler	nented bit, rea	id as '0'		
-n = Value a	It POR	'1' = Bit is set		ʻ0' = Bit is cle		x = Bit is unki	nown	
bit 7	ROON: Refe	rence Oscillato	r Output Enabl	e bit				
	1 = Referenc	e oscillator out e oscillator out	out is available					
bit 6		• • <b>ted:</b> Read as '						
bit 5	-	eference Oscilla		p in Sleep bit				
		e oscillator con	•	• •				
	0 = Referenc	e oscillator is d	isabled in Slee	ep				
	<b>ROSEL:</b> Reference Oscillator Source Select bit <sup>(1)</sup>							
bit 4	ROSEL: Refe	erence Oscillato	or Source Sele	ect bit <sup>(1)</sup>				
bit 4	1 = Primary o	oscillator (EC or	·HS) is used a	is the base clo		ck switching of	the device	
	1 = Primary c 0 = System c	oscillator (EC or lock is used as	<sup>-</sup> HS) is used a the base clock	is the base clo k; base clock r		ock switching of	the device	
bit 4 bit 3-0	1 = Primary c 0 = System c <b>RODIV&lt;3:0&gt;</b>	oscillator (EC or lock is used as : Reference Os	HS) is used a the base clock	is the base clo k; base clock r Select bits		ock switching of	the device	
	1 = Primary c 0 = System c <b>RODIV&lt;3:0&gt;</b> 1111 = Base	oscillator (EC or lock is used as	HS) is used a the base clock cillator Divisor vided by 32,768	is the base clo k; base clock r Select bits 8		ock switching of	the device	
	1 = Primary c 0 = System c <b>RODIV&lt;3:0&gt;</b> 1111 = Base 1110 = Base	oscillator (EC or lock is used as : Reference Os clock value div	HS) is used a the base clock cillator Divisor vided by 32,768 vided by 16,384	is the base clo k; base clock r Select bits 8		ock switching of	the device	
	1 = Primary c 0 = System c <b>RODIV&lt;3:0&gt;</b> 1111 = Base 1110 = Base 1101 = Base 1100 = Base	oscillator (EC or lock is used as : Reference Os clock value div clock value div clock value div clock value div	HS) is used a the base clock cillator Divisor vided by 32,768 vided by 16,384 vided by 8,192 vided by 4,096	is the base clo k; base clock r Select bits 8		ock switching of	the device	
	1 = Primary c 0 = System c RODIV<3:0> 1111 = Base 1110 = Base 1101 = Base 1100 = Base 1011 = Base	scillator (EC or lock is used as Reference Os clock value div clock value div clock value div clock value div clock value div clock value div	HS) is used a the base clock cillator Divisor rided by 32,768 rided by 16,384 rided by 8,192 rided by 4,096 rided by 2,048	is the base clo k; base clock r Select bits 8		ock switching of	the device	
	1 = Primary c 0 = System c RODIV<3:0> 1111 = Base 1110 = Base 1101 = Base 1100 = Base 1011 = Base 1011 = Base	scillator (EC or lock is used as : Reference Os clock value div clock value div clock value div clock value div clock value div clock value div clock value div	HS) is used a the base clock cillator Divisor vided by 32,768 vided by 16,384 vided by 8,192 vided by 4,096 vided by 2,048 vided by 1,024	is the base clo k; base clock r Select bits 8		ock switching of	the device	
	1 = Primary of 0 = System of RODIV<3:0> 1111 = Base 1110 = Base 1101 = Base 1010 = Base 1011 = Base 1010 = Base 1001 = Base	scillator (EC or lock is used as : Reference Os clock value div clock value div	HS) is used a the base clock cillator Divisor vided by 32,768 vided by 16,384 vided by 8,192 vided by 4,096 vided by 2,048 vided by 1,024 vided by 512	is the base clo k; base clock r Select bits 8		ock switching of	the device	
	1 = Primary of 0 = System of <b>RODIV&lt;3:0&gt;</b> 1111 = Base 1100 = Base 1001 = Base 1011 = Base 1010 = Base 1010 = Base 1001 = Base 1000 = Base	scillator (EC or lock is used as : Reference Os clock value div clock value div	HS) is used a the base clock cillator Divisor vided by 32,768 vided by 16,384 vided by 8,192 vided by 4,096 vided by 2,048 vided by 1,024 vided by 512 vided by 256	is the base clo k; base clock r Select bits 8		ock switching of	the device	
	1 = Primary of 0 = System of RODIV<3:0> 1111 = Base 1100 = Base 1001 = Base 1011 = Base 1010 = Base 1001 = Base 1001 = Base 1000 = Base 0111 = Base	scillator (EC or lock is used as : Reference Os clock value div clock value div	HS) is used a the base clock cillator Divisor vided by 32,768 vided by 16,384 vided by 8,192 vided by 4,096 vided by 2,048 vided by 1,024 vided by 512 vided by 256 vided by 128	is the base clo k; base clock r Select bits 8		ock switching of	the device	
	1 = Primary of 0 = System of RODIV<3:0> 1111 = Base 1100 = Base 1001 = Base 1011 = Base 1001 = Base 1001 = Base 1000 = Base 0111 = Base 0110 = Base	scillator (EC or lock is used as : Reference Os clock value div clock value div	HS) is used a the base clock cillator Divisor vided by 32,768 vided by 16,384 vided by 8,192 vided by 4,096 vided by 2,048 vided by 1,024 vided by 512 vided by 512 vided by 256 vided by 128 vided by 64	is the base clo k; base clock r Select bits 8		ock switching of	the device	
	1 = Primary of 0 = System of RODIV<3:0> 1111 = Base 1100 = Base 1001 = Base 1001 = Base 1001 = Base 1000 = Base 0111 = Base 0110 = Base 0110 = Base 0101 = Base 0101 = Base	scillator (EC or lock is used as : Reference Os clock value div clock value div	HS) is used a the base clock cillator Divisor vided by 32,768 vided by 16,384 vided by 8,192 vided by 4,096 vided by 2,048 vided by 1,024 vided by 512 vided by 512 vided by 256 vided by 128 vided by 64 vided by 32 vided by 16	is the base clo k; base clock r Select bits 8		ock switching of	the device	
	1 = Primary of 0 = System of RODIV<3:0> 1111 = Base 1100 = Base 1001 = Base 1001 = Base 1001 = Base 1000 = Base 0111 = Base 0111 = Base 0110 = Base 0101 = Base 0101 = Base 0101 = Base 0101 = Base	scillator (EC or lock is used as : Reference Os clock value div clock value div	HS) is used a the base clock cillator Divisor vided by 32,768 vided by 16,384 vided by 8,192 vided by 4,096 vided by 2,048 vided by 1,024 vided by 512 vided by 512 vided by 256 vided by 128 vided by 64 vided by 32 vided by 16 vided by 8	is the base clo k; base clock r Select bits 8		ock switching of	the device	
	1 = Primary of 0 = System of RODIV<3:0> 1111 = Base 1100 = Base 1001 = Base 1001 = Base 1001 = Base 0101 = Base 0111 = Base 0111 = Base 0110 = Base 0101 = Base 0101 = Base 0101 = Base 0101 = Base 0101 = Base	scillator (EC or lock is used as : Reference Os clock value div clock value div	HS) is used a the base clock cillator Divisor vided by 32,768 vided by 16,384 vided by 8,192 vided by 4,096 vided by 2,048 vided by 2,048 vided by 512 vided by 512 vided by 512 vided by 256 vided by 128 vided by 64 vided by 32 vided by 16 vided by 8 vided by 4	is the base clo k; base clock r Select bits 8		ock switching of	the device	

## REGISTER 3-4: REFOCON: REFERENCE OSCILLATOR CONTROL REGISTER

**Note 1:** For ROSEL (REVOCON<4>), the primary oscillator is available only when configured as the default via the FOSC settings. This is regardless of whether the device is in Sleep mode.

# 3.8 Effects of Power-Managed Modes on the Various Clock Sources

When PRI\_IDLE mode is selected, the designated primary oscillator continues to run without interruption. For all other power-managed modes, the oscillator using the OSC1 pin is disabled. The OSC1 pin (and OSC2 pin if used by the oscillator) will stop oscillating.

In secondary clock modes (SEC\_RUN and SEC\_IDLE), the SOSC oscillator is operating and providing the device clock. The SOSC oscillator may also run in all power-managed modes if required to clock SOSC.

In RC\_RUN and RC\_IDLE modes, the internal oscillator provides the device clock source. The 31 kHz LF-INTOSC output can be used directly to provide the clock and may be enabled to support various special features, regardless of the power-managed mode (see Section 28.2 "Watchdog Timer (WDT)" through Section 28.5 "Fail-Safe Clock Monitor" for more information on WDT, Fail-Safe Clock Monitor and Two-Speed Start-up).

If Sleep mode is selected, all clock sources are stopped. Since all the transistor switching currents have been stopped, Sleep mode achieves the lowest current consumption of the device (only leakage currents).

Enabling any on-chip feature that will operate during Sleep will increase the current consumed during Sleep. The INTOSC is required to support WDT operation. The SOSC oscillator may be operating to support a Real-Time Clock (RTC). Other features may be operating that do not require a device clock source (i.e., MSSP slave, INTx pins and others). Peripherals that may add significant current consumption are listed in Section 31.2 "DC Characteristics: Power-Down and Supply Current PIC18F87K22 Family (Industrial/ Extended)".

# 3.9 Power-up Delays

Power-up delays are controlled by two timers, so that no external Reset circuitry is required for most applications. The delays ensure that the device is kept in Reset until the device power supply is stable under normal circumstances and the primary clock is operating and stable. For additional information on power-up delays, see **Section 5.6 "Power-up Timer (PWRT)**".

The first timer is the Power-up Timer (PWRT), which provides a fixed delay on a power-up time of about 64 ms (Parameter 33, Table 31-13); it is always enabled.

The second timer is the Oscillator Start-up Timer (OST), intended to keep the chip in Reset until the crystal oscillator is stable (HS, XT or LP modes). The OST does this by counting 1,024 oscillator cycles before allowing the oscillator to clock the device.

There is a delay of interval, TCSD (Parameter 38, Table 31-13), following POR, while the controller becomes ready to execute instructions.

TABLE 3-4: C	DSC1 AND OSC2 PIN STATES IN SLEEP MODE
--------------	--

Oscillator Mode	OSC1 Pin	OSC2 Pin
EC, ECPLL	Floating, pulled by external clock	At logic low (clock/4 output)
HS, HSPLL	Feedback inverter is disabled at quiescent voltage level	Feedback inverter is disabled at quiescent voltage level
INTOSC, INTPLL1/2	I/O pin, RA6, direction is controlled by TRISA<6>	I/O pin, RA6, direction is controlled by TRISA<7>

Note: See Section 5.0 "Reset" for time-outs due to Sleep and MCLR Reset.

NOTES:

# 4.0 POWER-MANAGED MODES

The PIC18F87K22 family of devices offers a total of seven operating modes for more efficient power management. These modes provide a variety of options for selective power conservation in applications where resources may be limited (such as battery-powered devices).

There are three categories of power-managed mode:

- Run modes
- Idle modes
- Sleep mode

There is an Ultra Low-Power Wake-up (ULPWU) for waking from the Sleep mode.

These categories define which portions of the device are clocked, and sometimes, at what speed. The Run and Idle modes may use any of the three available clock sources (primary, secondary or internal oscillator block). The Sleep mode does not use a clock source.

The ULPWU mode, on the RA0 pin, enables a slow falling voltage to generate a wake-up, even from Sleep, without excess current consumption. (See Section 4.7 "Ultra Low-Power Wake-up".)

The power-managed modes include several powersaving features offered on previous PIC<sup>®</sup> devices. One is the clock switching feature, offered in other PIC18 devices. This feature allows the controller to use the SOSC oscillator instead of the primary one. Another power-saving feature is Sleep mode, offered by all PIC devices, where all device clocks are stopped.

# 4.1 Selecting Power-Managed Modes

Selecting a power-managed mode requires two decisions:

- · Will the CPU be clocked or not
- · What will be the clock source

The IDLEN bit (OSCCON<7>) controls CPU clocking, while the SCS<1:0> bits (OSCCON<1:0>) select the clock source. The individual modes, bit settings, clock sources and affected modules are summarized in Table 4-1.

### 4.1.1 CLOCK SOURCES

The SCS<1:0> bits select one of three clock sources for power-managed modes. Those sources are:

- The primary clock as defined by the FOSC<3:0> Configuration bits
- The secondary clock (the SOSC oscillator)
- The internal oscillator block (for LF-INTOSC modes)

#### 4.1.2 ENTERING POWER-MANAGED MODES

Switching from one power-managed mode to another begins by loading the OSCCON register. The SCS<1:0> bits select the clock source and determine which Run or Idle mode is used. Changing these bits causes an immediate switch to the new clock source, assuming that it is running. The switch may also be subject to clock transition delays. These considerations are discussed in Section 4.1.3 "Clock Transitions and Status Indicators" and subsequent sections.

Entering the power-managed Idle or Sleep modes is triggered by the execution of a SLEEP instruction. The actual mode that results depends on the status of the IDLEN bit.

Depending on the current and impending mode, a change to a power-managed mode does not always require setting all of the previously discussed bits. Many transitions can be done by changing the oscillator select bits, or changing the IDLEN bit, prior to issuing a SLEEP instruction. If the IDLEN bit is already configured as desired, it may only be necessary to perform a SLEEP instruction to switch to the desired mode.

Mode	OSCCON Bits		Module Clocking		Available Clock and Oscillator Source	
Mode	IDLEN<7> <sup>(1)</sup>	SCS<1:0>	CPU	Peripherals	Available Clock and Oscillator Source	
Sleep	0	N/A	Off	Off	None – All clocks are disabled	
PRI_RUN	N/A	00	Clocked	Clocked	Primary – XT, LP, HS, EC, RC and PLL modes. This is the normal, Full-Power Execution mode.	
SEC_RUN	N/A	01	Clocked	Clocked	Secondary – SOSC Oscillator	
RC_RUN	N/A	1x	Clocked	Clocked	Internal oscillator block <sup>(2)</sup>	
PRI_IDLE	1	00	Off	Clocked	Primary – LP, XT, HS, RC, EC	
SEC_IDLE	1	01	Off	Clocked	Secondary – SOSC oscillator	
RC_IDLE	1	lx	Off	Clocked	Internal oscillator block <sup>(2)</sup>	

TABLE 4-1:POWER-MANAGED MODES

Note 1: IDLEN reflects its value when the SLEEP instruction is executed.

<sup>2:</sup> Includes INTOSC (HF-INTOSC and MG-INTOSC) and INTOSC postscaler, as well as the LF-INTOSC source.

# 4.1.3 CLOCK TRANSITIONS AND STATUS INDICATORS

The length of the transition between clock sources is the sum of two cycles of the old clock source and three to four cycles of the new clock source. This formula assumes that the new clock source is stable. The HF-INTOSC and MF-INTOSC are termed as INTOSC in this chapter.

Three bits indicate the current clock source and its status, as shown in Table 4-2. The three bits are:

- OSTS (OSCCON<3>)
- HFIOFS (OSCCON<2>)
- SOSCRUN (OSCCON2<6>)

TABLE 4-2:	SYSTEM CLOCK INDICATOR
------------	------------------------

Main Clock Source	osts	HFIOFSor MFIOFS	SOSCRUN
Primary Oscillator	1	0	0
INTOSC (HF-INTOSC or MF-INTOSC)	0	1	0
Secondary Oscillator	0	0	1
MF-INTOSC or HF-INTOSC as Primary Clock Source	1	1	0
LF-INTOSC is Running or INTOSC is Not Yet Stable	0	0	0

When the OSTS bit is set, the primary clock is providing the device clock. When the HFIOFS or MFIOFS bit is set, the INTOSC output is providing a stable 16 MHz clock source to a divider that actually drives the device clock. When the SOSCRUN bit is set, the SOSC oscillator is providing the clock. If none of these bits are set, either the LF-INTOSC clock source is clocking the device or the INTOSC source is not yet stable.

If the internal oscillator block is configured as the primary clock source by the FOSC<3:0> Configuration bits (CONFIG1H<3:0>), then the OSTS and HFIOFS or MFIOFS bits can be set when in PRI\_RUN or PRI\_IDLE modes. This indicates that the primary clock (INTOSC output) is generating a stable 16 MHz output. Entering another INTOSC power-managed mode at the same frequency would clear the OSTS bit.

- Note 1: Caution should be used when modifying a single IRCF bit. At a lower VDD, it is possible to select a higher clock speed than is supportable by that VDD. Improper device operation may result if the VDD/ Fosc specifications are violated.
  - 2: Executing a SLEEP instruction does not necessarily place the device into Sleep mode. It acts as the trigger to place the controller into either the Sleep mode or one of the Idle modes, depending on the setting of the IDLEN bit.

#### 4.1.4 MULTIPLE SLEEP COMMANDS

The power-managed mode that is invoked with the SLEEP instruction is determined by the setting of the IDLEN bit at the time the instruction is executed. If another SLEEP instruction is executed, the device will enter the power-managed mode specified by IDLEN at that time. If IDLEN has changed, the device will enter the new power-managed mode specified by the new setting.

## 4.2 Run Modes

In the Run modes, clocks to both the core and peripherals are active. The difference between these modes is the clock source.

### 4.2.1 PRI\_RUN MODE

The PRI\_RUN mode is the normal, Full-Power Execution mode of the microcontroller. This is also the default mode upon a device Reset, unless Two-Speed Start-up is enabled. (For details, see Section 28.4 "Two-Speed Start-up".) In this mode, the OSTS bit is set. The HFIOFS or MFIOFS bit may be set if the internal oscillator block is the primary clock source. (See Section 3.2 "Control Registers".)

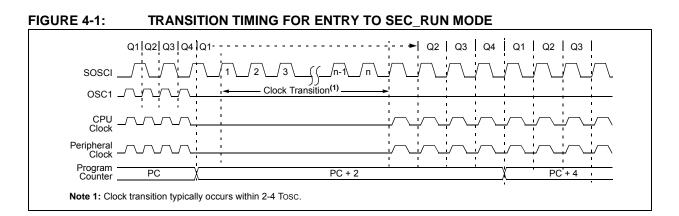
### 4.2.2 SEC\_RUN MODE

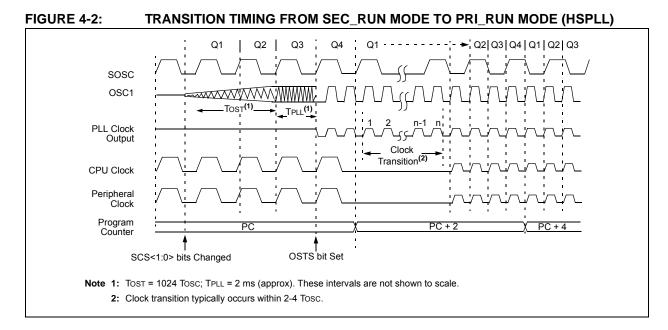
The SEC\_RUN mode is the compatible mode to the "clock-switching" feature offered in other PIC18 devices. In this mode, the CPU and peripherals are clocked from the SOSC oscillator. This enables lower power consumption while retaining a high-accuracy clock source.

SEC\_RUN mode is entered by setting the SCS<1:0> bits to '01'. The device clock source is switched to the SOSC oscillator (see Figure 4-1), the primary oscillator is shut down, the SOSCRUN bit (OSCCON2<6>) is set and the OSTS bit is cleared.

Note:	The SOSC oscillator can be enabled by setting the SOSCGO bit (OSCCON2<3>). If this bit is set, the clock switch to the SEC_RUN mode can switch immediately
	once SCS<1:0> are set to '01'.

On transitions from SEC\_RUN mode to PRI\_RUN mode, the peripherals and CPU continue to be clocked from the SOSC oscillator while the primary clock is started. When the primary clock becomes ready, a clock switch back to the primary clock occurs (see Figure 4-2). When the clock switch is complete, the SOSCRUN bit is cleared, the OSTS bit is set and the primary clock is providing the clock. The IDLEN and SCS bits are not affected by the wake-up and the SOSC oscillator continues to run.





# 4.2.3 RC\_RUN MODE

In RC\_RUN mode, the CPU and peripherals are clocked from the internal oscillator block using the INTOSC multiplexer. In this mode, the primary clock is shut down. When using the LF-INTOSC source, this mode provides the best power conservation of all the Run modes, while still executing code. It works well for user applications which are not highly timing-sensitive or do not require high-speed clocks at all times.

If the primary clock source is the internal oscillator block – either LF-INTOSC or INTOSC (MF-INTOSC or HF-INTOSC) – there are no distinguishable differences between the PRI\_RUN and RC\_RUN modes during execution. Entering or exiting RC\_RUN mode, however, causes a clock switch delay. Therefore, if the primary clock source is the internal oscillator block, using RC\_RUN mode is not recommended.

This mode is entered by setting the SCS1 bit to '1'. To maintain software compatibility with future devices, it is recommended that the SCS0 bit also be cleared, even though the bit is ignored. When the clock source is switched to the INTOSC multiplexer (see Figure 4-3),

the primary oscillator is shut down and the OSTS bit is cleared. The IRCF bits may be modified at any time to immediately change the clock speed.

Note:	Caution should be used when modifying a					
	single IRCF bit. At a lower VDD, it is					
	possible to select a higher clock speed					
	than is supportable by that VDD. Improper					
	device operation may result if the VDD/					
	Fosc specifications are violated.					

If the IRCF bits and the INTSRC bit are all clear, the INTOSC output (HF-INTOSC/MF-INTOSC) is not enabled and the HFIOFS and MFIOFS bits will remain clear. There will be no indication of the current clock source. The LF-INTOSC source is providing the device clocks.

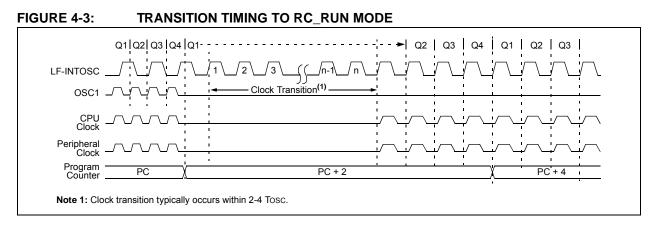
If the IRCF bits are changed from all clear (thus, enabling the INTOSC output), or if INTSRC or MFIOSEL is set, the HFIOFS or MFIOFS bit is set after the INTOSC output becomes stable. For details, see Table 4-3.

TABLE 4-3:	INTERNAL OSCILLATOR FREQUENCY STABILITY BITS

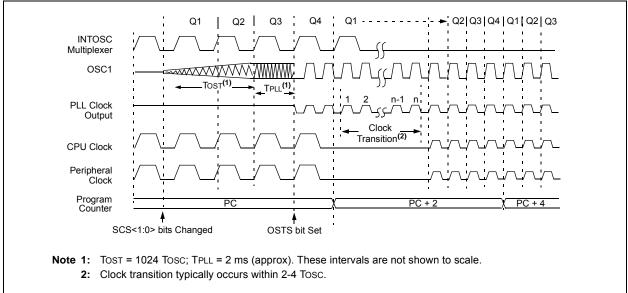
IRC	F<2:0>	INTSRC	MFIOSEL	Status of MFIOFS or HFIOFS when INTOSC is Stable	
(	000	0	x	MFIOFS = 0, HFIOFS = 0 and clock source is LF-INTOSC	
C	000	1	0	MFIOFS = 0, HFIOFS = 1 and clock source is HF-INTOSC	
C	000	1	1	MFIOFS = 1, HFIOFS = 0 and clock source is MF-INTOSC	
Nor	n-Zero	x	0	MFIOFS = 0, HFIOFS = 1 and clock source is HF-INTOSC	
Nor	n-Zero	x	1	MFIOFS = 1, HFIOFS = 0 and clock source is MF-INTOSC	

Clocks to the device continue while the INTOSC source stabilizes after an interval of TIOBST (Parameter 39, Table 31-13).

If the IRCF bits were previously at a non-zero value, or if INTSRC was set before setting SCS1 and the INTOSC source was already stable, the HFIOFS or MFIOFS bit will remain set. On transitions from RC\_RUN mode to PRI\_RUN mode, the device continues to be clocked from the INTOSC multiplexer while the primary clock is started. When the primary clock becomes ready, a clock switch to the primary clock occurs (see Figure 4-4). When the clock switch is complete, the HFIOFS or MFIOFS bit is cleared, the OSTS bit is set and the primary clock is providing the device clock. The IDLEN and SCS bits are not affected by the switch. The LF-INTOSC source will continue to run if either the WDT or the Fail-Safe Clock Monitor is enabled.







# 4.3 Sleep Mode

The power-managed Sleep mode in the PIC18F87K22 family of devices is identical to the legacy Sleep mode offered in all other PIC devices. It is entered by clearing the IDLEN bit (the default state on device Reset) and executing the SLEEP instruction. This shuts down the selected oscillator (Figure 4-5). All clock source status bits are cleared.

Entering Sleep mode from any other mode does not require a clock switch. This is because no clocks are needed once the controller has entered Sleep. If the WDT is selected, the LF-INTOSC source will continue to operate. If the SOSC oscillator is enabled, it will also continue to run.

When a wake event occurs in Sleep mode (by interrupt, Reset or WDT time-out), the device will not be clocked until the clock source selected by the SCS<1:0> bits becomes ready (see Figure 4-6). Alternately, the device will be clocked from the internal oscillator block if either the Two-Speed Start-up or the Fail-Safe Clock Monitor is enabled (see Section 28.0 "Special Features of the CPU"). In either case, the OSTS bit is set when the primary clock is providing the device clocks. The IDLEN and SCS bits are not affected by the wake-up.

## 4.4 Idle Modes

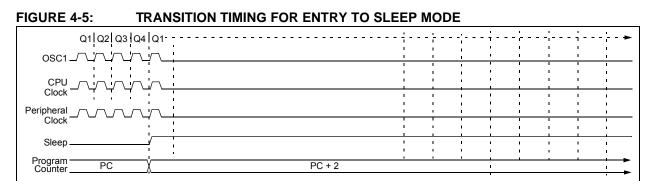
The Idle modes allow the controller's CPU to be selectively shut down while the peripherals continue to operate. Selecting a particular Idle mode allows users to further manage power consumption.

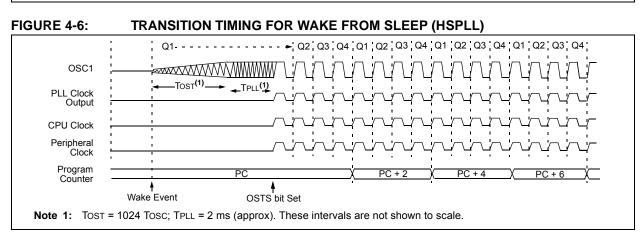
If the IDLEN bit is set to a '1' when a SLEEP instruction is executed, the peripherals will be clocked from the clock source selected using the SCS<1:0> bits. The CPU, however, will not be clocked. The clock source status bits are not affected. This approach is a quick method to switch from a given Run mode to its corresponding Idle mode.

If the WDT is selected, the LF-INTOSC source will continue to operate. If the SOSC oscillator is enabled, it will also continue to run.

Since the CPU is not executing instructions, the only exits from any of the Idle modes are by interrupt, WDT time-out or a Reset. When a wake event occurs, CPU execution is delayed by an interval of TCSD (Parameter 38, Table 31-13) while it becomes ready to execute code. When the CPU begins executing code, it resumes with the same clock source for the current Idle mode. For example, when waking from RC\_IDLE mode, the internal oscillator block will clock the CPU and peripherals (in other words, RC\_RUN mode). The IDLEN and SCS bits are not affected by the wake-up.

While in any Idle mode or Sleep mode, a WDT timeout will result in a WDT wake-up to the Run mode currently specified by the SCS<1:0> bits.





#### 4.4.1 PRI\_IDLE MODE

This mode is unique among the three low-power Idle modes, in that it does not disable the primary device clock. For timing-sensitive applications, this allows for the fastest resumption of device operation with its more accurate, primary clock source, since the clock source does not have to "warm-up" or transition from another oscillator.

PRI\_IDLE mode is entered from PRI\_RUN mode by setting the IDLEN bit and executing a SLEEP instruction. If the device is in another Run mode, set IDLEN first, then clear the SCS bits and execute SLEEP. Although the CPU is disabled, the peripherals continue to be clocked from the primary clock source specified by the FOSC<3:0> Configuration bits. The OSTS bit remains set (see Figure 4-7).

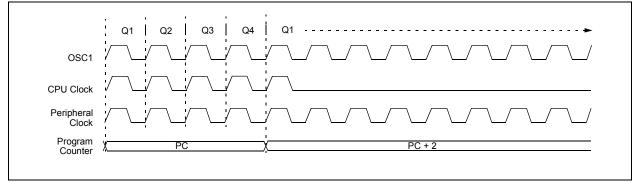
When a wake event occurs, the CPU is clocked from the primary clock source. A delay of interval, TCSD (Parameter 39, Table 31-13), is required between the wake event and the start of code execution. This is required to allow the CPU to become ready to execute instructions. After the wake-up, the OSTS bit remains set. The IDLEN and SCS bits are not affected by the wake-up (see Figure 4-8).

## 4.4.2 SEC\_IDLE MODE

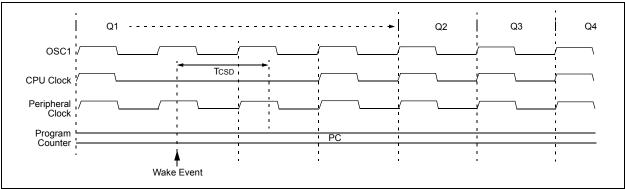
In SEC\_IDLE mode, the CPU is disabled but the peripherals continue to be clocked from the SOSC oscillator. This mode is entered from SEC\_RUN by setting the IDLEN bit and executing a SLEEP instruction. If the device is in another Run mode, set the IDLEN bit first, then set the SCS<1:0> bits to '01' and execute SLEEP. When the clock source is switched to the SOSC oscillator, the primary oscillator is shut down, the OSTS bit is cleared and the SOSCRUN bit is set.

When a wake event occurs, the peripherals continue to be clocked from the SOSC oscillator. After an interval of TCSD following the wake event, the CPU begins executing code being clocked by the SOSC oscillator. The IDLEN and SCS bits are not affected by the wakeup and the SOSC oscillator continues to run (see Figure 4-8).

#### FIGURE 4-7: TRANSITION TIMING FOR ENTRY TO IDLE MODE



### FIGURE 4-8: TRANSITION TIMING FOR WAKE FROM IDLE TO RUN MODE



# 4.4.3 RC\_IDLE MODE

In RC\_IDLE mode, the CPU is disabled but the peripherals continue to be clocked from the internal oscillator block using the INTOSC multiplexer. This mode provides controllable power conservation during Idle periods.

From RC\_RUN, this mode is entered by setting the IDLEN bit and executing a SLEEP instruction. If the device is in another Run mode, first set IDLEN, then set the SCS1 bit and execute SLEEP. To maintain software compatibility with future devices, it is recommended that SCS0 also be cleared, though its value is ignored. The INTOSC multiplexer may be used to select a higher clock frequency by modifying the IRCF bits before executing the SLEEP instruction. When the clock source is switched to the INTOSC multiplexer, the primary oscillator is shut down and the OSTS bit is cleared.

If the IRCF bits are set to any non-zero value, or the INTSRC/MFIOSEL bit is set, the INTOSC output is enabled. The HFIOFS/MFIOFS bits become set, after the INTOSC output becomes stable, after an interval of TIOBST (Parameter 38, Table 31-13). (For information on the HFIOFS/MFIOFS bits, see Table 4-3.)

Clocks to the peripherals continue while the INTOSC source stabilizes. The HFIOFS/MFIOFS bits will remain set if the IRCF bits were previously at a non-zero value or if INTSRC was set before the SLEEP instruction was executed and the INTOSC source was already stable. If the IRCF bits and INTSRC are all clear, the INTOSC output will not be enabled, the HFIOFS/MFIOFS bits will remain clear and there will be no indication of the current clock source.

When a wake event occurs, the peripherals continue to be clocked from the INTOSC multiplexer. After a delay of TCSD (Parameter 38, Table 31-13) following the wake event, the CPU begins executing code clocked by the INTOSC multiplexer. The IDLEN and SCS bits are not affected by the wake-up. The INTRC source will continue to run if either the WDT or the Fail-Safe Clock Monitor is enabled.

## 4.5 Selective Peripheral Module Control

Idle mode allows users to substantially reduce power consumption by stopping the CPU clock. Even so, peripheral modules still remain clocked, and thus, consume power. There may be cases where the application needs what this mode does not provide: the allocation of power resources to the CPU processing with minimal power consumption from the peripherals.

PIC18F87K22 family devices address this requirement by allowing peripheral modules to be selectively disabled, reducing or eliminating their power consumption. This can be done with two control bits:

- Peripheral Enable bit, generically named XXXEN Located in the respective module's main control register
- Peripheral Module Disable (PMD) bit, generically named, XXXMD – Located in one of the PMDx Control registers (PMD0, PMD1, PMD2 or PMD3)

Disabling a module by clearing its XXXEN bit disables the module's functionality, but leaves its registers available to be read and written to. This reduces power consumption, but not by as much as the second approach.

Most peripheral modules have an enable bit.

In contrast, setting the PMD bit for a module disables all clock sources to that module, reducing its power consumption to an absolute minimum. In this state, the control and status registers associated with the peripheral are also disabled, so writes to those registers have no effect and read values are invalid. Many peripheral modules have a corresponding PMD bit.

There are four PMD registers in the PIC18F87K22 family devices: PMD0, PMD1, PMD2 and PMD3. These registers have bits associated with each module for disabling or enabling a particular peripheral.

### REGISTER 4-1: PMD3: PERIPHERAL MODULE DISABLE REGISTER 3

R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0
CCP10MD <sup>(1)</sup>	CCP9MD <sup>(1)</sup>	CCP8MD	CCP7MD	CCP6MD	CCP5MD	CCP4MD	TMR12MD <sup>(1)</sup>
bit 7	•	•	•			•	bit 0
Legend:							
R = Readable	bit	W = Writable	bit	U = Unimpler	nented bit, read	as '0'	
-n = Value at P	OR	'1' = Bit is set		'0' = Bit is cle	ared	x = Bit is unk	nown
bit 7	CCP10MD: P	MD CCP10 Er	able/Disable I	bit <sup>(1)</sup>			
				enabled for CC	P10, disabling a	all of its clock s	sources
		lisabled for CC		(1)			
bit 6		ID CCP9 Enab			P9, disabling al	of its slock or	
		lisabled for CC			r9, uisabiling al		Juices
bit 5		ID CCP8 Enab					
	1 = Periphera	al Module Disa	ble (PMD) is e	enabled for CC	P8, disabling al	of its clock so	ources
	0 = PMD is d	isabled for CC	P8				
bit 4		ID CCP7 Enab					
		al Module Disa lisabled for CC		enabled for CC	P7, disabling al	of its clock so	ources
bit 3		ID CCP6 Enab					
DIL 3				anabled for CC	P6, disabling al	of its clock se	
	•	lisabled for CC			r o, disabiling al		Juices
bit 2	CCP5MD: PN	ID CCP5 Enab	le/Disable bit				
				enabled for CC	P5, disabling al	of its clock so	ources
		lisabled for CC	-				
bit 1		ID CCP4 Enab					
		al Module Disa lisabled for CC		enabled for CC	P4, disabling al	of its clock so	ources
bit 0	<b>TMR12MD:</b> ⊤	MR12MD Disa	ble bit <sup>(1)</sup>				
		nabled and all			e disabled		
	0 = PMD is d	lisabled and TN	/IR12MD is en	abled			

Note 1: Unimplemented on devices with a program memory of 32 Kbytes (PIC18FX5K22).

R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0		
TMR10MD <sup>(1)</sup>	TMR8MD	TMR7MD <sup>(1)</sup>	TMR6MD	TMR5MD	CMP3MD	CMP2MD	CMP1MD		
bit 7							bit (		
Legend:			.,						
R = Readable bit		W = Writable bit		U = Unimplemented bit, read					
-n = Value at POR		'1' = Bit is set	Bit is set '0' = Bit is cleared		x = Bit is unknown				
bit 7	TMR10MD: TMR10MD Disable bit <sup>(1)</sup>								
	1 = Peripheral Module Disable (PMD) is enabled and all TMR10MD clock sources are disabled								
	0 = PMD is disabled and TMR10MD is enabled								
bit 6	TMR8MD: TMR8MD Disable bit								
	1 = PMD is enabled and all TMR8MD clock sources are disabled								
	0 = PMD is disabled and TMR8MD is enabled								
bit 5	TMR7MD: TMR7MD Disable bit <sup>(1)</sup>								
	1 = PMD is enabled and all TMR7MD clock sources are disabled								
	0 = PMD is disabled and TMR7MD is enabled								
bit 4	TMR6MD: TMR6MD Disable bit								
	<ul> <li>1 = PMD is enabled and all TMR6MD clock sources are disabled</li> <li>0 = PMD is disabled and TMR6MD is enabled</li> </ul>								
bit 3	TMR5MD: TMR5MD Disable bit								
bit 5	1 = PMD is enabled and all TMR5MD clock sources are disabled								
	0 = PMD is disabled and TMR5MD is enabled								
bit 2	CMP3MD: PMD Comparator 3 Enable/Disable bit								
	1 = PMD is enabled for Comparator 3, disabling all of its clock sources								
	0 = PMD is disabled for Comparator 3								
bit 1	CMP2MD: PMD Comparator 3 Enable/Disable bit								
	1 = PMD is enabled for Comparator 2, disabling all of its clock sources								
	0 = PMD is	disabled for Com	parator 2						
bit 0	CMP1MD: PMD Comparator 3 Enable/Disable bit								
	1 = PMD is enabled for Comparator 1, disabling all of its clock sources								
	0 = PMD is	disabled for Com	parator 1						
Note 1. Uni	mnlemented o	n devices with a	nrogram me	mory of 32 Kby	tos (DIC18EVE	K22)			

#### REGISTER 4-2: PMD2: PERIPHERAL MODULE DISABLE REGISTER 2

Note 1: Unimplemented on devices with a program memory of 32 Kbytes (PIC18FX5K22).

# REGISTER 4-3: PMD1: PERIPHERAL MODULE DISABLE REGISTER 1

R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0		
PSPMD	CTMUMD	RTCCMD <sup>(1)</sup>	TMR4MD	TMR3MD	TMR2MD	TMR1MD	EMBMD		
bit 7							bit (		
Legend:									
R = Readable		W = Writable bit		U = Unimplemented bit, read as '0'					
-n = Value at POR		'1' = Bit is set		'0' = Bit is cleared		x = Bit is unknown			
L:+ 7									
bit 7	<b>PSPMD:</b> Peripheral Module Disable (PMD) PSP Enable/Disable bit								
	<ul> <li>1 = PMD is enabled for PSP, disabling all of its clock sources</li> <li>0 = PMD is disabled for PSP</li> </ul>								
bit 6	CTMUMD: P	CTMUMD: PMD CTMU Enable/Disable bit							
	1 = PMD is enabled for CTMU, disabling all of its clock sources								
	0 = PMD is disabled for CTMU								
bit 5	RTCCMD: PI	MD RTCC Enab	le/Disable bit	(1)					
	1 = PMD is enabled for RTCC, disabling all of its clock sources								
	0 = PMD is disabled for RTCC								
bit 4	TMR4MD: TMR4MD Disable bit								
	<ul> <li>1 = PMD is enabled and all TMR4MD clock sources are disabled</li> <li>0 = PMD is disabled and TMR4MD is enabled</li> </ul>								
				ibied					
bit 3	TMR3MD: TMR3MD Disable bit								
	<ul> <li>1 = PMD is enabled and all TMR3MD clock sources are disabled</li> <li>0 = PMD is disabled and TMR3MD is enabled</li> </ul>								
bit 2		MR2MD Disable		ibied					
511 2	1 = PMD is enabled and all TMR2MD clock sources are disabled								
	0 = PMD is disabled and TMR2MD is enabled								
bit 1	TMR1MD: T	MR1MD Disable	bit						
	1 = PMD is enabled and all TMR1MD clock sources are disabled								
	0 = PMD is disabled and TMR1MD is enabled								
bit 0		D EMB Enable/							
		nabled for EMB		of its clock sou	irces				
	0 = PMD is d	isabled for EME							

**Note 1:** RICCMD can only be set to '1' after an EECON2 unlock sequence. Refer to Section 18.0 "Real-Clock and Calendar (RTCC)" for the unlock sequence (Example 18-1).

R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0		
CCP3MD	CCP2MD	CCP1MD	UART2MD	UART1MD	SSP2MD	SSP1MD	ADCMD		
bit 7							bit 0		
Legend:									
R = Readable		W = Writable		U = Unimplemented bit, read as '0'					
-n = Value at POR		'1' = Bit is set		'0' = Bit is cleared		x = Bit is unknown			
bit 7			ahle/Disable hi	i+					
Dit 7		ID ECCP3 Enable/Disable bit Module Disable (PMD) is enabled for ECCP3, disabling all of its clock sources							
	<ul> <li>1 = Peripheral Module Disable (PMD) is enabled for ECCP3, disabling all of its clock sources</li> <li>0 = PMD is disabled for ECCP3</li> </ul>								
bit 6	CCP2MD: PM	ID ECCP2 Ena	able/Disable bi	it					
		nabled for EC	•	all of its clock	sources				
	0 = PMD is d	isabled for EC	CP2						
bit 5		ID ECCP1 Ena		-					
		nabled for EC		all of its clock	sources				
bit 4		MD UART2 EI		hit					
Dit 4		nabled for UA			SOURCES				
		lisabled for UA			3001003				
bit 3	UART1MD: PMD UART1 Enable/Disable bit								
	1 = PMD is enabled for UART1, disabling all of its clock sources								
		isabled for UA							
bit 2 SSP2MD: PMD MSSP2 Enable/Disable bit									
		nabled for MS		all of its clock	sources				
bit 1	· · · · · · · ·	1D MSSP1 Ena		+					
		nabled for MS			SOURCES				
		lisabled for MS			3001053				
bit 0	ADCMD: PMI	D Analog/Digita	al Converter P	MD Enable/Dis	able bit				
		•••			abling all of its	clock sources			
	0 = PMD is d	lisabled for the	Analog/Digita	I Converter	-				

### REGISTER 4-4: PMD0: PERIPHERAL MODULE DISABLE REGISTER 0

## 4.6 Exiting Idle and Sleep Modes

An exit from Sleep mode or any of the Idle modes is triggered by an interrupt, a Reset or a WDT time-out. This section discusses the triggers that cause exits from power-managed modes. The clocking subsystem actions are discussed in each of the power-managed modes (see Section 4.2 "Run Modes", Section 4.3 "Sleep Mode" and Section 4.4 "Idle Modes").

#### 4.6.1 EXIT BY INTERRUPT

Any of the available interrupt sources can cause the device to exit from an Idle mode or Sleep mode to a Run mode. To enable this functionality, an interrupt source must be enabled by setting its enable bit in one of the INTCONx or PIEx registers. The exit sequence is initiated when the corresponding interrupt flag bit is set.

On all exits from Idle or Sleep modes by interrupt, code execution branches to the interrupt vector if the GIE/ GIEH bit (INTCON<7>) is set. Otherwise, code execution continues or resumes without branching (see **Section 11.0 "Interrupts"**).

### 4.6.2 EXIT BY WDT TIME-OUT

A WDT time-out will cause different actions depending on which power-managed mode the device is in when the time-out occurs.

If the device is not executing code (all Idle modes and Sleep mode), the time-out will result in an exit from the power-managed mode (see Section 4.2 "Run Modes" and Section 4.3 "Sleep Mode"). If the device is executing code (all Run modes), the time-out will result in a WDT Reset (see Section 28.2 "Watchdog Timer (WDT)").

Executing a SLEEP or CLRWDT instruction clears the WDT timer and postscaler, loses the currently selected clock source (if the Fail-Safe Clock Monitor is enabled) and modifies the IRCF bits in the OSCCON register (if the internal oscillator block is the device clock source).

### 4.6.3 EXIT BY RESET

Normally, the device is held in Reset by the Oscillator Start-up Timer (OST) until the primary clock becomes ready. At that time, the OSTS bit is set and the device begins executing code. If the internal oscillator block is the new clock source, the HFIOFS/MFIOFS bits are set instead.

The exit delay time from Reset to the start of code execution depends on both the clock sources before and after the wake-up, and the type of oscillator if the new clock source is the primary clock. Exit delays are summarized in Table 4-4.

Code execution can begin before the primary clock becomes ready. If either the Two-Speed Start-up (see Section 28.4 "Two-Speed Start-up") or Fail-Safe Clock Monitor (see Section 28.5 "Fail-Safe Clock Monitor") is enabled, the device may begin execution as soon as the Reset source has cleared. Execution is clocked by the INTOSC multiplexer driven by the internal oscillator block. Execution is clocked by the internal oscillator block until either the primary clock becomes ready or a power-managed mode is entered before the primary clock becomes ready; the primary clock is then shut down.

#### 4.6.4 EXIT WITHOUT AN OSCILLATOR START-UP DELAY

Certain exits from power-managed modes do not invoke the OST at all. The two cases are:

- When in PRI\_IDLE mode, where the primary clock source is not stopped
- When the primary clock source is not any of the LP, XT, HS or HSPLL modes

In these instances, the primary clock source either does not require an oscillator start-up delay, since it is already running (PRI\_IDLE), or normally does not require an oscillator start-up delay (RC, EC and INTIO Oscillator modes). However, a fixed delay of interval, TCSD, following the wake event, is still required when leaving Sleep and Idle modes to allow the CPU to prepare for execution. Instruction execution resumes on the first clock cycle following this delay.

# 4.7 Ultra Low-Power Wake-up

The Ultra Low-Power Wake-up (ULPWU) on pin, RA0, allows a slow falling voltage to generate an interrupt without excess current consumption.

To use this feature:

- 1. Charge the capacitor on RA0 by configuring the RA0 pin to an output and setting it to '1'.
- 2. Stop charging the capacitor by configuring RA0 as an input.
- 3. Discharge the capacitor by setting the ULPEN and ULPSINK bits in the WDTCON register.
- 4. Configure Sleep mode.
- 5. Enter Sleep mode.

When the voltage on RA0 drops below VIL, the device wakes up and executes the next instruction.

This feature provides a low-power technique for periodically waking up the device from Sleep mode.

The time-out is dependent on the discharge time of the RC circuit on RA0.

When the ULPWU module wakes the device from Sleep mode, the ULPLVL bit (WDTCON<5>) is set. Software can check this bit upon wake-up to determine the wake-up source.

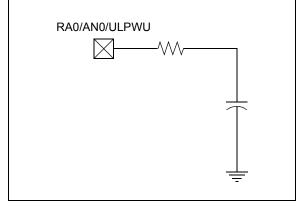
See Example 4-1 for initializing the ULPWU module.

#### EXAMPLE 4-1: ULTRA LOW-POWER WAKE-UP INITIALIZATION

```
//Charge the capacitor on RAO
       TRISAbits.TRISA0 = 0;
PORTAbits.RA0 = 1;
for(i = 0; i < 10000; i++) Nop();</pre>
       //Stop Charging the capacitor
       //on RAO
       TRISAbits.TRISA0 = 1;
       //Enable the Ultra Low Power
       //Wakeup module and allow
       //capacitor discharge
       WDTCONbits ULPEN = 1;
WDTCONbits.ULPSINK = 1;
       //For Sleep
OSCCONDits.IDLEN = 0;
       //Enter Sleep Mode
       11
Sleep();
       //for sleep, execution will
       //resume here
```

A series resistor, between RA0 and the external capacitor, provides overcurrent protection for the RA0/AN0/ ULPWU pin and enables software calibration of the time-out (see Figure 4-9).

#### FIGURE 4-9: ULTRA LOW-POWER WAKE-UP INITIALIZATION



A timer can be used to measure the charge time and discharge time of the capacitor. The charge time can then be adjusted to provide the desired delay in Sleep. This technique compensates for the affects of temperature, voltage and component accuracy. The peripheral can also be configured as a simple Programmable Low-Voltage Detect (LVD) or temperature sensor.

Note: For more information, see AN879, "Using the Microchip Ultra Low-Power Wake-up Module" (DS00879).

# TABLE 4-4:EXIT DELAY ON WAKE-UP BY RESET FROM SLEEP MODE OR ANY IDLE MODE<br/>(BY CLOCK SOURCES)

Power-Managed Mode	Clock Source <sup>(5)</sup>	Exit Delay	Clock Ready Status Bits	
	LP, XT, HS			
	HSPLL		OSTS	
	EC, RC	HF-INTOSC <sup>(2)</sup>		
PRI_IDLE mode	HF-INTOSC <sup>(2)</sup>	ICSD(**	HFIOFS	
	MF-INTOSC <sup>(2)</sup>		MFIOFS	
	LF-INTOSC		None	
SEC_IDLE mode	SOSC	Tcsd <sup>(1)</sup>	SOSCRUN	
	HF-INTOSC <sup>(2)</sup>		HFIOFS	
EC_IDLE mode	MF-INTOSC <sup>(2)</sup>	Tcsd <sup>(1)</sup>	MFIOFS	
	LF-INTOSC		None	
	LP, XT, HS	Tost <sup>(3)</sup>		
	HSPLL	Tost + t <sub>rc</sub> <sup>(3)</sup>	OSTS	
	EC, RC	Tcsd <sup>(1)</sup>		
Sleep mode	HF-INTOSC <sup>(2)</sup>		HFIOFS	
	MF-INTOSC <sup>(2)</sup>	TIOBST <sup>(4)</sup>	MFIOFS	
	LF-INTOSC	1	None	

**Note 1:** TCSD (Parameter 38, Table 31-13) is a required delay when waking from Sleep and all Idle modes, and runs concurrently with any other required delays (see Section 4.4 "Idle Modes").

2: Includes postscaler derived frequencies. On Reset, INTOSC defaults to HF-INTOSC at 8 MHz.

**3:** TOST is the Oscillator Start-up Timer (Parameter 32, Table 31-13). TRC is the PLL Lock-out Timer (Parameter F12, Table 31-7); it is also designated as TPLL.

4: Execution continues during TIOBST (Parameter 39, Table 31-13), the INTOSC stabilization period.

**5:** The clock source is dependent upon the settings of the SCS (OSCCON<1:0>), IRCF (OSCCON<6:4>) and FOSC (CONFIG1H<3:0>) bits.

NOTES:

## 5.0 RESET

The PIC18F87K22 family of devices differentiates between various kinds of Reset:

- a) Power-on Reset (POR)
- b) MCLR Reset during normal operation
- c) MCLR Reset during power-managed modes
- d) Watchdog Timer (WDT) Reset (during execution)
- e) Configuration Mismatch (CM) Reset
- f) Brown-out Reset (BOR)
- g) RESET Instruction
- h) Stack Full Reset
- i) Stack Underflow Reset

This section discusses Resets generated by MCLR, POR and BOR, and covers the operation of the various start-up timers. Stack Reset events are covered in **Section 6.1.3.4 "Stack Full and Underflow Resets"**. WDT Resets are covered in **Section 28.2 "Watchdog Timer (WDT)**".

A simplified block diagram of the on-chip Reset circuit is shown in Figure 5-1.

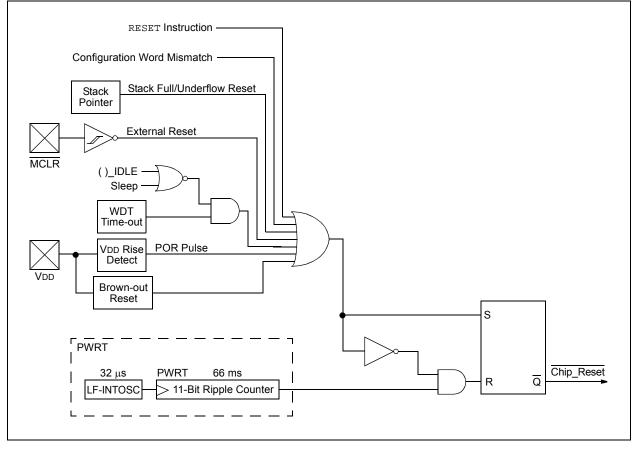
# 5.1 RCON Register

Device Reset events are tracked through the RCON register (Register 5-1). The lower five bits of the register indicate that a specific Reset event has occurred. In most cases, these bits can only be set by the event and must be cleared by the application after the event.

The state of these flag bits, taken together, can be read to indicate the type of Reset that just occurred. This is described in more detail in **Section 5.7** "**Reset State of Registers**".

The RCON register also has a control bit for setting interrupt priority (IPEN). Interrupt priority is discussed in **Section 11.0 "Interrupts**".

### FIGURE 5-1: SIMPLIFIED BLOCK DIAGRAM OF ON-CHIP RESET CIRCUIT



#### R/W-0 R/W-1 R/W-1 R/W-1 R-1 R-1 R/W-0 R/W-0 CM RI TO PD POR **IPEN** SBOREN BOR bit 7 bit 0 Legend: R = Readable bit W = Writable bit U = Unimplemented bit, read as '0' -n = Value at POR '1' = Bit is set '0' = Bit is cleared x = Bit is unknown bit 7 IPEN: Interrupt Priority Enable bit 1 = Enable priority levels on interrupts 0 = Disable priority levels on interrupts (PIC16CXXX Compatibility mode) bit 6 SBOREN: BOR Software Enable bit If BOREN<1:0> = 01: 1 = BOR is enabled 0 = BOR is disabled If BOREN<1:0> = 00. 10 or 11: Bit is disabled and read as '0'. CM: Configuration Mismatch Flag bit bit 5 1 = A Configuration Mismatch Reset has not occurred 0 = A Configuration Mismatch Reset has occurred (must be set in software after a Configuration Mismatch Reset occurs) bit 4 RI: RESET Instruction Flag bit 1 = The RESET instruction was not executed (set by firmware only) 0 = The RESET instruction was executed causing a device Reset (must be set in software after a Brown-out Reset occurs) bit 3 TO: Watchdog Time-out Flag bit 1 = Set by power-up, CLRWDT instruction or SLEEP instruction 0 = A WDT time-out occurred bit 2 PD: Power-Down Detection Flag bit 1 = Set by power-up or by the CLRWDT instruction 0 = Set by execution of the SLEEP instruction POR: Power-on Reset Status bit bit 1 1 = A Power-on Reset has not occurred (set by firmware only) 0 = A Power-on Reset occurred (must be set in software after a Power-on Reset occurs) bit 0 BOR: Brown-out Reset Status bit 1 = A Brown-out Reset has not occurred (set by firmware only) 0 = A Brown-out Reset occurred (must be set in software after a Brown-out Reset occurs)

#### REGISTER 5-1: RCON: RESET CONTROL REGISTER

**Note 1:** It is recommended that the POR bit be set after a Power-on Reset has been detected, so that subsequent Power-on Resets may be detected.

2: Brown-out Reset is said to have occurred when BOR is '0' and POR is '1' (assuming that POR was set to '1' by software immediately after a Power-on Reset).

# 5.2 Master Clear (MCLR)

The MCLR pin provides a method for triggering a hard external Reset of the device. A Reset is generated by holding the pin low. PIC18 extended microcontroller devices have a noise filter in the MCLR Reset path which detects and ignores small pulses.

The  $\overline{\text{MCLR}}$  pin is not driven low by any internal Resets, including the WDT.

## 5.3 Power-on Reset (POR)

A Power-on Reset condition is generated on-chip whenever VDD rises above a certain threshold. This allows the device to start in the initialized state when VDD is adequate for operation.

To take advantage of the POR circuitry, tie the  $\overline{\text{MCLR}}$  pin through a resistor (1 k $\Omega$  to 10 k $\Omega$ ) to VDD. This will eliminate external RC components usually needed to create a Power-on Reset delay. A minimum rise rate for VDD is specified (Parameter D004). For a slow rise time, see Figure 5-2.

When the device starts normal operation (exiting the Reset condition), device operating parameters (such as voltage, frequency and temperature) must be met to ensure operation. If these conditions are not met, the device must be held in Reset until the operating conditions are met.

Power-on Reset events are captured by the  $\overrightarrow{POR}$  bit (RCON<1>). The state of the bit is set to '0' whenever a Power-on Reset occurs and does not change for any other Reset event.  $\overrightarrow{POR}$  is not reset to '1' by any hardware event. To capture multiple events, the user manually resets the bit to '1' in software following any Power-on Reset.

## 5.4 Brown-out Reset (BOR)

The PIC18F87K22 family has four BOR Power modes:

- High-Power BOR
- Medium Power BOR
- Low-Power BOR
- Zero-Power BOR

Each power Mode is selected by the BORPWR<1:0> bits setting (CONFIG2L<6:5>). For low, medium and high-power BOR, the module monitors the VDD depending on the BORV<1:0> setting (CONFIG1L<3:2>). A BOR event re-arms the Power-on Reset. It also causes a Reset, depending on which of the trip levels has been set: 1.8V, 2V, 2.7V or 3V.

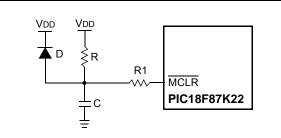
BOR is enabled by the BOREN<1:0> bits (CONFIG2L<2:1>) and the SBOREN bit (RCON<6>). Typical power consumption is listed as Parameter D022A in Section 31.0 "Electrical Characteristics".

In Zero-Power BOR (ZPBORMV), the module monitors the VDD voltage and re-arms the POR at about 2V. ZPBORMV does not cause a Reset, but re-arms the POR.

The BOR accuracy varies with its power level. The lower the power setting, the less accurate the BOR trip levels are. Therefore, the high-power BOR has the highest accuracy and the low-power BOR has the lowest accuracy. The trip levels (BVDD, Parameter D005), current consumption (Section 31.2 "DC Characteristics: Power-Down and Supply Current PIC18F87K22 Family (Industrial/Extended)") and time required below BVDD (TBOR, Parameter 35) can all be found in Section 31.0 "Electrical Characteristics".

# FIGURE 5-2:

#### EXTERNAL POWER-ON RESET CIRCUIT (FOR SLOW VDD POWER-UP)



- Note 1: External Power-on Reset circuit is required only if the VDD power-up slope is too slow. The diode, D, helps discharge the capacitor quickly when VDD powers down.
  - 2:  $R < 40 \text{ k}\Omega$  is recommended to make sure that the voltage drop across R does not violate the device's electrical specification.
  - **3:**  $R1 \ge 1 \ k\Omega$  will limit any current flowing into  $\overline{MCLR}$  from external capacitor, C, in the event of  $\overline{MCLR}/VPP$  pin breakdown due to Electrostatic Discharge (ESD) or Electrical Overstress (EOS).

### 5.4.1 DETECTING BOR

The BOR bit always resets to '0' on any Brown-out Reset or Power-on Reset event. This makes it difficult to determine if a Brown-out Reset event has occurred just by reading the state of BOR alone. A more reliable method is to simultaneously check the state of both POR and BOR. This assumes that the POR bit is reset to '1' in software immediately after any Power-on Reset event. If BOR is '0' while POR is '1', it can be reliably assumed that a Brown-out Reset event has occurred.

LP-BOR cannot be detected with the  $\overline{\text{BOR}}$  bit in the RCON register. LP-BOR can rearm the  $\overline{\text{POR}}$  and can cause a Power-on Reset.

## 5.5 Configuration Mismatch (CM)

The Configuration Mismatch (CM) Reset is designed to detect, and attempt to recover from, random, memory corrupting events. These include Electrostatic Discharge (ESD) events that can cause widespread, single bit changes throughout the device and result in catastrophic failure.

In PIC18F87K22 family Flash devices, the device Configuration registers (located in the configuration memory space) are continuously monitored during operation by comparing their values to complimentary shadow registers. If a mismatch is detected between the two sets of registers, a CM Reset automatically occurs. These events are captured by the CM bit (RCON<5>). The state of the bit is set to '0' whenever a CM event occurs and does not change for any other Reset event.

A CM Reset behaves similarly to a Master Clear Reset, RESET instruction, WDT time-out or Stack Event Reset. As with all hard and power Reset events, the device Configuration Words are reloaded from the Flash Configuration Words in program memory as the device restarts.

### 5.6 **Power-up Timer (PWRT)**

PIC18F87K22 family devices incorporate an on-chip Power-up Timer (PWRT) to help regulate the Power-on Reset process. The PWRT is enabled by setting the PWRTEN bit (CONFIG2L<0>). The main function is to ensure that the device voltage is stable before code is executed.

The Power-up Timer (PWRT) of the PIC18F87K22 family devices is a 13-bit counter that uses the LF-INTOSC source as the clock input. This yields an approximate time interval of 2,048 x 32  $\mu$ s = 66 ms. While the PWRT is counting, the device is held in Reset.

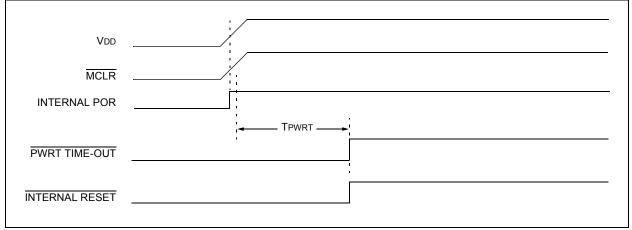
The power-up time delay depends on the LF-INTOSC clock and will vary from chip-to-chip due to temperature and process variation. See DC Parameter 33 for details.

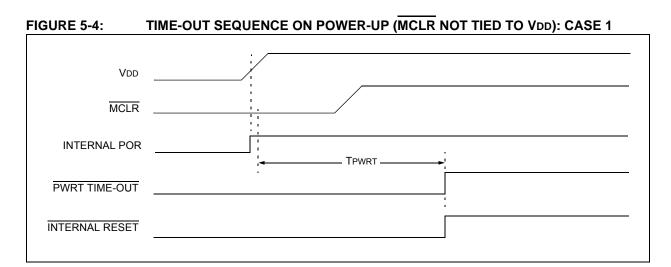
#### 5.6.1 TIME-OUT SEQUENCE

If enabled, the PWRT time-out is invoked after the POR pulse has cleared. The total time-out will vary based on the status of the PWRT. Figure 5-3, Figure 5-4, Figure 5-5 and Figure 5-6 all depict time-out sequences on power-up with the Power-up Timer enabled.

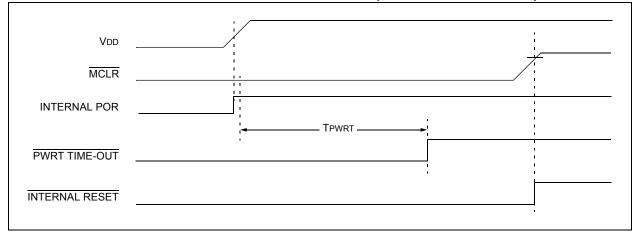
Since the time-outs occur from the POR pulse, if MCLR is kept low long enough, the PWRT will expire. Bringing MCLR high will begin execution immediately (Figure 5-5). This is useful for testing purposes or for synchronizing more than one PIC18 device operating in parallel.

### FIGURE 5-3: TIME-OUT SEQUENCE ON POWER-UP (MCLR TIED TO VDD, VDD RISE < TPWRT)

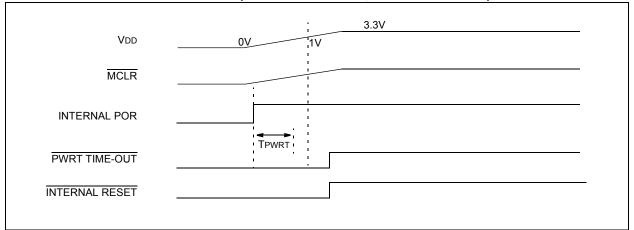




#### FIGURE 5-5: TIME-OUT SEQUENCE ON POWER-UP (MCLR NOT TIED TO VDD): CASE 2



## FIGURE 5-6: SLOW RISE TIME (MCLR TIED TO VDD, VDD RISE > TPWRT)



### 5.7 Reset State of Registers

Most registers are unaffected by a Reset. Their status is unknown on POR and unchanged by all other Resets. The other registers are forced to a "Reset state" depending on the type of Reset that occurred.

Most registers are not affected by a WDT wake-up, since this is viewed as the resumption of normal operation. Status bits from the RCON register ( $\overline{CM}$ ,  $\overline{RI}$ ,  $\overline{TO}$ ,  $\overline{PD}$ ,  $\overline{POR}$  and  $\overline{BOR}$ ) are set or cleared differently in

different Reset situations, as indicated in Table 5-1. These bits are used in software to determine the nature of the Reset.

Table 5-2 describes the Reset states for all of the Special Function Registers. These are categorized by Power-on and Brown-out Resets, Master Clear and WDT Resets, and WDT wake-ups.

# TABLE 5-1:STATUS BITS, THEIR SIGNIFICANCE AND THE INITIALIZATION CONDITION FOR<br/>RCON REGISTER

Condition Program			RCON Register				STKPTR Register		
Condition	Counter <sup>(1)</sup>	СМ	RI	то	PD	POR	BOR	STKFUL	STKUNF
Power-on Reset	0000h	1	1	1	1	0	0	0	0
RESET instruction	0000h	u	0	u	u	u	u	u	u
Brown-out Reset	0000h	1	1	1	1	u	0	u	u
Configuration Mismatch Reset	0000h	0	u	u	u	u	u	u	u
MCLR Reset during power-managed Run modes	0000h	u	u	1	u	u	u	u	u
MCLR Reset during power- managed Idle modes and Sleep mode	0000h	u	u	1	0	u	u	u	u
MCLR Reset during full-power execution	0000h	u	u	u	u	u	u	u	u
Stack Full Reset (STVREN = 1)	0000h	u	u	u	u	u	u	1	u
Stack Underflow Reset (STVREN = 1)	0000h	u	u	u	u	u	u	u	1
Stack Underflow Error (not an actual Reset, STVREN = 0)	0000h	u	u	u	u	u	u	u	1
WDT time-out during full-power or power-managed Run modes	0000h	u	u	0	u	u	u	u	u
WDT time-out during power-managed Idle or Sleep modes	PC + 2	u	u	0	0	u	u	u	u
Interrupt exit from power-managed modes	PC + 2	u	u	u	0	u	u	u	u

Legend: u = unchanged

**Note 1:** When the wake-up is due to an interrupt and the GIEH or GIEL bit is set, the PC is loaded with the interrupt vector (0008h or 0018h).

TABLE 5-2: INITIALIZATION CONDITIONS FOR ALL REGISTERS					
Register	Applicabl	e Devices	Power-on Reset, Brown-out Reset	MCLR Resets, WDT Reset, RESET Instruction, Stack Resets, CM Resets	Wake-up via WDT or Interrupt
TOSU	PIC18F6XK22	PIC18F8XK22	0 0000	0 0000	0 uuuu <b>(1)</b>
TOSH	PIC18F6XK22	PIC18F8XK22	0000 0000	0000 0000	uuuu uuuu <sup>(1)</sup>
TOSL	PIC18F6XK22	PIC18F8XK22	0000 0000	0000 0000	uuuu uuuu <sup>(1)</sup>
STKPTR	PIC18F6XK22	PIC18F8XK22	00-0 0000	uu-0 0000	uu-u uuuu <b>(1)</b>
PCLATU	PIC18F6XK22	PIC18F8XK22	0 0000	0 0000	u uuuu
PCLATH	PIC18F6XK22	PIC18F8XK22	0000 0000	0000 0000	uuuu uuuu
PCL	PIC18F6XK22	PIC18F8XK22	0000 0000	0000 0000	PC + 2 <sup>(2)</sup>
TBLPTRU	PIC18F6XK22	PIC18F8XK22	00 0000	00 0000	uu uuuu
TBLPTRH	PIC18F6XK22	PIC18F8XK22	0000 0000	0000 0000	uuuu uuuu
TBLPTRL	PIC18F6XK22	PIC18F8XK22	0000 0000	0000 0000	uuuu uuuu
TABLAT	PIC18F6XK22	PIC18F8XK22	0000 0000	0000 0000	uuuu uuuu
PRODH	PIC18F6XK22	PIC18F8XK22	XXXX XXXX	սսսս սսսս	uuuu uuuu
PRODL	PIC18F6XK22	PIC18F8XK22	XXXX XXXX	սսսս սսսս	uuuu uuuu
INTCON	PIC18F6XK22	PIC18F8XK22	0000 000x	0000 000u	uuuu uuuu <sup>(3)</sup>
INTCON2	PIC18F6XK22	PIC18F8XK22	1111 1111	1111 1111	uuuu uuuu <sup>(3)</sup>
INTCON3	PIC18F6XK22	PIC18F8XK22	1100 0000	1100 0000	uuuu uuuu <sup>(3)</sup>
INDF0	PIC18F6XK22	PIC18F8XK22	N/A	N/A	N/A
POSTINC0	PIC18F6XK22	PIC18F8XK22	N/A	N/A	N/A
POSTDEC0	PIC18F6XK22	PIC18F8XK22	N/A	N/A	N/A
PREINC0	PIC18F6XK22	PIC18F8XK22	N/A	N/A	N/A
PLUSW0	PIC18F6XK22	PIC18F8XK22	N/A	N/A	N/A
FSR0H	PIC18F6XK22	PIC18F8XK22	0000	0000	uuuu
FSR0L	PIC18F6XK22	PIC18F8XK22	XXXX XXXX	սսսս սսսս	uuuu uuuu
WREG	PIC18F6XK22	PIC18F8XK22	XXXX XXXX	սսսս սսսս	uuuu uuuu
INDF1	PIC18F6XK22	PIC18F8XK22	N/A	N/A	N/A
POSTINC1	PIC18F6XK22	PIC18F8XK22	N/A	N/A	N/A
POSTDEC1	PIC18F6XK22	PIC18F8XK22	N/A	N/A	N/A
PREINC1	PIC18F6XK22	PIC18F8XK22	N/A	N/A	N/A
PLUSW1	PIC18F6XK22	PIC18F8XK22	N/A	N/A	N/A
FSR1H	PIC18F6XK22	PIC18F8XK22	0000	0000	uuuu
FSR1L	PIC18F6XK22	PIC18F8XK22	xxxx xxxx	uuuu uuuu	սսսս սսսս
BSR	PIC18F6XK22	PIC18F8XK22	0000	0000	uuuu

#### TABLE 5-2: INITIALIZATION CONDITIONS FOR ALL REGISTERS

**Legend:** u = unchanged, x = unknown, - = unimplemented bit, read as '0', q = value depends on condition. Shaded cells indicate conditions do not apply for the designated device.

**Note 1:** When the wake-up is due to an interrupt, and the GIEL or GIEH bit is set, the TOSU, TOSH and TOSL are updated with the current value of the PC. The STKPTR is modified to point to the next location in the hardware stack.

2: When the wake-up is due to an interrupt and the GIEL or GIEH bit is set, the PC is loaded with the interrupt vector (0008h or 0018h).

- 3: One or more bits in the INTCONx or PIRx registers will be affected (to cause wake-up).
- 4: See Table 5-1 for Reset value for specific condition.

TABLE 5-2: INITIALIZATION CONDITIONS FOR ALL REGISTERS (CONTINUED)					ED)
Register	Applicabl	e Devices	Power-on Reset, Brown-out Reset	MCLR Resets, WDT Reset, RESET Instruction, Stack Resets, CM Resets	Wake-up via WDT or Interrupt
INDF2	PIC18F6XK22	PIC18F8XK22	N/A	N/A	N/A
POSTINC2	PIC18F6XK22	PIC18F8XK22	N/A	N/A	N/A
POSTDEC2	PIC18F6XK22	PIC18F8XK22	N/A	N/A	N/A
PREINC2	PIC18F6XK22	PIC18F8XK22	N/A	N/A	N/A
PLUSW2	PIC18F6XK22	PIC18F8XK22	N/A	N/A	N/A
FSR2H	PIC18F6XK22	PIC18F8XK22	xxxx	uuuu	uuuu
FSR2L	PIC18F6XK22	PIC18F8XK22	XXXX XXXX	սսսս սսսս	uuuu uuuu
STATUS	PIC18F6XK22	PIC18F8XK22	x xxxx	u uuuu	u uuuu
TMR0H	PIC18F6XK22	PIC18F8XK22	0000 0000	սսսս սսսս	uuuu uuuu
TMR0L	PIC18F6XK22	PIC18F8XK22	XXXX XXXX	սսսս սսսս	uuuu uuuu
TOCON	PIC18F6XK22	PIC18F8XK22	1111 1111	1111 1111	uuuu uuuu
SPBRGH1	PIC18F6XK22	PIC18F8XK22	0000 0000	0000 0000	uuuu uuuu
OSCCON	PIC18F6XK22	PIC18F8XK22	0110 q000	0110 q000	uuuu quuu
IPR5	PIC18F65K22	PIC18F85K22	1 -111	1 -111	u -uuu
IPR5	PIC18F66K22 PIC18F67K22	PIC18F86K22 PIC18F87K22	1000 0000	1000 0000	սսսս սսսս
WDTCON	PIC18F6XK22	PIC18F8XK22	0-x0 -000	0-x0 -000	u-uu -uuu
RCON	PIC18F6XK22	PIC18F8XK22	0111 11qq	0uqq qquu	uuuu qquu
TMR1H	PIC18F6XK22	PIC18F8XK22	XXXX XXXX	uuuu uuuu	uuuu uuuu
TMR1L	PIC18F6XK22	PIC18F8XK22	XXXX XXXX	uuuu uuuu	uuuu uuuu
T1CON	PIC18F6XK22	PIC18F8XK22	0000 0000	uuuu uuuu	uuuu uuuu
TMR2	PIC18F6XK22	PIC18F8XK22	0000 0000	0000 0000	uuuu uuuu
PR2	PIC18F6XK22	PIC18F8XK22	1111 1111	1111 1111	uuuu uuuu
T2CON	PIC18F6XK22	PIC18F8XK22	-000 0000	-000 0000	-uuu uuuu
SSP1BUF	PIC18F6XK22	PIC18F8XK22	XXXX XXXX	սսսս սսսս	uuuu uuuu
SSP1ADD	PIC18F6XK22	PIC18F8XK22	0000 0000	0000 0000	uuuu uuuu
SSP1STAT	PIC18F6XK22	PIC18F8XK22	0000 0000	0000 0000	uuuu uuuu
SSP1CON1	PIC18F6XK22	PIC18F8XK22	0000 0000	0000 0000	uuuu uuuu
SSP1CON2	PIC18F6XK22	PIC18F8XK22	0000 0000	0000 0000	uuuu uuuu
ADRESH	PIC18F6XK22	PIC18F8XK22	XXXX XXXX	սսսս սսսս	uuuu uuuu
ADRESL	PIC18F6XK22	PIC18F8XK22	XXXX XXXX	սսսս սսսս	uuuu uuuu
ADCON0	PIC18F6XK22	PIC18F8XK22	-000 0000	-000 0000	-uuu uuuu
ADCON1	PIC18F6XK22	PIC18F8XK22	0000 0000	0000 0000	uuuu uuuu
ADCON2	PIC18F6XK22	PIC18F8XK22	0-00 0000	0-00 0000	u-uu uuuu

### TABLE 5-2: INITIALIZATION CONDITIONS FOR ALL REGISTERS (CONTINUED)

**Legend:** u = unchanged, x = unknown, - = unimplemented bit, read as '0', q = value depends on condition. Shaded cells indicate conditions do not apply for the designated device.

**Note 1:** When the wake-up is due to an interrupt, and the GIEL or GIEH bit is set, the TOSU, TOSH and TOSL are updated with the current value of the PC. The STKPTR is modified to point to the next location in the hardware stack.

2: When the wake-up is due to an interrupt and the GIEL or GIEH bit is set, the PC is loaded with the interrupt vector (0008h or 0018h).

3: One or more bits in the INTCONx or PIRx registers will be affected (to cause wake-up).

TABLE 5-2:         INITIALIZATION CONDITIONS FOR ALL REGISTERS (CONTINUED)					
Register	Applicabl	e Devices	Power-on Reset, Brown-out Reset	MCLR Resets, WDT Reset, RESET Instruction, Stack Resets, CM Resets	Wake-up via WDT or Interrupt
ECCP1AS	PIC18F6XK22	PIC18F8XK22	0000 0000	0000 0000	uuuu uuuu
ECCP1DEL	PIC18F6XK22	PIC18F8XK22	0000 0000	0000 0000	uuuu uuuu
CCPR1H	PIC18F6XK22	PIC18F8XK22	XXXX XXXX	սսսս սսսս	uuuu uuuu
CCPR1L	PIC18F6XK22	PIC18F8XK22	XXXX XXXX	սսսս սսսս	uuuu uuuu
CCP1CON	PIC18F6XK22	PIC18F8XK22	0000 0000	0000 0000	uuuu uuuu
PIR5	PIC18F65K22	PIC18F85K22	0 -000	0 -000	u -uuu
PIR5	PIC18F66K22 PIC18F67K22	PIC18F86K22 PIC18F87K22	0000 0000	0000 0000	uuuu uuuu
PIE5	PIC18F65K22	PIC18F85K22	0 0000	0 0000	u uuuu <sup>(1)</sup>
PIE5	PIC18F66K22 PIC18F67K22	PIC18F86K22 PIC18F87K22	0000 000	0000 0000	uuuu uuuu <sup>(1)</sup>
IPR4	PIC18F65K22	PIC18F85K22	11 1111	11 1111	uu uuuu
IPR4	PIC18F66K22 PIC18F67K22	PIC18F86K22 PIC18F87K22	1111 1111	1111 1111	uuuu uuuu
PIR4	PIC18F65K22	PIC18F85K22	00 0000	00 0000	uu uuuu <b>(1)</b>
PIR4	PIC18F66K22 PIC18F67K22	PIC18F86K22 PIC18F87K22	0000 0000	0000 0000	uuuu uuuu <sup>(1)</sup>
PIE4	PIC18F65K22	PIC18F85K22	00 0000	00 0000	uu uuuu
PIE4	PIC18F66K22 PIC18F67K22	PIC18F86K22 PIC18F87K22	0000 0000	0000 0000	uuuu uuuu
CVRCON	PIC18F6XK22	PIC18F8XK22	0000 0000	0000 0000	uuuu uuuu
CMSTAT	PIC18F6XK22	PIC18F8XK22	xxx	xxx	uuu
TMR3H	PIC18F6XK22	PIC18F8XK22	XXXX XXXX	սսսս սսսս	uuuu uuuu
TMR3L	PIC18F6XK22	PIC18F8XK22	XXXX XXXX	uuuu uuuu	uuuu uuuu
T3CON	PIC18F6XK22	PIC18F8XK22	0000 0000	00x0 0x00	սսսս սսսս
T3GCON	PIC18F6XK22	PIC18F8XK22	0000 0x00	0000 0000	սսսս սսսս
SPBRG1	PIC18F6XK22	PIC18F8XK22	0000 0000	0000 0000	սսսս սսսս
RCREG1	PIC18F6XK22	PIC18F8XK22	0000 0000	0000 0000	սսսս սսսս
TXREG1	PIC18F6XK22	PIC18F8XK22	XXXX XXXX	XXXX XXXX	uuuu uuuu
TXSTA1	PIC18F6XK22	PIC18F8XK22	0000 0010	0000 0010	սսսս սսսս
RCSTA1	PIC18F6XK22	PIC18F8XK22	0000 000x	0000 000x	սսսս սսսս
T1GCON	PIC18F6XK22	PIC18F8XK22	0000 0x00	00x0 0x00	uuuu -uuu
IPR6	PIC18F6XK22	PIC18F8XK22	1 -111	1 -111	u -uuu
HLVDCON	PIC18F6XK22	PIC18F8XK22	0000 0101	0000 0101	սսսս սսսս
PSPCON	PIC18F6XK22	PIC18F8XK22	0000	0000	uuuu
PIR6	PIC18F6XK22	PIC18F8XK22	0 -000	0 -000	u -uuu

### TABLE 5-2: INITIALIZATION CONDITIONS FOR ALL REGISTERS (CONTINUED)

**Legend:** u = unchanged, x = unknown, - = unimplemented bit, read as '0', q = value depends on condition. Shaded cells indicate conditions do not apply for the designated device.

**Note 1:** When the wake-up is due to an interrupt, and the GIEL or GIEH bit is set, the TOSU, TOSH and TOSL are updated with the current value of the PC. The STKPTR is modified to point to the next location in the hardware stack.

2: When the wake-up is due to an interrupt and the GIEL or GIEH bit is set, the PC is loaded with the interrupt vector (0008h or 0018h).

3: One or more bits in the INTCONx or PIRx registers will be affected (to cause wake-up).

-up via WDT Interrupt
uu uuuu
-u uuuu
uu uuuu
uu uuuu
uu uuuu
-u uuuu
-u uuuu
uu uuuu
-u uuuu
uu uuu-
uu uuuu
uu uuu-
uu uuuu
uu uuuu

### TABLE 5-2: INITIALIZATION CONDITIONS FOR ALL REGISTERS (CONTINUED)

**Legend:** u = unchanged, x = unknown, - = unimplemented bit, read as '0', q = value depends on condition. Shaded cells indicate conditions do not apply for the designated device.

**Note 1:** When the wake-up is due to an interrupt, and the GIEL or GIEH bit is set, the TOSU, TOSH and TOSL are updated with the current value of the PC. The STKPTR is modified to point to the next location in the hardware stack.

2: When the wake-up is due to an interrupt and the GIEL or GIEH bit is set, the PC is loaded with the interrupt vector (0008h or 0018h).

3: One or more bits in the INTCONx or PIRx registers will be affected (to cause wake-up).

Register	INITIALIZATION CONDIT		Power-on Reset,	MCLR Resets, WDT Reset, RESET Instruction,	Wake-up via WDT
Register			Brown-out Reset	Stack Resets, CM Resets	or Interrupt
PORTC	PIC18F6XK22	PIC18F8XK22	xxxx xxxx	սսսս սսսս	uuuu uuuu
PORTB	PIC18F6XK22	PIC18F8XK22	XXXX XXXX	uuuu uuuu	uuuu uuuu
PORTA	PIC18F6XK22	PIC18F8XK22	xx0x 0000	uu0u 0000	uuuu uuuu
EECON1	PIC18F6XK22	PIC18F8XK22	xx-0 x000	uu-0 u000	uu-u uuuu
EECON2	PIC18F6XK22	PIC18F8XK22			
TMR5H	PIC18F6XK22	PIC18F8XK22	XXXX XXXX	uuuu uuuu	uuuu uuuu
TMR5L	PIC18F6XK22	PIC18F8XK22	XXXX XXXX	uuuu uuuu	uuuu uuuu
T5CON	PIC18F6XK22	PIC18F8XK22	0000 0000	uuuu uuuu	uuuu uuuu
T5GCON	PIC18F6XK22	PIC18F8XK22	00x0 0x00	uuuu uuuu	uuuu uuuu
CCPR4H	PIC18F6XK22	PIC18F8XK22	xxxx xxxx	uuuu uuuu	uuuu uuuu
CCPR4L	PIC18F6XK22	PIC18F8XK22	XXXX XXXX	uuuu uuuu	uuuu uuuu
CCP4CON	PIC18F6XK22	PIC18F8XK22	00 0000	00 0000	uu uuuu
CCPR5H	PIC18F6XK22	PIC18F8XK22	XXXX XXXX	uuuu uuuu	uuuu uuuu
CCPR5L	PIC18F6XK22	PIC18F8XK22	XXXX XXXX	uuuu uuuu	uuuu uuuu
CCP5CON	PIC18F6XK22	PIC18F8XK22	00 0000	00 0000	uu uuuu
CCPR6H	PIC18F6XK22	PIC18F8XK22	XXXX XXXX	uuuu uuuu	uuuu uuuu
CCPR6L	PIC18F6XK22	PIC18F8XK22	XXXX XXXX	uuuu uuuu	uuuu uuuu
CCP6CON	PIC18F6XK22	PIC18F8XK22	00 0000	00 0000	uu uuuu
CCPR7H	PIC18F6XK22	PIC18F8XK22	xxxx xxxx	uuuu uuuu	uuuu uuuu
CCPR7L	PIC18F6XK22	PIC18F8XK22	XXXX XXXX	uuuu uuuu	uuuu uuuu
CCP7CON	PIC18F6XK22	PIC18F8XK22	00 0000	00 0000	uu uuuu
TMR4	PIC18F6XK22	PIC18F8XK22	xxxx xxxx	xxxx xxxx	uuuu uuuu
PR4	PIC18F6XK22	PIC18F8XK22	1111 1111	1111 1111	1111 1111
T4CON	PIC18F6XK22	PIC18F8XK22	-111 1111	-111 1111	-uuu uuuu
SSP2BUF	PIC18F6XK22	PIC18F8XK22	xxxx xxxx	uuuu uuuu	uuuu uuuu
SSP2ADD	PIC18F6XK22	PIC18F8XK22	0000 0000	0000 0000	uuuu uuuu
SSP2STAT	PIC18F6XK22	PIC18F8XK22	0000 0000	0000 0000	uuuu uuuu
SSP2CON1	PIC18F6XK22	PIC18F8XK22	0000 0000	0000 0000	uuuu uuuu
SSP2CON2	PIC18F6XK22	PIC18F8XK22	0100 0000	0000 0000	uuuu uuuu
BAUDCON1	PIC18F6XK22	PIC18F8XK22	0100 0-00	0100 0-00	uuuu u-uu
OSCCON2	PIC18F6XK22	PIC18F8XK22	-0 0-x0	-0 0-u0	-u u-uu
EEADRH	PIC18F6XK22	PIC18F8XK22	00	00	uu
EEADR	PIC18F6XK22	PIC18F8XK22	0000 0000	0000 0000	uuuu uuuu
EEDATA	PIC18F6XK22	PIC18F8XK22	0000 0000	0000 0000	uuuu uuuu
PIE6	PIC18F6XK22	PIC18F8XK22	0 -000	0 -000	u -uuu

#### TABLE 5-2: INITIALIZATION CONDITIONS FOR ALL REGISTERS (CONTINUED)

**Legend:** u = unchanged, x = unknown, - = unimplemented bit, read as '0', q = value depends on condition. Shaded cells indicate conditions do not apply for the designated device.

**Note 1:** When the wake-up is due to an interrupt, and the GIEL or GIEH bit is set, the TOSU, TOSH and TOSL are updated with the current value of the PC. The STKPTR is modified to point to the next location in the hardware stack.

2: When the wake-up is due to an interrupt and the GIEL or GIEH bit is set, the PC is loaded with the interrupt vector (0008h or 0018h).

3: One or more bits in the INTCONx or PIRx registers will be affected (to cause wake-up).

TABLE 5-2:	INITIALIZATION CONDITIONS FOR ALL REGISTERS (CONTINUED)				
			MCLR Resets,		
Register	Applicable Devices	Power-on Reset,	WDT Reset, RESET Instruction.	Wake-up	

Register	Applicable Devices		Power-on Reset, Brown-out Reset	MCLR Resets, WDT Reset, RESET Instruction, Stack Resets, CM Resets	Wake-up via WDT or Interrupt
RTCCFG	PIC18F6XK22	PIC18F8XK22	0-00 0000	u-uu uuuu	u-uu uuuu
RTCCAL	PIC18F6XK22	PIC18F8XK22	0000 0000	uuuu uuuu	uuuu uuuu
RTCVALH	PIC18F6XK22	PIC18F8XK22	XXXX XXXX	uuuu uuuu	uuuu uuuu
RTCVALL	PIC18F6XK22	PIC18F8XK22	0000 0000	uuuu uuuu	uuuu uuuu
ALRMCFG	PIC18F6XK22	PIC18F8XK22	0000 0000	uuuu uuuu	uuuu uuuu
ALRMRPT	PIC18F6XK22	PIC18F8XK22	0000 0000	uuuu uuuu	uuuu uuuu
ALRMVALH	PIC18F6XK22	PIC18F8XK22	xxxx xxxx	uuuu uuuu	uuuu uuuu
ALRMVALL	PIC18F6XK22	PIC18F8XK22	xxxx xxxx	uuuu uuuu	uuuu uuuu
CTMUCONH	PIC18F6XK22	PIC18F8XK22	0-00 0000	0-00 0000	u-uu uuuu
CTMUCONL	PIC18F6XK22	PIC18F8XK22	0000 00xx	0000 00xx	uuuu uuuu
CTMUICONH	PIC18F6XK22	PIC18F8XK22	0000 0000	0000 0000	uuuu uuuu
CM1CON	PIC18F6XK22	PIC18F8XK22	0001 1111	0001 1111	սսսս սսսս
PADCFG1	PIC18F6XK22	PIC18F8XK22	0000-	uuuu-	uuuu-
PADCFG1	PIC18F6XK22	PIC18F8XK22	00000-	uuuuu-	uuuuu-
ECCP2AS	PIC18F6XK22	PIC18F8XK22	0000 0000	0000 0000	uuuu uuuu
ECCP2DEL	PIC18F6XK22	PIC18F8XK22	0000 0000	0000 0000	uuuu uuuu
CCPR2H	PIC18F6XK22	PIC18F8XK22	xxxx xxxx	uuuu uuuu	uuuu uuuu
CCPR2L	PIC18F6XK22	PIC18F8XK22	XXXX XXXX	սսսս սսսս	uuuu uuuu
CCP2CON	PIC18F6XK22	PIC18F8XK22	0000 0000	0000 0000	uuuu uuuu
ECCP3AS	PIC18F6XK22	PIC18F8XK22	0000 0000	0000 0000	uuuu uuuu
ECCP3DEL	PIC18F6XK22	PIC18F8XK22	0000 0000	0000 0000	uuuu uuuu
CCPR3H	PIC18F6XK22	PIC18F8XK22	XXXX XXXX	սսսս սսսս	uuuu uuuu
CCPR3L	PIC18F6XK22	PIC18F8XK22	xxxx xxxx	uuuu uuuu	uuuu uuuu
CCP3CON	PIC18F6XK22	PIC18F8XK22	0000 0000	0000 0000	uuuu uuuu
CCPR8H	PIC18F6XK22	PIC18F8XK22	xxxx xxxx	uuuu uuuu	uuuu uuuu
CCPR8L	PIC18F6XK22	PIC18F8XK22	xxxx xxxx	uuuu uuuu	uuuu uuuu
CCP8CON	PIC18F6XK22	PIC18F8XK22	00 0000	00 0000	uu uuuu
CCPR9H	PIC18F66K22 PIC18F67K22	PIC18F86K22 PIC18F87K22	XXXX XXXX	սսսս սսսս	սսսս սսսս
CCPR9L		PIC18F86K22 PIC18F87K22	xxxx xxxx	սսսս սսսս	սսսս սսսս
CCP9CON	PIC18F66K22 PIC18F67K22	PIC18F86K22 PIC18F87K22	00 0000	00 0000	uu uuuu
CCPR10H	PIC18F66K22 PIC18F67K22	PIC18F86K22 PIC18F87K22	XXXX XXXX	սսսս սսսս	սսսս սսսս
CCPR10L	PIC18F66K22 PIC18F67K22	PIC18F86K22 PIC18F87K22	XXXX XXXX	սսսս սսսս	սսսս սսսս

Legend: u = unchanged, x = unknown, - = unimplemented bit, read as '0', q = value depends on condition. Shaded cells indicate conditions do not apply for the designated device.

Note 1: When the wake-up is due to an interrupt, and the GIEL or GIEH bit is set, the TOSU, TOSH and TOSL are updated with the current value of the PC. The STKPTR is modified to point to the next location in the hardware stack.

2: When the wake-up is due to an interrupt and the GIEL or GIEH bit is set, the PC is loaded with the interrupt vector (0008h or 0018h).

3: One or more bits in the INTCONx or PIRx registers will be affected (to cause wake-up).

Register	Applicabl	e Devices	Power-on Reset, Brown-out Reset	MCLR Resets, WDT Reset, RESET Instruction, Stack Resets, CM Resets	Wake-up via WDT or Interrupt
CCP10CON	PIC18F66K22 PIC18F67K22	PIC18F86K22 PIC18F87K22	00 0000	00 0000	uu uuuu
TMR7H	PIC18F66K22 PIC18F67K22	PIC18F86K22 PIC18F87K22	XXXX XXXX	սսսս սսսս	սսսս սսսս
TMR7L	PIC18F66K22 PIC18F67K22	PIC18F86K22 PIC18F87K22	XXXX XXXX	սսսս սսսս	սսսս սսսս
T7CON	PIC18F66K22 PIC18F67K22	PIC18F86K22 PIC18F87K22	0000 0000	սսսս սսսս	սսսս սսսս
T7GCON	PIC18F66K22 PIC18F67K22	PIC18F86K22 PIC18F87K22	0000 0x00	0000 0x00	սսսս սսսս
TMR6	PIC18F6XK22	PIC18F8XK22	0000 0000	0000 0000	uuuu uuuu
PR6	PIC18F6XK22	PIC18F8XK22	1111 1111	1111 1111	uuuu uuuu
T6CON	PIC18F6XK22	PIC18F8XK22	-000 0000	-000 0000	-uuu uuuu
TMR8	PIC18F6XK22	PIC18F8XK22	0000 0000	0000 0000	uuuu uuuu
PR8	PIC18F6XK22	PIC18F8XK22	1111 1111	1111 1111	uuuu uuuu
T8CON	PIC18F6XK22	PIC18F8XK22	-000 0000	-000 0000	-uuu uuuu
TMR10	PIC18F66K22 PIC18F67K22	PIC18F86K22 PIC18F87K22	0000 0000	0000 0000	սսսս սսսս
PR10	PIC18F66K22 PIC18F67K22	PIC18F86K22 PIC18F87K22	1111 1111	1111 1111	սսսս սսսս
T10CON	PIC18F66K22 PIC18F67K22	PIC18F86K22 PIC18F87K22	-000 0000	-000 0000	-uuu uuuu
TMR12	PIC18F66K22 PIC18F67K22	PIC18F86K22 PIC18F87K22	0000 0000	0000 0000	սսսս սսսս
PR12	PIC18F66K22 PIC18F67K22	PIC18F86K22 PIC18F87K22	1111 1111	1111 1111	uuuu uuuu
T12CON	PIC18F66K22 PIC18F67K22	PIC18F86K22 PIC18F87K22	-000 0000	-000 0000	-uuu uuuu
CM2CON	PIC18F6XK22	PIC18F8XK22	0001 1111	0001 1111	uuuu uuuu
CM3CON	PIC18F6XK22	PIC18F8XK22	0001 1111	0001 1111	uuuu uuuu
CCPTMRS0	PIC18F6XK22	PIC18F8XK22	0000 0000	սսսս սսսս	սսսս սսսս
CCPTMRS1	PIC18F6XK22	PIC18F8XK22	00-0 -000	uu-u -uuu	uu-u -uuu
CCPTMRS2	PIC18F66K22 PIC18F67K22	PIC18F86K22 PIC18F87K22	0 -000	u -uuu	u -uuu
CCPTMRS2	PIC18F65K22	PIC18F85K22	00	uu	uu
REFOCON	PIC18F6XK22	PIC18F8XK22	0-00 0000	u-uu uuuu	u-uu uuuu
ODCON1	PIC18F6XK22	PIC18F8XK22	0000	uuuu	uuuu

#### TABLE 5-2: INITIALIZATION CONDITIONS FOR ALL REGISTERS (CONTINUED)

**Note 1:** When the wake-up is due to an interrupt, and the GIEL or GIEH bit is set, the TOSU, TOSH and TOSL are updated with the current value of the PC. The STKPTR is modified to point to the next location in the hardware stack.

2: When the wake-up is due to an interrupt and the GIEL or GIEH bit is set, the PC is loaded with the interrupt vector (0008h or 0018h).

3: One or more bits in the INTCONx or PIRx registers will be affected (to cause wake-up).

TABLE 5-2:	INITIALIZA	INITIALIZATION CONDITIONS FOR ALL REGISTERS (CONTINUED)						
Register	Applicable Devices		Power-on Reset, Brown-out Reset	MCLR Resets, WDT Reset, RESET Instruction, Stack Resets, CM Resets	Wake-up via WDT or Interrupt			
ODCON2	PIC18F66K22 PIC18F67K22	PIC18F86K22 PIC18F87K22	0000 0000	սսսս սսսս	սսսս սսսս			
ODCON2	PIC18F65K22	PIC18F85K22	00 0000	uu uuuu	uu uuuu			
ODCON3	PIC18F6XK22	PIC18F8XK22	000	uuu	uuu			
MEMCON	PIC18F6XK22	PIC18F8XK22	0-0000	0-0000	u-uuuu			
ANCON0	PIC18F6XK22	PIC18F8XK22	1111 1111	սսսս սսսս	uuuu uuuu			
ANCON1	PIC18F6XK22	PIC18F8XK22	1111 1111	սսսս սսսս	uuuu uuuu			
ANCON2	PIC18F6XK22	PIC18F8XK22	1111 1111	սսսս սսսս	uuuu uuuu			
RCSTA2	PIC18F6XK22	PIC18F8XK22	0000 000x	0000 000x	սսսս սսսս			
TXSTA2	PIC18F6XK22	PIC18F8XK22	0000 0010	0000 0010	uuuu uuuu			
BAUDCON2	PIC18F6XK22	PIC18F8XK22	0100 0-00	0100 0-00	uuuu u-uu			
SPBRGH2	PIC18F6XK22	PIC18F8XK22	0000 0000	0000 0000	սսսս սսսս			
SPBRG2	PIC18F6XK22	PIC18F8XK22	0000 0000	0000 0000	սսսս սսսս			
RCREG2	PIC18F6XK22	PIC18F8XK22	0000 0000	0000 0000	սսսս սսսս			
TXREG2	PIC18F6XK22	PIC18F8XK22	xxxx xxxx	xxxx xxxx	uuuu uuuu			
PSTR2CON	PIC18F6XK22	PIC18F8XK22	00-0 0001	00-0 0001	uu-u uuuu			
PSTR3CON	PIC18F6XK22	PIC18F8XK22	00-0 0001	00-0 0001	uu-u uuuu			
PMD0	PIC18F6XK22	PIC18F8XK22	0000 0000	0000 0000	uuuu uuuu			
PMD1	PIC18F6XK22	PIC18F8XK22	0000 0000	0000 0000	սսսս սսսս			
PMD2	PIC18F66K22 PIC18F67K22	PIC18F86K22 PIC18F87K22	0000 0000	0000 0000	սսսս սսսս			
PMD2	PIC18F65K22	PIC18F85K22	-0-0 0000	-0-0 0000	-u-u uuuu			
PMD3	PIC18F66K22 PIC18F67K22	PIC18F86K22 PIC18F87K22	0000 0000	0000 0000	սսսս սսսս			
PMD3	PIC18F65K22	PIC18F85K22	00 000-	00 000-	uu uuu-			

#### TABLE 5-2: INITIALIZATION CONDITIONS FOR ALL REGISTERS (CONTINUED)

**Legend:** u = unchanged, x = unknown, - = unimplemented bit, read as '0', q = value depends on condition. Shaded cells indicate conditions do not apply for the designated device.

**Note 1:** When the wake-up is due to an interrupt, and the GIEL or GIEH bit is set, the TOSU, TOSH and TOSL are updated with the current value of the PC. The STKPTR is modified to point to the next location in the hardware stack.

2: When the wake-up is due to an interrupt and the GIEL or GIEH bit is set, the PC is loaded with the interrupt vector (0008h or 0018h).

3: One or more bits in the INTCONx or PIRx registers will be affected (to cause wake-up).

### 6.0 MEMORY ORGANIZATION

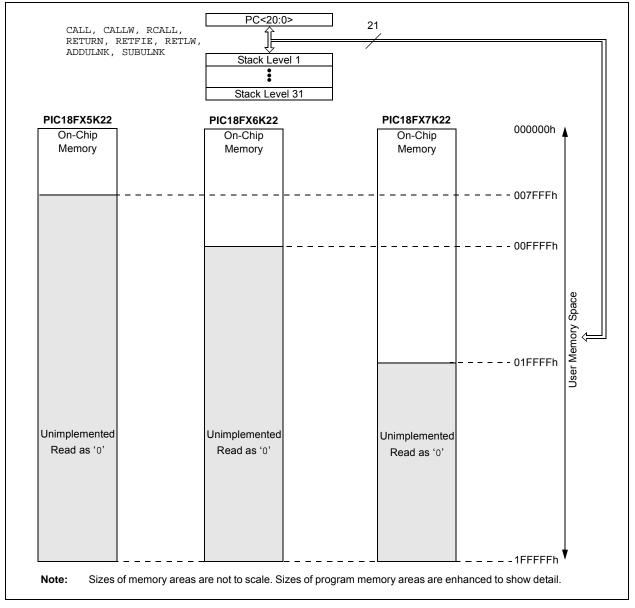
PIC18F87K22 family devices have these types of memory:

- Program Memory
- Data RAM
- Data EEPROM

As Harvard architecture devices, the data and program memories use separate busses. This enables concurrent access of the two memory spaces. The data EEPROM, for practical purposes, can be regarded as a peripheral device because it is addressed and accessed through a set of control registers.

Additional detailed information on the operation of the Flash program memory is provided in **Section 7.0 "Flash Program Memory"**. The data EEPROM is discussed separately in **Section 9.0 "Data EEPROM Memory"**.

#### FIGURE 6-1: MEMORY MAPS FOR PIC18F87K22 FAMILY DEVICES



### 6.1 Program Memory Organization

PIC18 microcontrollers implement a 21-bit Program Counter that is capable of addressing a 2-Mbyte program memory space. Accessing a location between the upper boundary of the physically implemented memory and the 2-Mbyte address will return all '0's (a NOP instruction).

The entire PIC18F87K22 family offers a range of on-chip Flash program memory sizes, from 32 Kbytes (up to 16,384 single-word instructions) to 128 Kbytes (65,536 single-word instructions).

- PIC18F65K22 and PIC18F85K22 32 Kbytes of Flash memory, storing up to 16,384 single-word instructions
- PIC18F66K22 and PIC18F86K22 64 Kbytes of Flash memory, storing up to 32,768 single-word instructions
- PIC18F67K22 and PIC18F87K22 128 Kbytes of Flash memory, storing up to 65,536 single-word instructions

The program memory maps for individual family members are shown in Figure 6-1.

#### 6.1.1 HARD MEMORY VECTORS

All PIC18 devices have a total of three hard-coded return vectors in their program memory space. The Reset vector address is the default value to which the Program Counter returns on all device Resets; it is located at 0000h.

PIC18 devices also have two interrupt vector addresses for handling high-priority and low-priority interrupts. The high-priority interrupt vector is located at 0008h and the low-priority interrupt vector is at 0018h. The locations of these vectors are shown, in relation to the program memory map, in Figure 6-2.

#### FIGURE 6-2: HARD VECTOR FOR PIC18F87K22 FAMILY DEVICES

	Reset Vector	0000h
	High-Priority Interrupt Vector	0008h
	Low-Priority Interrupt Vector	0018h
	On-Chip Program Memory	
	Read '0'	
		1FFFFFh
Le	gend: (Top of Memory) repr of on-chip program n Figure 6-1 for device Shaded area represe memory. Areas are n	-specific values). ents unimplemented

### 6.1.2 PROGRAM COUNTER

The Program Counter (PC) specifies the address of the instruction to fetch for execution. The PC is 21 bits wide and contained in three separate 8-bit registers.

The low byte, known as the PCL register, is both readable and writable. The high byte, or PCH register, contains the PC<15:8> bits and is not directly readable or writable. Updates to the PCH register are performed through the PCLATH register. The upper byte is called PCU. This register contains the PC<20:16> bits; it is also not directly readable or writable. Updates to the PCU register are performed through the PCLATU register.

The contents of PCLATH and PCLATU are transferred to the Program Counter by any operation that writes PCL. Similarly, the upper two bytes of the Program Counter are transferred to PCLATH and PCLATU by an operation that reads PCL. This is useful for computed offsets to the PC (see Section 6.1.5.1 "Computed GOTO").

The PC addresses bytes in the program memory. To prevent the PC from becoming misaligned with word instructions, the Least Significant bit of PCL is fixed to a value of '0'. The PC increments by two to address sequential instructions in the program memory.

The CALL, RCALL, GOTO and program branch instructions write to the Program Counter directly. For these instructions, the contents of PCLATH and PCLATU are not transferred to the Program Counter.

#### 6.1.3 RETURN ADDRESS STACK

The return address stack enables execution of any combination of up to 31 program calls and interrupts. The PC is pushed onto the stack when a CALL or RCALL instruction is executed or an interrupt is Acknowledged. The PC value is pulled off the stack on a RETURN, RETLW or a RETFIE instruction. The value also is pulled off the stack on ADDULNK and SUBULNK instructions, if the extended instruction set is enabled. PCLATU and PCLATH are not affected by any of the RETURN or CALL instructions.

The stack operates as a 31-word by 21-bit RAM and a 5-bit Stack Pointer, STKPTR. The stack space is not part of either program or data space. The Stack Pointer is readable and writable and the address on the top of the stack is readable and writable through the Top-of-Stack Special Function Registers. Data can also be pushed to, or popped from, the stack using these registers.

A CALL type instruction causes a push onto the stack. The Stack Pointer is first incremented and the location pointed to by the Stack Pointer is written with the contents of the PC (already pointing to the instruction following the CALL). A RETURN type instruction causes a pop from the stack. The contents of the location pointed to by the STKPTR are transferred to the PC and then the Stack Pointer is decremented.

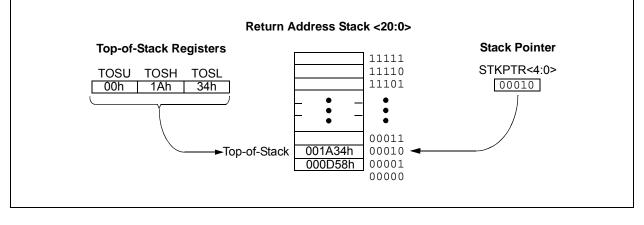
The Stack Pointer is initialized to '00000' after all Resets. There is no RAM associated with the location corresponding to a Stack Pointer value of '00000'; this is only a Reset value. Status bits indicate if the stack is full, has overflowed or has underflowed.

#### 6.1.3.1 Top-of-Stack Access

Only the top of the return address stack (TOS) is readable and writable. A set of three registers, TOSU:TOSH:TOSL, holds the contents of the stack location pointed to by the STKPTR register (Figure 6-3). This allows users to implement a software stack, if necessary. After a CALL, RCALL or interrupt (or ADDULNK and SUBULNK instructions, if the extended instruction set is enabled), the software can read the pushed value by reading the TOSU:TOSH:TOSL registers. These values can be placed on a user-defined software stack. At return time, the software can return these values to TOSU:TOSH:TOSL and do a return.

While accessing the stack, users must disable the Global Interrupt Enable bits to prevent inadvertent stack corruption.





### 6.1.3.2 Return Stack Pointer (STKPTR)

The STKPTR register (Register 6-1) contains the Stack Pointer value, the STKFUL (Stack Full) status bit and the STKUNF (Stack Underflow) status bits. The value of the Stack Pointer can be 0 through 31. The Stack Pointer increments before values are pushed onto the stack and decrements after values are popped off the stack. On Reset, the Stack Pointer value will be zero.

The user may read and write the Stack Pointer value. This feature can be used by a Real-Time Operating System (RTOS) for return stack maintenance.

After the PC is pushed onto the stack 31 times (without popping any values off the stack), the STKFUL bit is set. The STKFUL bit is cleared by software or by a POR.

What happens when the stack becomes full depends on the state of the STVREN (Stack Overflow Reset Enable) Configuration bit. (For a description of the device Configuration bits, see **Section 28.1 "Configuration Bits"**.) If STVREN is set (default), the 31st push will push the (PC + 2) value onto the stack, set the STKFUL bit and reset the device. The STKFUL bit will remain set and the Stack Pointer will be set to zero.

If STVREN is cleared, the STKFUL bit will be set on the 31st push and the Stack Pointer will increment to 31. Any additional pushes will not overwrite the 31st push and the STKPTR will remain at 31.

When the stack has been popped enough times to unload the stack, the next pop will return a value of zero to the PC and set the STKUNF bit while the Stack Pointer remains at zero. The STKUNF bit will remain set until cleared by software or until a POR occurs.

Note:	Returning a value of zero to the PC on an underflow has the effect of vectoring the program to the Reset vector, where the stack conditions can be verified and appropriate actions can be taken. This is not the same as a Reset as the contents
	of the SFRs are not affected.

### 6.1.3.3 PUSH and POP Instructions

Since the Top-of-Stack (TOS) is readable and writable, the ability to push values onto the stack and pull values off of the stack, without disturbing normal program execution, is a desirable feature. The PIC18 instruction set includes two instructions, PUSH and POP, that permit the TOS to be manipulated under software control. TOSU, TOSH and TOSL can be modified to place data or a return address on the stack.

The PUSH instruction places the current PC value onto the stack. This increments the Stack Pointer and loads the current PC value onto the stack.

The POP instruction discards the current TOS by decrementing the Stack Pointer. The previous value pushed onto the stack then becomes the TOS value.

#### R/C-0 R/C-0 R/W-0 R/W-0 U-0 R/W-0 R/W-0 R/W-0 STKFUL<sup>(1)</sup> STKUNF<sup>(1)</sup> SP4 SP3 SP2 SP1 SP0 bit 7 bit 0 Legend: C = Clearable bit R = Readable bit W = Writable bit U = Unimplemented bit, read as '0' -n = Value at POR '0' = Bit is cleared '1' = Bit is set x = Bit is unknown STKFUL: Stack Full Flag bit<sup>(1)</sup> bit 7 1 = Stack has become full or overflowed 0 = Stack has not become full or overflowed STKUNF: Stack Underflow Flag bit<sup>(1)</sup> bit 6 1 = Stack underflow has occurred

## REGISTER 6-1: STKPTR: STACK POINTER REGISTER

bit 5 Unimplemented: Read as '0' bit 4-0 SP<4:0>: Stack Pointer Location bits

0 = Stack underflow did not occur

**Note 1:** Bit 7 and bit 6 are cleared by user software or by a POR.

#### 6.1.3.4 Stack Full and Underflow Resets

Device Resets on stack overflow and stack underflow conditions are enabled by setting the STVREN bit (CONFIG4L<0>). When STVREN is set, a full or underflow condition will set the appropriate STKFUL or STKUNF bit and then cause a device Reset. When STVREN is cleared, a full or underflow condition will set the appropriate STKFUL or STKUNF bit, but not cause a device Reset. The STKFUL or STKUNF bits are cleared by the user software or a Power-on Reset.

#### 6.1.4 FAST REGISTER STACK

A Fast Register Stack is provided for the STATUS, WREG and BSR registers to provide a "fast return" option for interrupts. This stack is only one level deep and is neither readable nor writable. It is loaded with the current value of the corresponding register when the processor vectors for an interrupt. All interrupt sources will push values into the Stack registers. The values in the registers are then loaded back into the working registers if the RETFIE, FAST instruction is used to return from the interrupt.

If both low and high-priority interrupts are enabled, the Stack registers cannot be used reliably to return from low-priority interrupts. If a high-priority interrupt occurs while servicing a low-priority interrupt, the Stack register values stored by the low-priority interrupt will be overwritten. In these cases, users must save the key registers in software during a low-priority interrupt.

If interrupt priority is not used, all interrupts may use the Fast Register Stack for returns from interrupt. If no interrupts are used, the Fast Register Stack can be used to restore the STATUS, WREG and BSR registers at the end of a subroutine call. To use the Fast Register Stack for a subroutine call, a CALL label, FAST instruction must be executed to save the STATUS, WREG and BSR registers to the Fast Register Stack. A RETURN, FAST instruction is then executed to restore these registers from the Fast Register Stack.

Example 6-1 shows a source code example that uses the Fast Register Stack during a subroutine call and return.

#### EXAMPLE 6-1: FAST REGISTER STACK CODE EXAMPLE

CALL SUB1, FAST	;STATUS, WREG, BSR
•	;SAVED IN FAST REGISTER
SUB1 •	;STACK
RETURN FAST	;RESTORE VALUES SAVED ;IN FAST REGISTER STACK

#### 6.1.5 LOOK-UP TABLES IN PROGRAM MEMORY

There may be programming situations that require the creation of data structures, or look-up tables, in program memory. For PIC18 devices, look-up tables can be implemented in two ways:

- Computed GOTO
- Table Reads

#### 6.1.5.1 Computed GOTO

A computed GOTO is accomplished by adding an offset to the Program Counter. An example is shown in Example 6-2.

A look-up table can be formed with an ADDWF PCL instruction and a group of RETLW nn instructions. The W register is loaded with an offset into the table before executing a call to that table. The first instruction of the called routine is the ADDWF PCL instruction. The next instruction executed will be one of the RETLW nn instructions that returns the value, 'nn', to the calling function.

The offset value (in WREG) specifies the number of bytes that the Program Counter should advance and should be multiples of two (LSb = 0).

In this method, only one data byte may be stored in each instruction location and room on the return address stack is required.

#### EXAMPLE 6-2: COMPUTED GOTO USING AN OFFSET VALUE

	MOVF	OFFSET, W
	CALL	TABLE
ORG	nn00h	
TABLE	ADDWF	PCL
	RETLW	nnh
	RETLW	nnh
	RETLW	nnh

#### 6.1.5.2 Table Reads

A better method of storing data in program memory allows two bytes of data to be stored in each instruction location.

Look-up table data may be stored, two bytes per program word, while programming. The Table Pointer (TBLPTR) specifies the byte address and the Table Latch (TABLAT) contains the data that is read from the program memory. Data is transferred from program memory, one byte at a time.

The table read operation is discussed further in **Section 7.1 "Table Reads and Table Writes**".

### 6.2 PIC18 Instruction Cycle

#### 6.2.1 CLOCKING SCHEME

The microcontroller clock input, whether from an internal or external source, is internally divided by four to generate four non-overlapping, quadrature clocks (Q1, Q2, Q3 and Q4). Internally, the Program Counter is incremented on every Q1, with the instruction fetched from the program memory and latched into the Instruction Register (IR) during Q4.

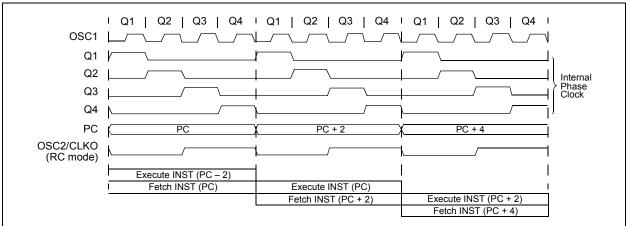
The instruction is decoded and executed during the following Q1 through Q4. The clocks and instruction execution flow are shown in Figure 6-4.

#### 6.2.2 INSTRUCTION FLOW/PIPELINING

An "Instruction Cycle" consists of four Q cycles, Q1 through Q4. The instruction fetch and execute are pipelined in such a manner that a fetch takes one instruction cycle, while the decode and execute take another instruction cycle. However, due to the pipelining, each instruction effectively executes in one cycle. If an instruction (such as GOTO) causes the Program Counter to change, two cycles are required to complete the instruction. (See Example 6-3.)

A fetch cycle begins with the Program Counter (PC) incrementing in Q1.

In the execution cycle, the fetched instruction is latched into the Instruction Register (IR) in cycle Q1. This instruction is then decoded and executed during the Q2, Q3 and Q4 cycles. Data memory is read during Q2 (operand read) and written during Q4 (destination write).



#### FIGURE 6-4: CLOCK/INSTRUCTION CYCLE

#### EXAMPLE 6-3: INSTRUCTION PIPELINE FLOW

-	Тсү0	TCY1	Tcy2	TCY3	TCY4	TcY5
1. MOVLW 55h	etch 1	Execute 1				
2. MOVWF PORTB		Fetch 2	Execute 2		_	
3. BRA SUB_1	_		Fetch 3	Execute 3		
4. BSF PORTA, BIT3 (Force	ed NOP)			Fetch 4	Flush (NOP)	
5. Instruction @ address SU	UB_1				Fetch SUB_1	Execute SUB_1

All instructions are single cycle, except for any program branches. These take two cycles since the fetch instruction is "flushed" from the pipeline while the new instruction is being fetched and then executed.

#### 6.2.3 INSTRUCTIONS IN PROGRAM MEMORY

The program memory is addressed in bytes. Instructions are stored as two or four bytes in program memory. The Least Significant Byte of an instruction word is always stored in a program memory location with an even address (LSB = 0). To maintain alignment with instruction boundaries, the PC increments in steps of two and the LSB will always read '0' (see Section 6.1.2 "Program Counter").

Figure 6-5 shows an example of how instruction words are stored in the program memory.

The CALL and GOTO instructions have the absolute program memory address embedded into the instruction. Since instructions are always stored on word boundaries, the data contained in the instruction is a word address. The word address is written to PC<20:1> which accesses the desired byte address in program memory. Instruction #2 in Figure 6-5 shows how the instruction, GOTO 0006h, is encoded in the program memory. Program branch instructions, which encode a relative address offset, operate in the same manner. The offset value stored in a branch instruction represents the number of single-word instructions that the PC will be offset by. For more details on the instruction set, see **Section 29.0 "Instruction Set Summary"**.

FIGURE 6-5:	INSTRUCTIONS IN PROGRAM MEMORY

			LSB = 1	LSB = 0	Word Address $\downarrow$
	Program N	lemory			000000h
	Byte Locat	ions $\rightarrow$			000002h
					000004h
					000006h
Instruction 1:	MOVLW	055h	0Fh	55h	000008h
Instruction 2:	GOTO	0006h	EFh	03h	00000Ah
			F0h	00h	00000Ch
Instruction 3:	MOVFF	123h, 456h	C1h	23h	00000Eh
			F4h	56h	000010h
					000012h
					000014h

### 6.2.4 TWO-WORD INSTRUCTIONS

The standard PIC18 instruction set has four, two-word instructions: CALL, MOVFF, GOTO and LSFR. In all cases, the second word of the instructions always has '1111' as its four Most Significant bits. The other 12 bits are literal data, usually a data memory address.

The use of '1111' in the 4 MSbs of an instruction specifies a special form of NOP. If the instruction is executed in proper sequence, immediately after the first word, the data in the second word is accessed and

used by the instruction sequence. If the first word is skipped, for some reason, and the second word is executed by itself, a NOP is executed instead. This is necessary for cases when the two-word instruction is preceded by a conditional instruction that changes the PC. Example 6-4 shows how this works.

Note: For information on two-word instructions in the extended instruction set, see Section 6.5 "Program Memory and the Extended Instruction Set".

EXAMPLE 6-4:	TWO-WORD INSTRUCTIONS

CASE 1:		
Object Code	Source Code	
0110 0110 0000 0000	TSTFSZ REG1 ;	is RAM location 0?
1100 0001 0010 0011	MOVFF REG1, REG2 ;	No, skip this word
1111 0100 0101 0110	;	Execute this word as a NOP
0010 0100 0000 0000	ADDWF REG3 ;	continue code
CASE 2:		
Object Code	Source Code	
0110 0110 0000 0000	TSTFSZ REG1 ;	is RAM location 0?
1100 0001 0010 0011	MOVFF REG1, REG2 ;	Yes, execute this word
1111 0100 0101 0110	;	2nd word of instruction
0010 0100 0000 0000	ADDWF REG3 ;	continue code

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#### 6.3 Data Memory Organization

Note:	The operation of some aspects of data								
	memory are changed when the PIC18								
	extended instruction set is enabled. See								
	Section 6.6 "Data Memory and the								
	Extended Instruction Set" for more								
	information.								

The data memory in PIC18 devices is implemented as static RAM. Each register in the data memory has a 12-bit address, allowing up to 4,096 bytes of data memory. The memory space is divided into as many as 16 banks that contain 256 bytes each. PIC18FX6K22 and PIC18FX7K22 devices implement all 16 complete banks, for a total of 4 Kbytes. PIC18FX5K22 devices implement only the first eight complete banks, for a total of 2 Kbytes.

Figure 6-6 and Figure 6-7 show the data memory organization for the devices.

The data memory contains Special Function Registers (SFRs) and General Purpose Registers (GPRs). The SFRs are used for control and status of the controller and peripheral functions, while GPRs are used for data storage and scratchpad operations in the user's application. Any read of an unimplemented location will read as '0's.

The instruction set and architecture allow operations across all banks. The entire data memory may be accessed by Direct, Indirect or Indexed Addressing modes. Addressing modes are discussed later in this section.

To ensure that commonly used registers (select SFRs and select GPRs) can be accessed in a single cycle, PIC18 devices implement an Access Bank. This is a 256-byte memory space that provides fast access to select SFRs and the lower portion of GPR Bank 0 without using the Bank Select Register. For details on the Access RAM, see Section 6.3.2 "Access Bank".

#### 6.3.1 BANK SELECT REGISTER

Large areas of data memory require an efficient addressing scheme to make it possible for rapid access to any address. Ideally, this means that an entire address does not need to be provided for each read or write operation. For PIC18 devices, this is accomplished with a RAM banking scheme. This divides the memory space into 16 contiguous banks of 256 bytes. Depending on the instruction, each location can be addressed directly by its full 12-bit address, or an 8-bit, low-order address and a 4-bit Bank Pointer.

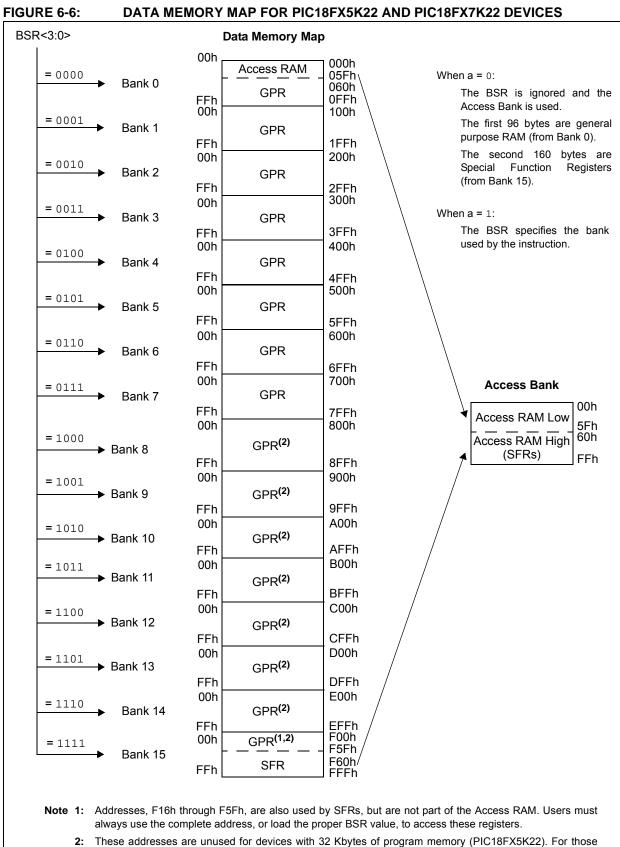
Most instructions in the PIC18 instruction set make use of the Bank Pointer, known as the Bank Select Register (BSR). This SFR holds the four Most Significant bits of a location's address. The instruction itself includes the eight Least Significant bits. Only the four lower bits of the BSR are implemented (BSR<3:0>). The upper four bits are unused, always read as '0' and cannot be written to. The BSR can be loaded directly by using the MOVLB instruction.

The value of the BSR indicates the bank in data memory. The eight bits in the instruction show the location in the bank and can be thought of as an offset from the bank's lower boundary. The relationship between the BSR's value and the bank division in data memory is shown in Figure 6-7.

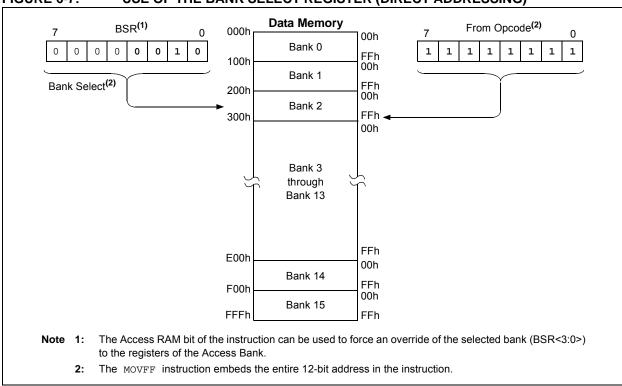
Since up to 16 registers may share the same low-order address, the user must always be careful to ensure that the proper bank is selected before performing a data read or write. For example, writing what should be program data to an eight-bit address of F9h, while the BSR is 0Fh, will end up resetting the Program Counter.

While any bank can be selected, only those banks that are actually implemented can be read or written to. Writes to unimplemented banks are ignored, while reads from unimplemented banks will return '0's. Even so, the STATUS register will still be affected as if the operation was successful. The data memory map in Figure 6-6 indicates which banks are implemented.

In the core PIC18 instruction set, only the MOVFF instruction fully specifies the 12-bit address of the source and target registers. When this instruction executes, it ignores the BSR completely. All other instructions include only the low-order address as an operand and must use either the BSR or the Access Bank to locate their target registers.



 These addresses are unused for devices with 32 Kbytes of program memory (PIC18FX5K22). For those devices, read these addresses at 00h.



#### FIGURE 6-7: USE OF THE BANK SELECT REGISTER (DIRECT ADDRESSING)

#### 6.3.2 ACCESS BANK

While the use of the BSR, with an embedded 8-bit address, allows users to address the entire range of data memory, it also means that the user must ensure that the correct bank is selected. If not, data may be read from, or written to, the wrong location. This can be disastrous if a GPR is the intended target of an operation, but an SFR is written to instead. Verifying and/or changing the BSR for each read or write to data memory can become very inefficient.

To streamline access for the most commonly used data memory locations, the data memory is configured with an Access Bank, which allows users to access a mapped block of memory without specifying a BSR. The Access Bank consists of the first 96 bytes of memory (00h-5Fh) in Bank 0 and the last 160 bytes of memory (60h-FFh) in Bank 15. The lower half is known as the "Access RAM" and is composed of GPRs. The upper half is where the device's SFRs are mapped. These two areas are mapped contiguously in the Access Bank and can be addressed in a linear fashion by an eight-bit address (Figure 6-6).

The Access Bank is used by core PIC18 instructions that include the Access RAM bit (the 'a' parameter in the instruction). When 'a' is equal to '1', the instruction uses the BSR and the 8-bit address included in the opcode for the data memory address. When 'a' is '0',

however, the instruction is forced to use the Access Bank address map. In that case, the current value of the BSR is ignored entirely.

Using this "forced" addressing allows the instruction to operate on a data address in a single cycle without updating the BSR first. For 8-bit addresses of 60h and above, this means that users can evaluate and operate on SFRs more efficiently. The Access RAM below 60h is a good place for data values that the user might need to access rapidly, such as immediate computational results or common program variables.

Access RAM also allows for faster and more code efficient context saving and switching of variables.

The mapping of the Access Bank is slightly different when the extended instruction set is enabled (XINST Configuration bit = 1). This is discussed in more detail in Section 6.6.3 "Mapping the Access Bank in Indexed Literal Offset Mode".

#### 6.3.3 GENERAL PURPOSE REGISTER FILE

PIC18 devices may have banked memory in the GPR area. This is data RAM which is available for use by all instructions. GPRs start at the bottom of Bank 0 (address 000h) and grow upwards towards the bottom of the SFR area. GPRs are not initialized by a Power-on Reset and are unchanged on all other Resets.

#### 6.3.4 SPECIAL FUNCTION REGISTERS

The Special Function Registers (SFRs) are registers used by the CPU and peripheral modules for controlling the desired operation of the device. These registers are implemented as static RAM. SFRs start at the top of data memory (FFFh) and extend downward to occupy all of Bank 15 (F00h to FFFh) and the top part of Bank 14 (EF4h to EFFh).

A list of these registers is given in Table 6-1 and Table 6-2.

The SFRs can be classified into two sets: those associated with the "core" device functionality (ALU, Resets and interrupts) and those related to the peripheral functions. The Reset and Interrupt registers are described in their respective chapters, while the ALU's STATUS register is described later in this section. Registers related to the operation of the peripheral features are described in the chapter for that peripheral.

The SFRs are typically distributed among the peripherals whose functions they control. Unused SFR locations are unimplemented and read as '0's.

Addr.	Name	Addr.	Name	Addr.	Name	Addr.	Name	Addr.	Name	Addr.	Name <sup>(4)</sup>
FFFh	TOSU	FDFh	INDF2 <sup>(1)</sup>	FBFh	ECCP1AS	F9Fh	IPR1	F7Fh	EECON1	F5Fh	RTCCFG
FFEh	TOSH	FDEh	POSTINC2 <sup>(1)</sup>	FBEh	ECCP1DEL	F9Eh	PIR1	F7Eh	EECON2	F5Eh	RTCCAL
FFDh	TOSL	FDDh	POSTDEC2 <sup>(1)</sup>	FBDh	CCPR1H	F9Dh	PIE1	F7Dh	TMR5H	F5Dh	RTCVALH
FFCh	STKPTR	FDCh	PREINC2 <sup>(1)</sup>	FBCh	CCPR1L	F9Ch	PSTR1CON	F7Ch	TMR5L	F5Ch	RTCVALL
FFBh	PCLATU	FDBh	PLUSW2 <sup>(1)</sup>	FBBh	CCP1CON	F9Bh	OSCTUNE	F7Bh	T5CON	F5Bh	ALRMCFG
FFAh	PCLATH	FDAh	FSR2H	FBAh	PIR5	F9Ah	TRISJ <sup>(2)</sup>	F7Ah	T5GCON	F5Ah	ALRMRPT
FF9h	PCL	FD9h	FSR2L	FB9h	PIE5	F99h	TRISH <sup>(2)</sup>	F79h	CCPR4H	F59h	ALRMVALH
FF8h	TBLPTRU	FD8h	STATUS	FB8h	IPR4	F98h	TRISG	F78h	CCPR4L	F58h	ALRMVALL
FF7h	TBLPTRH	FD7h	TMR0H	FB7h	PIR4	F97h	TRISF	F77h	CCP4CON	F57h	CTMUCONH
FF6h	TBLPTRL	FD6h	TMR0L	FB6h	PIE4	F96h	TRISE	F76h	CCPR5H	F56h	CTMUCONL
FF5h	TABLAT	FD5h	TOCON	FB5h	CVRCON	F95h	TRISD	F75h	CCPR5L	F55h	CTMUICONH
FF4h	PRODH	FD4h	SPBRGH1	FB4h	CMSTAT	F94h	TRISC	F74h	CCP5CON	F54h	CM1CON
FF3h	PRODL	FD3h	OSCCON	FB3h	TMR3H	F93h	TRISB	F73h	CCPR6H	F53h	PADCFG1
FF2h	INTCON	FD2h	IPR5	FB2h	TMR3L	F92h	TRISA	F72h	CCPR6L	F52h	ECCP2AS
FF1h	INTCON2	FD1h	WDTCON	FB1h	T3CON	F91h	LATJ <sup>(2)</sup>	F71h	CCP6CON	F51h	ECCP2DEL
FF0h	INTCON3	FD0h	RCON	FB0h	T3GCON	F90h	LATH <sup>(2)</sup>	F70h	CCPR7H	F50h	CCPR2H
FEFh	INDF0 <sup>(1)</sup>	FCFh	TMR1H	FAFh	SPBRG1	F8Fh	LATG	F6Fh	CCPR7L	F4Fh	CCPR2L
	POSTINC0 <sup>(1)</sup>	FCEh	TMR1L	FAEh	RCREG1	F8Eh	LATF	F6Eh	CCP7CON	F4Eh	CCP2CON
FEDh	POSTDEC0 <sup>(1)</sup>	FCDh	T1CON	FADh	TXREG1	F8Dh	LATE	F6Dh	TMR4	F4Dh	ECCP3AS
FECh	PREINC0 <sup>(1)</sup>	FCCh	TMR2	FACh	TXSTA1	F8Ch	LATD	F6Ch	PR4	F4Ch	ECCP3DEL
FEBh	PLUSW0 <sup>(1)</sup>	FCBh	PR2	FABh	RCSTA1	F8Bh	LATC	F6Bh	T4CON	F4Bh	CCPR3H
FEAh	FSR0H	FCAh	T2CON	FAAh	T1GCON	F8Ah	LATB	F6Ah	SSP2BUF	F4Ah	CCPR3L
FE9h	FSR0L	FC9h	SSP1BUF	FA9h	IPR6	F89h	LATA	F69h	SSP2ADD	F49h	CCP3CON
FE8h	WREG	FC8h	SSP1ADD	FA8h	HLVDCON	F88h	PORTJ <sup>(2)</sup>	F68h	SSP2STAT	F48h	CCPR8H
FE7h	INDF1 <sup>(1)</sup>	FC7h	SSP1STAT	FA7h	PSPCON	F87h	PORTH <sup>(2)</sup>	F67h	SSP2CON1	F47h	CCPR8L
FE6h	POSTINC1 <sup>(1)</sup>	FC6h	SSP1CON1	FA6h	PIR6	F86h	PORTG	F66h	SSP2CON2	F46h	CCP8CON
FE5h	POSTDEC1 <sup>(1)</sup>	FC5h	SSP1CON2	FA5h	IPR3	F85h	PORTF	F65h	BAUDCON1	F45h	CCPR9H <sup>(3)</sup>
FE4h	PREINC1 <sup>(1)</sup>	FC4h	ADRESH	FA4h	PIR3	F84h	PORTE	F64h	OSCCON2	F44h	CCPR9L <sup>(3)</sup>
FE3h	PLUSW1 <sup>(1)</sup>	FC3h	ADRESL	FA3h	PIE3	F83h	PORTD	F63h	EEADRH	F43h	CCP9CON <sup>(3)</sup>
FE2h	FSR1H	FC2h	ADCON0	FA2h	IPR2	F82h	PORTC	F62h	EEADR	F42h	CCPR10H <sup>(3)</sup>
FE1h	FSR1L	FC1h	ADCON1	FA1h	PIR2	F81h	PORTB	F61h	EEDATA	F41h	
FE0h	BSR	FC0h	ADCON2	FA0h	PIE2	F80h	PORTA	F60h	PIE6	F40h	CCP10CON <sup>(3)</sup>

 TABLE 6-1:
 SPECIAL FUNCTION REGISTER MAP FOR PIC18F87K22 FAMILY

**Note 1:** This is not a physical register.

2: Unimplemented on 64-pin devices (PIC18F6XK22), read as '0'.

3: This register is not available on devices with a program memory of 32 Kbytes (PIC18FX5K22).

4: Addresses, F16h through F5Fh, are also used by SFRs, but are not part of the Access RAM. To access these registers, users must always load the proper BSR value.

#### TABLE 6-1: SPECIAL FUNCTION REGISTER MAP FOR PIC18F87K22 FAMILY (CONTINUED)

Addr.	Name	Addr.	Name	Addr.	Name	Addr.	Name
F3Fh	TMR7H <sup>(3)</sup>	F32h	TMR12 <sup>(3)</sup>	F25h	ANCON0	F18h	PMD1
F3Eh	TMR7L <sup>(3)</sup>	F31h	PR12 <sup>(3)</sup>	F24h	ANCON1	F17h	PMD2
F3Dh	T7CON <sup>(3)</sup>	F30h	T12CON <sup>(3)</sup>	F23h	ANCON2	F16h	PMD3
F3Ch	T7GCON <sup>(3)</sup>	F2Fh	CM2CON	F22h	RCSTA2	_	
F3Bh	TMR6	F2Eh	CM3CON	F21h	TXSTA2		
F3Ah	PR6	F2Dh	CCPTMRS0	F20h	BAUDCON2		
F39H	T6CON	F2Ch	CCPTMRS1	F1Fh	SPBRGH2		
F38h	TMR8	F2Bh	CCPTMRS2	F1Eh	SPBRG2		
F37h	PR8	F2Ah	REFOCON	F1Dh	RCREG2		
F36h	T8CON	F29H	ODCON1	F1Ch	TXREG2		
F35h	TMR10 <sup>(3)</sup>	F28h	ODCON2	F1Bh	PSTR2CON		
F34h	PR10 <sup>(3)</sup>	F27h	ODCON3	F1Ah	PSTR3CON		
F33h	T10CON <sup>(3)</sup>	F26h	MEMCON <sup>(3)</sup>	F19h	PMD0		

**Note 1:** This is not a physical register.

2: Unimplemented on 64-pin devices (PIC18F6XK22), read as '0'.

- 3: This register is not available on devices with a program memory of 32 Kbytes (PIC18FX5K22).
- 4: Addresses, F16h through F5Fh, are also used by SFRs, but are not part of the Access RAM. To access these registers, users must always load the proper BSR value.

Address	File Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Value on POR, BOR
FFFh	TOSU	_	_	_	Top-of-Stack U	pper Byte (TO	S<20:16>)			0 0000
FFEh	TOSH	Top-of-Stack H	ligh Byte (TOS	S<15:8>)						0000 0000
FFDh	TOSL	Top-of-Stack L	ow Byte (TOS	<7:0>)						0000 0000
FFCh	STKPTR	STKFUL	STKUNF	_	Return Stack F	Pointer				uu-0 0000
FFBh	PCLATU	—	_	_	Holding Regist	er for PC<20:1	6>			0 0000
FFAh	PCLATH	Holding Regis	ter for PC<15:	8>						0000 0000
FF9h	PCL	PC Low Byte	(PC<7:0>)							0000 0000
FF8h	TBLPTRU	_	_	bit 21	Program Mem	ory Table Point	ter Upper Byte	(TBLPTR<20:	16>)	00 0000
FF7h	TBLPTRH	Program Mem	ory Table Poin	ter High Byte	(TBLPTR<15:8	>)				0000 0000
FF6h	TBLPTRL	Program Mem	ory Table Poin	nter Low Byte (	TBLPTR<7:0>)					0000 0000
FF5h	TABLAT	Program Mem	Program Memory Table Latch							
FF4h	PRODH	Product Regis	ter High Byte							XXXX XXXX
FF3h	PRODL	Product Regis	ter Low Byte							XXXX XXXX
FF2h	INTCON	GIE/GIEH	PEIE/GIEL	TMR0IE	<b>INTOIE</b>	RBIE	TMR0IF	INTOIF	RBIF	0000 000x
FF1h	INTCON2	RBPU	INTEDG0	INTEDG1	INTEDG2	INTEDG3	TMR0IP	INT3IP	RBIP	1111 1111
FF0h	INTCON3	INT2IP	INT1IP	INT3IE	INT2IE	INT1IE	INT3IF	INT2IF	INT1IF	1100 0000
FEFh	INDF0	Uses contents	of FSR0 to ac	dress data me	emory – value o	f FSR0 not cha	anged (not a pł	nysical register	)	
FEEh	POSTINC0	Uses contents	of FSR0 to ac	ldress data me	emory – value o	f FSR0 post-in	cremented (no	t a physical reg	gister)	
FEDh	POSTDEC0	Uses contents	of FSR0 to ac	dress data me	emory – value o	f FSR0 post-de	ecremented (no	ot a physical re	gister)	
FECh	PREINC0	Uses contents	of FSR0 to ac	ldress data me	emory – value o	f FSR0 pre-inc	remented (not	a physical regi	ster)	
FEBh	PLUSW0	Uses contents FSR0 offset by		ldress data me	emory – value of	FSR0 pre-incr	remented (not a	a physical regis	ster) – value of	
FEAh	FSR0H	_	_	_	_	Indirect Data	Memory Addre	ess Pointer 0 H	igh	0000
FE9h	FSR0L	Indirect Data N	lemory Addre	ss Pointer 0 Lo	ow Byte					XXXX XXXX
FE8h	WREG	Working Regis	ster							XXXX XXXX
FE7h	INDF1	Uses contents	of FSR1 to ac	dress data me	emory – value o	f FSR1 not cha	anged (not a pl	nysical register	)	

#### TABLE 6-2: PIC18F87K22 FAMILY REGISTER FILE SUMMARY

Note 1: This bit is available when Master Clear is disabled (MCLRE = 0). When MCLRE is set, the bit is unimplemented.

2: Unimplemented on 64-pin devices (PIC18F6XK22), read as '0'.

3: Unimplemented on devices with a program memory of 32 Kbytes (PIC18FX5K22).

Address	File Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Value on POR, BOR
FE6h	POSTINC1	Uses contents	of FSR1 to ad	ldress data me	emory – value o	f FSR1 post-in	cremented (no	t a physical reg	jister)	
FE5h	POSTDEC1	Uses contents	of FSR1 to ac	ldress data me	emory – value o	f FSR1 post-de	ecremented (no	ot a physical re	gister)	
FE4h	PREINC1	Uses contents	of FSR1 to ac	ldress data me	emory – value o	f FSR1 pre-inc	remented (not	a physical regi	ster)	
FE3h	PLUSW1	Uses contents FSR1 offset by		dress data me	mory – value of	FSR1 pre-incr	remented (not a	a physical regis	ster) – value of	
FE2h	FSR1H	—	—	—	—	Indirect Data	Memory Addre	ss Pointer 1 H	igh	xxxx
FE1h	FSR1L	Indirect Data N	Memory Addres	ss Pointer 1 Lo	ow Byte					XXXX XXXX
FE0h	BSR	—	—	—	_	Bank Select F	Register			0000
FDFh	INDF2	Uses contents	of FSR2 to ac	ldress data me	emory – value o	f FSR2 not cha	anged (not a pł	nysical register	)	
FDEh	POSTINC2	Uses contents	of FSR2 to ac	ldress data me	emory – value o	f FSR2 post-in	cremented (no	t a physical reg	jister)	
FDDh	POSTDEC2	Uses contents	of FSR2 to ad	ldress data me	emory – value o	f FSR2 post-de	ecremented (no	ot a physical re	gister)	
FDCh	PREINC2	Uses contents	of FSR2 to ad	ldress data me	emory – value o	f FSR2 pre-inc	remented (not	a physical regi	ster)	
FDBh	PLUSW2	Uses contents FSR2 offset by		ldress data me	mory – value of	FSR2 pre-incr	remented (not a	a physical regis	ster) – value of	
FDAh	FSR2H	_	_	_	_	Indirect Data	Memory Addre	ss Pointer 2 H	igh	xxxx
FD9h	FSR2L	Indirect Data M	Aemory Addres	ss Pointer 2 Lo	ow Byte					XXXX XXXX
FD8h	STATUS	_	_	_	Ν	OV	Z	DC	С	x xxxx
FD7h	TMR0H	Timer0 Regist	er High Byte							0000 0000
FD6h	TMR0L	Timer0 Regist	er Low Byte							xxxx xxxx
FD5h	T0CON	TMR00N	T08BIT	T0CS	T0SE	PSA	TOPS2	TOPS1	TOPS0	1111 1111
FD4h	SPBRGH1	USART1 Bau	d Rate Genera	tor High Byte						0000 0000
FD3h	OSCCON	IDLEN	IRCF2	IRCF1	IRCF0	OSTS	HFIOFS	SCS1	SCS0	0110 q000
FD2h	IPR5	TMR7GIP <sup>(3)</sup>	TMR12IP <sup>(3)</sup>	TMR10IP <sup>(3)</sup>	TMR8IP	TMR7IP <sup>(3)</sup>	TMR6IP	TMR5IP	TMR4IP	1111 1111
FD1h	WDTCON	REGSLP	_	ULPLVL	SRETEN	_	ULPEN	ULPSINK	SWDTEN	0-x0 -000
FD0h	RCON	IPEN	SBOREN	CM	RI	TO	PD	POR	BOR	0111 11qq
FCFh	TMR1H	Timer1 Regist		OW	Tu .	10	TD	TOR	BOIL	xxxx xxxx
FCEh	TMR1L	Timer1 Regist	• •							XXXX XXXX
FCDh	T1CON	TMR1CS1	TMR1CS0	T1CKPS1	T1CKPS0	SOSCEN	T1SYNC	RD16	TMR10N	
	TMR2			TICKPST	TICKP50	SUSCEN	TISTING	RD10	TIVIRIUN	0000 0000
FCCh	PR2	Timer2 Regist								0000 0000
FCBh		Timer2 Period	T2OUTPS3	T2OUTPS2			TMDOON	T2CKPS1	TOOLDOO	1111 1111
FCAh	T2CON				T2OUTPS1	T2OUTPS0	TMR2ON	1206951	T2CKPS0	-000 0000
FC9h	SSP1BUF		e Buffer/Trans	U	1. 00D4 D		D	M		XXXX XXXX
FC8h	SSP1ADD			_	ode. SSP1 Bau	1				0000 0000
FC7h	SSP1STAT	SMP	CKE	D/A	Р	S	R/W	UA	BF	0000 0000
FC6h	SSP1CON1	WCOL	SSPOV	SSPEN	CKP	SSPM3	SSPM2	SSPM1	SSPM0	0000 0000
FC5h	SSP1CON2	GCEN	ACKSTAT	ACKDT	ACKEN	RCEN	PEN	RSEN	SEN	0000 0000
FC4h	ADRESH		gister High By							XXXX XXXX
FC3h	ADRESL	A/D Result Re	gister Low Byt	е		1	r	r		XXXX XXXX
FC2h	ADCON0	—	CHS4	CHS3	CHS2	CHS1	CHS0	GO/DONE	ADON	-000 0000
FC1h	ADCON1	TRIGSEL1	TRIGSEL0	VCFG1	VCFG0	VNCFG	CHSN2	CHSN1	CHSN0	0000 0000
FC0h	ADCON2	ADFM	—	ACQT2	ACQT1	ACQT0	ADCS2	ADCS1	ADCS0	0-00 0000
FBFh	ECCP1AS	ECCP1ASE	ECCP1AS2	ECCP1AS1	ECCP1AS0	PSS1AC1	PSS1AC0	PSS1BD1	PSS1BD0	0000 0000
FBEh	ECCP1DEL	P1RSEN	P1DC6	P1DC5	P1DC4	P1DC3	P1DC2	P1DC1	P1DC0	0000 0000
FBDh	CCPR1H	Capture/Comp	bare/PWM Reg	gister1 High By	/te					XXXX XXXX
FBCh	CCPR1L	Capture/Comp	bare/PWM Reg	gister1 Low By	te	1		1		XXXX XXXX
FBBh	CCP1CON	P1M1	P1M0	DC1B1	DC1B0	CCP1M3	CCP1M2	CCP1M1	CCP1M0	0000 0000
FBAh	PIR5	TMR7GIF <sup>(3)</sup>	TMR12IF <sup>(3)</sup>	TMR10IF <sup>(3)</sup>	TMR8IF	TMR7IF <sup>(3)</sup>	TMR6IF	TMR5IF	TMR4IF	0000 0000
FB9h	PIE5	TMR7GIE <sup>(3)</sup>	TMR12IE <sup>(3)</sup>	TMR10IE <sup>(3)</sup>	TMR8IE	TMR7IE <sup>(3)</sup>	TMR6IE	TMR5IE	TMR4IE	0000 0000
FB8h	IPR4	CCP10IP <sup>(3)</sup>	CCP9IP <sup>(3)</sup>	CCP8IP	CCP7IP	CCP6IP	CCP5IP	CCP4IP	CCP3IP	1111 1111
FB7h	PIR4	CCP10IF <sup>(3)</sup>	CCP9IF <sup>(3)</sup>	CCP8IF	CCP7IF	CCP6IF	CCP5IF	CCP4IF	CCP3IF	0000 0000

#### PIC18E87K22 FAMILY REGISTER FILE SUMMARY (CONTINUED) TABLE 6-2

This bit is available when Master Clear is disabled (MCLRE = 0). When MCLRE is set, the bit is unimplemented. Unimplemented on 64-pin devices (PIC18F6XK22), read as '0'. Unimplemented on devices with a program memory of 32 Kbytes (PIC18FX5K22). Note 1:

2: 3:

TABLE 6-2: PIC18F87K22 FAMILY REGISTER FILE SUMMARY (CONTIN
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Address	File Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Value on POR, BOR
FB6h	PIE4	CCP10IE <sup>(3)</sup>	CCP9IE <sup>(3)</sup>	CCP8IE	CCP7IE	CCP6IE	CCP5IE	CCP4IE	CCP3IE	0000 0000
FB5h	CVRCON	CVREN	CVROE	CVRSS	CVR4	CVR3	CVR2	CVR1	CVR0	0000 0000
FB4h	CMSTAT	CMP3OUT	CMP2OUT	CMP1OUT	_		_		_	xxx
FB3h	TMR3H	Timer3 Regist	er High Byte							XXXX XXXX
FB2h	TMR3L	Timer3 Regist	er Low Byte							XXXX XXXX
FB1h	T3CON	TMR3CS1	TMR3CS0	T3CKPS1	T3CKPS0	SOSCEN	T3SYNC	RD16	TMR3ON	0000 0000
FB0h	T3GCON	TMR3GE	T3GPOL	T3GTM	T3GSPM	T3GGO/ T3DONE	T3GVAL	T3GSS1	T3GSS0	00x0 0x00
FAFh	SPBRG1	USART1 Bau	RT1 Baud Rate Generator							
FAEh	RCREG1	USART1 Rece	eive Register							0000 0000
FADh	TXREG1	USART1 Tran	smit Register							XXXX XXXX
FACh	TXSTA1	CSRC	TX9	TXEN	SYNC	SENDB	BRGH	TRMT	TX9D	0000 0010
FABh	RCSTA1	SPEN	RX9	SREN	CREN	ADDEN	FERR	OERR	RX9D	0000 000x
FAAh	T1GCON	TMR1GE	T1GPOL	T1GTM	T1GSPM	T1GGO/ T1DONE	T1GVAL	T1GSS1	T1GSS0	00x0 0x00
FA9h	IPR6	—	—		EEIP		CMP3IP	CMP2IP	CMP1IP	1 -111
FA8h	HLVDCON	VDIRMAG	BGVST	IRVST	HLVDEN	HLVDL3	HLVDL2	HLVDL1	HLVDL0	0000 0000
FA7h	PSPCON	IBF	OBF	IBOV	PSPMODE	_	_	_	_	0000
FA6h	PIR6	_	_	_	EEIF	_	CMP3IF	CMP2IF	CMP1IF	0 -000
FA5h	IPR3	TMR5GIP	_	RC2IP	TX2IP	CTMUIP	CCP2IP	CCP1IP	RTCCIP	1-11 1111
FA4h	PIR3	TMR5GiF	_	RC2IF	TX2IF	CTMUIF	CCP2IF	CCP1IF	RTCCIF	0-00 0000
FA3h	PIE3	TMR5GIE	_	RC2IE	TX2IE	CTMUIE	CCP2IE	CCP1IE	RTCCIE	0-00 0000
FA2h	IPR2	OSCFIP	—	SSP2IP	BCL2IP	BCL1IP	HLVDIP	TMR3IP	TMR3GIP	1-11 1111
FA1h	PIR2	OSCFIF	—	SSP2IF	BCL2IF	BCL1IF	HLVDIF	TMR3IF	TMR3GIF	0-00 0000
FA0h	PIE2	OSCFIE	_	SSP2IE	BCL2IE	BCL1IE	HLVDIE	TMR3IE	TMR3GIE	0-00 0000
F9Fh	IPR1	PSPIP	ADIP	RC1IP	TX1IP	SSP1IP	TMR1GIP	TMR2IP	TMR1IP	1111 1111
F9Eh	PIR1	PSPIF	ADIF	RC1IF	TX1IF	SSP1IF	TMR1GIF	TMR2IF	TMR1IF	0000 0000
F9Dh	PIE1	PSPIE	ADIE	RC1IE	TX1IE	SSP1IE	TMR1GIE	TMR2IE	TMR1IE	0000 0000
F9Ch	PSTR1CON	CMPL1	CMPL0	_	STRSYNC	STRD	STRC	STRB	STRA	00-0 0001
F9Bh	OSCTUNE	INTSRC	PLLEN	TUN5	TUN4	TUN3	TUN2	TUN1	TUN0	0000 0000
F9Ah	TRISJ <sup>(2)</sup>	TRISJ7	TRISJ6	TRISJ5	TRISJ4	TRISJ3	TRISJ2	TRISJ1	TRISJ0	1111 1111
F99h	TRISH <sup>(2)</sup>	TRISH7	TRISH6	TRISH5	TRISH4	TRISH3	TRISH2	TRISH1	TRISH0	1111 1111
F98h	TRISG	_	_	_	TRISG4	TRISG3	TRISG2	TRISG1	TRISG0	1 1111
F97h	TRISF	TRISF7	TRISF6	TRISF5	TRISF4	TRISF3	TRISF2	TRISF1	_	1111 111-
F96h	TRISE	TRISE7	TRISE6	TRISE5	TRISE4	TRISE3	TRISE2	TRISE1	_	1111 111-
F95h	TRISD	TRISD7	TRISD6	TRISD5	TRISD4	TRISD3	TRISD2	TRISD1	TRISD0	1111 1111
F94h	TRISC	TRISC7	TRISC6	TRISC5	TRISC4	TRISC3	TRISC2	TRISC1	TRISC0	1111 1111
F93h	TRISB	TRISB7	TRISB6	TRISB5	TRISB4	TRISB3	TRISB2	TRISB1	TRISB0	1111 1111
F92h	TRISA	TRISA7	TRISA6	TRISA5	TRISA4	TRISA3	TRISA2	TRISA1	TRISA0	1111 1111
F91h	LATJ <sup>(2)</sup>	LATJ7	LATJ6	LATJ5	LATJ4	LATJ3	LATJ2	LATJ1	LATJ0	xxxx xxxx
F90h	LATH <sup>(2)</sup>	LATH7	LATH6	LATH5	LATH4	LATH3	LATH2	LATH1	LATH0	xxxx xxxx
F8Fh	LATG	_	_	_	LATG4	LATG3	LATG2	LATG1	LATG0	x xxxx
F8Eh	LATF	LATF7	LATF6	LATF5	LATF4	LATF3	LATF2	LATF1	—	xxxx xxx-
F8Dh	LATE	LATE7	LATE6	LATE5	LATE4	LATE3	LATE2	LATE1	LATE0	xxxx xxxx
F8Ch	LATD	LATD7	LATD6	LATD5	LATD4	LATD3	LATD2	LATD1	LATD0	xxxx xxxx
F8Bh	LATC	LATC7	LATC6	LATC5	LATC4	LATC3	LATC2	LATC1	LATCO	xxxx xxxx
F8Ah	LATB	LATB7	LATB6	LATB5	LATB4	LATB3	LATB2	LATB1	LATB0	XXXX XXXX
F89h	LATA	LATA7	LATA6	LATA5	LATA4	LATA3	LATA2	LATA1	LATAO	XXXX XXXX
F88h	PORTJ <sup>(2)</sup>	RJ7	RJ6	RJ5	RJ4	RJ3	RJ2	RJ1	RJ0	XXXX XXXX
F87h	PORTH <sup>(2)</sup>	RH7	RH6	RH5	RH4	RH3	RH2	RH1	RH0	XXXX XXXX

This bit is available when Master Clear is disabled (MCLRE = 0). When MCLRE is set, the bit is unimplemented. Unimplemented on 64-pin devices (PIC18F6XK22), read as '0'. Unimplemented on devices with a program memory of 32 Kbytes (PIC18FX5K22). Note 1: 2: 3:

Address	File Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Value on POR, BOR
F86h	PORTG	_	_	RG5 <sup>(1)</sup>	RG4	RG3	RG2	RG1	RG0	xx xxxx
F85h	PORTF	RF7	RF6	RF5	RF4	RF3	RF2	RF1	_	xxxx xxx-
F84h	PORTE	RE7	RE6	RE5	RE4	RE3	RE2	RE1	RE0	XXXX XXXX
F83h	PORTD	RD7	RD6	RD5	RD4	RD3	RD2	RD1	RD0	XXXX XXXX
F82h	PORTC	RC7	RC6	RC5	RC4	RC3	RC2	RC1	RC0	XXXX XXXX
F81h	PORTB	RB7	RB6	RB5	RB4	RB3	RB2	RB1	RB0	XXXX XXXX
F80h	PORTA	RA7	RA6	RA5	RA4	RA3	RA2	RA1	RA0	XXXX XXXX
F7Fh	EECON1	EEPGD	CFGS	_	FREE	WRERR	WREN	WR	RD	xx-0 x000
F7Eh	EECON2	EEPROM Cor	ntrol Register 2	(not a physica	al register)					
F7Dh	TMR5H	Timer5 Regist	er High Byte							XXXX XXXX
F7Ch	TMR5L	Timer5 Regist	er Low Byte							xxxx xxxx
F7Bh	T5CON	TMR5CS1	TMR5CS0	T5CKPS1	T5CKPS0	SOSCEN	T5SYNC	RD16	TMR5ON	0000 0000
F7Ah	T5GCON	TMR5GE	T5GPOL	T5GTM	T5GSPM	T5GGO/ T5DONE	T5GVAL	T5GSS1	T5GSS0	0000 0x00
F79h	CCPR4H	Capture/Comp	bare/PWM Reg	gister 4 High B	yte					xxxx xxxx
F78h	CCPR4L	Capture/Comp	bare/PWM Reg	gister 4 Low By	/te					xxxx xxxx
F77h	CCP4CON	_	—	DC4B1	DC4B0	CCP4M3	CCP4M2	CCP4M1	CCP4M0	00 0000
F76h	CCPR5H	Capture/Compare/PWM Register 5 High Byte							xxxx xxxx	
F75h	CCPR5L	Capture/Com	oare/PWM Reg	gister 5 Low By	/te					xxxx xxxx
F74h	CCP5CON	_	_	DC5B1	DC5B0	CCP5M3	CCP5M2	CCP5M1	CCP5M0	00 0000
F73h	CCPR6H	Capture/Com	Capture/Compare/PWM Register 6 High Byte							xxxx xxxx
F72h	CCPR6L	Capture/Com	oare/PWM Reg	gister 6 Low By	/te					xxxx xxxx
F71h	CCP6CON	_	_	DC6B1	DC6B0	CCP6M3	CCP6M2	CCP6M1	CCP6M0	00 0000
F70h	CCPR7H	Capture/Comp	oare/PWM Reg	gister 7 High B	yte					XXXX XXXX
F6Fh	CCPR7L	Capture/Com	bare/PWM Reg	gister 7 Low By	/te					xxxx xxxx
F6Eh	CCP7CON	_	_	DC7B1	DC7B0	CCP7M3	CCP7M2	CCP7M1	CCP7M0	00 0000
F6Dh	TMR4	Timer4 Regist	er		•					xxxx xxxx
F6Ch	PR4	Timer4 Period	Register							1111 1111
F6Bh	T4CON	_	T4OUTPS3	T4OUTPS2	T4OUTPS1	T4OUTPS0	TMR4ON	T4CKPS1	T4CKPS0	-111 1111
F6Ah	SSP2BUF	MSSP Receiv	e Buffer/Trans	mit Register	•					XXXX XXXX
F69h	SSP2ADD	MSSP Addres	s Register in l	<sup>2</sup> C <sup>™</sup> Slave Mo	de. MSSP1 Ba	ud Rate Reloa	d Register in I <sup>2</sup>	C Master Mod	e.	0000 0000
F68h	SSP2STAT	SMP	CKE	D/Ā	Р	S	R/W	UA	BF	0000 0000
F67h	SSP2CON1	WCOL	SSPOV	SSPEN	CKP	SSPM3	SSPM2	SSPM1	SSPM0	0000 0000
F66h	SSP2CON2	GCEN	ACKSTAT	ACKDT	ACKEN	RCEN	PEN	RSEN	SEN	0100 0000
F65h	BAUDCON1	ABDOVF	RCIDL	RXDTP	TXCKP	BRG16	_	WUE	ABDEN	0100 0-00
F64h	OSCCON2	_	SOSCRUN	_	_	SOSCGO	_	MFIOFS	MFIOSEL	-0 0-x0
F63h	EEADRH	EEPROM Address Register High Byte							0000 0000	
F62h	EEADR	EEPROM Address Register Low Byte							0000 0000	
F61h	EEDATA	EEPROM Data Register							0000 0000	
F60h	PIE6	_	_	_	EEIE	_	CMP3IE	CMP2IE	CMP1IE	0 -000
F5Fh	RTCCFG	RTCEN	_	RTCWREN	RTCSYNC	HALFSEC	RTCOE	RTCPTR1	RTCPTR0	0-00 0000
F5Eh	RTCCAL	CAL7	CAL6	CAL5	CAL4	CAL3	CAL2	CAL1	CAL0	0000 0000
F5Dh	RTCVALH	RTCC Value H	ligh Register V	Vindow Based	on RTCPTR<1	:0>			•	xxxx xxxx

#### PIC18F87K22 FAMILY REGISTER FILE SUMMARY (CONTINUED) **TABLE 6-2**:

This bit is available when Master Clear is disabled (MCLRE = 0). When MCLRE is set, the bit is unimplemented. Unimplemented on 64-pin devices (PIC18F6XK22), read as '0'. Unimplemented on devices with a program memory of 32 Kbytes (PIC18FX5K22). Note 1: 2: 3:

Address	File Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Value on POR, BOR
F5Ch	RTCVALL	RTCC Value L	ow Register W	Vindow Based	on RTCPTR<1	:0>				0000 0000
F5Bh	ALRMCFG	ALRMEN	CHIME	AMASK3	AMASK2	AMASK1	AMASK0	ALRMPTR1	ALRMPTR0	0000 0000
F5Ah	ALRMRPT	ARPT7	ARPT6	ARPT5	ARPT4	ARPT3	ARPT2	ARPT1	ARPT0	0000 0000
F59h	ALRMVALH	Alarm Value H	ligh Register V	Vindow Based	on APTR<1:0>					XXXX XXXX
F58h	ALRMVALL	Alarm Value L	ow Register W	/indow Based	on APTR<1:0>					xxxx xxxx
F57h	CTMUCONH	CTMUEN	_	CTMUSIDL	TGEN	EDGEN	EDGSEQEN	IDISSEN	CTTRIG	0-00 0000
F56h	CTMUCONL	EDG2POL	EDG2SEL1	EDG2SEL0	EDG1POL	EDG1SEL1	EDG1SEL0	EDG2STAT	EDG1STAT	0000 0000
F55h	CTMUICONH	ITRIM5	ITRIM4	ITRIM3	ITRIM2	ITRIM1	ITRIM0	IRNG1	IRNG0	0000 0000
F54h	CM1CON	CON	COE	CPOL	EVPOL1	EVPOL0	CREF	CCH1	CCH0	0001 1111
F53h	PADCFG1	RDPU	REPU	rjpu <sup>(2)</sup>	—	_	RTSECSEL1	RTSECSEL0	_	00000-
F52h	ECCP2AS	ECCP2ASE	ECCP2AS2	ECCP2AS1	ECCP2AS0	PSS2AC1	PSS2AC0	PSS2BD1	PSS2BD0	0000 0000
F51h	ECCP2DEL	P2RSEN	P2DC6	P2DC5	P2DC4	P2DC3	P2DC2	P2DC1	P2DC0	0000 0000
F50h	CCPR2H	Capture/Comp	oare/PWM Reg	gister 2 High B	yte					XXXX XXXX
F4Fh	CCPR2L	Capture/Comp	oare/PWM Reg	gister 2 Low By	/te					XXXX XXXX
F4Eh	CCP2CON	P2M1	P2M0	DC2B1	DC2B0	CCP2M3	CCP2M2	CCP2M1	CCP2M0	0000 0000
F4Dh	ECCP3AS	ECCP3ASE	ECCP3AS2	ECCP3AS1	ECCP3AS0	PSS3AC1	PSS3AC0	PSS3BD1	PSS3BD0	0000 0000
F4Ch	ECCP3DEL	P3RSEN	P3DC6	P3DC5	P3DC4	P3DC3	P3DC2	P3DC1	P3DC0	0000 0000
F4Bh	CCPR3H	Capture/Comp	pare/PWM Reg	gister 3 High B	yte					xxxx xxxx
F4Ah	CCPR3L	Capture/Comp	oare/PWM Reg	gister 3 Low By	/te					xxxx xxxx
F49h	CCP3CON	P3M1	P3M0	DC3B1	DC3B0	CCP3M3	CCP3M2	CCP3M1	CCP3M0	0000 0000
F48h	CCPR8H	Capture/Comp	pare/PWM Reg	gister 8 High B	yte					xxxx xxxx
F47h	CCPR8L	Capture/Comp	oare/PWM Reg	gister 8 Low By	/te					xxxx xxxx
F46h	CCP8CON		_	DC8B1	DC8B0	CCP8M3	CCP8M2	CCP8M1	CCP8M0	00 0000
F45h	CCPR9H <sup>(3)</sup>	Capture/Comp	oare/PWM Reg	gister 9 High B	yte					xxxx xxxx
F44h	CCPR9L <sup>(3)</sup>	Capture/Comp	oare/PWM Reg	gister 9 Low By	/te					xxxx xxxx
F43h	CCP9CON <sup>(3)</sup>		_	DC9B1	DC9B0	CCP9M3	CCP9M2	CCP9M1	CCP9M0	00 0000
F42h	CCPR10H(3)	Capture/Comp	are/PWM Reg	gister 10 High	Byte					xxxx xxxx
F41h	CCPR10L <sup>(3)</sup>	Capture/Comp	oare/PWM Reg	gister 10 Low E	Byte					xxxx xxxx
F40h	CCP10CON <sup>(3)</sup>	_	_	DC10B1	DC10B0	CCP10M3	CCP10M2	CCP10M1	CCP10M0	00 0000
F3Fh	TMR7H <sup>(3)</sup>	Timer7 Regist	er High Byte							xxxx xxxx
F3Eh	TMR7L <sup>(3)</sup>	Timer7 Regist	er Low Byte							0000 0000
F3Dh	T7CON <sup>(3)</sup>	TMR7CS1	TMR7CS0	T7CKPS1	T7CKPS0	SOSCEN	T7SYNC	RD16	TMR70N	0000 0000
F3Ch	T7GCON <sup>(3)</sup>	TMR7GE	T7GPOL	T7GTM	T7GSPM	T7GGO/ T7DONE	T7GVAL	T7GSS1	T7GSS0	0000 0x00
F3Bh	TMR6	Timer6 Regist	er							0000 0000
F3Ah	PR6	Timer6 Period								1111 1111
F39h	T6CON	_	T6OUTPS3	T6OUTPS2	T6OUTPS1	T6OUTPS0	TMR6ON	T6CKPS1	T6CKPS0	-000 0000
F38h	TMR8	Timer8 Regist	er							0000 0000
F37h	PR8	Timer8 Period	Register							1111 1111
F36h	T8CON	_	T8OUTPS3	T8OUTPS2	T8OUTPS1	T8OUTPS0	TMR8ON	T8CKPS1	T8CKPS0	-000 0000
F35h	TMR10 <sup>(3)</sup>	TMR10 Register							0000 0000	
F34h	PR10 <sup>(3)</sup>	Timer10 Perio	d Register							1111 1111
F33h	T10CON <sup>(3)</sup>	_	T10OUTPS3	T10OUTPS2	T10OUTPS1	T10OUTPS0	TMR100N	T10CKPS1	T10CKPS0	-000 0000
F32h	TMR12 <sup>(3)</sup>								0000 0000	
F31h	PR12 <sup>(3)</sup>								1111 1111	
F30h	T12CON <sup>(3)</sup>	_	2	T12OUTPS2	T12OUTPS1	T12OUTPS0	TMR12ON	T12CKPS1	T12CKPS0	-000 0000
F2Fh	CM2CON	CON	COE	CPOL	EVPOL1	EVPOL0	CREF	CCH1	CCH0	0001 1111
F2Eh	CM3CON	CON	COE	CPOL	EVPOL1	EVPOL0	CREF	CCH1	CCH0	0001 1111
	CCPTMRS0	C3TSEL1	C3TSEL0	C2TSEL2	C2TSEL1	C2TSEL0	C1TSEL2	C1TSEL1	C1TSEL0	0000 0000

#### TARI E 6-2. PIC18F87K22 FAMILY REGISTER FILE SUMMARY (CONTINUED)

This bit is available when Master Clear is disabled (MCLRE = 0). When MCLRE is set, the bit is unimplemented. Unimplemented on 64-pin devices (PIC18F6XK22), read as '0'. Unimplemented on devices with a program memory of 32 Kbytes (PIC18FX5K22). Note 1:

2: 3:

IADEE	TABLE 0-2. FICTOFOTR22 FAMILT REGISTER FILE SUMMART (CONTINUED)									
Address	File Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Value on POR, BOR
F2Ch	CCPTMRS1	C7TSEL1	C7TSEL0	-	C6TSEL0	—	C5TSEL0	C4TSEL1	C4TSEL0	00-0 -000
F2Bh	CCPTMRS2	_	_	_	C10TSEL0 <sup>(3)</sup>	_	C9TSEL0 <sup>(3)</sup>	C8TSEL1	C8TSEL0	0 -000
F2Ah	REFOCON	ROON	_	ROSSLP	ROSEL	RODIV3	RODIV2	RODIV1	RODIV0	0-00 0000
F29h	ODCON1	SSP10D	CCP2OD	CCP10D	_	—	_	_	SSP2OD	0000
F28h	ODCON2	CCP100D <sup>(3)</sup>	CCP9OD <sup>(3)</sup>	CCP8OD	CCP7OD	CCP6OD	CCP5OD	CCP4OD	CCP3OD	0000 0000
F27h	ODCON3	U2OD	U10D	_	_	_	_	_	CTMUDS	000
F26h	MEMCON <sup>(2)</sup>	EBDIS	_	WAIT1	WAIT0	_	_	WM1	WM0	0-0000
F25h	ANCON0	ANSEL7	ANSEL6	ANSEL5	ANSEL4	ANSEL3	ANSEL2	ANSEL1	ANSEL0	1111 1111
F24h	ANCON1	ANSEL15	ANSEL14	ANSEL13	ANSEL12	ANSEL11	ANSEL10	ANSEL9	ANSEL8	1111 1111
F23h	ANCON2	ANSEL23	ANSEL22	ANSEL21	ANSEL20	ANSEL19	ANSEL18	ANSEL17	ANSEL16	1111 1111
F22h	RCSTA2	SPEN	RX9	SREN	CREN	ADDEN	FERR	OERR	RX9D	0000 000x
F21h	TXSTA2	CSRC	TX9	TXEN	SYNC	SENDB	BRGH	TRMT	TX9D	0000 0010
F20h	BAUDCON2	ABDOVF	RCIDL	RXDTP	TXCKP	BRG16	_	WUE	ABDEN	0100 0-00
F1Fh	SPBRGH2	USART2 Baud Rate Generator High Byte							0000 0000	
F1Eh	SPBRG2	USART2 Bau	d Rate Genera	tor						0000 0000
F1Dh	RCREG2	Receive Data	FIFO							0000 0000
F1Ch	TXREG2	Transmit Data	FIFO							XXXX XXXX
F1Bh	PSTR2CON	CMPL1	CMPL0	_	STRSYNC	STRD	STRC	STRB	STRA	00-0 0001
F1Ah	PSTR3CON	CMPL1	CMPL0	_	STRSYNC	STRD	STRC	STRB	STRA	00-0 0001
F19h	PMD0	CCP3MD	CCP2MD	CCP1MD	UART2MD	UART1MD	SSP2MD	SSP1MD	ADCMD	0000 0000
F18h	PMD1	PSPMD	CTMUMD	RTCCMD	TMR4MD	TMR3MD	TMR2MD	TMR1MD	EMBMD	0000 0000
F17h	PMD2	TMR10MD <sup>(3)</sup>	TMR8MD	TMR7MD <sup>(3)</sup>	TMR6MD	TMR5MD	CMP3MD	CMP2MD	CMP1MD	0000 0000
F16h	PMD3	CCP10MD <sup>(3)</sup>	CCP9MD <sup>(3)</sup>	CCP8MD	CCP7MD	CCP6MD	CCP5MD	CCP4MD	TMR12MD <sup>(3)</sup>	0000 0000

#### **TABLE 6-2:** PIC18F87K22 FAMILY REGISTER FILE SUMMARY (CONTINUED)

This bit is available when Master Clear is disabled (MCLRE = 0). When MCLRE is set, the bit is unimplemented. Unimplemented on 64-pin devices (PIC18F6XK22), read as '0'. Unimplemented on devices with a program memory of 32 Kbytes (PIC18FX5K22). Note 1: 2: 3:

#### 6.3.5 STATUS REGISTER

The STATUS register, shown in Register 6-2, contains the arithmetic status of the ALU. The STATUS register can be the operand for any instruction, as with any other register. If the STATUS register is the destination for an instruction that affects the Z, DC, C, OV or N bits, the write to these five bits is disabled.

These bits are set or cleared according to the device logic. Therefore, the result of an instruction with the STATUS register as destination may be different than intended. For example, CLRF STATUS will set the Z bit but leave the other bits unchanged. The STATUS register then reads back as '000u uluu'.

It is recommended, therefore, that only BCF, BSF, SWAPF, MOVFF and MOVWF instructions be used to alter the STATUS register because these instructions do not affect the Z, C, DC, OV or N bits in the STATUS register.

For other instructions not affecting any Status bits, see the instruction set summaries in Table 29-2 and Table 29-3.

Note: The C and DC bits operate, in subtraction, as borrow and digit borrow bits, respectively.

### REGISTER 6-2: STATUS REGISTER

U-0	U-0	U-0	R/W-x	R/W-x	R/W-x	R/W-x	R/W-x
	_	—	N	OV	Z	DC <sup>(1)</sup>	C <sup>(2)</sup>
bit 7							bit 0
Legend:							
R = Read	able bit	W = Writable	bit	U = Unimpler	nented bit, rea	ad as '0'	
-n = Valu	e at POR	'1' = Bit is se	t	'0' = Bit is cle		x = Bit is unkr	nown
bit 7-5	Unimpler	nented: Read as	·0'				
bit 4	N: Negati						
	•	used for signed a	rithmetic (2's co	omplement). It i	ndicates whe	ther the result wa	as negative
		t was negative t was positive					
bit 3	OV: Over	flow bit					
		used for signed a which causes the				verflow of the se	even-bit
		ow occurred for si erflow occurred	gned arithmeti	c (in this arithm	etic operation	)	
bit 2	Z: Zero bi	t					
		esult of an arithme esult of an arithme	• •		<sup>-</sup> О		
bit 1		Carry/Borrow bit <sup>(1</sup>					
		F, ADDLW, SUBI					
		y-out from the 4th rry-out from the 4t			urrea		
bit 0		Borrow bit <sup>(2)</sup>					
		F, ADDLW, SUBI	Lw and SUBWF i	instructions:			
		y-out from the Mo			occurred		
	0 <b>= No ca</b>	rry-out from the M	ost Significant	bit of the result	occurred		
Note 1:	For borrow, th operand.	e polarity is revers	ed. A subtractio	on is executed b	by adding the 2	2's complement o	of the second
2:	·	e polarity is revers	ed. A subtracti	on is executed	by adding the	2's complement	of the second

#### 6.4 Data Addressing Modes

Note:	The execution of some instructions in the
	core PIC18 instruction set are changed
	when the PIC18 extended instruction set is
	enabled. For more information, see
	Section 6.6 "Data Memory and the
	Extended Instruction Set".

While the program memory can be addressed in only one way, through the Program Counter, information in the data memory space can be addressed in several ways. For most instructions, the addressing mode is fixed. Other instructions may use up to three modes, depending on which operands are used and whether or not the extended instruction set is enabled.

The addressing modes are:

- Inherent
- Literal
- Direct
- Indirect

An additional addressing mode, Indexed Literal Offset, is available when the extended instruction set is enabled (XINST Configuration bit = 1). For details on this mode's operation, see **Section 6.6.1 "Indexed Addressing with Literal Offset**".

# 6.4.1 INHERENT AND LITERAL ADDRESSING

Many PIC18 control instructions do not need any argument at all. They either perform an operation that globally affects the device or they operate implicitly on one register. This addressing mode is known as Inherent Addressing. Examples of this mode include SLEEP, RESET and DAW.

Other instructions work in a similar way, but require an additional explicit argument in the opcode. This method is known as the Literal Addressing mode because the instructions require some literal value as an argument. Examples of this include ADDLW and MOVLW, which respectively, add or move a literal value to the W register. Other examples include CALL and GOTO, which include a 20-bit program memory address.

#### 6.4.2 DIRECT ADDRESSING

Direct Addressing specifies all or part of the source and/or destination address of the operation within the opcode itself. The options are specified by the arguments accompanying the instruction.

In the core PIC18 instruction set, bit-oriented and byte-oriented instructions use some version of Direct Addressing by default. All of these instructions include some 8-bit literal address as their Least Significant Byte. This address specifies the instruction's data source as either a register address in one of the banks of data RAM (see Section 6.3.3 "General Purpose Register File") or a location in the Access Bank (see Section 6.3.2 "Access Bank").

The Access RAM bit, 'a', determines how the address is interpreted. When 'a' is '1', the contents of the BSR (Section 6.3.1 "Bank Select Register") are used with the address to determine the complete 12-bit address of the register. When 'a' is '0', the address is interpreted as being a register in the Access Bank. Addressing that uses the Access RAM is sometimes also known as Direct Forced Addressing mode.

A few instructions, such as MOVFF, include the entire 12-bit address (either source or destination) in their opcodes. In these cases, the BSR is ignored entirely.

The destination of the operation's results is determined by the destination bit, 'd'. When 'd' is '1', the results are stored back in the source register, overwriting its original contents. When 'd' is '0', the results are stored in the W register. Instructions without the 'd' argument have a destination that is implicit in the instruction, either the target register being operated on or the W register.

#### 6.4.3 INDIRECT ADDRESSING

Indirect Addressing allows the user to access a location in data memory without giving a fixed address in the instruction. This is done by using File Select Registers (FSRs) as pointers to the locations to be read or written to. Since the FSRs are themselves located in RAM as Special Function Registers, they can also be directly manipulated under program control. This makes FSRs very useful in implementing data structures such as tables and arrays in data memory.

The registers for Indirect Addressing are also implemented with Indirect File Operands (INDFs) that permit automatic manipulation of the pointer value with auto-incrementing, auto-decrementing or offsetting with another value. This allows for efficient code using loops, such as the example of clearing an entire RAM bank in Example 6-5. It also enables users to perform Indexed Addressing and other Stack Pointer operations for program memory in data memory.

#### EXAMPLE 6-5: HOW TO CLEAR RAM (BANK 1) USING INDIRECT ADDRESSING

	LFSR	FSR0, 100h	;	
NEXT	CLRF	POSTINC0	;	Clear INDF
			;	register then
			;	inc pointer
	BTFSS	FSROH, 1	;	All done with
			;	Bank1?
	BRA	NEXT	;	NO, clear next
CONTINU	JE		;	YES, continue
1				

# 6.4.3.1 FSR Registers and the INDF Operand

At the core of Indirect Addressing are three sets of registers: FSR0, FSR1 and FSR2. Each represents a pair of 8-bit registers: FSRnH and FSRnL. The four upper bits of the FSRnH register are not used, so each FSR pair holds a 12-bit value. This represents a value that can address the entire range of the data memory in a linear fashion. The FSR register pairs, then, serve as pointers to data memory locations.

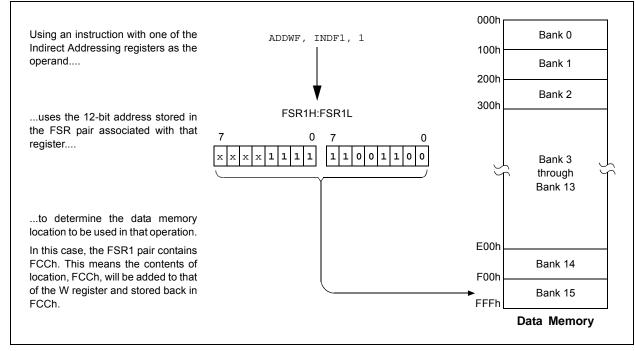
Indirect Addressing is accomplished with a set of Indirect File Operands, INDF0 through INDF2. These can be thought of as "virtual" registers. The operands are

#### FIGURE 6-8: INDIRECT ADDRESSING

mapped in the SFR space, but are not physically implemented. Reading or writing to a particular INDF register actually accesses its corresponding FSR register pair. A read from INDF1, for example, reads the data at the address indicated by FSR1H:FSR1L.

Instructions that use the INDF registers as operands actually use the contents of their corresponding FSR as a pointer to the instruction's target. The INDF operand is just a convenient way of using the pointer.

Because Indirect Addressing uses a full 12-bit address, data RAM banking is not necessary. Thus, the current contents of the BSR and the Access RAM bit have no effect on determining the target address.



#### 6.4.3.2 FSR Registers and POSTINC, POSTDEC, PREINC and PLUSW

In addition to the INDF operand, each FSR register pair also has four additional indirect operands. Like INDF, these are "virtual" registers that cannot be indirectly read or written to. Accessing these registers actually accesses the associated FSR register pair, but also performs a specific action on its stored value.

These operands are:

- POSTDEC Accesses the FSR value, then automatically decrements it by '1' afterwards
- POSTINC Accesses the FSR value, then automatically increments it by '1' afterwards
- PREINC Increments the FSR value by '1', then uses it in the operation
- PLUSW Adds the signed value of the W register (range of -127 to 128) to that of the FSR and uses the new value in the operation

In this context, accessing an INDF register uses the value in the FSR registers without changing them. Similarly, accessing a PLUSW register gives the FSR value, offset by the value in the W register, with neither value actually changed in the operation. Accessing the other virtual registers changes the value of the FSR registers.

Operations on the FSRs with POSTDEC, POSTINC and PREINC affect the entire register pair. Rollovers of the FSRnL register, from FFh to 00h, carry over to the FSRnH register. On the other hand, results of these operations do not change the value of any flags in the STATUS register (for example, Z, N and OV bits).

The PLUSW register can be used to implement a form of Indexed Addressing in the data memory space. By manipulating the value in the W register, users can reach addresses that are fixed offsets from pointer addresses. In some applications, this can be used to implement some powerful program control structure, such as software stacks, inside of data memory.

### 6.4.3.3 Operations by FSRs on FSRs

Indirect Addressing operations that target other FSRs or virtual registers represent special cases. For example, using an FSR to point to one of the virtual registers will not result in successful operations.

As a specific case, assume that the FSR0H:FSR0L registers contain FE7h, the address of INDF1. Attempts to read the value of the INDF1, using INDF0 as an operand, will return 00h. Attempts to write to INDF1, using INDF0 as the operand, will result in a NOP.

On the other hand, using the virtual registers to write to an FSR pair may not occur as planned. In these cases, the value will be written to the FSR pair, but without any incrementing or decrementing. Thus, writing to INDF2 or POSTDEC2 will write the same value to the FSR2H:FSR2L.

Since the FSRs are physical registers mapped in the SFR space, they can be manipulated through all direct operations. Users should proceed cautiously when working on these registers, however, particularly if their code uses Indirect Addressing.

Similarly, operations by Indirect Addressing are generally permitted on all other SFRs. Users should exercise the appropriate caution, so that they do not inadvertently change settings that might affect the operation of the device.

### 6.5 Program Memory and the Extended Instruction Set

The operation of program memory is unaffected by the use of the extended instruction set.

Enabling the extended instruction set adds five additional two-word commands to the existing PIC18 instruction set: ADDFSR, CALLW, MOVSF, MOVSS and SUBFSR. These instructions are executed as described in Section 6.2.4 "Two-Word Instructions".

# 6.6 Data Memory and the Extended Instruction Set

Enabling the PIC18 extended instruction set (XINST Configuration bit = 1) significantly changes certain aspects of data memory and its addressing. Using the Access Bank for many of the core PIC18 instructions introduces a new addressing mode for the data memory space. This mode also alters the behavior of Indirect Addressing using FSR2 and its associated operands.

What does not change is just as important. The size of the data memory space is unchanged, as well as its linear addressing. The SFR map remains the same. Core PIC18 instructions can still operate in both Direct and Indirect Addressing mode. Inherent and literal instructions do not change at all. Indirect Addressing with FSR0 and FSR1 also remains unchanged.

#### 6.6.1 INDEXED ADDRESSING WITH LITERAL OFFSET

Enabling the PIC18 extended instruction set changes the behavior of Indirect Addressing using the FSR2 register pair and its associated file operands. Under the proper conditions, instructions that use the Access Bank – that is, most bit-oriented and byte-oriented instructions – can invoke a form of Indexed Addressing using an offset specified in the instruction. This special addressing mode is known as Indexed Addressing with Literal Offset or the Indexed Literal Offset mode.

When using the extended instruction set, this addressing mode requires the following:

- Use of the Access Bank ('a' = 0)
- A file address argument that is less than or equal to 5Fh

Under these conditions, the file address of the instruction is not interpreted as the lower byte of an address (used with the BSR in Direct Addressing) or as an 8-bit address in the Access Bank. Instead, the value is interpreted as an offset value to an Address Pointer specified by FSR2. The offset and the contents of FSR2 are added to obtain the target address of the operation.

### 6.6.2 INSTRUCTIONS AFFECTED BY INDEXED LITERAL OFFSET MODE

Any of the core PIC18 instructions that can use Direct Addressing are potentially affected by the Indexed Literal Offset Addressing mode. This includes all byte-oriented and bit-oriented instructions, or almost one-half of the standard PIC18 instruction set. Instructions that only use Inherent or Literal Addressing modes are unaffected.

Additionally, byte-oriented and bit-oriented instructions are not affected if they do not use the Access Bank (Access RAM bit = 1), or include a file address of 60h or above. Instructions meeting these criteria will continue to execute as before. A comparison of the different possible addressing modes when the extended instruction set is enabled is shown in Figure 6-9.

Those who desire to use byte-oriented or bit-oriented instructions in the Indexed Literal Offset mode should note the changes to assembler syntax for this mode. This is described in more detail in Section 29.2.1 "Extended Instruction Syntax".

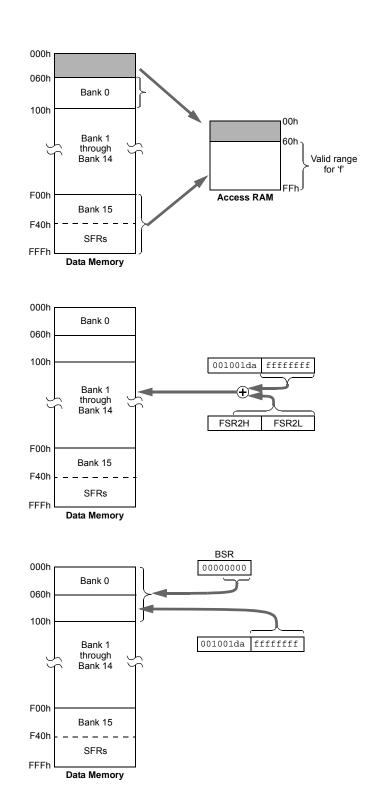
#### FIGURE 6-9: COMPARING ADDRESSING OPTIONS FOR BIT-ORIENTED AND BYTE-ORIENTED INSTRUCTIONS (EXTENDED INSTRUCTION SET ENABLED)

**EXAMPLE INSTRUCTION:** ADDWF, f, d, a (Opcode: 0010 01da ffff ffff)

#### When a = 0 and $f \ge 60h$ :

The instruction executes in Direct Forced mode. 'f' is interpreted as a location in the Access RAM, between 060h and FFFh. This is the same as locations, F60h to FFFh (Bank 15), of data memory.

Locations below 060h are not available in this addressing mode.



#### When a = 0 and $f \le 5Fh$ :

The instruction executes in Indexed Literal Offset mode. 'f' is interpreted as an offset to the address value in FSR2. The two are added together to obtain the address of the target register for the instruction. The address can be anywhere in the data memory space.

Note that in this mode, the correct syntax is now: ADDWF [k], d where 'k' is the same as 'f'.

#### When a = 1 (all values of f):

The instruction executes in Direct mode (also known as Direct Long mode). 'f' is interpreted as a location in one of the 16 banks of the data memory space. The bank is designated by the Bank Select Register (BSR). The address can be in any implemented bank in the data memory space.

## 6.6.3 MAPPING THE ACCESS BANK IN INDEXED LITERAL OFFSET MODE

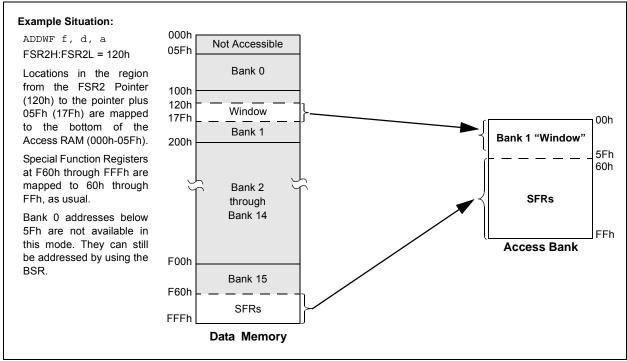
The use of Indexed Literal Offset Addressing mode effectively changes how the lower part of Access RAM (00h to 5Fh) is mapped. Rather than containing just the contents of the bottom part of Bank 0, this mode maps the contents from Bank 0 and a user-defined "window" that can be located anywhere in the data memory space.

The value of FSR2 establishes the lower boundary of the addresses mapped into the window, while the upper boundary is defined by FSR2 plus 95 (5Fh). Addresses in the Access RAM above 5Fh are mapped as previously described. (See **Section 6.3.2 "Access Bank**".) An example of Access Bank remapping in this addressing mode is shown in Figure 6-10. Remapping the Access Bank applies *only* to operations using the Indexed Literal Offset mode. Operations that use the BSR (Access RAM bit = 1) will continue to use Direct Addressing as before. Any Indirect or Indexed Addressing operation that explicitly uses any of the indirect file operands (including FSR2) will continue to operate as standard Indirect Addressing. Any instruction that uses the Access Bank, but includes a register address of greater than 05Fh, will use Direct Addressing and the normal Access Bank map.

## 6.6.4 BSR IN INDEXED LITERAL OFFSET MODE

Although the Access Bank is remapped when the extended instruction set is enabled, the operation of the BSR remains unchanged. Direct Addressing, using the BSR to select the data memory bank, operates in the same manner as previously described.

## FIGURE 6-10: REMAPPING THE ACCESS BANK WITH INDEXED LITERAL OFFSET ADDRESSING



## 7.0 FLASH PROGRAM MEMORY

The Flash program memory is readable, writable and erasable during normal operation over the entire  $\mathsf{V}\mathsf{D}\mathsf{D}$  range.

A read from program memory is executed on one byte at a time. For execution of a write to, or erasure of, program memory:

- Memory of 32 Kbytes and 64 Kbytes (PIC18FX5K22 and PIC18FX6K22 devices) – Blocks of 64 bytes
- Memory of 128 Kbytes (PIC18FX7K22 devices) Blocks of 128 bytes

Writing or erasing program memory will cease instruction fetches until the operation is complete. The program memory cannot be accessed during the write or erase, therefore, code cannot execute. An internal programming timer terminates program memory writes and erases.

A value written to program memory does not need to be a valid instruction. Executing a program memory location that forms an invalid instruction results in a NOP.

## 7.1 Table Reads and Table Writes

In order to read and write program memory, there are two operations that allow the processor to move bytes between the program memory space and the data RAM:

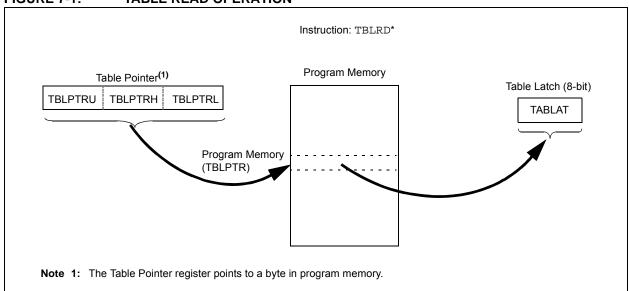
- Table Read (TBLRD)
- Table Write (TBLWT)

The program memory space is 16 bits wide, while the data RAM space is 8 bits wide. Table reads and table writes move data between these two memory spaces through an 8-bit register (TABLAT).

Table read operations retrieve data from program memory and place it into the data RAM space. Figure 7-1 shows the operation of a table read with program memory and data RAM.

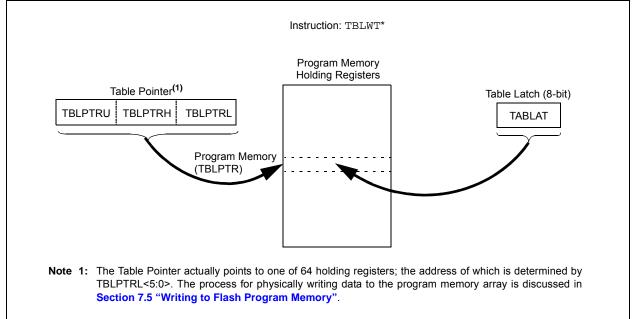
Table write operations store data from the data memory space into holding registers in program memory. The procedure to write the contents of the holding registers into program memory is detailed in **Section 7.5** "Writing **to Flash Program Memory**". Figure 7-2 shows the operation of a table write with program memory and data RAM.

Table operations work with byte entities. A table block containing data, rather than program instructions, is not required to be word-aligned. Therefore, a table block can start and end at any byte address. If a table write is being used to write executable code into program memory, program instructions will need to be word-aligned.



#### FIGURE 7-1: TABLE READ OPERATION

## FIGURE 7-2: TABLE WRITE OPERATION



## 7.2 Control Registers

Several control registers are used in conjunction with the  ${\tt TBLRD}$  and  ${\tt TBLWT}$  instructions. These include the:

- · EECON1 register
- EECON2 register
- TABLAT register
- TBLPTR registers

## 7.2.1 EECON1 AND EECON2 REGISTERS

The EECON1 register (Register 7-1) is the control register for memory accesses. The EECON2 register, not a physical register, is used exclusively in the memory write and erase sequences. Reading EECON2 will read all '0's.

The EEPGD control bit determines if the access is a program or data EEPROM memory access. When clear, any subsequent operations operate on the data EEPROM memory. When set, any subsequent operations operate on the program memory.

The CFGS control bit determines if the access is to the Configuration/Calibration registers or to program memory/data EEPROM memory. When set, subsequent operations operate on Configuration registers regardless of EEPGD (see Section 28.0 "Special Features of the CPU"). When clear, memory selection access is determined by EEPGD.

The FREE bit, when set, allows a program memory erase operation. When FREE is set, the erase operation is initiated on the next WR command. When FREE is clear, only writes are enabled.

The WREN bit, when set, allows a write operation. On power-up, the WREN bit is clear. The WRERR bit is set in hardware when the WR bit is set and cleared when the internal programming timer expires and the write operation is complete.

Note:	During normal operation, the WRERR is									
	read as '1'. This can indicate that a write									
	operation was prematurely terminated by									
	a Reset or a write operation was									
	attempted improperly.									

The WR control bit initiates write operations. The bit cannot be cleared, only set, in software. It is cleared in hardware at the completion of the write operation.

Note: The EEIF interrupt flag bit (PIR6<4>) is set when the write is complete. It must be cleared in software.

#### REGISTER 7-1: EECON1: EEPROM CONTROL REGISTER 1

R/W-x	R/W-x	U-0	R/W-0	R/W-x	R/W-0	R/S-0	R/S-0
EEPGD	CFGS	—	FREE	WRERR <sup>(1)</sup>	WREN	WR	RD
bit 7							bit 0

Legend:	S = Settable bit		
R = Readable bit	W = Writable bit	U = Unimplemented bit,	read as '0'
-n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown

bit 7	EEPGD: Flash Program or Data EEPROM Memory Select bit
	1 = Access Flash program memory
	0 = Access data EEPROM memory
bit 6	CFGS: Flash Program/Data EEPROM or Configuration Select bit
	1 = Access Configuration registers
	0 = Access Flash program or data EEPROM memory
bit 5	Unimplemented: Read as '0'
bit 4	FREE: Flash Row Erase Enable bit
	1 = Erase the program memory row addressed by TBLPTR on the next WR command
	(cleared by completion of erase operation)
	0 = Perform write-only
bit 3	WRERR: Flash Program/Data EEPROM Error Flag bit <sup>(1)</sup>
	1 = A write operation is prematurely terminated (any Reset during self-timed programming in normal
	operation or an improper write attempt)
	0 = The write operation completed
bit 2	WREN: Flash Program/Data EEPROM Write Enable bit
	1 = Allows write cycles to Flash program/data EEPROM
	0 = Inhibits write cycles to Flash program/data EEPROM
bit 1	WR: Write Control bit
	1 = Initiates a data EEPROM erase/write cycle or a program memory erase cycle or write cycle
	(The operation is self-timed and the bit is cleared by hardware once the write is complete.
	The WR bit can only be set (not cleared) in software.)
	0 = Write cycle to the EEPROM is complete
bit 0	RD: Read Control bit
	1 = Initiates an EEPROM read (Read takes one cycle. RD is cleared in hardware. The RD bit can only
	be set (not cleared) in software. The RD bit cannot be set when EEPGD = 1 or CFGS = 1.)
	0 = Does not initiate an EEPROM read

**Note 1:** When a WRERR occurs, the EEPGD and CFGS bits are not cleared. This allows tracing of the error condition.

## 7.2.2 TABLAT – TABLE LATCH REGISTER

The Table Latch (TABLAT) is an eight-bit register mapped into the SFR space. The Table Latch register is used to hold 8-bit data during data transfers between program memory and data RAM.

#### 7.2.3 TBLPTR – TABLE POINTER REGISTER

The Table Pointer (TBLPTR) register addresses a byte within the program memory. The TBLPTR is comprised of three SFR registers: Table Pointer Upper Byte, Table Pointer High Byte and Table Pointer Low Byte (TBLPTRU:TBLPTRH:TBLPTRL). These three registers join to form a 22-bit wide pointer. The low-order 21 bits allow the device to address up to 2 Mbytes of program memory space. The 22nd bit allows access to the Device ID, the User ID and the Configuration bits.

The Table Pointer register, TBLPTR, is used by the TBLRD and TBLWT instructions. These instructions can update the TBLPTR in one of four ways, based on the table operation. These operations are shown in Table 7-1 and only affect the low-order 21 bits.

## 7.2.4 TABLE POINTER BOUNDARIES

The TBLPTR is used in reads, writes and erases of the Flash program memory.

When a TBLRD is executed, all 22 bits of the TBLPTR determine which byte is read from program memory into the TABLAT.

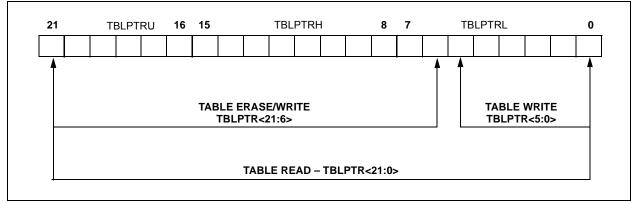
When a TBLWT is executed, the six LSbs of the Table Pointer register (TBLPTR<5:0>) determine which of the 64 program memory holding registers is written to. When the timed write to program memory begins (via the WR bit), the 16 MSbs of the TBLPTR (TBLPTR<21:6>) determine which program memory block of 64 bytes is written to. For more detail, see Section 7.5 "Writing to Flash Program Memory".

When an erase of program memory is executed, the 16 MSbs of the Table Pointer register (TBLPTR<21:6>) point to the 64-byte block that will be erased. The Least Significant bits (TBLPTR<5:0>) are ignored.

Figure 7-3 describes the relevant boundaries of the TBLPTR based on Flash program memory operations.

Example	Operation on Table Pointer
TBLRD* TBLWT*	TBLPTR is not modified
TBLRD*+ TBLWT*+	TBLPTR is incremented after the read/write
TBLRD*- TBLWT*-	TBLPTR is decremented after the read/write
TBLRD+* TBLWT+*	TBLPTR is incremented before the read/write

#### FIGURE 7-3: TABLE POINTER BOUNDARIES BASED ON OPERATION



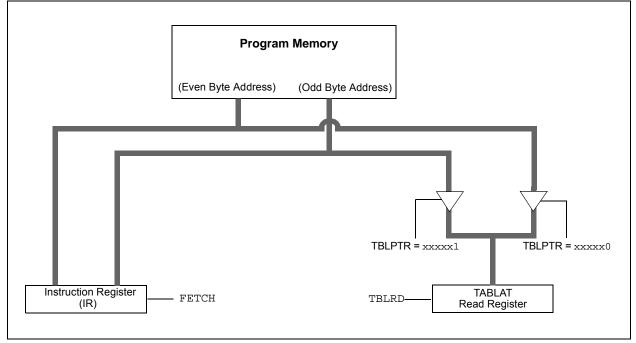
### 7.3 Reading the Flash Program Memory

The TBLRD instruction is used to retrieve data from program memory and places it into data RAM. Table reads from program memory are performed, one byte at a time.

TBLPTR points to a byte address in program space. Executing TBLRD places the byte pointed to into TABLAT. In addition, the TBLPTR can be modified automatically for the next table read operation.

The internal program memory is typically organized by words. The Least Significant bit of the address selects between the high and low bytes of the word. Figure 7-4 shows the interface between the internal program memory and the TABLAT.

## FIGURE 7-4: READS FROM FLASH PROGRAM MEMORY



#### EXAMPLE 7-1: READING A FLASH PROGRAM MEMORY WORD

	BCF	EECON1, CFGS	; point to Flash program memory
	BSF	EECON1, EEPGD	; access Flash program memory
	MOVLW	CODE_ADDR_UPPER	; Load TBLPTR with the base
	MOVWF	TBLPTRU	; address of the word
	MOVLW	CODE_ADDR_HIGH	
	MOVWF	TBLPTRH	
	MOVLW	CODE_ADDR_LOW	
	MOVWF	TBLPTRL	
READ_WORD			
	TBLRD*-	÷	; read into TABLAT and increment
	MOVF	TABLAT, W	; get data
	MOVWF	WORD_EVEN	
	TBLRD*-	÷	; read into TABLAT and increment
	MOVF	TABLAT, W	; get data
	MOVF	WORD_ODD	

## 7.4 Erasing Flash Program Memory

The erase blocks are:

- PIC18FX5K22 and PIC18FX6K22 32 words or 64 bytes
- PIC18FX7K22 64 words or 128 bytes

Word erase in the Flash array is not supported.

When initiating an erase sequence from the microcontroller itself, a block of 64 or 128 bytes of program memory is erased. The Most Significant 16 bits of the TBLPTR<21:6> point to the block being erased. The TBLPTR<5:0> bits are ignored.

The EECON1 register commands the erase operation. The EEPGD bit must be set to point to the Flash program memory. The WREN bit must be set to enable write operations. The FREE bit is set to select an erase operation.

For protection, the write initiate sequence for EECON2 must be used.

A long write is necessary for erasing the internal Flash. Instruction execution is halted while in a long write cycle. The long write will be terminated by the internal programming timer.

#### 7.4.1 FLASH PROGRAM MEMORY ERASE SEQUENCE

The sequence of events for erasing a block of internal program memory location is:

- 1. Load the Table Pointer register with the address of the row to be erased.
- 2. Set the EECON1 register for the erase operation:
  - · Set the EEPGD bit to point to program memory
  - Clear the CFGS bit to access program memory
  - · Set the WREN bit to enable writes
  - Set the FREE bit to enable the erase
- 3. Disable the interrupts.
- 4. Write 0x55 to EECON2.
- 5. Write 0xAA to EECON2.
- Set the WR bit. This begins the row erase cycle. The CPU will stall for the duration of the erase for TIW. (See Parameter D133A.)
- 7. Re-enable interrupts.

	MOVLW MOVWF MOVLW MOVWF MOVLW MOVWF	CODE_ADDR_UPPER TBLPTRU CODE_ADDR_HIGH TBLPTRH CODE_ADDR_LOW TBLPTRL	; load TBLPTR with the base ; address of the memory block
ERASE_ROW			
	BSF	EECON1, EEPGD	; point to Flash program memory
	BCF	EECON1, CFGS	; access Flash program memory
	BSF	EECON1, WREN	; enable write to memory
	BSF	EECON1, FREE	; enable Row Erase operation
	BCF	INTCON, GIE	; disable interrupts
Required	MOVLW	0x55	
Sequence	MOVWF	EECON2	; write 55h
	MOVLW	0xAA	
	MOVWF	EECON2	; write OAAh
	BSF	EECON1, WR	; start erase (CPU stall)
	BSF	INTCON, GIE	; re-enable interrupts

## EXAMPLE 7-2: ERASING A FLASH PROGRAM MEMORY ROW

## 7.5 Writing to Flash Program Memory

The programming blocks are:

- PIC18FX5K22 and PIC18FX6K22 32 words or 64 bytes
- PIC18FX7K22 64 words or 128 bytes

Word or byte programming is not supported.

Table writes are used internally to load the holding registers needed to program the Flash memory. The number of holding registers used for programming by the table writes are:

- PIC18FX5K22 and PIC18FX6K22 64
- PIC18FX7K22 128

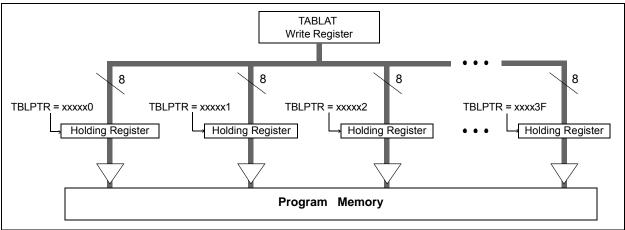
Since the Table Latch (TABLAT) is only a single byte, the TBLWT instruction may need to be executed 64 times for each programming operation. All of the table write operations will essentially be short writes because only the holding registers are written. At the end of updating the 64 or 128 holding registers, the EECON1 register must be written to in order to start the programming operation with a long write.

The long write is necessary for programming the internal Flash. Instruction execution is halted while in a long write cycle. The long write is terminated by the internal programming timer.

The EEPROM on-chip timer controls the write time. The write/erase voltages are generated by an on-chip charge pump, rated to operate over the voltage range of the device.

**Note:** The default value of the holding registers on device Resets, and after write operations, is FFh. A write of FFh to a holding register does not modify that byte. This means that individual bytes of program memory may be modified, provided that the change does not attempt to change any bit from a '0' to a '1'. When modifying individual bytes, it is not necessary to load all 64 or 128 holding registers before executing a write operation.





## 7.5.1 FLASH PROGRAM MEMORY WRITE SEQUENCE

The sequence of events for programming an internal program memory location should be:

- 1. Read the 64 or 128 bytes into RAM.
- 2. Update the data values in RAM as necessary.
- 3. Load the Table Pointer register with the address being erased.
- 4. Execute the row erase procedure.
- 5. Load the Table Pointer register with the address of the first byte being written.
- 6. Write the 64 or 128 bytes into the holding registers with auto-increment.
- 7. Set the EECON1 register for the write operation:
  - Set the EEPGD bit to point to program memory
  - · Clear the CFGS bit to access program memory
  - · Set the WREN to enable byte writes
- 8. Disable the interrupts.
- 9. Write 0x55 to EECON2.
- 10. Write 0xAA to EECON2.
- Set the WR bit. This will begin the write cycle. The CPU will stall for duration of the write for Tiw (see Parameter D133A).
- 12. Re-enable the interrupts.
- 13. Verify the memory (table read).

An example of the required code is shown in Example 7-3.

**Note:** Before setting the WR bit, the Table Pointer address needs to be within the intended address range of the 64 or 128 bytes in the holding register.

Note: Self-write execution to Flash and EEPROM memory cannot be done while running in LP Oscillator mode (Low-Power mode). Therefore, executing a self-write will put the device into High-Power mode.

EXAMPLE 7-3:	WRI	TING TO FLASH PROC	GRAM MEMORY				
	MOVLW MOVWF	SIZE_OF_BLOCK COUNTER	;	number of bytes in erase block			
	MOVLW MOVWF	BUFFER_ADDR_HIGH FSR0H	;	point to buffer			
	MOVLW	BUFFER_ADDR_LOW					
	MOVWF	FSROL					
	MOVLW	CODE_ADDR_UPPER		Load TBLPTR with the base			
	MOVWF MOVLW	TBLPTRU CODE_ADDR_HIGH	,	address of the memory block			
	MOVWF	TBLPTRH					
	MOVLW	CODE_ADDR_LOW					
	MOVWF	TBLPTRL					
READ_BLOCK							
	TBLRD*+ MOVF	- TABLAT, W		read into TABLAT, and inc get data			
	MOVIF	POSTINCO		store data			
		COUNTER		done?			
	BRA	READ_BLOCK	;	repeat			
MODIFY_WORD	MO1	DIMI IDDD		and the head of the second			
	MOVLW MOVWF	DATA_ADDR_HIGH FSR0H	;	point to buffer			
	MOVWF MOVLW	FSRUH DATA_ADDR_LOW					
	MOVWF	FSR0L					
	MOVLW	NEW_DATA_LOW	;	update buffer word			
	MOVWF	POSTINC0					
	MOVLW	NEW_DATA_HIGH					
ERASE_BLOCK	MOVWF	INDF0					
EKADE_DIOCK	MOVLW	CODE_ADDR_UPPER	;	load TBLPTR with the base			
	MOVWF	TBLPTRU	;	address of the memory block			
	MOVLW	CODE_ADDR_HIGH					
	MOVWF	TBLPTRH					
	MOVLW MOVWF	CODE_ADDR_LOW TBLPTRL					
	BSF	EECON1, EEPGD	;	point to Flash program memory			
	BCF	EECON1, CFGS		access Flash program memory			
	BSF	EECON1, WREN	;	enable write to memory			
	BSF	EECON1, FREE		enable Row Erase operation			
	BCF	INTCON, GIE 0x55	;	disable interrupts			
Required	MOVLW MOVWF	EECON2	;	write 55h			
Sequence	MOVLW	OxAA					
	MOVWF	EECON2		write 0AAh			
	BSF	EECON1, WR		start erase (CPU stall)			
	BSF TBLRD*-	INTCON, GIE		re-enable interrupts dummy read decrement			
	MOVLW	- BUFFER_ADDR_HIGH		point to buffer			
	MOVWF	FSR0H		-			
	MOVLW	BUFFER_ADDR_LOW					
	MOVWF	FSROL					
WRITE_BUFFER_B	ACK MOVLW	SIZE_OF_BLOCK		number of bytes in holding register			
	MOVLW MOVWF	COUNTER	'	number of bytes in notating register			
WRITE_BYTE_TO_							
	MOVFF	POSTINC0, WREG		get low byte of buffer data			
	MOVWF	TABLAT		present data to table latch			
	TBLWT+*	r		write data, perform a short write			
	DECESZ	COUNTER		to internal TBLWT holding register. loop until buffers are full			
	BRA	WRITE_BYTE_TO_HREGS	,	log and partors are rait			

#### EXAMPLE 7-3: WRITING TO FLASH PROGRAM MEMORY

#### EXAMPLE 7-3: WRITING TO FLASH PROGRAM MEMORY (CONTINUED)

PROGRAM_MEMOR	ζ.			
	BSF	EECON1,	EEPGD	; point to Flash program memory
	BCF	EECON1,	CFGS	; access Flash program memory
	BSF	EECON1,	WREN	; enable write to memory
	BCF	INTCON,	GIE	; disable interrupts
	MOVLW	0x55		
Required	MOVWF	EECON2		; write 55h
Sequence	MOVLW	0xAA		
	MOVWF	EECON2		; write OAAh
	BSF	EECON1,	WR	; start program (CPU stall)
	BSF	INTCON,	GIE	; re-enable interrupts
	BCF	EECON1,	WREN	; disable write to memory

#### 7.5.2 WRITE VERIFY

Depending on the application, good programming practice may dictate that the value written to the memory should be verified against the original value. This should be used in applications where excessive writes can stress bits near the specification limit.

## 7.5.3 UNEXPECTED TERMINATION OF WRITE OPERATION

If a write is terminated by an unplanned event, such as loss of power or an unexpected Reset, the memory location just programmed should be verified and reprogrammed if needed. If the write operation is interrupted by a MCLR Reset or a WDT Time-out Reset during normal operation, the user can check the WRERR bit and rewrite the location(s) as needed.

#### 7.5.4 PROTECTION AGAINST SPURIOUS WRITES

To protect against spurious writes to Flash program memory, the write initiate sequence must also be followed. See **Section 28.0** "**Special Features of the CPU**" for more details.

### 7.6 Flash Program Operation During Code Protection

See Section 28.6 "Program Verification and Code Protection" for details on code protection of Flash program memory.

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0			
TBLPTRU	_	_	bit 21 <sup>(1)</sup>	bit 21 <sup>(1)</sup> Program Memory Table Pointer Upper Byte (TBLPTR<20:16>)							
TBPLTRH	Program Memory Table Pointer High Byte (TBLPTR<15:8>)										
TBLPTRL	Program Me	mory Table Po	pinter Low	Byte (TBLPT	R<7:0>)						
TABLAT	Program Me	mory Table La	atch								
INTCON	GIE/GIEH	PEIE/GIEL	TMR0IE	INT0IE	RBIE	TMR0IF	INT0IF	RBIF			
EECON2	EEPROM Co	ontrol Registe	r 2 (not a p	hysical regist	er)						
EECON1	EEPGD	CFGS		– FREE WRERR WREN WR				RD			
IPR6	—	_	_	EEIP	_	CMP3IP	CMP2IP	CMP1IP			
PIR6	_	_		EEIF	_	CMP3IF	CMP2IF	CMP1IF			
PIE6	—			EEIE		CMP3IE	CMP2IE	CMP1IE			

#### TABLE 7-2: REGISTERS ASSOCIATED WITH PROGRAM FLASH MEMORY

Legend: — = unimplemented, read as '0'. Shaded cells are not used during Flash/EEPROM access.

**Note 1:** Bit 21 of the TBLPTRU allows access to the device Configuration bits.

## 8.0 EXTERNAL MEMORY BUS

Note:	The	External	Memory	Bus	is	not		
	implemented on 64-pin devices.							

The External Memory Bus (EMB) allows the device to access external memory devices (such as Flash, EPROM or SRAM) as program or data memory. It supports both 8 and 16-Bit Data Width modes and three address widths of up to 20 bits.

The bus is implemented with 28 pins, multiplexed across four I/O ports. Three ports (PORTD, PORTE and PORTH) are multiplexed with the address/data bus for a total of 20 available lines, while PORTJ is multiplexed with the bus control signals.

A list of the pins and their functions is provided in Table 8-1.

<b>TABLE 8-1:</b>	PIC18F87K22 FAMILY EXTERNAL BUS – I/O PORT FUNCTIONS

Name	Port	Bit	External Memory Bus Function
RD0/AD0	PORTD	0	Address Bit 0 or Data Bit 0
RD1/AD1	PORTD	1	Address Bit 1 or Data Bit 1
RD2/AD2	PORTD	2	Address Bit 2 or Data Bit 2
RD3/AD3	PORTD	3	Address Bit 3 or Data Bit 3
RD4/AD4	PORTD	4	Address Bit 4 or Data Bit 4
RD5/AD5	PORTD	5	Address Bit 5 or Data Bit 5
RD6/AD6	PORTD	6	Address Bit 6 or Data Bit 6
RD7/AD7	PORTD	7	Address Bit 7 or Data Bit 7
RE0/AD8	PORTE	0	Address Bit 8 or Data Bit 8
RE1/AD9	PORTE	1	Address Bit 9 or Data Bit 9
RE2/AD10	PORTE	2	Address Bit 10 or Data Bit 10
RE3/AD11	PORTE	3	Address Bit 11 or Data Bit 11
RE4/AD12	PORTE	4	Address Bit 12 or Data Bit 12
RE5/AD13	PORTE	5	Address Bit 13 or Data Bit 13
RE6/AD14	PORTE	6	Address Bit 14 or Data Bit 14
RE7/AD15	PORTE	7	Address Bit 15 or Data Bit 15
RH0/A16	PORTH	0	Address Bit 16
RH1/A17	PORTH	1	Address Bit 17
RH2/A18	PORTH	2	Address Bit 18
RH3/A19	PORTH	3	Address Bit 19
RJ0/ALE	PORTJ	0	Address Latch Enable (ALE) Control pin
RJ1/OE	PORTJ	1	Output Enable (OE) Control pin
RJ2/WRL	PORTJ	2	Write Low (WRL) Control pin
RJ3/WRH	PORTJ	3	Write High (WRH) Control pin
RJ4/BA0	PORTJ	4	Byte Address Bit 0 (BA0)
RJ5/CE	PORTJ	5	Chip Enable (CE) Control pin
RJ6/LB	PORTJ	6	Lower Byte Enable (IB) Control pin
RJ7/UB	PORTJ	7	Upper Byte Enable (UB) Control pin

**Note:** For the sake of clarity, only I/O port and external bus assignments are shown here. One or more additional multiplexed features may be available on some pins.

## 8.1 External Memory Bus Control

The operation of the interface is controlled by the MEMCON register (Register 8-1). This register is available in all program memory operating modes except Microcontroller mode. In this mode, the register is disabled and cannot be written to.

The EBDIS bit (MEMCON<7>) controls the operation of the bus and related port functions. Clearing EBDIS enables the interface and disables the I/O functions of the ports, as well as any other functions multiplexed to those pins. Setting the bit enables the I/O ports and other functions, but allows the interface to override everything else on the pins when an external memory operation is required. By default, the external bus is always enabled and disables all other I/O. The operation of the EBDIS bit is also influenced by the program memory mode being used. This is discussed in more detail in Section 8.5 "Program Memory Modes and the External Memory Bus".

The WAIT bits allow for the addition of Wait states to external memory operations. The use of these bits is discussed in **Section 8.3** "Wait States".

The WM bits select the particular operating mode used when the bus is operating in 16-Bit Data Width mode. These bits are discussed in more detail in **Section 8.6 "16-Bit Data Width Modes"**. These bits have no effect when an 8-Bit Data Width mode is selected.

## REGISTER 8-1: MEMCON: EXTERNAL MEMORY BUS CONTROL REGISTER<sup>(1)</sup>

R/W-0	U-0	R/W-0	R/W-0	U-0	U-0	R/W-0	R/W-0	
EBDIS	—	WAIT1	WAIT0	—	—	WM1	WM0	
bit 7		•					bit 0	
Legend:	Legend:							
R = Readable	e bit	W = Writable	bit	U = Unimplem	nented bit, read	l as '0'		
-n = Value at	POR	'1' = Bit is set		'0' = Bit is clea	ared	x = Bit is unkr	nown	
bit 7	EBDIS: Extern	nal Bus Disable	e bit					
	1 = External	bus is enabled	when microco	ontroller access	es external me	mory; otherwis	se, all external	
		rs are mapped						
	0 = External I	bus is always e	nabled, I/O po	rts are disabled	1			
bit 6	Unimplement	ted: Read as 'o	)'					
bit 5-4	WAIT<1:0>: Table Reads and Writes Bus Cycle Wait Count bits							
	11 = Table rea	ads and writes	will wait 0 TCY					
	10 = Table rea	ads and writes	will wait 1 TCY					
	01 = Table rea	ads and writes	will wait 2 TCY					
	00 = Table rea	ads and writes	will wait 3 TCY					
bit 3-2	Unimplemented: Read as '0'							
bit 1-0	WM<1:0>: TBLWT Operation with 16-Bit Data Bus Width Select bits							
	1x = Word W	/rite mode: TAE	BLAT word out	out; WRH is act	ive when TABL	AT is written		
	01 = Byte Se	lect mode: TAB	LAT data is cop	oied on both MS	B and LSB; WF	RH and (UB or L	B) will activate	
	00 = Byte Wr	rite mode: TAB	LAT data is co	pied on both MS	SB and LSB; W	/RH or WRL w	ill activate	

Note 1: Unimplemented on 64-pin devices (PIC18F6XK22), read as '0'.

## 8.2 Address and Data Width

The PIC18F87K22 family of devices can be independently configured for different address and data widths on the same memory bus. Both address and data width are set by Configuration bits in the CONFIG3L register. As Configuration bits, this means that these options can only be configured by programming the device and are not controllable in software.

The BW bit selects an 8-bit or 16-bit data bus width. Setting this bit (default) selects a data width of 16 bits.

The ABW<1:0> bits determine both the program memory operating mode and the address bus width. The available options are 20-bit, 16-bit and 12-bit, as well as Microcontroller mode (external bus is disabled). Selecting a 16-bit or 12-bit width makes a corresponding number of high-order lines available for I/O functions. These pins are no longer affected by the setting of the EBDIS bit. For example, selecting a 16-Bit Addressing mode (ABW<1:0> = 01) disables A<19:16> and allows PORTH<3:0> to function without interruptions from the bus. Using the smaller address widths allows users to tailor the memory bus to the size of the external memory space for a particular design while freeing up pins for dedicated I/O operation.

Because the ABW bits have the effect of disabling pins for memory bus operations, it is important to always select an address width at least equal to the data width. If a 12-bit address width is used with a 16-bit data width, the upper four bits of data will not be available on the bus.

All combinations of address and data widths require multiplexing of address and data information on the same lines. The address and data multiplexing, as well as I/O ports made available by the use of smaller address widths, are summarized in Table 8-2.

## 8.2.1 ADDRESS SHIFTING ON THE EXTERNAL BUS

By default, the address presented on the external bus is the value of the PC. In practical terms, this means that addresses in the external memory device, below the top of on-chip memory, are unavailable to the microcontroller. To access these physical locations, the glue logic between the microcontroller and the external memory must somehow translate addresses.

To simplify the interface, the external bus offers an extension of Extended Microcontroller mode that automatically performs address shifting. This feature is controlled by the EASHFT Configuration bit. Setting this bit offsets addresses on the bus by the size of the microcontroller's on-chip program memory and sets the bottom address at 0000h. This allows the device to use the entire range of physical addresses of the external memory.

## 8.2.2 21-BIT ADDRESSING

As an extension of 20-bit address width operation, the External Memory Bus can also fully address a 2-Mbyte memory space. This is done by using the Bus Address Bit 0 (BA0) control line as the Least Significant bit of the address. The UB and LB control signals may also be used with certain memory devices to select the upper and lower bytes within a 16-bit wide data word.

This addressing mode is available in both 8-Bit and certain 16-Bit Data Width modes. Additional details are provided in Section 8.6.3 "16-Bit Byte Select Mode" and Section 8.7 "8-Bit Data Width Mode".

TABLE 0-2. AD	ADDRESS AND DATA LINES FOR DITTERENT ADDRESS AND DATA WIDTIS							
Data Width	Address Width	Multiplexed Data and Address Lines (and Corresponding Ports)	Address Only Lines (and Corresponding Ports)	Ports Available for I/O				
	12-bit		AD<11:8> (PORTE<3:0>)	PORTE<7:4>, All of PORTH				
8-bit	16-bit	AD<7:0> (PORTD<7:0>)	AD<15:8> (PORTE<7:0>)	All of PORTH				
	20-bit		A<19:16>, AD<15:8> (PORTH<3:0>, PORTE<7:0>)	_				
	16-bit	AD<15:0>	—	All of PORTH				
16-bit	20-bit	(PORTD<7:0>, PORTE<7:0>)	A<19:16> (PORTH<3:0>)	_				

## TABLE 8-2: ADDRESS AND DATA LINES FOR DIFFERENT ADDRESS AND DATA WIDTHS

## 8.3 Wait States

While it may be assumed that external memory devices will operate at the microcontroller clock rate, this is often not the case. In fact, many devices require longer times to write or retrieve data than the time allowed by the execution of table read or table write operations.

To compensate for this, the External Memory Bus can be configured to add a fixed delay to each table operation using the bus. Wait states are enabled by setting the WAIT Configuration bit. When enabled, the amount of delay is set by the WAIT<1:0> bits (MEMCON<5:4>). The delay is based on multiples of microcontroller instruction cycle time and is added following the instruction cycle when the table operation is executed. The range is from no delay to 3 Tcy (default value).

## 8.4 Port Pin Weak Pull-ups

With the exception of the upper address lines, A<19:16>, the pins associated with the External Memory Bus are equipped with weak pull-ups. The pull-ups are controlled by the upper three bits of the PADCFG1 register (PADCFG1<7:5>). They are named RDPU, REPU and RJPU, and control pull-ups on PORTD, PORTE and PORTJ, respectively. Setting one of these bits enables the corresponding pull-ups for that port. All pull-ups are disabled by default on all device Resets.

In Extended Microcontroller mode, the port pull-ups can be useful in preserving the memory state on the external bus while the bus is temporarily disabled (EBDIS = 1).

## 8.5 Program Memory Modes and the External Memory Bus

The PIC18F87K22 family of devices is capable of operating in one of two program memory modes, using combinations of on-chip and external program memory. The functions of the multiplexed port pins depend on the program memory mode selected, as well as the setting of the EBDIS bit.

In **Microcontroller Mode**, the bus is not active and the pins have their port functions only. Writes to the MEMCOM register are not permitted. The Reset value of EBDIS ('0') is ignored and the ABW pins behave as I/O ports.

In **Extended Microcontroller Mode**, the external program memory bus shares I/O port functions on the pins. When the device is fetching or doing table read/table write operations on the external program memory space, the pins will have the external bus function.

If the device is fetching and accessing internal program memory locations only, the EBDIS control bit will change the pins from external memory to I/O port functions. When EBDIS = 0, the pins function as the external bus. When EBDIS = 1, the pins function as I/O ports.

If the device fetches or accesses external memory while EBDIS = 1, the pins will switch to the external bus. If the EBDIS bit is set by a program executing from external memory, the action of setting the bit will be delayed until the program branches into the internal memory. At that time, the pins will change from external bus to I/O ports.

If the device is executing out of internal memory when EBDIS = 0, the memory bus address/data and control pins will not be active. They will go to a state where the active address/data pins are tri-state, the  $\overline{CE}$ ,  $\overline{OE}$ , WRH, WRL, UB and LB signals are '1', and ALE and BA0 are '0'. Note that only those pins associated with the current address width are forced to tri-state; the other pins continue to function as I/O. In the case of 16-bit address width, for example, only AD<15:0> (PORTD and PORTE) are affected; A<19:16> (PORTH<3:0>) continue to function as I/O.

In all external memory modes, the bus takes priority over any other peripherals that may share pins with it. This includes the Parallel Master Port (PMP) and serial communication modules which would otherwise take priority over the I/O port.

## 8.6 16-Bit Data Width Modes

In 16-Bit Data Width mode, the external memory interface can be connected to external memories in three different configurations:

- 16-Bit Byte Write
- 16-Bit Word Write
- 16-Bit Byte Select

The configuration to be used is determined by the WM<1:0> bits in the MEMCON register (MEMCON<1:0>). These three different configurations allow the designer maximum flexibility in using both 8-bit and 16-bit devices with 16-bit data.

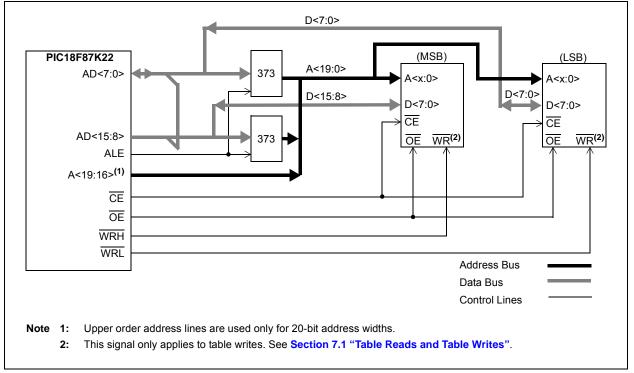
For all 16-bit modes, the Address Latch Enable (ALE) pin indicates that the Address bits, AD<15:0>, are available on the external memory interface bus. Following the address latch, the Output Enable (OE) signal will enable both bytes of program memory at once to form a 16-bit instruction word. The Chip Enable (CE signal) is active at any time that the microcontroller accesses external memory, whether reading or writing; it is inactive (asserted high) whenever the device is in Sleep mode.

In Byte Select mode, JEDEC standard Flash memories will require BA0 for the byte address line and one I/O line to select between Byte and Word mode. The other 16-bit modes do not need BA0. JEDEC standard static RAM memories will use the UB or LB signals for byte selection.

### 8.6.1 16-BIT BYTE WRITE MODE

Figure 8-1 shows an example of 16-Bit Byte Write mode for PIC18F87K22 family devices. This mode is used for two separate 8-bit memories connected for 16-bit operation. This generally includes basic EPROM and Flash devices. It allows table writes to byte-wide external memories. During a TBLWT instruction cycle, the TABLAT data is presented on the upper and <u>lower bytes</u> of the AD<15:0> bus. The appropriate WRH or WRL control line is strobed on the LSb of the TBLPTR.





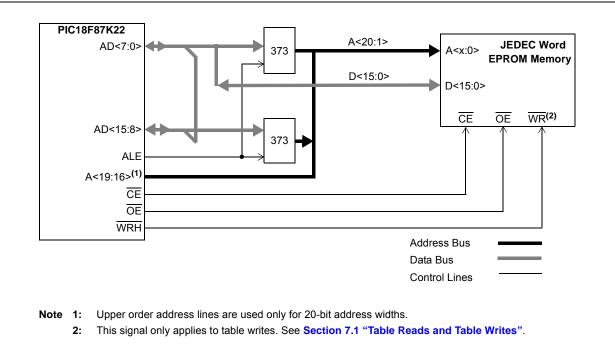
## 8.6.2 16-BIT WORD WRITE MODE

Figure 8-2 shows an example of 16-Bit Word Write mode for PIC18F87K22 family devices. This mode is used for word-wide memories, which includes some of the EPROM and Flash type memories. This mode allows opcode fetches and table reads from all forms of 16-bit memory, and table writes to any type of word-wide external memories. This method makes a distinction between TBLWT cycles to even or odd addresses.

During a TBLWT cycle to an even address (TBLPTR<0>= 0), the TABLAT data is transferred to a holding latch and the external address data bus is tri-stated for the data portion of the bus cycle. No write signals are activated.

During a TBLWT cycle to an odd address (TBLPTR<0> = 1), the TABLAT data is presented on the upper byte of the AD<15:0> bus. The contents of the holding latch are presented on the lower byte of the AD<15:0> bus.

<u>The WRH</u> signal is strobed for each write cycle; the WRL pin is unused. The signal on the BA0 pin indicates the LSb of the TBLPTR, but it is left unconnected. Instead, the UB and LB signals are active to select both bytes. The obvious limitation to this method is that the table write must be done in pairs on a specific word boundary to correctly write a word location.



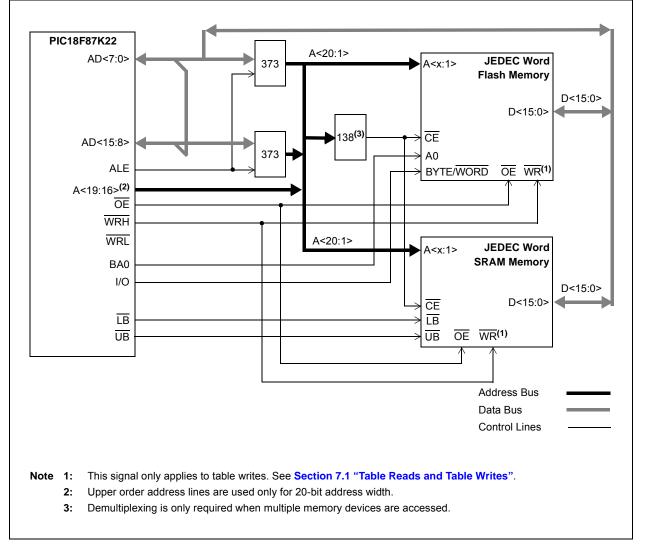
#### FIGURE 8-2: 16-BIT WORD WRITE MODE EXAMPLE

#### 8.6.3 16-BIT BYTE SELECT MODE

Figure 8-3 shows an example of 16-Bit Byte Select mode. This mode allows table write operations to word-wide external memories with byte selection capability. This generally includes both word-wide Flash and SRAM devices.

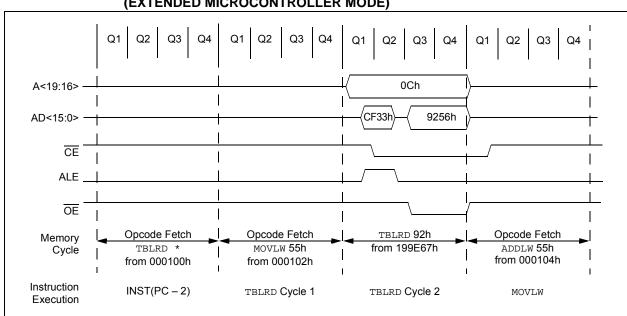
During a TBLWT cycle, the TABLAT data is presented on the upper and lower byte of the AD<15:0> bus. The WRH signal is strobed for each write cycle; the WRL pin is not used. The BA0 or UB/LB signals are used to select the byte to be written, based on the Least Significant bit of the TBLPTR register. Flash and SRAM devices use different control signal combinations to implement Byte Select mode. JEDEC standard Flash memories require that a controller I/O port pin be connected to the memory's BYTE/WORD pin to provide the select signal. They also use the BA0 signal from the controller as a byte address. JEDEC standard static RAM memories, on the other hand, use the UB or LB signals to select the byte.





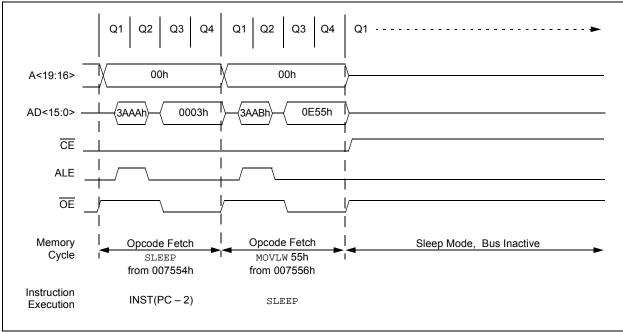
#### 8.6.4 16-BIT MODE TIMING

The presentation of control signals on the External Memory Bus is different for the various operating modes. Typical signal timing diagrams are shown in Figure 8-4 and Figure 8-5.



#### FIGURE 8-4: EXTERNAL MEMORY BUS TIMING FOR TBLRD (EXTENDED MICROCONTROLLER MODE)

#### FIGURE 8-5: EXTERNAL MEMORY BUS TIMING FOR SLEEP (EXTENDED MICROCONTROLLER MODE)



## 8.7 8-Bit Data Width Mode

In 8-Bit Data Width mode, the External Memory Bus operates only in Multiplexed mode; that is, data shares the 8 Least Significant bits of the address bus.

Figure 8-6 shows an example of 8-Bit Multiplexed mode for PIC18F8XK22 devices. This mode is used for a single, 8-bit memory connected for 16-bit operation. The instructions will be fetched as two 8-bit bytes on a shared data/address bus. The two bytes are sequentially fetched within one instruction cycle (TcY). Therefore, the designer must choose external memory devices according to timing calculations based on 1/2 TcY (2 times the instruction rate). For proper memory speed selection, glue logic propagation delay times must be considered, along with setup and hold times.

The Address Latch Enable (ALE) pin indicates that the Address bits, AD<15:0>, are available on the External Memory Bus interface. The Output Enable (OE) signal

will enable one byte of program memory for a portion of the instruction cycle, then BA0 will change and the second byte will be enabled to form the 16-bit instruction word. The Least Significant bit of the address, BA0, must be connected to the memory devices in this mode. The Chip Enable (CE) signal is active at any time that the microcontroller accesses external memory, whether reading or writing. It is inactive (asserted high) whenever the device is in Sleep mode.

This generally includes basic EPROM and Flash devices. It allows table writes to byte-wide external memories.

During a TBLWT instruction cycle, the TABLAT data is presented on the upper and lower bytes of the AD<15:0> bus. The appropriate level of the BA0 control line is strobed on the LSb of the TBLPTR.

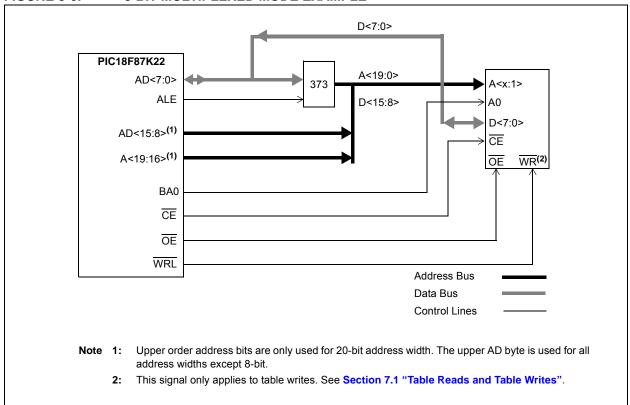
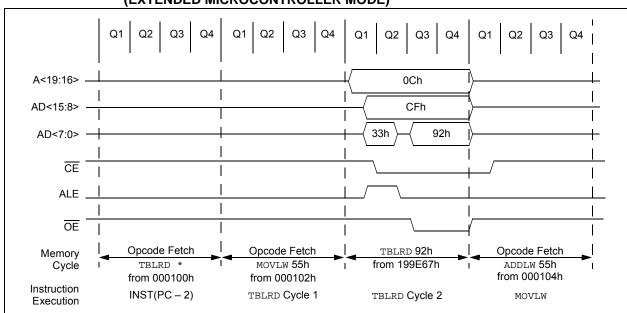


FIGURE 8-6: 8-BIT MULTIPLEXED MODE EXAMPLE

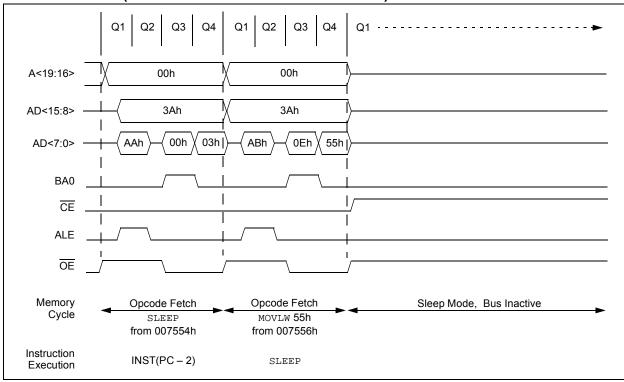
#### 8.7.1 8-BIT MODE TIMING

The presentation of control signals on the External Memory Bus is different for the various operating modes. Typical signal timing diagrams are shown in Figure 8-7 and Figure 8-8.



#### FIGURE 8-7: EXTERNAL MEMORY BUS TIMING FOR TBLRD (EXTENDED MICROCONTROLLER MODE)

## FIGURE 8-8: EXTERNAL MEMORY BUS TIMING FOR SLEEP (EXTENDED MICROCONTROLLER MODE)



### 8.8 Operation in Power-Managed Modes

In alternate, power-managed Run modes, the external bus continues to operate normally. If a clock source with a lower speed is selected, bus operations will run at that speed. In these cases, excessive access times for the external memory may result if Wait states have been enabled and added to external memory operations. If operations in a lower power Run mode are anticipated, users should provide in their applications for adjusting memory access times at the lower clock speeds. In Sleep and Idle modes, the microcontroller core does not need to access data; bus operations are suspended. The state of the external bus is frozen, with the address/data pins, and most of the control pins, holding at the same state they were in when the mode was invoked. The only potential changes are to the  $\overline{CE}$ ,  $\overline{LB}$  and  $\overline{UB}$  pins, which are held at logic high.

<b>TABLE 8-3</b> :	REGISTERS ASSOCIATED WITH THE EXTERNAL MEMORY BUS

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
MEMCON <sup>(1)</sup>	EBDIS	_	WAIT1	WAIT0		—	WM1	WM0
PADCFG1	RDPU	REPU	RJPU <sup>(1)</sup>	_	—	RTSECSEL1	RTSECSEL0	
PMD1	PSPMD	CTMUMD	RTCCMD	TMR4MD	TMR3MD	TMR2MD	TMR1MD	EMBMD

**Legend:** — = unimplemented, read as '0'. Shaded cells are not used during External Memory Bus access.

Note 1: Unimplemented in 64-pin devices (PIC18F6XK22), read as '0'.

NOTES:

## 9.0 DATA EEPROM MEMORY

The data EEPROM is a nonvolatile memory array, separate from the data RAM and program memory, that is used for long-term storage of program data. It is not directly mapped in either the register file or program memory space, but is indirectly addressed through the Special Function Registers (SFRs). The EEPROM is readable and writable during normal operation over the entire VDD range.

Five SFRs are used to read and write to the data EEPROM, as well as the program memory. They are:

- EECON1
- EECON2
- EEDATA
- EEADR
- EEADRH

The data EEPROM allows byte read and write. When interfacing to the data memory block, EEDATA holds the 8-bit data for read/write and the EEADRH:EEADR register pair holds the address of the EEPROM location being accessed.

The EEPROM data memory is rated for high erase/write cycle endurance. A byte write automatically erases the location and writes the new data (erase-before-write). The write time is controlled by an on-chip timer; it will vary with voltage and temperature, as well as from chip-to-chip. Please refer to Parameter D122 (Table 31-1 in Section 31.0 "Electrical Characteristics") for exact limits.

## 9.1 EEADR and EEADRH Registers

The EEADRH:EEADR register pair is used to address the data EEPROM for read and write operations. EEADRH holds the two MSbs of the address; the upper 6 bits are ignored. The 10-bit range of the pair can address a memory range of 1024 bytes (00h to 3FFh).

## 9.2 EECON1 and EECON2 Registers

Access to the data EEPROM is controlled by two registers: EECON1 and EECON2. These are the same registers which control access to the program memory and are used in a similar manner for the data EEPROM.

The EECON1 register (Register 9-1) is the control register for data and program memory access. Control bit, EEPGD, determines if the access will be to program memory or data EEPROM memory. When clear, operations will access the data EEPROM memory. When set, program memory is accessed.

Control bit, CFGS, determines if the access will be to the Configuration registers or to program memory/data EEPROM memory. When set, subsequent operations access Configuration registers. When CFGS is clear, the EEPGD bit selects either program Flash or data EEPROM memory.

The WREN bit, when set, will allow a write operation. On power-up, the WREN bit is clear. The WRERR bit is set in hardware when the WREN bit is set, and cleared, when the internal programming timer expires and the write operation is complete.

Note:	During normal operation, the WRERR is				
	read as '1'. This can indicate that a write				
	operation was prematurely terminated by				
	a Reset or a write operation was				
	attempted improperly.				

The WR control bit initiates write operations. The bit cannot be cleared, only set, in software. It is cleared in hardware at the completion of the write operation.

Note:	The EEIF interrupt flag bit (PIR6<4>) is
	set when the write is complete; it must be
	cleared in software.

Control bits, RD and WR, start read and erase/write operations, respectively. These bits are set by firmware and cleared by hardware at the completion of the operation.

The RD bit cannot be set when accessing program memory (EEPGD = 1). Program memory is read using table read instructions. See **Section 7.1 "Table Reads and Table Writes"** regarding table reads.

The EECON2 register is not a physical register. It is used exclusively in the memory write and erase sequences. Reading EECON2 will read all '0's.

R/W-x	R/W-x	U-0	R/W-0	R/W-x	R/W-0	R/S-0	R/S-0		
EEPGD	CFGS	_	FREE	WRERR <sup>(1)</sup>	WREN	WR	RD		
bit 7	•						bit		
Legend:		S = Settable	bit						
R = Readabl	e bit	W = Writable		U = Unimpler	nented bit, rea	d as '0'			
-n = Value at	POR	'1' = Bit is set	t	'0' = Bit is cle		x = Bit is unkr	nown		
bit 7	EEPGD: F	ash Program or I	Data EEPROI	M Memory Sele	ct bit				
		Flash program r							
		data EEPROM I	-						
bit 6		sh Program/Data		Configuration S	Select bit				
		Configuration re Flash program o	•	OM memory					
bit 5		ented: Read as '		emmennery					
bit 4	-	sh Row Erase En							
	1 = Erase	L = Erase the program memory row addressed by TBLPTR on the next WR command (cleared by							
	•	completion of erase operation)							
		m write-only		(1)					
bit 3		lash Program/Da							
		e operation is pre ion or an imprope	•	· •	set during sen-	umed programi	ning in norma		
		rite operation has							
bit 2	WREN: Flash Program/Data EEPROM Write Enable bit								
	1 = Allows	write cycles to F	lash program	/data EEPROM					
		s write cycles to F	-lash program	n/data EEPROM	1				
bit 1	WR: Write								
		s a data EEPROI							
	•	(The operation is self-timed and the bit is cleared by hardware once the write is complete. The WR bit can only be set (not cleared) in software.)							
		cycle to the EEPF			,				
bit 0	RD: Read	Control bit							
		s an EEPROM re		al in the analysis are					
		takes one cycle. re. The RD bit ca					tot cleared)		
		not initiate an EE			- 5, 5, 5, 50 - 1	•••			

#### REGISTER 9-1: EECON1: DATA EEPROM CONTROL REGISTER 1

**Note 1:** When a WRERR occurs, the EEPGD and CFGS bits are not cleared. This allows tracing of the error condition.

## 9.3 Reading the Data EEPROM Memory

To read a data memory location, the user must write the address to the EEADRH:EEADR register pair, clear the EEPGD control bit (EECON1<7>) and then set control bit, RD (EECON1<0>). The data is available in the EEDATA register after one cycle; therefore, it can be read after one NOP instruction. EEDATA will hold this value until another read operation or until it is written to by the user (during a write operation).

The basic process is shown in Example 9-1.

## 9.4 Writing to the Data EEPROM Memory

To write an EEPROM data location, the address must first be written to the EEADRH:EEADR register pair and the data written to the EEDATA register. The sequence in Example 9-2 must be followed to initiate the write cycle.

The write will not begin if this sequence is not exactly followed (write 0x55 to EECON2, write 0xAA to EECON2, then set WR bit) for each byte. It is strongly recommended that interrupts be disabled during this code segment.

Additionally, the WREN bit in EECON1 must be set to enable writes. This mechanism prevents accidental writes to data EEPROM due to unexpected code execution (i.e., runaway programs). The WREN bit should be kept clear at all times, except when updating the EEPROM. The WREN bit is not cleared by hardware. After a write sequence has been initiated, EECON1, EEADRH:EEADR and EEDATA cannot be modified. The WR bit will be inhibited from being set unless the WREN bit is set. The WREN bit must be set on a previous instruction. Both WR and WREN cannot be set with the same instruction.

At the completion of the write cycle, the WR bit is cleared in hardware and the EEPROM Interrupt Flag bit (EEIF) is set. The user may either enable this interrupt, or poll this bit. EEIF must be cleared by software.

## 9.5 Write Verify

Depending on the application, good programming practice may dictate that the value written to the memory should be verified against the original value. This should be used in applications where excessive writes can stress bits near the specification limit.

Note: Self-write execution to Flash and EEPROM memory cannot be done while running in LP Oscillator mode (Low-Power mode). Therefore, executing a self-write will put the device into High-Power mode.

#### EXAMPLE 9-1: DATA EEPROM READ

MOVLW	DATA_EE_ADDRH	;
MOVWF	EEADRH	; Upper bits of Data Memory Address to read
MOVLW	DATA_EE_ADDR	i
MOVWF	EEADR	; Lower bits of Data Memory Address to read
BCF	EECON1, EEPGD	; Point to DATA memory
BCF	EECON1, CFGS	; Access EEPROM
BSF	EECON1, RD	; EEPROM Read
NOP		
MOVF	EEDATA, W	; W = EEDATA

#### EXAMPLE 9-2: DATA EEPROM WRITE

	MOVLW MOVWF MOVLW MOVWF MOVLW MOVWF	DATA_EE_ADDRH EEADRH DATA_EE_ADDR EEADR DATA_EE_DATA EEDATA	; ; Upper bits of Data Memory Address to write ; ; Lower bits of Data Memory Address to write ; ; Data Memory Value to write
	BCF	EECON1, EEPGD	; Point to DATA memory
	BCF	EECON1, CFGS	; Access EEPROM
	BSF	EECON1, WREN	; Enable writes
	BCF	INTCON, GIE	; Disable Interrupts
	MOVLW	0x55	;
Required	MOVWF	EECON2	; Write 55h
Sequence	MOVLW	0xAA	i
	MOVWF	EECON2	; Write OAAh
	BSF	EECON1, WR	; Set WR bit to begin write
	BTFSC	EECON1, WR	; Wait for write to complete GOTO \$-2
	BSF	INTCON, GIE	; Enable Interrupts
			· Harr godo ovogution
	DOD		; User code execution
	BCF	EECON1, WREN	; Disable writes on write complete (EEIF set)

## 9.6 Operation During Code-Protect

Data EEPROM memory has its own code-protect bits in the Configuration Words. External read and write operations are disabled if code protection is enabled.

The microcontroller itself can both read and write to the internal data EEPROM, regardless of the state of the code-protect Configuration bit. Refer to Section 28.0 "Special Features of the CPU" for additional information.

## 9.7 Protection Against Spurious Write

There are conditions when the device may not want to write to the data EEPROM memory. To protect against spurious EEPROM writes, various mechanisms have been implemented. On power-up, the WREN bit is cleared. In addition, writes to the EEPROM are blocked during the Power-up Timer period (TPWRT, Parameter 33 in Table 31-13).

The write initiate sequence, and the WREN bit together, help prevent an accidental write during brown-out, power glitch or software malfunction.

## 9.8 Using the Data EEPROM

The data EEPROM is a high-endurance, byte addressable array that has been optimized for the storage of frequently changing information (e.g., program variables or other data that is updated often). Frequently changing values will typically be updated more often than Specification D124. If this is the case, an array refresh must be performed. For this reason, variables that change infrequently (such as constants, IDs, calibration, etc.) should be stored in Flash program memory.

A simple data EEPROM refresh routine is shown in Example 9-3.

Note: If data EEPROM is only used to store constants and/or data that changes often, an array refresh is likely not required. See Specification D124.

	CLRF	EEADR EEADRH	; Start at address 0 ;
	BCF	EECON1, CFGS	
	BCF		; Set for Data EEPROM
	BCF	INTCON, GIE	; Disable interrupts
	BSF	EECON1, WREN	; Enable writes
LOOP			; Loop to refresh array
	BSF	EECON1, RD	; Read current address
	MOVLW	0x55	;
	MOVWF	EECON2	; Write 55h
	MOVLW	0xAA	;
	MOVWF	EECON2	; Write OAAh
	BSF	EECON1, WR	; Set WR bit to begin write
	BTFSC	EECON1, WR	; Wait for write to complete
	BRA	\$-2	
	INCFSZ	EEADR, F	; Increment address
	BRA	LOOP	; Not zero, do it again
	INCFSZ	EEADRH, F	; Increment the high address
	BRA	LOOP	; Not zero, do it again
	BCF	EECON1, WREN	; Disable writes
	BSF	INTCON, GIE	; Enable interrupts

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
INTCON	GIE/GIEH	PEIE/GIEL	TMR0IE	INT0IE	RBIE	TMR0IF	INT0IF	RBIF
EEADRH	EEPROM A	EEPROM Address Register High Byte						
EEADR	EEPROM A	EEPROM Address Register Low Byte						
EEDATA	EEPROM Data Register							
EECON2	EEPROM Control Register 2 (not a physical register)							
EECON1	EEPGD	CFGS	_	FREE	WRERR	WREN	WR	RD
IPR6	_	_	_	EEIP	—	CMP3IP	CMP2IP	CMP1IP
PIR6	_		_	EEIF	_	CMP3IF	CMP2IF	CMP1IF
PIE6		_	_	EEIE	_	CMP3IE	CMP2IE	CMP1IE

Legend: — = unimplemented, read as '0'. Shaded cells are not used during Flash/EEPROM access.

## 10.0 8 x 8 HARDWARE MULTIPLIER

## 10.1 Introduction

All PIC18 devices include an 8 x 8 hardware multiplier as part of the ALU. The multiplier performs an unsigned operation and yields a 16-bit result that is stored in the product register pair, PRODH:PRODL. The multiplier's operation does not affect any flags in the STATUS register.

Making multiplication a hardware operation allows it to be completed in a single instruction cycle. This has the advantages of higher computational throughput and reduced code size for multiplication algorithms and allows PIC18 devices to be used in many applications previously reserved for digital-signal processors. A comparison of various hardware and software multiply operations, along with the savings in memory and execution time, is shown in Table 10-1.

## 10.2 Operation

Example 10-1 shows the instruction sequence for an 8 x 8 unsigned multiplication. Only one instruction is required when one of the arguments is already loaded in the WREG register.

Example 10-2 shows the sequence to do an 8 x 8 signed multiplication. To account for the sign bits of the arguments, each argument's Most Significant bit (MSb) is tested and the appropriate subtractions are done.

#### EXAMPLE 10-1: 8 x 8 UNSIGNED MULTIPLY ROUTINE

MOVF	ARG1,	W	;
MULWF	ARG2		; ARG1 * ARG2 ->
			; PRODH:PRODL

#### EXAMPLE 10-2: 8 x 8 SIGNED MULTIPLY ROUTINE

MOVF	ARG1, W			
MULWF	ARG2	;	ARG1 * ARG2 ->	
		;	PRODH:PRODL	
BTFSC	ARG2, SB	;	Test Sign Bit	
SUBWF	PRODH, F	;	PRODH = PRODH	
		;	- ARG1	
MOVF	ARG2, W			
BTFSC	ARG1, SB	;	Test Sign Bit	
SUBWF	PRODH, F	;	PRODH = PRODH	
		;	- ARG2	

		Program		Time				
Routine	Multiply Method	Memory (Words)	Cycles (Max)	@ 64 MHz	@ 48 MHz	@ 10 MHz	@ 4 MHz	
9 x 9 Lineigned	Without Hardware Multiply	13	69	4.3 μs	5.7 μs	27.6 μs	69 μs	
8 x 8 Unsigned	Hardware Multiply	1	1	62.5 ns	83.3 ns	400 ns	1 μs	
8 x 8 Signed	Without Hardware Multiply	33	91	5.6 μs	7.5 μs	36.4 μs	91 μs	
	Hardware Multiply	6	6	375 ns	500 ns	2.4 μs	6 μs	
16 x 16	Without Hardware Multiply	21	242	15.1 μs	20.1 μs	96.8 μs	242 μs	
Unsigned	Hardware Multiply	28	28	1.7 μs	2.3 μs	11.2 μs	28 μs	
10 × 10 Signad	Without Hardware Multiply	52	254	15.8 μs	21.2 μs	101.6 μs	254 μs	
16 x 16 Signed	Hardware Multiply	35	40	2.5 μs	3.3 μs	16.0 μs	40 μs	

## TABLE 10-1: PERFORMANCE COMPARISON FOR VARIOUS MULTIPLY OPERATIONS

Example 10-3 shows the sequence to do a 16 x 16 unsigned multiplication. Equation 10-1 shows the algorithm that is used. The 32-bit result is stored in four registers (RES3:RES0).

#### EQUATION 10-1: 16 x 16 UNSIGNED MULTIPLICATION ALGORITHM

RES3:RES0	=	ARG1H:ARG1L • ARG2H:ARG2L
	=	$(ARG1H \bullet ARG2H \bullet 2^{16}) +$
		$(ARG1H \bullet ARG2L \bullet 2^8) +$
		$(ARG1L \bullet ARG2H \bullet 2^8) +$
		(ARG1L • ARG2L)

#### EXAMPLE 10-3: 16 x 16 UNSIGNED MULTIPLY ROUTINE

	MOVF	ARG1L, W	
	MULWF	ARG2L	; ARG1L * ARG2L->
			; PRODH:PRODL
	MOVFF	PRODH, RES1	;
	MOVFF	PRODL, RESO	;
;			
	MOVF	ARG1H, W	
	MULWF	ARG2H	; ARG1H * ARG2H->
			; PRODH:PRODL
	MOVFF	PRODH, RES3	;
	MOVFF	PRODL, RES2	;
;			
	MOVF	ARG1L, W	
	MULWF	ARG2H	; ARG1L * ARG2H->
			; PRODH:PRODL
	MOVF	PRODL, W	;
	ADDWF	RES1, F	; Add cross
		PRODH, W	; products
	ADDWFC	RES2, F	;
	CLRF	WREG	;
	ADDWFC	RES3, F	;
;			
		ARG1H, W	;
	MULWF	ARG2L	; ARG1H * ARG2L->
			; PRODH:PRODL
	MOVF		;
	ADDWF	RES1, F	; Add cross
			; products
	ADDWFC	RES2, F	;
	CLRF		;
	ADDWFC	RES3, F	;

Example 10-4 shows the sequence to do a 16 x 16 signed multiply. Equation 10-2 shows the algorithm used. The 32-bit result is stored in four registers (RES3:RES0). To account for the sign bits of the arguments, the MSb for each argument pair is tested and the appropriate subtractions are done.

#### EQUATION 10-2: 16 x 16 SIGNED MULTIPLICATION ALGORITHM

RES3:RES0=	ARG1H:ARG1L • ARG2H:ARG2L
=	$(ARG1H \bullet ARG2H \bullet 2^{16}) +$
	$(ARG1H \bullet ARG2L \bullet 2^8) +$
	$(ARG1L \bullet ARG2H \bullet 2^8) +$
	$(ARG1L \bullet ARG2L) +$
	$(-1 \bullet ARG2H < 7 > \bullet ARG1H:ARG1L \bullet 2^{16}) +$
	$(-1 \bullet ARG1H < 7 > \bullet ARG2H: ARG2L \bullet 2^{16})$

#### EXAMPLE 10-4: 16 x 16 SIGNED MULTIPLY ROUTINE

	MOVF	ARG1L, W		
	MULWF	ARG2L	;	ARG1L * ARG2L ->
				PRODH:PRODL
	MOVFF	PRODH, RES1	;	
	MOVFF	PRODL, RESO		
;		- ,		
	MOVF	ARG1H, W		
	MULWF	ARG2H	;	ARG1H * ARG2H ->
	ноциг	AROZII		PRODH:PRODL
	MOVFF	PRODH, RES3		TRODITIRODE
	MOVFF			
	MOVEL	FRODE, RESZ	'	
;	MOVE			
		ARG1L, W		
	MULWF	ARG2H		ARG1L * ARG2H ->
				PRODH:PRODL
	MOVF	PRODL, W	;	
		RES1, F		Add cross
	MOVF		;	products
	ADDWFC	RES2, F	;	
	CLRF	WREG	;	
	ADDWFC	RES3, F	;	
;				
	MOVF	ARG1H, W	;	
	MULWF	ARG2L	;	ARG1H * ARG2L ->
			;	PRODH:PRODL
	MOVF	PRODL, W	;	
	ADDWF	RES1, F	;	Add cross
	MOVF	PRODH, W	;	products
	ADDWFC	RES2, F	;	-
	CLRF	WREG	;	
	ADDWFC	RES3, F	;	
;				
	BTFSS	ARG2H, 7	;	ARG2H:ARG2L neg?
	BRA	SIGN_ARG1		no, check ARG1
	MOVF	ARG1L, W	;	, 0110011 11101
	SUBWF	RES2	;	
	MOVF	ARG1H, W	;	
	SUBWFB	RES3		SIGN ARG1
	BTFSS			ARG1H:ARG1L neg?
	BRA	CONT_CODE		no, done
	MOVF	ARG2L, W	;	
	SUBWF	RES2	;	
	MOVF	ARG2H, W	;	
	SUBWFB	RES3		
;				
CC	NT_CODE			
	:			

## 11.0 INTERRUPTS

Members of the PIC18F87K22 family of devices have multiple interrupt sources and an interrupt priority feature that allows most interrupt sources to be assigned a high-priority level or a low-priority level. The high-priority interrupt vector is at 0008h and the low-priority interrupt vector is at 0018h. High-priority interrupt events will interrupt any low-priority interrupts that may be in progress.

The registers for controlling interrupt operation are:

- RCON
- INTCON
- INTCON2
- INTCON3
- PIR1, PIR2, PIR3
- PIE1, PIE2, PIE3
- IPR1, IPR2, IPR3

It is recommended that the Microchip header files supplied with MPLAB<sup>®</sup> IDE be used for the symbolic bit names in these registers. This allows the assembler/compiler to automatically take care of the placement of these bits within the specified register.

In general, interrupt sources have three bits to control their operation. They are:

- Flag bit Indicating that an interrupt event occurred
- Enable bit Enabling program execution to branch to the interrupt vector address when the flag bit is set
- Priority bit Specifying high priority or low priority

The interrupt priority feature is enabled by setting the IPEN bit (RCON<7>). When interrupt priority is enabled, there are two bits that enable interrupts globally. Setting the GIEH bit (INTCON<7>) enables all interrupts that have the priority bit set (high priority). Setting the GIEL bit (INTCON<6>) and GIEH bit (INTCON<7>) enables all interrupts that have the priority bit cleared (low priority). When the interrupt flag, enable bit and appropriate Global Interrupt Enable bit are set, the interrupt will vector immediately to address, 0008h or 0018h, depending on the priority bit setting. Individual interrupts can be disabled through their corresponding enable bits.

When the IPEN bit is cleared (default state), the interrupt priority feature is disabled and interrupts are compatible with PIC<sup>®</sup> mid-range devices. In Compatibility mode, the interrupt priority bits for each source have no effect. INTCON<6> is the PEIE bit that enables/disables all peripheral interrupt sources. INTCON<7> is the GIE bit that enables/disables all interrupt sources. All interrupts branch to address, 0008h, in Compatibility mode.

When an interrupt is responded to, the Global Interrupt Enable bit is cleared to disable further interrupts. If the IPEN bit is cleared, this is the GIE bit. If interrupt priority levels are used, this will be either the GIEH or GIEL bit. High-priority interrupt sources can interrupt a low-priority interrupt. Low-priority interrupts are not processed while high-priority interrupts are in progress.

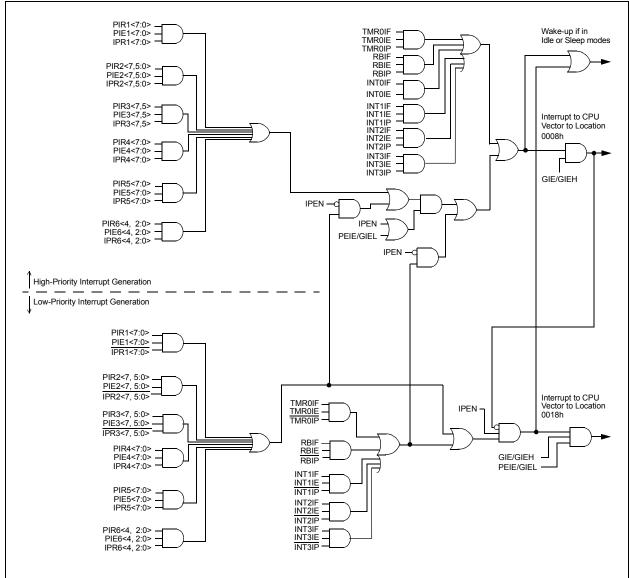
The return address is pushed onto the stack and the PC is loaded with the interrupt vector address (0008h or 0018h). Once in the Interrupt Service Routine (ISR), the source(s) of the interrupt can be determined by polling the interrupt flag bits. The interrupt flag bits must be cleared in software, before re-enabling interrupts, to avoid recursive interrupts.

The "return from interrupt" instruction, RETFIE, exits the interrupt routine and sets the GIE bit (GIEH or GIEL if priority levels are used) that re-enables interrupts.

For external interrupt events, such as the INTx pins or the PORTB input change interrupt, the interrupt latency will be three to four instruction cycles. The exact latency is the same for one or two-cycle instructions. Individual interrupt flag bits are set regardless of the status of their corresponding enable bit or the GIE bit.

**Note:** Do not use the MOVFF instruction to modify any of the Interrupt Control registers while **any** interrupt is enabled. Doing so may cause erratic microcontroller behavior.

## FIGURE 11-1: PIC18F87K22 FAMILY INTERRUPT LOGIC



### 11.1 INTCON Registers

The INTCON registers are readable and writable registers that contain various enable, priority and flag bits.

Note: Interrupt flag bits are set when an interrupt condition occurs regardless of the state of its corresponding enable bit or the Global Interrupt Enable bit. User software should ensure the appropriate interrupt flag bits are clear prior to enabling an interrupt. This feature allows for software polling.

#### REGISTER 11-1: INTCON: INTERRUPT CONTROL REGISTER

GIE/GIEH         PEIE/GIEL         TMR0IE         INT0IE         RBIE         TMR0IF         INT0IF         RBIF <sup>(1)</sup> bit 7         bit 0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-x
bit 7 bit 0	GIE/GIEH	PEIE/GIEL	TMR0IE	INT0IE	RBIE	TMR0IF	INT0IF	RBIF <sup>(1)</sup>
	bit 7							bit 0

Legend:			
R = Readable bit	W = Writable bit	U = Unimplemented bit	, read as '0'
-n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown

bit 7	GIE/GIEH: Global Interrupt Enable bit
	When IPEN = <u>0</u> :
	1 = Enables all unmasked interrupts
	0 = Disables all interrupts
	<u>When IPEN = 1:</u> 1 = Enables all high-priority interrupts
	0 = Disables all interrupts
bit 6	PEIE/GIEL: Peripheral Interrupt Enable bit
	When IPEN = $0$ :
	1 = Enables all unmasked peripheral interrupts
	0 = Disables all peripheral interrupts
	When IPEN = 1:
	<ol> <li>1 = Enables all low-priority peripheral interrupts</li> <li>0 = Disables all low-priority peripheral interrupts</li> </ol>
bit 5	TMR0IE: TMR0 Overflow Interrupt Enable bit
	1 = Enables the TMR0 overflow interrupt
	0 = Disables the TMR0 overflow interrupt
bit 4	INTOIE: INTO External Interrupt Enable bit
	1 = Enables the INT0 external interrupt
	0 = Disables the INT0 external interrupt
bit 3	RBIE: RB Port Change Interrupt Enable bit
	1 = Enables the RB port change interrupt
1.11.0	0 = Disables the RB port change interrupt
bit 2	<b>TMR0IF:</b> TMR0 Overflow Interrupt Flag bit
	<ul> <li>1 = TMR0 register has overflowed (must be cleared in software)</li> <li>0 = TMR0 register did not overflow</li> </ul>
bit 1	INTOIF: INTO External Interrupt Flag bit
	1 = The INT0 external interrupt occurred (must be cleared in software)
	0 = The INT0 external interrupt did not occur
bit 0	RBIF: RB Port Change Interrupt Flag bit <sup>(1)</sup>
	1 = At least one of the RB<7:4> pins changed state (must be cleared in software)
	0 = None of the RB<7:4> pins have changed state
Note 1:	A mismatch condition will continue to set this bit. Reading PORTB, and then waiting one additional instruction
	avela, will and the minimatch condition and allow the bit to be degred

cycle, will end the mismatch condition and allow the bit to be cleared.

R/W-1	R/W-1	R/W-1	R/W-1	R/W-1	R/W-1	R/W-1	R/W-1	
RBPU	INTEDG0	INTEDG1	INTEDG2	INTEDG3	TMR0IP	INT3IP	RBIP	
bit 7							bit (	
Legend:								
R = Readable bit		W = Writable bit		U = Unimplemented bit, read as '0'				
-n = Value at POR		'1' = Bit is set		'0' = Bit is cleared		x = Bit is unknown		
bit 7	RBPU: PORT	B Pull-up Enat	ole bit					
	1 = All PORTB pull-ups are disabled							
	0 = PORTB pull-ups are enabled by individual TRIS register values							
bit 6	INTEDG0: External Interrupt 0 Edge Select bit							
	1 = Interrupt on rising edge							
	0 = Interrupt on falling edge							
bit 5	INTEDG1: External Interrupt 1 Edge Select bit							
	1 = Interrupt on rising edge 0 = Interrupt on falling edge							
bit 4	INTEDG2: External Interrupt 2 Edge Select bit							
	1 = Interrupt on rising edge							
	0 = Interrupt on falling edge							
bit 3	INTEDG3: External Interrupt 3 Edge Select bit							
	1 = Interrupt on rising edge							
	0 = Interrupt on falling edge							
bit 2		R0 Overflow Int	errupt Priority	bit				
	1 = High priority 0 = Low priority							
bit 1	•	2	upt Drigrity bit					
	INT3IP: INT3 External Interrupt Priority bit 1 = High priority							
	0 = Low priority							
bit 0	RBIP: RB Port Change Interrupt Priority bit							
	1 = High priority							
	0 = Low priority							

#### REGISTER 11-2: INTCON2: INTERRUPT CONTROL REGISTER 2

**Note:** Interrupt flag bits are set when an interrupt condition occurs, regardless of the state of its corresponding enable bit or the Global Interrupt Enable bit. User software should ensure the appropriate interrupt flag bits are clear prior to enabling an interrupt. This feature allows for software polling.

#### REGISTER 11-3: INTCON3: INTERRUPT CONTROL REGISTER 3

R/W-1	R/W-1	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0
INT2IP	INT1IP	INT3IE	INT2IE	INT1IE	INT3IF	INT2IF	INT1IF
bit 7							bit
Legend:							
R = Reada	able bit	W = Writable	bit	U = Unimpler	nented bit, rea	d as '0'	
-n = Value		'1' = Bit is set		'0' = Bit is cle		x = Bit is unkr	nown
bit 7	1 = High prio	2	upt Priority bit				
bit 6	0 = Low prior INT1IP: INT1 1 = High prio 0 = Low prior	External Interr	upt Priority bit				
bit 5	INT3IE: INT3 1 = Enables 1	External Interr the INT3 extern the INT3 extern	nal interrupt				
bit 4	1 = Enables	External Interr the INT2 extern the INT2 exter	nal interrupt				
bit 3	1 = Enables	External Interr the INT1 extern the INT1 exter	nal interrupt				
bit 2	1 = The INT3	External Interr external interr external interr	upt occurred (	must be cleare	d in software)		
bit 1	INT2IF: INT2 1 = The INT2	External Interr	upt Flag bit upt occurred (	must be cleare	d in software)		
bit 0	INT1IF: INT1 1 = The INT1	External Interr	upt Flag bit upt occurred (	must be cleare	d in software)		
Note:	Interrupt flag bits enable bit or the 0 are clear prior to	Global Interrupt	Enable bit. Us	er software sho	ould ensure the	e appropriate int	

#### 11.2 PIR Registers

The PIR registers contain the individual flag bits for the peripheral interrupts. Due to the number of peripheral interrupt sources, there are six Peripheral Interrupt Request (Flag) registers (PIR1 through PIR6).

- Note 1: Interrupt flag bits are set when an interrupt condition occurs, regardless of the state of its corresponding enable bit or the Global Interrupt Enable bit, GIE (INTCON<7>).
  - 2: User software should ensure the appropriate interrupt flag bits are cleared prior to enabling an interrupt and after servicing that interrupt.

#### REGISTER 11-4: PIR1: PERIPHERAL INTERRUPT REQUEST (FLAG) REGISTER 1

R/W-0	R/W-0	R-0	R-0	R/W-0	R/W-0	R/W-0	R/W-0
PSPIF	ADIF	RC1IF	TX1IF	SSP1IF	TMR1GIF	TMR2IF	TMR1IF
bit 7							bit 0

Legend:			
R = Readable bit	W = Writable bit	U = Unimplemented bit, read	1 as '0'
-n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown

bit 7	PSPIF: Parallel Slave Port Read/Write Interrupt Flag bit
	<ul> <li>1 = A read or write operation has taken place (must be cleared in software)</li> <li>0 = No read or write operation has occurred</li> </ul>
bit 6	ADIF: A/D Converter Interrupt Flag bit
	<ul> <li>1 = An A/D conversion completed (must be cleared in software)</li> <li>0 = The A/D conversion is not complete</li> </ul>
bit 5	RC1IF: EUSART Receive Interrupt Flag bit
	<ul> <li>1 = The EUSART receive buffer, RCREG1, is full (cleared when RCREG1 is read)</li> <li>0 = The EUSART receive buffer is empty</li> </ul>
bit 4	TX1IF: EUSART Transmit Interrupt Flag bit
	<ul> <li>1 = The EUSART transmit buffer, TXREG1, is empty (cleared when TXREG1 is written)</li> <li>0 = The EUSART transmit buffer is full</li> </ul>
bit 3	SSP1IF: Master Synchronous Serial Port Interrupt Flag bit
	<ul> <li>1 = The transmission/reception is complete (must be cleared in software)</li> <li>0 = Waiting to transmit/receive</li> </ul>
bit 2	TMR1GIF: Timer1 Gate Interrupt Flag bit
	<ul><li>1 = Timer gate interrupt occurred (must be cleared in software)</li><li>0 = No timer gate interrupt occurred</li></ul>
bit 1	TMR2IF: TMR2 to PR2 Match Interrupt Flag bit
	<ul><li>1 = TMR2 to PR2 match occurred (must be cleared in software)</li><li>0 = No TMR2 to PR2 match occurred</li></ul>
bit 0	TMR1IF: TMR1 Overflow Interrupt Flag bit
	1 = TMR1 register overflowed (must be cleared in software)
	0 = TMR1 register did not overflow

#### REGISTER 11-5: PIR2: PERIPHERAL INTERRUPT REQUEST (FLAG) REGISTER 2

R/W-0	U-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0
OSCFIF	—	SSP2IF	BCL2IF	BCL1IF	HLVDIF	TMR3IF	TMR3GIF
bit 7							bit 0

Legend:			
R = Readable bit	W = Writable bit	U = Unimplemented bit, rea	d as '0'
-n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown

bit 7	OSCFIF: Oscillator Fail Interrupt Flag bit
	<ul> <li>1 = Device oscillator failed, clock input has changed to INTOSC (must be cleared in software)</li> <li>0 = Device clock is operating</li> </ul>
bit 6	Unimplemented: Read as '0'
bit 5	SSP2IF: Master Synchronous Serial Port Interrupt Flag bit
	<ul><li>1 = The transmission/reception has been completed (must be cleared in software)</li><li>0 = Waiting to transmit/receive</li></ul>
bit 4	BCL2IF: Bus Collision Interrupt Flag bit
	<ul><li>1 = A bus collision occurred (must be cleared in software)</li><li>0 = No bus collision occurred</li></ul>
bit 3	BCL1IF: Bus Collision Interrupt Flag bit
	<ul><li>1 = A bus collision occurred (must be cleared in software)</li><li>0 = No bus collision occurred</li></ul>
bit 2	HLVDIF: High/Low-Voltage Detect Interrupt Flag bit
	<ul> <li>1 = A low-voltage condition occurred (must be cleared in software)</li> <li>0 = The device voltage is above the regulator's low-voltage trip point</li> </ul>
bit 1	TMR3IF: TMR3 Overflow Interrupt Flag bit
	<ul><li>1 = TMR3 register overflowed (must be cleared in software)</li><li>0 = TMR3 register did not overflow</li></ul>
bit 0	TMR3GIF: TMR3 Gate Interrupt Flag bit
	<ul> <li>1 = Timer gate interrupt occurred (must be cleared in software)</li> <li>0 = No timer gate interrupt occurred</li> </ul>

R/W-0	U-0	R-0	R-0	R/W-0	R/W-0	R/W-0	R/W-0			
TMR5GIF		RC2IF	TX2IF	CTMUIF	CCP2IF	CCP1IF	RTCCIF			
bit 7		·					bit			
Legend:										
R = Readabl	le bit	W = Writable	bit	U = Unimplem	nented bit, read	d as '0'				
-n = Value at	t POR	'1' = Bit is set		'0' = Bit is clea	ared	x = Bit is unkr	nown			
bit 7	TMP5CIE	Timer5 Gate Inte	rrupt Elag bit							
		ate interrupt occ		e cleared in soft	ware)					
	-	er gate interrupt of								
bit 6	Unimpleme	ented: Read as '	0'							
bit 5	RC2IF: EUS	SART Receive In	terrupt Flag bi	t						
		ISART receive b		2, is full (cleared	when RCREG	G2 is read)				
L:1 4		ISART receive b								
bit 4	<b>TX2IF:</b> EUSART Transmit Interrupt Flag bit 1 = The EUSART transmit buffer, TXREG2, is empty (cleared when TXREG2 is written)									
		ISART transmit b					)			
bit 3	CTMUIF: CTMU Interrupt Flag bit									
	<ul> <li>1 = CTMU interrupt occurred (must be cleared in software)</li> <li>0 = No CTMU interrupt occurred</li> </ul>									
bit 2	CCP2IF: ECCP2 Interrupt Flag bit									
	Capture mo	-								
	1 = A TMR	register capture		st be cleared in	software)					
		R register captur	e occurred							
	$\frac{\text{Compare } m}{1 = \Delta TMR}$	<u>ode:</u> register compar	e match occur	red (must be cle	ared in softwa	re)				
		R register compare								
	PWM mode	-								
	Unused in th									
bit 1	CCP1IF: ECCP1 Interrupt Flag bit									
	<u>Capture mode:</u> 1 = A TMR register capture occurred (must be cleared in software)									
	1 = A  find register capture occurred (must be cleared in software) 0 =  No TMR register capture occurred									
	<u>Compare mode:</u> 1 = A TMR register compare match occurred (must be cleared in software)									
		register compar R register compa		•	eared in softwa	re)				
	<u>PWM mode</u>	• ·		inco						
	Unused in the									
bit 0	RTCCIF: RT	FCC Interrupt Fla	ag bit							
		nterrupt occurre		ared in software	e)					
	0 = No RTC	CC interrupt occu	irred							

#### REGISTER 11-7: PIR4: PERIPHERAL INTERRUPT REQUEST (FLAG) REGISTER 4

R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0
CCP10IF <sup>(1)</sup>	CCP9IF <sup>(1)</sup>	CCP8IF	CCP7IF	CCP6IF	CCP5IF	CCP4IF	CCP3IF
bit 7							bit 0
Legend:							
R = Readable I		W = Writable		-	mented bit, read	d as '0'	
-n = Value at P	OR	'1' = Bit is set		'0' = Bit is cle	ared	x = Bit is unk	nown
bit 0	0 = No TMR i <u>Compare Mod</u> 1 = A TMR re 0 = No TMR i <u>PWM Mode</u> : Not used in P <sup>V</sup> <b>CCP3IF</b> : ECC <u>Capture Mode</u> 1 = A TMR re <u>Compare Mod</u> 1 = A TMR re	2: egister capture register captur egister compar- register compar- register compa WM mode. CP3 Interrupt F 2: egister capture register captur fegister captur	occurred (mu e occurred e match occur ire match occu lag bit occurred (mu e occurred e match occur	ist be cleared in rred (must be c urred ist be cleared in rred (must be c	cleared in softw		

Note 1: Unimplemented on devices with a program memory of 32 Kbytes (PIC18FX5K22).

R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0		
TMR7GIF <sup>(1)</sup>	TMR12IF <sup>(1)</sup>	TMR10IF <sup>(1)</sup>	TMR8IF	TMR7IF <sup>(1)</sup>	TMR6IF	TMR5IF	TMR4IF		
bit 7							bit (		
Legend:									
R = Readable		W = Writable b	it	U = Unimplem					
-n = Value at F	OR	'1' = Bit is set		'0' = Bit is clea	ared	x = Bit is unkr	nown		
bit 7	TMR7GIF: TM	IR7 Gate Interro	upt Flag bits	1)					
	1 = TMR gate		red (bit mus	t be cleared in s	oftware)				
bit 6	TMR12IF: TM	R12 to PR12 M	atch Interrup	ot Flag bit <sup>(1)</sup>					
		PR12 match o 2 to PR12 matc	•	st be cleared in	software)				
bit 5	TMR10IF: TM	R10 to PR10 M	atch Interrup	ot Flag bit <sup>(1)</sup>					
		PR10 match o 0 to PR10 matc	· ·	st be cleared in	software)				
bit 4	TMR8IF: TMR	R8 to PR8 Match	n Interrupt Fl	ag bit					
		PR8 match occ to PR8 match	``	be cleared in so	ftware)				
bit 3	TMR7IF: TMR	R7 Overflow Inte	rrupt Flag bi	t <sup>(1)</sup>					
			•	leared in softwa	ire)				
	-	gister did not ov							
bit 2		R6 to PR6 Match	•	•					
		PR6 match occ to PR6 match	``	be cleared in so	ftware)				
bit 1	TMR5IF: TMR	MR5IF: TMR5 Overflow Interrupt Flag bit							
		gister overflowe gister did not ov		leared in softwa	ire)				
bit 0	TMR4IF: TMR	R4 to PR4 Match	n Interrupt Fl	ag bit					
	1 = TMR4 to		urred (must	be cleared in so	ftware)				
Note 1 Uni	mplemented on	devices with a	nrogram me	mory of 32 Kby		5422)			

#### REGISTER 11-8: PIR5: PERIPHERAL INTERRUPT REQUEST (FLAG) REGISTER 5

**Note 1:** Unimplemented on devices with a program memory of 32 Kbytes (PIC18FX5K22).

#### REGISTER 11-9: PIR6: PERIPHERAL INTERRUPT REQUEST (FLAG) REGISTER 6

U-0	U-0	U-0	R/W-0	U-0	R/W-0	R/W-0	R/W-0
—	—	—	EEIF	—	CMP3IF	CMP2IF	CMP1IF
bit 7							bit 0

Legend:			
R = Readable bit	W = Writable bit	U = Unimplemented bit,	, read as '0'
-n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown

bit 7-5	Unimplemented: Read as '0'
bit 4	EEIF: Data EEDATA/Flash Write Operation Interrupt Flag bit
	<ul> <li>1 = The write operation is complete (must be cleared in software)</li> <li>0 = The write operation is not complete or has not been started</li> </ul>
bit 3	Unimplemented: Read as '0'
bit 2	CMP3IF: CMP3 Interrupt Flag bit
	<ul> <li>1 = CMP3 interrupt occurred (must be cleared in software)</li> <li>0 = No CMP3 interrupt occurred</li> </ul>
bit 1	CMP2IF: CMP2 Interrupt Flag bit
	<ul> <li>1 = CMP2 interrupt occurred (must be cleared in software)</li> <li>0 = No CMP2 interrupt occurred</li> </ul>
bit 0	CMP1IF: CM1 Interrupt Flag bit
	<ul> <li>1 = CMP1 interrupt occurred (must be cleared in software)</li> <li>0 = No CMP1 interrupt occurred</li> </ul>

#### 11.3 PIE Registers

The PIE registers contain the individual enable bits for the peripheral interrupts. Due to the number of peripheral interrupt sources, there are six Peripheral Interrupt Enable registers (PIE1 through PIE6). When IPEN (RCON<7>) = 0, the PEIE bit must be set to enable any of these peripheral interrupts.

#### REGISTER 11-10: PIE1: PERIPHERAL INTERRUPT ENABLE REGISTER 1

R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0		
PSPIE	ADIE	RC1IE	TX1IE	SSP1IE	TMR1GIE	TMR2IE	TMR1IE		
bit 7		RONE	IXIL	001 IIL	INITIOL		bit C		
Legend:									
R = Reada	ble bit	W = Writable	bit	•	mented bit, read	1 as '0'			
-n = Value	at POR	'1' = Bit is set		'0' = Bit is cle	ared	x = Bit is unkr	nown		
bit 7	PSPIE: Para	allel Slave Port F	ead/Write Inte	errupt Enable b	it				
		the PSP read/w the PSP read/w							
bit 6	ADIE: A/D C	converter Interru	pt Enable bit						
		the A/D interrup							
		the A/D interrup							
bit 5		ART Receive In	•						
		the EUSART re the EUSART re							
bit 4		ART Transmit Interrupt Enable bit							
		the EUSART tra							
bit 3	SSP1IE: Ma	ster Synchronou	s Serial Port	nterrupt Enable	e bit				
		the MSSP interr	•						
		the MSSP inter	•						
bit 2		MR1 Gate Inter	rupt Enable bi	t					
	1 = Enables 0 = Disabled								
bit 1		MR2 to PR2 Match Interrupt Enable bit							
		the TMR2 to PF	•						
		the TMR2 to PF							
bit 0	TMR1IE: TM	IR1 Overflow Int	errupt Enable	bit					
		the TMR1 overf							
	0 = Disables	the TMR1 over	low interrupt						

#### REGISTER 11-11: PIE2: PERIPHERAL INTERRUPT ENABLE REGISTER 2

R/W-0	U-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0
OSCFIE	_	SSP2IE	BCL2IE	BCL1IE	HLVDIE	TMR3IE	TMR3GIE
bit 7							bit 0
Legend:							
R = Readable	e bit	W = Writable I	bit	U = Unimplem	ented bit, read	d as '0'	
-n = Value at	POR	'1' = Bit is set		'0' = Bit is clea	ared	x = Bit is unkr	nown
bit 7	<b>OSCFIE:</b> Os 1 = Enableo 0 = Disable		rupt Enable bi	t			
bit 6	Unimpleme	nted: Read as '0	)'				
bit 5	SSP2IE: Ma	ster Synchronou	s Serial Port 2	Interrupt Enab	le bit		
		s the MSSP inter s the MSSP inter					
bit 4	1 = Enables	s Collision Interr s the bus collision s the bus collisio	n interrupt				
bit 3	BCL1IE: Bu 1 = Enablec 0 = Disable	-	upt Enable bit				
bit 2	HLVDIE: Hig 1 = Enablec 0 = Disable	-	Detect Interrup	t Enable bit			
bit 1	TMR3IE: TM 1 = Enablec 0 = Disable		errupt Enable	bit			
bit 0	<b>TMR3GIE:</b> 1 1 = Enabled 0 = Disable	-	rupt Enable b	it			

R/W-0	U-0	R-0	R-0	R/W-0	R/W-0	R/W-0	R/W-0
TMR5GIE	_	RC2IE	TX2IE	CTMUIE	CCP2IE	CCP1IE	RTCCIE
bit 7							bit 0
Legend:							
R = Readable	e bit	W = Writable	bit	U = Unimplem		d as '0'	
-n = Value at	POR	'1' = Bit is set		'0' = Bit is clea	ared	x = Bit is unkr	nown
bit 7	TMR5GIE: Tir	ner5 Gate Inter	rupt Enable bit	t			
	1 = Enabled 0 = Disabled						
bit 6	Unimplemen	ted: Read as '	0'				
bit 5	RC2IE: EUSA	ART Receive In	iterrupt Enable	e bit			
	1 = Enabled						
	0 = Disabled						
bit 4		RT Transmit Ir	terrupt Enable	e bit			
	1 = Enabled 0 = Disabled						
bit 3		MU Interrupt Er	nahle hit				
	1 = Enabled						
	0 = Disabled						
bit 2	CCP2IE: ECO	CP2 Interrupt E	nable bit				
	1 = Enabled						
	0 = Disabled						
bit 1		CP1 Interrupt E	nable bit				
	1 = Enabled 0 = Disabled						
bit 0	RTCCIE: RTC	CC Interrupt En	able bit				
	1 = Enabled 0 = Disabled						

#### REGISTER 11-12: PIE3: PERIPHERAL INTERRUPT ENABLE REGISTER 3

#### REGISTER 11-13: PIE4: PERIPHERAL INTERRUPT ENABLE REGISTER 4

R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0			
CCP10IE <sup>(1)</sup>	CCP9IE <sup>(1)</sup>	CCP8IE	CCP7IE	CCP6IE	CCP5IE	CCP4IE	CCP3IE			
bit 7	bit 7 bit 0									
Legend:										
R = Readable b	oit	W = Writable	bit	U = Unimplemented bit, read as '0'						
-n = Value at POR '1' = Bit is set			'0' = Bit is cleared x = Bit is unknown							

bit 7-0 CCP<10:3>IE: CCP<10:3> Interrupt Enable bits<sup>(1)</sup>

1 = Enabled

0 = Disabled

Note 1: Unimplemented on devices with a program memory of 32 Kbytes (PIC18FX5K22).

#### REGISTER 11-14: PIE5: PERIPHERAL INTERRUPT ENABLE REGISTER 5

R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0
TMR7GIE <sup>(1)</sup>	TMR12IE <sup>(1)</sup>	TMR10IE <sup>(1)</sup>	TMR8IE	TMR7IE <sup>(1)</sup>	TMR6IE	TMR5IE	TMR4IE
bit 7				1			bit 0
Legend:							
R = Readable	bit	W = Writable b	bit	U = Unimplem	nented bit, read	d as '0'	
-n = Value at F	POR	'1' = Bit is set		'0' = Bit is clea	ared	x = Bit is unkr	nown
bit 7		MR7 Gate Interr	upt Enable b	it <sup>(1)</sup>			
	1 = Enabled 0 = Disabled						
bit 6		/IR12 to PR12 N	Aatch Interru	ot Enable bit(1)			
bit o		the TMR12 to F					
		the TMR12 to F					
bit 5	TMR10IE: TM	/IR10 to PR10 N	atch Interrup	ot Enable bit <sup>(1)</sup>			
		the TMR10 to F					
		the TMR10 to F		•			
bit 4		R8 to PR8 Matc	•				
		the TMR8 to PF the TMR8 to PI					
bit 3		R7 Overflow Inte					
		the TMR7 overf	•				
	0 = Disables	the TMR7 over	flow interrupt	:			
bit 2	TMR6IE: TM	R6 to PR6 Matc	h Interrupt E	nable bit			
		the TMR6 to PF					
		the TMR6 to P		•			
bit 1		R5 Overflow Internation	•				
		the TMR5 overf the TMR5 over					
bit 0		R4 to PR4 Matc					
		the TMR4 to PF					
	0 = Disables			•			

Note 1: Unimplemented on devices with a program memory of 32 Kbytes (PIC18FX5K22).

U-0	U-0	U-0	R/W-0	U-0	R/W-0	R/W-0	R/W-0
_	—	—	EEIE	—	CMP3IE	CMP2IE	CMP1IE
bit 7	·			·		•	bit 0
Legend:							
R = Readal	ole bit	W = Writable	bit	U = Unimpler	mented bit, read	as '0'	
-n = Value a	at POR	'1' = Bit is set		'0' = Bit is cle	ared	x = Bit is unkr	nown
bit 7-5	Unimplemen	ted: Read as '	0'				
bit 4	EEIE: Data E	EDATA/Flash V	Vrite Operatio	n Enable bit			
	1 = Interrupt	is enabled					
	0 = interrupt	is disabled					
bit 3	Unimplemen	ted: Read as '	0'				
bit 2	CMP3IE: CM	P3 Enable bit					
	1 = Interrupt						
	0 = interrupt	is disabled					
bit 1	CMP2E: CMF	P2 Enable bit					
	1 = Interrupt						
	0 = interrupt						
bit 0	CMP1IE: CM						
	1 = Interrupt	is enabled					

#### REGISTER 11-15: PIE6: PERIPHERAL INTERRUPT ENABLE REGISTER 6

Interrupt is enabled

0 = interrupt is disabled

#### 11.4 IPR Registers

The IPR registers contain the individual priority bits for the peripheral interrupts. Due to the number of peripheral interrupt sources, there are six Peripheral Interrupt Priority registers (IPR1 through IPR6). Using the priority bits requires that the Interrupt Priority Enable (IPEN) bit (RCON<7>) be set.

#### REGISTER 11-16: IPR1: PERIPHERAL INTERRUPT PRIORITY REGISTER 1

R/W-1	R/W-1	R/W-1	R/W-1	R/W-1	R/W-1	R/W-1	R/W-1
PSPIP	ADIP	RC1IP	TX1IP	SSP1IP	TMR1GIP	TMR2IP	TMR1IP
bit 7							bit 0

Legend:			
R = Readable bit	W = Writable bit	U = Unimplemented bit, read	d as '0'
-n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown

bit 7	PSPIP: Parallel Slave Port Read/Write Interrupt Priority bit
	1 = High priority
	0 = Low priority
bit 6	ADIP: A/D Converter Interrupt Priority bit
	1 = High priority
	0 = Low priority
bit 5	RC1IP: EUSART Receive Interrupt Priority bit
	1 = High priority
	0 = Low priority
bit 4	<b>TX1IP:</b> EUSART Transmit Interrupt Priority bit
	1 = High priority
	0 = Low priority
bit 3	SSP1IP: Master Synchronous Serial Port Interrupt Priority bit
	1 = High priority
	1 = High priority 0 = Low priority
bit 2	
bit 2	<ul> <li>0 = Low priority</li> <li>TMR1GIP: Timer1 Gate Interrupt Priority bit</li> <li>1 = High priority</li> </ul>
bit 2	<ul><li>0 = Low priority</li><li>TMR1GIP: Timer1 Gate Interrupt Priority bit</li></ul>
bit 2 bit 1	<ul> <li>0 = Low priority</li> <li>TMR1GIP: Timer1 Gate Interrupt Priority bit</li> <li>1 = High priority</li> </ul>
	<ul> <li>0 = Low priority</li> <li>TMR1GIP: Timer1 Gate Interrupt Priority bit</li> <li>1 = High priority</li> <li>0 = Low priority</li> <li>TMR2IP: TMR2 to PR2 Match Interrupt Priority bit</li> <li>1 = High priority</li> </ul>
bit 1	<ul> <li>0 = Low priority</li> <li>TMR1GIP: Timer1 Gate Interrupt Priority bit</li> <li>1 = High priority</li> <li>0 = Low priority</li> <li>TMR2IP: TMR2 to PR2 Match Interrupt Priority bit</li> <li>1 = High priority</li> <li>0 = Low priority</li> </ul>
	<ul> <li>0 = Low priority</li> <li>TMR1GIP: Timer1 Gate Interrupt Priority bit</li> <li>1 = High priority</li> <li>0 = Low priority</li> <li>TMR2IP: TMR2 to PR2 Match Interrupt Priority bit</li> <li>1 = High priority</li> <li>0 = Low priority</li> <li>TMR1IP: TMR1 Overflow Interrupt Priority bit</li> </ul>
bit 1	<ul> <li>0 = Low priority</li> <li>TMR1GIP: Timer1 Gate Interrupt Priority bit</li> <li>1 = High priority</li> <li>0 = Low priority</li> <li>TMR2IP: TMR2 to PR2 Match Interrupt Priority bit</li> <li>1 = High priority</li> <li>0 = Low priority</li> </ul>

R/W-1	U-0	R/W-1	R/W-1	R/W-1	R/W-1	R/W-1	R/W-1
OSCFIP		SSP2IP	BCL2IP	BCL1IP	HLVDIP	TMR3IP	TMR3GIP
bit 7		·		•			bit 0
Legend:							
R = Readab	le bit	W = Writable	bit	•	nented bit, read	d as '0'	
-n = Value a	It POR	'1' = Bit is set		'0' = Bit is clea	ared	x = Bit is unki	nown
bit 7		Scillator Fail Inter	rupt Priority bi	it			
	1 = High p 0 = Low pr	•					
bit 6	Unimplem	ented: Read as '	כי				
bit 5	SSP2IP: M	aster Synchronou	us Serial Port 2	2 Interrupt Prior	ity bit		
	1 = High p	•					
	0 = Low pr	-					
bit 4		us Collision Interr	upt priority bit	(MSSP)			
	1 = High p 0 = Low pr						
bit 3	BCL1IP: B	us Collision Interr	upt Priority bit				
	1 = High p	•					
	0 = Low pr	•					
bit 2		igh/Low-Voltage I	Detect Interrup	t Priority bit			
	1 = High p 0 = Low pr						
bit 1	•	MR3 Overflow Inf	errunt Priority	hit			
	1 = High p		enuptritionty	bit			
	0 = Low pr	•					
bit 0	·	TMR3 Gate Inter	rupt Priority bi	t			
	1 = High p						
	0 = Low pr	•					

#### REGISTER 11-17: IPR2: PERIPHERAL INTERRUPT PRIORITY REGISTER 2

#### REGISTER 11-18: IPR3: PERIPHERAL INTERRUPT PRIORITY REGISTER 3

R/W-1	U-0	R-1	R-1	R/W-1	R/W-1	R/W-1	R/W-1
TMR5GIP	—	RC2IP	TX2IP	CTMUIP	CCP2IP	CCP1IP	RTCCIP
bit 7							bit (
Legend:							
R = Readabl	e bit	W = Writable	bit	U = Unimplem	nented bit, rea	d as '0'	
-n = Value at	POR	'1' = Bit is set		'0' = Bit is clea	ared	x = Bit is unkr	nown
bit 7		mar Cata inte	www.et Daieaite/h	:.			
	1 = High prio	mer5 Gate inter	rupt Priority b	אונ			
	0 = Low prior						
bit 6	Unimplemen	ted: Read as '	)'				
bit 5	RC2IP: EUSA	ART Receive P	riority Flag bit				
	1 = High prio						
	0 = Low prior	•					
bit 4		RT Transmit In	terrupt Priority	y bit			
	1 = High prio 0 = Low prior						
bit 3	•	MU Interrupt Pr	iority hit				
	1 = High prio		ionty bit				
	0 = Low prior	•					
bit 2	CCP2IP: ECC	CP2 Interrupt P	riority bit				
	1 = High prio						
	0 = Low prior	•					
bit 1		CP1 Interrupt P	riority bit				
	1 = High prio 0 = Low prior						
bit 0	•	CC Interrupt Pri	oritv bit				
	1 = High prio						
	0 = Low prior	•					

#### **REGISTER 11-19: IPR4: PERIPHERAL INTERRUPT PRIORITY REGISTER 4**

R/W-1	R/W-1	R/W-1	R/W-1	R/W-1	R/W-1	R/W-1	R/W-1
CCP10IP <sup>(1)</sup>	CCP9IP <sup>(1)</sup>	CCP8IP	CCP7IP	CCP6IP	CCP5IP	CCP4IP	CCP3IP
bit 7							bit 0

Legend:					
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'			
-n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown		

bit 7-0 CCP<10:3>IP: CCP<10:3> Interrupt Priority bits<sup>(1)</sup>

1 = High priority

0 = Low priority

Note 1: Unimplemented on devices with a program memory of 32 Kbytes (PIC18FX5K22).

R/W-1	R/W-1	R/W-1	R/W-1	R/W-1	R/W-1	R/W-1	R/W-1					
TMR7GIP <sup>(1)</sup>	TMR12IP <sup>(1)</sup>	TMR10IP <sup>(1)</sup>	TMR8IP	TMR7IP <sup>(1)</sup>	TMR6IP	TMR5IP	TMR4IP					
bit 7	•	•					bit					
Legend:												
R = Readable	bit	W = Writable b	oit	U = Unimplem	nented bit, read	d as '0'						
-n = Value at F	POR	'1' = Bit is set		'0' = Bit is clea	ared	x = Bit is unkr	nown					
bit 7	TMR7GIP: TN	/IR7 Gate Interr	upt Priority b	it <sup>(1)</sup>								
	1 = High priority											
	0 = Low prior	•										
bit 6	<b>TMR12IP:</b> TMR12 to PR12 Match Interrupt Priority bit <sup>(1)</sup>											
	1 = High priority 0 = Low priority											
bit 5	<b>TMR10IP:</b> TMR10 to PR10 Match Interrupt Priority bit <sup>(1)</sup>											
	1 = High priority											
	0 = Low priority											
bit 4	TMR8IP: TMR8 to PR8 Match Interrupt Priority bit											
	1 = High priority											
1.11.0	0 = Low prior	•		L ::( <b>1</b> )								
bit 3	<b>TMR7IP:</b> TMR7 Overflow Interrupt Priority bit <sup>(1)</sup>											
	1 = High priority 0 = Low priority											
bit 2	•	•	h Interrupt Pi	rioritv bit								
	<b>TMR6IP:</b> TMR6 to PR6 Match Interrupt Priority bit 1 = High priority											
	0 = Low prior	ity										
bit 1	TMR5IP: TMR5 Overflow Interrupt Priority bit											
	1 = High prio											
	0 = Low prior	2										
bit 0		R4 to PR4 Matc	h Interrupt Pi	riority bit								
	1 = High prio 0 = Low prior											
		ity										

#### REGISTER 11-20: IPR5: PERIPHERAL INTERRUPT PRIORITY REGISTER 5

Note 1: Unimplemented on devices with a program memory of 32 Kbytes (PIC18FX5K22).

#### REGISTER 11-21: IPR6: PERIPHERAL INTERRUPT PRIORITY REGISTER 6

U-0	U-0	U-0	R/W-1	U-0	R/W-1	R/W-1	R/W-1					
_	_		EEIP		CMP3IP	CMP2IP	CMP1IP					
bit 7			•	-			bit 0					
Legend:												
R = Readab	le bit	W = Writable	bit	U = Unimpler	mented bit, read	1 as '0'						
-n = Value a	It POR	'1' = Bit is set		'0' = Bit is cle		x = Bit is unkr	nown					
bit 7-5	Unimplem	ented: Read as '	0'									
bit 4	EEIP: EE li	EEIP: EE Interrupt Priority bit										
	1 = High p	1 = High priority										
	0 = Low pr	riority										
bit 3	Unimplem	ented: Read as '	0'									
bit 2	CMP3IP: C	CMP3IP: CMP3 Interrupt Priority bit										
	1 = High p	1 = High priority										
	0 = Low priority											
bit 1	CMP2IP: C	CMP2IP: CMP2 Interrupt Priority bit										
		1 = High priority										
	0 = Low pr	riority										
bit 0		MP1 Interrupt Pr	iority bit									
	1 = High p											
	0 = Low pr	riority										

#### 11.5 RCON Register

**—** 

The RCON register contains the bits used to determine the cause of the last Reset, or wake-up from Idle or Sleep modes. RCON also contains the bit that enables interrupt priorities (IPEN).

#### REGISTER 11-22: RCON: RESET CONTROL REGISTER

R/W-0	R/W-1	R/W-1	R/W-1	R-1	R-1	R/W-0	R/W-0
IPEN	SBOREN	CM	RI	TO	PD	POR	BOR
bit 7							bit 0

Legend:			
R = Readable bit	W = Writable bit	, read as '0'	
-n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown

bit 7	IPEN: Interrupt Priority Enable bit
	1 = Enable priority levels on interrupts
	0 = Disable priority levels on interrupts (PIC16CXXX Compatibility mode)
bit 6	SBOREN: Software BOR Enable bit
	For details of bit operation, see Register 5-1.
bit 5	CM: Configuration Mismatch Flag bit
	<ul> <li>1 = A Configuration Mismatch Reset has not occurred</li> <li>0 = A Configuration Mismatch Reset has occurred (must be subsequently set in software)</li> </ul>
bit 4	RI: RESET Instruction Flag bit
	For details of bit operation, see Register 5-1.
bit 3	TO: Watchdog Timer Time-out Flag bit
	For details of bit operation, see Register 5-1.
bit 2	PD: Power-Down Detection Flag bit
	For details of bit operation, see Register 5-1.
bit 1	POR: Power-on Reset Status bit
	For details of bit operation, see Register 5-1.
bit 0	BOR: Brown-out Reset Status bit
	For details of bit operation, see Register 5-1.

#### 11.6 INTx Pin Interrupts

External interrupts on the RB0/INT0, RB1/INT1, RB2/INT2 and RB3/INT3 pins are edge-triggered. If the corresponding INTEDGx bit in the INTCON2 register is set (= 1), the interrupt is triggered by a rising edge. If that bit is clear, the trigger is on the falling edge.

When a valid edge appears on the RBx/INTx pin, the corresponding flag bit, INTxIF, is set. This interrupt can be disabled by clearing the corresponding enable bit, INTxIE. Before re-enabling the interrupt, the flag bit (INTxIF) must be cleared in software in the Interrupt Service Routine.

All external interrupts (INT0, INT1, INT2 and INT3) can wake up the processor from the power-managed modes if bit, INTxIE, was set prior to going into the power-managed modes. If the Global Interrupt Enable bit (GIE) is set, the processor will branch to the interrupt vector following wake-up.

The interrupt priority for INT1, INT2 and INT3 is determined by the value contained in the Interrupt Priority bits, INT1IP (INTCON3<6>), INT2IP (INTCON3<7>) and INT3IP (INTCON2<1>).

There is no priority bit associated with INTO. It is always a high-priority interrupt source.

#### 11.7 TMR0 Interrupt

In 8-bit mode (the default), an overflow in the TMR0 register (FFh  $\rightarrow$  00h) will set flag bit, TMR0IF. In 16-bit mode, an overflow in the TMR0H:TMR0L register pair (FFFFh  $\rightarrow$  0000h) will set TMR0IF.

The interrupt can be enabled/disabled by setting/clearing enable bit, TMR0IE (INTCON<5>). Interrupt priority for Timer0 is determined by the value contained in the interrupt priority bit, TMR0IP (INTCON2<2>). For further details on the Timer0 module, see **Section 13.0 "Timer0 Module"**.

#### 11.8 PORTB Interrupt-on-Change

An input change on PORTB<7:4> sets flag bit, RBIF (INTCON<0>). The interrupt can be enabled/disabled by setting/clearing enable bit, RBIE (INTCON<3>).

Interrupt priority for PORTB interrupt-on-change is determined by the value contained in the interrupt priority bit, RBIP (INTCON2<0>).

#### 11.9 Context Saving During Interrupts

During interrupts, the return PC address is saved on the stack. Additionally, the WREG, STATUS and BSR registers are saved on the Fast Return Stack.

If a fast return from interrupt is not used (see **Section 6.3 "Data Memory Organization**"), the user may need to save the WREG, STATUS and BSR registers on entry to the Interrupt Service Routine (ISR). Depending on the user's application, other registers may also need to be saved.

Example 11-1 saves and restores the WREG, STATUS and BSR registers during an Interrupt Service Routine.

MOVWF MOVFF MOVFF ;	W_TEMP STATUS, STATUS_TEMP BSR, BSR_TEMP	; W_TEMP is in virtual bank ; STATUS_TEMP located anywhere ; BSR_TMEP located anywhere
; USER	ISR CODE	
;		
MOVFF	BSR_TEMP, BSR	; Restore BSR
MOVF	W_TEMP, W	; Restore WREG
MOVFF	STATUS_TEMP, STATUS	; Restore STATUS

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
INTCON	GIE/GIEH	PEIE/GIEL	TMR0IE	INT0IE	RBIE	TMR0IF	INT0IF	RBIF
INTCON2	RBPU	INTEDG0	INTEDG1	INTEDG2	INTEDG3	TMR0IP	INT3IP	RBIP
INTCON3	INT2IP	INT1IP	INT3IE	INT2IE	INT1IE	INT3IF	INT2IF	INT1IF
PIR1	PSPIP	ADIF	RC1IF	TX1IF	SSP1IF	TMR1GIF	TMR2IF	TMR1IF
PIR2	OSCFIF	_	SSP2IF	BCL2IF	BCL1IF	HLVDIF	TMR3IF	TMR3GIF
PIR3	TMR5GIF	_	RC2IF	TX2IF	CTMUIF	CCP2IF	CCP1IF	RTCCIF
PIR4	CCP10IF <sup>(1)</sup>	CCP9IF <sup>(1)</sup>	CCP8IF	CCP7IF	CCP6IF	CCP5IF	CCP4IF	CCP3IF
PIR5	TMR7GIF <sup>(1)</sup>	TMR12IF <sup>(1)</sup>	TMR10IF <sup>(1)</sup>	TMR8IF	TMR7IF <sup>(1)</sup>	TMR6IF	TMR5IF	TMR4IF
PIR6	—	_	_	EEIF	—	CMP3IF	CMP2IF	CMP1IF
PIE1	PSPIE	ADIE	RC1IE	TX1IE	SSP1IE	TMR1GIE	TMR2IE	TMR1IE
PIE2	OSCFIE	_	SSP2IE	BCL2IE	BCL1IE	HLVDIE	TMR3IE	TMR3GIE
PIE3	TMR5GIE		RC2IE	TX2IE	CTMUIE	CCP2IE	CCP1IE	RTCCIE
PIE4	CCP10IE <sup>(1)</sup>	CCP9IE <sup>(1)</sup>	CCP8IE	CCP7IE	CCP6IE	CCP5IE	CCP4IE	CCP3IE
PIE5	TMR7GIE <sup>(1)</sup>	TMR12IE <sup>(1)</sup>	TMR10IE <sup>(1)</sup>	TMR8IE	TMR7IE <sup>(1)</sup>	TMR6IE	TMR5IE	TMR4IE
PIE6	—		_	EEIE	—	CMP3IE	CMP2IE	CMP1IE
IPR1	PSPIP	ADIP	RC1IP	TX1IP	SSP1IP	TMR1GIP	TMR2IP	TMR1IP
IPR2	OSCFIP	_	SSP2IP	BCL2IP	BCL1IP	HLVDIP	TMR3IP	TMR3GIP
IPR3	TMR5GIP	_	RC2IP	TX2IP	CTMUIP	CCP2IP	CCP1IP	RTCCIP
IPR4	CCP10IP <sup>(1)</sup>	CCP9IP <sup>(1)</sup>	CCP8IP	CCP7IP	CCP6IP	CCP5IP	CCP4IP	CCP3IP
IPR5	TMR7GIP <sup>(1)</sup>	TMR12IP <sup>(1)</sup>	TMR10IP <sup>(1)</sup>	TMR8IP	TMR7IP <sup>(1)</sup>	TMR6IP	TMR5IP	TMR4IP
IPR6				EEIP		CMP3IP	CMP2IP	CMP1IP
RCON	IPEN	SBOREN	CM	RI	TO	PD	POR	BOR

#### TABLE 11-1: SUMMARY OF REGISTERS ASSOCIATED WITH INTERRUPTS

Legend: Shaded cells are not used by the interrupts.

Note 1: Unimplemented on devices with a program memory of 32 Kbytes (PIC18FX5K22).

### 12.0 I/O PORTS

Depending on the device selected and features enabled, there are up to nine ports available. Some pins of the I/O ports are multiplexed with an alternate function from the peripheral features on the device. In general, when a peripheral is enabled, that pin may not be used as a general purpose I/O pin.

Each port has three memory mapped registers for its operation:

- TRIS register (Data Direction register)
- PORT register (reads the levels on the pins of the device)
- LAT register (Output Latch register)

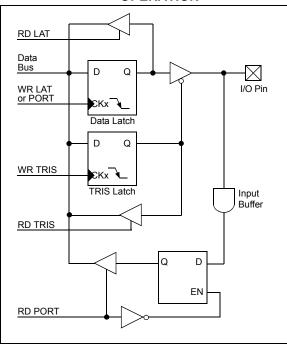
Reading the PORT register reads the current status of the pins, whereas writing to the PORT register writes to the Output Latch (LAT) register.

Setting a TRIS bit (= 1) makes the corresponding port pin an input (putting the corresponding output driver in a High-Impedance mode). Clearing a TRIS bit (= 0) makes the corresponding port pin an output (i.e., puts the contents of the corresponding LAT bit on the selected pin).

The Output Latch (LAT register) is useful for read-modify-write operations on the value that the I/O pins are driving. Read-modify-write operations on the LAT register read and write the latched output value for the PORT register.

A simplified model of a generic I/O port, without the interfaces to other peripherals, is shown in Figure 12-1.

FIGURE 12-1: GENERIC I/O PORT OPERATION



#### 12.1 I/O Port Pin Capabilities

When developing an application, the capabilities of the port pins must be considered. Outputs on some pins have higher output drive strength than others. Similarly, some pins can tolerate higher than VDD input levels.

All of the digital ports are 5.5V input tolerant. The analog ports have the same tolerance – having clamping diodes implemented internally.

#### 12.1.1 PIN OUTPUT DRIVE

When used as digital I/O, the output pin drive strengths vary, according to the pins' grouping, to meet the needs for a variety of applications. In general, there are two classes of output pins, in terms of drive capability:

- Outputs designed to drive higher current loads, such as LEDs:
  - PORTA PORTB
  - PORTC
- Outputs with lower drive levels, but capable of driving normal digital circuit loads with a high input impedance. Able to drive LEDs, but only those with smaller current requirements:
  - PORTD PORTE
  - PORTE PORTG
  - PORTH<sup>(†)</sup> PORTJ<sup>(†)</sup>
    - † These ports are not available on 64-pin devices.

#### 12.1.2 PULL-UP CONFIGURATION

Four of the I/O ports (PORTB, PORTD, PORTE and PORTJ) implement configurable weak pull-ups on all pins. These are internal pull-ups that allow floating digital input signals to be pulled to a consistent level without the use of external resistors.

The pull-ups are enabled with a single bit for each of the ports: RBPU (INTCON2<7>) for PORTB, and RDPU, REPU and RJPU (PADCFG1<7:5>) for the other ports.

#### REGISTER 12-1: PADCFG1: PAD CONFIGURATION REGISTER

R/W-0	R/W-0	R/W-0	U-0	U-0	R/W-0	R/W-0	U-0						
RDPU	REPU	RJPU <sup>(2)</sup>	_	_	RTSECSEL1 <sup>(1)</sup>	RTSECSEL0 <sup>(1)</sup>	_						
bit 7						· · · · · · · · · · · · · · · · · · ·	bit 0						
Legend:													
R = Readab	le bit	W = Writable	bit	U = Unimple	mented bit, read a	s '0'							
-n = Value a	t POR	'1' = Bit is set		'0' = Bit is cl	eared	x = Bit is unknown							
bit 7	RDPU: POR	RTD Pull-up Ena	able bit										
		1 = PORTD pull-up resistors are enabled by individual port latch values											
	0 = All POR	TD pull-up resis	tors are disa	abled									
bit 6	REPU: PORTE Pull-up Enable bit												
	<ul> <li>1 = PORTE pull-up resistors are enabled by individual port latch values</li> <li>0 = All PORTE pull-up resistors are disabled</li> </ul>												
bit 5		TJ Pull-up Ena											
	1 = PORTJ pull-up resistors are enabled by individual port latch values												
	0 = All PORTJ pull-up resistors are disabled												
bit 4-3	Unimpleme	nted: Read as	'0'										
bit 2-1	RTSECSEL	<1:0>: RTCC S	econds Cloc	k Output Sele	ect bits <sup>(1)</sup>								
	11 = Reserved; do not use												
		10 = RTCC source clock is selected for the RTCC pin (the pin can be LF-INTOSC or SOSC, depending on the RTCOSC (CONFIG3L<1>) bit setting)											
		seconds clock		•	bin								
		alarm pulse is											
bit 0	Unimpleme	nted: Read as	'0'										
	•												

- **Note 1:** To enable the actual RTCC output, the RTCOE (RTCCFG<2>) bit must be set.
  - 2: Unimplemented on 64-pin devices (PIC18F6XK22), read as '0'.

#### 12.1.3 OPEN-DRAIN OUTPUTS

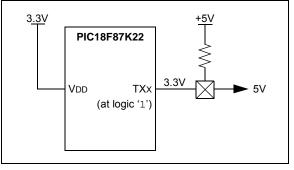
The output pins for several peripherals are also equipped with a configurable, open-drain output option. This allows the peripherals to communicate with external digital logic, operating at a higher voltage level, without the use of level translators.

The open-drain option is implemented on port pins specifically associated with the data and clock outputs of the USARTs, the MSSP module (in SPI mode) and the CCP modules. This option is selectively enabled by setting the open-drain control bits in the registers, ODCON1, ODCON2 and ODCON3.

When the open-drain option is required, the output pin must also be tied through an external pull-up resistor provided by the user to a higher voltage level, up to 5V (Figure 12-2). When a digital logic high signal is output, it is pulled up to the higher voltage level.

#### FIGURE 12-2: USING THE OPEN-DRAIN OUTPUT (USART SHOWN

#### AS EXAMPLE)



#### REGISTER 12-2: ODCON1: PERIPHERAL OPEN-DRAIN CONTROL REGISTER 1

R/W-0	R/W-0	R/W-0	U-0	U-0	U-0	U-0	R/W-0
SSP10D	CCP2OD	CCP10D	_	—	—	—	SSP2OD
bit 7							bit 0

Legend:			
R = Readable bit	W = Writable bit	U = Unimplemented bit	, read as '0'
-n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown

bit 7	SSP10D: MSSP1 Open-Drain Output Enable bit
	1 = Open-drain capability is enabled
	0 = Open-drain capability is disabled
bit 6	CCP2OD: ECCP2 Open-Drain Output Enable bit
	1 = Open-drain capability is enabled
	0 = Open-drain capability is disabled
bit 5	CCP10D: ECCP1 Open-Drain Output Enable bit
	1 = Open-drain capability is enabled
	0 = Open-drain capability is disabled
bit 4-1	Unimplemented: Read as '0'
bit 0	SSP2OD: MSSP2 Open-Drain Output Enable bit
	1 = Open-drain capability is enabled
	0 = Open-drain canability is disabled

0 = Open-drain capability is disabled

R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0			
CCP100D <sup>(1)</sup>	CCP9OD <sup>(1)</sup>	CCP8OD	CCP7OD	CCP6OD	CCP5OD	CCP4OD	CCP3OD			
bit 7							bit			
Legend:										
R = Readable	bit	W = Writable	bit	•	nented bit, rea	d as '0'				
-n = Value at P	OR	'1' = Bit is set		'0' = Bit is cle	ared	x = Bit is unknown				
L:1 7	0004000-00									
bit 7	<b>CCP10OD:</b> CCP10 Open-Drain Output Enable bit <sup>(1)</sup> 1 = Open-drain capability is enabled									
	0 = Open-drai									
bit 6	CCP9OD: CC	P9 Open-Drai	n Output Enat	ole bit <sup>(1)</sup>						
	1 = Open-drain capability is enabled									
	0 = Open-drain capability is disabled									
bit 5	CCP8OD: CCP8 Open-Drain Output Enable bit									
	<ul> <li>1 = Open-drain capability is enabled</li> <li>0 = Open-drain capability is disabled</li> </ul>									
bit 4	CCP7OD: CCI			a hit						
DIL 4		•		Die Dit						
	<ul> <li>1 = Open-drain capability is enabled</li> <li>0 = Open-drain capability is disabled</li> </ul>									
bit 3	CCP6OD: CC	P6 Open-Drai	n Output Enat	ole bit						
	1 = Open-drai	n capability is	enabled							
	0 = Open-drai	in capability is	disabled							
bit 2	CCP5OD: CC	•	•	ole bit						
	<ol> <li>1 = Open-drain capability is enabled</li> <li>0 = Open-drain capability is disabled</li> </ol>									
<b>L</b> :4	•			la hit						
bit 1	<b>CCP4OD:</b> CCP4 Open-Drain Output Enable bit 1 = Open-drain capability is enabled									
	0 =  Open-drain capability is disabled									
bit 0	CCP3OD: EC			able bit						
	1 = Open-drai	•								
	0 = Open-drai									

#### REGISTER 12-3: ODCON2: PERIPHERAL OPEN-DRAIN CONTROL REGISTER 2

Note 1: Not implemented on devices with 32-byte program memory (PIC18FX5K22).

#### REGISTER 12-4: ODCON3: PERIPHERAL OPEN-DRAIN CONTROL REGISTER 3

		-	-							
R/W-0	R/W-0	U-0	U-0	U-0	U-0	U-0	R/W-0			
U2OD	U10D		_	_	—	—	CTMUDS			
bit 7							bit (			
Legend:										
R = Readable	bit	W = Writable	bit	U = Unimplei	mented bit, rea	d as '0'				
-n = Value at F	POR	'1' = Bit is set		'0' = Bit is cle	eared	x = Bit is unknown				
	1 = Open-dra	ART2 Open-Dra ain capability is ain capability is	enabled							
bit 6	U10D: EUSART1 Open-Drain Output Enable bit									
		ain capability is ain capability is								
bit 5-1	Unimplemen	ted: Read as '	0'							
bit 0	CTMUDS: CTMU Pulse Delay Enable bit									
		lay input for CT lay input for CT		•						

#### 12.1.4 ANALOG AND DIGITAL PORTS

Many of the ports multiplex analog and digital functionality, providing a lot of flexibility for hardware designers. PIC18F87K22 family devices can make any analog pin analog or digital, depending on an application's needs. The ports' analog/digital functionality is controlled by registers: ANCON0, ANCON1 and ANCON2. Setting these registers makes the corresponding pins analog and clearing the registers makes the ports digital. For details on these registers, see Section 23.0 "12-Bit Analog-to-Digital Converter (A/D) Module".

### 12.2 PORTA, TRISA and LATA Registers

PORTA is an 8-bit wide, bidirectional port. The corresponding Data Direction and Output Latch registers are TRISA and LATA.

RA4/T0CKI is a Schmitt Trigger input. All other PORTA pins have TTL input levels and full CMOS output drivers.

RA5 and RA<3:0> are multiplexed with analog inputs for the A/D Converter.

The operation of the analog inputs as A/D Converter inputs is selected by clearing or setting the ANSEL control bits in the ANCON1 register. The corresponding TRISA bits control the direction of these pins, even when they are being used as analog inputs. The user must ensure the bits in the TRISA register are maintained set when using them as analog inputs.

Note:	RA5 and RA<3:0> are configured as					
	analog inputs on any Reset and are read					
	as '0'. RA4 is configured as a digital inpu					

OSC2/CLKO/RA6 and OSC1/CLKI/RA7 normally serve as the external circuit connections for the external (primary) oscillator circuit (HS Oscillator modes), or the external clock input and output (EC Oscillator modes). In these cases, RA6 and RA7 are not available as digital I/O and their corresponding TRIS and LAT bits are read as '0'. When the device is configured to use HF-INTOSC, MF-INTOSC or LF-INTOSC as the default oscillator mode, RA6 and RA7 are automatically configured as digital I/O; the oscillator and clock in/clock out functions are disabled.

RA5 has additional functionality for Timer1 and Timer3. It can be configured as the Timer1 clock input or the Timer3 external clock gate input.

CLRF	PORTA	; Initialize PORTA by
		; clearing output latches
CLRF	LATA	; Alternate method to
		; clear output data latches
BANKSEL	ANCON1	; Select bank with ANCON1 register
MOVLW	00h	; Configure A/D
MOVWF	ANCON1	; for digital inputs
BANKSEL	TRISA	; Select bank with TRISA register
MOVLW	OBFh	; Value used to initialize
		; data direction
MOVWF	TRISA	; Set RA<7, 5:0> as inputs,
		; RA<6> as output

TADLE 12-1.	FURIAT			-	
Pin Name	Function	TRIS Setting	I/O	I/O Type	Description
RA0/AN0/ULPWU	RA0	0	0	DIG	LATA<0> data output; not affected by analog input.
		1	I	TTL	PORTA<0> data input; disabled when analog input is enabled.
	AN0	1	I	ANA	A/D Input Channel 0. Default input configuration on POR; does not affect digital output.
	ULPWU	1	Ι	ANA	Ultra Low-Power Wake-up input.
RA1/AN1	RA1	0	0	DIG	LATA<1> data output; not affected by analog input.
		1	I	TTL	PORTA<1> data input; disabled when analog input is enabled.
	AN1	1	I	ANA	A/D Input Channel 1. Default input configuration on POR; does not affect digital output.
RA2/AN2/VREF-	RA2	0	0	DIG	LATA<2> data output; not affected by analog input.
		1	I	TTL	PORTA<2> data input; disabled when analog functions are enabled.
	AN2	1	I	ANA	A/D Input Channel 2. Default input configuration on POR.
	VREF-	1	I	ANA	A/D and comparator low reference voltage input.
RA3/AN3/VREF+	RA3	0	0	DIG	LATA<3> data output; not affected by analog input.
		1	I	TTL	PORTA<3> data input; disabled when analog input is enabled.
	AN3	1	I	ANA	A/D Input Channel 3. Default input configuration on POR.
	VREF+	1	I	ANA	A/D and comparator high reference voltage input.
RA4/T0CKI	RA4	0	0	DIG	LATA<4> data output.
		1	I	ST	PORTA<4> data input. Default configuration on POR.
	TOCKI	x	I	ST	Timer0 clock input.
RA5/AN4/T1CKI/	RA5	0	0	DIG	LATA<5> data output; not affected by analog input.
T3G/HLVDIN		1	I	TTL	PORTA<5> data input; disabled when analog input is enabled.
	AN4	1	I	ANA	A/D Input Channel 4. Default configuration on POR.
	T1CKI	x	I	ST	Timer1 clock input.
	T3G	х	I	ST	Timer3 external clock gate input.
	HLVDIN	1	I	ANA	High/Low-Voltage Detect (HLVD) external trip point input.
OSC2/CLKO/RA6	OSC2	x	0	ANA	Main oscillator feedback output connection (HS, XT and LP modes).
	CLKO	х	0	DIG	System cycle clock output (Fosc/4, EC and INTOSC modes).
	RA6	0	0	DIG	LATA<6> data output; disabled when FOSC2 Configuration bit is set.
		1	I	TTL	PORTA<6> data input; disabled when FOSC2 Configuration bit is set.
OSC1/CLKI/RA7	OSC1	х	I	ANA	Main oscillator input connection (HS, XT and LP modes).
	CLKI	х	I	ANA	Main external clock source input (EC modes).
	RA7	0	0	DIG	LATA<7> data output; disabled when FOSC2 Configuration bit is set.
		1	Ι	TTL	PORTA<7> data input; disabled when FOSC2 Configuration bit is set.

TABLE 12-1: PORTA FUNCTIONS

**Legend:** O = Output, I = Input, ANA = Analog Signal, DIG = Digital Output, ST = Schmitt Trigger Buffer Input, TTL = TTL Buffer Input, x = Don't care (TRIS bit does not affect port direction or is overridden for this option).

<b>TABLE 12-2:</b>	SUMMARY OF REGISTERS ASSOCIATED WITH PORTA
--------------------	--

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
PORTA	RA7 <sup>(1)</sup>	RA6 <sup>(1)</sup>	RA5	RA4	RA3	RA2	RA1	RA0
LATA	LATA7 <sup>(1)</sup>	LATA6 <sup>(1)</sup>	LATA5	LATA4	LATA3	LATA2	LATA1	LATA0
TRISA	TRISA7 <sup>(1)</sup>	TRISA6 <sup>(1)</sup>	TRISA5	TRISA4	TRISA3	TRISA2	TRISA1	TRISA0
ANCON0	ANSEL7	ANSEL6	ANSEL5	ANSEL4	ANSEL3	ANSEL2	ANSEL1	ANSEL0

**Legend:** — = unimplemented, read as '0'. Shaded cells are not used by PORTA.

**Note 1:** These bits are enabled depending on the oscillator mode selected. When not enabled as PORTA pins, they are disabled and read as 'x'.

### 12.3 PORTB, TRISB and LATB Registers

PORTB is an eight-bit wide, bidirectional port. The corresponding Data Direction and Output Latch registers are TRISB and LATB. All pins on PORTB are digital only.

CLRF	PORTB	; Initialize PORTB by ; clearing output
CLRF	LATB	; data latches ; Alternate method
		; to clear output ; data latches
MOVLW	OCFh	; Value used to ; initialize data
		; direction
MOVWF	TRISB	; Set RB<3:0> as inputs
		; RB<5:4> as outputs ; RB<7:6> as inputs

Each of the PORTB pins has a weak internal pull-up. A single control bit can turn on all the pull-ups. This is performed by clearing bit, RBPU (INTCON2<7>). The weak pull-up is automatically turned off when the port pin is configured as an output. The pull-ups are disabled on a Power-on Reset.

Four of the PORTB pins (RB<7:4>) have an interrupt-on-change feature. Only pins configured as inputs can cause this interrupt to occur. Any RB<7:4> pin configured as an output will be excluded from the interrupt-on-change comparison.

Comparisons with the input pins (of RB<7:4>) are made with the old value latched on the last read of PORTB. The "mismatch" outputs of RB<7:4> are ORed together to generate the RB Port Change Interrupt with Flag bit, RBIF (INTCON<0>).

This interrupt can wake the device from power-managed modes. To clear the interrupt in the Interrupt Service Routine:

- a) Any read or write of PORTB (except with the MOVFF (ANY), PORTB instruction). This will end the mismatch condition.
- b) Wait one instruction cycle (such as executing a NOP instruction).
- c) Clear flag bit, RBIF.

A mismatch condition will continue to set flag bit, RBIF. Reading PORTB will end the mismatch condition and allow flag bit, RBIF, to be cleared after one Tcy delay.

The interrupt-on-change feature is recommended for wake-up on key depression operation and operations where PORTB is only used for the interrupt-on-change feature. Polling of PORTB is not recommended while using the interrupt-on-change feature.

The RB<3:2> pins are multiplexed as CTMU edge inputs. RB5 has an additional function for Timer3 and Timer1. It can be configured for Timer3 clock input or Timer1 external clock gate input.

Pin Name	Function	TRIS Setting	I/O	I/O Type	Description	
RB0/INT0/FLT0	RB0	0	0	DIG	LATB<0> data output.	
		1	I	TTL	PORTB<0> data input; weak pull-up when RBPU bit is cleared.	
	INT0	1	I	ST	External Interrupt 0 input.	
	FLT0	x	Ι	ST	Enhanced PWM Fault input for ECCPx.	
RB1/INT1	RB1	0	0	DIG	LATB<1> data output.	
		1	Ι	TTL	PORTB<1> data input; weak pull-up when RBPU bit is cleared.	
	INT1	1	Ι	ST	External Interrupt 1 input.	
RB2/INT2/CTED1	RB2	0	0	DIG	LATB<2> data output.	
		1	Ι	TTL	PORTB<2> data input; weak pull-up when RBPU bit is cleared.	
	INT2	1	Ι	ST	External Interrupt 2 input.	
	CTED1	x	Ι	ST	CTMU Edge 1 input.	

#### TABLE 12-3: PORTB FUNCTIONS

Legend: O = Output, I = Input, ANA = Analog Signal, DIG = Digital Output, ST = Schmitt Trigger Buffer Input,

TTL = TTL Buffer Input, x = Don't care (TRIS bit does not affect port direction or is overridden for this option).

Note 1: Alternate assignment for ECCP2 when the CCP2MX Configuration bit is cleared and in Extended Microcontroller mode.

Pin Name	Function	TRIS Setting	I/O	I/O Type	Description
RB3/INT3/CTED2/	RB3	0	0	DIG	LATB<3> data output.
ECCP2/P2A		1	I	TTL	PORTB<3> data input; weak pull-up when RBPU bit is cleared.
	INT3	1	I	ST	External Interrupt 3 input.
	CTED2	x	Ι	ST	CTMU Edge 2 input.
	ECCP2 <sup>(1)</sup>	0	0	DIG	ECCP2 compare output and ECCP2 PWM output. Takes priority over port data.
		1	Ι	ST	ECCP2 capture input.
	P2A	0	0	DIG	ECCP2 Enhanced PWM output, Channel A. May be configured for tri-state during Enhanced PWM shutdown events. Takes priority over port data.
RB4/KBI0	RB4	0	0	DIG	LATB<4> data output.
		1	I	TTL	PORTB<4> data input; weak pull-up when RBPU bit is cleared.
	KBI0	1	Ι	TTL	Interrupt-on-pin change.
RB5/KBI1/T3CKI/	RB5	0	0	DIG	LATB<5> data output.
T1G		1	I	TTL	PORTB<5> data input; weak pull-up when RBPU bit is cleared.
	KBI1	1	I	TTL	Interrupt-on-pin change.
	T3CKI	x	Ι	ST	Timer3 clock input.
	T1G	x	Ι	ST	Timer1 external clock gate input.
RB6/KBI2/PGC	RB6	0	0	DIG	LATB<6> data output.
		1	I	TTL	PORTB<6> data input; weak pull-up when $\overline{\text{RBPU}}$ bit is cleared.
	KBI2	1	I	TTL	Interrupt-on-pin change.
	PGC	x	-	ST	Serial execution (ICSP™) clock input for ICSP and ICD operations.
RB7/KBI3/PGD	RB7	0	0	DIG	LATB<7> data output.
		1	Ι	TTL	PORTB<7> data input; weak pull-up when RBPU bit is cleared.
	KBI3	1	I	TTL	Interrupt-on-pin change.
	PGD	х	0	DIG	Serial execution data output for ICSP and ICD operations.
		x	Ι	ST	Serial execution data input for ICSP and ICD operations.

TABLE 12-3: PORTB FUNCTIONS (CONTINUED)

Legend: O = Output, I = Input, ANA = Analog Signal, DIG = Digital Output, ST = Schmitt Trigger Buffer Input, TTL = TTL Buffer Input, x = Don't care (TRIS bit does not affect port direction or is overridden for this option).

Note 1: Alternate assignment for ECCP2 when the CCP2MX Configuration bit is cleared and in Extended Microcontroller mode.

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
PORTB	RB7	RB6	RB5	RB4	RB3	RB2	RB1	RB0
LATB	LATB7	LATB6	LATB5	LATB4	LATB3	LATB2	LATB1	LATB0
TRISB	TRISB7	TRISB6	TRISB5	TRISB4	TRISB3	TRISB2	TRISB1	TRISB0
INTCON	GIE/GIEH	PEIE/GIEL	TMR0IE	INT0IE	RBIE	TMR0IF	INT0IF	RBIF
INTCON2	RBPU	INTEDG0	INTEDG1	INTEDG2	INTEDG3	TMR0IP	INT3IP	RBIP
INTCON3	INT2IP	INT1IP	INT3IE	INT2IE	INT1IE	INT3IF	INT2IF	INT1IF
ODCON1	SSP10D	CCP2OD	CCP10D		_	_		SSP2OD

Legend: Shaded cells are not used by PORTB.

### 12.4 PORTC, TRISC and LATC Registers

PORTC is an eight-bit wide, bidirectional port. The corresponding Data Direction and Output Latch registers are TRISC and LATC. Only PORTC pins, RC2 through RC7, are digital only pins.

PORTC is multiplexed with ECCP, MSSP and EUSART peripheral functions (Table 12-5). The pins have Schmitt Trigger input buffers. The pins for ECCP, SPI and EUSART are also configurable for open-drain output whenever these functions are active. Open-drain configuration is selected by setting the SPIOD, CCPxOD and U1OD control bits in the registers, ODCON1 and ODCON3.

RC1 is normally configured as the default peripheral pin for the ECCP2 module. The assignment of ECCP2 is controlled by Configuration bit, CCP2MX (default state, CCP2MX = 1). When enabling peripheral functions, use care in defining TRIS bits for each PORTC pin. Some peripherals can override the TRIS bit to make a pin an output or input. Consult the corresponding peripheral section for the correct TRIS bit settings.

Note:	These pins are configured as digital inputs
	on any device Reset.

The contents of the TRISC register are affected by peripheral overrides. Reading TRISC always returns the current contents, even though a peripheral device may be overriding one or more of the pins.

EXAMPLE 12-3:	INITIALIZING PORTC

CLRF	PORTC	; Initialize PORTC by ; clearing output
		; data latches
CLRF	LATC	; Alternate method
CLRF	LAIC	, Alternate method
		; to clear output
		; data latches
MOVLW	0CFh	; Value used to
		; initialize data
		; direction
MOVWF	TRISC	; Set RC<3:0> as inputs
		; RC<5:4> as outputs
		; RC<7:6> as inputs

Pin Name	Function	TRIS Setting	I/O	I/O Type	Description
RC0/SOSCO/	RC0	0	0	DIG	LATC<0> data output.
SCLKI/		1	Ι	ST	PORTC<0> data input.
	SOSCO	1	Ι	ST	SOSC oscillator output.
	SCLKI	1	Ι	ST	Digital clock input; enabled when SOSC oscillator is disabled.
RC1/SOSCI/	RC1	0	0	DIG	LATC<1> data output.
ECCP2/P2A		1	Ι	ST	PORTC<1> data input.
	SOSCI	x	Ι	ANA	SOSC oscillator input.
	ECCP2 <sup>(1)</sup>	0	0	DIG	ECCP2 compare output and ECCP2 PWM output; takes priority over port data.
	1 I		ST	ECCP2 capture input.	
	P2A	0	0	DIG	ECCP2 Enhanced PWM output, Channel A. May be configured for tri-state during Enhanced PWM shutdown events; takes priority over port data.
RC2/ECCP1/	RC2	0	0	DIG	LATC<2> data output.
P1A		1	Ι	ST	PORTC<2> data input.
	ECCP1	0	0	DIG	ECCP1 compare output and ECCP1 PWM output; takes priority over port data.
		1	Ι	ST	ECCP1 capture input.
	P1A	0	0	DIG	ECCP1 Enhanced PWM output, Channel A. May be configured for tri-state during Enhanced PWM shutdown events; takes priority over port data.

#### TABLE 12-5:PORTC FUNCTIONS

**Legend:** O = Output, I = Input, ANA = Analog Signal, DIG = Digital Output, ST = Schmitt Trigger Buffer Input, TTL = TTL Buffer Input,  $I^2C = I^2C^{TM}$ /SMBus Buffer Input, x = Don't care (TRIS bit does not affect port direction or is overridden for this option).

**Note 1:** Default assignment for ECCP2 when the CCP2MX Configuration bit is set.

Pin Name	Function	TRIS Setting	I/O	l/O Type	Description
RC3/SCK1/	RC3	0	0	DIG	LATC<3> data output.
SCL1		1	Ι	ST	PORTC<3> data input.
	SCK1	0	0	DIG	SPI clock output (MSSP module); takes priority over port data.
		1	Ι	ST	SPI clock input (MSSP module).
	SCL1	0	0	DIG	I <sup>2</sup> C clock output (MSSP module); takes priority over port data.
		1	Ι	l <sup>2</sup> C	I <sup>2</sup> C clock input (MSSP module); input type depends on module setting.
RC4/SDI1/	RC4	0	0	DIG	LATC<4> data output.
SDA1		1	Ι	ST	PORTC<4> data input.
SDI			Ι	ST	SPI data input (MSSP module).
	SDA1	1	0	DIG	I <sup>2</sup> C data output (MSSP module); takes priority over port data.
		1	Ι	l <sup>2</sup> C	I <sup>2</sup> C data input (MSSP module); input type depends on module setting.
RC5/SDO1 RC5		0	0	DIG	LATC<5> data output.
		1	Ι	ST	PORTC<5> data input.
	SDO1	0	0	DIG	SPI data output (MSSP module).
RC6/TX1/CK1	RC6	0	0	DIG	LATC<6> data output.
		1	-	ST	PORTC<6> data input.
	TX1	1	0	DIG	Synchronous serial data output (EUSART module); takes priority over port data.
	CK1	1	0	DIG	Synchronous serial data input (EUSART module); user must configure as an input.
		1	Ι	ST	Synchronous serial clock input (EUSART module).
RC7/RX1/DT1	RC7	0	0	DIG	LATC<7> data output.
		1	Ι	ST	PORTC<7> data input.
	RX1	1	Ι	ST	Asynchronous serial receive data input (EUSART module).
	DT1	1	0	DIG	Synchronous serial data output (EUSART module); takes priority over port data.
		1	Ι	ST	Synchronous serial data input (EUSART module); user must configure as an input.

TABLE 12-5: PORTC FUNCTIONS (CONTINUED)

**Legend:** O = Output, I = Input, ANA = Analog Signal, DIG = Digital Output, ST = Schmitt Trigger Buffer Input, TTL = TTL Buffer Input,  $I^2C = I^2C^{TM}$ /SMBus Buffer Input, x = Don't care (TRIS bit does not affect port direction or is overridden for this option).

**Note 1:** Default assignment for ECCP2 when the CCP2MX Configuration bit is set.

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
PORTC	RC7	RC6	RC5	RC4	RC3	RC2	RC1	RC0
LATC	LATC7	LATBC6	LATC5	LATCB4	LATC3	LATC2	LATC1	LATC0
TRISC	TRISC7	TRISC6	TRISC5	TRISC4	TRISC3	TRISC2	TRISC1	TRISC0
ODCON1	SSP10D	CCP2OD	CCP10D	_	_	_	_	SSP2OD
ODCON3	U2OD	U10D	_	_	_	_	_	CTMUDS

Legend: Shaded cells are not used by PORTC.

#### 12.5 PORTD, TRISD and LATD Registers

PORTD is an 8-bit wide, bidirectional port. The corresponding Data Direction and Output Latch registers are TRISD and LATD.

All pins on PORTD are implemented with Schmitt Trigger input buffers. Each pin is individually configurable as an input or output.

Note:	These pins are configured as digital inputs
	on any device Reset.

Each of the PORTD pins has a weak internal pull-up. A single control bit can turn off all the pull-ups. This is performed by setting bit, RDPU (PADCFG1<7>). The weak pull-up is automatically turned off when the port pin is configured as an output. The pull-ups are disabled on all device Resets.

On 80-pin devices, PORTD is multiplexed with the system bus as part of the external memory interface. The I/O port and other functions are only available when the interface is disabled by setting the EBDIS bit (MEMCON<7>). When the interface is enabled, PORTD is the low-order byte of the multiplexed address/data bus (AD<7:0>). The TRISD bits are also overridden.

PORTD can also be configured as an 8-bit wide microprocessor port (Parallel Slave Port) by setting control bit, PSPMODE (TRISE<4>). In this mode, the input buffers are TTL. For additional information, see Section 12.11 "Parallel Slave Port".

The PORTD also has the I<sup>2</sup>C and SPI functionality on RD4, RD5 and RD6. The pins for SPI are also configurable for open-drain output. Open-drain configuration is selected by setting bit, SSP2OD (ODCON1<0>).

RD0 has a CTMU functionality. RD1 has the functionality for the Timer5 clock input and Timer7 external clock gate input.

EXAMP	LE 12-4:		INITIALIZING PORTD
CLRF	PORTD	;	Initialize PORTD by
		;	clearing output
		;	data latches
CLRF	LATD	;	Alternate method
		;	to clear output
		;	data latches
MOVLW	0CFh	;	Value used to
		;	initialize data
		;	direction
MOVWF	TRISD	;	Set RD<3:0> as inputs

; RD<5:4> as outputs

; RD<7:6> as inputs

Pin Name	Function	TRIS Setting	I/O	l/O Type	Description
RD0/PSP0/	RD0	0	0	DIG	LATD<0> data output.
AD0/CTPLS		1	Ι	ST	PORTD<0> data input.
	PSP0 <sup>(1)</sup>	x	I/O	TTL	Parallel Slave Port data.
	AD0 <sup>(2)</sup>	x	I/O	TTL	External Memory Address/Data 0.
	CTPLS	x	0	DIG	CTMU pulse generator output.
RD1/PSP1/	RD1	0	0	DIG	LATD<1> data output.
AD1/T5CKI/		1	Ι	ST	PORTD<1> data input.
T7G	PSP1 <sup>(1)</sup>	x	I/O	TTL	Parallel Slave Port data.
	AD1 <sup>(2)</sup>	x	I/O	TTL	External Memory Address/Data 1.
	T5CKI	x	Ι	ST	Timer5 clock input.
	T7G	x	Ι	ST	Timer7 external clock gate input.
RD2/PSP2/AD2	RD2	0	0	DIG	LATD<2> data output.
		1	Ι	ST	PORTD<2> data input.
	PSP2 <sup>(1)</sup>	x	I/O	TTL	Parallel Slave Port data.
	AD2 <sup>(2)</sup>	х	I/O	TTL	External Memory Address/Data 2.

#### TABLE 12-7: PORTD FUNCTIONS

Legend: O = Output, I = Input, ANA = Analog Signal, DIG = Digital Output, ST = Schmitt Trigger Buffer Input,  $I^2C = I^2C^{TM}/SMBus Buffer Input, x = Don't care (TRIS bit does not affect port direction or is overridden for this option).$ 

Note 1: The Parallel Slave Port (PSP) is available only in Microcontroller mode.

2: This feature is available only on PIC18F8XK22 devices.

IABLE 12-7:	PORTD FUNCTIONS (CONTINUED)								
Pin Name	Function	TRIS Setting	I/O	l/O Type	Description				
RD3/PSP3/AD3	RD3	0	0	DIG	LATD<3> data output.				
		1	I	ST	PORTD<3> data input.				
	PSP3 <sup>(1)</sup>	x	I/O	TTL	Parallel Slave Port data.				
	AD3 <sup>(2)</sup>	x	I/O	TTL	External Memory Address/Data 3.				
RD4/PSP4/	RD4	0	0	DIG	LATD<4> data output.				
AD4/SDO2		1	I	ST	PORTD<4> data input.				
	PSP4 <sup>(1)</sup>	x	I/O	TTL	Parallel Slave Port data.				
	AD4 <sup>(2)</sup>	x	I/O	TTL	External Memory Address/Data 4.				
	SDO2	0	Р	DOG	SPI data output (MSSP module).				
RD5/PSP5/	RD5	0	0	DIG	LATD<5> data output.				
AD5/SDI2/		1	I	ST	PORTD<5> data input.				
SDA2	PSP5 <sup>(1)</sup>	х	I/O	TTL	Parallel Slave Port data.				
	AD5 <sup>(2)</sup>	x	I/O	TTL	External Memory Address/Data 5.				
	SDI2	1	-	ST	SPI data input (MSSP module).				
	SDA2	0	0	l <sup>2</sup> C	I <sup>2</sup> C data input (MSSP module). Input type depends on module setting.				
RD6/PSP6/	RD6	0	0	DIG	LATD<6> data output.				
AD6/SCK2/ SCL2		1	-	ST	PORTD<6> data input.				
30L2	PSP6 <sup>(1)</sup>	x	I/O	TTL	Parallel Slave Port data.				
	AD6 <sup>(2)</sup>	x	I/O	TTL	External Memory Address/Data 6.				
	SCK2	0	0	DIG	SPI clock output (MSSP module); takes priority over port data.				
		1	Ι	ST	SPI clock input (MSSP module).				
	SCL2	0	0	DIG	I <sup>2</sup> C clock output (MSSP module); takes priority over port data.				
		1	I	l <sup>2</sup> C	I <sup>2</sup> C clock input (MSSP module). Input type depends on module setting.				
RD7/PSP7/	RD7	0	0	DIG	LATD<7> data output.				
AD7/SS2		1	Ι	ST	PORTD<7> data input.				
	PSP7 <sup>(1)</sup>	x	I/O	TTL	Parallel Slave Port data.				
	AD7 <sup>(2)</sup>	x	I/O	TTL	External Memory Address/Data 7.				
	SS2	1	I	TTL	Slave select input for MSSP module.				

<b>TABLE 12-7:</b>	PORTD FUNCTIONS	(CONTINUED)
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**Legend:** O = Output, I = Input, ANA = Analog Signal, DIG = Digital Output, ST = Schmitt Trigger Buffer Input,  $I^2C = I^2C^{TM}$ /SMBus Buffer Input, x = Don't care (TRIS bit does not affect port direction or is overridden for this option).

**Note** 1: The Parallel Slave Port (PSP) is available only in Microcontroller mode.

2: This feature is available only on PIC18F8XK22 devices.

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
PORTD	RD7	RD6	RD5	RD4	RD3	RD2	RD1	RD0
LATD	LATD7	LATD6	LATD5	LATD4	LATD3	LATD2	LATD1	LATD0
TRISD	TRISD7	TRISD6	TRISD5	TRISD4	TRISD3	TRISD2	TRISD1	TRISD0
PADCFG1	RDPU	REPU	RJPU <sup>(1)</sup>		_	RTSECSEL1	RESECSEL0	_
ODCON1	SSP10D	CCP2OD	CCP10D		_	_	—	SSP2OD

Legend: Shaded cells are not used by PORTD.

Note 1: Unimplemented on PIC18F6XK22 devices, read as '0'.

### 12.6 PORTE, TRISE and LATE Registers

PORTE is an eight-bit wide, bidirectional port. The corresponding Data Direction and Output Latch registers are TRISE and LATE.

All pins on PORTE are implemented with Schmitt Trigger input buffers. Each pin is individually configurable as an input or output. The RE7 pin is also configurable for open-drain output when ECCP2 is active on this pin. Open-drain configuration is selected by setting the CCP2OD control bit (ODCON1<6>)

Note:	These pins are configured as digital inputs
	on any device Reset.

Each of the PORTE pins has a weak internal pull-up. A single control bit can turn off all the pull-ups. This is performed by setting bit, REPU (PADCFG1<6>). The weak pull-up is automatically turned off when the port pin is configured as an output. The pull-ups are disabled on any device Reset.

PORTE is also multiplexed with Enhanced PWM Outputs, B and C for ECCP1 and ECCP3, for Outputs, B, C and D for ECCP2. For all devices, their default assignments are on PORTE<6:0>.

On 80-pin devices, the multiplexing for the outputs of ECCP1 and ECCP3 is controlled by the ECCPMX Configuration bit. Clearing this bit re-assigns the P1B/P1C and P3B/P3C outputs to PORTH.

For devices operating in Microcontroller mode, the RE7 pin can be configured as the alternate peripheral pin for the ECCP2 module and Enhanced PWM Output 2A. This is done by clearing the CCP2MX Configuration bit. PORTE is also multiplexed with the Parallel Slave Port address lines. RE1 and RE0 are multiplexed with the control signals, WR and RD.

RE3 can also be configured as the Reference Clock Output (REFO) from the system clock. For further details, see **Section 3.7 "Reference Clock Output"**.

EXAMPLE 12-5:	INITIALIZING PORTE

CLRF	PORTE	; Initialize PORTE by
		; clearing output
		; data latches
CLRF	LATE	; Alternate method
		; to clear output
		; data latches
MOVLW	03h	; Value used to
		; initialize data
		; direction
MOVWF	TRISE	; Set RE<1:0> as inputs
		; RE<7:2> as outputs

Function	TRIS Setting	I/O	I/O Type	Description
RE0	0	0	DIG	LATE<0> data output.
	1	Ι	ST	PORTE<0> data input.
RD	х	0	DIG	Parallel Slave Port read strobe pin.
	x	Ι	TTL	Parallel Slave Port read pin.
P2D	0	0		ECCP2 PWM Output D. May be configured for tri-state during Enhanced PWM shutdown events.
AD8 <sup>(2)</sup>	x	0	DIG	External memory interface, Data Bit 8 output.
	x	Ι	TTL	External memory interface, Data Bit 8 input.
RE1	0	0	DIG	LATE<1> data output.
	1	Ι	ST	PORTE<1> data input.
P2C	0	0	_	ECCP2 PWM Output C. May be configured for tri-state during Enhanced PWM shutdown events.
WR	х	0	DIG	Parallel Slave Port write strobe pin.
	x	I	TTL	Parallel Slave Port write pin.
AD9 <sup>(2)</sup>	x	0	DIG	External memory interface, Data Bit 9 output.
	x	Ι	TTL	External memory interface, Data Bit 9 input.
	RE0 RD P2D AD8 <sup>(2)</sup> RE1 P2C WR	Function         Setting           RE0         0           I         I           RD         x           P2D         0           AD8 <sup>(2)</sup> x           RE1         0           P2C         0           WR         x           X         x           X         X           X         X           X         X           X         X           X         X           X         X           X         X           X         X           X         X	Function         Setting         I/O           RE0         0         0           1         1         1           RD         x         0           AD8 <sup>(2)</sup> x         0           RE1         0         0           P2C         0         0           WR         x         0           WR         x         0           X         1         1	Function         Setting         I/O         Type           RE0         0         O         DIG           1         I         ST           RD         x         O         DIG           RD         x         O         DIG           RD         x         O         DIG           RD         x         I         TTL           P2D         0         O         —           AD8 <sup>(2)</sup> x         O         DIG           X         I         TTL           P2D         0         O         DIG           X         I         TTL           RE1         0         O         DIG           P2C         0         O         DIG           WR         x         O         DIG           X         X         DIG         —           WR         x         O         DIG           X         I         TTL         —

#### TABLE 12-9: PORTE FUNCTIONS

Legend: O = Output, I = Input, ANA = Analog Signal, DIG = Digital Output, ST = Schmitt Trigger Buffer Input, x = Don't care (TRIS bit does not affect port direction or is overridden for this option).

Note 1: Alternate assignment for ECCP2 when the CCP2MX Configuration bit is cleared and in Microcontroller mode.

2: This feature is only available on PIC18F8XKXX devices.

TABLE 12-9:     PORTE FUNCTIONS (CONTINUED)       Pin Name     Function       TRIS     I/O       VO     Description					
Pin Name	Function	Setting	1/0	Туре	Description
RE2/CS/P2B/	RE2	0	0	DIG	LATE<2> data output.
CCP10/AD10		1	Ι	ST	PORTE<2> data input.
	CS	x	Ι	TTL	Parallel Slave Port chip select.
	P2B	0	0	—	ECCP2 PWM Output B. May be configured for tri-state during Enhanced PWM shutdown events
	CCP10	1	I/O	ST	Capture 10 input/Compare 10 output/PWM10 output.
	AD10 <sup>(2)</sup>	x	0	DIG	External memory interface, Address/Data Bit 10 output.
		x	-	TTL	External memory interface, Data Bit 10 input.
RE3/P3C/	RE3	0	0	DIG	LATE<3> data output.
CCP9/REFO/		1	Ι	ST	PORTE<3> data input.
AD11	P3C	0	0	—	ECCP3 PWM Output C. May be configured for tri-state during Enhanced PWM shutdown events
	CCP9	0	0	DIG	CCP9 Compare/PWM output; takes priority over port data.
		1	Ι	ST	CCP9 capture input.
	REFO	x	0	DIG	Reference output clock.
	AD11 <sup>(2)</sup>	х	0	DIG	External memory interface, Address/Data Bit 11 output.
		x	Ι	TTL	External memory interface, Data Bit 11 input.
RE4/P3B/	RE4	0	0	DIG	LATE<4> data output.
CCP8/AD12		1	Ι	ST	PORTE<4> data input.
	P3B	0	0	—	ECCP3 PWM Output B. May be configured for tri-state during Enhanced PWM shutdown events
	CCP8	0	0	DIG	CCP8 compare/PWM output; takes priority over port data.
		1	I	ST	CCP8 capture input.
	AD12 <sup>(2)</sup>	x	0	DIG	External memory interface, Address/Data Bit 12 output.
		x	I	TTL	External memory interface, Data Bit 12 input.
RE5/P1C/	RE5	0	0	DIG	LATE<5> data output.
CCP7/AD13		1	Ι	ST	PORTE<5> data input.
	P1C	0	0	—	ECCP1 PWM Output C. May be configured for tri-state during Enhanced PWM shutdown events
	CCP7	0	0	DIG	CCP7 compare/PWM output; takes priority over port data.
		1	I	ST	CCP7 capture input.
	AD13 <sup>(2)</sup>	x	0	DIG	External memory interface, Address/Data Bit 13 output.
		x	1	TTL	External memory interface, Data Bit 13 input.
RE6/P1B/	RE6	0	0	DIG	LATE<6> data output.
CCP6/AD14		1	Ι	ST	PORTE<6> data input.
	P1B	0	0	—	ECCP1 PWM Output B. May be configured for tri-state during Enhanced PWM shutdown events
	CCP6	0	0	DIG	CCP6 compare/PWM output; takes priority over port data.
		1	Ι	ST	CCP9 capture input.
	AD14 <sup>(2)</sup>	x	0	DIG	External memory interface, Address/Data Bit 14 output.
		x		TTL	External memory interface, Data Bit 14 input.

#### TABLE 12-9: PORTE FUNCTIONS (CONTINUED)

Legend: O = Output, I = Input, ANA = Analog Signal, DIG = Digital Output, ST = Schmitt Trigger Buffer Input, x = Don't care (TRIS bit does not affect port direction or is overridden for this option).

Note 1: Alternate assignment for ECCP2 when the CCP2MX Configuration bit is cleared and in Microcontroller mode.

2: This feature is only available on PIC18F8XKXX devices.

Pin Name	Function	TRIS Setting	I/O	I/O Type	Description
RE7/ECCP2/	RE7	0	0	DIG	LATE<7> data output.
P2A/AD15		1	Ι	ST	PORTE<7> data input.
	ECCP2 <sup>(1)</sup>	0	0	DIG	ECCP2 compare/PWM output; takes priority over port data.
		1	Ι	ST	ECCP2 capture input.
	P2A	0	0		ECCP2 PWM Output A. May be configured for tri-state during Enhanced PWM shutdown event.
	AD15 <sup>(2)</sup>	x	0	DIG	External memory interface, Address/Data Bit 15 output.
		x	Ι	TTL	External memory interface, Data Bit 15 input.

#### TABLE 12-9: PORTE FUNCTIONS (CONTINUED)

**Legend:** O = Output, I = Input, ANA = Analog Signal, DIG = Digital Output, ST = Schmitt Trigger Buffer Input, x = Don't care (TRIS bit does not affect port direction or is overridden for this option).

Note 1: Alternate assignment for ECCP2 when the CCP2MX Configuration bit is cleared and in Microcontroller mode.

2: This feature is only available on PIC18F8XKXX devices.

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
PORTE	RE7	RE6	RE5	RE4	RE3	RE2	RE1	RE0
LATE	LATE7	LATE6	LATE5	LATE4	LATE3	LATE2	LATE1	LATE0
TRISE	TRISE7	TRISE6	TRISE5	TRISE4	TRISE3	TRISE2	TRISE1	TRISE0
PADCFG1	RDPU	REPU	RJPU <sup>(2)</sup>	_	_	RTSECSEL1	RTSECSEL0	_
ODCON1	SSP10D	CCP2OD	CCP10D	_	_	—	—	SSP2OD
ODCON2	CCP100D <sup>(1)</sup>	CCP9OD <sup>(1)</sup>	CCP8OD	CCP7OD	CCP6OD	CCP5OD	CCP4OD	CCP3OD

Legend: Shaded cells are not used by PORTE.

Note 1: Unimplemented on PIC18FX5K22 devices, read as '0'.

2: Unimplemented on 64-pin devices (PIC18F6XK22), read as '0'.

#### 12.7 PORTF, LATF and TRISF Registers

PORTF is a 7-bit wide, bidirectional port. The corresponding Data Direction and Output Latch registers are TRISF and LATF. All pins on PORTF are implemented with Schmitt Trigger input buffers. Each pin is individually configurable as an input or output.

Pins, RF1 through RF6, may be used as comparator inputs or outputs by setting the appropriate bits in the CMCON register. To use RF<7:1> as digital inputs, it is also necessary to turn off the comparators.

- **Note 1:** On device Resets, pins, RF<7:1>, are configured as analog inputs and are read as '0'.
  - To configure PORTF as a digital I/O, turn off the comparators and clear ANCON1 and ANCON2 to digital.

#### EXAMPLE 12-6: INITIALIZING PORTF

CLRF	PORTF	; Initialize PORTF by ; clearing output ; data latches
CLRF	LATF	; Alternate method ; to clear output ; data latches
BANKSEL	ANCON1	; Select bank with ANCON1 register
MOVLW	1Fh	; Make AN6, AN7 and AN5 digital
MOVWF	ANCON1	;
MOVLW	OFh	; Make AN8, AN9, AN10 and AN11 digital
MOVWF	ANCON	; Set PORTF as digital I/O
BANKSEL	TRISF	; Select bank with TRISF register
MOVLW	0CEh	; Value used to
		; initialize data
		; direction
MOVWF	TRISF	; Set RF3:RF1 as inputs
		; RF5:RF4 as outputs
		; RF7:RF6 as inputs

Pin Name	Function	TRIS Setting	I/O	l/O Type	Description				
RF1/AN6/C2OUT/	RF1	0	0	DIG	LATF<1> data output; not affected by analog input.				
CTDIN		1	I	ST	PORTF<1> data input; disabled when analog input is enabled.				
	AN6	1	Ι	ANA	A/D Input Channel 6. Default configuration on POR.				
	C2OUT	0	0	DIG	Comparator 2 output; takes priority over port data.				
	CTDIN	1	I	ST	CTMU pulse delay input.				
RF2/AN7/C1OUT	RF2	0	0	DIG	LATF<2> data output; not affected by analog input.				
		1	I	ST	PORTF<2> data input; disabled when analog input is enabled.				
	AN7	1	I	ANA	A/D Input Channel 7. Default configuration on POR.				
	C10UT	0	0	DIG	Comparator 1 output; takes priority over port data.				
RF3/AN8/C2INB/	RF3	0	0	DIG	LATF<3> data output; not affected by analog input.				
CTMUI		1	I	ST	PORTF<3> data input; disabled when analog input is enabled.				
-	AN8	1	I	ANA	A/D Input Channel 8 and Comparator C2+ input. Default input configuration on POR; not affected by analog output.				
	C2INB	1	Ι	ANA	Comparator 2 Input B.				
	CTMUI	х	0	—	CTMU pulse generator charger for the C2INB comparator input.				
RF4/AN9/C2INA	RF4	0	0	DIG	LATF<4> data output; not affected by analog input.				
		1	-	ST	PORTF<4> data input; disabled when analog input is enabled.				
	AN9	1	I	ANA	A/D Input Channel 9 and Comparator C2- input. Default input configuration on POR; does not affect digital output.				
	C2INA	1	I	ANA	Comparator 2 Input A.				
RF5/AN10/CVREF/ C1INB	RF5	0	0	DIG	LATF<5> data output; not affected by analog input. Disabled when CVREF output is enabled.				
		1	I	ST	PORTF<5> data input; disabled when analog input is enabled. Disabled when CVREF output is enabled.				
	AN10	1	I	ANA	A/D Input Channel 10 and Comparator C1+ input. Default input configuration on POR.				
	CVREF	х	0	ANA	Comparator voltage reference output. Enabling this feature disables digital I/O.				
	C1INB	1	I	ANA	Comparator 1 Input B.				

#### TABLE 12-11: PORTF FUNCTIONS

Legend: O = Output, I = Input, ANA = Analog Signal, DIG = Digital Output, ST = Schmitt Trigger Buffer Input,

TTL = TTL Buffer Input, x = Don't care (TRIS bit does not affect port direction or is overridden for this option).

Pin Name	Function	TRIS Setting	I/O	l/O Type	Description
RF6/AN11/C1INA	RF6	0	0	DIG	LATF<6> data output; not affected by analog input.
		1	I	ST	PORTF<6> data input; disabled when analog input is enabled.
	AN11	1	I	ANA	A/D Input Channel 11 and Comparator C1- input. Default input configuration on POR; does not affect digital output.
	C1INA	1	I	ANA	Comparator 1 Input A.
RF7/AN5/SS1	RF7	0	0	DIG	LATF<7> data output; not affected by analog input.
		1	I	ST	PORTF<7> data input; disabled when analog input is enabled.
	AN5	1	Ι	ANA	A/D Input Channel 5. Default configuration on POR.
	SS1	1	I	TTL	Slave select input for MSSP module.

### TABLE 12-11: PORTF FUNCTIONS (CONTINUED)

Legend: O = Output, I = Input, ANA = Analog Signal, DIG = Digital Output, ST = Schmitt Trigger Buffer Input, TTL = TTL Buffer Input, x = Don't care (TRIS bit does not affect port direction or is overridden for this option).

#### TABLE 12-12: SUMMARY OF REGISTERS ASSOCIATED WITH PORTF

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
PORTF	RF7	RF6	RF5	RF4	RF3	RF2	RF1	—
LATF	LATF7	LATF6	LATF5	LATF4	LATF3	LATF2	LATF1	—
TRISF	TRISF7	TRISF6	TRISF5	TRISF4	TRISF3	TRISF2	TRISF1	_
ANCON0	ANSEL7	ANSEL6	ANSEL5	ANSEL4	ANSEL3	ANSEL2	ANSEL1	ANSEL0
ANCON1	ANSEL15	ANSEL14	ANSEL13	ANSEL12	ANSEL11	ANSEL10	ANSEL9	ANSEL8

**Legend:** — = unimplemented, read as '0'. Shaded cells are not used by PORTF.

# 12.8 PORTG, TRISG and LATG Registers

PORTG is a 5-bit wide, bidirectional port. The corresponding Data Direction and Output Latch registers are TRISG and LATG.

PORTG is multiplexed with the AUSART and CCP, ECCP, Analog, Comparator, RTCC and Timer input functions (Table 12-13). When operating as I/O, all PORTG pins have Schmitt Trigger input buffers. The open-drain functionality for the CCPx and UART can be configured using ODCONx.

When enabling peripheral functions, care should be taken in defining TRIS bits for each PORTG pin. Some peripherals override the TRIS bit to make a pin an output, while other peripherals override the TRIS bit to make a pin an input. The user should refer to the corresponding peripheral section for the correct TRIS bit settings. The pin override value is not loaded into the TRIS register. This allows read-modify-write of the TRIS register without concern due to peripheral overrides.

#### EXAMPLE 12-7: INITIALIZING PORTG

CLRF	PORTG	;	Initialize PORTG by
		;	clearing output
		;	data latches
BCF	CM1CON, C	ON ;	disable
		;	comparator 1
CLRF	LATG	;	Alternate method
		;	to clear output
		;	data latches
BANKSEL	ANCON2	;	Select bank with ACON2 register
MOVLW	0F0h	;	make AN16 to AN19
		;	digital
MOVWF	ANCON2		
BANKSEL	TRISG	;	Select bank with TRISG register
MOVLW	04h	;	Value used to
		;	initialize data
		;	direction
MOVWF	TRISG	;	Set RG1:RG0 as
		;	outputs
		;	RG2 as input
		;	RG4:RG3 as inputs

#### TABLE 12-13: PORTG FUNCTIONS

Pin Name	Function	TRIS Setting	I/O	I/O Type	Description
RG0/ECCP3/	RG0	0	0	DIG	LATG<0> data output.
P3A		1	Ι	ST	PORTG<0> data input.
	ECCP3	0	0	DIG	ECCP3 compare output and ECCP3 PWM output; takes priority over port data.
		1	Ι	ST	ECCP3 capture input.
	P3A	0	0	—	ECCP3 PWM Output A. May be configured for tri-state during Enhanced PWM shutdown events.
RG1/TX2/CK2/	RG1	0	0	DIG	LATG<1> data output.
AN19/C3OUT		1	Ι	ST	PORTG<1> data input.
	TX2	1	0	DIG	Synchronous serial data output (EUSART module); takes priority over port data.
	CK2	1	0	DIG	Synchronous serial data input (EUSART module); user must configure as an input.
		1	I	ST	Synchronous serial clock input (EUSART module).
	AN19	1	Ι	ANA	A/D Input Channel 19. Default input configuration on POR. Does not affect digital output.
	C3OUT	x	0	DIG	Comparator 3 output.

Legend: O = Output, I = Input, ANA = Analog Signal, DIG = Digital Output, ST = Schmitt Trigger Buffer Input,

x = Don't care (TRIS bit does not affect port direction or is overridden for this option).

# TABLE 12-13: PORTG FUNCTIONS (CONTINUED)

Pin Name	Function	TRIS Setting	I/O	l/O Type	Description				
RG2/RX2/DT2/	RG2	0	0	DIG	LATG<2> data output.				
AN18/C3INA		1	Ι	ST	PORTG<2> data input.				
Ì	RX2	1	I	ST	Asynchronous serial receive data input (EUSART module).				
	DT2	1	0	DIG	Synchronous serial data output (EUSART module); takes priority over port data.				
		1	I	ST	Synchronous serial data input (EUSART module); user must configure as an input.				
	AN18	1	Ι	ANA	A/D Input Channel 18. Default input configuration on POR; does not affect digital output.				
	C3INA	x	Ι	ANA	Comparator 3 Input A.				
RG3/CCP4/AN17/	RG3	0	0	DIG	LATG<3> data output.				
P3D/C3INB		1	Ι	ST	PORTG<3> data input.				
Ì	CCP4	0	0	DIG	CCP4 compare/PWM output; takes priority over port data.				
		1	Ι	CCP4 capture input.					
	AN17	1	I	ANA	A/D Input Channel 17. Default input configuration on POR; does not affect digital output.				
	P3D	0	0		ECCP3 PWM Output D. May be configured for tri-state during Enhanced PWM.				
Ì	C3INB	x	Ι	ANA	Comparator 3 Input B.				
RG4/RTCC/	RG4	0	0	DIG	LATG<4> data output.				
T7CKI/T5G/		1	Ι	ST	PORTG<4> data input.				
CCP5/AN16/ P1D/C3INC	RTCC	x	0	DIG	RTCC output.				
	T7CKI	x	Ι	ST	Timer7 clock input.				
Ì	T5G	x	Ι	ST	Timer5 external clock gate input.				
Ì	CCP5	0	0	DIG	CCP5 compare/PWM output; takes priority over port data.				
		1	Ι	ST	CCP5 capture input.				
	AN16	1	Ι	ANA	A/D Input Channel 17. Default input configuration on POR; does not affect digital output.				
	P1D	0	0		ECCP1 PWM Output D. May be configured for tri-state during Enhanced PWM.				
Ī	C3INC	х	Ι	ANA	Comparator 3 Input C.				

Legend: O = Output, I = Input, ANA = Analog Signal, DIG = Digital Output, ST = Schmitt Trigger Buffer Input, x = Don't care (TRIS bit does not affect port direction or is overridden for this option).

#### TABLE 12-14: SUMMARY OF REGISTERS ASSOCIATED WITH PORTG

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
PORTG	—	_	RG5 <sup>(2)</sup>	RG4	RG3	RG2	RG1	RG0
TRISG	—		_	TRISG4	TRISG3	TRISG2	TRISG1	TRISG0
ANCON2	ANSEL23	ANSEL22	ANSEL21	ANSEL20	ANSEL19	ANSEL18	ANSEL17	ANSEL16
ODCON1	SSP10D	CCP2OD	CCP10D	_				SSP2OD
ODCON2	CCP100D <sup>(1)</sup>	CCP9OD <sup>(1)</sup>	CCP8OD	CCP7OD	CCP6OD	CCP5OD	CCP4OD	CCP3OD
ODCON3	U2OD	U10D	_		—	_	_	CTMUDS

**Legend:** — = unimplemented, read as '0'. Shaded cells are not used by PORTG.

Note 1: Unimplemented on PIC18FX5K22 devices, read as '0'.

2: This bit is available when Master Clear is disabled (MCLRE = 0). When MCLRE is set, the bit is unimplemented.

#### 12.9 PORTH, LATH and TRISH Registers

Note:	PORTH is	available	only	on	the	80-pin
	devices.					

PORTH is an 8-bit wide, bidirectional I/O port. The corresponding Data Direction and Output Latch registers are TRISH and LATH.

All pins on PORTH are implemented with Schmitt Trigger input buffers. Each pin is individually configurable as an input or output.

#### EXAMPLE 12-8: INITIALIZING PORTH

CLRF	PORTH	;	Initialize PORTH by clearing output data latches
CLRF	LATH	;;	Alternate method to clear output data latches
BANKSEL	ANCON2	;	Select bank with ANCON2 register
MOVLW	0Fh	;	Configure PORTH as
MOVWF	ANCON2	;	digital I/O
MOVLW	0Fh	;	Configure PORTH as
MOVWF	ANCON1	;	digital I/O
BANKSEL	TRISH	;	Select bank with TRISH register
MOVLW	0CFh	;	Value used to
		;	initialize data
		;	direction
MOVWF	TRISH	;	Set RH3:RH0 as inputs
		;	RH5:RH4 as outputs
		;	RH7:RH6 as inputs

Pin Name	Function	TRIS Setting	I/O	l/O Type	Description			
RH0/AN23/A16	RH0	0 0 O		DIG	LATH<0> data output.			
		1	Ι	ST	PORTH<0> data input.			
	AN23	1	Ι	ANA	A/D Input Channel 23. Default input configuration on POR; does not affect digital input.			
	A16	x	0	DIG	External memory interface, Address Line 16; takes priority over port data.			
RH1/AN22/A17	RH1	0	0	DIG	LATH<1> data output.			
		1	Ι	ST	PORTH<1> data input.			
	AN22	1	Ι	ANA	A/D Input Channel 22. Default input configuration on POR; does not affect digital input.			
	A17	x	0	DIG	External memory interface, Address Line 17; takes priority over port data.			
RH2/AN21/A18	RH2	0	0	DIG	LATH<2> data output.			
		1	I	ST	PORTH<2> data input.			
	AN21	1	Ι	ANA	A/D Input Channel 21. Default input configuration on POR; does not affect digital input.			
	A18	x	0	DIG	External memory interface, Address Line 18; takes priority over port data.			
RH3/AN20/A19	RH3	0	0	DIG	LATH<3> data output.			
		1	I	ST	PORTH<3> data input.			
	AN20	1	Ι	ANA	A/D Input Channel 20. Default input configuration on POR; does not affect digital input.			
	A19	x	0	DIG	External memory interface, Address Line 19; takes priority over port data.			

#### TABLE 12-15: PORTH FUNCTIONS

Legend: O = Output, I = Input, ANA = Analog Signal, DIG = Digital Output, ST = Schmitt Trigger Buffer Input, x = Don't care (TRIS bit does not affect port direction or is overridden for this option).

# TABLE 12-15: PORTH FUNCTIONS (CONTINUED)

Pin Name	Function	TRIS Setting	I/O	I/O Type	Description			
RH4/CCP9/	RH4	0	0	DIG	LATH<4> data output.			
P3C/AN12/		1	Ι	ST	PORTH<4> data input.			
C2INC	CCP9	0	0	DIG	CCP9 compare/PWM output; takes priority over port data.			
		1	I	ST	CCP9 capture input.			
	P3C	0	0	_	ECCP3 PWM Output C. May be configured for tri-state during Enhanced PWM.			
	AN12	1	Ι	ANA	A/D Input Channel 12. Default input configuration on POR; does not affect digital input.			
	C2INC	x	Ι	ANA	Comparator 2 Input C.			
RH5/CCP8/	RH5	0	0	DIG	LATH<5> data output.			
P3B/AN13/ C2IND		1	Ι	ST	PORTH<5> data input.			
	CCP8	28 0 O DIG CCP8 compare/PWM output;		DIG	CCP8 compare/PWM output; takes priority over port data.			
		1	Ι	ST	CCP8 capture input.			
-	P3B	0	0	—	ECCP3 PWM Output B. May be configured for tri-state during Enhanced PWM.			
	AN13	1	I	ANA	A/D Input Channel 13. Default input configuration on POR; does not affect digital input.			
	C2IND	x	Ι	ANA	Comparator 2 Input D.			
RH6/CCP7/	RH6	0	0	DIG	LATH<6> data output.			
P1C/AN14/		1	I	ST	PORTH<6> data input.			
C1INC	CCP7	0	0	DIG	CCP7 compare/PWM output; takes priority over port data.			
		1	Ι	ST	CCP7 capture input.			
	P1C	0	0	—	ECCP1 PWM Output C. May be configured for tri-state during Enhanced PWM.			
	AN14	1	Ι	ANA	A/D Input Channel 14. Default input configuration on POR; does not affect digital input.			
	C1INC	x	-	ANA	Comparator 1 Input C.			
RH7/CCP6/	RH7	0	0	DIG	LATH<7> data output.			
P1B/AN15		1	Ι	ST	PORTH<7> data input.			
	CCP6	0	0	DIG	CCP6 compare/PWM output; takes priority over port data.			
		1	Ι	ST	CCP6 capture input.			
	P1B	0	0	—	ECCP1 PWM Output B. May be configured for tri-state during Enhanced PWM.			
	AN15	1	I	ANA	A/D Input Channel 15. Default input configuration on POR; does not affect digital input.			

**Legend:** O = Output, I = Input, ANA = Analog Signal, DIG = Digital Output, ST = Schmitt Trigger Buffer Input, x = Don't care (TRIS bit does not affect port direction or is overridden for this option).

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
PORTH <sup>(1)</sup>	RH7	RH6	RH5	RH4	RH3	RH2	RH1	RH0
LATH <sup>(1)</sup>	LATH7	LATH6	LATH5	LATH4	LATH3	LATH2	LATH1	LATH0
TRISH <sup>(1)</sup>	TRISH7	TRISH6	TRISH5	TRISH4	TRISH3	TRISH2	TRISH1	TRISH0
ANCON1	ANSEL15	ANSEL14	ANSEL13	ANSEL12	ANSEL11	ANSEL10	ANSEL9	ANSEL8
ANCON2	ANSEL23	ANSEL22	ANSEL21	ANSEL20	ANSEL19	ANSEL18	ANSEL17	ANSEL16
ODCON2	CCP10OD <sup>(2)</sup>	CCP9OD <sup>(2)</sup>	CCP8OD	CCP7OD	CCP6OD	CCP5OD	CCP4OD	CCP3OD

#### TABLE 12-16: SUMMARY OF REGISTERS ASSOCIATED WITH PORTH

**Note 1:** Unimplemented on 64-pin devices (PIC18F6XK22), read as '0'.

2: Unimplemented on PIC18FX5K22 devices, read as '0'.

# 12.10 PORTJ, TRISJ and LATJ Registers

#### Note: PORTJ is available only on 80-pin devices.

PORTJ is an 8-bit wide, bidirectional port. The corresponding Data Direction and Output Latch registers are TRISJ and LATJ.

All pins on PORTJ are implemented with Schmitt Trigger input buffers. Each pin is individually configurable as an input or output.

Note:	These pins are configured as digital inputs
	on any device Reset.

When the external memory interface is enabled, all of the PORTJ pins function as control outputs for the interface. This occurs automatically when the interface is enabled by clearing the EBDIS control bit (MEMCON<7>). The TRISJ bits are also overridden. Each of the PORTJ pins has a weak internal pull-up. The pull-ups are provided to keep the inputs at a known state for the external memory interface while powering up. A single control bit can turn off all the pull-ups. This is performed by clearing bit, RJPU (PADCFG1<5>). The weak pull-up is automatically turned off when the port pin is configured as an output. The pull-ups are disabled on any device Reset.

CLRF	PORTJ	; Initialize PORTJ by
CLRF	LATJ	; clearing output latches ; Alternate method
		; to clear output latches
MOVLW	0CFh	; Value used to
		; initialize data
		; direction
MOVWF	TRISJ	; Set RJ3:RJ0 as inputs
		; RJ5:RJ4 as output
		; RJ7:RJ6 as inputs

Pin Name	Function	TRIS Setting	I/O	I/O Type	Description
RJ0/ALE	RJ0	0	0	DIG	LATJ<0> data output.
		1	I	ST	PORTJ<0> data input.
	ALE	x	0	DIG	External memory interface address latch enable control output; takes priority over digital I/O.
RJ1/OE	RJ1	0	0	DIG	LATJ<1> data output.
		1	Ι	ST	PORTJ<1> data input.
	ŌĒ	х	0	DIG	External memory interface output enable control output; takes priority over digital I/O.
RJ2/WRL	RJ2	0	0	DIG	LATJ<2> data output.
		1	I	ST	PORTJ<2> data input.
	WRL	x	0	DIG	External Memory Bus write low byte control; takes priority over digital I/O.
RJ3/WRH	RJ3	0	0	DIG	LATJ<3> data output.
		1	I	ST	PORTJ<3> data input.
	WRH	x	0	DIG	External memory interface write high-byte control; takes priority over digital I/O.
RJ4/BA0	RJ4	0	0	DIG	LATJ<4> data output.
		1	I	ST	PORTJ<4> data input.
	BA0	x	0	DIG	External Memory Interface Byte Address 0 control output; takes priority over digital I/O.
RJ5/CE	RJ5	0	0	DIG	LATJ<5> data output.
		1	I	ST	PORTJ<5> data input.
	CE	x	0	DIG	External memory interface chip enable control output; takes priority over digital I/O.
RJ6/LB	RJ6	0	0	DIG	LATJ<6> data output.
		1	I	ST	PORTJ<6> data input.
	LB	x	0	DIG	External memory interface lower byte enable control output; takes priority over digital I/O.
RJ7/UB	RJ7	0	0	DIG	LATJ<7> data output.
		1	I	ST	PORTJ<7> data input.
	UB	x	0	DIG	External memory interface upper byte enable control output; takes priority over digital I/O.

### TABLE 12-17: PORTJ FUNCTIONS

Legend: O = Output, I = Input, ANA = Analog Signal, DIG = Digital Output, ST = Schmitt Trigger Buffer Input,

x = Don't care (TRIS bit does not affect port direction or is overridden for this option).

<b>TABLE 12-18:</b>	SUMMARY OF REGISTERS ASSOCIATED WITH PORTJ
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Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
PORTJ <sup>(1)</sup>	RJ7	RJ6	RJ5	RJ4	RJ3	RJ2	RJ1	RJ0
LATJ <sup>(1)</sup>	LATJ7	LATJ6	LATJ5	LATJ4	LATJ3	LATJ2	LATJ1	LATJ0
TRISJ <sup>(1)</sup>	TRISJ7	TRISJ6	TRISJ5	TRISJ4	TRISJ3	TRISJ2	TRISJ1	TRISJ0
PADCFG1	RDPU	REPU	rjpu <sup>(1)</sup>	_	_	RTSECSEL1	RTSECSEL0	—

Legend: Shaded cells are not used by PORTJ.

Note 1: Unimplemented on 64-pin devices (PIC18F6XK22), read as '0'.

# 12.11 Parallel Slave Port

PORTD can function as an 8-bit-wide Parallel Slave Port (PSP), or microprocessor port, when control bit, PSPMODE (PSPCON<4>), is set. The port is asynchronously readable and writable by the external world through the RD control input pin (RE0/P2D/RD/AD8) and WR control input pin (RE1/P2C/WR/AD9).

Note:	The Parallel Slave Port is available only in
	Microcontroller mode.

The PSP can directly interface to an 8-bit microprocessor data bus. The external microprocessor can read or write the PORTD latch as an eight-bit latch.

Setting bit, PSPMODE, enables port pin, RE0/P2D/RD/AD8, to be the RD input, RE1/P2C/WR/AD9 to be the WR input and RE2/P2B/CCP10/CS/AD10 to be the CS (Chip Select) input. For this functionality, the corresponding data direction bits of the TRISE register (TRISE<2:0>) must be configured as inputs (= 111).

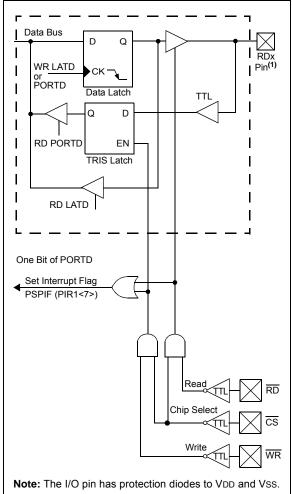
A write to the PSP occurs when both the  $\overline{CS}$  and  $\overline{WR}$  lines are first detected low and ends when either are detected high. The PSPIF and IBF flag bits (PIR1<7> and PSPCON<7>, respectively) are set when the write ends.

A read from the PSP occurs when both the  $\overline{CS}$  and  $\overline{RD}$  lines are first detected low. The data in PORTD is read out and the OBF bit (PSPCON<6>) is set. If the user writes new data to PORTD to set OBF, the data is immediately read out, but the OBF bit is not set.

When either the  $\overline{CS}$  or  $\overline{RD}$  line is detected high, the PORTD pins return to the input state and the PSPIF bit is set. User applications should wait for PSPIF to be set before servicing the PSP. When this happens, the IBF and OBF bits can be polled and the appropriate action taken.

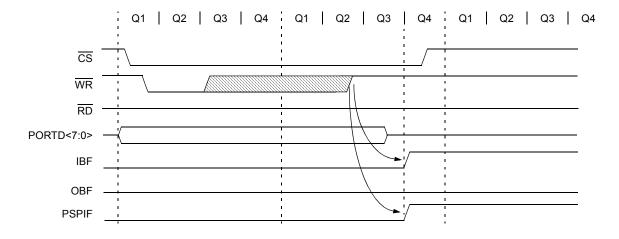
The timing for the control signals in Write and Read modes is shown in Figure 12-4 and Figure 12-5, respectively.

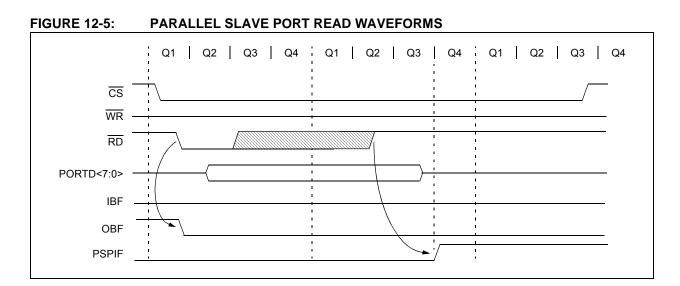
# FIGURE 12-3: PORTD AND PORTE BLOCK DIAGRAM (PARALLEL SLAVE PORT)



### REGISTER 12-5: PSPCON: PARALLEL SLAVE PORT CONTROL REGISTER

R-0	R-0	R/W-0	R/W-0	U-0	U-0	U-0	U-0
IBF	OBF	IBOV	PSPMODE	—	—	—	_
bit 7							bit C
Legend:							
R = Readable	e bit	W = Writable	bit	U = Unimpler	nented bit, read	as '0'	
-n = Value at	POR	'1' = Bit is se	t	'0' = Bit is cle	ared	x = Bit is unkno	own
bit 7	IBF: Input But	ffer Full Status	bit				
	-	is been receiv	ed and is waiti	ng to be read l	by the CPU		
bit 6	OBF: Output	Buffer Full Sta	tus bit				
	1 = The outpu 0 = The outpu		olds a previous een read	ly written word			
bit 5	IBOV: Input B	uffer Overflow	Detect bit				
	1 = A write oc 0 = No overflo		a previously inp	out word had n	ot been read (m	nust be cleared i	in software)
bit 4	PSPMODE: P	Parallel Slave I	Port Mode Sele	ect bit			
	1 = Parallel S		•				
	0 = General P	-					
bit 3-0	Unimplement	ted: Read as	0.				
FIGURE 12-	4: PARAI	LLEL SLAVI		TE WAVEFO	RMS		





#### TABLE 12-19: REGISTERS ASSOCIATED WITH PARALLEL SLAVE PORT

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
PORTD	RD7	RD6	RD5	RD4	RD3	RD2	RD1	RD0
LATD	LATD7	LATD6	LATD5	LATD4	LATD3	LATD2	LATD1	LATD0
TRISD	TRISD7	TRISD6	TRISD5	TRISD4	TRISD3	TRISD2	TRISD1	TRISD0
PORTE	RE7	RE6	RE5	RE4	RE3	RE2	RE1	RE0
LATE	LATE7	LATE6	LATE5	LATE4	LATE3	LATE2	LATE1	LATE0
TRISE	TRISE7	TRISE6	TRISE5	TRISE4	TRISE3	TRISE2	TRISE1	TRISE0
PSPCON	IBF	OBF	IBOV	PSPMODE		—	_	—
INTCON	GIE/GIEH	PEIE/GIEL	TMR0IE	INT0IE	RBIE	TMR0IF	INT0IF	RBIF
PIR1	PSPIF	ADIF	RC1IF	TX1IF	SSP1IF	TMR1GIF	TMR2IF	TMR1IF
PIE1	PSPIE	ADIE	RC1IE	TX1IE	SSP1IE	TMR1GIE	TMR2IE	TMR1IE
IPR1	PSPIP	ADIP	RC1IP	TX1IP	SSP1IP	TMR1GIP	TMR2IP	TMR1IP
PMD1	PSPMD	CTMUMD	RTCCMD	TMR4MD	TMR3MD	TMR2MD	TMR1MD	EMBDM

**Legend:** — = unimplemented, read as '0'. Shaded cells are not used by the Parallel Slave Port.

NOTES:

# 13.0 TIMER0 MODULE

The Timer0 module incorporates the following features:

- Software-selectable operation as a timer or counter in both 8-bit or 16-bit modes
- · Readable and writable registers
- Dedicated 8-bit, software programmable
   prescaler
- Selectable clock source (internal or external)
- · Edge select for external clock
- Interrupt-on-overflow

The T0CON register (Register 13-1) controls all aspects of the module's operation, including the prescale selection. It is both readable and writable.

Figure 13-1 provides a simplified block diagram of the Timer0 module in 8-bit mode. Figure 13-2 provides a simplified block diagram of the Timer0 module in 16-bit mode.

#### REGISTER 13-1: TOCON: TIMER0 CONTROL REGISTER

R/W-1	R/W-1	R/W-1	R/W-1	R/W-1	R/W-1	R/W-1	R/W-1
TMR00N	T08BIT	TOCS	TOSE	PSA	T0PS2	T0PS1	T0PS0
bit 7							bit 0

Legend:									
R = Reada	able bit	W = Writable bit	U = Unimplemented bit	, read as '0'					
-n = Value	at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown					
bit 7	TMR0ON	I: Timer0 On/Off Control bit							
	1 = Enab	les Timer0							
	0 = Stops	s Timer0							
bit 6	<b>T08BIT</b> : 1	Fimer0 8-Bit/16-Bit Control bi	t						
	1 = Timer	r0 is configured as an 8-bit ti	mer/counter						
		r0 is configured as a 16-bit ti							
bit 5	TOCS: Tir	mer0 Clock Source Select bit	t						
	1 = Trans	sition on TOCKI pin input edg	e						
		nal clock (Fosc/4)							
bit 4	TOSE: Tir	mer0 Source Edge Select bit							
	1 = Increi	ment on high-to-low transition	n on T0CKI pin						
	0 = Increi	ment on low-to-high transition	n on T0CKI pin						
bit 3	PSA: Tim	ner0 Prescaler Assignment bi	it						
	1 = Timer	r0 prescaler is not assigned;	Timer0 clock input bypasses	orescaler					
	0 = Timer	0 prescaler is assigned; Tim	er0 clock input comes from pr	escaler output					
bit 2-0	T0PS<2:0>: Timer0 Prescaler Select bits								
	111 <b>= 1</b> :2	256 Prescale value							
	110 <b>= 1</b> :1	128 Prescale value							
		64 Prescale value							
		32 Prescale value							
	•	16 Prescale value							
		B Prescale value							
	001 = 1:4 000 = 1:2	Prescale value Prescale value							

### 13.1 Timer0 Operation

Timer0 can operate as either a timer or a counter. The mode is selected with the T0CS bit (T0CON<5>). In Timer mode (T0CS = 0), the module increments on every clock by default unless a different prescaler value is selected (see **Section 13.3 "Prescaler**"). If the TMR0 register is written to, the increment is inhibited for the following two instruction cycles. The user can work around this by writing an adjusted value to the TMR0 register.

The Counter mode is selected by setting the T0CS bit (= 1). In this mode, Timer0 increments either on every rising edge or falling edge of the T0CKI pin. The incrementing edge is determined by the Timer0 Source Edge Select bit, T0SE (T0CON<4>); clearing this bit selects the rising edge. Restrictions on the external clock input are discussed below.

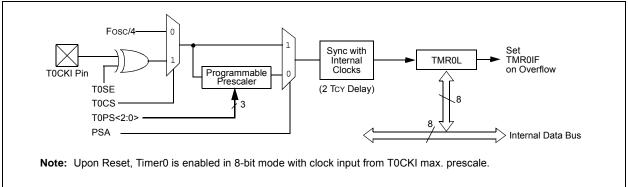
An external clock source can be used to drive Timer0; however, it must meet certain requirements to ensure that the external clock can be synchronized with the internal phase clock (Tosc). There is a delay between synchronization and the onset of incrementing the timer/counter.

# 13.2 Timer0 Reads and Writes in 16-Bit Mode

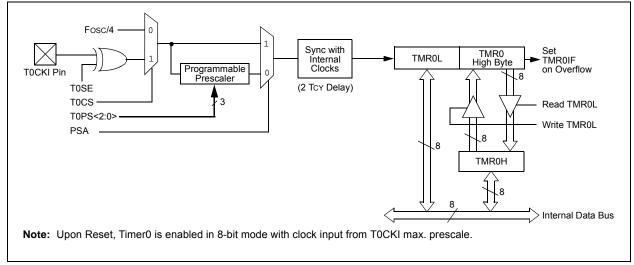
TMR0H is not the actual high byte of Timer0 in 16-bit mode. It is actually a buffered version of the real high byte of Timer0, which is not directly readable nor writable (see Figure 13-2). TMR0H is updated with the contents of the high byte of Timer0 during a read of TMR0L. This provides the ability to read all 16 bits of Timer0 without having to verify that the read of the high and low byte was valid, due to a rollover between successive reads of the high and low byte.

Similarly, a write to the high byte of Timer0 must also take place through the TMR0H Buffer register. The high byte is updated with the contents of TMR0H when a write occurs to TMR0L. This allows all 16 bits of Timer0 to be updated at once.

#### FIGURE 13-1: TIMER0 BLOCK DIAGRAM (8-BIT MODE)



#### FIGURE 13-2: TIMER0 BLOCK DIAGRAM (16-BIT MODE)



#### 13.3 Prescaler

An 8-bit counter is available as a prescaler for the Timer0 module. The prescaler is not directly readable or writable. Its value is set by the PSA and T0PS<2:0> bits (T0CON<3:0>), which determine the prescaler assignment and prescale ratio.

Clearing the PSA bit assigns the prescaler to the Timer0 module. When it is assigned, prescale values from 1:2 through 1:256, in power-of-two increments, are selectable.

When assigned to the Timer0 module, all instructions writing to the TMR0 register (for example, CLRF TMR0, MOVWF TMR0, BSF TMR0) clear the prescaler count.

Note:	Writing to TMR0 when the prescaler is
	assigned to Timer0 will clear the prescaler
	count but will not change the prescaler
	assignment.

#### 13.3.1 SWITCHING PRESCALER ASSIGNMENT

The prescaler assignment is fully under software control and can be changed "on-the-fly" during program execution.

### 13.4 Timer0 Interrupt

The TMR0 interrupt is generated when the TMR0 register overflows from FFh to 00h in 8-bit mode, or from FFFFh to 0000h in 16-bit mode. This overflow sets the TMR0IF flag bit. The interrupt can be masked by clearing the TMR0IE bit (INTCON<5>). Before reenabling the interrupt, the TMR0IF bit must be cleared in software by the Interrupt Service Routine (ISR).

Since Timer0 is shut down in Sleep mode, the TMR0 interrupt cannot awaken the processor from Sleep.

TABLE 13-1:	REGISTERS ASSOCIATED WITH TIMER0

Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
Immed         Timer0 Register Low Byte							
Timer0 Register High Byte							
GIE/GIEH	PEIE/GIEL	TMR0IE	INTOIE	RBIE	TMR0IF	<b>INT0IF</b>	RBIF
TMR0ON	T08BIT	TOCS	TOSE	PSA	T0PS2	T0PS1	T0PS0
	Timer0 Regis Timer0 Regis GIE/GIEH	Timer0 Register Low Byte Timer0 Register High Byte GIE/GIEH PEIE/GIEL	Timer0 Register Low Byte Timer0 Register High Byte GIE/GIEH PEIE/GIEL TMR0IE	Timer0 Register Low Byte Timer0 Register High Byte GIE/GIEH PEIE/GIEL TMR0IE INT0IE	Timer0 Register Low Byte       Timer0 Register High Byte       GIE/GIEH     PEIE/GIEL       TMR0IE     INT0IE	Timer0 Register Low Byte       Timer0 Register High Byte       GIE/GIEH     PEIE/GIEL       TMR0IE     INT0IE       RBIE     TMR0IF	Timer0 Register Low Byte       Timer0 Register High Byte       GIE/GIEH     PEIE/GIEL       TMR0IE     INT0IE       RBIE     TMR0IF

Legend: — = unimplemented, read as '0'. Shaded cells are not used by Timer0.

NOTES:

# 14.0 TIMER1 MODULE

The Timer1 timer/counter module incorporates these features:

- Software-selectable operation as a 16-bit timer or counter
- Readable and writable 8-bit registers (TMR1H and TMR1L)
- Selectable clock source (internal or external) with device clock or SOSC oscillator internal options
- Interrupt-on-overflow
- Reset on ECCP Special Event Trigger
- · Timer with gated control

Figure 14-1 displays a simplified block diagram of the Timer1 module.

The Timer1 oscillator can also be used as a low-power clock source for the microcontroller in power-managed operation. The Timer1 can also work on the SOSC oscillator.

Timer1 is controlled through the T1CON Control register (Register 14-1). It also contains the Secondary Oscillator Enable bit (SOSCEN). Timer1 can be enabled or disabled by setting or clearing control bit, TMR10N (T1CON<0>).

The FOSC clock source should not be used with the ECCP capture/compare features. If the timer will be used with the capture or compare features, always select one of the other timer clocking options.

	1.1.1.00			OIOTEN			
R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0
TMR1CS1	TMR1CS0	T1CKPS1	T1CKPS0	SOSCEN	T1SYNC	RD16	TMR10N
bit 7							bit 0

#### REGISTER 14-1: T1CON: TIMER1 CONTROL REGISTER

Legend:			
R = Readable bit	W = Writable bit	U = Unimplemented bit, read	l as '0'
-n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown

bit 7-6	<pre>TMR1CS&lt;1:0&gt;: Timer1 Clock Source Select bits 10 = Timer1 clock source is either from a pin or oscillator, depending on the SOSCEN bit: SOSCEN = 0: External clock from the T1CKI pin (on the rising edge). SOSCEN = 1: Depending on the SOSCSEL Configuration bit, the clock source is either a crystal oscillator on the SOSCI/SOSCO pins or an internal clock from the SCLKI pin. 01 = Timer1 clock source is the system clock (Fosc)<sup>(1)</sup> 00 = Timer1 clock source is the instruction clock (Fosc/4)</pre>
bit 5-4	<b>T1CKPS&lt;1:0&gt;:</b> Timer1 Input Clock Prescale Select bits 11 = 1:8 Prescale value 10 = 1:4 Prescale value 01 = 1:2 Prescale value 00 = 1:1 Prescale value
bit 3	<b>SOSCEN:</b> SOSC Oscillator Enable bit 1 = SOSC is enabled and available for Timer1 0 = SOSC is disabled for Timer1 The oscillator inverter and feedback resistor are turned off to eliminate power drain.
bit 2	<b>T1SYNC:</b> Timer1 External Clock Input Synchronization Select bit $\underline{TMR1CS<1:0> = 10:}$ 1 = Do not synchronize external clock input0 = Synchronize external clock input $\underline{TMR1CS<1:0> = 0x:}$ This bit is ignored. Timer1 uses the internal clock when TMR1CS<1:0> = 1x.
bit 1	<b>RD16:</b> 16-Bit Read/Write Mode Enable bit 1 = Enables register read/write of Timer1 in one 16-bit operation 0 = Enables register read/write of Timer1 in two 8-bit operations
bit 0	TMR1ON: Timer1 On bit 1 = Enables Timer1 0 = Stops Timer1

Note 1: The FOSC clock source should not be selected if the timer will be used with the ECCP capture/compare features.

#### 14.1 Timer1 Gate Control Register

The Timer1 Gate Control register (T1GCON), displayed in Register 14-2, is used to control the Timer1 gate.

# REGISTER 14-2: T1GCON: TIMER1 GATE CONTROL REGISTER<sup>(1)</sup>

R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R-x	R/W-0	R/W-0
TMR1GE	T1GPOL	T1GTM	T1GSPM	T1GGO/T1DONE	T1GVAL	T1GSS1	T1GSS0
bit 7							bit 0
Legend:	1. 1.4				1.1.11	(0)	
R = Readab		W = Writable		U = Unimplemented	d dit, read as		
-n = Value a	IT POR	'1' = Bit is set		'0' = Bit is cleared		x = Bit is unkr	iown
bit 7	TMR1GE: Ti	mer1 Gate Ena	able bit				
	If TMR10N =						
	This bit is igr						
	If TMR10N =		rollod by the	Timer1 gate function			
		ounts regardle					
bit 6	T1GPOL: Tir	mer1 Gate Pol	arity bit				
	•		•	ounts when gate is hig			
	•		•	unts when gate is low	()		
bit 5		er1 Gate Togg					
		Gate Toggle m		d d and toggle flip-flop	ic cloared		
		flip-flop toggle			13 Cleared		
bit 4	T1GSPM: Ti	mer1 Gate Sin	gle Pulse Mo	de bit			
		•		nabled and is controll	ing Timer1 g	ate	
		Bate Single Pul					
bit 3			•	ulse Acquisition Statu			
				is ready, waiting for a has completed or ha		tarted	
				GSPM is cleared.		laiteu	
bit 2	T1GVAL: Tir	ner1 Gate Cur	rent State bit				
				gate that could be p	provided to T	MR1H:TMR1L;	unaffected by
L:1 4 0		Enable (TMR1		-4 6 4-			
bit 1-0		Timer1 Gate	Source Sele	CT DITS			
		rator 2 output rator 1 output					
		o match PR2 of	output				
	00 <b>= Timer1</b>	gate pin					
Note 1: F	Programming th	e T1GCON pr	ior to T1CON	is recommended			

**Note 1:** Programming the T1GCON prior to T1CON is recommended.

# 14.2 Timer1 Operation

The Timer1 module is an 8 or 16-bit incrementing counter that is accessed through the TMR1H:TMR1L register pair.

When used with an internal clock source, the module is a timer and increments on every instruction cycle. When used with an external clock source, the module can be used as either a timer or counter. It increments on every selected edge of the external source.

Timer1 is enabled by configuring the TMR1ON and TMR1GE bits in the T1CON and T1GCON registers, respectively.

When SOSC is selected as Crystal mode (by SOSCSEL), the RC1/SOSCI/ECCP2/P2A and RC0/ SOSCO/SCLKI pins become inputs. This means the values of TRISC<1:0> are ignored and the pins are read as '0'.

# 14.3 Clock Source Selection

The TMR1CS<1:0> and SOSCEN bits of the T1CON register are used to select the clock source for Timer1. Register 14-1 displays the clock source selections.

#### 14.3.1 INTERNAL CLOCK SOURCE

When the internal clock source is selected, the TMR1H:TMR1L register pair will increment on multiples of Fosc, as determined by the Timer1 prescaler.

### 14.3.2 EXTERNAL CLOCK SOURCE

When the external clock source is selected, the Timer1 module may work as a timer or a counter.

When enabled to count, Timer1 is incremented on the rising edge of the external clock input, T1CKI. Either of these external clock sources can be synchronized to the microcontroller system clock or they can run asynchronously.

When used as a timer with a clock oscillator, an external, 32.768 kHz crystal can be used in conjunction with the dedicated internal oscillator circuit.

**Note:** In Counter mode, a falling edge must be registered by the counter prior to the first incrementing rising edge after any one or more of the following conditions:

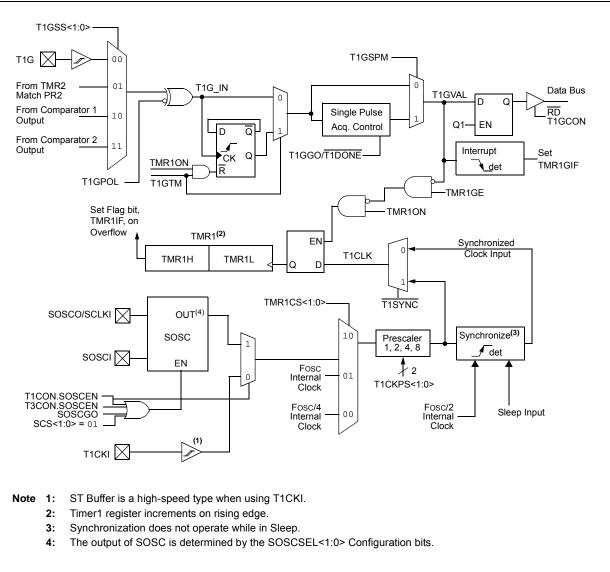
- Timer1 enabled after POR Reset
- Write to TMR1H or TMR1L
- Timer1 is disabled
- Timer1 is disabled (TMR1ON = 0)

When T1CKI is high, Timer1 is enabled (TMR1ON = 1) when T1CKI is low.

TMR1CS1	TMR1CS0	SOSCEN Clock Source					
0	1	x	Clock Source (Fosc)				
0	0	x Instruction Clock (Fosc/4)					
1	0	0 External Clock on T1CKI Pin					
1	0	1	Oscillator Circuit on SOSCI/SOSCO Pins				

#### TABLE 14-1: TIMER1 CLOCK SOURCE SELECTION





### 14.4 Timer1 16-Bit Read/Write Mode

Timer1 can be configured for 16-bit reads and writes. When the RD16 control bit (T1CON<1>) is set, the address for TMR1H is mapped to a buffer register for the high byte of Timer1. A read from TMR1L loads the contents of the high byte of Timer1 into the Timer1 High Byte Buffer register. This provides the user with the ability to accurately read all 16 bits of Timer1 without having to determine whether a read of the high byte, followed by a read of the low byte, has become invalid due to a rollover between reads.

A write to the high byte of Timer1 must also take place through the TMR1H Buffer register. The Timer1 high byte is updated with the contents of TMR1H when a write occurs to TMR1L. This allows a user to write all 16 bits at once to both the high and low bytes of Timer1.

The high byte of Timer1 is not directly readable or writable in this mode. All reads and writes must take place through the Timer1 High Byte Buffer register. Writes to TMR1H do not clear the Timer1 prescaler; the prescaler is only cleared on writes to TMR1L.

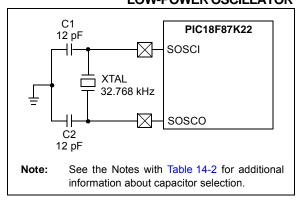
#### 14.5 SOSC Oscillator

An on-chip crystal oscillator circuit is incorporated between pins, SOSCI (input) and SOSCO (amplifier output). It is enabled by any peripheral that requests it. There are eight ways the SOSC can be enabled: if the SOSC is selected as the source by any of the odd timers, which is done by each respective SOSCEN bit (TxCON<3>), if the SOSC is selected as the RTCC source by the RTCOSC Configuration bit (CONFIG3L<1>), if the SOSC is selected as the CPU clock source by the SCS bits (OSCCON<1:0>) or if the SOSCGO bit is set (OSCCON2<3>). The SOSCGO bit is used to warm up the SOSC so that it is ready before any peripheral requests it. The oscillator is a low-power circuit, rated for 32 kHz crystals. It will continue to run during all power-managed modes. The circuit for a typical low-power oscillator is depicted in Figure 14-2. Table 14-2 provides the capacitor selection for the SOSC oscillator.

The user must provide a software time delay to ensure proper start-up of the SOSC oscillator.

#### FIGURE 14-2:

#### EXTERNAL COMPONENTS FOR THE SOSC LOW-POWER OSCILLATOR



# TABLE 14-2: CAPACITOR SELECTION FOR THE TIMER OSCILLATOR<sup>(2,3,4,5)</sup>

Oscillator Type	Freq.	C1	C2					
LP	32 kHz	12 pF <sup>(1)</sup>	12 pF <sup>(1)</sup>					
Note 1: Microchip suggests these values as a starting point in validating the oscillator circuit.								
i	Higher capacit ty of the oscilla start-up time.							

- 3: Since each resonator/crystal has its own characteristics, the user should consult the resonator/crystal manufacturer for appropriate values of external components.
- 4: Capacitor values are for design guidance only. Values listed would be typical of a CL = 10 pF rated crystal when SOSCSEL<1:0> = 11.
- Incorrect capacitance value may result in a frequency not meeting the crystal manufacturer's tolerance specification.

The SOSC crystal oscillator drive level is determined based on the SOSCSEL<1:0> (CONFIG1L<4:3>) Configuration bits. The Higher Drive Level mode, SOSCSEL<1:0> = 11, is intended to drive a wide variety of 32.768 kHz crystals with a variety of Capacitance Load (CL) ratings.

The Lower Drive Level mode is highly optimized for extremely low-power consumption. It is not intended to drive all types of 32.768 kHz crystals. In the Low Drive Level mode, the crystal oscillator circuit may not work correctly if excessively large discrete capacitors are placed on the SOSCO and SOSCI pins. This mode is designed to work only with discrete capacitances of approximately 3 pF-10 pF on each pin.

Crystal manufacturers usually specify a CL (Capacitance Load) rating for their crystals. This value is related to, but not necessarily the same as, the values that should be used for C1 and C2 in Figure 14-2.

For more details on selecting the optimum C1 and C2 for a given crystal, see the crystal manufacturer's applications information. The optimum value depends, in part, on the amount of parasitic capacitance in the circuit, which is often unknown. For that reason, it is highly recommended that thorough testing and validation of the oscillator be performed after values have been selected.

#### 14.5.1 USING SOSC AS A CLOCK SOURCE

The SOSC oscillator is also available as a clock source in power-managed modes. By setting the System Clock Select bits, SCS<1:0> (OSCCON<1:0>), to '01', the device switches to SEC\_RUN mode, and both the CPU and peripherals are clocked from the SOSC oscillator. If the IDLEN bit (OSCCON<7>) is cleared and a SLEEP instruction is executed, the device enters SEC\_IDLE mode. Additional details are available in Section 4.0 "Power-Managed Modes".

Whenever the SOSC oscillator is providing the clock source, the SOSC System Clock Status flag, SOSCRUN (OSCCON2<6>), is set. This can be used to determine the controller's current clocking mode. It can also indicate the clock source currently being used by the Fail-Safe Clock Monitor (FSCM).

If the Clock Monitor is enabled and the SOSC oscillator fails while providing the clock, polling the SOCSRUN bit will indicate whether the clock is being provided by the SOSC oscillator or another source.

# 14.5.2 SOSC OSCILLATOR LAYOUT CONSIDERATIONS

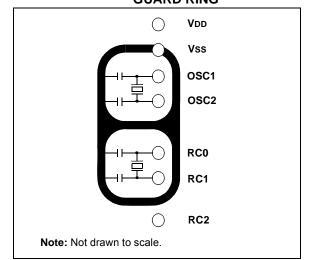
The SOSC oscillator circuit draws very little power during operation. Due to the low-power nature of the oscillator, it may also be sensitive to rapidly changing signals in close proximity. This is especially true when the oscillator is configured for extremely Low-Power mode, SOSCSEL<1:0> (CONFIG1L<4:3>) = 01.

The oscillator circuit, displayed in Figure 14-2, should be located as close as possible to the microcontroller. There should be no circuits passing within the oscillator circuit boundaries other than Vss or VDD.

If a high-speed circuit must be located near the oscillator, it may help to have a grounded guard ring around the oscillator circuit. The guard, as displayed in Figure 14-3, could be used on a single-sided PCB or in addition to a ground plane. (Examples of a high-speed circuit include the ECCP1 pin, in Output Compare or PWM mode, or the primary oscillator using the OSC2 pin.)



#### : OSCILLATOR CIRCUIT WITH GROUNDED GUARD RING



In the Low Drive Level mode, SOSCSEL<1:0> = 01, it is critical that RC2 I/O pin signals be kept away from the oscillator circuit. Configuring RC2 as a digital output, and toggling it, can potentially disturb the oscillator circuit, even with a relatively good PCB layout. If possible, either leave RC2 unused or use it as an input pin with a slew rate limited signal source. If RC2 must be used as a digital output, it may be necessary to use the Higher Drive Level Oscillator mode (SOSCSEL<1:0> = 11) with many PCB layouts.

Even in the Higher Drive Level mode, careful layout procedures should still be followed when designing the oscillator circuit.

In addition to dV/dt induced noise considerations, it is important to ensure that the circuit board is clean. Even a very small amount of conductive soldering flux residue can cause PCB leakage currents that can overwhelm the oscillator circuit.

# 14.6 Timer1 Interrupt

The TMR1 register pair (TMR1H:TMR1L) increments from 0000h to FFFFh and rolls over to 0000h. The Timer1 interrupt, if enabled, is generated on overflow which is latched in interrupt flag bit, TMR1IF (PIR1<0>). This interrupt can be enabled or disabled by setting or clearing the Timer1 Interrupt Enable bit, TMR1IE (PIE1<0>).

### 14.7 Resetting Timer1 Using the ECCP Special Event Trigger

If ECCP modules are configured to use Timer1 and to generate a Special Event Trigger in Compare mode (CCPxM<3:0> = 1011), this signal will reset Timer1. The trigger from ECCP2 will also start an A/D conversion, if the A/D module is enabled. (For more information, see **Section 20.3.4 "Special Event Trigger**".)

To take advantage of this feature, the module must be configured as either a timer or a synchronous counter. When used this way, the CCPRxH:CCPRxL register pair effectively becomes a period register for Timer1.

If Timer1 is running in Asynchronous Counter mode, this Reset operation may not work.

In the event that a write to Timer1 coincides with a Special Event Trigger, the write operation will take precedence.

Note:	The Special Eve	ent Trigger	from the			
	ECCPx module wi	Il only clear	the TMR1			
	register's content, but not set the TMR1IF					
	interrupt flag bit (P	IR1<0>).				

# 14.8 Timer1 Gate

Timer1 can be configured to count freely or the count can be enabled and disabled using the Timer1 gate circuitry. This is also referred to as Timer1 gate count enable.

Timer1 gate can also be driven by multiple selectable sources.

# 14.8.1 TIMER1 GATE COUNT ENABLE

The Timer1 Gate Enable mode is enabled by setting the TMR1GE bit of the T1GCON register. The polarity of the Timer1 Gate Enable mode is configured using the T1GPOL bit (T1GCON<6>).

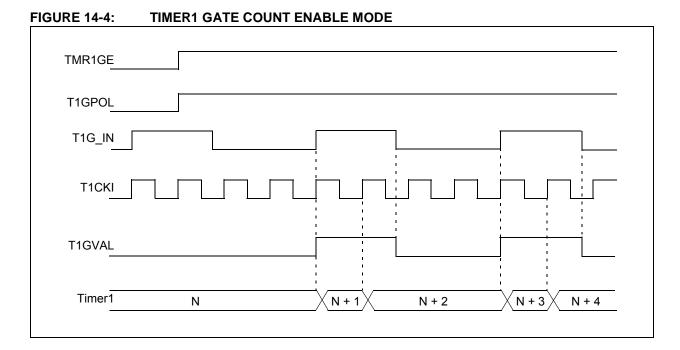
When Timer1 Gate Enable mode is enabled, Timer1 will increment on the rising edge of the Timer1 clock source. When Timer1 Gate Enable mode is disabled, no incrementing will occur and Timer1 will hold the current count. See Figure 14-4 for timing details.

TABLE 14-3:	TIMER1 GATE ENABLE
	SELECTIONS

T1CLK <sup>(†)</sup>	T1GPOL (T1GCON<6>)	T1G Pin	Timer1 Operation
1	0	0	Counts
$\uparrow$	0	1	Holds Count
$\uparrow$	1	0	Holds Count
$\uparrow$	1	1	Counts

† The clock on which TMR1 is running. For more information, see Figure 14-1.

Note: The CCP and ECCP modules use Timers, 1 through 8, for some modes. The assignment of a particular timer to a CCP/ECCP module is determined by the Timer to CCP enable bits in the CCPTMRSx registers. For more details, see Register 19-2, Register 19-3 and Register 20-2.



#### 14.8.2 TIMER1 GATE SOURCE SELECTION

The Timer1 gate source can be selected from one of four sources. Source selection is controlled by the T1GSSx (T1GCON<1:0>) bits (see Table 14-4).

TABLE 14-4:	TIMER1 GATE SOURCES
-------------	---------------------

T1GSS<1:0>	Timer1 Gate Source
00	Timer1 Gate Pin
01	TMR2 to Match PR2 (TMR2 increments to match PR2)
10	Comparator 1 Output (comparator logic high output)
11	Comparator 2 Output (comparator logic high output)

The polarity for each available source is also selectable, controlled by the T1GPOL bit (T1GCON<6>).

#### 14.8.2.1 T1G Pin Gate Operation

The T1G pin is one source for Timer1 gate control. It can be used to supply an external source to the Timer1 gate circuitry.

#### 14.8.2.2 Timer2 Match Gate Operation

The TMR2 register will increment until it matches the value in the PR2 register. On the very next increment cycle, TMR2 will be reset to 00h. When this Reset

occurs, a low-to-high pulse will automatically be generated and internally supplied to the Timer1 gate circuitry. The pulse will remain high for one instruction cycle and will return back to a low state until the next match.

The T1GPOL bit determines when the Timer1 counter increments based on this pulse. When T1GPOL = 1, Timer1 increments for a single instruction cycle following a TMR2 match with PR2. When T1GPOL = 0, Timer1 increments continuously, except for the cycle following the match, when the gate signal goes from low-to-high.

#### 14.8.2.3 Comparator 1 Output Gate Operation

The output of Comparator 1 can be internally supplied to the Timer1 gate circuitry. After setting up Comparator 1 with the CM1CON register, Timer1 will increment depending on the transitions of the CMP1OUT (CMSTAT<5>) bit.

#### 14.8.2.4 Comparator 2 Output Gate Operation

The output of Comparator 2 can be internally supplied to the Timer1 gate circuitry. After setting up Comparator 2 with the CM2CON register, Timer1 will increment depending on the transitions of the CMP2OUT (CMSTAT<6>) bit.

#### 14.8.3 TIMER1 GATE TOGGLE MODE

When Timer1 Gate Toggle mode is enabled, it is possible to measure the full cycle length of a Timer1 gate signal, as opposed to the duration of a single level pulse.

The Timer1 gate source is routed through a flip-flop that changes state on every incrementing edge of the signal. (For timing details, see Figure 14-5.)

The T1GVAL bit (T1GCON<2>) indicates when the Toggled mode is active and the timer is counting.

The Timer1 Gate Toggle mode is enabled by setting the T1GTM bit (T1GCON<5>). When T1GTM is cleared, the flip-flop is cleared and held clear. This is necessary in order to control which edge is measured.

TMR1GE	
T1GPOL	
T1GTM	
T1G_IN	
Т1СКІ	
T1GVAL	
Timer1 N	$\underbrace{\times N+1}_{N+2} \underbrace{\times N+3}_{N+4} \underbrace{\times N+5}_{N+5} \underbrace{\times N+6}_{N+7} \underbrace{\times N+8}_{N+8}$

#### FIGURE 14-5: TIMER1 GATE TOGGLE MODE

#### 14.8.4 TIMER1 GATE SINGLE PULSE MODE

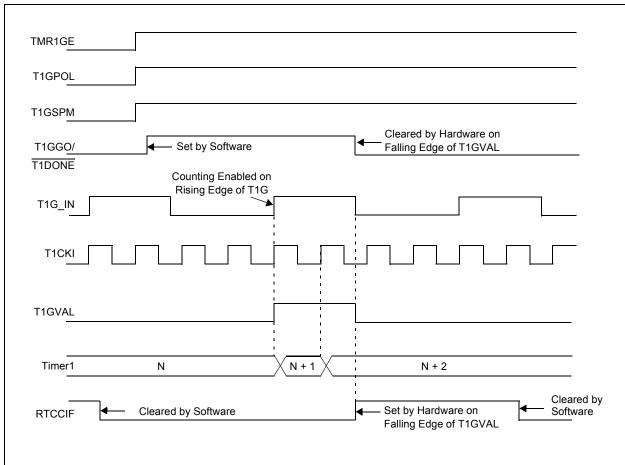
When Timer1 Gate Single Pulse mode is enabled, it is possible to capture a single pulse gate event. Timer1 Gate Single Pulse mode is enabled by setting the T1GSPM bit (T1GCON<4>) and the T1GGO/T1DONE bit (T1GCON<3>). The Timer1 will be fully enabled on the next incrementing edge.

On the next trailing edge of the pulse, the T1GGO/ T1DONE bit will automatically be cleared. No other gate events will be allowed to increment Timer1 until the T1GGO/T1DONE bit is once again set in software. Clearing the T1GSPM <u>bit of the</u> T1GCON register will also clear the T1GGO/T1DONE bit. (For timing details, see Figure 14-6.)

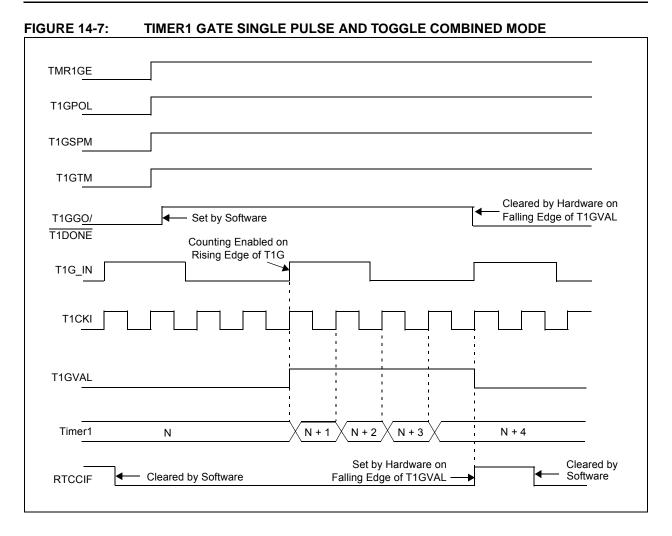
Simultaneously enabling the Toggle and Single Pulse modes will permit both sections to work together. This allows the cycle times on the Timer1 gate source to be measured. (For timing details, see Figure 14-7.)

#### 14.8.5 TIMER1 GATE VALUE STATUS

When the Timer1 gate value status is utilized, it is possible to read the most current level of the gate control value. The value is stored in the T1GVAL bit (T1GCON<2>). This bit is valid even when the Timer1 gate is not enabled (TMR1GE bit is cleared).



#### FIGURE 14-6: TIMER1 GATE SINGLE PULSE MODE



#### TABLE 14-5: REGISTERS ASSOCIATED WITH TIMER1 AS A TIMER/COUNTER

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
INTCON	GIE/GIEH	PEIE/GIEL	TMR0IE	INT0IE	RBIE	TMR0IF	INT0IF	RBIF
PIR1	PSPIF	ADIF	RC1IF	TX1IF	SSP1IF	TMR1GIF	TMR2IF	TMR1IF
PIE1	PSPIE	ADIE	RC1IE	TX1IE	SSP1IE	TMR1GIE	TMR2IE	TMR1IE
IPR1	PSPIP	ADIP	RC1IP	TX1IP	SSP1IP	TMR1GIP	TMR2IP	TMR1IP
TMR1L	Timer1 Regi	Timer1 Register Low Byte						
TMR1H	Timer1 Regi	ster High Byte	е					
T1CON	TMR1CS1	TMR1CS0	T1CKPS1	T1CKPS0	SOSCEN	T1SYNC	RD16	TMR10N
T1GCON	TMR1GE	T1GPOL	T1GTM	T1GSPM	T1GGO/ T1DONE	T1GVAL	T1GSS1	T1GSS0
OSCCON2	_	SOSCRUN	_	—	SOSCGO	_	MFIOFS	MFIOSEL
PMD1	PSPMD	CTMUMD	RTCCMD	TMR4MD	TMR3MD	TMR2MD	TMR1MD	EMBDM

Legend: Shaded cells are not used by the Timer1 module.

Note 1: Unimplemented on 32-Kbyte devices (PIC18FX5K22).

NOTES:

# 15.0 TIMER2 MODULE

The Timer2 module incorporates the following features:

- Eight-bit Timer and Period registers (TMR2 and PR2, respectively)
- Both registers are readable and writable
- Software programmable prescaler (1:1, 1:4 and 1:16)
- Software programmable postscaler (1:1 through 1:16)
- Interrupt on TMR2 to PR2 match
- Optional use as the shift clock for the MSSP modules

This module is controlled through the T2CON register (Register 15-1) that enables or disables the timer, and configures the prescaler and postscaler. Timer2 can be shut off by clearing control bit, TMR2ON (T2CON<2>), to minimize power consumption.

A simplified block diagram of the module is shown in Figure 15-1.

# 15.1 Timer2 Operation

In normal operation, TMR2 is incremented from 00h on each clock (Fosc/4). A four-bit counter/prescaler on the clock input gives the prescale options of direct input, divide-by-4 or divide-by-16. These are selected by the prescaler control bits, T2CKPS<1:0> (T2CON<1:0>).

The value of TMR2 is compared to that of the Period register, PR2, on each clock cycle. When the two values match, the comparator generates a match signal as the timer output. This signal also resets the value of TMR2 to 00h on the next cycle and drives the output counter/ postscaler. (See Section 15.2 "Timer2 Interrupt".)

The TMR2 and PR2 registers are both directly readable and writable. The TMR2 register is cleared on any device Reset, while the PR2 register initializes at FFh. Both the prescaler and postscaler counters are cleared on the following events:

- A write to the TMR2 register
- · A write to the T2CON register
- Any device Reset Power-on Reset (POR), MCLR Reset, Watchdog Timer Reset (WDTR) or Brown-out Reset (BOR)

TMR2 is not cleared when T2CON is written.

Note: The CCP and ECCP modules use Timers, 1 through 8, for some modes. The assignment of a particular timer to a CCP/ECCP module is determined by the Timer to CCP enable bits in the CCPTMRSx registers. For more details, see Register 20-2, Register 19-2 and Register 19-3.

#### REGISTER 15-1: T2CON: TIMER2 CONTROL REGISTER

U-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0
—	T2OUTPS3	T2OUTPS2	T2OUTPS1	T2OUTPS0	TMR2ON	T2CKPS1	T2CKPS0
bit 7							bit 0

Legend:			
R = Readable bit	W = Writable bit	U = Unimplemented bi	it, read as '0'
-n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown

bit 7 bit 6-3	Unimplemented: Read as '0' T2OUTPS<3:0>: Timer2 Output Postscale Select bits 0000 = 1:1 Postscale 0001 = 1:2 Postscale
	• • 1111 = 1:16 Postscale
bit 2	<b>TMR2ON:</b> Timer2 On bit 1 = Timer2 is on 0 = Timer2 is off
bit 1-0	<b>T2CKPS&lt;1:0&gt;:</b> Timer2 Clock Prescale Select bits 00 = Prescaler is 1 01 = Prescaler is 4 1x = Prescaler is 16

### 15.2 Timer2 Interrupt

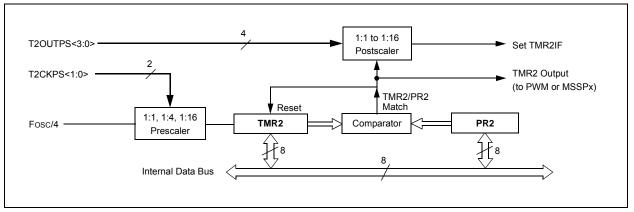
Timer2 can also generate an optional device interrupt. The Timer2 output signal (TMR2 to PR2 match) provides the input for the 4-bit output counter/postscaler. This counter generates the TMR2 match interrupt flag, which is latched in TMR2IF (PIR1<1>). The interrupt is enabled by setting the TMR2 Match Interrupt Enable bit, TMR2IE (PIE1<1>).

A range of 16 postscaler options (from 1:1 through 1:16 inclusive) can be selected with the postscaler control bits, T2OUTPS<3:0> (T2CON<6:3>).

# 15.3 Timer2 Output

The unscaled output of TMR2 is available primarily to the ECCP modules, where it is used as a time base for operations in PWM mode.

Timer2 can optionally be used as the shift clock source for the MSSP modules operating in SPI mode. Additional information is provided in Section 21.0 "Master Synchronous Serial Port (MSSP) Module".



# FIGURE 15-1: TIMER2 BLOCK DIAGRAM

#### TABLE 15-1: REGISTERS ASSOCIATED WITH TIMER2 AS A TIMER/COUNTER

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
INTCON	GIE/GIEH	PEIE/GIEL	TMR0IE	INT0IE	RBIE	TMR0IF	INT0IF	RBIF
PIR1	PSPIF	ADIF	RC1IF	TX1IF	SSP1IF	TMR1GIF	TMR2IF	TMR1IF
PIE1	PSPIE	ADIE	RC1IE	TX1IE	SSP1IE	TMR1GIE	TMR2IE	TMR1IE
IPR1	PSPIP	ADIP	RC1IP	TX1IP	SSP1IP	TMR1GIP	TMR2IP	TMR1IP
TMR2	Timer2 Register							
T2CON	—	T2OUTPS3	T2OUTPS2	T2OUTPS1	T2OUTPS0	TMR2ON	T2CKPS1	T2CKPS0
PR2	Timer2 Period Register							
PMD1	PSPMD	CTMUMD	RTCCMD	TMR4MD	TMR3MD	TMR2MD	TMR1MD	EMBDM

**Legend:** — = unimplemented, read as '0'. Shaded cells are not used by the Timer2 module.

# 16.0 TIMER3/5/7 MODULES

The Timer3/5/7 timer/counter modules incorporate these features:

- Software-selectable operation as a 16-bit timer or counter
- Readable and writable eight-bit registers (TMRxH and TMRxL)
- Selectable clock source (internal or external) with device clock or SOSC oscillator internal options
- Interrupt-on-overflow
- · Module Reset on ECCP Special Event Trigger

Timer7 is unimplemented for devices with a program memory of 32 Kbytes (PIC18FX5K22).

**Note:** Throughout this section, generic references are used for register and bit names that are the same – except for an 'x' variable that indicates the item's association with the Timer3, Timer5 or Timer7 module. For example, the control register is named TxCON and refers to T3CON, T5CON and T7CON.

A simplified block diagram of the Timer3/5/7 module is shown in Figure 16-1.

The Timer3/5/7 module is controlled through the TxCON register (Register 16-1). It also selects the clock source options for the ECCP modules. (For more information, see Section 20.1.1 "ECCP Module and Timer Resources".)

The Fosc clock source should not be used with the ECCP capture/compare features. If the timer will be used with the capture or compare features, always select one of the other timer clocking options.

R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0
TMRxCS1	TMRxCS0	TxCKPS1	TxCKPS0	SOSCEN	TxSYNC	RD16	TMRxON
bit 7							bit 0
Legend:							
R = Readable	e bit	W = Writable	bit	-	nented bit, read	d as '0'	
-n = Value at POR		'1' = Bit is set		'0' = Bit is clea	ared	x = Bit is unki	nown
bit 7-6	10 = Timer1 c SOSCEN = 02 External clock SOSCEN = 12 Depending on	from the T1Cl : the SOSCSEL	pends on the S KI pin (on the ri	SOSCEN bit:	or on the SOS	CI/SOSCO pine	s or an external
	00 = Timerx c	lock source is t lock source is t	the instruction	clock (Fosc/4)			
bit 5-4	<b>TxCKPS&lt;1:0&gt;</b> : Timerx Input Clock Prescale Select bits 11 = 1:8 Prescale value 10 = 1:4 Prescale value 01 = 1:2 Prescale value 00 = 1:1 Prescale value						
bit 3	<b>SOSCEN</b> : SOSC Oscillator Enable bit 1 = SOSC/SCLKI are enabled for Timerx (based on the SOSCSEL fuses) 0 = SOSC/SCLKI are disabled for Timerx and TxCKI is enabled						
bit 2	(Not usable if <u>When TMRxC</u> 1 = Do not syn 0 = Synchroni <u>When TMRxC</u>	the device cloo CS<1:0> = 10: nchronize exter ize external clo	ck comes from rnal clock input ck input				
bit 1	<b>RD16:</b> 16-Bit 1 = Enables re	Read/Write Mc egister read/wr	de Enable bit ite of Timerx in	one 16-bit ope two eight-bit o			
bit 0	TMRxON: Tin 1 = Enables T 0 = Stops Tim	imerx					

#### **REGISTER 16-1:** TxCON: TIMERx CONTROL REGISTER

**Note 1:** The Fosc clock source should not be selected if the timer will be used with the ECCP capture/compare features.

### 16.1 Timer3/5/7 Gate Control Register

The Timer3/5/7 Gate Control register (TxGCON), provided in Register 14-2, is used to control the Timerx gate.

# REGISTER 16-2: TxGCON: TIMERx GATE CONTROL REGISTER<sup>(1)</sup>

R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R-x	R/W-0	R/W-0
TMRxGE	TxGPOL	TxGTM	TxGSPM	TxGGO/TxDONE	TxGVAL	TxGSS1	TxGSS0
bit 7							bit 0

Legend:			
R = Readable bit	W = Writable bit	U = Unimplemented bit, re	ad as '0'
-n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown

bit 7	TMRxGE: Timerx Gate Enable bit
	<u>If TMRxON = 0:</u>
	This bit is ignored.
	<u>If TMRxON = 1:</u>
	1 = Timerx counting is controlled by the Timerx gate function
	0 = Timerx counts regardless of Timerx gate function
bit 6	TxGPOL: Timerx Gate Polarity bit
	<ul> <li>1 = Timerx gate is active-high (Timerx counts when gate is high)</li> <li>0 = Timerx gate is active-low (Timerx counts when gate is low)</li> </ul>
bit 5	TxGTM: Timerx Gate Toggle Mode bit
	1 = Timerx Gate Toggle mode is enabled.
	0 = Timerx Gate Toggle mode is disabled and toggle flip-flop is cleared
	Timerx gate flip-flop toggles on every rising edge.
bit 4	TxGSPM: Timerx Gate Single Pulse Mode bit
	<ul> <li>1 = Timerx Gate Single Pulse mode is enabled and is controlling Timerx gate</li> <li>0 = Timerx Gate Single Pulse mode is disabled</li> </ul>
bit 3	TxGGO/TxDONE: Timerx Gate Single Pulse Acquisition Status bit
	1 = Timerx gate single pulse acquisition is ready, waiting for an edge
	0 = Timerx gate single pulse acquisition has completed or has not been started
	This bit is automatically cleared when TxGSPM is cleared.
bit 2	TxGVAL: Timerx Gate Current State bit
	Indicates the current state of the Timerx gate that could be provided to TMRxH:TMRxL. Unaffected by the Timerx Gate Enable (TMRxGE) bit.
bit 1-0	TxGSS<1:0>: Timerx Gate Source Select bits
	11 = Comparator 2 output
	10 = Comparator 1 output
	01 = TMR(x+1) to match PR(x+1) output <sup>(2)</sup>
	00 = Timer1 gate pin The Watchdog Timer oscillator is turned on if TMRxGE = 1, regardless of the state of TMRxON.
Note 1:	Programming the TxGCON prior to TxCON is recommended

2: Timer(x+1) will be Timer4/6/8 for Timerx (Timer3/5/7), respectively.

#### REGISTER 16-3: OSCCON2: OSCILLATOR CONTROL REGISTER 2

U-0	R-0	U-0	U-0	R/W-0	U-0	R-x	R/W-0
_	SOSCRUN	—		SOSCGO	_	MFIOFS	MFIOSEL
bit 7							bit 0
Legend:							
R = Readable bit		W = Writable	oit	U = Unimplemented bit, read as '0'			
-n = Value at POR		'1' = Bit is set		'0' = Bit is cleared		x = Bit is unknown	
bit 7	Unimplemented: Read as '0'						
bit 6	SOSCRUN: SOSC Run Status bit						
	1 = System clock comes from a secondary SOSC						
	0 = System clock comes from an oscillator other than SOSC						
bit 5-4	Unimplemented: Read as '0'						
bit 3	SOSCGO: Oscillator Start Control bit						
	1 = Oscillator is running even if no other sources are requesting it						
	<ul> <li>0 = Oscillator is shut off if no other sources are requesting it (When the SOSC is selected to run from a digital clock input, rather than an external crystal, this bit has no effect.)</li> </ul>						
bit 2	Unimplemented: Read as '0'						
bit 1	MFIOFS: MF-INTOSC Frequency Stable bit						
	1 = MF-INTOSC is stable						
	0 = MF-INTOSC is not stable						
bit 0	MFIOSEL: MF-INTOSC Select bit						
	1 = MF-INTOSC is used in place of HF-INTOSC frequencies of 500 kHz, 250 kHz and 31.25 kHz						
	0 = MF-INTC	SC is not used					

The operating mode is determined by the clock select

bits, TMRxCSx (TxCON<7:6>). When the TMRxCSx bits

are cleared (= 00), Timer3/5/7 increments on every inter-

nal instruction cycle (Fosc/4). When TMRxCSx = 01, the

Timer3/5/7 clock source is the system clock (Fosc), and when it is '10', Timer3/5/7 works as a counter from the

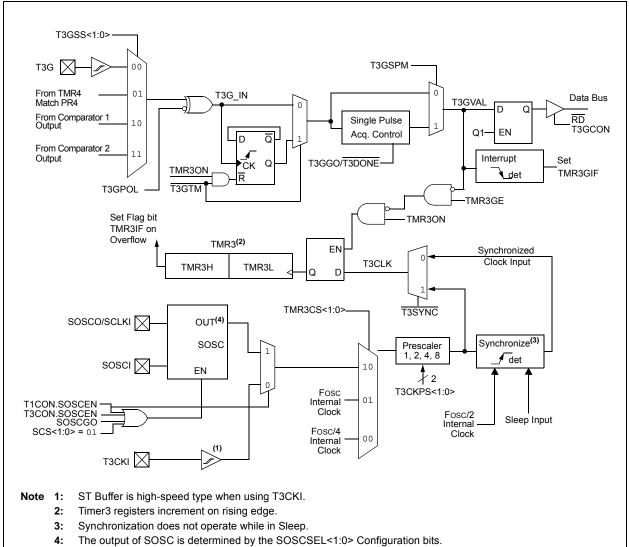
external clock from the TxCKI pin (on the rising edge after

the first falling edge) or the SOSC oscillator.

# 16.2 Timer3/5/7 Operation

Timer3, Timer5 and Timer7 can operate in these modes:

- Timer
- Synchronous Counter
- Asynchronous Counter
- · Timer with Gated Control



#### FIGURE 16-1: TIMER3/5/7 BLOCK DIAGRAM

### 16.3 Timer3/5/7 16-Bit Read/Write Mode

Timer3/5/7 can be configured for 16-bit reads and writes (see Figure 16.3). When the RD16 control bit (TxCON<1>) is set, the address for TMRxH is mapped to a buffer register for the high byte of Timer3/5/7. A read from TMRxL will load the contents of the high byte of Timer3/5/7 into the Timerx High Byte Buffer register. This provides users with the ability to accurately read all 16 bits of Timer3/5/7 without having to determine whether a read of the high byte, followed by a read of the low byte, has become invalid due to a rollover between reads.

A write to the high byte of Timer3/5/7 must also take place through the TMRxH Buffer register. The Timer3/5/7 high byte is updated with the contents of TMRxH when a write occurs to TMRxL. This allows users to write all 16 bits to both the high and low bytes of Timer3/5/7 at once.

The high byte of Timer3/5/7 is not directly readable or writable in this mode. All reads and writes must take place through the Timerx High Byte Buffer register.

Writes to TMRxH do not clear the Timer3/5/7 prescaler. The prescaler is only cleared on writes to TMRxL.

### 16.4 Using the SOSC Oscillator as the Timer3/5/7 Clock Source

The SOSC internal oscillator may be used as the clock source for Timer3/5/7. The SOSC oscillator is enabled by any peripheral that requests it. There are eight ways the SOSC can be enabled: if the SOSC is selected as the source by any of the odd timers, which is done by each respective SOSCEN bit (TxCON<3>), if the SOSC is selected as the RTCC source by the RTCOSC Configuration bit (CONFIG3L<1>), if the SOSC is selected as the CPU clock source by the SCS bits (OSCCON<1:0>) or if the SOSCGO bit is set (OSCCON2<3>). The SOSCGO bit is used to warm up the SOSC so that it is ready before any peripheral requests it. To use it as the Timer3/5/7 clock source, the TMRxCS bit must also be set. As previously noted, this also configures Timer3/5/7 to increment on every rising edge of the oscillator source.

The SOSC oscillator is described in **Section 14.5** "SOSC Oscillator".

#### 16.5 Timer3/5/7 Gates

Timer3/5/7 can be configured to count freely or the count can be enabled and disabled using the Timer3/5/7 gate circuitry. This is also referred to as the Timer3/5/7 gate count enable.

The Timer3/5/7 gate can also be driven by multiple selectable sources.

#### 16.5.1 TIMER3/5/7 GATE COUNT ENABLE

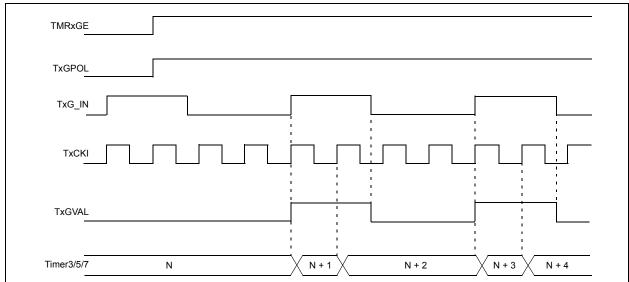
The Timerx Gate Enable mode is enabled by setting the TMRxGE bit (TxGCON<7>). The polarity of the Timerx Gate Enable mode is configured using the TxGPOL bit (TxGCON<6>).

When Timerx Gate Enable mode is enabled, Timer3/5/7 will increment on the rising edge of the Timer3/5/7 clock source. When Timerx Gate Enable mode is disabled, no incrementing will occur and Timer3/5/7 will hold the current count. See Figure 16-2 for timing details.

### TABLE 16-1: TIMER3/5/7 GATE ENABLE SELECTIONS

TxCLK <sup>(†)</sup>	TxGPOL (TxGCON<6>)	TxG Pin	Timerx Operation
1	0	0	Counts
$\uparrow$	0	1	Holds Count
$\uparrow$	1	0	Holds Count
$\uparrow$	1	1	Counts

† The clock on which TMR3/5/7 is running. For more information, see TxCLK in Figure 16-1.



#### FIGURE 16-2: TIMER3/5/7 GATE COUNT ENABLE MODE

#### 16.5.2 TIMER3/5/7 GATE SOURCE SELECTION

The Timer3/5/7 gate source can be selected from one of four different sources. Source selection is controlled by the TxGSS<1:0> bits (TxGCON<1:0>). The polarity for each available source is also selectable and is controlled by the TxGPOL bit (TxGCON <6>).

TABLE 16-2:	<b>TIMER3/5/7 GATE SOURCES</b>

TxGSS<1:0>	Timerx Gate Source
00	Timerx Gate Pin
01	TMR(x+1) to Match PR(x+1) (TMR(x+1) increments to match PR(x+1))
10	Comparator 1 Output (comparator logic high output)
11	Comparator 2 Output (comparator logic high output)

16.5.2.1 TxG Pin Gate Operation

The TxG pin is one source for Timer3/5/7 gate control. It can be used to supply an external source to the Timerx gate circuitry.

#### 16.5.2.2 Timer4/6/8 Match Gate Operation

The TMR(x+1) register will increment until it matches the value in the PR(x+1) register. On the very next increment cycle, TMR2 will be reset to 00h. When this Reset occurs, a low-to-high pulse will automatically be generated and internally supplied to the Timerx gate circuitry. The pulse will remain high for one instruction cycle and will return back to a low state until the next match.

Depending on TxGPOL, Timerx increments differently when TMR(x+1) matches PR(x+1). When TxGPOL = 1, Timerx increments for a single instruction

cycle following a TMR(x+1) match with PR(x+1). When TxGPOL = 0, Timerx increments continuously, except for the cycle following the match, when the gate signal goes from low-to-high.

#### 16.5.2.3 Comparator 1 Output Gate Operation

The output of Comparator 1 can be internally supplied to the Timerx gate circuitry. After setting up Comparator 1 with the CM1CON register, Timerx will increment depending on the transitions of the CMP1OUT (CMSTAT<5>) bit.

#### 16.5.2.4 Comparator 2 Output Gate Operation

The output of Comparator 2 can be internally supplied to the Timerx gate circuitry. After setting up Comparator 2 with the CM2CON register, Timerx will increment depending on the transitions of the CMP2OUT (CMSTAT<6>) bit.

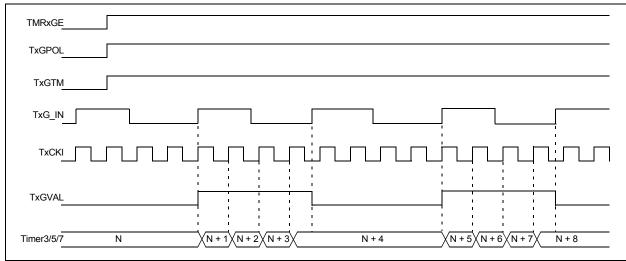
#### 16.5.3 TIMER3/5/7 GATE TOGGLE MODE

When Timer3/5/7 Gate Toggle mode is enabled, it is possible to measure the full cycle length of a Timer3/5/7 gate signal, as opposed to the duration of a single level pulse.

The Timerx gate source is routed through a flip-flop that changes state on every incrementing edge of the signal. (For timing details, see Figure 16-3.)

The TxGVAL bit will indicate when the Toggled mode is active and the timer is counting.

Timer3/5/7 Gate Toggle mode is enabled by setting the TxGTM bit (TxGCON<5>). When the TxGTM bit is cleared, the flip-flop is cleared and held clear. This is necessary in order to control which edge is measured.



#### FIGURE 16-3: TIMER3/5/7 GATE TOGGLE MODE

### 16.5.4 TIMER3/5/7 GATE SINGLE PULSE MODE

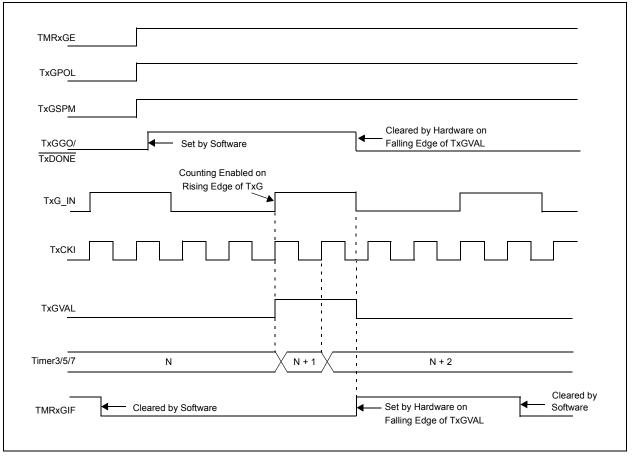
When Timer3/5/7 Gate Single Pulse mode is enabled, it is possible to capture a single pulse gate event. Timer3/5/7 Gate Single Pulse mode is first enabled by setting the TxGSPM bit (TxGCON<4>). Next, the TxGGO/TxDONE bit (TxGCON<3>) must be set.

The Timer3/5/7 will be fully enabled on the next incrementing edge. On the next trailing edge of the pulse, the TxGGO/TxDONE bit will automatically be cleared. No other gate events will <u>be allowed</u> to increment Timer3/5/7 until the TxGGO/TxDONE bit is once again set in software.

Clearing the TxGSPM bit also will clear the TxGGO/ TxDONE bit. (For timing details, see Figure 16-4.)

Simultaneously enabling the Toggle mode and the Single Pulse mode will permit both sections to work together. This allows the cycle times on the Timer3/5/7 gate source to be measured. (For timing details, see Figure 16-5.)

#### FIGURE 16-4: TIMER3/5/7 GATE SINGLE PULSE MODE



TMRxGE						
TxGPOL						
TxGSPM						
TxGTM						
TxGGO/ TxDONE	Set by Software					Hardware on ge of TxGVAL
TxG_IN	Rising Edge of TxG				]	1
ТхСКІ						
TxGVAL					; 1;	
imer3/5/7	Ν	N+1	N + 2	N + 3	N + 4	
	Cleared by Software		Set by H Falling Edge	lardware on	, , ,	Cleared by Software

#### FIGURE 16-5: TIMER3/5/7 GATE SINGLE PULSE AND TOGGLE COMBINED MODE

#### 16.5.5 TIMER3/5/7 GATE VALUE STATUS

When Timer3/5/7 gate value status is utilized, it is possible to read the most current level of the gate control value. The value is stored in the TxGVAL bit (TxGCON<2>). The TxGVAL bit is valid even when the Timer3/5/7 gate is not enabled (TMRxGE bit is cleared).

#### 16.5.6 TIMER3/5/7 GATE EVENT INTERRUPT

When the Timer3/5/7 gate event interrupt is enabled, it is possible to generate an interrupt upon the completion of a gate event. When the falling edge of TxGVAL occurs, the TMRxGIF flag bit in the PIRx register will be set. If the TMRxGIE bit in the PIEx register is set, then an interrupt will be recognized.

The TMRxGIF flag bit operates even when the Timer3/5/7 gate is not enabled (TMRxGE bit is cleared).

#### 16.6 Timer3/5/7 Interrupt

The TMRx register pair (TMRxH:TMRxL) increments from 0000h to FFFFh and overflows to 0000h. The Timerx interrupt, if enabled, is generated on overflow and is latched in the interrupt flag bit, TMRxIF. Table 16-3 gives each module's flag bit.

#### TABLE 16-3: TIMER3/5/7 INTERRUPT FLAG BITS

Timer Module	Flag Bit
3	PIR2<1>
5	PIR5<1>
7	PIR5<3>

This interrupt can be enabled or disabled by setting or clearing the TMRxIE bit, respectively. Table 16-4 gives each module's enable bit.

### TABLE 16-4:TIMER3/5/7 INTERRUPTENABLE BITS

Timer Module	Flag Bit
3	PIE2<1>
5	PIE5<1>
7	PIE5<3>

#### 16.7 Resetting Timer3/5/7 Using the ECCP Special Event Trigger

If the ECCP modules are configured to use Timerx and to generate a Special Event Trigger in Compare mode (CCPxM<3:0> = 1011), this signal will reset Timerx. The trigger from ECCP2 will also start an A/D conversion if the A/D module is enabled. (For more information, see Section 20.3.4 "Special Event Trigger".)

The module must be configured as either a timer or synchronous counter to take advantage of this feature. When used this way, the CCPRxH:CCPRxL register pair effectively becomes a Period register for Timerx.

If Timerx is running in Asynchronous Counter mode, the Reset operation may not work.

In the event that a write to Timerx coincides with a Special Event Trigger from an ECCP module, the write will take precedence.

Note:	The Special Event Triggers from the				
	ECCPx module will only clear the TMR3				
	register's content, but not set the TMR3IF				
	interrupt flag bit (PIR1<0>).				

Note: The CCP and ECCP modules use Timers, 1 through 8, for some modes. The assignment of a particular timer to a CCP/ECCP module is determined by the Timer to CCP enable bits in the CCPTMRSx registers. For more details, see Register 19-2, Register 19-3 and Register 20-2

IADLE 10-	ABLE 16-5: REGISTERS ASSOCIATED WITH TIMER3/5/7 AS A TIMER/COUNTER							
Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
INTCON	GIE/GIEH	PEIE/GIEL	TMR0IE	<b>INT0IE</b>	RBIE	TMR0IF	INT0IF	RBIF
PIR5	TMR7GIF <sup>(1)</sup>	TMR12IF <sup>(1)</sup>	TMR10IF <sup>(1)</sup>	TMR8IF	TMR7IF <sup>(1)</sup>	TMR6IF	TMR5IF	TMR4IF
IPR5	TMR7GIP <sup>(1)</sup>	TMR12IP <sup>(1)</sup>	TMR10IP <sup>(1)</sup>	TMR8IP	TMR7IP <sup>(1)</sup>	TMR6IP	TMR5IP	TMR4IP
PIE5	TMR7GIE <sup>(1)</sup>	TMR12IE <sup>(1)</sup>	TMR10IE <sup>(1)</sup>	TMR8IE	TMR7IE <sup>(1)</sup>	TMR6IE	TMR5IE	TMR4IE
PIE3	TMR5GIE	_	RC2IE	TX2IE	CTMUIE	CCP2IE	CCP1IE	RTCCIE
IPR3	TMR5GIP	_	RC2IP	TX2IP	CTMUIP	CCP2IP	CCP1IP	RTCCIP
PIR3	TMR5GIF	_	RC2IF	TX2IF	CTMUIF	CCP2IF	CCP1IF	RTCCIF
PIE2	OSCFIE	_	SSP2IE	BCL2IE	BCL1IE	HLVDIE	TMR3IE	TMR3GIE
PIR2	OSCFIF	_	SSP2IF	BCL2IF	BCL1IF	HLVDIF	TMR3IF	TMR3GIF
IPR2	OSCFIP	_	SSP2IP	BCL2IP	BCL1IP	HLVDIP	TMR3IP	TMR3GIP
TMR3H	Timer3 Regist	er High Byte						
TMR3L	Timer3 Regist	er Low Byte						
T3GCON	TMR3GE	T3GPOL	T3GTM	T3GSPM	T3GGO/ T3DONE	T3GVAL	T3GSS1	T3GSS0
T3CON	TMR3CS1	TMR3CS0	T3CKPS1	T3CKPS0	SOSCEN	T3SYNC	RD16	TMR3ON
TMR5H	Timer5 Regist	er High Byte						
TMR5L	Timer5 Regist	er Low Byte						
T5GCON	TMR5GE	T5GPOL	T5GTM	T5GSPM	T5GGO/ T5DONE	T5GVAL	T5GSS1	T5GSS0
T5CON	TMR5CS1	TMR5CS0	T5CKPS1	T5CKPS0	SOSCEN	T5SYNC	RD16	TMR5ON
TMR7H <sup>(1)</sup>	Timer7 Regist	er High Byte						
TMR7L <sup>(1)</sup>	Timer7 Register Low Byte							
T7GCON <sup>(1)</sup>	TMR7GE	T7GPOL	T7GTM	T7GSPM	T7GGO/ T7DONE	T7GVAL	T7GSS1	T7GSS0
T7CON <sup>(1)</sup>	TMR7CS1	TMR7CS0	T7CKPS1	T7CKPS0	SOSCEN	T7SYNC	RD16	TMR7ON
OSCCON2	—	SOSCRUN	—	—	SOSCGO	—	MFIOFS	MFIOSEL
PMD1	PSPMD	CTMUMD	RTCCMD	TMR4MD	TMR3MD	TMR2MD	TMR1MD	EMBMD
PMD2	TMR10MD <sup>(1)</sup>	TMR8MD	TMR7MD <sup>(1)</sup>	TMR6MD	TMR5MD	CMP3MD	CMP2MD	CMP2MD

#### TABLE 16-5: REGISTERS ASSOCIATED WITH TIMER3/5/7 AS A TIMER/COUNTER

**Legend:** — = unimplemented, read as '0'. Shaded cells are not used by the Timer3/5/7 module.

Note 1: Unimplemented on devices with a program memory of 32 Kbytes (PIC18FX5K22).

### 17.0 TIMER4/6/8/10/12 MODULES

The Timer4/6/8/10/12 timer modules have the following features:

- Eight-bit Timer register (TMRx)
- Eight-bit Period register (PRx)
- Readable and writable (all registers)
- Software programmable prescaler (1:1, 1:4, 1:16)
- Software programmable postscaler (1:1 to 1:16)
- Interrupt on TMRx match of PRx

Timer10 and Timer12 are unimplemented for devices with a program memory of 32 Kbytes (PIC18FX5K22).

Note:	Throughout this section, generic references
	are used for register and bit names that are the
	same, except for an 'x' variable that indicates
	the item's association with the Timer4, Timer6,
	Timer8, Timer10 or Timer12 module. For
	example, the control register is named TxCON
	and refers to T4CON, T6CON, T8CON,
	T10CON and T12CON.

The Timer4/6/8/10/12 modules have a control register, which is shown in Register 17-1. Timer4/6/8/10/12 can be shut off by clearing control bit, TMRxON (TxCON<2>), to minimize power consumption. The prescaler and post-scaler selection of Timer4/6/8/10/12 are also controlled by this register. Figure 17-1 is a simplified block diagram of the Timer4/6/8/10/12 modules.

#### 17.1 Timer4/6/8/10/12 Operation

Timer4/6/8/10/12 can be used as the PWM time base for the PWM mode of the ECCP modules. The TMRx registers are readable and writable, and are cleared on any device Reset. The input clock (Fosc/4) has a prescale option of 1:1, 1:4 or 1:16, selected by control bits, TxCKPS<1:0> (TxCON<1:0>). The match output of TMRx goes through a four-bit postscaler (that gives a 1:1 to 1:16 inclusive scaling) to generate a TMRx interrupt, latched in the flag bit, TMRxIF. Table 17-1 shows each module's flag bit.

Timer Module	Flag Bit PIR5 <x></x>	Timer Module	Flag Bit PIR5 <x></x>
4	0	10	5
6	2	12	6
8	4		

TABLE 17-1: TIMER4/6/8/10/12 FLAG BITS

The interrupt can be enabled or disabled by setting or clearing the Timerx Interrupt Enable bit (TMRxIE), shown in Table 17-2.

TABLE 17-2:	TIMER4/6/8/10/12 INTERRUPT
	ENABLE BITS

Timer Module	Flag Bit PIE5 <x></x>	Timer Module	Flag Bit PIE5 <x></x>
4	0	10	5
6	2	12	6
8	4		

The prescaler and postscaler counters are cleared when any of the following occurs:

- A write to the TMRx register
- A write to the TxCON register
- <u>Any device Reset Power-on Reset (POR),</u> MCLR Reset, Watchdog Timer Reset (WDTR) or Brown-out Reset (BOR)

A TMRx is not cleared when a TxCON is written.

Note: The CCP and ECCP modules use Timers, 1 through 8, for some modes. The assignment of a particular timer to a CCP/ECCP module is determined by the Timer to CCP enable bits in the CCPTMRSx registers. For more details, see Register 19-2, Register 19-3 and Register 20-2.

#### REGISTER 17-1: TxCON: TIMERx CONTROL REGISTER

U-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0
—	TxOUTPS3	TxOUTPS2	TxOUTPS1	TxOUTPS0	TMRxON	TxCKPS1	TxCKPS0
bit 7							bit 0
Legend:							

R = Readable bit	W = Writable bit	U = Unimplemented bit, rea	d as '0'
-n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown

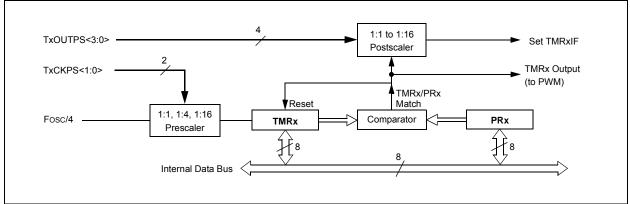
bit 7 bit 6-3	Unimplemented: Read as '0' TxOUTPS<3:0>: Timerx Output Postscale Select bits 0000 = 1:1 Postscale 0001 = 1:2 Postscale
	•
	1111 = 1:16 Postscale
bit 2	TMRxON: Timerx On bit 1 = Timerx is on 0 = Timerx is off
bit 1-0	<b>TxCKPS&lt;1:0&gt;:</b> Timerx Clock Prescale Select bits 00 = Prescaler is 1 01 = Prescaler is 4 1x = Prescaler is 16

#### 17.2 Timer4/6/8/10/12 Interrupt

The Timer4/6/8/10/12 modules have eight-bit Period registers, PRx, that are both readable and writable. Timer4/6/8/10/12 increment from 00h until they match PR4/6/8/10/12 and then reset to 00h on the next increment cycle. The PRx registers are initialized to FFh upon Reset.

#### 17.3 Output of TMRx

The outputs of TMRx (before the postscaler) are used only as a PWM time base for the ECCP modules. They are not used as baud rate clocks for the MSSP modules as is the Timer2 output.



#### FIGURE 17-1: TIMER4 BLOCK DIAGRAM

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
INTCON	GIE/GIEH	PEIE/GIEL	TMR0IE	INT0IE	RBIE	TMR0IF	INT0IF	RBIF
IPR5	TMR7GIP <sup>(1)</sup>	TMR12IP <sup>(1)</sup>	TMR10IP <sup>(1)</sup>	TMR8IP	TMR7IP <sup>(1)</sup>	TMR6IP	TMR5IP	TMR4IP
PIR5	TMR7GIF <sup>(1)</sup>	TMR12IF <sup>(1)</sup>	TMR10IF <sup>(1)</sup>	TMR8IF	TMR7IF <sup>(1)</sup>	TMR6IF	TMR5IF	TMR4IF
PIE5	TMR7GIE <sup>(1)</sup>	TMR12IE <sup>(1)</sup>	TMR10IE <sup>(1)</sup>	TMR8IE	TMR7IE <sup>(1)</sup>	TMR6IE	TMR5IE	TMR4IE
TMR4	Timer4 Regis	ter						
T4CON	_	T4OUTPS3	T4OUTPS2	T4OUTPS1	T4OUTPS0	TMR4ON	T4CKPS1	T4CKPS0
PR4	Timer4 Period	d Register						
TMR6	Timer6 Regis	ter						
T6CON	—	T6OUTPS3	T6OUTPS2	T6OUTPS1	T6OUTPS0	TMR6ON	T6CKPS1	T6CKPS0
PR6	Timer6 Period	d Register						
TMR8	Timer8 Regis	ter						
T8CON	—	T8OUTPS3	T8OUTPS2	T8OUTPS1	T8OUTPS0	TMR8ON	T8CKPS1	T8CKPS0
PR8	Timer8 Period	d Register						
TMR10 <sup>(1)</sup>	Timer10 Regi	ster						
T10CON <sup>(1)</sup>	—	T10OUTPS3	T10OUTPS2	T10OUTPS1	T10OUTPS0	TMR10ON	T10CKPS1	T10CKPS0
PR10 <sup>(1)</sup>	Timer10 Peric	od Register						
TMR12 <sup>(1)</sup>	Timer12 Regi	ster						
T12CON <sup>(1)</sup>	—	T12OUTPS3	T12OUTPS2	T12OUTPS1	T12OUTPS0	TMR12ON	T12CKPS1	T12CKPS0
PR12 <sup>(1)</sup>	Timer12 Perio	od Register						
PMD1	PSPMD	CTMUMD	RTCCMD	TMR4MD	TMR3MD	TMR2MD	TMR1MD	EMBMD
PMD2	TMR10MD <sup>(1)</sup>		TMR7MD <sup>(1)</sup>	TMR6MD	TMR5MD	CMP3MD	CMP2MD	CMP2MD
PMD3	CCP10MD <sup>(1)</sup>	CCP9MD <sup>(1)</sup>	CCP8MD	CCP7MD	CCP6MD	CCP5MD	CCP4MD	TMR12MD <sup>(1)</sup>
					a a d las s Ala a Tis			

TABLE 17-3:	REGISTERS ASSOCIATED WITH TIMER4/6/8/10/12 AS A TIMER/COUNTER
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Legend: — = unimplemented, read as '0'. Shaded cells are not used by the Timer4/6/8/10/12 module.

**Note 1:** Unimplemented on devices with a program memory of 32 Kbytes (PIC18FX5K22).

NOTES:

### 18.0 REAL-TIME CLOCK AND CALENDAR (RTCC)

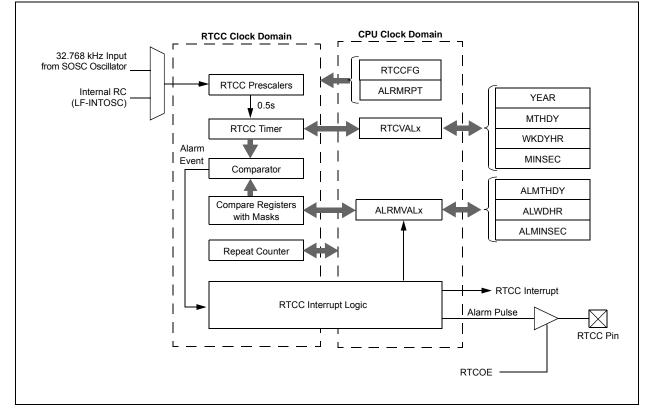
The key features of the Real-Time Clock and Calendar (RTCC) module are:

- Time: hours, minutes and seconds
- Twenty-four hour format (military time)
- Calendar: weekday, date, month and year
- · Alarm configurable
- Year range: 2000 to 2099
- · Leap year correction
- · BCD format for compact firmware
- · Optimized for low-power operation
- · User calibration with auto-adjust
- Calibration range: ±2.64 seconds error per month
- Requirements: external 32.768 kHz clock crystal
- Alarm pulse or seconds clock output on RTCC pin

The RTCC module is intended for applications where accurate time must be maintained for extended period with minimum to no intervention from the CPU. The module is optimized for low-power usage in order to provide extended battery life while keeping track of time.

The module is a 100-year clock and calendar with automatic leap year detection. The range of the clock is from 00:00:00 (midnight) on January 1, 2000 to 23:59:59 on December 31, 2099.

Hours are measured in 24-hour (military time) format. The clock provides a granularity of one second with half-second visibility to the user.



#### FIGURE 18-1: RTCC BLOCK DIAGRAM

### 18.1 RTCC MODULE REGISTERS

The RTCC module registers are divided into the following categories:

#### **RTCC Control Registers**

- RTCCFG
- RTCCAL
- PADCFG1
- ALRMCFG
- ALRMRPT

#### **RTCC Value Registers**

- RTCVALH
- RTCVALL

Both registers access the following registers:

- YEAR
- MONTH
- DAY
- WEEKDAY
- HOUR
- MINUTE
- SECOND

#### **Alarm Value Registers**

- ALRMVALH
- ALRMVALL
   Both registers access the following registers:
  - ALRMMNTH
  - ALRMDAY
  - ALRMWD
  - ALRMHR
  - ALRMMIN
  - ALRMSEC
- Note: The RTCVALH and RTCVALL registers can be accessed through RTCRPT<1:0> (RTCCFG<1:0>). ALRMVALH and ALRMVALL can be accessed through ALRMPTR<1:0> (ALRMCFG<1:0>).

#### 18.1.1 RTCC CONTROL REGISTERS

REGISTER		CFG: RTCC CO					
R/W-0	U-0	R/W-0	R-0	R-0	R/W-0	R/W-0	R/W-0
RTCEN <sup>(2</sup>	) _	RTCWREN <sup>(4)</sup>	RTCSYNC	HALFSEC <sup>(3)</sup>	RTCOE	RTCPTR1	RTCPTR0
bit 7							bit (
Legend:							
R = Readab	ole bit	W = Writable	bit	U = Unimplem	ented bit, rea	d as '0'	
-n = Value a	It POR	'1' = Bit is set		'0' = Bit is clea	ired	x = Bit is unki	nown
bit 7	RTCEN: RT	CC Enable bit <sup>(2)</sup>					
		module is enable module is disable					
bit 6	Unimpleme	ented: Read as '	)'				
bit 5	RTCWREN	RTCC Value Re	gisters Write	Enable bit <sup>(4)</sup>			
		LH and RTCVAL	•		•		
		LH and RTCVAL	-		-	en to by the use	r
bit 4		RTCC Value Re	-	-			
	results	LH, RTCVALL a in an invalid data	read. If the r				
		assumed to be v LH, RTCVALL a		PT registers ca	n be read wit	hout concern c	over a rollove
bit 3	HALFSEC:	Half-Second Sta	tus bit <sup>(3)</sup>				
		I half period of a Ilf period of a sec					
bit 2	RTCOE: RT	CC Output Enab	le bit				
		clock output is er					
		clock output is di					
bit 1-0		:0>: RTCC Value	-				
		ne correspondino he RTCPTR<1:0 '.					
	RTCVALH:						
	00 = Minute	-					
	01 = Weekd	lay					
	10 = Month 11 = Reserv	hav					
	RTCVALL:	/cu					
	00 = Second	ds					
	01 = Hours						
	10 <b>= Day</b>						
	11 <b>= Year</b>						
Note 1: 7	he RTCCFG reg	gister is only affe	cted by a PO	R.			
		CEN bit is only a					
3: 1	his bit is read-o	nly; it is cleared t	o '0' on a wri	te to the lower h	alf of the MIN	SEC register.	

#### PTCCEC, PTCCCCONFICURATION PECIETED(1)

4: RTCWREN can only be written with the unlock sequence (see Example 18-1).

Legend:							
bit 7							bit 0
CAL7	CAL6	CAL5	CAL4	CAL3	CAL2	CAL1	CAL0
R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0

#### REGISTER 18-2: RTCCAL: RTCC CALIBRATION REGISTER

Legena.			
R = Readable bit	W = Writable bit	U = Unimplemented bit, read	d as 'O'
-n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown

#### bit 7-0 CAL<7:0>: RTC Drift Calibration bits

.

01111111 = Maximum positive adjustment. Adds 508 RTC clock pulses every minute.

• 00000001 = Minimum positive adjustment. Adds four RTC clock pulses every minute. 00000000 = No adjustment
11111111 = Minimum negative adjustment. Subtracts four RTC clock pulses every minute.
•
10000000 = Maximum negative adjustment. Subtracts 512 RTC clock pulses every minute.

#### REGISTER 18-3: PADCFG1: PAD CONFIGURATION REGISTER

R/W-0	R/W-0	R/W-0	U-0	U-0	R/W-0	R/W-0	U-0
RDPU	REPU	RJPU <sup>(2)</sup>			RTSECSEL1 <sup>(1)</sup>	RTSECSEL0 <sup>(1)</sup>	_
bit 7							bit 0
Legend:							
R = Reada	able bit	W = Writable I	oit	U = Unimplen	nented bit, read	as '0'	
-n = Value	at POR	'1' = Bit is set		'0' = Bit is cle	ared	x = Bit is unknow	n
bit 7	RDPU: POI	RTD Pull-up Ena	ble bit				
		) pull-up resistor			oort latch values		
		RTD pull-up resis	tors are disab	led			
bit 6	REPU: POP	RTE Pull-up Enal	ole bit				
		E pull-up resistor			oort latch values		
		RTE pull-up resis		led			
bit 5		RTJ Pull-up Enab					
		pull-up resistors			ort latch values		
		RTJ pull-up resist			(4)		
bit 2-1	RTSECSEL	.<1:0>: RTCC Se	econds Clock	Output Select	bits <sup>(1)</sup>		
		ved; do not use					
					oin can be LF-IN	TOSC or SOSC, d	epending on
		COSC (CONFIC seconds clock is	,	0,			
		alarm pulse is s		•			
bit 0		ented: Read as '		F			
	•						
Note 1:	To enable the a	actual RTCC outp	out, the RTCO	E (RTCCFG<	2>) bit must be s	set.	

2: Unimplemented on 64-pin devices (PIC18F6XK22), read as '0'.

R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0			
ALRMEN	CHIME	AMASK3	AMASK2	AMASK1	AMASK0	ALRMPTR1	ALRMPTR0			
bit 7							bit 0			
Legend: R = Readable	hit	W = Writable	hit	II = I Inimpler	nented bit, rea	d as '0'				
-n = Value at		'1' = Bit is set		'0' = Bit is cle		x = Bit is unkr				
							lowin			
bit 7	ALRMEN: A	larm Enable bit								
		enabled (cleare	ed automatica	lly after an alar	m event when	ever ARPT<7:0	> = 00			
	0 = Alarm is	,								
bit 6	CHIME: Chir	me Enable bit								
		s enabled; ALRI s disabled; ALR				m 00h to FFh				
bit 5-2	AMASK<3:0>: Alarm Mask Configuration bits									
	0000 = Every half second									
	0001 = Every second									
	0010 = Every 10 seconds									
	0011 = Every minute									
	0100 = Every 10 minutes									
	0101 = Every hour									
	0110 = Once a day									
	0111 = Once a week 1000 = Once a month									
	1000 – Once a month 1001 = Once a year (except when configured for February 29 <sup>th</sup> , once every four years)									
	101x = Reserved - Do not use									
	11xx = Reserved – Do not use									
bit 1-0	ALRMPTR<1:0>: Alarm Value Register Window Pointer bits									
	Points to the corresponding Alarm Value registers when reading the ALRMVALH and ALRMVALL registers. The ALRMPTR<1:0> value decrements on every read or write of ALRMVALH until it reaches									
	'00'.									
	ALRMVALH:									
	00 = ALRMN 01 = ALRMV									
	10 = ALRMN									
	11 = Unimple									
	ALRMVALL:									
	00 = ALRMS									
	01 = ALRMH									
	10 = ALRME									

### REGISTER 18-4: ALRMCFG: ALARM CONFIGURATION REGISTER

| R/W-0 |
|-------|-------|-------|-------|-------|-------|-------|-------|
| ARPT7 | ARPT6 | ARPT5 | ARPT4 | ARPT3 | ARPT2 | ARPT1 | ARPT0 |
| bit 7 |       |       |       |       |       |       | bit 0 |

Legend:			
R = Readable bit	W = Writable bit	U = Unimplemented bit, read	l as '0'
-n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown

### bit 7-0 **ARPT<7:0>:** Alarm Repeat Counter Value bits

11111111 = Alarm will repeat 255 more times

. . 00000000 = Alarm will not repeat

The counter decrements on any alarm event. The counter is prevented from rolling over from 00h to FFh unless CHIME = 1.

#### 18.1.2 RTCVALH AND RTCVALL REGISTER MAPPINGS

#### REGISTER 18-6: RESERVED REGISTER

U-0	U-0	U-0	U-0	U-0	U-0	U-0	U-0
—	—	—	_	—	—	—	—
bit 7							bit 0

Legend:			
R = Readable bit	W = Writable bit	U = Unimplemented bit, rea	d as '0'
-n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown

bit 7-0 Unimplemented: Read as '0'

#### REGISTER 18-7: YEAR: YEAR VALUE REGISTER<sup>(1)</sup>

| R/W-x  |
|--------|--------|--------|--------|--------|--------|--------|--------|
| YRTEN3 | YRTEN2 | YRTEN1 | YRTEN0 | YRONE3 | YRONE2 | YRONE1 | YRONE0 |
| bit 7  |        |        |        |        |        |        | bit 0  |

Legend:			
R = Readable bit	W = Writable bit	U = Unimplemented bit, read	as '0'
-n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown

bit 7-4 YRTEN<3:0>: Binary Coded Decimal Value of Year's Tens Digit bits Contains a value from 0 to 9.

bit 3-0 **YRONE<3:0>:** Binary Coded Decimal Value of Year's Ones Digit bits Contains a value from 0 to 9.

**Note 1:** A write to the YEAR register is only allowed when RTCWREN = 1.

#### REGISTER 18-8: MONTH: MONTH VALUE REGISTER<sup>(1)</sup>

U-0	U-0	U-0	R/W-x	R/W-x	R/W-x	R/W-x	R/W-x
—	—	—	MTHTEN0	MTHONE3	MTHONE2	MTHONE1	MTHONE0
bit 7							bit 0

Legend:			
R = Readable bit	W = Writable bit	U = Unimplemented bit, read	d as '0'
-n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown

Unimplemented: Read as '0'
MTHTEN0: Binary Coded Decimal Value of Month's Tens Digit bit
Contains a value of 0 or 1.
MTHONE<3:0>: Binary Coded Decimal Value of Month's Ones Digit bits
Contains a value from 0 to 9.

**Note 1:** A write to this register is only allowed when RTCWREN = 1.

#### REGISTER 18-9: DAY: DAY VALUE REGISTER<sup>(1)</sup>

U-0	U-0	R/W-x	R/W-x	R/W-x	R/W-x	R/W-x	R/W-x
—	—	DAYTEN1	DAYTEN0	DAYONE3	DAYONE2	DAYONE1	DAYONE0
bit 7							bit 0

Legend:			
R = Readable bit	W = Writable bit	U = Unimplemented bit	t, read as '0'
-n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown

bit 7-6	Unimplemented: Read as '0'
bit 5-4	DAYTEN<1:0>: Binary Coded Decimal value of Day's Tens Digit bits
	Contains a value from 0 to 3.
bit 3-0	DAYONE<3:0>: Binary Coded Decimal Value of Day's Ones Digit bits
	Contains a value from 0 to 9.

**Note 1:** A write to this register is only allowed when RTCWREN = 1.

#### REGISTER 18-10: WEEKDAY: WEEKDAY VALUE REGISTER<sup>(1)</sup>

U-0	U-0	U-0	U-0	U-0	R/W-x	R/W-x	R/W-x
—	—			—	WDAY2	WDAY1	WDAY0
bit 7							bit 0

Legend:			
R = Readable bit	W = Writable bit	U = Unimplemented bit,	read as '0'
-n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown

bit 7-3 Unimplemented: Read as '0'

bit 2-0 **WDAY<2:0>:** Binary Coded Decimal Value of Weekday Digit bits Contains a value from 0 to 6.

**Note 1:** A write to this register is only allowed when RTCWREN = 1.

#### REGISTER 18-11: HOUR: HOUR VALUE REGISTER<sup>(1)</sup>

U-0	U-0	R/W-x	R/W-x	R/W-x	R/W-x	R/W-x	R/W-x
—	—	HRTEN1	HRTEN0	HRONE3	HRONE2	HRONE1	HRONE0
bit 7							bit 0

Legend:			
R = Readable bit	W = Writable bit	U = Unimplemented bit,	read as '0'
-n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown

bit 7-6 Unimplemented: Read as '0'

bit 5-4	HRTEN<1:0>: Binary Coded Decimal Value of Hour's Tens Digit bits
	Contains a value from 0 to 2.
bit 3-0	HRONE<3:0>: Binary Coded Decimal Value of Hour's Ones Digit bits
	Contains a value from 0 to 9.

**Note 1:** A write to this register is only allowed when RTCWREN = 1.

#### **REGISTER 18-12: MINUTE: MINUTE VALUE REGISTER**

U-0	R/W-x						
—	MINTEN2	MINTEN1	MINTEN0	MINONE3	MINONE2	MINONE1	MINONE0
bit 7							bit 0

Legend:			
R = Readable bit	W = Writable bit	U = Unimplemented bit	, read as '0'
-n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown

bit 7	Unimplemented: Read as '0'
bit 6-4	MINTEN<2:0>: Binary Coded Decimal Value of Minute's Tens Digit bits
	Contains a value from 0 to 5.
bit 3-0	MINONE<3:0>: Binary Coded Decimal Value of Minute's Ones Digit bits
	Contains a value from 0 to 9.

#### **REGISTER 18-13: SECOND: SECOND VALUE REGISTER**

U-0	R/W-x						
—	SECTEN2	SECTEN1	SECTEN0	SECONE3	SECONE2	SECONE1	SECONE0
bit 7							bit 0

Legend:			
R = Readable bit	W = Writable bit	U = Unimplemented bit,	, read as '0'
-n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown

bit 7	Unimplemented: Read as '0'
bit 6-4	SECTEN<2:0>: Binary Coded Decimal Value of Second's Tens Digit bits
	Contains a value from 0 to 5.
bit 3-0	<b>SECONE&lt;3:0&gt;:</b> Binary Coded Decimal Value of Second's Ones Digit bits Contains a value from 0 to 9.

#### 18.1.3 ALRMVALH AND ALRMVALL REGISTER MAPPINGS

### REGISTER 18-14: ALRMMNTH: ALARM MONTH VALUE REGISTER<sup>(1)</sup>

U-0	U-0	R/W-x	R/W-x	R/W-x	R/W-x	R/W-x
—	—	MTHTEN0	MTHONE3	MTHONE2	MTHONE1	MTHONE0
						bit 0
bit	W = Writable	bit	U = Unimpler	nented bit, read	as '0'	
-n = Value at POR '1' = Bit is set			'0' = Bit is cleared		x = Bit is unknown	
	-	OR '1' = Bit is set	bit W = Writable bit	bit W = Writable bit U = Unimpler OR '1' = Bit is set '0' = Bit is cle	bit W = Writable bit U = Unimplemented bit, read OR '1' = Bit is set '0' = Bit is cleared	bit W = Writable bit U = Unimplemented bit, read as '0' OR '1' = Bit is set '0' = Bit is cleared x = Bit is unkr

bit 7-5	Unimplemented: Read as '0'
bit 4	MTHTEN0: Binary Coded Decimal Value of Month's Tens Digit bits
	Contains a value of 0 or 1.
bit 3-0	MTHONE<3:0>: Binary Coded Decimal Value of Month's Ones Digit bits
	Contains a value from 0 to 9.

**Note 1:** A write to this register is only allowed when RTCWREN = 1.

### REGISTER 18-15: ALRMDAY: ALARM DAY VALUE REGISTER<sup>(1)</sup>

U-0	U-0	R/W-x	R/W-x	R/W-x	R/W-x	R/W-x	R/W-x
—	—	DAYTEN1	DAYTEN0	DAYONE3	DAYONE2	DAYONE1	DAYONE0
bit 7							bit 0

Legend:			
R = Readable bit	W = Writable bit	U = Unimplemented bit, read	d as '0'
-n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown

bit 7-6	Unimplemented: Read as '0'
bit 5-4	DAYTEN<1:0>: Binary Coded Decimal Value of Day's Tens Digit bits
	Contains a value from 0 to 3.
bit 3-0	DAYONE<3:0>: Binary Coded Decimal Value of Day's Ones Digit bits

Contains a value from 0 to 9.

**Note 1:** A write to this register is only allowed when RTCWREN = 1.

#### REGISTER 18-16: ALRMWD: ALARM WEEKDAY VALUE REGISTER<sup>(1)</sup>

-         -         -         WDAY2         WDAY1         WDAY0           bit 7             bit 0	U-0	U-0	U-0	U-0	U-0	R/W-x	R/W-x	R/W-x
bit 7 bit 0	—	—				WDAY2	WDAY1	WDAY0
	bit 7							bit 0

Legend:			
R = Readable bit	W = Writable bit	U = Unimplemented bit,	, read as '0'
-n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown

bit 7-3 Unimplemented: Read as '0'

bit 2-0 **WDAY<2:0>:** Binary Coded Decimal Value of Weekday Digit bits Contains a value from 0 to 6.

**Note 1:** A write to this register is only allowed when RTCWREN = 1.

#### REGISTER 18-17: ALRMHR: ALARM HOURS VALUE REGISTER<sup>(1)</sup>

U-0	U-0	R/W-x	R/W-x	R/W-x	R/W-x	R/W-x	R/W-x
—	_	HRTEN1	HRTEN0	HRONE3	HRONE2	HRONE1	HRONE0
bit 7							bit 0

Legend:			
R = Readable bit	W = Writable bit	U = Unimplemented bit,	read as '0'
-n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown

bit 7-6	Unimplemented: Read as '0'
bit 5-4	HRTEN<1:0>: Binary Coded Decimal Value of Hour's Tens Digit bits
	Contains a value from 0 to 2

	Contains a value from 0 to 2.
bit 3-0	HRONE<3:0>: Binary Coded Decimal Value of Hour's Ones Digit bits
	Contains a value from 0 to 9.

**Note 1:** A write to this register is only allowed when RTCWREN = 1.

#### REGISTER 18-18: ALRMMIN: ALARM MINUTES VALUE REGISTER

U-0	R/W-x						
—	MINTEN2	MINTEN1	MINTEN0	MINONE3	MINONE2	MINONE1	MINONE0
bit 7							bit 0

Legend:			
R = Readable bit	W = Writable bit	U = Unimplemented bit, read	l as '0'
-n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown

bit 7 Unimplemented: Read as '0'

- bit 6-4 **MINTEN<2:0>:** Binary Coded Decimal Value of Minute's Tens Digit bits Contains a value from 0 to 5.
- bit 3-0 MINONE<3:0>: Binary Coded Decimal Value of Minute's Ones Digit bits Contains a value from 0 to 9.

#### REGISTER 18-19: ALRMSEC: ALARM SECONDS VALUE REGISTER

U-0	R/W-x						
—	SECTEN2	SECTEN1	SECTEN0	SECONE3	SECONE2	SECONE1	SECONE0
bit 7							bit 0

Legend:				
R = Reada	ble bit	W = Writable bit	U = Unimplemented bit,	read as '0'
-n = Value at POR '1' = Bit is set		'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown
bit 7	Unimple	mented: Read as '0'		
	-			
bit 6-4	SECTEN	<2:0>: Binary Coded Decim	al Value of Second's Tens Dig	jit bits
	Contains	a value from 0 to 5.		

bit 3-0 SECONE<3:0>: Binary Coded Decimal Value of Second's Ones Digit bits Contains a value from 0 to 9.

#### 18.1.4 RTCEN BIT WRITE

RTCWREN (RTCCFG<5>) must be set before a write to RTCEN can take place. Any write to the RTCEN bit, while RTCWREN = 0, will be ignored.

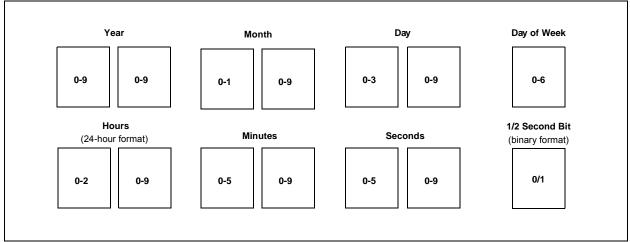
Like the RTCEN bit, the RTCVALH and RTCVALL registers can only be written to when RTCWREN = 1. A write to these registers, while RTCWREN = 0, will be ignored.

FIGURE 18-2: TIMER DIGIT FORMAT

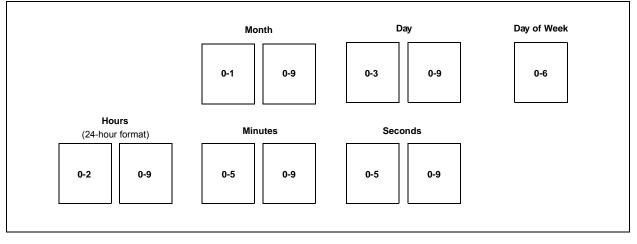
#### 18.2 Operation

#### 18.2.1 REGISTER INTERFACE

The register interface for the RTCC and alarm values is implemented using the Binary Coded Decimal (BCD) format. This simplifies the firmware when using the module, as each of the digits is contained within its own 4-bit value (see Figure 18-2 and Figure 18-3).



#### FIGURE 18-3: ALARM DIGIT FORMAT

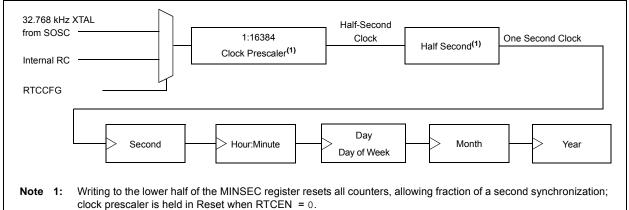


#### 18.2.2 CLOCK SOURCE

As previously mentioned, the RTCC module is intended to be clocked by an external Real-Time Clock (RTC) crystal, oscillating at 32.768 kHz, but an internal oscillator can be used. The RTCC clock selection is decided by the RTCOSC bit (CONFIG3L<0>).

Calibration of the crystal can be done through this module to yield an error of 3 seconds or less per month. (For further details, see Section 18.2.9 "Calibration".)





#### 18.2.2.1 Real-Time Clock Enable

The RTCC module can be clocked by an external, 32.768 kHz crystal (SOSC oscillator) or the LF-INTOSC oscillator, which can be selected in CONFIG3L<0>.

If the external clock is used, the SOSC oscillator should be enabled. If LF-INTOSC is providing the clock, the INTOSC clock can be brought out to the RTCC pin by the RTSECSEL<1:0> bits (PADCFG<2:1>).

#### 18.2.3 DIGIT CARRY RULES

This section explains which timer values are affected when there is a rollover:

- Time of Day: From 23:59:59 to 00:00:00 with a carry to the Day field
- Month: From 12/31 to 01/01 with a carry to the Year field
- Day of Week: From 6 to 0 with no carry (see Table 18-1)
- Year Carry: From 99 to 00; this also surpasses the use of the RTCC

For the day-to-month rollover schedule, see Table 18-2.

Because the following values are in BCD format, the carry to the upper BCD digit occurs at the count of 10, not 16 (SECONDS, MINUTES, HOURS, WEEKDAY, DAYS and MONTHS).

#### TABLE 18-1: DAY OF WEEK SCHEDULE

Day of Week						
Sunday	0					
Monday	1					
Tuesday	2					
Wednesday	3					
Thursday	4					
Friday	5					
Saturday	6					

### TABLE 18-2:DAY TO MONTH ROLLOVER<br/>SCHEDULE

Maximum Day Field
31
28 or 29 <sup>(1)</sup>
31
30
31
30
31
31
30
31
30
31

Note 1: See Section 18.2.4 "Leap Year".

#### 18.2.4 LEAP YEAR

Since the year range on the RTCC module is 2000 to 2099, the leap year calculation is determined by any year divisible by four in the above range. Only February is affected in a leap year.

February will have 29 days in a leap year and 28 days in any other year.

#### 18.2.5 GENERAL FUNCTIONALITY

All Timer registers containing a time value of seconds or greater are writable. The user configures the time by writing the required year, month, day, hour, minutes and seconds to the Timer registers, via register pointers. (See Section 18.2.8 "Register Mapping".)

The timer uses the newly written values and proceeds with the count from the required starting point.

The RTCC is enabled by setting the RTCEN bit (RTCCFG<7>). If enabled, while adjusting these registers, the timer still continues to increment. However, any time the MINSEC register is written to, both of the timer prescalers are reset to '0'. This allows fraction of a second synchronization.

The Timer registers are updated in the same cycle as the write instruction's execution by the CPU. The user must ensure that when RTCEN = 1, the updated registers will not be incremented at the same time. This can be accomplished in several ways:

- By checking the RTCSYNC bit (RTCCFG<4>)
- By checking the preceding digits from which a carry can occur
- By updating the registers immediately following the seconds pulse (or an alarm interrupt)

The user has visibility to the half-second field of the counter. This value is read-only and can be reset only by writing to the lower half of the SECONDS register.

#### 18.2.6 SAFETY WINDOW FOR REGISTER READS AND WRITES

The RTCSYNC bit indicates a time window during which the RTCC Clock Domain registers can be safely read and written without concern about a rollover. When RTCSYNC = 0, the registers can be safely accessed by the CPU.

Whether RTCSYNC = 1 or 0, the user should employ a firmware solution to ensure that the data read did not fall on a rollover boundary, resulting in an invalid or partial read. This firmware solution would consist of reading each register twice and then comparing the two values. If the two values matched, then a rollover did not occur.

#### 18.2.7 WRITE LOCK

In order to perform a write to any of the RTCC Timer registers, the RTCWREN bit (RTCCFG<5>) must be set.

To avoid accidental writes to the RTCC Timer register, it is recommended that the RTCWREN bit (RTCCFG<5>) be kept clear when not writing to the register. For the RTCWREN bit to be set, there is only one instruction cycle time window allowed between the 55h/AA sequence and the setting of RTCWREN. For that reason, it is recommended that users follow the code example in Example 18-1.

### EXAMPLE 18-1: SETTING THE RTCWREN BIT

movlw	0x55
movwf	EECON2
movlw	0xAA
movwf	EECON2
bsf	RTCCFG, RTCWREN

#### 18.2.8 REGISTER MAPPING

To limit the register interface, the RTCC Timer and Alarm Timer registers are accessed through corresponding register pointers. The RTCC Value register window (RTCVALH and RTCVALL) uses the RTCPTRx bits (RTCCFG<1:0>) to select the required Timer register pair.

By reading or writing to the RTCVALH register, the RTCC Pointer value (RTCPTR<1:0>) decrements by '1' until it reaches '00'. When '00' is reached, the MINUTES and SECONDS value is accessible through RTCVALH and RTCVALL until the pointer value is manually changed.

TABLE 18-3: RTCVALH AND RTCVALL REGISTER MAPPING

RTCPTR<1:0>	RTCC Value Register Window				
KICFIK(I.0>	RTCVALH	RTCVALL			
00	MINUTES	SECONDS			
01	WEEKDAY	HOURS			
10	MONTH	DAY			
11	—	YEAR			

The Alarm Value register windows (ALRMVALH and ALRMVALL) use the ALRMPTR bits (ALRMCFG<1:0>) to select the desired alarm register pair.

By reading or writing to the ALRMVALH register, the Alarm Pointer value, ALRMPTR<1:0>, decrements by one until it reaches '00'. When it reaches '00', the ALRMMIN and ALRMSEC values are accessible through ALRMVALH and ALRMVALL until the pointer value is manually changed.

## TABLE 18-4:ALRMVAL REGISTER<br/>MAPPING

ALRMPTR<1:0>	Alarm Value Register Window				
ALRIVIPIR (1.0>	ALRMVALH	ALRMVALL			
0 0	ALRMMIN	ALRMSEC			
01	ALRMWD	ALRMHR			
10	ALRMMNTH	ALRMDAY			
11	—	—			

#### 18.2.9 CALIBRATION

The real-time crystal input can be calibrated using the periodic auto-adjust feature. When properly calibrated, the RTCC can provide an error of less than three seconds per month.

To perform this calibration, find the number of error clock pulses and store the value into the lower half of the RTCCAL register. The eight-bit, signed value, loaded into RTCCAL, is multiplied by four and will be either added or subtracted from the RTCC timer, once every minute.

To calibrate the RTCC module:

- 1. Use another timer resource on the device to find the error of the 32.768 kHz crystal.
- 2. Convert the number of error clock pulses per minute (see Equation 18-1).

#### EQUATION 18-1: CONVERTING ERROR CLOCK PULSES

(Ideal Frequency (32,758) – Measured Frequency) \* 60 = Error Clocks per Minute

- If the oscillator is *faster* than ideal (negative result from Step 2), the RCFGCALL register value needs to be negative. This causes the specified number of clock pulses to be subtracted from the timer counter, once every minute.
- If the oscillator is *slower* than ideal (positive result from Step 2), the RCFGCALL register value needs to be positive. This causes the specified number of clock pulses to be added to the timer counter, once every minute.
- 3. Load the RTCCAL register with the correct value.

Writes to the RTCCAL register should occur only when the timer is turned off or immediately after the rising edge of the seconds pulse.

Note:	In determining the crystal's error value, it						
	is the user's responsibility to include the						
	crystal's initial error from drift due to						
	temperature or crystal aging.						

#### 18.3 Alarm

The Alarm features and characteristics are:

- Configurable from half a second to one year
- Enabled using the ALRMEN bit (ALRMCFG<7>, Register 18-4)
- · Offers one-time and repeat alarm options

#### 18.3.1 CONFIGURING THE ALARM

The alarm feature is enabled using the ALRMEN bit. This bit is cleared when an alarm is issued. The bit will not be cleared if the CHIME bit = 1 or if ALRMRPT  $\neq 0$ .

The interval selection of the alarm is configured through the ALRMCFG bits (AMASK<3:0>); see Figure 18-5. These bits determine which, and how many, digits of the alarm must match the clock value for the alarm to occur.

The alarm can also be configured to repeat based on a preconfigured interval. The number of times this occurs, after the alarm is enabled, is stored in the ALRMRPT register.

Note: While the alarm is enabled (ALRMEN = 1), changing any of the registers, other than the RTCCAL, ALRMCFG and ALRMRPT registers and the CHIME bit, can result in a false alarm event leading to a false alarm interrupt. To avoid this, only change the timer and alarm values while the alarm is disabled (ALRMEN = 0). It is recommended that the ALRMCFG and ALRMRPT registers and CHIME bit be changed when RTCSYNC = 0.



Alarm Mask Setting AMASK<3:0>	Day of the Week	e Month	Day	Hours	Minutes Seconds	
0000 – Every half second 0001 – Every second					:	
0010 – Every 10 seconds						
0011 – Every minute					s s	
0100 – Every 10 minutes					m s s	
0101 – Every hour					mm:ss	
0110 – Every day				h h	: m m : s s	
0111 – Every week	d			h h	: m m : s s	
1000 – Every month			d d	h h	: m m : s s	
1001 – Every year <sup>(1)</sup>		<b>m m</b> /	d d	h h	: m m : s s	
Note 1: Annually, except when c	onfigured fo	r February 29.				

When ALRMCFG = 00 and the CHIME bit = 0 (ALRMCFG<6>), the repeat function is disabled and only a single alarm will occur. The alarm can be repeated up to 255 times by loading the ALRMRPT register with FFh.

After each alarm is issued, the ALRMRPT register is decremented by one. Once the register has reached '00', the alarm will be issued one last time.

After the alarm is issued a last time, the ALRMEN bit is cleared automatically and the alarm is turned off. Indefinite repetition of the alarm can occur if the CHIME bit = 1.

When CHIME = 1, the alarm is not disabled when the ALRMRPT register reaches '00', but it rolls over to FF and continues counting indefinitely.

#### 18.3.2 ALARM INTERRUPT

At every alarm event, an interrupt is generated. Additionally, an alarm pulse output is provided that operates at half the frequency of the alarm.

The alarm pulse output is completely synchronous with the RTCC clock and can be used as a trigger clock to other peripherals. This output is available on the RTCC pin. The output pulse is a clock with a 50% duty cycle and a frequency half that of the alarm event (see Figure 18-6).

The RTCC pin also can output the seconds clock. The user can select between the alarm pulse, generated by the RTCC module, or the seconds clock output.

The RTSECSEL<1:0> bits (PADCFG1<2:1>) select between these two outputs:

- Alarm pulse RTSECSEL<1:0> = 00
- Seconds clock RTSECSEL<1:0> = 01

#### FIGURE 18-6: TIMER PULSE GENERATION

RTCEN bit	
ALRMEN bit	
RTCC Alarm Event	
RTCC Pin	

#### 18.4 Sleep Mode

The timer and alarm continue to operate while in Sleep mode. The operation of the alarm is not affected by Sleep, as an alarm event can always wake up the CPU.

The Idle mode does not affect the operation of the timer or alarm.

#### 18.5 Reset

#### 18.5.1 DEVICE RESET

When a device Reset occurs, the ALRMRPT register is forced to its Reset state, causing the alarm to be disabled (if enabled prior to the Reset). If the RTCC was enabled, it will continue to operate when a basic device Reset occurs.

#### 18.5.2 POWER-ON RESET (POR)

The RTCCFG and ALRMRPT registers are reset only on a POR. Once the device exits the POR state, the clock registers should be reloaded with the desired values.

The timer prescaler values can be reset only by writing to the SECONDS register. No device Reset can affect the prescalers.

#### 18.6 Register Maps

Table 18-5, Table 18-6 and Table 18-7 summarize theregisters associated with the RTCC module.

File Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
RTCCFG	RTCEN	_	RTCWREN	RTCSYNC	HALFSEC	RTCOE	RTCPTR1	RTCPTR0
RTCCAL	CAL7	CAL6	CAL5	CAL4	CAL3	CAL2	CAL1	CAL0
PADCFG1	RDPU	REPU	RJPU <sup>(1)</sup>		_	RTSECSEL1	RTSECSEL0	_
ALRMCFG	ALRMEN	CHIME	AMASK3	AMASK2	AMASK1	AMASK0	ALRMPTR1	ALRMPTR0
ALRMRPT	ARPT7	ARPT6	ARPT5	ARPT4	ARPT3	ARPT2	ARPT1	ARPT0
PMD1	PSPMD	CTMUMD	RTCCMD	TMR4MD	TMR3MD	TMR2MD	TMR1MD	EMBDM

#### TABLE 18-5: RTCC CONTROL REGISTERS

**Legend:** — = unimplemented, read as '0'. Reset values are shown in hexadecimal for 80-pin devices.

Note 1: Unimplemented on 64-pin devices (PIC18F6XK22), read as '0'.

#### TABLE 18-6: RTCC VALUE REGISTERS

File Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
RTCVALH	RTCC Value	RTCC Value High Register Window Based on RTCPTR<1:0>						
RTCVALL	RTCC Value	RTCC Value Low Register Window Based on RTCPTR<1:0>						

#### TABLE 18-7: ALARM VALUE REGISTERS

File Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
ALRMVALH	ALRMVALH Alarm Value High Register Window Based on ALRMPTR<1:0>							
ALRMVALL	LL Alarm Value Low Register Window Based on ALRMPTR<1:0>							

NOTES:

#### 19.0 CAPTURE/COMPARE/PWM (CCP) MODULES

PIC18F87K22 family devices have seven CCP (Capture/Compare/PWM) modules, designated CCP4 through CCP10. All the modules implement standard Capture, Compare and Pulse-Width Modulation (PWM) modes.

**Note:** Throughout this section, generic references are used for register and bit names that are the same, except for an 'x' variable that indicates the item's association with the specific CCP module. For example, the control register is named CCPxCON and refers to CCP4CON through CCP10CON.

Each CCP module contains a 16-bit register that can operate as a 16-bit Capture register, a 16-bit Compare register or a PWM Master/Slave Duty Cycle register. For the sake of clarity, all CCP module operation in the following sections is described with respect to CCP4, but is equally applicable to CCP5 through CCP10.

Note: The CCP9 and CCP10 modules are disabled on the devices with 32 Kbytes of program memory (PIC18FX5K22).

#### REGISTER 19-1: CCPxCON: CCPx CONTROL REGISTER (CCP4-CCP10 MODULES)<sup>(1)</sup>

R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0
PxM1	PxM0	DCxB1	DCxB0	CCPxM3 <sup>(2)</sup>	CCPxM2 <sup>(2)</sup>	CCPxM1 <sup>(2)</sup>	CCPxM0 <sup>(2)</sup>
bit 7							bit 0

Legend:			
R = Readable bit	W = Writable bit	U = Unimplemented bit, rea	d as '0'
-n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown

bit 7-6 PxM<1:0>: PWM Output Configuration bits If CCPxM<3:2> = 00, 01, 10: xx = PxA is assigned as a capture/compare input/output; PxB, PxC and PxD are assigned as port pins If CCPxM<3:2> = 11: 00 = Single output: PxA, PxB, PxC and PxD are controlled by steering 01 = Full-bridge output forward: PxD is modulated; PxA is active; PxB, PxC are inactive 10 = Half-bridge output: PxA, PxB are modulated with dead-band control; PxC and PxD are assigned as port pins 11 = Full-bridge output reverse: PxB is modulated; PxC is active; PxA and PxD are inactive bit 5-4 DCxB<1:0>: PWM Duty Cycle bit 1 and bit 0 for CCPx Module Capture mode: Unused. Compare mode: Unused. PWM mode: These bits are the two Least Significant bits (bit 1 and bit 0) of the 10-bit PWM duty cycle. The eight Most Significant bits (DCx<9:2>) of the duty cycle are found in CCPRxL. Note 1: The CCP9 and CCP10 modules are not available on devices with 32 Kbytes of program memory

- (PIC18FX5K22).
  - 2: CCPxM<3:0> = 1011 will only reset the timer and not start AN A/D conversion on CCPx match.

#### REGISTER 19-1: CCPxCON: CCPx CONTROL REGISTER (CCP4-CCP10 MODULES)<sup>(1)</sup>

#### bit 3-0 CCPxM<3:0>: CCPx Module Mode Select bits<sup>(2)</sup>

- 0000 = Capture/Compare/PWM disabled (resets CCPx module)
- 0001 = Reserved
- 0010 = Compare mode, toggle output on match (CCPxIF bit is set)
- 0011 = Reserved
- 0100 = Capture mode: every falling edge
- 0101 = Capture mode: every rising edge
- 0110 = Capture mode: every 4th rising edge
- 0111 = Capture mode: every 16th rising edge
- 1000 = Compare mode: initialize CCPx pin low; on compare match, force CCPx pin high (CCPxIF bit is set)
- 1001 = Compare mode: initialize CCPx pin high; on compare match, force CCPx pin low (CCPxIF bit is set)
- 1010 = Compare mode: generate software interrupt on compare match (CCPxIF bit is set, CCPx pin reflects I/O state)
- 1011 = Compare mode: Special Event Trigger; reset timer on CCPx match (CCPxIF bit is set)
- 11xx = PWM mode

**2:** CCPxM<3:0> = 1011 will only reset the timer and not start AN A/D conversion on CCPx match.

#### REGISTER 19-2: CCPTMRS1: CCP TIMER SELECT REGISTER 1

R/W-0	R/W-0	U-0	R/W-0	U-0	R/W-0	R/W-0	R/W-0
C7TSEL1	C7TSEL0	—	C6TSEL0	—	C5TSEL0	C4TSEL1	C4TSEL0
bit 7							bit 0

Legend:				
R = Readable bit W = Writable bit		U = Unimplemented bit,	read as '0'	
-n = Value	at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown
bit 7-6	C7TSEL	<1:0>: CCP7 Timer Selection	n bits	
	01 = CC 10 = CC	P7 is based off of TMR1/TMF P7 is based off of TMR5/TMF P7 is based off of TMR5/TMF P7 is based off of TMR5/TMF	R4 R6	
bit 5	Unimple	mented: Read as '0'		
bit 4	C6TSEL	0: CCP6 Timer Selection bit		
	• •••	P6 is based off of TMR1/TMR2 P6 is based off of TMR5/TMR2		
bit 3	Unimple	mented: Read as '0'		
bit 2	0 = CCF	<b>0:</b> CCP5 Timer Selection bit 25 is based off of TMR1/TMR 25 is based off of TMR5/TMR	_	
bit 1-0	00 = CC 01 = CC 10 = CC	<1:0>: CCP4 Timer Selection P4 is based off of TMR1/TMF P4 is based off of TMR3/TMF P4 is based off of TMR3/TMF served; do not use	82 84	

**Note 1:** The CCP9 and CCP10 modules are not available on devices with 32 Kbytes of program memory (PIC18FX5K22).

#### REGISTER 19-3: CCPTMRS2: CCP TIMER SELECT REGISTER 2

U-0	U-0	U-0	R/W-0	U-0	R/W-0	R/W-0	R/W-0
—	—	_	C10TSEL0 <sup>(1)</sup>	—	C9TSEL0 <sup>(1)</sup>	C8TSEL1	C8TSEL0
bit 7							bit 0

Legend:					
R = Readable bit	t W = Writable bit	U = Unimplemented bit, read as '0'			
-n = Value at PO	R '1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown		
bit 7-5 <b>l</b>	Jnimplemented: Read as '0'				

DIL 7-5	Unimplemented: Read as 0
bit 4	C10TSEL0: CCP10 Timer Selection bit <sup>(1)</sup>
	0 = CCP10 is based off of TMR1/TMR2
	1 = CCP10 is based off of TMR7/TMR2
bit 3	Unimplemented: Read as '0'
bit 2	C9TSEL0: CCP9 Timer Selection bit <sup>(1)</sup>
	0 = CCP9 is based off of TMR1/TMR2
	1 = CCP9 is based off of TMR7/TMR4
bit 1-0	C8TSEL<1:0>: CCP8 Timer Selection bits
	On Non 32-Byte Device Variants:
	00 = CCP8 is based off of TMR1/TMR2
	01 = CCP8 is based off of TMR7/TMR4
	10 = CCP8 is based off of TMR7/TMR6
	11 = Reserved; do not use
	On 32-Byte Device Variants (PIC18F65K22 and PIC18F85K22):
	00 = CCP8 is based off of TMR1/TMR2
	01 = CCP8 is based off of TMR1/TMR4
	10 = CCP8 is based off of TMR1/TMR6

- 11 = Reserved; do not use
- **Note 1:** This bit is unimplemented and reads as '0' on devices with 32 Kbytes of program memory (PIC18FX5K22).

#### REGISTER 19-4: CCPRxL: CCPx PERIOD LOW BYTE REGISTER

| R/W-x   |
|---------|---------|---------|---------|---------|---------|---------|---------|
| CCPRxL7 | CCPRxL6 | CCPRxL5 | CCPRxL4 | CCPRxL3 | CCPRxL2 | CCPRxL1 | CCPRxL0 |
| bit 7   |         |         |         |         |         |         | bit 0   |

Legend:			
R = Readable bit	W = Writable bit	U = Unimplemented bit,	read as '0'
-n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown

bit 7-0 CCPRxL<7:0>: CCPx Period Register Low Byte bits Capture Mode: Capture Register Low Byte Compare Mode: Compare Register Low Byte PWM Mode: Duty Cycle Register

#### REGISTER 19-5: CCPRxH: CCPx PERIOD HIGH BYTE REGISTER

| R/W-x   |
|---------|---------|---------|---------|---------|---------|---------|---------|
| CCPRxH7 | CCPRxH6 | CCPRxH5 | CCPRxH4 | CCPRxH3 | CCPRxH2 | CCPRxH1 | CCPRxH0 |
| bit 7   |         |         |         |         |         |         | bit 0   |

Legend:			
R = Readable bit	W = Writable bit	U = Unimplemented bit,	, read as '0'
-n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown

bit 7-0 CCPRxH<7:0>: CCPx Period Register High Byte bits Capture Mode: Capture Register High Byte Compare Mode: Compare Register High Byte PWM Mode: Duty Cycle Buffer Register

**—** 

#### **19.1 CCP Module Configuration**

Each Capture/Compare/PWM module is associated with a control register (generically, CCPxCON) and a data register (CCPRx). The data register, in turn, is comprised of two 8-bit registers: CCPRxL (low byte) and CCPRxH (high byte). All registers are both readable and writable.

#### 19.1.1 CCP MODULES AND TIMER RESOURCES

The CCP modules utilize Timers, 1 through 8, which vary with the selected mode. Various timers are available to the CCP modules in Capture, Compare or PWM modes, as shown in Table 19-1.

#### TABLE 19-1: CCP MODE – TIMER RESOURCE

CCP Mode	Timer Resource					
Capture						
Compare	Timer1, Timer3, Timer 5 or Timer7					
PWM	Timer2, Timer4, Timer 6 or Timer8					

The assignment of a particular timer to a module is determined by the timer to CCP enable bits in the CCPTMRSx registers (see Register 19-2 and Register 19-3). All of the modules may be active at once and may share the same timer resource if they are configured to operate in the same mode (Capture/Compare or PWM) at the same time.

The CCPTMRS1 register selects the timers for CCP modules, 7, 6, 5 and 4, and the CCPTMRS2 register selects the timers for CCP modules, 10, 9 and 8. The possible configurations are shown in Table 19-2 and Table 19-3.

#### TABLE 19-2: TIMER ASSIGNMENTS FOR CCP MODULES 4, 5, 6 AND 7

	CCPTMRS1 Register										
CCP4 CCP5				CCP6		CCP7					
C4TSEL <1:0>	Capture/ Compare Mode	PWM Mode	C5TSEL0	Capture/ Compare Mode	PWM Mode	C6TSEL0	Capture/ Compare Mode	PWM Mode	C7TSEL <1:0>	Capture/ Compare Mode	
0 0	TMR1	TMR2	0	TMR1	TMR2	0	TMR1	TMR2	0 0	TMR1	TMR2
0 1	TMR3	TMR4	1	TMR5	TMR4	1	TMR5	TMR2	0 1	TMR5	TMR4
1 0	TMR3	TMR6							1 0	TMR5	TMR6
1 1 Reserved <sup>(1)</sup>								1 1	TMR5	TMR8	

Note 1: Do not use the reserved bits.

#### TABLE 19-3: TIMER ASSIGNMENTS FOR CCP MODULES 8, 9 AND 10

	CCPTMRS2 Register										
CCP8 CCP8 CCP8 Devices with 32 Kbytes					CCP9 <sup>(1)</sup>		С	CCP10 <sup>(1)</sup>			
C8TSEL <1:0>	Capture/ Compare Mode	PWM Mode	C8TSEL <1:0>	Capture/ Compare Mode	PWM Mode	C9TSEL0 Compare Mode		C10TSEL0	Capture/ Compare Mode	PWM Mode	
0 0	TMR1	TMR2	0 0	TMR1	TMR2	0	TMR1	TMR2	0	TMR1	TMR2
0 1	TMR7	TMR4	0 1	TMR1	TMR4	1	TMR7	TMR4	1	TMR7	TMR2
1 0	TMR7	TMR6	1 0	TMR1	TMR6						
1 1	Reserv	ed <sup>(2)</sup>	1 1	1 1 Reserved <sup>(2)</sup>							

**Note 1:** The module is not available for devices with 32 Kbytes of program memory (PIC18F65K22 and PIC18F85K22).

2: Do not use the reserved setting.

#### 19.1.2 OPEN-DRAIN OUTPUT OPTION

When operating in Output mode (the Compare or PWM modes), the drivers for the CCPx pins can be optionally configured as open-drain outputs. This feature allows the voltage level on the pin to be pulled to a higher level through an external pull-up resistor and allows the output to communicate with external circuits without the need for additional level shifters.

The open-drain output option is controlled by the CCPxOD bits (ODCON2<7:0>). Setting the appropriate bit configures the pin for the corresponding module for open-drain operation.

#### 19.1.3 PIN ASSIGNMENT FOR CCP6, CCP7, CCP8 AND CCP9

The pin assignment for CCP6/7/8/9 (Capture input, Compare and PWM output) can change, based on the device configuration.

The ECCPMX Configuration bit (CONFIG3H<1>) determines the pin to which CCP6/7/8/9 is multiplexed. The pin assignments for these CCP modules are given in Table 19-4.

TABLE 19-4: CCP PIN ASSIGNMENT

ECCPMX	Pin Mapped to								
Value	CCP6	CCP7	CCP8	CC9					
1 (Default)	RE6	RE5	RE4	RE3					
0	RH7	RH6	RH5	RH4					

#### 19.2 Capture Mode

In Capture mode, the CCPR4H:CCPR4L register pair captures the 16-bit value of the Timer register selected in the CCPTMRS1 when an event occurs on the CCP4 pin. An event is defined as one of the following:

- · Every falling edge
- Every rising edge
- · Every 4th rising edge
- · Every 16th rising edge

The event is selected by the mode select bits, CCP4M<3:0> (CCP4CON<3:0>). When a capture is made, the interrupt request flag bit, CCP4IF (PIR4<1>), is set. (It must be cleared in software.) If another capture occurs before the value in CCPR4 is read, the old captured value is overwritten by the new captured value.

Figure 19-1 shows the Capture mode block diagram.

#### 19.2.1 CCP PIN CONFIGURATION

In Capture mode, the appropriate CCPx pin should be configured as an input by setting the corresponding TRIS direction bit.

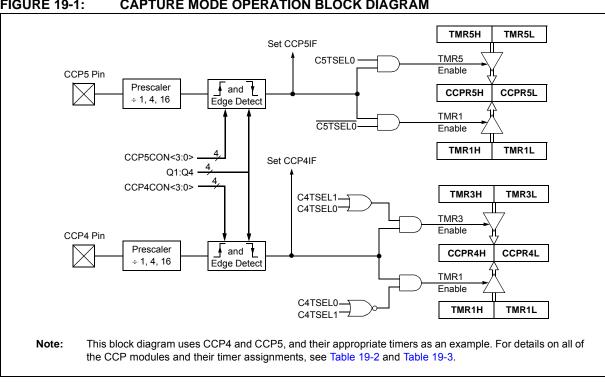
Note:	If RC1	or	RE7 is	s co	onfigu	ured a	as a CCF	P4
	output,	а	write	to	the	port	causes	а
	capture condition.							

#### 19.2.2 TIMER1/3/5/7 MODE SELECTION

For the available timers (1/3/5/7) to be used for the capture feature, the used timers must be running in Timer mode or Synchronized Counter mode. In Asynchronous Counter mode, the capture operation may not work.

The timer to be used with each CCP module is selected in the CCPTMRSx registers. (See Section 19.1.1 "CCP Modules and Timer Resources".)

Details of the timer assignments for the CCP modules are given in Table 19-2 and Table 19-3.



#### **FIGURE 19-1: CAPTURE MODE OPERATION BLOCK DIAGRAM**

#### 19.2.3 SOFTWARE INTERRUPT

When the Capture mode is changed, a false capture interrupt may be generated. The user should keep the CCP4IE bit (PIE4<1>) clear to avoid false interrupts and should clear the flag bit, CCP4IF, following any such change in operating mode.

#### 19.2.4 CCP PRESCALER

There are four prescaler settings in Capture mode. They are specified as part of the operating mode selected by the mode select bits (CCP4M<3:0>). Whenever the CCP module is turned off, or the CCP module is not in Capture mode, the prescaler counter is cleared. This means that any Reset will clear the prescaler counter.

Switching from one capture prescaler to another may generate an interrupt. Doing that will also not clear the prescaler counter - meaning the first capture may be from a non-zero prescaler.

Example 19-1 shows the recommended method for switching between capture prescalers. This example also clears the prescaler counter and will not generate the "false" interrupt.

EXAM	PLE 19-1:		HANGING BETWEEN APTURE PRESCALERS
CLRF	CCP4CON	;	Turn CCP module off
MOVLW	NEW_CAPT_PS	;	Load WREG with the
		;	new prescaler mode
		;	value and CCP ON
MOVWF	CCP4CON	;	Load CCP4CON with

; this value

#### **19.3 Compare Mode**

In Compare mode, the 16-bit CCPR4 register value is constantly compared against the Timer register pair value selected in the CCPTMR1 register. When a match occurs, the CCP4 pin can be:

- Driven high
- Driven low
- Toggled (high-to-low or low-to-high)
- Unchanged (that is, reflecting the state of the I/O latch)

The action on the pin is based on the value of the mode select bits (CCP4M<3:0>). At the same time, the interrupt flag bit, CCP4IF, is set.

Figure 19-2 gives the Compare mode block diagram

#### 19.3.1 CCP PIN CONFIGURATION

The user must configure the CCPx pin as an output by clearing the appropriate TRIS bit.

Note: Clearing the CCP4CON register will force the RC1 or RE7 compare output latch (depending on device configuration) to the default low level. This is not the PORTC or PORTE I/O data latch.

#### 19.3.2 TIMER1/3/5/7 MODE SELECTION

If the CCP module is using the compare feature in conjunction with any of the Timer1/3/5/7 timers, the timers must be running in Timer mode or Synchronized Counter mode. In Asynchronous Counter mode, the compare operation may not work.

**Note:** Details of the timer assignments for the CCP modules are given in Table 19-2 and Table 19-3.

#### 19.3.3 SOFTWARE INTERRUPT MODE

When the Generate Software Interrupt mode is chosen (CCP4M<3:0> = 1010), the CCP4 pin is not affected. Only a CCP interrupt is generated, if enabled, and the CCP4IE bit is set.

#### 19.3.4 SPECIAL EVENT TRIGGER

Both CCP modules are equipped with a Special Event Trigger. This is an internal hardware signal generated in Compare mode to trigger actions by other modules. The Special Event Trigger is enabled by selecting the Compare Special Event Trigger mode (CCP4M<3:0> = 1011).

For either CCP module, the Special Event Trigger resets the Timer register pair for whichever timer resource is currently assigned as the module's time base. This allows the CCPRx registers to serve as a programmable Period register for either timer.

The Special Event Trigger for CCP4 cannot start an A/D conversion.

Note: The Special Event Trigger of ECCP2 can start an A/D conversion, but the A/D Converter must be enabled. For more information, see Section 19.0 "Capture/Compare/PWM (CCP) Modules".

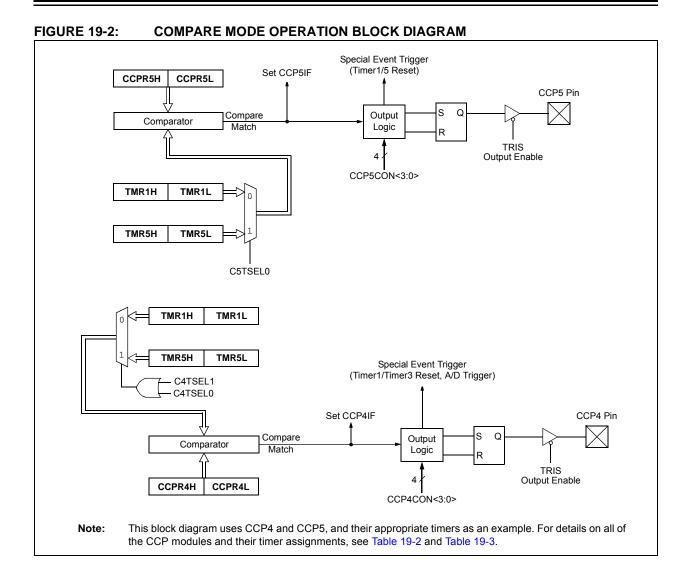


 TABLE 19-5:
 REGISTERS ASSOCIATED WITH CAPTURE, COMPARE, TIMER1/3/5/7

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
INTCON	GIE/GIEH	PEIE/GIEL	TMR0IE	INT0IE	RBIE	TMR0IF	INT0IF	RBIF
RCON	IPEN	SBOREN	CM	RI	TO	PD	POR	BOR
PIR4	CCP10IF <sup>(1)</sup>	CCP9IF <sup>(1)</sup>	CCP8IF	CCP7IF	CCP6IF	CCP5IF	CCP4IF	CCP3IF
PIE4	CCP10IE <sup>(1)</sup>	CCP9IE <sup>(1)</sup>	CCP8IE	CCP7IE	CCP6IE	CCP5IE	CCP4IE	CCP3IE
IPR4	CCP10IP <sup>(1)</sup>	CCP9IP <sup>(1)</sup>	CCP8IP	CCP7IP	CCP6IP	CCP5IP	CCP4IP	CCP3IP
TRISB	TRISB7	TRISB6	TRISB5	TRISB4	TRISB3	TRISB2	TRISB1	TRISB0
TRISC	TRISC7	TRISC6	TRISC5	TRISC4	TRISC3	TRISC2	TRISC1	TRISC0
TRISE	TRISE7	TRISE6	TRISE5	TRISE4	TRISE3	TRISE2	TRISE1	TRISE0
TRISH <sup>(2)</sup>	TRISH7	TRISH6	TRISH5	TRISH4	TRISH3	TRISH2	TRISH1	TRISH0
TMR1L	Timer1 Register Low Byte							

**Legend:** — = unimplemented, read as '0'. Shaded cells are not used by Capture/Compare or Timer1/3/5/7.

Note 1: Unimplemented on devices with a program memory of 32 Kbytes (PIC18F65K22 and PIC18F85K22).

2: Unimplemented on 64-pin devices (PIC18F6XK22), read as '0'.

TABLE 19-5:	REGISTERS ASSOCIATED WITH CAPTURE, COMPARE, TIMER1/3/5/7 (CONTINUED)							
Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
TMR1H	Timer1 Regis	Timer1 Register High Byte						
TMR3L	Timer3 Regis	ter Low Byte						
TMR3H	Timer3 Regis	ter High Byte	ł					
TMR5L	Timer5 Regis	ter Low Byte						
TMR5H	Timer5 Regis	ter High Byte	ł					
TMR7L <sup>(1)</sup>	Timer7 Regis	ter Low Byte						
TMR7H <sup>(1)</sup>	Timer7 Regis	ter High Byte	ł					
T1CON	TMR1CS1	TMR1CS0	T1CKPS1	T1CKPS0	SOSCEN	T1SYNC	RD16	TMR10N
T3CON	TMR3CS1	TMR3CS0	T3CKPS1	T3CKPS0	SOSCEN	T3SYNC	RD16	TMR3ON
T5CON	TMR5CS1	TMR5CS0	T5CKPS1	T5CKPS0	SOSCEN	T5SYNC	RD16	TMR5ON
T7CON <sup>(1)</sup>	TMR7CS1	TMR7CS0	T7CKPS1	T7CKPS0	SOSCEN	T7SYNC	RD16	TMR7ON
CCPR4L	Capture/Com	pare/PWM R	egister 4 Lo	ow Byte				
CCPR4H	Capture/Com	pare/PWM R	egister 4 Hi	gh Byte				
CCPR5L	Capture/Com	pare/PWM R	egister 5 Lo	ow Byte				
CCPR5H	Capture/Com	pare/PWM R	egister 5 Hi	gh Byte				
CCPR6L	Capture/Com	pare/PWM R	egister 6 Lo	ow Byte				
CCPR6H	Capture/Com	pare/PWM R	egister 6 Hi	gh Byte				
CCPR7L	Capture/Com	pare/PWM R	egister 7 Lo	ow Byte				
CCPR7H	Capture/Com	pare/PWM R	egister 7 Hi	gh Byte				
CCPR8L	Capture/Com	pare/PWM R	egister 8 Lo	ow Byte				
CCPR8H	Capture/Com	pare/PWM R	egister 8 Hi	gh Byte				
CCPR9L <sup>(1)</sup>	Capture/Com	pare/PWM R	egister 9 Lo	ow Byte				
CCPR9H <sup>(1)</sup>	Capture/Com	pare/PWM R	egister 9 Hi	gh Byte				
CCPR10L <sup>(1)</sup>	Capture/Com	pare/PWM R	egister 10 L	ow Byte				
CCPR10H <sup>(1)</sup>	Capture/Com	pare/PWM R	egister 10 H	ligh Byte				
CCP4CON			DC4B1	DC4B0	CCP4M3	CCP4M2	CCP4M1	CCP4M0
CCP5CON	_	—	DC5B1	DC5B0	CCP5M3	CCP5M2	CCP5M1	CCP5M0
CCP6CON	_	_	DC6B1	DC6B0	CCP6M3	CCP6M2	CCP6M1	CCP6M0
CCP7CON			DC7B1	DC7B0	CCP7M3	CCP7M2	CCP7M1	CCP7M0
CCP8CON	_	_	DC8B1	DC8B0	CCP8M3	CCP8M2	CCP8M1	CCP8M0
CCP9CON <sup>(1)</sup>		_	DC9B1	DC9B0	CCP9M3	CCP9M2	CCP9M1	CCP9M0
CCP10CON <sup>(1)</sup>	_	_	DC10B1	DC10B0	CCP10M3	CCP10M2	CCP10M1	CCP10M0
CCPTMRS1	C7TSEL1	C7TSEL0	_	C6TSEL0	_	C5TSEL0	C4TSEL1	C4TSEL0
CCPTMRS2	_	_	_	C10TSEL0 <sup>(1)</sup>	_	C9TSEL0 <sup>(1)</sup>	C8TSEL1	C8TSEL0
PMD3	CCP10MD <sup>(1)</sup>	CCP9MD <sup>(1)</sup>	CCP8MD	CCP7MD	CCP6MD	CCP5MD	CCP4MD	TMR12MD <sup>(1)</sup>

## TABLE 19-5: REGISTERS ASSOCIATED WITH CAPTURE, COMPARE, TIMER1/3/5/7 (CONTINUED)

Legend: — = unimplemented, read as '0'. Shaded cells are not used by Capture/Compare or Timer1/3/5/7.

Note 1: Unimplemented on devices with a program memory of 32 Kbytes (PIC18F65K22 and PIC18F85K22).

2: Unimplemented on 64-pin devices (PIC18F6XK22), read as '0'.

## 19.4 PWM Mode

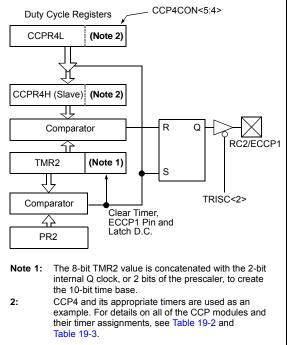
In Pulse-Width Modulation (PWM) mode, the CCP4 pin produces up to a 10-bit resolution PWM output. Since the CCP4 pin is multiplexed with a PORTC or PORTE data latch, the appropriate TRIS bit must be cleared to make the CCP4 pin an output.

Note:	Clearing the CCP4CON register will force the RC1 or RE7 output latch (depending on device configuration) to the default low level. This is not the PORTC or PORTE
	I/O data latch.

Figure 19-3 shows a simplified block diagram of the ECCP1 module in PWM mode.

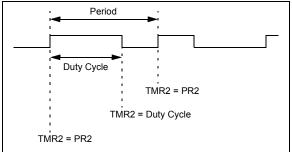
For a step-by-step procedure on how to set up the CCP module for PWM operation, see Section 19.4.3 "Setup for PWM Operation".





A PWM output (Figure 19-4) has a time base (period) and a time that the output stays high (duty cycle). The frequency of the PWM is the inverse of the period (1/period).

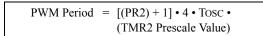
## FIGURE 19-4: PWM OUTPUT



## 19.4.1 PWM PERIOD

The PWM period is specified by writing to the PR2 register. The PWM period can be calculated using the following formula:

## EQUATION 19-1:



PWM frequency is defined as 1/[PWM period].

When TMR2 is equal to PR2, the following three events occur on the next increment cycle:

- TMR2 is cleared
- · The CCP4 pin is set

(An exception: If PWM duty cycle = 0%, the CCP4 pin will not be set)

 The PWM duty cycle is latched from CCPR4L into CCPR4H

Note:	The	Timer2	postscalers	(see		
	Section	n 15.0 "Tin	n <mark>er2 Module</mark> ") a	are not		
	used ir	n the deter	mination of the	PWM		
	frequency. The postscaler could be used					
	to have	a servo up	odate rate at a di	fferent		
	frequer	icy than the	e PWM output.			

## 19.4.2 PWM DUTY CYCLE

The PWM duty cycle is specified to use CCP4, as an example, by writing to the CCPR4L register and to the CCP4CON<5:4> bits. Up to 10-bit resolution is available. The CCPR4L contains the eight MSbs and the CCP4CON<5:4> bits contain the two LSbs. This 10-bit value is represented by CCPR4L:CCP4CON<5:4>. The following equation is used to calculate the PWM duty cycle in time:

## EQUATION 19-2:

PWM Duty Cycle = (CCPR4L:CCP4CON<5:4>) • Tosc • (TMR2 Prescale Value)

CCPR4L and CCP4CON<5:4> can be written to at any time, but the duty cycle value is not latched into CCPR4H until after a match between PR2 and TMR2 occurs (that is, the period is complete). In PWM mode, CCPR4H is a read-only register.

The CCPR4H register and a two-bit internal latch are used to double-buffer the PWM duty cycle. This double-buffering is essential for glitchless PWM operation.

When the CCPR4H and two-bit latch match TMR2, concatenated with an internal two-bit Q clock or two bits of the TMR2 prescaler, the CCP4 pin is cleared.

The maximum PWM resolution (bits) for a given PWM frequency is shown in Equation 19-3:

## **EQUATION 19-3:**

PWM Resolution (max) = 
$$\frac{\log(\frac{FOSC}{FPWM})}{\log(2)}$$
 bits

Note: If the PWM duty cycle value is longer than the PWM period, the CCP4 pin will not be cleared.

## TABLE 19-6: EXAMPLE PWM FREQUENCIES AND RESOLUTIONS AT 40 MHz

PWM Frequency	2.44 kHz	9.77 kHz	39.06 kHz	156.25 kHz	312.50 kHz	416.67 kHz
Timer Prescaler (1, 4, 16)	16	4	1	1	1	1
PR2 Value	FFh	FFh	FFh	3Fh	1Fh	17h
Maximum Resolution (bits)	10	10	10	8	7	6.58

## 19.4.3 SETUP FOR PWM OPERATION

To configure the CCP module for PWM operation, using CCP4 as an example:

- 1. Set the PWM period by writing to the PR2 register.
- 2. Set the PWM duty cycle by writing to the CCPR4L register and CCP4CON<5:4> bits.
- 3. Make the CCP4 pin an output by clearing the appropriate TRIS bit.
- 4. Set the TMR2 prescale value, then enable Timer2 by writing to T2CON.
- 5. Configure the CCP4 module for PWM operation.

### TABLE 19-7: REGISTERS ASSOCIATED WITH PWM AND TIMERS

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
INTCON	GIE/GIEH	PEIE/GIEL	TMR0IE	INT0IE	RBIE	TMR0IF	INT0IF	RBIF
RCON	IPEN	SBOREN	CM	RI	TO	PD	POR	BOR
PIR4	CCP10IF <sup>(1)</sup>	CCP9IF <sup>(1)</sup>	CCP8IF	CCP7IF	CCP6IF	CCP5IF	CCP4IF	CCP3IF
PIE4	CCP10IE <sup>(1)</sup>	CCP9IE <sup>(1)</sup>	CCP8IE	CCP7IE	CCP6IE	CCP5IE	CCP4IE	CCP3IE
IPR4	CCP10IP <sup>(1)</sup>	CCP9IP <sup>(1)</sup>	CCP8IP	CCP7IP	CCP6IP	CCP5IP	CCP4IP	CCP3IP
TRISB	TRISB7	TRISB6	TRISB5	TRISB4	TRISB3	TRISB2	TRISB1	TRISB0
TRISC	TRISC7	TRISC6	TRISC5	TRISC4	TRISC3	TRISC2	TRISC1	TRISC0
TRISE	TRISE7	TRISE6	TRISE5	TRISE4	TRISE3	TRISE2	TRISE1	TRISE0

**Legend:** — = unimplemented, read as '0'. Shaded cells are not used by PWM or Timer2/4/6/8.

Note 1: Unimplemented on devices with a program memory of 32 Kbytes (PIC18F65K22 and PIC18F85K22).

2: Unimplemented on 64-pin devices (PIC18F6XK22), read as '0'.

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	
TRISH <sup>(2)</sup>	TRISH7	TRISH6	TRISH5	TRISH4	TRISH3	TRISH2	TRISH1	TRISH0	
TMR2	Timer2 Regis	Timer2 Register							
TMR4	Timer4 Regis	ter							
TMR6	Timer6 Regis	ter							
TMR8	Timer8 Regis	ter							
PR2	Timer2 Perio	d Register							
PR4	Timer4 Perio	d Register							
PR6	Timer6 Perio	d Register							
PR8	Timer8 Perio	d Register							
T2CON		T2OUTPS3	T2OUTPS2	T2OUTPS1	T2OUTPS0	TMR2ON	T2CKPS1	T2CKPS0	
T4CON	_	T4OUTPS3	T4OUTPS2	T4OUTPS1	T4OUTPS0	TMR4ON	T4CKPS1	T4CKPS0	
T6CON		T6OUTPS3	T6OUTPS2	T6OUTPS1	T6OUTPS0	TMR6ON	T6CKPS1	T6CKPS0	
T8CON		T8OUTPS3	T8OUTPS2	T8OUTPS1	T8OUTPS0	TMR8ON	T8CKPS1	T8CKPS0	
CCPR4L	Capture/Com	pare/PWM F	Register 4 Lo	w Byte					
CCPR4H	Capture/Com	pare/PWM F	Register 4 Hig	gh Byte					
CCPR5L	Capture/Com	pare/PWM F	Register 5 Lo	w Byte					
CCPR5H	Capture/Com	pare/PWM F	Register 5 Hig	gh Byte					
CCPR6L	Capture/Com	pare/PWM F	Register 6 Lo	w Byte					
CCPR6H	Capture/Com	pare/PWM F	Register 6 Hig	gh Byte					
CCPR7L	Capture/Com	pare/PWM F	Register 7 Lo	w Byte					
CCPR7H	Capture/Com	pare/PWM F	Register 7 Hig	gh Byte					
CCPR8L	Capture/Com	pare/PWM F	Register 8 Lo	w Byte					
CCPR8H	Capture/Com	pare/PWM F	Register 8 Hig	gh Byte					
CCPR9L <sup>(1)</sup>	Capture/Com	pare/PWM F	Register 9 Lo	w Byte					
CCPR9H <sup>(1)</sup>	Capture/Com	pare/PWM F	Register 9 Hig	gh Byte					
CCPR10L <sup>(1)</sup>	Capture/Com	pare/PWM F	Register 10 L	ow Byte					
CCPR10H <sup>(1)</sup>	Capture/Com	pare/PWM F	Register 10 H	ligh Byte					
CCP4CON		—	DC4B1	DC4B0	CCP4M3	CCP4M2	CCP4M1	CCP4M0	
CCP5CON	_	_	DC5B1	DC5B0	CCP5M3	CCP5M2	CCP5M1	CCP5M0	
CCP6CON	_	_	DC6B1	DC6B0	CCP6M3	CCP6M2	CCP6M1	CCP6M0	
CCP7CON			DC7B1	DC7B0	CCP7M3	CCP7M2	CCP7M1	CCP7M0	
CCP8CON	_	_	DC8B1	DC8B0	CCP8M3	CCP8M2	CCP8M1	CCP8M0	
CCP9CON <sup>(1)</sup>		_	DC9B1	DC9B0	CCP9M3	CCP9M2	CCP9M1	CCP9M0	
CCP10CON <sup>(1)</sup>		_	DC10B1	DC10B0	CCP10M3	CCP10M2	CCP10M1	CCP10M0	
CCPTMRS1	C7TSEL1	C7TSEL0	_	C6TSEL0	_	C5TSEL0	C4TSEL1	C4TSEL0	
CCPTMRS2				C10TSEL0 <sup>(1)</sup>		C9TSEL0 <sup>(1)</sup>	C8TSEL1	C8TSEL0	
PMD3	CCP10MD <sup>(1)</sup>	CCP9MD <sup>(1)</sup>	CCP8MD	CCP7MD	CCP6MD	CCP5MD	CCP4MD	TMR12MD <sup>(1</sup>	

Legend: — = unimplemented, read as '0'. Shaded cells are not used by PWM or Timer2/4/6/8.

Note 1: Unimplemented on devices with a program memory of 32 Kbytes (PIC18F65K22 and PIC18F85K22).

2: Unimplemented on 64-pin devices (PIC18F6XK22), read as '0'.

NOTES:

## 20.0 ENHANCED CAPTURE/COMPARE/PWM (ECCP) MODULE

PIC18F87K22 family devices have three Enhanced Capture/Compare/PWM (ECCP) modules: ECCP1, ECCP2 and ECCP3. These modules contain a 16-bit register, which can operate as a 16-bit Capture register, a 16-bit Compare register or a PWM Master/Slave Duty Cycle register. These ECCP modules are upward compatible with CCP.

**Note:** Throughout this section, generic references are used for register and bit names that are the same, except for an 'x' variable that indicates the item's association with the ECCP1, ECCP2 or ECCP3 module. For example, the control register is named CCPxCON and refers to CCP1CON, CCP2CON and CCP3CON.

ECCP1, ECCP2 and ECCP3 are implemented as standard CCP modules with Enhanced PWM capabilities. These include:

- Provision for two or four output channels
- · Output Steering modes
- · Programmable polarity
- Programmable dead-band control
- · Automatic shutdown and restart

The enhanced features are discussed in detail in Section 20.4 "PWM (Enhanced Mode)".

The ECCP1, ECCP2 and ECCP3 modules use the control registers: CCP1CON, CCP2CON and CCP3CON. The control registers, CCP4CON through CCP10CON, are for the modules, CCP4 through CCP10.

R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0			
PxM1	PxM0	DCxB1	DCxB0	CCPxM3	CCPxM2	CCPxM1	CCPxM0			
bit 7							bit (			
Legend:										
R = Readabl	le bit	W = Writable		U = Unimpler	mented bit, read	d as '0'				
-n = Value at	t POR	'1' = Bit is set		'0' = Bit is cle	ared	x = Bit is unk	nown			
bit 7-6	<b>PxM&lt;1:0&gt;:</b> E	Inhanced PWM	l Output Confi	guration bits						
		<b>2&gt; =</b> 00, 01, 10	•	0						
	xx = PxA is a	assigned as ca	oture/compare	e input/output; F	PxB, PxC and P	xD are assigne	ed as port pine			
	-		B, PxC and F	PxD are contro	lled by steering	(see Section	20.4.7 "Puls			
	01 = Full-brid	lge output forw			is active; PxB, F ith dead-band					
		ed as port pins lge output reve	rse: PxB is m	odulated; PxC	is active; PxA a	nd PxD are ina	ctive			
bit 5-4		PWM Duty Cyd								
	<u>Capture mod</u> Unused.									
	<u>Compare mo</u> Unused.	<u>de:</u>								
	<u>PWM mode:</u> These bits are in CCPRxL.	e the two LSbs	of the 10-bit F	PWM duty cycle	e. The eight MSI	bs of the duty c	ycle are found			
bit 3-0	CCPxM<3:0>	-: ECCPx Mode	e Select bits							
		0000 = Capture/Compare/PWM off (resets ECCPx module) 0001 = Reserved								
		0010 = Compare mode: toggle output on match								
	0011 = Capt		<b>f</b> allia a salara							
		ure mode: ever ure mode: ever								
		ure mode: ever		edae						
	0111 = Capt	ure mode: ever	y 16 <sup>th</sup> rising e	edge						
					put on compare	•	,			
					output on comp , ECCPx pin re					
	1011 <b>= Com</b>				esets TMR1 or					
		,	nd PxC are ac	tive-high; PxB	and PxD are ac	tive-high				
	1101 = PWM	1 mode: PxA ar	nd PxC are ac	tive-high; PxB	and PxD are ac	tive-low				
					nd PxD are act nd PxD are act					

## REGISTER 20-1: CCPxCON: ENHANCED CAPTURE/COMPARE/PWMx CONTROL

## REGISTER 20-2: CCPTMRS0: CCP TIMER SELECT 0 REGISTER

110 = Reserved; do not use 111 = Reserved; do not use

110 = Reserved; do not use 111 = Reserved; do not use

C1TSEL<2:0>: ECCP1 Timer Selection bits 000 = ECCP1 is based off of TMR1/TMR2 001 = ECCP1 is based off of TMR3/TMR4 010 = ECCP1 is based off of TMR3/TMR6 011 = ECCP1 is based off of TMR3/TMR8

bit 2-0

R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0
C3TSEL1	C3TSEL0	C2TSEL2	C2TSEL1	C2TSEL0	C1TSEL2	C1TSEL1	C1TSEL0
bit 7							bit C
Legend:							
R = Readable	bit	W = Writable	bit	U = Unimpler	nented bit, read	as '0'	
-n = Value at F	POR	'1' = Bit is set		'0' = Bit is cleared		x = Bit is unknown	
	10 = ECCP3	is based off of is based off of is based off of	TMR3/TMR6				
bit 5-3	000 = ECCP 001 = ECCP 010 = ECCP	>: ECCP2 Time 2 is based off of 2 is based off of 2 is based off of 2 is based off of 2 is based off of	of TMR1/TMR of TMR3/TMR of TMR3/TMR	2 4 6			

100 = ECCP1 is based off of TMR3/TMR10: option reserved on the 32-Kbyte device variant; do not use 101 = ECCP1 is based off of TMR3/TMR12: option reserved on the 32-Kbyte device variant; do not use

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In addition to the expanded range of modes available through the CCPxCON and ECCPxAS registers, the ECCP modules have two additional registers associated with Enhanced PWM operation and auto-shutdown features. They are:

- ECCPxDEL Enhanced PWM Control
- PSTRxCON Pulse Steering Control

## 20.1 ECCP Outputs and Configuration

The Enhanced CCP module may have up to four PWM outputs, depending on the selected operating mode. The CCPxCON register is modified to allow control over four PWM outputs: ECCPx/PxA, PxB, PxC and PxD. Applications can use one, two or four of these outputs.

The outputs that are active depend on the ECCP selected operating mode. The pin assignments are summarized in Table 20-3.

To configure the I/O pins as PWM outputs, the proper PWM mode must be selected by setting the PxM<1:0> and CCPxM<3:0> bits. The appropriate TRIS direction bits for the port pins must also be set as outputs.

## 20.1.1 ECCP MODULE AND TIMER RESOURCES

The ECCP modules use Timers, 1, 2, 3, 4, 6, 8, 10 or 12, depending on the mode selected. These timers are available to CCP modules in Capture, Compare or PWM modes, as shown in Table 20-1.

## TABLE 20-1: ECCP MODE – TIMER RESOURCE

ECCP Mode	Timer Resource
Capture	Timer1 or Timer3
Compare	Timer1 or Timer3
PWM	Timer2, Timer4, Timer6, Timer8, Timer10 or Timer12

The assignment of a particular timer to a module is determined by the timer to ECCP enable bits in the CCPTMRSx register (Register 20-2). The interactions between the two modules are depicted in Figure 20-1. Capture operations are designed to be used when the timer is configured for Synchronous Counter mode. Capture operations may not work as expected if the associated timer is configured for Asynchronous Counter mode.

## 20.1.2 ECCP PIN ASSIGNMENT

The pin assignment for ECCPx (Capture input, Compare and PWM output) can change, based on device configuration. The ECCPMX (CONFIG3H<1>) Configuration bit determines which pins ECCP1 and ECCP3 are multiplexed to.

- Default/ECCPMX = 1:
  - ECCP1 (P1B/P1C) is multiplexed onto RE6 and RE5
  - ECCP3 (P3B/P3C) is multiplexed onto RE4 and RE3
- ECCPMX = 0:
  - ECCP1 (P1B/P1C) is multiplexed onto RH7 and RH6
  - ECCP3 (P3B/P3C) is multiplexed onto RH5 and RH4.

The pin assignment for ECCP2 (Capture input, Compare and PWM output) can change, based on the device configuration.

The CCP2MX Configuration bit (CONFIG3H<0>) determines which pin ECCP2 is multiplexed to.

- If CCP2MX = 1 (default) ECCP2 is multiplexed to RC1
- If CCP2MX = 0 ECCP2 is multiplexed to RE7

## 20.2 Capture Mode

In Capture mode, the CCPRxH:CCPRxL register pair captures the 16-bit value of the TMR1 or TMR3 registers when an event occurs on the corresponding ECCPx pin. An event is defined as one of the following:

- Every falling edge
- · Every rising edge
- Every fourth rising edge
- Every 16<sup>th</sup> rising edge

The event is selected by the mode select bits, CCPxM<3:0> (CCPxCON<3:0>). When a capture is made, the interrupt request flag bit, CCPxIF, is set (see Table 20-2). The flag must be cleared by software. If another capture occurs before the value in the CCPRxH/L register is read, the old captured value is overwritten by the new captured value.

#### TABLE 20-2: ECCP1/2/3 INTERRUPT FLAG BITS

ECCP Module	Flag Bit
1	PIR3<1>
2	PIR3<2>
3	PIR4<0>

## 20.2.1 ECCP PIN CONFIGURATION

In Capture mode, the appropriate ECCPx pin should be configured as an input by setting the corresponding TRIS direction bit.

Note:	If the ECCPx pin is configured as an out-
	put, a write to the port can cause a capture
	condition.

## 20.2.2 TIMER1/2/3/4/6/8/10/12 MODE SELECTION

The timers that are to be used with the capture feature (Timer1 2, 3, 4, 6, 8, 10 or 12) must be running in Timer mode or Synchronized Counter mode. In Asynchronous Counter mode, the capture operation may not work. The timer to be used with each ECCP module is selected in the CCPTMRS0 register (Register 20-2).

## 20.2.3 SOFTWARE INTERRUPT

When the Capture mode is changed, a false capture interrupt may be generated. The user should keep the CCPxIE interrupt enable bit clear to avoid false interrupts. The interrupt flag bit, CCPxIF, should also be cleared following any such change in operating mode.

## 20.2.4 ECCP PRESCALER

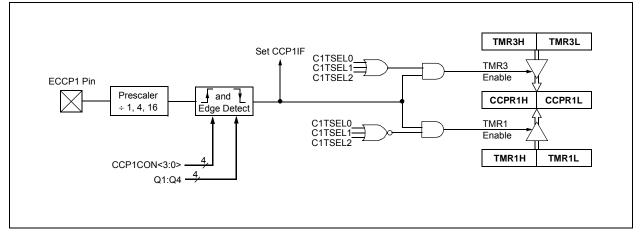
There are four prescaler settings in Capture mode; they are specified as part of the operating mode selected by the mode select bits (CCPxM<3:0>). Whenever the ECCP module is turned off, or Capture mode is disabled, the prescaler counter is cleared. This means that any Reset will clear the prescaler counter.

Switching from one capture prescaler to another may generate an interrupt. Also, the prescaler counter will not be cleared; therefore, the first capture may be from a non-zero prescaler. Example 20-1 provides the recommended method for switching between capture prescalers. This example also clears the prescaler counter and will not generate the "false" interrupt.

## EXAMPLE 20-1: CHANGING BETWEEN CAPTURE PRESCALERS

CLRF	CCP1CON	;	Turn ECCP module off
MOVLW	NEW_CAPT_PS	;	Load WREG with the
		;	new prescaler mode
		;	value and ECCP ON
MOVWF	CCP1CON	;	Load CCP1CON with
		;	this value

## FIGURE 20-1: CAPTURE MODE OPERATION BLOCK DIAGRAM



## 20.3 Compare Mode

In Compare mode, the 16-bit CCPRx register value is constantly compared against the Timer register pair value selected in the CCPTMR1 register. When a match occurs, the ECCPx pin can be:

- Driven high
- Driven low
- Toggled (high-to-low or low-to-high)
- Unchanged (that is, reflecting the state of the I/O latch)

The action on the pin is based on the value of the mode select bits (CCPxM<3:0>). At the same time, the interrupt flag bit, CCPxIF, is set.

### 20.3.1 ECCP PIN CONFIGURATION

Users must configure the ECCPx pin as an output by clearing the appropriate TRIS bit.

Note:	Clearing the	CCPxCON re	egister wil	I force	
	the ECCPx	compare	output	latch	
	(depending on device configuration) to the				
	default low le	evel. This is	not the F	PORTX	
	I/O data latch				

## 20.3.2 TIMER1/2/3/4/6/8/10/12 MODE SELECTION

Timer1 2, 3, 4, 6, 8, 10 or 12 must be running in Timer mode or Synchronized Counter mode if the ECCP module is using the compare feature. In Asynchronous Counter mode, the compare operation will not work reliably.

## 20.3.3 SOFTWARE INTERRUPT MODE

When the Generate Software Interrupt mode is chosen (CCPxM<3:0> = 1010), the ECCPx pin is not affected; only the CCPxIF interrupt flag is affected.

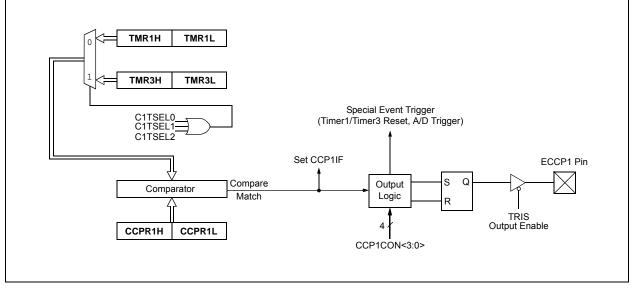
## 20.3.4 SPECIAL EVENT TRIGGER

The ECCP module is equipped with a Special Event Trigger. This is an internal hardware signal generated in Compare mode to trigger actions by other modules. The Special Event Trigger is enabled by selecting the Compare Special Event Trigger mode (CCPxM<3:0> = 1011).

The Special Event Trigger resets the Timer register pair for whichever timer resource is currently assigned as the module's time base. This allows the CCPRx registers to serve as a programmable Period register for either timer.

The Special Event Trigger can also start an A/D conversion. In order to do this, the A/D Converter must already be enabled.

## FIGURE 20-2: COMPARE MODE OPERATION BLOCK DIAGRAM



## 20.4 PWM (Enhanced Mode)

The Enhanced PWM mode can generate a PWM signal on up to four different output pins with up to 10 bits of resolution. It can do this through four different PWM Output modes:

- Single PWM
- Half-Bridge PWM
- Full-Bridge PWM, Forward mode
- Full-Bridge PWM, Reverse mode

To select an Enhanced PWM mode, the PxM bits of the CCPxCON register must be set appropriately.

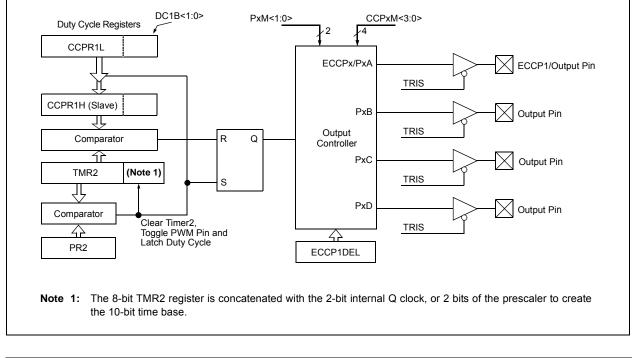
The PWM outputs are multiplexed with I/O pins and are designated: PxA, PxB, PxC and PxD. The polarity of the PWM pins is configurable and is selected by setting the CCPxM bits in the CCPxCON register appropriately.

Table 20-1provides the pin assignments for eachEnhanced PWM mode.

Figure 20-3 provides an example of a simplified block diagram of the Enhanced PWM module.

**Note:** To prevent the generation of an incomplete waveform when the PWM is first enabled, the ECCP module waits until the start of a new PWM period before generating a PWM signal.

## FIGURE 20-3: EXAMPLE SIMPLIFIED BLOCK DIAGRAM OF THE ENHANCED PWM MODE



Note 1: The TRIS register value for each PWM output must be configured appropriately.2: Any pin not used by an Enhanced PWM mode is available for alternate pin functions.

ECCP Mode	PxM<1:0>	PxA	PxB	PxC	PxD
Single	0 0	Yes <sup>(1)</sup>	Yes <sup>(1)</sup>	Yes <sup>(1)</sup>	Yes <sup>(1)</sup>
Half-Bridge	10	Yes	Yes	No	No
Full-Bridge, Forward	01	Yes	Yes	Yes	Yes
Full-Bridge, Reverse	11	Yes	Yes	Yes	Yes

## TABLE 20-3: EXAMPLE PIN ASSIGNMENTS FOR VARIOUS PWM ENHANCED MODES

**Note 1:** Outputs are enabled by pulse steering in Single mode (see Register 20-5).

## FIGURE 20-4: EXAMPLE PWM (ENHANCED MODE) OUTPUT RELATIONSHIPS (ACTIVE-HIGH STATE)

		PxA Modulated			' 
00	(Single Output)		Delay <sup>(1)</sup>	Delay <sup>(1)</sup>	/ /
		PxA Modulated		<b>▲►</b>	· · ·
10	(Half-Bridge)	PxB Modulated	_ !		<u> </u>
		PxA Active	;		
01	(Full-Bridge, Forward)	PxB Inactive	- ;		
01	Forward)	PxC Inactive			
		PxD Modulated			1 1 1
		PxA Inactive			1 1 1
11	(Full-Bridge, Reverse)	PxB Modulated		 	1 1 
	Reverse)	PxC Active	- ' - ' '		
		PxD Inactive		1 	1 1 1
Rela	tionships:	c * (PR2 + 1) * (TMR2 Pre			'

Dead-Band Delay Mode").

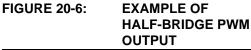
			-	Width	- Period	
00	(Single Output)	PxA Modulated			,	
		PxA Modulated		► ay(1)	Delay <sup>(1)</sup>	
10	(Half-Bridge)	PxB Modulated		ay		
		PxA Active				
01	(Full-Bridge, Forward)	PxB Inactive			   	<i>i</i> I I
01	Forward)	PxC Inactive	<u> </u>			I 
		PxD Modulated				 
		PxA Inactive	i		1 1 1	1 1 1
11	(Full-Bridge, Reverse)	PxB Modulated				
	(tevelse)	PxC Active	;		1 1	1 
		PxD Inactive	;			
	<ul> <li>Pulse Width = Tos</li> <li>Delay = 4 * Tosc</li> </ul>	* (PR2 + 1) * (TMR2 Pre: sc * (CCPRxL<7:0>:CCP * (ECCPxDEL<6:0>) delay is programmed usir	xCON<5:4>)*(	•		ammable Dead-Banc

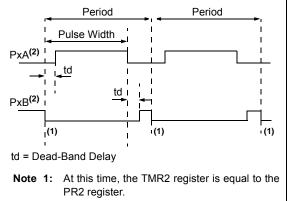
## FIGURE 20-5: EXAMPLE ENHANCED PWM OUTPUT RELATIONSHIPS (ACTIVE-LOW STATE)

## 20.4.1 HALF-BRIDGE MODE

In Half-Bridge mode, two pins are used as outputs to drive push-pull loads. The PWM output signal is output on the PxA pin, while the complementary PWM output signal is output on the PxB pin (see Figure 20-6). This mode can be used for half-bridge applications, as shown in Figure 20-7, or for full-bridge applications, where four power switches are being modulated with two PWM signals.

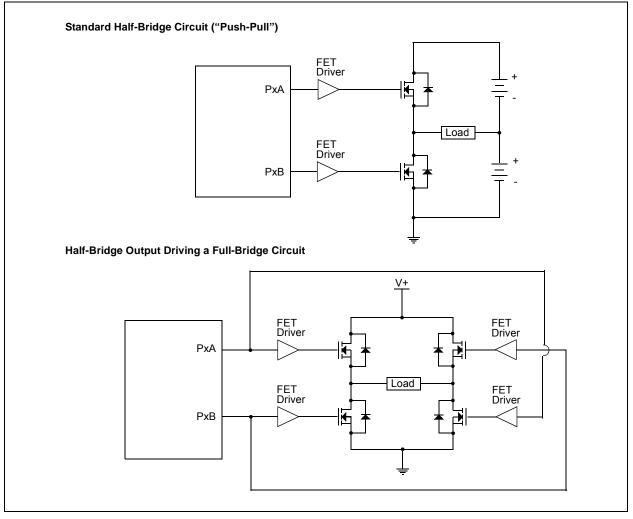
In Half-Bridge mode, the programmable dead-band delay can be used to prevent shoot-through current in half-bridge power devices. The value of the PxDC<6:0> bits of the ECCPxDEL register sets the number of instruction cycles before the output is driven active. If the value is greater than the duty cycle, the corresponding output remains inactive during the entire cycle. For more details on the dead-band delay operations, see Section 20.4.6 "Programmable Dead-Band Delay Mode". Since the PxA and PxB outputs are multiplexed with the port data latches, the associated TRIS bits must be cleared to configure PxA and PxB as outputs.





**<sup>2:</sup>** Output signals are shown as active-high.

## FIGURE 20-7: EXAMPLE OF HALF-BRIDGE APPLICATIONS



In the Reverse mode, the PxC pin is driven to its active state and the PxB pin is modulated, while the PxA and

PxD pins are driven to their inactive state, as provided

The PxA, PxB, PxC and PxD outputs are multiplexed

with the port data latches. The associated TRIS bits

must be cleared to configure the PxA, PxB, PxC and

## 20.4.2 FULL-BRIDGE MODE

In Full-Bridge mode, all four pins are used as outputs. An example of a full-bridge application is provided in Figure 20-8.

In the Forward mode, the PxA pin is driven to its active state and the PxD pin is modulated, while the PxB and PxC pins are driven to their inactive state, as provided in Figure 20-9.

FIGURE 20-8: EXAMPLE OF FULL-BRIDGE APPLICATION

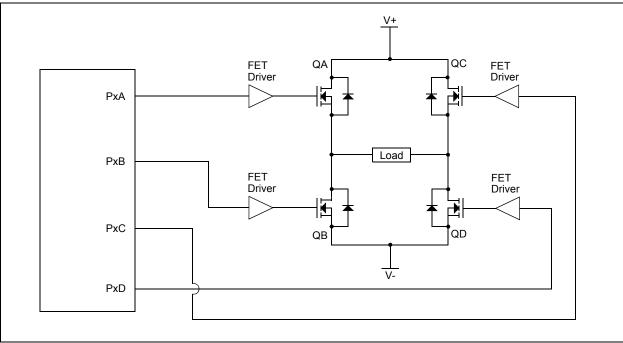
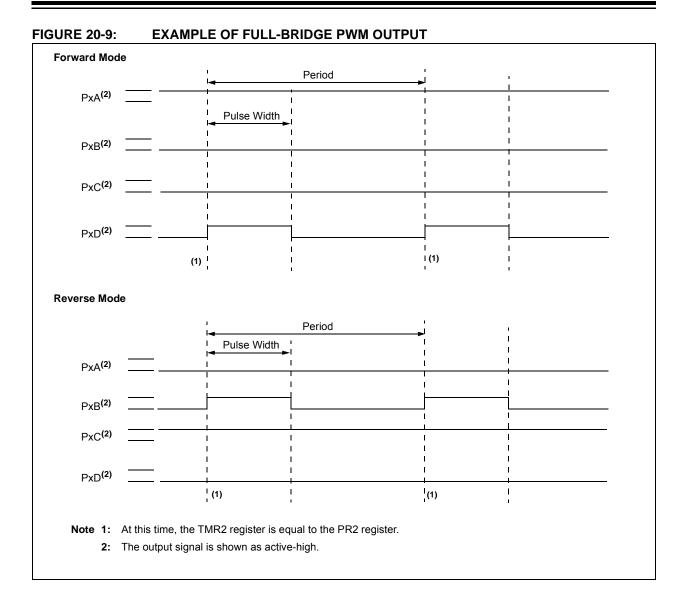


Figure 20-9.

PxD pins as outputs.



## 20.4.2.1 Direction Change in Full-Bridge Mode

In Full-Bridge mode, the PxM1 bit in the CCPxCON register allows users to control the forward/reverse direction. When the application firmware changes this direction control bit, the module will change to the new direction on the next PWM cycle.

A direction change is initiated in software by changing the PxM1 bit of the CCPxCON register. The following sequence occurs prior to the end of the current PWM period:

- The modulated outputs (PxB and PxD) are placed in their inactive state.
- The associated unmodulated outputs (PxA and PxC) are switched to drive in the opposite direction.
- PWM modulation resumes at the beginning of the next period.

For an illustration of this sequence, see Figure 20-10.

The Full-Bridge mode does not provide a dead-band delay. As one output is modulated at a time, a dead-band delay is generally not required. There is a situation where a dead-band delay is required. This situation occurs when both of the following conditions are true:

- The direction of the PWM output changes when the duty cycle of the output is at or near 100%.
- The turn-off time of the power switch, including the power device and driver circuit, is greater than the turn-on time.

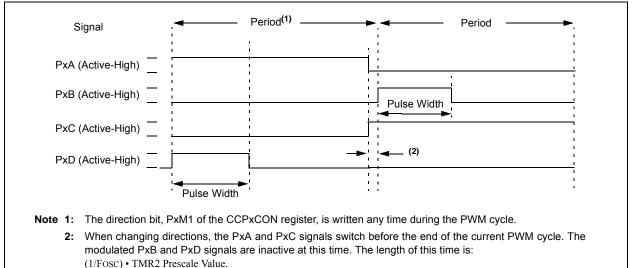
Figure 20-11 shows an example of the PWM direction changing from forward to reverse, at a near 100% duty cycle. In this example, at time, t1, the PxA and PxD outputs become inactive, while the PxC output becomes active. Since the turn-off time of the power devices is longer than the turn-on time, a shoot-through current will flow through power devices, QC and QD (see Figure 20-8), for the duration of 't'. The same phenomenon will occur to power devices, QA and QB, for PWM direction change from reverse to forward.

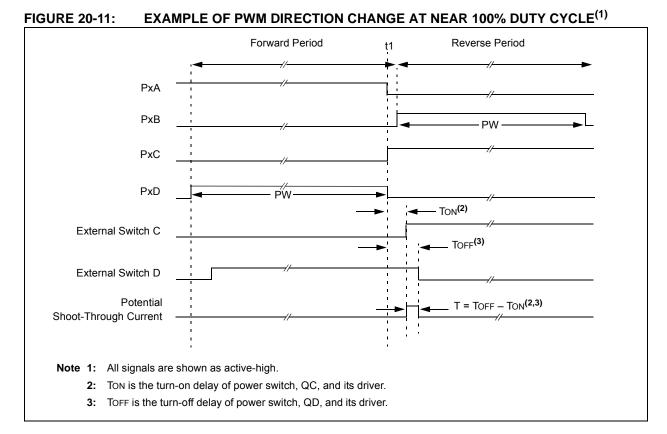
If changing PWM direction at high duty cycle is required for an application, two possible solutions for eliminating the shoot-through current are:

- Reduce PWM duty cycle for one PWM period before changing directions.
- Use switch drivers that can drive the switches off faster than they can drive them on.

Other options to prevent shoot-through current may exist.

## FIGURE 20-10: EXAMPLE OF PWM DIRECTION CHANGE





## 20.4.3 START-UP CONSIDERATIONS

When any PWM mode is used, the application hardware must use the proper external pull-up and/or pull-down resistors on the PWM output pins.

Note:	When the microcontroller is released from Reset, all of the I/O pins are in the
	· · · · · ·
	high-impedance state. The external
	circuits must keep the power switch
	devices in the OFF state until the micro-
	controller drives the I/O pins with the
	proper signal levels or activates the PWM
	output(s).

The CCPxM<1:0> bits of the CCPxCON register allow the user to choose whether the PWM output signals are active-high or active-low for each pair of PWM output pins (PxA/PxC and PxB/PxD). The PWM output polarities must be selected before the PWM pin output drivers are enabled. Changing the polarity configuration while the PWM pin output drivers are enabled is not recommended since it may result in damage to the application circuits.

The PxA, PxB, PxC and PxD output latches may not be in the proper states when the PWM module is initialized. Enabling the PWM pin output drivers at the same time as the Enhanced PWM modes may cause damage to the application circuit. The Enhanced PWM modes must be enabled in the proper Output mode and complete a full PWM cycle before enabling the PWM pin output drivers. The completion of a full PWM cycle is indicated by the TMR2IF or TMR4IF bit of the PIR1 or PIR5 register being set as the second PWM period begins.

## 20.4.4 ENHANCED PWM AUTO-SHUTDOWN MODE

The PWM mode supports an Auto-Shutdown mode that will disable the PWM outputs when an external shutdown event occurs. Auto-Shutdown mode places the PWM output pins into a predetermined state. This mode is used to help prevent the PWM from damaging the application.

The auto-shutdown sources are selected using the ECCPxAS<2:0> bits (ECCPxAS<6:4>). A shutdown event may be generated by:

- A logic '0' on the pin that is assigned the FLT0 input function
- C1 Comparator
- · C2 Comparator
- · Setting the ECCPxASE bit in firmware

A shutdown condition is indicated by the ECCPxASE (Auto-Shutdown Event Status) bit (ECCPxAS<7>). If the bit is a '0', the PWM pins are operating normally. If the bit is a '1', the PWM outputs are in the shutdown state.

When a shutdown event occurs, two things happen:

- The ECCPxASE bit is set to '1'. The ECCPxASE will remain set until cleared in firmware or an auto-restart occurs. (See Section 20.4.5 "Auto-Restart Mode".)
- The enabled PWM pins are asynchronously placed in their shutdown states. The PWM output pins are grouped into pairs (PxA/PxC) and (PxB/PxD). The state of each pin pair is determined by the PSSxAC and PSSxBD bits (ECCPxAS<3:2> and <1:0>, respectively).

Each pin pair may be placed into one of three states:

- Drive logic '1'
- Drive logic '0'
- Tri-state (high-impedance)

## REGISTER 20-3: ECCPxAS: ECCPx AUTO-SHUTDOWN CONTROL REGISTER

R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0
ECCPxASE	ECCPxAS2	ECCPxAS1	ECCPxAS0	PSSxAC1	PSSxAC0	PSSxBD1	PSSxBD0
bit 7 bit 0							

Legend:				
R = Readable bit		W = Writable bit	U = Unimplemented bit,	, read as '0'
-n = Value	at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown
bit 7	1 = A shu	<b>SE:</b> ECCP Auto-Shutdown tdown event has occurred; Poutputs are operating	Event Status bit ECCP outputs are in a shutdov	wn state
bit 6-4	ECCPxA	S<2:0>: ECCP Auto-Shutdo	own Source Select bits	
	001 <b>= Co</b>	to-shutdown is disabled mparator C1OUT output is mparator C2OUT output is	0	

- 011 = Either Comparator C1OUT or C2OUT is high
- 100 = VIL on FLTO pin
- 101 = VIL on FLT0 pin or Comparator C1OUT output is high
- 110 = VIL on FLT0 pin or Comparator C2OUT output is high
- 111 = VIL on FLT0 pin or Comparator C1OUT or Comparator C2OUT is high
- bit 3-2 **PSSxAC<1:0>:** PxA and PxC Pins Shutdown State Control bits
  - 00 = Drive pins, PxA and PxC, to '0'
  - 01 = Drive pins, PxA and PxC, to '1'
  - lx = PxA and PxC pins tri-state

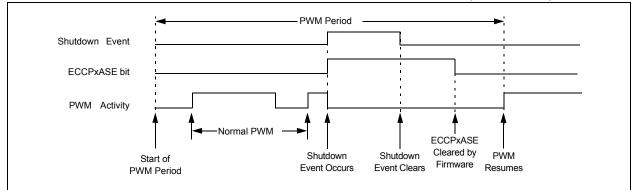
### bit 1-0 **PSSxBD<1:0>:** Pins PxB and PxD Shutdown State Control bits

- 00 = Drive pins, PxB and PxD, to '0'
- 01 = Drive pins, PxB and PxD, to '1'
- $\mathtt{lx}$  = PxB and PxD pins tri-state

## **Note 1:** The auto-shutdown condition is a level-based signal, not an edge-based signal. As long as the level is present, the auto-shutdown will persist.

- 2: Writing to the ECCPxASE bit is disabled while an auto-shutdown condition persists.
- **3:** Once the auto-shutdown condition has been removed and the PWM restarted (either through firmware or auto-restart), the PWM signal will always restart at the beginning of the next PWM period.

## FIGURE 20-12: PWM AUTO-SHUTDOWN WITH FIRMWARE RESTART (PxRSEN = 0)

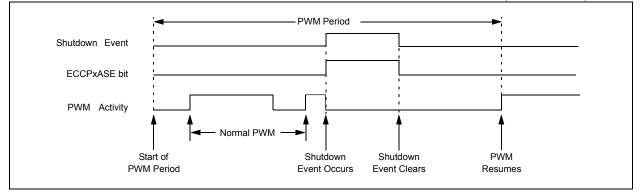


## 20.4.5 AUTO-RESTART MODE

The Enhanced PWM can be configured to automatically restart the PWM signal once the auto-shutdown condition has been removed. Auto-restart is enabled by setting the PxRSEN bit (ECCPxDEL<7>).

If auto-restart is enabled, the ECCPxASE bit will remain set as long as the auto-shutdown condition is active. When the auto-shutdown condition is removed, the ECCPxASE bit will be cleared via hardware and normal operation will resume. The module will wait until the next PWM period begins, however, before re-enabling the output pin. This behavior allows the auto-shutdown with auto-restart features to be used in applications based on current mode of PWM control.

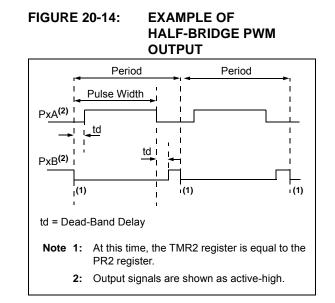
## FIGURE 20-13: PWM AUTO-SHUTDOWN WITH AUTO-RESTART ENABLED (PxRSEN = 1)



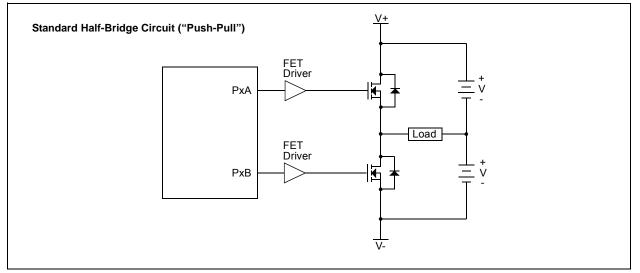
## 20.4.6 PROGRAMMABLE DEAD-BAND DELAY MODE

In half-bridge applications, where all power switches are modulated at the PWM frequency, the power switches normally require more time to turn off than to turn on. If both the upper and lower power switches are switched at the same time (one turned on and the other turned off), both switches may be on for a short period until one switch completely turns off. During this brief interval, a very high current (shoot-through current) will flow through both power switches, shorting the bridge supply. To avoid this potentially destructive shoot-through current from flowing during switching, turning on either of the power switches is normally delayed to allow the other switch to completely turn off.

In Half-Bridge mode, a digitally programmable dead-band delay is available to avoid shoot-through current from destroying the bridge power switches. The delay occurs at the signal transition from the non-active state to the active state. For an illustration, see Figure 20-14. The lower seven bits of the associated ECCPxDEL register (Register 20-4) set the delay period in terms of microcontroller instruction cycles (TcY or 4 Tosc).



## FIGURE 20-15: EXAMPLE OF HALF-BRIDGE APPLICATIONS



R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0
PxRSEN	PxDC6	PxDC5	PxDC4	PxDC3	PxDC2	PxDC1	PxDC0
bit 7							bit 0

Legend:			
R = Readable bit	W = Writable bit	U = Unimplemented bit, rea	d as '0'
-n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown

bit 7 PxRSEN: PWM Restart Enable bit

- 1 = Upon auto-shutdown, the ECCPxASE bit clears automatically once the shutdown event goes away; the PWM restarts automatically
- 0 = Upon auto-shutdown, ECCPxASE must be cleared by software to restart the PWM

bit 6-0 PxI

PxDC<6:0>: PWM Delay Count bits

PxDCn = Number of Fosc/4 (4 \* Tosc) cycles between the scheduled time when a PWM signal **should** transition active and the **actual** time it does transition active.

## 20.4.7 PULSE STEERING MODE

In Single Output mode, pulse steering allows any of the PWM pins to be the modulated signal. Additionally, the same PWM signal can simultaneously be available on multiple pins.

Once the Single Output mode is selected (CCPxM<3:2> = 11 and PxM<1:0> = 00 of the CCPxCON register), the user firmware can bring out the same PWM signal to one, two, three or four output pins by setting the appropriate STR<D:A> bits (PSTRxCON<3:0>), as provided in Table 20-3.

**Note:** The associated TRIS bits must be set to output ('0') to enable the pin output driver in order to see the PWM signal on the pin.

While the PWM Steering mode is active, the CCPxM<1:0> bits (CCPxCON<1:0>) select the PWM output polarity for the Px<D:A> pins.

The PWM auto-shutdown operation also applies to the PWM Steering mode, as described in Section 20.4.4 "Enhanced PWM Auto-shutdown mode". An auto-shutdown event will only affect pins that have PWM outputs enabled.

## REGISTER 20-5: PSTRxCON: PULSE STEERING CONTROL<sup>(1)</sup>

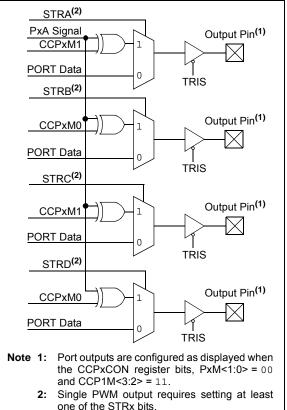
bit 7							bit 0
CMPL1	CMPL0	—	STRSYNC	STRD	STRC	STRB	STRA
R/W-0	R/W-0	U-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-1

Legend:			
R = Readable bit	W = Writable bit	U = Unimplemented bit, read	d as '0'
-n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown

CMPL<1:0>: Complementary Mode Output Assignment Steering Sync bits
00 = See STR <d:a></d:a>
01 = PA and PB are selected as the complementary output pair
10 = PA and PC are selected as the complementary output pair
11 = PA and PD are selected as the complementary output pair
Unimplemented: Read as '0'
STRSYNC: Steering Sync bit
1 = Output steering update occurs on the next PWM period
0 = Output steering update occurs at the beginning of the instruction cycle boundary
STRD: Steering Enable bit D
1 = PxD pin has the PWM waveform with polarity control from CCPxM<1:0>
0 = PxD pin is assigned to the port pin
STRC: Steering Enable bit C
1 = PxC pin has the PWM waveform with polarity control from CCPxM<1:0>
0 = PxC pin is assigned to the port pin
STRB: Steering Enable bit B
1 = PxB pin has the PWM waveform with polarity control from CCPxM<1:0>
0 = PxB pin is assigned to the port pin
STRA: Steering Enable bit A
1 = PxA pin has the PWM waveform with polarity control from CCPxM<1:0>
0 = PxA pin is assigned to the port pin
The PWM Steering mode is available only when the CCPxCON register bits, CCPxM<3:2> = 11 and

PxM<1:0> = 00.

## FIGURE 20-16: SIMPLIFIED STEERING BLOCK DIAGRAM



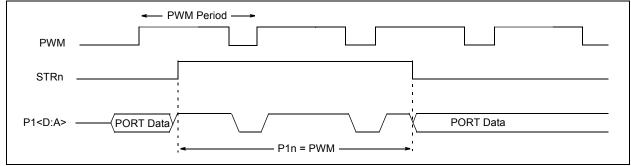
## 20.4.7.1 Steering Synchronization

The STRSYNC bit of the PSTRxCON register gives the user two choices for when the steering event will happen. When the STRSYNC bit is '0', the steering event will happen at the end of the instruction that writes to the PSTRxCON register. In this case, the output signal at the Px<D:A> pins may be an incomplete PWM waveform. This operation is useful when the user firmware needs to immediately remove a PWM signal from the pin.

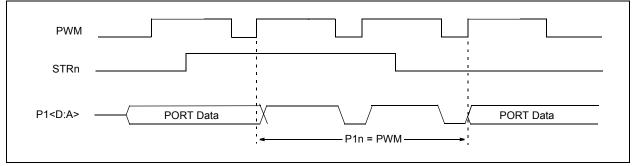
When the STRSYNC bit is '1', the effective steering update will happen at the beginning of the next PWM period. In this case, steering on/off the PWM output will always produce a complete PWM waveform.

Figures 20-17 and 20-18 illustrate the timing diagrams of the PWM steering depending on the STRSYNC setting.

## FIGURE 20-17: EXAMPLE OF STEERING EVENT AT END OF INSTRUCTION (STRSYNC = 0)



## FIGURE 20-18: EXAMPLE OF STEERING EVENT AT BEGINNING OF INSTRUCTION (STRSYNC = 1)



### 20.4.8 OPERATION IN POWER-MANAGED MODES

In Sleep mode, all clock sources are disabled. Timer2/4/6/8 will not increment and the state of the module will not change. If the ECCPx pin is driving a value, it will continue to drive that value. When the device wakes up, it will continue from this state. If Two-Speed Start-ups are enabled, the initial start-up frequency from HF-INTOSC and the postscaler may not be immediately stable.

In PRI\_IDLE mode, the primary clock will continue to clock the ECCPx module without change.

## 20.4.8.1 Operation with Fail-Safe Clock Monitor (FSCM)

If the Fail-Safe Clock Monitor (FSCM) is enabled, a clock failure will force the device into the power-managed RC\_RUN mode and the OSCFIF bit of the PIR2 register will be set. The ECCPx will then be clocked from the internal oscillator clock source, which may have a different clock frequency than the primary clock.

## 20.4.9 EFFECTS OF A RESET

Both Power-on Reset and subsequent Resets will force all ports to Input mode and the ECCP registers to their Reset states. This forces the ECCP module to reset to a state compatible with previous, non-Enhanced CCP modules used on other PIC18 and PIC16 devices.

File Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
INTCON	GIE/GIEH	PEIE/GIEL	TMR0IE	INT0IE	RBIE	TMR0IF	INT0IF	RBIF
RCON	IPEN	SBOREN	CM	RI	TO	PD	POR	BOR
PIR3	TMR5GIF	—	RC2IF	TX2IF	CTMUIF	CCP2IF	CCP1IF	RTCCIF
PIR4	CCP10IF <sup>(1)</sup>	CCP9IF <sup>(1)</sup>	CCP8IF	CCP7IF	CCP6IF	CCP5IF	CCP4IF	CCP3IF
PIE3	TMR5GIE	—	RC2IE	TX2IE	CTMUIE	CCP2IE	CCP1IE	RTCCIE
PIE4	CCP10IE <sup>(1)</sup>	CCP9IE <sup>(1)</sup>	CCP8IE	CCP7IE	CCP6IE	CCP5IE	CCP4IE	CCP3IE
IPR3	TMR5GIP	—	RC2IP	TX2IP	CTMUIP	CCP2IP	CCP1IP	RTCCIP
IPR4	CCP10IP <sup>(1)</sup>	CCP9IP <sup>(1)</sup>	CCP8IP	CCP7IP	CCP6IP	CCP5IP	CCP4IP	CCP3IP
TRISB	TRISB7	TRISB6	TRISB5	TRISB4	TRISB3	TRISB2	TRISB1	TRISB0
TRISC	TRISC7	TRISC6	TRISC5	TRISC4	TRISC3	TRISC2	TRISC1	TRISC0
TRISE	TRISE7	TRISE6	TRISE5	TRISE4	TRISE3	TRISE2	TRISE1	TRISE0
TRISH <sup>(2)</sup>	TRISH7	TRISH6	TRISH5	TRISH4	TRISH3	TRISH2	TRISH1	TRISH0
TMR1H	Timer1 Register High Byte							
TMR1L	Timer1 Register	Low Byte						
TMR2	Timer2 Register							
TMR3H	Timer3 Register	High Byte						
TMR3L	Timer3 Register	Low Byte						
TMR4	Timer4 Register							
TMR6	Timer6 Register							
TMR8	Timer8 Register							
TMR10 <sup>(1)</sup>	TMR10 Register	r						
TMR12 <sup>(1)</sup>	TMR10 Register							
PR2	Timer2 Period Register							
PR4	Timer4 Period Register							
PR6	Timer6 Period Register							
PR8	Timer8 Period Register							
PR10 <sup>(1)</sup>	Timer10 Period	Register						
PR12 <sup>(1)</sup>	Timer12 Period	Register						

### TABLE 20-4: REGISTERS ASSOCIATED WITH ECCP1/2/3 MODULE AND TIMER1/2/3/4/6/8/10/12

Note 1: Unimplemented on devices with a program memory of 32 Kbytes (PIC18F65K22 and PIC18F85K22).

**2:** Unimplemented on 64-pin devices (PIC18F6XK22), read as '0'.

## TABLE 20-4: REGISTERS ASSOCIATED WITH ECCP1/2/3 MODULE AND TIMER1/2/3/4/6/8/10/12 (CONTINUED)

File Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
T1CON	TMR1CS1	TMR1CS0	T1CKPS1	T1CKPS0	SOSCEN	T1SYNC	RD16	TMR10N
T2CON	—	T2OUTPS3	T2OUTPS2	T2OUTPS1	T2OUTPS0	TMR2ON	T2CKPS1	T2CKPS0
T3CON	TMR3CS1	TMR3CS0	T3CKPS1	T3CKPS0	SOSCEN	<b>T3SYNC</b>	RD16	TMR3ON
T4CON	—	T4OUTPS3	T4OUTPS2	T4OUTPS1	T4OUTPS0	TMR4ON	T4CKPS1	T4CKPS0
T6CON	—	T6OUTPS3	T6OUTPS2	T6OUTPS1	T6OUTPS0	TMR6ON	T6CKPS1	T6CKPS0
T8CON	—	T8OUTPS3	T8OUTPS2	T8OUTPS1	T8OUTPS0	TMR8ON	T8CKPS1	T8CKPS0
T10CON <sup>(1)</sup>	—	T10OUTPS3	T10OUTPS2	T10OUTPS1	T10OUTPS0	TMR100N	T10CKPS1	T10CKPS0
T12CON <sup>(1)</sup>	—	T12OUTPS3	T12OUTPS2	T12OUTPS1	T12OUTPS0	TMR12ON	T12CKPS1	T12CKPS0
CCPR1H	Capture/Compare/PWM Register 1 High Byte							
CCPR1L	Capture/Compa	Capture/Compare/PWM Register 1 Low Byte						
CCPR2H	Capture/Compa	Capture/Compare/PWM Register 2 High Byte						
CCPR2L	Capture/Compa	re/PWM Regis	ter 2 Low Byte	е				
CCPR3H	Capture/Compa	re/PWM Regis	ter 3 High Byt	e				
CCPR3L	Capture/Compa	re/PWM Regis	ter 3 Low Byte	е				
CCP1CON	P1M1	P1M0	DC1B1	DC1B0	CCP1M3	CCP1M2	CCP1M1	CCP1M0
CCP2CON	P2M1	P2M0	DC2B1	DC2B0	CCP2M3	CCP2M2	CCP2M1	CCP2M0
CCP3CON	CCP3MD	CCP2MD	CCP1MD	UART2MD	UART1MD	SSP2MD	SSP1MD	ADCMD
PMD0	CCP3MD	CCP2MD	CCP1MD	UART2MD	UART1MD	SSP2MD	SSP1MD	ADCMD

Note 1: Unimplemented on devices with a program memory of 32 Kbytes (PIC18F65K22 and PIC18F85K22).

2: Unimplemented on 64-pin devices (PIC18F6XK22), read as '0'.

## 21.0 MASTER SYNCHRONOUS SERIAL PORT (MSSP) MODULE

## 21.1 Master SSP (MSSP) Module Overview

The Master Synchronous Serial Port (MSSP) module is a serial interface, useful for communicating with other peripheral or microcontroller devices. These peripheral devices may be serial EEPROMs, shift registers, display drivers, A/D Converters, etc. The MSSP module can operate in one of two modes:

- Serial Peripheral Interface (SPI)
- Inter-Integrated Circuit<sup>™</sup> (I<sup>2</sup>C<sup>™</sup>)
  - Full Master mode
  - Slave mode (with general address call)

The  $I^2C$  interface supports the following modes in hardware:

- Master mode
- · Multi-Master mode
- Slave mode with 5-bit and 7-bit address masking (with address masking for both 10-bit and 7-bit addressing)

All members of the PIC18F87K22 family have two MSSP modules, designated as MSSP1 and MSSP2. Each module operates independently of the other.

Note: Throughout this section, generic references to an MSSP module in any of its operating modes may be interpreted as being equally applicable to MSSP1 or MSSP2. Register names and module I/O signals use the generic designator 'x' to indicate the use of a numeral to distinguish a particular module when required. Control bit names are not individuated.

## 21.2 Control Registers

Each MSSP module has three associated control registers. These include a status register (SSPxSTAT) and two control registers (SSPxCON1 and SSPxCON2). The use of these registers and their individual configuration bits differ significantly depending on whether the MSSP module is operated in SPI or I<sup>2</sup>C mode.

Additional details are provided under the individual sections.

Note:	In devices with more than one MSSP
	module, it is very important to pay close
	attention to SSPxCON register names.
	SSP1CON1 and SSP1CON2 control
	different operational aspects of the same
	module, while SSP1CON1 and
	SSP2CON1 control the same features for
	two different modules.

Note: The SSPxBUF register cannot be used with read-modify-write instructions, such as BCF, COMF, etc.

To avoid lost data in Master mode, a read of the SSPxBUF must be performed to clear the Buffer Full (BF) detect bit (SSPSTAT<0>) between each transmission.

## 21.3 SPI Mode

The SPI mode allows 8 bits of data to be synchronously transmitted and received simultaneously. All four modes of SPI are supported. To accomplish communication, typically three pins are used:

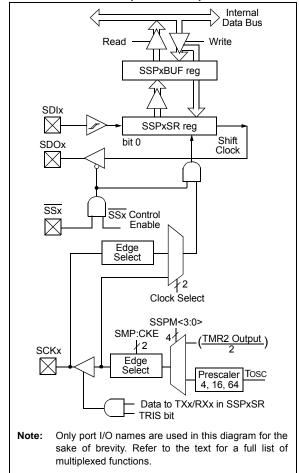
- Serial Data Out (SDOx) RC5/SDO1 or RD4/PSP4/SDO2
- Serial Data In (SDIx) RC4/SDI1/SDA1 or RD5/PSP5/SDI2/SDA2
- Serial Clock (SCKx) RC3/SCK1/SCL1 or RD6/PSP6/SCK2/SCL2

Additionally, a fourth pin may be used when in a Slave mode of operation:

• Slave Select (SSx) – RF7/AN5/SS1 or RD7/SS2

Figure 21-1 shows the block diagram of the MSSP module when operating in SPI mode.

## FIGURE 21-1: MSSP BLOCK DIAGRAM (SPI MODE)



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## 21.3.1 REGISTERS

Each MSSP module has four registers for SPI mode operation. These are:

- MSSPx Control Register 1 (SSPxCON1)
- MSSPx Status Register (SSPxSTAT)
- Serial Receive/Transmit Buffer Register (SSPxBUF)
- MSSPx Shift Register (SSPxSR) Not directly accessible

SSPxCON1 and SSPxSTAT are the control and status registers in SPI mode operation. The SSPxCON1 register is readable and writable. The lower 6 bits of the SSPxSTAT are read-only. The upper two bits of the SSPxSTAT are read/write.

SSPxSR is the shift register used for shifting data in or out. SSPxBUF is the buffer register to which data bytes are written to or read from.

In receive operations, SSPxSR and SSPxBUF together create a double-buffered receiver. When SSPxSR receives a complete byte, it is transferred to SSPxBUF and the SSPxIF interrupt is set.

During transmission, the SSPxBUF is not double-buffered. A write to SSPxBUF will write to both SSPxBUF and SSPxSR.

## REGISTER 21-1: SSPxSTAT: MSSPx STATUS REGISTER (SPI MODE)

				•	,			
R/W-0	R/W-0	R-0	R-0	R-0	R-0	R-0	R-0	
SMP	CKE <sup>(1)</sup>	D/A	Р	S	R/W	UA	BF	
bit 7							bit C	
Legend:								
R = Readab		W = Writable	bit	•	nented bit, rea			
-n = Value a	t POR	'1' = Bit is set		'0' = Bit is cle	ared	x = Bit is unkr	lown	
=								
bit 7	SMP: Sample							
	<u>SPI Master n</u>		the end of de	to output time				
	•	a is sampled at		data output time	2			
	SPI Slave mo			und output link	-			
		e cleared when	SPI is used in	Slave mode.				
bit 6	CKE: SPI CI	ock Select bit <sup>(1)</sup>						
	1 = Transmit	occurs on the ti	ansition from	active to Idle cl	ock state			
			ansition from	Idle to active cl	ock state			
bit 5	D/A: Data/Ad							
	Used in I <sup>2</sup> C <sup>™</sup>	mode only.						
bit 4	P: Stop bit							
	Used in I <sup>2</sup> C r	Used in I <sup>2</sup> C mode only. This bit is cleared when the MSSPx module is disabled; SSPEN is cleared.						
bit 3	S: Start bit	S: Start bit						
		Used in I <sup>2</sup> C mode only.						
bit 2		Vrite Informatior	ı bit					
	Used in I <sup>2</sup> C r	Used in I <sup>2</sup> C mode only.						
bit 1	UA: Update Address bit							
	Used in I <sup>2</sup> C r	Used in I <sup>2</sup> C mode only.						
bit 0	BF: Buffer Fi	ull Status bit (Re	ceive mode o	only)				
		is complete, SS						
	0 = Receive	is not complete,	SSPxBUF is	empty				
	) a la ritu of a la alu	atata ia aat hu th						

Note 1: Polarity of clock state is set by the CKP bit (SSPxCON1<4>).

REGISTER 21-2:	SSPxCON1: MSSPx CONTROL REGISTER 1 (SPI MODE)
----------------	---

R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0
WCOL	SSPOV <sup>(1)</sup>	SSPEN <sup>(2)</sup>	CKP	SSPM3 <sup>(3)</sup>	SSPM2 <sup>(3)</sup>	SSPM1 <sup>(3)</sup>	SSPM0 <sup>(3)</sup>
bit 7					•		bit (
Legend:							
R = Readab	ole bit	W = Writable I	oit	U = Unimplen	nented bit, rea	d as '0'	
-n = Value a	at POR	'1' = Bit is set		'0' = Bit is clea	ared	x = Bit is unkr	nown
bit 7		)		e it is still transm	nitting the prev	ious word (mus	t be cleared in
bit 6	<u>SPI Slave mo</u> 1 = A new by overflow,	yte is received the data in SSI xBUF, even if ).	while the SSI PxSR is lost.	PxBUF register Overflow can on tting data, to a	ly occur in Sla	ve mode. The u	user must rea
bit 5	1 = Enables		onfigures SC	Kx, SDOx, SDIx		erial port pins	
bit 4	<ul> <li>0 = Disables serial port and configures these pins as I/O port pins</li> <li>CKP: Clock Polarity Select bit</li> <li>1 = Idle state for the clock is a high level</li> <li>0 = Idle state for the clock is a low level</li> </ul>						
bit 3-0	1010 = SPI M 0101 = SPI S 0100 = SPI S 0011 = SPI M 0010 = SPI M 0001 = SPI M	laster mode: clo lave mode: clo	bck = Fosc/8 ck = SCKx pirch $ck = SCKx pirch bck = TMR2 cbck = Fosc/6bck = Fosc/10$	4	ol disabled; $\overline{SS}$	Sx can be used	as I/O pin
	n Master mode, t		is not set sind	ce each new rec	eption (and tra	ansmission) is i	nitiated by
	writing to the SSPxBUF register.						

- 2: When enabled, these pins must be properly configured as inputs or outputs.
- **3:** Bit combinations not specifically listed here are either reserved or implemented in  $I^2C^{TM}$  mode only.

## 21.3.2 OPERATION

When initializing the SPI, several options need to be specified. This is done by programming the appropriate control bits (SSPxCON1<5:0> and SSPxSTAT<7:6>). These control bits allow the following to be specified:

- Master mode (SCKx is the clock output)
- Slave mode (SCKx is the clock input)
- Clock Polarity (Idle state of SCKx)
- Data Input Sample Phase (middle or end of data output time)
- Clock Edge (output data on rising/falling edge of SCKx)
- Clock Rate (Master mode only)
- · Slave Select mode (Slave mode only)

Each MSSP module consists of a Transmit/Receive Shift register (SSPxSR) and a Serial Input Buffer register (SSPxBUF). The SSPxSR shifts the data in and out of the device, MSb first. The SSPxBUF holds the data that was written to the SSPxSR until the received data is ready. Once the 8 bits of data have been received, that byte is moved to the SSPxBUF register. Then, the Buffer Full detect bit, BF (SSPxSTAT<0>), and the interrupt flag bit, SSPxIF, are set. This double-buffering of the received data (SSPxBUF) allows the next byte to start reception before reading the data that was just received. Any write to the SSPxBUF register during transmission/reception of data will be ignored and the Write Collision Detect bit, WCOL (SSPxCON1<7>), will be set. User software must clear the WCOL bit so that it can be determined if the following write(s) to the SSPxBUF register completed successfully.

When the application software is expecting to receive valid data, the SSPxBUF should be read before the next byte of data to transfer is written to the SSPxBUF. The Buffer Full bit, BF (SSPxSTAT<0>), indicates when SSPxBUF has been loaded with the received data (transmission is complete). When the SSPxBUF is read, the BF bit is cleared. This data may be irrelevant if the SPI is only a transmitter. Generally, the MSSP interrupt is used to determine when the transmission/reception has completed. If the interrupt method is not going to be used, then software polling can be done to ensure that a write collision does not occur. Example 21-1 shows the loading of the SSPxBUF (SSPxSR) for data transmission.

The SSPxSR is not directly readable or writable and can only be accessed by addressing the SSPxBUF register. Additionally, the SSPxSTAT register indicates the various status conditions.

## 21.3.3 OPEN-DRAIN OUTPUT OPTION

The drivers for the SDOx output and SCKx clock pins can be optionally configured as open-drain outputs. This feature allows the voltage level on the pin to be pulled to a higher level through an external pull-up resistor, and allows the output to communicate with external circuits without the need for additional level shifters. For more information, see Section 12.1.3 "Open-Drain Outputs".

The open-drain output option is controlled by the SSP2OD and SSP1OD bits (ODCON3<1:0>). Setting an SSPxOD bit configures the SDOx and SCKx pins for the corresponding module for open-drain operation.

Note: To avoid lost data in Master mode, a read of the SSPxBUF must be performed to clear the Buffer Full (BF) detect bit (SSPxSTAT<0>) between each transmission.

### EXAMPLE 21-1: LOADING THE SSP1BUF (SSP1SR) REGISTER

LOOP	BTFSS BRA MOVF	SSP1STAT, BF LOOP SSP1BUF, W	<pre>;Has data been received (transmit complete)? ;No ;WREG reg = contents of SSP1BUF</pre>
	MOVWF	RXDATA	;Save in user RAM, if data is meaningful
	MOVF MOVWF	TXDATA, W SSP1BUF	;W reg = contents of TXDATA ;New data to xmit

## 21.3.4 ENABLING SPI I/O

To enable the serial port, MSSP Enable bit, SSPEN (SSPxCON1<5>), must be set. To reset or reconfigure SPI mode, clear the SSPEN bit, re-initialize the SSPxCON registers and then set the SSPEN bit. This configures the SDIx, SDOx, SCKx and SSx pins as serial port pins. For the pins to behave as the serial port function, some must have their data direction bits (in the TRIS register) appropriately programmed as follows:

- SDIx must have the TRISC<4> or TRISD<5> bit set
- SDOx must have the TRISC<5> or TRISD<4> bit cleared
- SCKx (Master mode) must have the TRISC<3> or TRISD<6>bit cleared
- SCKx (Slave mode) must have the TRISC<3> or TRISD<6> bit set
- SSx must have the TRISF<7> or TRISD<7> bit set

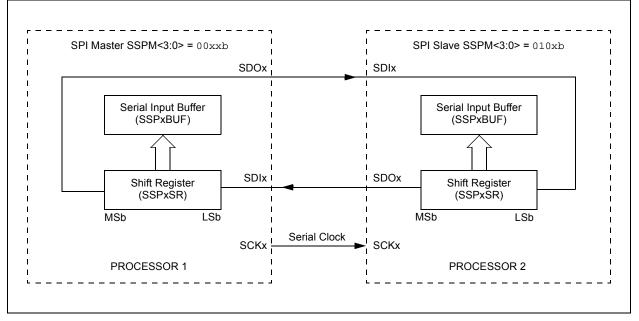
Any serial port function that is not desired may be overridden by programming the corresponding Data Direction (TRIS) register to the opposite value.

## 21.3.5 TYPICAL CONNECTION

Figure 21-2 shows a typical connection between two microcontrollers. The master controller (Processor 1) initiates the data transfer by sending the SCKx signal. Data is shifted out of both shift registers on their programmed clock edge and latched on the opposite edge of the clock. Both processors should be programmed to the same Clock Polarity (CKP), then both controllers would send and receive data at the same time. Whether the data is meaningful (or dummy data) depends on the application software. This leads to three scenarios for data transmission:

- · Master sends data Slave sends dummy data
- Master sends data Slave sends data
- Master sends dummy data Slave sends data

## FIGURE 21-2: SPI MASTER/SLAVE CONNECTION



## 21.3.6 MASTER MODE

The master can initiate the data transfer at any time because it controls the SCKx. The master determines when the slave (Processor 1, Figure 21-2) is to broadcast data by the software protocol.

In Master mode, the data is transmitted/received as soon as the SSPxBUF register is written to. If the SPI is only going to receive, the SDOx output could be disabled (programmed as an input). The SSPxSR register will continue to shift in the signal present on the SDIx pin at the programmed clock rate. As each byte is received, it will be loaded into the SSPxBUF register as if a normal received byte (interrupts and status bits appropriately set). This could be useful in receiver applications as a "Line Activity Monitor" mode.

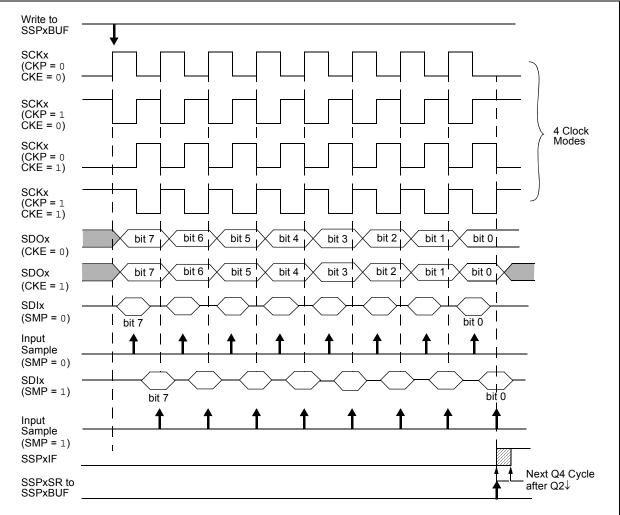
The clock polarity is selected by appropriately programming the CKP bit (SSPxCON1<4>). This, then, would give waveforms for SPI communication, as

shown in Figure 21-3, Figure 21-5 and Figure 21-6, where the MSB is transmitted first. In Master mode, the SPI clock rate (bit rate) is user-programmable to be one of the following:

- Fosc/4 (or Tcy)
- Fosc/16 (or 4 Tcy)
- Fosc/64 (or 16 Tcy)
- Timer2 output/2

This allows a maximum data rate (at 40 MHz) of 10.00 Mbps.

Figure 21-3 shows the waveforms for Master mode. When the CKE bit is set, the SDOx data is valid before there is a clock edge on SCKx. The change of the input sample is shown based on the state of the SMP bit. The time when the SSPxBUF is loaded with the received data is shown.





## 21.3.7 SLAVE MODE

In Slave mode, the data is transmitted and received as the external clock pulses appear on SCKx. When the last bit is latched, the SSPxIF interrupt flag bit is set.

While in Slave mode, the external clock is supplied by the external clock source on the SCKx pin. This external clock must meet the minimum high and low times as specified in the electrical specifications.

While in Sleep mode, the slave can transmit/receive data. When a byte is received, the device can be configured to wake up from Sleep.

## 21.3.8 SLAVE SELECT SYNCHRONIZATION

The  $\overline{SSx}$  pin allows a Synchronous Slave mode. The SPI must be in Slave mode with the  $\overline{SSx}$  pin control enabled (SSPxCON1<3:0> = 04h). When the  $\overline{SSx}$  pin is low, transmission and reception are enabled and the SDOx pin is driven. When the  $\overline{SSx}$  pin goes high, the SDOx pin is no longer driven, even if in the middle of a

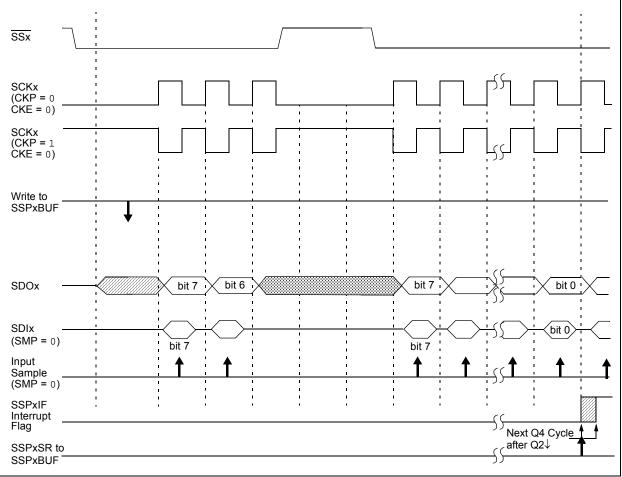
transmitted byte and becomes a floating output. External pull-up/pull-down resistors may be desirable depending on the application.

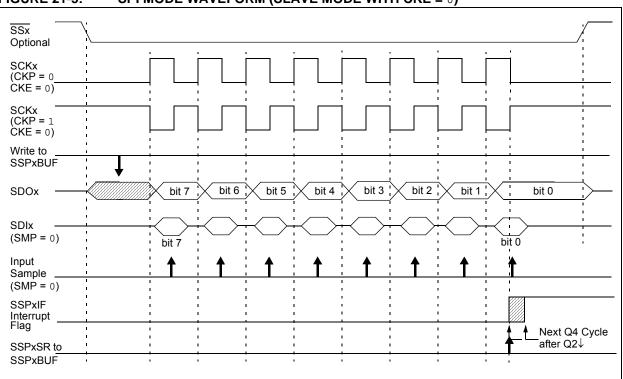
Note: When the SPI is in Slave mode with the SSx pin control enabled (SSPxCON1<3:0> = 0100), the SPI module will reset if the SSx pin is set to VDD.
 If the SPI is used in Slave mode with CKE set, then the SSx pin control must be enabled.

When the SPI module resets, the bit counter is forced to '0'. This can be done by either forcing the SSx pin to a high level or clearing the SSPEN bit.

To emulate two-wire communication, the SDOx pin can be connected to the SDIx pin. When the SPI needs to operate as a receiver, the SDOx pin can be configured as an input. This disables transmissions from the SDOx. The SDIx can always be left as an input (SDIx function) since it cannot create a bus conflict.

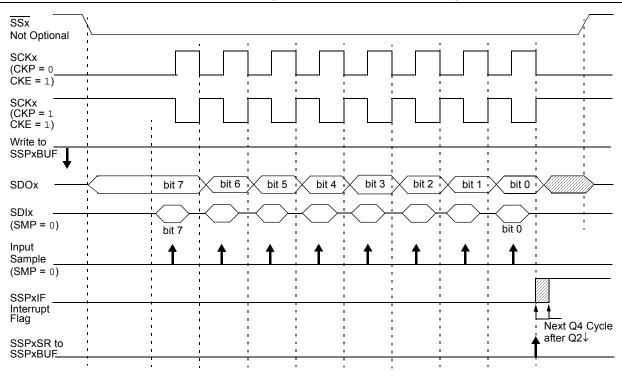






## FIGURE 21-5: SPI MODE WAVEFORM (SLAVE MODE WITH CKE = 0)





#### 21.3.9 OPERATION IN POWER-MANAGED MODES

In SPI Master mode, module clocks may be operating at a different speed than when in Full-Power mode. In the case of Sleep mode, all clocks are halted.

In Idle modes, a clock is provided to the peripherals. That clock can be from the primary clock source, the secondary clock (SOSC oscillator) or the INTOSC source. See **Section 3.3** "**Clock Sources and Oscillator Switching**" for additional information.

In most cases, the speed that the master clocks SPI data is not important; however, this should be evaluated for each system.

If MSSP interrupts are enabled, they can wake the controller from Sleep mode, or one of the Idle modes, when the master completes sending data. If an exit from Sleep or Idle mode is not desired, MSSP interrupts should be disabled.

If the Sleep mode is selected, all module clocks are halted and the transmission/reception will remain in that state until the device wakes. After the device returns to Run mode, the module will resume transmitting and receiving data.

In SPI Slave mode, the SPI Transmit/Receive Shift register operates asynchronously to the device. This allows the device to be placed in any power-managed mode and data to be shifted into the SPI Transmit/Receive Shift register. When all 8 bits have been received, the MSSP interrupt flag bit will be set, and if enabled, will wake the device.

#### 21.3.10 EFFECTS OF A RESET

A Reset disables the MSSP module and terminates the current transfer.

#### 21.3.11 BUS MODE COMPATIBILITY

Table 21-1shows the compatibility between thestandard SPI modes and the states of the CKP andCKE control bits.

Standard SPI Mode	Control Bits State				
Terminology	СКР	CKE			
0, 0	0	1			
0, 1	0	0			
1, 0	1	1			
1, 1	1	0			

There is also an SMP bit which controls when the data is sampled.

### 21.3.12 SPI CLOCK SPEED AND MODULE INTERACTIONS

Because MSSP1 and MSSP2 are independent modules, they can operate simultaneously at different data rates. Setting the SSPM<3:0> bits of the SSPxCON1 register determines the rate for the corresponding module.

An exception is when both modules use Timer2 as a time base in Master mode. In this instance, any changes to the Timer2 module's operation will affect both MSSP modules equally. If different bit rates are required for each module, the user should select one of the other three time base options for one of the modules.

TABLE 21-2. REGISTERS ASSOCIATED WITH SPICE RATION									
Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	
INTCON	GIE/GIEH	PEIE/GIEL	TMR0IE	INT0IE	RBIE	TMR0IF	INT0IF	RBIF	
PIR2	OSCFIF	—	SSP2IF	BCL2IF	BCL1IF	HLVDIF	TMR3IF	TMR3GIF	
PIE2	OSCFIE	—	SSP2IE	BCL2IE	BCL1IE	HLVDIE	TMR3IE	TMR3GIE	
IPR2	OSCFIP	—	SSP2IP	BCL2IP	BCL1IP	HLVDIP	TMR3IP	TMR3GIP	
PIR3	TMR5GIF	—	RC2IF	TX2IF	CTMUIF	CCP2IF	CCP1IF	RTCCIF	
PIE3	TMR5GIE	—	RC2IE	TX2IE	CTMUIE	CCP2IE	CCP1IE	RTCCIE	
IPR3	TMR5GIP	—	RC2IP	TX2IP	CTMUIP	CCP2IP	CCP1IP	RTCCIP	
TRISC	TRISC7	TRISC6	TRISC5	TRISC4	TRISC3	TRISC2	TRISC1	TRISC0	
TRISD	TRISD7	TRISD6	TRISD5	TRISD4	TRISD3	TRISD2	TRISD1	TRISD0	
TRISF	TRISF7	TRISF6	TRISF5	TRISF4	TRISF3	TRISF2	TRISF1	_	
SSP1BUF	MSSP1 Red	eive Buffer/T	ransmit Reg	ister					
SSP1CON1	WCOL	SSPOV	SSPEN	CKP	SSPM3	SSPM2	SSPM1	SSPM0	
SSP1CON2	GCEN	ACKSTAT	ACKDT	ACKEN	RCEN	PEN	RSEN	SEN	
SSP1STAT	SMP	CKE	D/A	Р	S	R/W	UA	BF	
SSP2CON1	WCOL	SSPOV	SSPEN	CKP	SSPM3	SSPM2	SSPM1	SSPM0	
SSP2CON2	GCEN	ACKSTAT	ACKDT	ACKEN	RCEN	PEN	RSEN	SEN	
SSP2STAT	SMP	CKE	D/A	Р	S	R/W	UA	BF	
SSP2BUF	MSSP2 Red	eive Buffer/T	ransmit Reg	ister					
ODCON1	SSP10D	CCP2OD	CCP10D	—	—	—	—	SSP2OD	
PMD0	CCP3MD	CCP2MD	CCP1MD	UART2MD	UART1MD	SSP2MD	SSP1MD	ADCMD	

#### TABLE 21-2: REGISTERS ASSOCIATED WITH SPI OPERATION

**Legend:** Shaded cells are not used by the MSSP module in SPI mode.

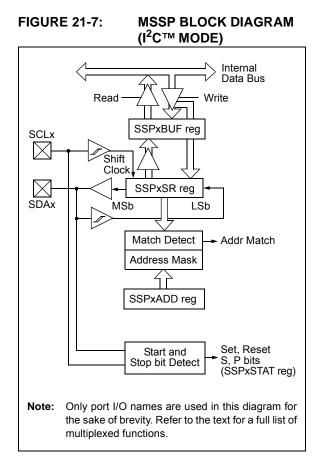
### 21.4 I<sup>2</sup>C Mode

The MSSP module in I<sup>2</sup>C mode fully implements all master and slave functions (including general call support), and provides interrupts on Start and Stop bits in hardware to determine a free bus (multi-master function). The MSSP module implements the standard mode specifications, as well as 7-bit and 10-bit addressing.

Two pins are used for data transfer:

- Serial Clock (SCLx) RC3/SCK1/SCL1 or RD6/SCK2/SCL2
- Serial Data (SDAx) RC4/SDI1/SDA1 or RD5/SDI2/SDA2

The user must configure these pins as inputs by setting the associated TRIS bits.



#### 21.4.1 REGISTERS

The MSSP module has seven registers for  ${\rm I}^2{\rm C}$  operation. These are:

- MSSPx Control Register 1 (SSPxCON1)
- MSSPx Control Register 2 (SSPxCON2)
- MSSPx Status Register (SSPxSTAT)
- Serial Receive/Transmit Buffer Register (SSPxBUF)
- MSSPx Shift Register (SSPxSR) Not directly accessible
- MSSPx Address Register (SSPxADD)
- I<sup>2</sup>C Slave Address Mask Register (SSPxMSK)

SSPxCON1, SSPxCON2 and SSPxSTAT are the control and status registers in  $I^2C$  mode operation. The SSPxCON1 and SSPxCON2 registers are readable and writable. The lower 6 bits of the SSPxSTAT are read-only. The upper two bits of the SSPxSTAT are read/write.

SSPxSR is the shift register used for shifting data in or out. SSPxBUF is the buffer register to which data bytes are written to or read from.

SSPxADD contains the slave device address when the MSSP is configured in  $I^2C$  Slave mode. When the MSSP is configured in Master mode, the lower seven bits of SSPxADD act as the Baud Rate Generator reload value.

SSPxMSK holds the slave address mask value when the module is configured for 7-Bit Address Masking mode. While it is a separate register, it shares the same SFR address as SSPxADD; it is only accessible when the SSPM<3:0> bits are specifically set to permit access. Additional details are provided in Section 21.4.3.4 "7-Bit Address Masking Mode".

In receive operations, SSPxSR and SSPxBUF together, create a double-buffered receiver. When SSPxSR receives a complete byte, it is transferred to SSPxBUF and the SSPxIF interrupt is set.

During transmission, the SSPxBUF is not double-buffered. A write to SSPxBUF will write to both SSPxBUF and SSPxSR.

R/W-0	R/W-0	R-0	R-0	R-0	R-0	R-0	R-0		
SMP	CKE	D/Ā	P <sup>(1)</sup>	S <sup>(1)</sup>	R/W <sup>(2,3)</sup>	UA	BF		
bit 7							bit		
Legend:									
R = Readab	le hit	W = Writable	hit	II = I Inimple	mented bit, rea	d as '0'			
-n = Value a		'1' = Bit is set		'0' = Bit is cl		x = Bit is unkr	NOWD		
					carco				
bit 7	SMP: Slew	Rate Control bit							
	1 = Slew ra	<u>Slave mode:</u> te control is disa te control is ena				d 1 MHz)			
bit 6	CKE: SMBu								
	1 = Enable S	<u>Slave mode:</u> SMBus-specific SMBus-specific							
bit 5	D/A: Data/A	ddress bit							
	<u>In Master m</u> Reserved.	<u>ode:</u>							
	In Slave mo		4	4					
		s that the last by s that the last by							
bit 4		<ul> <li>Indicates that the last byte received or transmitted was address</li> <li>P: Stop bit<sup>(1)</sup></li> </ul>							
		s that a Stop bit was not detecte		cted last					
bit 3	S: Start bit <sup>(1</sup>	)							
		s that a Start bit was not detecte		cted last					
bit 2	R/W: Read/	Write Information	n bit <mark>(2,3)</mark>						
	<u>In Slave mo</u> 1 = Read 0 = Write	<u>de:</u>							
	<u>In Master m</u> 1 = Transmi	<u>ode:</u> t is in progress t is not in progre	SS						
bit 1	UA: Update Address bit (10-Bit Slave mode only)								
		s that the user n does not need t		e the address i	n the SSPxADI	D register			
bit 0	BF: Buffer Full Status bit								
	<u>In Transmit</u> 1 = SSPxBU 0 = SSPxBU	JF is full							
		<u>node:</u> JF is full (does n JF is empty (doe							
Note 1: 7	his bit is cleare	ed on Reset and	when SSPEN	is cleared.					
		e R/W bit inform			ess match. This	bit is only valid	from the		
		o the next Start				•			

### REGISTER 21-3: SSPxSTAT: MSSPx STATUS REGISTER (I<sup>2</sup>C<sup>™</sup> MODE)

2: This bit holds the R/W bit information following the last address match. This bit is only valid from the address match to the next Start bit, Stop bit or not ACK bit.

3: ORing this bit with SEN, RSEN, PEN, RCEN or ACKEN will indicate if the MSSPx is in Active mode.

### REGISTER 21-4: SSPxCON1: MSSPx CONTROL REGISTER 1 (I<sup>2</sup>C<sup>™</sup> MODE)

R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0
WCOL	SSPOV	SSPEN <sup>(1)</sup>	CKP	SSPM3 <sup>(2)</sup>	SSPM2 <sup>(2)</sup>	SSPM1 <sup>(2)</sup>	SSPM0 <sup>(2)</sup>
bit 7							bit (
Legend:							
R = Reada	able bit	W = Writable b	it	U = Unimple	mented bit, rea	d as '0'	
-n = Value		'1' = Bit is set		'0' = Bit is cle		x = Bit is unk	nown
		1 Bit lo oot					
bit 7	WCOL: Write	e Collision Detect	bit				
		<u>ansmit mode:</u>			2		
		to the SSPxBUF				onditions were	not valid for a
	transmis 0 = No collis	ssion to be started	d (must be c	leared in softwa	are)		
	In Slave Trar						
		PxBUF register is	written while	e it is still trans	mitting the prev	vious word (mus	st be cleared in
	software	-			5		
	0 = No collis	sion					
	<u>In Receive m</u> This is a "doi	<u>node (Master or S</u> n't care" bit.	lave modes	<u>):</u>			
bit 6		ceive Overflow Ind	dicator bit				
	In Receive m	node:					
	-	s received while the	ne SSPxBUF	<sup>-</sup> register is still	holding the pre	vious byte (mu	st be cleared i
	software	/					
	0 = No over In Transmit r						
		n't care" bit in Tra	nsmit mode				
bit 5	SSPEN: Mas	ster Synchronous	Serial Port	Enable bit <sup>(1)</sup>			
	1 = Enables	the serial port and serial port and co	d configures	the SDAx and		he serial port p	ins
bit 4		Release Control I	-		port pino		
	In Slave mod						
	1 = Releases						
	0 = Holds clo	ock low (clock stre	etch), used t	o ensure data s	etup time		
	In Master mo Unused in th						
bit 3-0		: Master Synchro	nous Serial	Port Mode Sele	ect bits <sup>(2)</sup>		
		Slave mode: 10-b				enabled	
	1110 = I <sup>2</sup> C S	Slave mode: 7-bit	address wit	h Start and Stop	bit interrupts	enabled	
		Firmware Controll					
		SSPMSK registe					
		Master mode: cloo Slave mode: 10-b		(SSPXADD 1	- 1))		
		Slave mode: 7-bit					
Note 1:	When enabled, t	he SDAx and SC	Lx pins mus	t be configured	as inputs.		
2:		not specifically li	•	•	•	ted in SPI mod	e only.
3:	When SSPM<3:0	0> = 1001, any re					-
	SSPxMSK regist	ter.					
4:	TTUES A REPORT OF	y available when	<b>7</b> D'' A ' '				

R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0
GCEN	ACKSTAT	ACKDT <sup>(1)</sup>	ACKEN <sup>(2)</sup>	RCEN <sup>(2)</sup>	PEN <sup>(2)</sup>	RSEN <sup>(2)</sup>	SEN <sup>(2)</sup>
bit 7							bit
Legend:							
R = Reada	able bit	W = Writable	bit	U = Unimpler	nented bit, rea	d as '0'	
-n = Value	at POR	'1' = Bit is set		'0' = Bit is cle		x = Bit is unkn	own
bit 7	GCEN: Gene	ral Call Enable	bit				
	Unused in Ma	aster mode.					
bit 6	ACKSTAT: A	cknowledge Sta	atus bit (Master	Transmit mod	e only)		
		dge was not re dge was receiv		ave			
bit 5	ACKDT: Ackr	nowledge Data	bit (Master Red	ceive mode on	ly) <sup>(1)</sup>		
	1 = Not Ackno 0 = Acknowle	Ų					
bit 4	ACKEN: Ack	nowledge Sequ	ience Enable b	it <sup>(2)</sup>			
	Automati	Acknowledge cally cleared by	/ hardware.	SDAx and S	CLx pins and	I transmits AC	KDT data bi
bit 3		ve Enable bit (l		e mode only)(2)	)		
		Receive mode f		inclue entry			
bit 2		ondition Enable	bit <sup>(2)</sup>				
Sit 2		Stop condition o		CLx pins. Auto	matically clear	ed by hardware	
bit 1	RSEN: Repea	ated Start Conc	lition Enable bi	t <mark>(2)</mark>			
		Repeated Start d Start condition		DAx and SCLx	pins. Automa	tically cleared by	y hardware.
bit 0	SEN: Start Co	ondition Enable	bit <sup>(2)</sup>				
	1 = Initiates S	tart condition o		Clyping Auto	matically clear	ed by hardware	

### REGISTER 21-5: SSPxCON2: MSSPx CONTROL REGISTER 2 (I<sup>2</sup>C<sup>™</sup> MASTER MODE)

2: If the l<sup>2</sup>C module is active, these bits may not be set (no spooling) and the SSPxBUF may not be written (or writes to the SSPxBUF are disabled).

R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0
GCEN	ACKSTAT	ACKDT <sup>(1)</sup>	ACKEN <sup>(1)</sup>	RCEN <sup>(1)</sup>	PEN <sup>(1)</sup>	RSEN <sup>(1)</sup>	SEN <sup>(1)</sup>
bit 7							bit (
Legend:							
R = Readabl	le bit	W = Writable	bit	U = Unimplem	nented bit, rea	d as '0'	
-n = Value at	POR	'1' = Bit is set		'0' = Bit is clea	ared	x = Bit is unkn	own
bit 7	1 = Enables i	ral Call Enable nterrupt when a all address is c	general call a	ddress (0000h)	is received in	the SSPxSR	
bit 6	ACKSTAT: A Unused in Sla	cknowledge Sta ave mode.	atus bit				
bit 5	ACKDT: Ackr 1 = Not Ackro 0 = Acknowle	owledge	bit (Master Re	ceive mode onl	y) <sup>(1)</sup>		
bit 4	1 = Initiates Automati	nowledge Sequ Acknowledge cally cleared by edge sequence	sequence on / hardware.		CLx pins and	transmits ACk	(DT data bi
bit 3	RCEN: Recei 1 = Enables F 0 = Receive is	Receive mode f	Master Receive or I <sup>2</sup> C	e mode only) <sup>(1)</sup>			
bit 2				CLx pins. Autor	natically clear	ed by hardware.	
bit 1	1 = Initiates I	ated Start Conc Repeated Start d Start condition	condition on S		pins. Automa	tically cleared by	/ hardware.
bit 0	SEN: Stretch 1 = Clock stre 0 = Clock stre	etching is enabl		ve transmit and	slave receive	e (stretch enable	d)

**Note 1:** If the I<sup>2</sup>C module is active, this bit may not be set (no spooling) and the SSPxBUF may not be written (or writes to the SSPxBUF are disabled).

#### REGISTER 21-7: SSPxMSK: I<sup>2</sup>C<sup>™</sup> SLAVE ADDRESS MASK REGISTER (7-BIT MASKING MODE)<sup>(1)</sup>

					•		,	
R/W-1	R/W-1	R/W-1	R/W-1	R/W-1	R/W-1	R/W-1	R/W-1	
MSK7	MSK6	MSK5	MSK4	MSK3	MSK2	MSK1	MSK0 <sup>(2)</sup>	
bit 7							bit 0	
Legend:								
R = Readable bit W = Writable bit				U = Unimplemented bit, read as '0'				
-n = Value at POR '1' = Bit is set			'0' = Bit is clea	ared	x = Bit is unki	nown		

bit 7-0 MSK<7:0>: Slave Address Mask Select bit

1 = Masking of corresponding bit of SSPxADD is enabled

0 = Masking of corresponding bit of SSPxADD is disabled

Note 1: This register shares the same SFR address as SSPxADD and is only addressable in select MSSPx operating modes. See Section 21.4.3.4 "7-Bit Address Masking Mode" for more details.

**2:** MSK0 is not used as a mask bit in 7-bit addressing.

#### 21.4.2 OPERATION

The MSSP module functions are enabled by setting the MSSP Enable bit, SSPEN (SSPxCON1<5>).

The SSPxCON1 register allows control of the  $I^2C$  operation. Four mode selection bits (SSPxCON1<3:0>) allow one of the following  $I^2C$  modes to be selected:

- I<sup>2</sup>C Master mode, clock
- I<sup>2</sup>C Slave mode (7-bit address)
- I<sup>2</sup>C Slave mode (10-bit address)
- I<sup>2</sup>C Slave mode (7-bit address) with Start and Stop bit interrupts enabled
- I<sup>2</sup>C Slave mode (10-bit address) with Start and Stop bit interrupts enabled
- I<sup>2</sup>C Firmware Controlled Master mode, slave is Idle

Selection of any I<sup>2</sup>C mode with the SSPEN bit set forces the SCLx and SDAx pins to be open-drain, provided these pins are programmed as inputs by setting the appropriate TRISC or TRISD bits. To ensure proper operation of the module, pull-up resistors must be provided externally to the SCLx and SDAx pins.

#### 21.4.3 SLAVE MODE

In Slave mode, the SCLx and SDAx pins must be configured as inputs (TRISC<4:3> set). The MSSP module will override the input state with the output data when required (slave-transmitter).

The  $I^2C$  Slave mode hardware will always generate an interrupt on an address match. Address masking will allow the hardware to generate an interrupt for more than one address (up to 31 in 7-bit addressing and up to 63 in 10-bit addressing). Through the mode select bits, the user can also choose to interrupt on Start and Stop bits.

When an address is matched, or the data transfer after an address match is received, the hardware automatically will generate the Acknowledge (ACK) pulse and load the SSPxBUF register with the received value currently in the SSPxSR register.

Any combination of the following conditions will cause the MSSP module not to give this ACK pulse:

- The Buffer Full bit, BF (SSPxSTAT<0>), was set before the transfer was received.
- The overflow bit, SSPOV (SSPxCON1<6>), was set before the transfer was received.

In this case, the SSPxSR register value is not loaded into the SSPxBUF, but bit, SSPxIF, is set. The BF bit is cleared by reading the SSPxBUF register, while bit, SSPOV, is cleared through software.

The SCLx clock input must have a minimum high and low for proper operation. The high and low times of the  $I^2C$  specification, as well as the requirement of the MSSP module, are shown in timing Parameter 100 and Parameter 101.

#### 21.4.3.1 Addressing

Once the MSSP module has been enabled, it waits for a Start condition to occur. Following the Start condition, the 8 bits are shifted into the SSPxSR register. All incoming bits are sampled with the rising edge of the clock (SCLx) line. The value of register, SSPxSR<7:1>, is compared to the value of the SSPxADD register. The address is compared on the falling edge of the eighth clock (SCLx) pulse. If the addresses match and the BF and SSPOV bits are clear, the following events occur:

- 1. The SSPxSR register value is loaded into the SSPxBUF register.
- 2. The Buffer Full bit, BF, is set.
- 3. An ACK pulse is generated.
- 4. The MSSP Interrupt Flag bit, SSPxIF, is set (and an interrupt is generated, if enabled) on the falling edge of the ninth SCLx pulse.

In 10-Bit Addressing mode, two address bytes need to be received by the slave. The five Most Significant bits (MSbs) of the first address byte specify if this is a 10-bit address. The  $R/\overline{W}$  (SSPxSTAT<2>) bit must specify a write so the slave device will receive the second address byte. For a 10-bit address, the first byte would equal '11110 A9 A8 0', where 'A9' and 'A8' are the two MSbs of the address. The sequence of events for 10-bit addressing is as follows, with Steps 7 through 9 for the slave-transmitter:

- 1. Receive first (high) byte of address (bits, SSPxIF, BF and UA, are set on address match).
- 2. Update the SSPxADD register with second (low) byte of address (clears bit, UA, and releases the SCLx line).
- 3. Read the SSPxBUF register (clears bit, BF) and clear flag bit, SSPxIF.
- 4. Receive second (low) byte of address (bits, SSPxIF, BF and UA, are set).
- 5. Update the SSPxADD register with the first (high) byte of address. If match releases SCLx line, this will clear bit, UA.
- 6. Read the SSPxBUF register (clears bit, BF) and clear flag bit, SSPxIF.
- 7. Receive Repeated Start condition.
- 8. Receive first (high) byte of address (bits, SSPxIF and BF, are set).
- 9. Read the SSPxBUF register (clears bit, BF) and clear flag bit, SSPxIF.

#### 21.4.3.2 Address Masking Modes

Masking an address bit causes that bit to become a "don't care". When one address bit is masked, two addresses will be Acknowledged and cause an interrupt. It is possible to mask more than one address bit at a time, which greatly expands the number of addresses Acknowledged.

The  $l^2C$  slave behaves the same way whether address masking is used or not. However, when address masking is used, the  $l^2C$  slave can Acknowledge multiple addresses and cause interrupts. When this occurs, it is necessary to determine which address caused the interrupt by checking the SSPxBUF.

The PIC18F87K22 family of devices is capable of using two different Address Masking modes in  $I^2C$  slave operation: 5-Bit Address Masking and 7-Bit Address Masking. The Masking mode is selected at device configuration using the MSSPMSK Configuration bit. The default device configuration is 7-Bit Address Masking.

Both Masking modes, in turn, support address masking of 7-bit and 10-bit addresses. The combination of Masking modes and addresses provides different ranges of Acknowledgable addresses for each combination.

While both Masking modes function in roughly the same manner, the way they use address masks are different.

#### 21.4.3.3 5-Bit Address Masking Mode

As the name implies, 5-Bit Address Masking mode uses an address mask of up to 5 bits to create a range of addresses to be Acknowledged, using bits, 5 through 1, of the incoming address. This allows the module to Acknowledge up to 31 addresses when using 7-bit addressing, or 63 addresses with 10-bit addressing (see Example 21-2). This Masking mode is selected when the MSSPMSK Configuration bit is programmed ('0').

The address mask in this mode is stored in the SSPxCON2 register, which stops functioning as a control register in I<sup>2</sup>C Slave mode (Register 21-6). In 7-Bit Address Masking mode, address mask bits, ADMSK<5:1> (SSPxCON2<5:1>), mask the corresponding address bits in the SSPxADD register. For any ADMSK bits that are set (ADMSK<n> = 1), the corresponding address bit is ignored (SSPxADD<n> = x). For the module to issue an address Acknowledge, it is sufficient to match only on addresses that do not have an active address mask.

In 10-Bit Address Masking mode, bits, ADMSK<5:2>, mask the corresponding address bits in the SSPxADD register. In addition, ADMSK1 simultaneously masks the two LSbs of the address (SSPxADD<1:0>). For any ADMSK bits that are active (ADMSK<n> = 1), the corresponding address bit is ignored (SPxADD<n> = x). Also note, that although in 10-Bit Address Masking mode, the upper address bits reuse part of the SSPxADD register bits. The address mask bits do not interact with those bits; they only affect the lower address bits.

- Note 1: ADMSK1 masks the two Least Significant bits of the address.
  - The two Most Significant bits of the address are not affected by address masking.

#### EXAMPLE 21-2: ADDRESS MASKING EXAMPLES IN 5-BIT MASKING MODE

#### 7-Bit Addressing:

SSPxADD<7:1>= A0h (1010000) (SSPxADD<0> is assumed to be '0')

#### ADMSK<5:1> = 00111

Addresses Acknowledged: A0h, A2h, A4h, A6h, A8h, AAh, ACh, AEh

#### **10-Bit Addressing:**

SSPxADD<7:0> = A0h (10100000) (The two MSb of the address are ignored in this example, since they are not affected by masking)

ADMSK<5:1> = 00111

Addresses Acknowledged: A0h, A1h, A2h, A3h, A4h, A5h, A6h, A7h, A8h, A9h, AAh, ABh, ACh, ADh, AEh, AFh

#### 21.4.3.4 7-Bit Address Masking Mode

Unlike 5-bit masking, 7-Bit Address Masking mode uses a mask of up to 8 bits (in 10-bit addressing) to define a range of addresses that can be Acknowledged, using the lowest bits of the incoming address. This allows the module to Acknowledge up to 127 different addresses with 7-bit addressing, or 255 with 10-bit addressing (see Example 21-3). This mode is the default configuration of the module, which is selected when MSSPMSK is unprogrammed ('1').

The address mask for 7-Bit Address Masking mode is stored in the SSPxMSK register, instead of the SSPxCON2 register. SSPxMSK is a separate hardware register within the module, but it is not directly addressable. Instead, it shares an address in the SFR space with the SSPxADD register. To access the SSPxMSK register, it is necessary to select MSSP mode, '1001' (SSPxCON1<3:0> = 1001) and then read or write to the location of SSPxADD.

To use 7-Bit Address Masking mode, it is necessary to initialize SSPxMSK with a value before selecting the  $I^2C$  Slave Addressing mode. Thus, the required sequence of events is:

- 1. Select SSPxMSK Access mode (SSPxCON2<3:0> = 1001).
- 2. Write the mask value to the appropriate SSPADD register address (FC8h for MSSP1, F6Eh for MSSP2).
- 3. Set the appropriate I<sup>2</sup>C Slave mode (SSPxCON2<3:0> = 0111 for 10-bit addressing, '0110' for 7-bit addressing).

Setting or clearing mask bits in SSPxMSK behaves in the opposite manner of the ADMSK bits in 5-Bit Address Masking mode. That is, clearing a bit in SSPxMSK causes the corresponding address bit to be masked; setting the bit requires a match in that position. SSPxMSK resets to all '1's upon any Reset condition and, therefore, has no effect on the standard MSSP operation until written with a mask value.

With 7-bit addressing, SSPxMSK<7:1> bits mask the corresponding address bits in the SSPxADD register. For any SSPxMSK bits that are active (SSPxMSK<n> = 0), the corresponding SSPxADD address bit is ignored (SSPxADD<n> = x). For the module to issue an address Acknowledge, it is sufficient to match only on addresses that do not have an active address mask.

With 10-bit addressing, SSPxMSK<7:0> bits mask the corresponding address bits in the SSPxADD register. For any SSPxMSK bits that are active (= 0), the corresponding SSPxADD address bit is ignored (SSPxADD<br/>n = x).

Note: The two Most Significant bits of the address are not affected by address masking.

#### EXAMPLE 21-3: ADDRESS MASKING EXAMPLES IN 7-BIT MASKING MODE

#### 7-Bit Addressing:

SSPxADD<7:1> = 1010 000

SSPxMSK<7:1> = 1111 001

Addresses Acknowledged = ACh, A8h, A4h, A0h

#### 10-Bit Addressing:

SSPxADD<7:0> = 1010 0000 (The two MSb are ignored in this example since they are not affected)

SSPxMSK<5:1> = 1111 0011

Addresses Acknowledged = ACh, A8h, A4h, A0h

#### 21.4.3.5 Reception

When the  $R/\overline{W}$  bit of the address byte is clear and an address match occurs, the  $R/\overline{W}$  bit of the SSPxSTAT register is cleared. The received address is loaded into the SSPxBUF register and the SDAx line is held low (ACK).

When the address byte overflow condition exists, then the no Acknowledge (ACK) pulse is given. An overflow condition is defined as either bit, BF (SSPxSTAT<0>), is set or bit, SSPOV (SSPxCON1<6>), is set.

An MSSP interrupt is generated for each data transfer byte. The interrupt flag bit, SSPxIF, must be cleared in software. The SSPxSTAT register is used to determine the status of the byte.

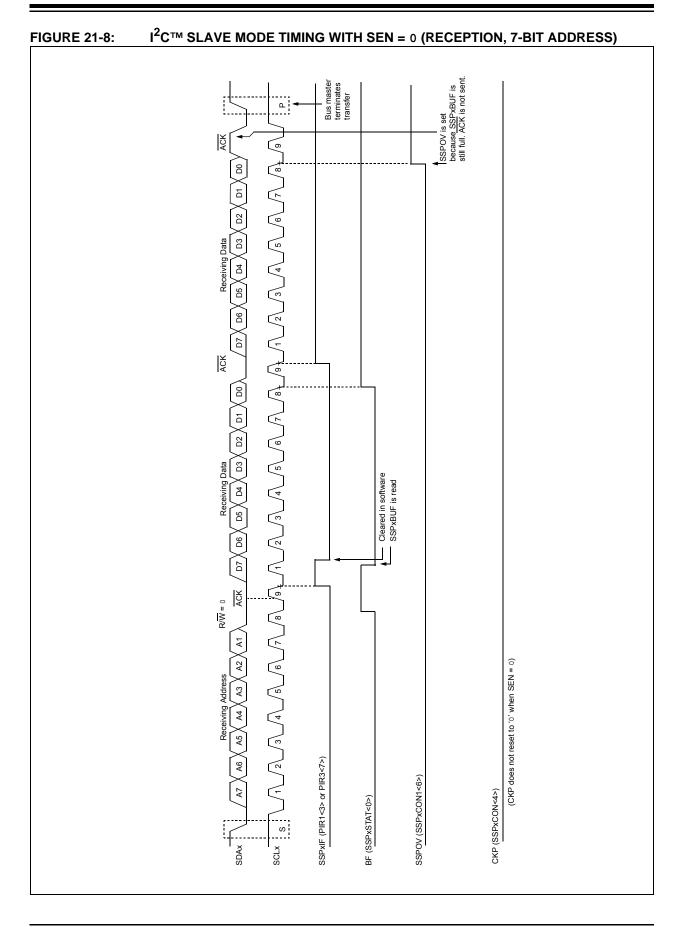
If SEN is enabled (SSPxCON2<0> = 1), SCLx will be held low (clock stretch) following each data transfer. The clock must be released by setting bit, CKP (SSPxCON1<4>). See Section 21.4.4 "Clock Stretching" for more details.

#### 21.4.3.6 Transmission

When the  $R/\overline{W}$  bit of the incoming address byte is set and an address match occurs, the  $R/\overline{W}$  bit of the SSPxSTAT register is set. The received address is loaded into the SSPxBUF register. The ACK pulse will be sent on the ninth bit and pin SCLx is held low regardless of SEN (see Section 21.4.4 "Clock Stretching" for more details). By stretching the clock, the master will be unable to assert another clock pulse until the slave is done preparing the transmit data. The transmit data must be loaded into the SSPxBUF register which also loads the SSPxSR register. Then, pin SCLx should be enabled by setting bit, CKP (SSPxCON1<4>). The eight data bits are shifted out on the falling edge of the SCLx input. This ensures that the SDAx signal is valid during the SCLx high time (Figure 21-10).

The  $\overline{ACK}$  pulse from the master-receiver is latched on the rising edge of the ninth SCLx input pulse. If the SDAx line is high (not  $\overline{ACK}$ ), then the data transfer is complete. In this case, when the  $\overline{ACK}$  is latched by the slave, the slave logic is reset and the slave monitors for another occurrence of the Start bit. If the SDAx line was low ( $\overline{ACK}$ ), the next transmit data must be loaded into the SSPxBUF register. Again, pin SCLx must be enabled by setting bit, CKP.

An MSSP interrupt is generated for each data transfer byte. The SSPxIF bit must be cleared in software and the SSPxSTAT register is used to determine the status of the byte. The SSPxIF bit is set on the falling edge of the ninth clock pulse.

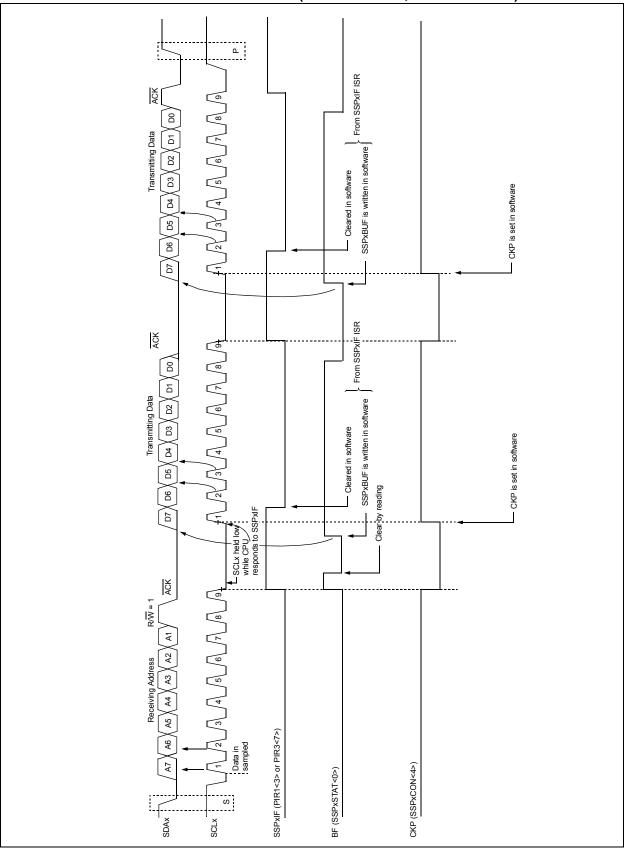


# **PIC18F87K22 FAMILY** $I^2C^{TM}$ SLAVE MODE TIMING WITH SEN = 0 AND ADMSK<5:1> = 01011

(RECEPTION, 7-BIT ADDRESS) SSPOV is set because SSPxBUF is still full. ACK is not sent. Bus master terminates transfer ٩ ACK ົ໑ 8 8 5 D2 D6 X D5 X D4 X D3 > Receiving Data 6 ACK ( DO ) In this example, an address equal to A7.A6.A5.X.A3.X.X will be Acknowledged and cause an interrupt. D2 X D1  $D5 \setminus D4 \setminus D3 \setminus$ Receiving Data Cleared in software SSPxBUF is read D6 X 1 x = Don't care (i.e., address bit can either be a '1' or a '0'). ACK R/W = 0 (CKP does not reset to '0' when SEN = 0) A5 X X X A3 ) Receiving Addre A6 SSPxIF (PIR1<3> or PIR3<7>) SSPOV (SSPxCON1<6>) A7 CKP (SSPxCON<4>) BF (SSPxSTAT<0>) S ÷ ∺ SDAX Note SCLX

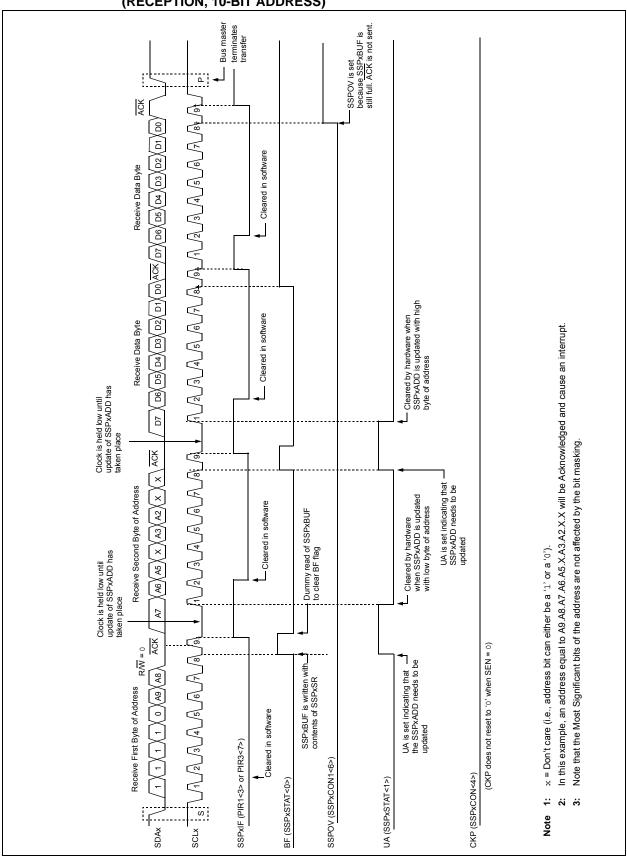
FIGURE 21-9:

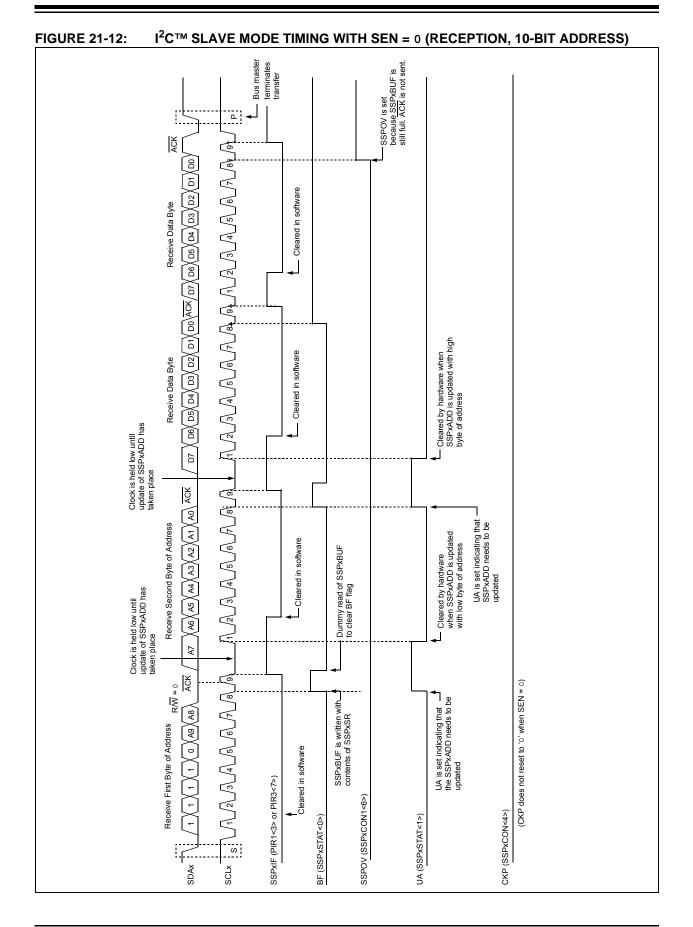


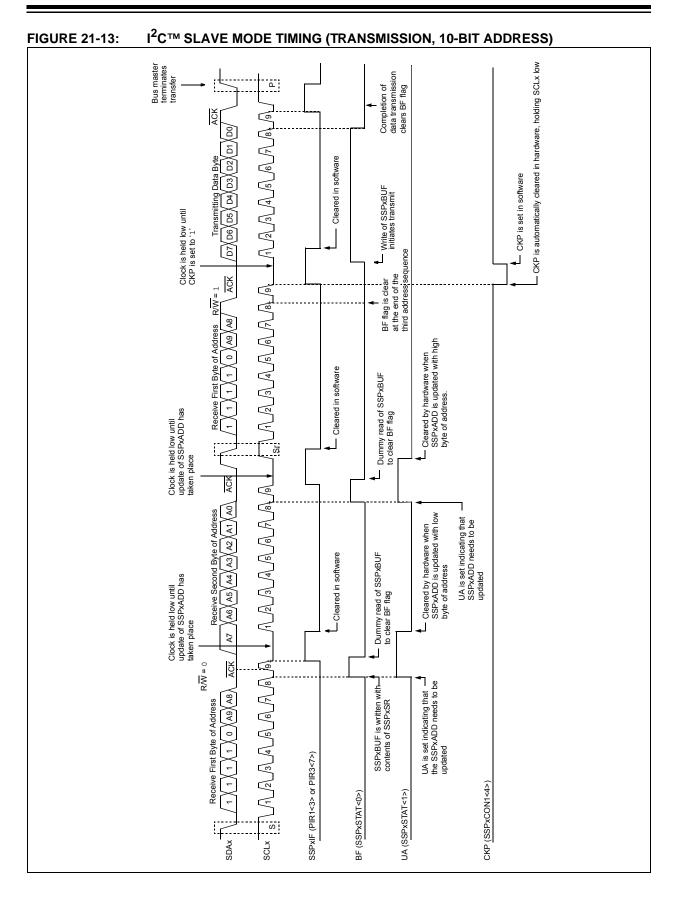


PIC18F87K22 FAMILY

FIGURE 21-11: I<sup>2</sup>C<sup>™</sup> SLAVE MODE TIMING WITH SEN = 0 AND ADMSK<5:1> = 01001 (RECEPTION, 10-BIT ADDRESS)







#### 21.4.4 CLOCK STRETCHING

Both 7-Bit and 10-Bit Slave modes implement automatic clock stretching during a transmit sequence.

The SEN bit (SSPxCON2<0>) allows clock stretching to be enabled during receives. Setting SEN will cause the SCLx pin to be held low at the end of each data receive sequence.

#### 21.4.4.1 Clock Stretching for 7-Bit Slave Receive Mode (SEN = 1)

In 7-Bit Slave Receive mode, <u>on the falling edge of the</u> ninth clock at the end of the ACK sequence, if the BF bit is set, the CKP bit in the SSPxCON1 register is automatically cleared, forcing the SCLx output to be held low. The CKP bit being cleared to '0' will assert the SCLx line low. The CKP bit must be set in the user's ISR before reception is allowed to continue. By holding the SCLx line low, the user has time to service the ISR and read the contents of the SSPxBUF before the master device can initiate another receive sequence. This will prevent buffer overruns from occurring (see Figure 21-15).

- Note 1: If the user reads the contents of the SSPxBUF before the falling edge of the ninth clock, thus clearing the BF bit, the CKP bit will not be cleared and clock stretching will not occur.
  - 2: The CKP bit can be set in software regardless of the state of the BF bit. The user should be careful to clear the BF bit in the ISR before the next receive sequence in order to prevent an overflow condition.

#### 21.4.4.2 Clock Stretching for 10-Bit Slave Receive Mode (SEN = 1)

In 10-Bit Slave Receive mode, during the address sequence, clock stretching automatically takes place but CKP is not cleared. During this time, if the UA bit is set after the ninth clock, clock stretching is initiated. The UA bit is set after receiving the upper byte of the 10-bit address and following the receive of the second byte of the 10-bit address with the R/W bit cleared to '0'. The release of the clock line occurs upon updating SSPxADD. Clock stretching will occur on each data receive sequence as described in 7-bit mode.

Note: If the user polls the UA bit and clears it by updating the SSPxADD register before the falling edge of the ninth clock occurs, and if the user hasn't cleared the BF bit by reading the SSPxBUF register before that time, then the CKP bit will still NOT be asserted low. Clock stretching, on the basis of the state of the BF bit, only occurs during a data sequence, not an address sequence.

#### 21.4.4.3 Clock Stretching for 7-Bit Slave Transmit Mode

The 7-Bit Slave Transmit mode implements clock stretching by clearing the CKP bit after the falling edge of the ninth clock if the BF bit is clear. This occurs regardless of the state of the SEN bit.

The user's ISR must set the CKP bit before transmission is allowed to continue. By holding the SCLx line low, the user has time to service the ISR and load the contents of the SSPxBUF before the master device can initiate another transmit sequence (see Figure 21-10).

- Note 1: If the user loads the contents of SSPxBUF, setting the BF bit before the falling edge of the ninth clock, the CKP bit will not be cleared and clock stretching will not occur.
  - **2:** The CKP bit can be set in software regardless of the state of the BF bit.

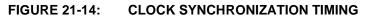
#### 21.4.4.4 Clock Stretching for 10-Bit Slave Transmit Mode

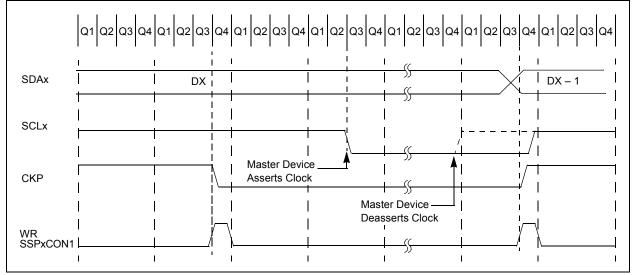
In 10-Bit Slave Transmit mode, clock stretching is controlled during the first two address sequences by the state of the UA bit, just as it is in 10-Bit Slave Receive mode. The first two addresses are followed by a third address sequence, which contains the high-order bits of the 10-bit address and the R/W bit set to '1'. After the third address sequence is performed, the UA bit is not set, the module is now configured in Transmit mode and clock stretching is controlled by the BF flag as in 7-Bit Slave Transmit mode (see Figure 21-13).

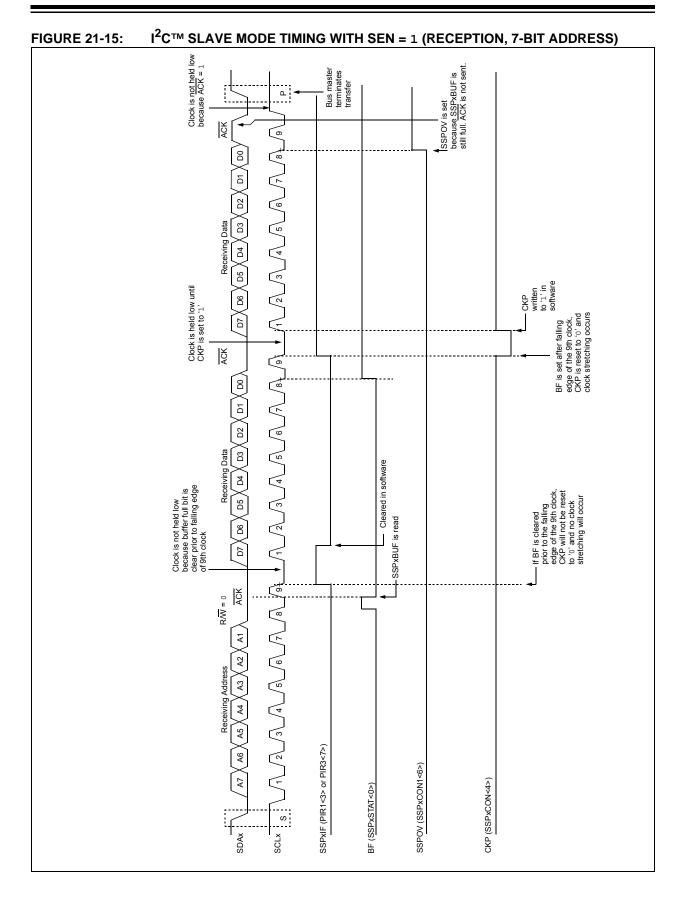
### 21.4.4.5 Clock Synchronization and the CKP bit

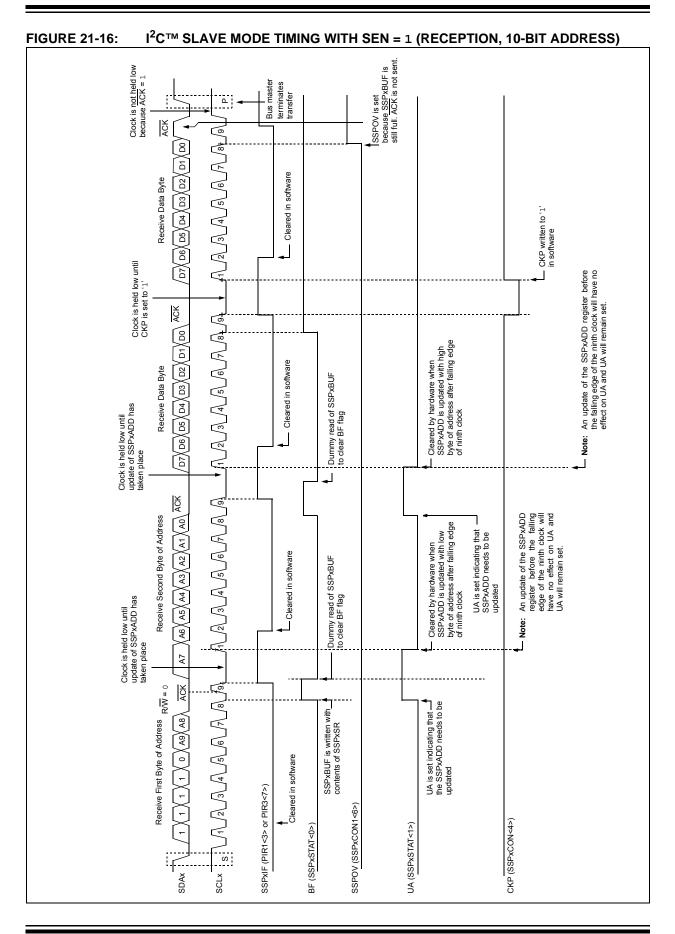
When the CKP bit is cleared, the SCLx output is forced to '0'. However, clearing the CKP bit will not assert the SCLx output low until the SCLx output is already sampled low. Therefore, the CKP bit will not assert the SCLx line until an external  $I^2C$  master device has

already asserted the SCLx line. The SCLx output will remain low until the CKP bit is set and all other devices on the  $I^2C$  bus have deasserted SCLx. This ensures that a write to the CKP bit will not violate the minimum high time requirement for SCLx (see Figure 21-14).









#### 21.4.5 GENERAL CALL ADDRESS SUPPORT

The addressing procedure for the  $I^2C$  bus is such that the first byte after the Start condition usually determines which device will be the slave addressed by the master. The exception is the general call address which can address all devices. When this address is used, all devices should, in theory, respond with an Acknowledge.

The general call address is one of eight addresses reserved for specific purposes by the  $I^2C$  protocol. It consists of all '0's with R/W = 0.

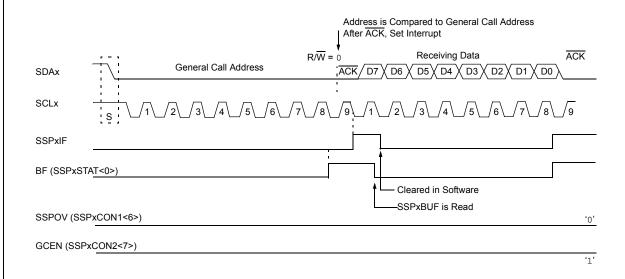
The general call address is recognized when the General Call Enable bit, GCEN, is enabled (SSPxCON2<7> set). Following a Start bit detect, eight bits are shifted into the SSPxSR and the address is compared against the SSPxADD. It is also compared to the general call address and fixed in hardware.

If the general call address matches, the SSPxSR is transferred to the SSPxBUF, the BF flag bit is set (eighth bit), and on the falling edge of the ninth bit (ACK bit), the SSPxIF interrupt flag bit is set.

When the interrupt is serviced, the source for the interrupt can be checked by reading the contents of the SSPxBUF. The value can be used to determine if the address was device-specific or a general call address.

In 10-Bit Addressing mode, the SSPxADD is required to be updated for the second half of the address to match and the UA bit is set (SSPxSTAT<1>). If the general call address is sampled when the GCEN bit is set, while the slave is configured in 10-Bit Addressing mode, then the second half of the address is not necessary, the UA bit will not be set and the slave will begin receiving data after the Acknowledge (Figure 21-17).





#### 21.4.6 MASTER MODE

Master mode is enabled by setting and clearing the appropriate SSPM bits in SSPxCON1, and by setting the SSPEN bit. In Master mode, the SCLx and SDAx lines are manipulated by the MSSP hardware if the TRIS bits are set.

The Master mode of operation is supported by interrupt generation on the detection of the Start and Stop conditions. The Stop (P) and Start (S) bits are cleared from a Reset or when the MSSP module is disabled. Control of the  $I^2C$  bus may be taken when the P bit is set, or the bus is Idle, with both the S and P bits clear.

In Firmware Controlled Master mode, user code conducts all  $I^2C$  bus operations based on Start and Stop bit conditions.

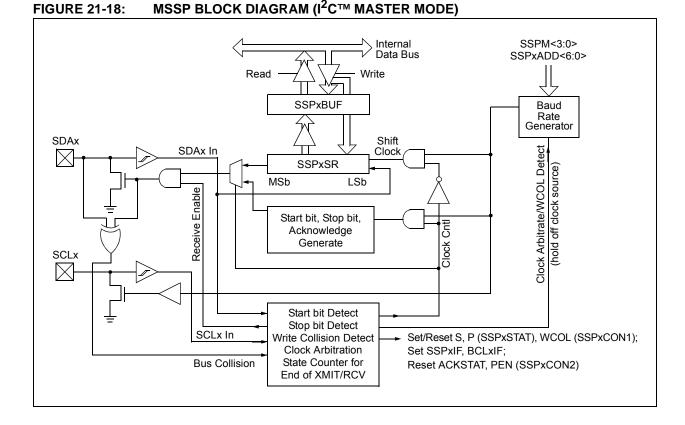
Once Master mode is enabled, the user has six options.

- 1. Assert a Start condition on SDAx and SCLx.
- 2. Assert a Repeated Start condition on SDAx and SCLx.
- 3. Write to the SSPxBUF register, initiating transmission of data/address.
- 4. Configure the I<sup>2</sup>C port to receive data.
- 5. Generate an Acknowledge condition at the end of a received byte of data.
- 6. Generate a Stop condition on SDAx and SCLx.

Note: The MSSP module, when configured in I<sup>2</sup>C Master mode, does not allow queueing of events. For instance, the user is not allowed to initiate a Start condition and immediately write the SSPxBUF register to initiate transmission before the Start condition is complete. In this case, the SSPxBUF will not be written to and the WCOL bit will be set, indicating that a write to the SSPxBUF did not occur.

The following events will cause the MSSP Interrupt Flag bit, SSPxIF, to be set (and MSSP interrupt, if enabled):

- · Start condition
- Stop condition
- Data transfer byte transmitted/received
- Acknowledge transmitted
- Repeated Start



#### 21.4.6.1 I<sup>2</sup>C<sup>™</sup> Master Mode Operation

The master device generates all of the serial clock pulses and the Start and Stop conditions. A transfer is ended with a Stop condition or with a Repeated Start condition. Since the Repeated Start condition is also the beginning of the next serial transfer, the I<sup>2</sup>C bus will not be released.

In Master Transmitter mode, serial data is output through SDAx while SCLx outputs the serial clock. The first byte transmitted contains the slave address of the receiving device (7 bits) and the Read/Write (R/W) bit. In this case, the R/W bit will be logic '0'. Serial data is transmitted, 8 bits at a time. After each byte is transmitted, an Acknowledge bit is received. Start and Stop conditions are output to indicate the beginning and the end of a serial transfer.

In Master Receive mode, the first byte transmitted contains the slave address of the transmitting device (7 bits) and the  $R/\overline{W}$  bit. In this case, the  $R/\overline{W}$  bit will be logic '1'. Thus, the first byte transmitted is a 7-bit slave address, followed by a '1' to indicate the receive bit. Serial data is received via SDAx, while SCLx outputs the serial clock. Serial data is received, 8 bits at a time. After each byte is received, an Acknowledge bit is transmitted. Start and Stop conditions indicate the beginning and end of transmission.

The Baud Rate Generator, used for the SPI mode operation, is used to set the SCLx clock frequency for either 100 kHz, 400 kHz or 1 MHz I<sup>2</sup>C operation. See **Section 21.4.7 "Baud Rate**" for more details.

A typical transmit sequence would go as follows:

- 1. The user generates a Start condition by setting the Start Enable bit, SEN (SSPxCON2<0>).
- 2. SSPxIF is set. The MSSP module will wait the required start time before any other operation takes place.
- 3. The user loads the SSPxBUF with the slave address to transmit.
- 4. Address is shifted out the SDAx pin until all 8 bits are transmitted.
- 5. The MSSP module shifts in the ACK bit from the slave device and writes its value into the SSPxCON2 register (SSPxCON2<6>).
- 6. The MSSP module generates an interrupt at the end of the ninth clock cycle by setting the SSPxIF bit.
- 7. The user loads the SSPxBUF with eight bits of data.
- 8. Data is shifted out the SDAx pin until all 8 bits are transmitted.
- The MSSP module shifts in the ACK bit from the slave device and writes its value into the SSPxCON2 register (SSPxCON2<6>).
- 10. The MSSP module generates an interrupt at the end of the ninth clock cycle by setting the SSPxIF bit.
- 11. The user generates a Stop condition by setting the Stop Enable bit, PEN (SSPxCON2<2>).
- 12. Interrupt is generated once the Stop condition is complete.

#### 21.4.7 BAUD RATE

In I<sup>2</sup>C Master mode, the Baud Rate Generator (BRG) reload value is placed in the lower 7 bits of the SSPxADD register (Figure 21-19). When a write occurs to SSPxBUF, the Baud Rate Generator will automatically begin counting. The BRG counts down to 0 and stops until another reload has taken place. The BRG count is decremented twice per instruction cycle (TcY) on the Q2 and Q4 clocks. In I<sup>2</sup>C Master mode, the BRG is reloaded automatically.

Once the given operation is complete (i.e., transmission of the last data bit is followed by ACK), the internal clock will automatically stop counting and the SCLx pin will remain in its last state.

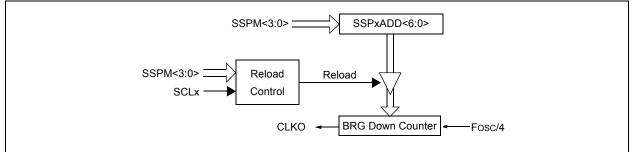
Table 21-3 demonstrates clock rates based on instruction cycles and the BRG value loaded into SSPxADD. The SSPxADD BRG value of  $0 \ge 00$  is not supported.

#### 21.4.7.1 Baud Rate and Module Interdependence

Because MSSP1 and MSSP2 are independent, they can operate simultaneously in I<sup>2</sup>C Master mode at different baud rates. This is done by using different BRG reload values for each module.

Because this mode derives its basic clock source from the system clock, any changes to the clock will affect both modules in the same proportion. It may be possible to change one or both baud rates back to a previous value by changing the BRG reload value.

#### FIGURE 21-19: BAUD RATE GENERATOR BLOCK DIAGRAM



#### TABLE 21-3: I<sup>2</sup>C<sup>™</sup> CLOCK RATE w/BRG

Fosc	Fcy	Fcy * 2	BRG Value	FscL (2 Rollovers of BRG)
40 MHz	10 MHz	20 MHz	18h	400 kHz
40 MHz	10 MHz	20 MHz	1Fh	312.5 kHz
40 MHz	10 MHz	20 MHz	63h	100 kHz
16 MHz	4 MHz	8 MHz	09h	400 kHz
16 MHz	4 MHz	8 MHz	0Ch	308 kHz
16 MHz	4 MHz	8 MHz	27h	100 kHz
4 MHz	1 MHz	2 MHz	02h	333 kHz
4 MHz	1 MHz	2 MHz	09h	100 kHz
16 MHz	4 MHz	8 MHz	03h	1 MHz <sup>(1)</sup>

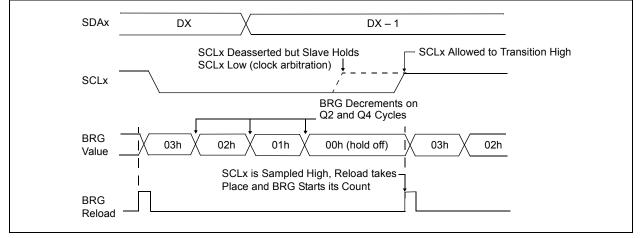
**Note 1:** A minimum of 16 MHz Fosc is required to get 1 MHz  $I^2C$ .

#### 21.4.7.2 Clock Arbitration

Clock arbitration occurs when the master, during any receive, transmit or Repeated Start/Stop condition, deasserts the SCLx pin (SCLx allowed to float high). When the SCLx pin is allowed to float high, the Baud Rate Generator (BRG) is suspended from counting until the SCLx pin is actually sampled high. When the

SCLx pin is sampled high, the Baud Rate Generator is reloaded with the contents of SSPxADD<6:0> and begins counting. This ensures that the SCLx high time will always be at least one BRG rollover count in the event that the clock is held low by an external device (Figure 21-20).





#### 21.4.8 I<sup>2</sup>C<sup>™</sup> MASTER MODE START CONDITION TIMING

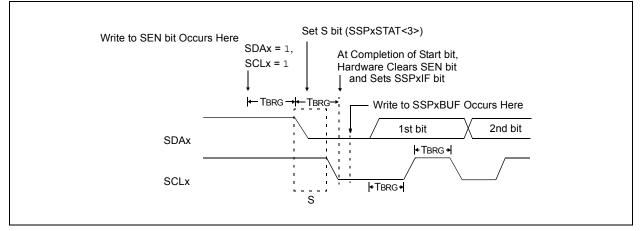
To initiate a Start condition, the user sets the Start Enable bit, SEN (SSPxCON2<0>). If the SDAx and SCLx pins are sampled high, the Baud Rate Generator is reloaded with the contents of SSPxADD<6:0> and starts its count. If SCLx and SDAx are both sampled high when the Baud Rate Generator times out (TBRG), the SDAx pin is driven low. The action of the SDAx being driven low while SCLx is high is the Start condition and causes the S bit (SSPxSTAT<3>) to be set. Following this, the Baud Rate Generator is reloaded with the contents of SSPxADD<6:0> and resumes its count. When the Baud Rate Generator times out (TBRG), the SEN bit (SSPxCON2<0>) will be automatically cleared by hardware. The Baud Rate Generator is suspended, leaving the SDAx line held low and the Start condition is complete.

Note: If, at the beginning of the Start condition, the SDAx and SCLx pins are already sampled low, or if during the Start condition, the SCLx line is sampled low before the SDAx line is driven low, a bus collision occurs, the Bus Collision Interrupt Flag, BCLxIF, is set, the Start condition is aborted and the I<sup>2</sup>C module is reset into its Idle state.

#### 21.4.8.1 WCOL Status Flag

If the user writes the SSPxBUF when a Start sequence is in progress, the WCOL bit is set and the contents of the buffer are unchanged (the write doesn't occur).

Note: Because queueing of events is not allowed, writing to the lower 5 bits of SSPxCON2 is disabled until the Start condition is complete.



#### FIGURE 21-21: FIRST START BIT TIMING

#### 21.4.9 I<sup>2</sup>C<sup>™</sup> MASTER MODE REPEATED START CONDITION TIMING

A Repeated Start condition occurs when the RSEN bit (SSPxCON2<1>) is programmed high and the I<sup>2</sup>C logic module is in the Idle state. When the RSEN bit is set, the SCLx pin is asserted low. When the SCLx pin is sampled low, the Baud Rate Generator is loaded with the contents of SSPxADD<5:0> and begins counting. The SDAx pin is released (brought high) for one Baud Rate Generator count (TBRG). When the Baud Rate Generator times out, and if SDAx is sampled high, the SCLx pin will be deasserted (brought high). When SCLx is sampled high, the Baud Rate Generator is reloaded with the contents of SSPxADD<6:0> and begins counting. SDAx and SCLx must be sampled high for one TBRG. This action is then followed by assertion of the SDAx pin (SDAx = 0) for one TBRG while SCLx is high. Following this, the RSEN bit (SSPxCON2<1>) will be automatically cleared and the Baud Rate Generator will not be reloaded, leaving the SDAx pin held low. As soon as a Start condition is detected on the SDAx and SCLx pins, the S bit (SSPxSTAT<3>) will be set. The SSPxIF bit will not be set until the Baud Rate Generator has timed out.

- **Note 1:** If RSEN is programmed while any other event is in progress, it will not take effect.
  - **2:** A bus collision during the Repeated Start condition occurs if:
    - SDAx is sampled low when SCLx goes from low-to-high.
    - SCLx goes low before SDAx is asserted low. This may indicate that another master is attempting to transmit a data '1'.

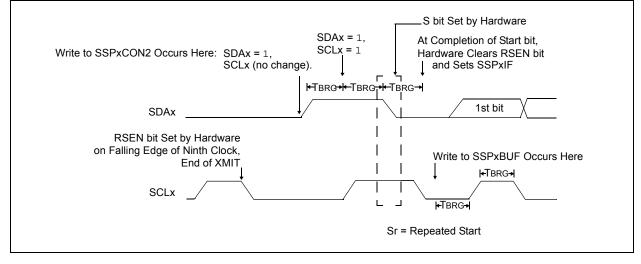
Immediately following the SSPxIF bit getting set, the user may write the SSPxBUF with the 7-bit address in 7-bit mode or the default first address in 10-bit mode. After the first eight bits are transmitted and an ACK is received, the user may then transmit an additional eight bits of address (10-bit mode) or eight bits of data (7-bit mode).

#### 21.4.9.1 WCOL Status Flag

If the user writes the SSPxBUF when a Repeated Start sequence is in progress, the WCOL is set and the contents of the buffer are unchanged (the write doesn't occur).

Note: Because queueing of events is not allowed, writing of the lower 5 bits of SSPxCON2 is disabled until the Repeated Start condition is complete.

### FIGURE 21-22: REPEATED START CONDITION WAVEFORM



#### 21.4.10 I<sup>2</sup>C<sup>™</sup> MASTER MODE TRANSMISSION

Transmission of a data byte, a 7-bit address or the other half of a 10-bit address, is accomplished by simply writing a value to the SSPxBUF register. This action will set the Buffer Full flag bit, BF, and allow the Baud Rate Generator to begin counting and start the next transmission. Each bit of address/data will be shifted out onto the SDAx pin after the falling edge of SCLx is asserted (see data hold time specification Parameter 106). SCLx is held low for one Baud Rate Generator rollover count (TBRG). Data should be valid before SCLx is released high (see data setup time specification Parameter 107). When the SCLx pin is released high, it is held that way for TBRG. The data on the SDAx pin must remain stable for that duration and some hold time after the next falling edge of SCLx. After the eighth bit is shifted out (the falling edge of the eighth clock), the BF flag is cleared and the master releases SDAx. This allows the slave device being addressed to respond with an ACK bit during the ninth bit time if an address match occurred, or if data was received properly. The status of ACK is written into the ACKDT bit on the falling edge of the ninth clock. If the master receives an Acknowledge, the Acknowledge Status bit, ACKSTAT, is cleared; if not, the bit is set. After the ninth clock, the SSPxIF bit is set and the master clock (Baud Rate Generator) is suspended until the next data byte is loaded into the SSPxBUF, leaving SCLx low and SDAx unchanged (Figure 21-23).

After the write to the SSPxBUF, each bit of the address will be shifted out on the falling edge of SCLx until all seven address bits and the R/W bit are completed. On the falling edge of the eighth clock, the master will deassert the SDAx pin, allowing the slave to respond with an Acknowledge. On the falling edge of the ninth clock, the master will sample the SDAx pin to see if the address was recognized by a slave. The status of the ACK bit is loaded into the ACKSTAT status bit (SSPxCON2<6>). Following the falling edge of the ninth clock transmission of the address, the SSPxIF flag is set, the BF flag is cleared and the Baud Rate Generator is turned off until another write to the SSPxBUF takes place, holding SCLx low and allowing SDAx to float.

#### 21.4.10.1 BF Status Flag

In Transmit mode, the BF bit (SSPxSTAT<0>) is set when the CPU writes to SSPxBUF and is cleared when all 8 bits are shifted out.

#### 21.4.10.2 WCOL Status Flag

If the user writes the SSPxBUF when a transmit is already in progress (i.e., SSPxSR is still shifting out a data byte), the WCOL bit is set and the contents of the buffer are unchanged (the write doesn't occur) after 2 TCY after the SSPxBUF write. If SSPxBUF is rewritten within 2 TCY, the WCOL bit is set and SSPxBUF is updated. This may result in a corrupted transfer.

The user should verify that the WCOL bit is clear after each write to SSPxBUF to ensure the transfer is correct. In all cases, WCOL must be cleared in software.

#### 21.4.10.3 ACKSTAT Status Flag

In Transmit mode, the ACKSTAT bit (SSPxCON2<6>) is cleared when the slave has sent an Acknowledge  $(\overline{ACK} = 0)$  and is set when the slave does not Acknowledge  $(\overline{ACK} = 1)$ . A slave sends an Acknowledge when it has recognized its address (including a general call), or when the slave has properly received its data.

### 21.4.11 I<sup>2</sup>C<sup>™</sup> MASTER MODE RECEPTION

Master mode reception is enabled by programming the Receive Enable bit, RCEN (SSPxCON2<3>).

Note: The MSSP module must be in an inactive state before the RCEN bit is set or the RCEN bit will be disregarded.

The Baud Rate Generator begins counting, and on each rollover, the state of the SCLx pin changes (high-to-low/low-to-high) and data is shifted into the SSPxSR. After the falling edge of the eighth clock, the receive enable flag is automatically cleared, the contents of the SSPxSR are loaded into the SSPxBUF, the BF flag bit is set, the SSPxIF flag bit is set and the Baud Rate Generator is suspended from counting, holding SCLx low. The MSSP is now in Idle state awaiting the next command. When the buffer is read by the CPU, the BF flag bit is automatically cleared. The user can then send an Acknowledge bit at the end of reception by setting the Acknowledge Sequence Enable bit, ACKEN (SSPxCON2<4>).

#### 21.4.11.1 BF Status Flag

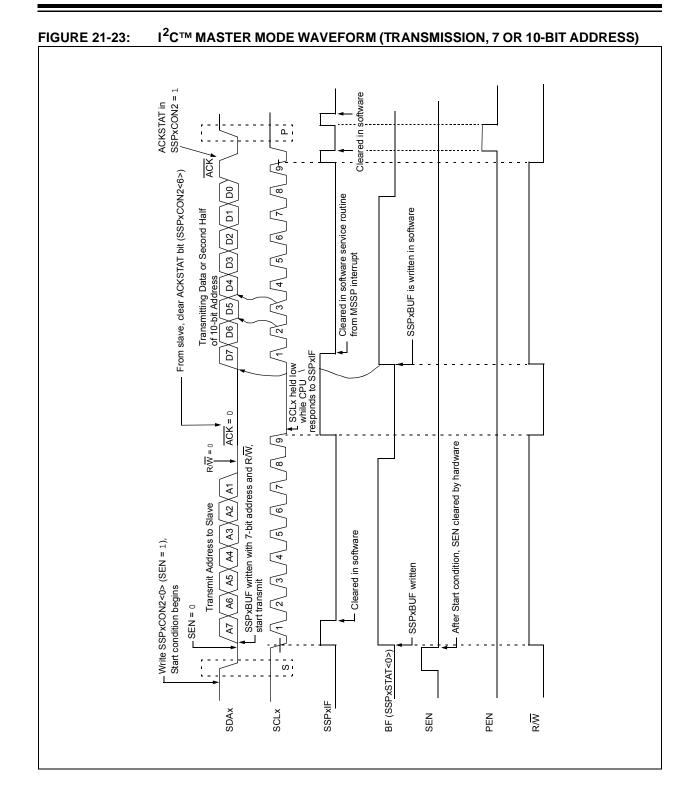
In receive operation, the BF bit is set when an address or data byte is loaded into SSPxBUF from SSPxSR. It is cleared when the SSPxBUF register is read.

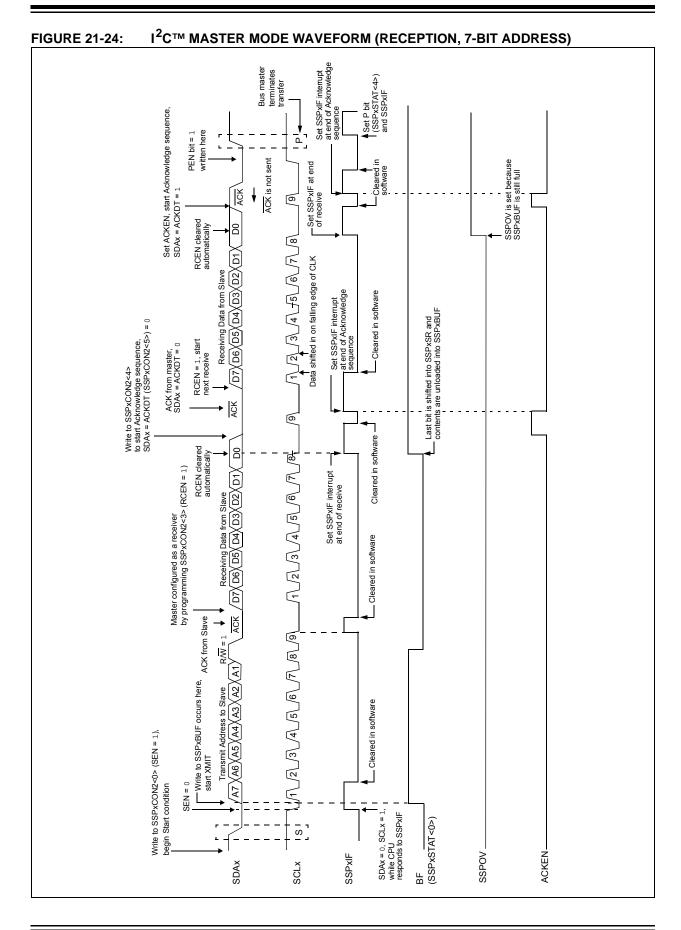
#### 21.4.11.2 SSPOV Status Flag

In receive operation, the SSPOV bit is set when 8 bits are received into the SSPxSR and the BF flag bit is already set from a previous reception.

#### 21.4.11.3 WCOL Status Flag

If the user writes the SSPxBUF when a receive is already in progress (i.e., SSPxSR is still shifting in a data byte), the WCOL bit is set and the contents of the buffer are unchanged (the write doesn't occur).





#### 21.4.12 ACKNOWLEDGE SEQUENCE TIMING

An Acknowledge sequence is enabled by setting the Acknowledge Seauence Enable bit. ACKEN (SSPxCON2<4>). When this bit is set, the SCLx pin is pulled low and the contents of the Acknowledge data bit are presented on the SDAx pin. If the user wishes to generate an Acknowledge, then the ACKDT bit should be cleared. If not, the user should set the ACKDT bit before starting an Acknowledge sequence. The Baud Rate Generator then counts for one rollover period (TBRG) and the SCLx pin is deasserted (pulled high). When the SCLx pin is sampled high (clock arbitration), the Baud Rate Generator counts for TBRG; the SCLx pin is then pulled low. Following this, the ACKEN bit is automatically cleared, the Baud Rate Generator is turned off and the MSSP module then goes into an inactive state (Figure 21-25).

#### 21.4.12.1 WCOL Status Flag

If the user writes the SSPxBUF when an Acknowledge sequence is in progress, then WCOL is set and the contents of the buffer are unchanged (the write doesn't occur).

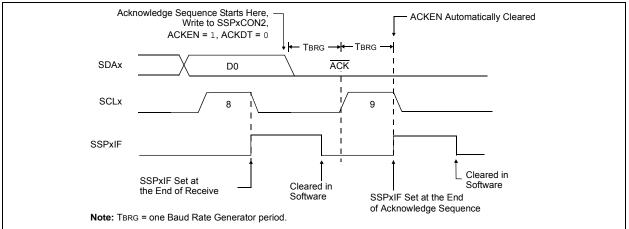
#### 21.4.13 STOP CONDITION TIMING

A Stop bit is asserted on the SDAx pin at the end of a receive/transmit by setting the Stop Sequence Enable bit, PEN (SSPxCON2<2>). At the end of a receive/transmit, the SCLx line is held low after the falling edge of the ninth clock. When the PEN bit is set, the master will assert the SDAx line low. When the SDAx line is sampled low, the Baud Rate Generator is reloaded and counts down to 0. When the Baud Rate Generator times out, the SCLx pin will be brought high and one TBRG (Baud Rate Generator rollover count) later, the SDAx pin will be deasserted. When the SDAx pin is sampled high while SCLx is high, the P bit (SSPxSTAT<4>) is set. A TBRG later, the PEN bit is cleared and the SSPxIF bit is set (see Figure 21-26).

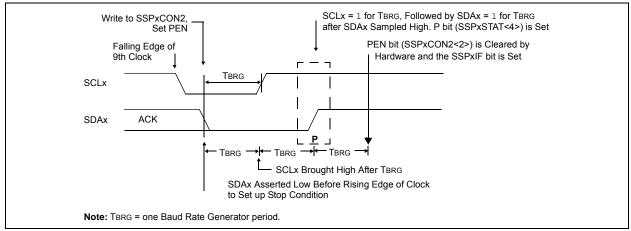
#### 21.4.13.1 WCOL Status Flag

If the user writes the SSPxBUF when a Stop sequence is in progress, then the WCOL bit is set and the contents of the buffer are unchanged (the write doesn't occur).

#### FIGURE 21-25: ACKNOWLEDGE SEQUENCE WAVEFORM



#### FIGURE 21-26: STOP CONDITION RECEIVE OR TRANSMIT MODE



#### 21.4.14 SLEEP OPERATION

While in Sleep mode, the  $I^2C$  module can receive addresses or data, and when an address match or complete byte transfer occurs, wake the processor from Sleep (if the MSSP interrupt is enabled).

#### 21.4.15 EFFECTS OF A RESET

A Reset disables the MSSP module and terminates the current transfer.

#### 21.4.16 MULTI-MASTER MODE

In Multi-Master mode, the interrupt generation on the detection of the Start and Stop conditions allows the determination of when the bus is free. The Stop (P) and Start (S) bits are cleared from a Reset or when the MSSP module is disabled. Control of the  $I^2C$  bus may be taken when the P bit (SSPxSTAT<4>) is set, or the bus is Idle, with both the S and P bits clear. When the bus is busy, enabling the MSSP interrupt will generate the interrupt when the Stop condition occurs.

In multi-master operation, the SDAx line must be monitored for arbitration to see if the signal level is the expected output level. This check is performed in hardware with the result placed in the BCLxIF bit.

The states where arbitration can be lost are:

- · Address Transfer
- · Data Transfer
- A Start Condition
- A Repeated Start Condition
- An Acknowledge Condition

#### 21.4.17 MULTI -MASTER COMMUNICATION, BUS COLLISION AND BUS ARBITRATION

Multi-Master mode support is achieved by bus arbitration. When the master outputs address/data bits onto the SDAx pin, arbitration takes place when the master outputs a '1' on SDAx, by letting SDAx float high, and another master asserts a '0'. When the SCLx pin floats high, data should be stable. If the expected data on SDAx is a '1' and the data sampled on the SDAx pin = 0, then a bus collision has taken place. The master will set the Bus Collision Interrupt Flag, BCLxIF, and reset the  $I^2C$  port to its Idle state (Figure 21-27).

If a transmit was in progress when the bus collision occurred, the transmission is halted, the BF flag is cleared, the SDAx and SCLx lines are deasserted and the SSPxBUF can be written to. When the user services the bus collision Interrupt Service Routine and if the  $I^2C$  bus is free, the user can resume communication by asserting a Start condition.

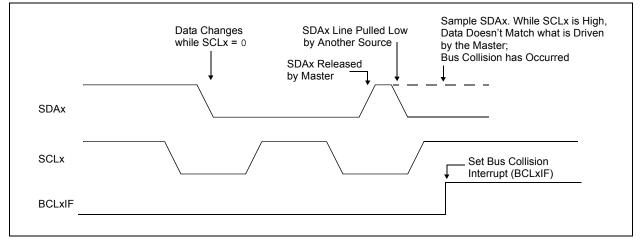
If a Start, Repeated Start, Stop or Acknowledge condition was in progress when the bus collision occurred, the condition is aborted, the SDAx and SCLx lines are deasserted and the respective control bits in the SSPxCON2 register are cleared. When the user services the bus collision Interrupt Service Routine (ISR), and if the  $I^2C$  bus is free, the user can resume communication by asserting a Start condition.

The master will continue to monitor the SDAx and SCLx pins. If a Stop condition occurs, the SSPxIF bit will be set.

A write to the SSPxBUF will start the transmission of data at the first data bit regardless of where the transmitter left off when the bus collision occurred.

In Multi-Master mode, the interrupt generation on the detection of Start and Stop conditions allows the determination of when the bus is free. Control of the I<sup>2</sup>C bus can be taken when the P bit is set in the SSPxSTAT register, or the bus is Idle and the S and P bits are cleared.

#### FIGURE 21-27: BUS COLLISION TIMING FOR TRANSMIT AND ACKNOWLEDGE



#### 21.4.17.1 Bus Collision During a Start Condition

During a Start condition, a bus collision occurs if:

- a) SDAx or SCLx is sampled low at the beginning of the Start condition (Figure 21-28).
- b) SCLx is sampled low before SDAx is asserted low (Figure 21-29).

During a Start condition, both the SDAx and the SCLx pins are monitored.

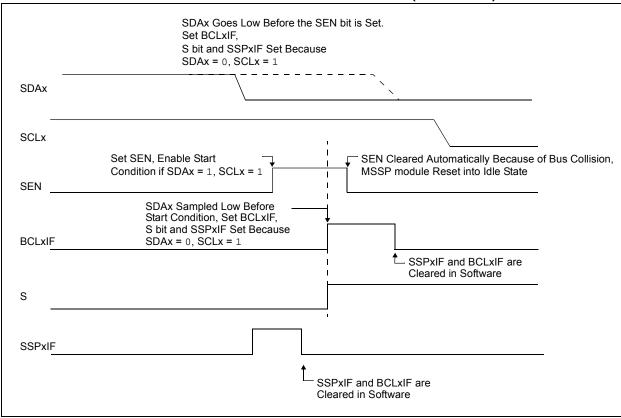
If the SDAx pin is already low, or the SCLx pin is already low, then all of the following occur:

- · The Start condition is aborted
- · The BCLxIF flag is set
- The MSSP module is reset to its inactive state (Figure 21-28)

The Start condition begins with the SDAx and SCLx pins deasserted. When the SDAx pin is sampled high, the Baud Rate Generator is loaded from SSPxADD<6:0> and counts down to 0. If the SCLx pin is sampled low while SDAx is high, a bus collision occurs because it is assumed that another master is attempting to drive a data '1' during the Start condition.

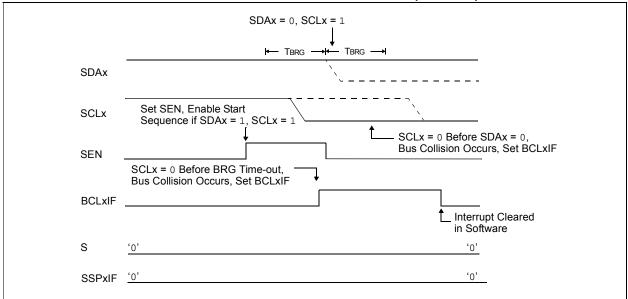
If the SDAx pin is sampled low during this count, the BRG is reset and the SDAx line is asserted early (Figure 21-30). If, however, a '1' is sampled on the SDAx pin, the SDAx pin is asserted low at the end of the BRG count. The Baud Rate Generator is then reloaded and counts down to 0. If the SCLx pin is sampled as '0' during this time, a bus collision does not occur. At the end of the BRG count, the SCLx pin is asserted low.

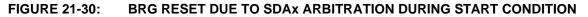
Note: The reason that a bus collision is not a factor during a Start condition is that no two bus masters can assert a Start condition at the exact same time. Therefore, one master will always assert SDAx before the other. This condition does not cause a bus collision because the two masters must be allowed to arbitrate the first address following the Start condition. If the address is the same, arbitration must be allowed to continue into the data portion, Repeated Start or Stop conditions.

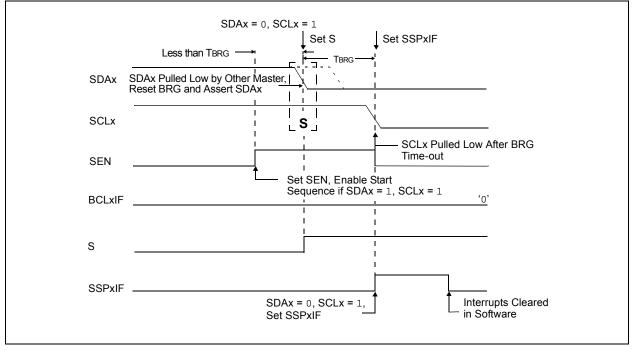


#### FIGURE 21-28: BUS COLLISION DURING START CONDITION (SDAx ONLY)









### 21.4.17.2 Bus Collision During a Repeated Start Condition

During a Repeated Start condition, a bus collision occurs if:

- a) A low level is sampled on SDAx when SCLx goes from a low level to a high level.
- b) SCLx goes low before SDAx is asserted low, indicating that another master is attempting to transmit a data '1'.

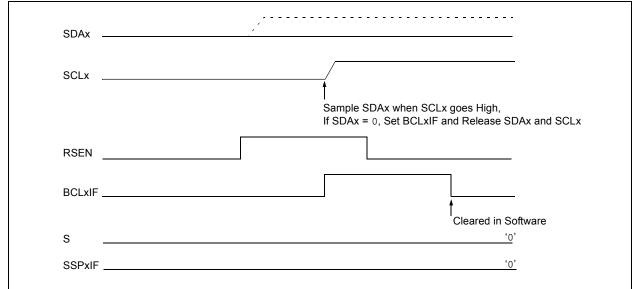
When the user deasserts SDAx and the pin is allowed to float high, the BRG is loaded with SSPxADD<6:0> and counts down to 0. The SCLx pin is then deasserted and when sampled high, the SDAx pin is sampled.

If SDAx is low, a bus collision has occurred (i.e., another master is attempting to transmit a data '0', Figure 21-31). If SDAx is sampled high, the BRG is reloaded and begins counting. If SDAx goes from high-to-low before the BRG times out, no bus collision occurs because no two masters can assert SDAx at exactly the same time.

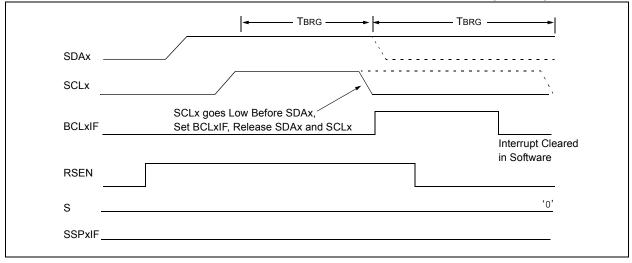
If SCLx goes from high-to-low before the BRG times out and SDAx has not already been asserted, a bus collision occurs. In this case, another master is attempting to transmit a data '1' during the Repeated Start condition (see Figure 21-32).

If, at the end of the BRG time-out, both SCLx and SDAx are still high, the SDAx pin is driven low and the BRG is reloaded and begins counting. At the end of the count, regardless of the status of the SCLx pin, the SCLx pin is driven low and the Repeated Start condition is complete.

#### FIGURE 21-31: BUS COLLISION DURING A REPEATED START CONDITION (CASE 1)



#### FIGURE 21-32: BUS COLLISION DURING REPEATED START CONDITION (CASE 2)



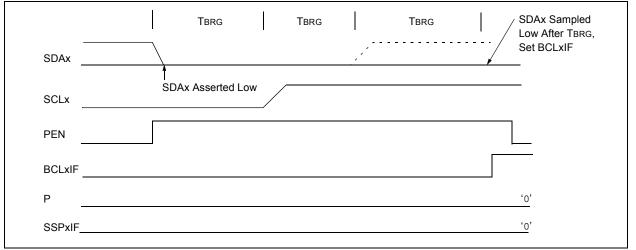
#### 21.4.17.3 Bus Collision During a Stop Condition

Bus collision occurs during a Stop condition if:

- a) After the SDAx pin has been deasserted and allowed to float high, SDAx is sampled low after the BRG has timed out.
- b) After the SCLx pin is deasserted, SCLx is sampled low before SDAx goes high.

The Stop condition begins with SDAx asserted low. When SDAx is sampled low, the SCLx pin is allowed to float. When the pin is sampled high (clock arbitration), the Baud Rate Generator is loaded with SSPxADD<6:0> and counts down to 0. After the BRG times out, SDAx is sampled. If SDAx is sampled low, a bus collision has occurred. This is due to another master attempting to drive a data '0' (Figure 21-33). If the SCLx pin is sampled low before SDAx is allowed to float high, a bus collision occurs. This is another case of another master attempting to drive a data '0' (Figure 21-34).

## FIGURE 21-33: BUS COLLISION DURING A STOP CONDITION (CASE 1)



## FIGURE 21-34: BUS COLLISION DURING A STOP CONDITION (CASE 2)

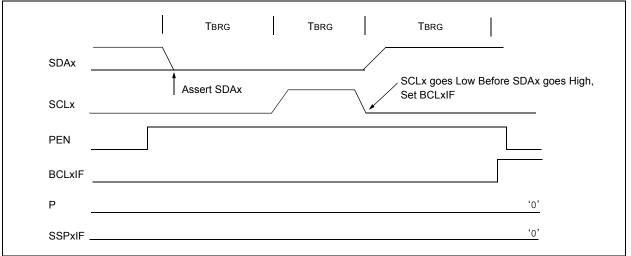


TABLE 21-4:	REGISTERS ASSOCIATED WITH I'C M OPERATION										
Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0			
INTCON	GIE/GIEH	PEIE/GIEL	TMR0IE	INT0IE	RBIE	TMR0IF	INT0IF	RBIF			
PIR1	PSPIF	ADIF	RC1IF	TX1IF	SSP1IF	TMR1GIF	TMR2IF	TMR1IF			
PIE1	PSPIE	ADIE	RC1IE	TX1IE	SSP1IE	TMR1GIE	TMR2IE	TMR1IE			
IPR1	PSPIP	ADIP	RC1IP	TX1IP	SSP1IP	TMR1GIP	TMR2IP	TMR1IP			
PIR2	OSCFIF		SSP2IF	BLC2IF	BCL1IF	HLVDIF	TMR3IF	TMR3GIF			
PIE2	OSCFIE		SSP2IE	BLC2IE	BCL1IE	HLVDIE	TMR3IE	TMR3GIE			
IPR2	OSCFIP		SSP2IP	BLC2IP	BCL1IP	HLVDIP	TMR3IP	TMR3GIP			
PIR3	TMR5GIF		RC2IF	TX2IF	CTMUIF	CCP2IF	CCP1IF	RTCCIF			
PIE3	TMR5GIE	—	RC2IE	TX2IE	CTMUIE	CCP2IE	CCP1IE	RTCCIE			
IPR3	TMR5GIP	_	RC2IP	TX2IP	CTMUIP	CCP2IP	CCP1IP	RTCCIP			
TRISC	TRISC7	TRISC6	TRISC5	TRISC4	TRISC3	TRISC2	TRISC1	TRISC0			
TRISD	TRISD7	TRISD6	TRISD5	TRISD4	TRISD3	TRISD2	TRISD1	TRISD0			
SSP1BUF	MSSP1 Receive Buffer/Transmit Register										
SSP1ADD			r (I <sup>2</sup> C Slave r ad Register (I		ode)						
SSP1MSK <sup>(1)</sup>	MSK7	MSK6	MSK5	MSK4	MSK3	MSK2	MSK1	MSK0			
SSP1CON1	WCOL	SSPOV	SSPEN	CKP	SSPM3	SSPM2	SSPM1	SSPM0			
SSP1CON2	GCEN	ACKSTAT	ACKDT	ACKEN	RCEN	PEN	RSEN	SEN			
	GCEN	ACKSTAT	ADMSK5 <sup>(2)</sup>	ADMSK4 <sup>(2)</sup>	ADMSK3 <sup>(2)</sup>	ADMSK2 <sup>(2)</sup>	ADMSK1 <sup>(2)</sup>	SEN			
SSP1STAT	SMP	CKE	D/A	Р	S	R/W	UA	BF			
SSP2BUF			ransmit Regi								
SSP2ADD	MSSP2 Add MSSP2 Bau	lress Registe Id Rate Reloa	r (I <sup>2</sup> C Slave r ad Register (I	node), <sup>2</sup> C Master mo	ode)						
SSP2MSK <sup>(1)</sup>	MSK7	MSK6	MSK5	MSK4	MSK3	MSK2	MSK1	MSK0			
SSP2CON1	WCOL	SSPOV	SSPEN	CKP	SSPM3	SSPM2	SSPM1	SSPM0			
SSP2CON2	GCEN	ACKSTAT	ACKDT	ACKEN	RCEN	PEN	RSEN	SEN			
	GCEN	ACKSTAT	ADMSK5 <sup>(2)</sup>	ADMSK4 <sup>(2)</sup>	ADMSK3 <sup>(2)</sup>	ADMSK2 <sup>(2)</sup>	ADMSK1 <sup>(2)</sup>	SEN			
SSP2STAT	SMP	CKE	D/A	Р	S	R/W	UA	BF			
PMD0	CCP3MD	CCP2MD	CCP1MD	UART2MD	UART1MD	SSP2MD	SSP1MD	ADCMD			
		tod rood oo (									

## TABLE 21-4: REGISTERS ASSOCIATED WITH I<sup>2</sup>C<sup>™</sup> OPERATION

**Legend:** -= unimplemented, read as '0'. Shaded cells are not used by the MSSP module in  $I^2C^{TM}$  mode.

Note 1: SSPxMSK shares the same address in SFR space as SSPxADD, but is only accessible in certain I<sup>2</sup>C Slave operating modes in 7-Bit Masking mode. See Section 21.4.3.4 "7-Bit Address Masking Mode" for more details.

2: Alternate bit definitions for use in I<sup>2</sup>C Slave mode operations only.

## 22.0 ENHANCED UNIVERSAL SYNCHRONOUS ASYNCHRONOUS RECEIVER TRANSMITTER (EUSART)

The Enhanced Universal Synchronous Asynchronous Receiver Transmitter (EUSART) module is one of two serial I/O modules. (Generically, the EUSART is also known as a Serial Communications Interface or SCI.) The EUSART can be configured as a full-duplex, asynchronous system that can communicate with peripheral devices, such as CRT terminals and personal computers. It can also be configured as a half-duplex synchronous system that can communicate with peripheral devices, such as A/D or D/A integrated circuits, serial EEPROMs, etc.

The Enhanced USART module implements additional features, including automatic baud rate detection and calibration, automatic wake-up on Sync Break reception and 12-bit Break character transmit. These make it ideally suited for use in Local Interconnect Network bus (LIN/J2602 bus) systems.

All members of the PIC18F87K22 family are equipped with two independent EUSART modules, referred to as EUSART1 and EUSART2. They can be configured in the following modes:

- Asynchronous (full duplex) with:
  - Auto-wake-up on character reception
  - Auto-baud calibration
  - 12-bit Break character transmission
- Synchronous Master (half duplex) with selectable clock polarity
- Synchronous Slave (half duplex) with selectable clock polarity

The pins of EUSART1 and EUSART2 are multiplexed with the functions of PORTC (RC6/TX1/CK1 and RC7/RX1/DT1) and PORTG (RG1/TX2/CK2/AN19/C3OUT and RG2/RX2/DT2/AN18/C3INA), respectively. In order to configure these pins as an EUSART:

- For EUSART1:
  - Bit, SPEN (RCSTA1<7>), must be set (= 1)
  - Bit, TRISC<7>, must be set (= 1)
  - Bit, TRISC<6>, must be cleared (= 0) for Asynchronous and Synchronous Master modes
  - Bit, TRISC<6>, must be set (= 1) for Synchronous Slave mode
- For EUSART2:
  - Bit, SPEN (RCSTA2<7>), must be set (= 1)
  - Bit, TRISG<2>, must be set (= 1)
  - Bit TRISG<1> must be cleared (= 0) for Asynchronous and Synchronous Master modes
  - Bit, TRISC<6>, must be set (= 1) for Synchronous Slave mode

Note: The EUSART control will automatically reconfigure the pin from input to output as needed.

The operation of each Enhanced USART module is controlled through three registers:

- Transmit Status and Control (TXSTAx)
- Receive Status and Control (RCSTAx)
- Baud Rate Control (BAUDCONx)

These are detailed on the following pages in Register 22-1, Register 22-2 and Register 22-3, respectively.

Note: Throughout this section, references to register and bit names that may be associated with a specific EUSART module are referred to generically by the use of 'x' in place of the specific module number. Thus, "RCSTAx" might refer to the Receive Status register for either EUSART1 or EUSART2.

R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R-1	R/W-0
CSRC	TX9	TXEN <sup>(1)</sup>	SYNC	SENDB	BRGH	TRMT	TX9D
bit 7	·		•				bit
L <b>egend:</b> R = Reada	blo bit	W = Writable	hit	II – Unimplor	nented bit, rea	d ac 'O'	
-n = Value		'1' = Bit is set		'0' = Bit is cle		x = Bit is unkr	
bit 7	CSRC: Cloc	k Source Select	bit				
	Asynchrono Don't care.	<u>us mode:</u>					
		<u>is mode:</u> mode (clock gen ode (clock from					
bit 6	<b>TX9:</b> 9-Bit T	ransmit Enable b	oit				
		9-bit transmissio 8-bit transmissio					
bit 5	TXEN: Tran	smit Enable bit <sup>(1</sup>	)				
		iit is enabled iit is disabled					
bit 4	SYNC: EUS	SART Mode Sele	ct bit				
	1 = Synchro						
L:1 0	-	onous mode	- 4 <b>h</b> : 4				
bit 3		end Break Charad	cter dit				
		/nc Break on ne> eak transmissior			ardware upon c	ompletion)	
	<u>Synchronou</u> Don't care.		·				
bit 2	BRGH: Higl	n Baud Rate Sele	ect bit				
	Asynchrono						
	1 = High sp 0 = Low spe						
	Synchronou						
	Unused in t						
bit 1	TRMT: Tran	smit Shift Regist	er Status bit				
	1 = TSR is 6 0 = TSR is 1						
bit 0	TX9D: 9th b	it of Transmit Da	ita				
	Can be add	ress/data bit or a	parity bit.				

## REGISTER 22-1: TXSTAX: TRANSMIT STATUS AND CONTROL REGISTER

R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R-0	R-0	R-x
SPEN	RX9	SREN	CREN	ADDEN	FERR	OERR	RX9D
oit 7							bit
_egend:							
R = Readab		W = Writable b	oit	U = Unimplem			
n = Value a	t POR	'1' = Bit is set		'0' = Bit is clea	ared	x = Bit is unkn	iown
bit 7	SPEN: Seria	al Port Enable bit					
510 1		ort is enabled					
		ort is disabled (h	eld in Reset)				
oit 6	RX9: 9-Bit F	Receive Enable b	it				
		9-bit reception					
		8-bit reception					
oit 5	Asynchrono	le Receive Enab	le bit				
	Don't care.	us mode.					
	Synchronou	s mode – Master					
		s single receive					
		s single receive eared after recep	tion is comple	ete.			
		s mode – Slave:					
oit 4	CREN: Cont	tinuous Receive	Enable bit				
	Asynchrono						
	1 = Enables						
	0 = Disable Synchronou						
		<u>s continuous rece</u>	eive until enat	ole bit, CREN, is	cleared (CREI	N overrides SR	EN)
		s continuous rec			, ,		,
oit 3	ADDEN: Ad	dress Detect Ena	able bit				
		us mode 9-Bit (R				<i>"</i>   <b>D</b> 0	<b>-</b>
		s address detecti s address detect					
		us mode 8-Bit (R					a party bit
	Don't care.	· · · · · · · · · · · · · · · · · · ·					
bit 2	FERR: Fram	ning Error bit					
	1 = Framing 0 = No fram	g error (can be cl ning error	eared by read	ding the RCREG	x register and	receiving the n	ext valid byte
bit 1	OERR: Ove	rrun Error bit					
		n error (can be cl	eared by clea	ring bit, CREN)			
hit O	0 = No over	rrun error bit of Received Da	ata				
bit 0	KAYD: YIN D	ni or Received Da	ald				

R/W-0	R-1	R/W-0	R/W-0	R/W-0	U-0	R/W-0	R/W-0
ABDOVF	RCIDL	RXDTP	TXCKP	BRG16	_	WUE	ABDEN
bit 7							bit
Legend:							
R = Readabl	e bit	W = Writable	bit	U = Unimplem	ented bit, read	d as '0'	
-n = Value at	POR	'1' = Bit is set		'0' = Bit is clea	red	x = Bit is unki	nown
bit 7	1 = A BRG ro		urred during A	Status bit uto-Baud Rate I	Detect mode (I	must be cleare	d in software)
hit C		rollover has oc					
bit 6		ive Operation I operation is Idle					
		peration is action					
bit 5		Receive Polar					
	Asynchronous		verted (active-				
		mode: <) is inverted (a <) is not inverte					
bit 4	TXCKP: Sync	hronous Clock	Polarity Selec	t bit			
		<u>s mode:</u> for transmit (T) for transmit (T)					
		<u>mode:</u> for clock (CKx) for clock (CKx)					
bit 3	<b>BRG16:</b> 16-B	it Baud Rate R	egister Enable	bit			
				Hx and SPBRG> only (Compatibl		RGHx value ig	nored
bit 2	Unimplemen	ted: Read as '	0'				
bit 1	WUE: Wake-u	up Enable bit					
	hardware		sing edge	RXx pin – interru letected	ipt generated	on falling edge	e; bit cleared i
	Synchronous Unused in this						
bit 0	ABDEN: Auto	-Baud Detect I	Enable bit				
	cleared ir	aud rate meas n hardware upo	on completion.	e next characte has completed	r. Requires re	ception of a Sy	ync field (55h
	Synchronous Unused in this	mode:		,			

## 22.1 Baud Rate Generator (BRG)

The BRG is a dedicated, 8-bit or 16-bit generator that supports both the Asynchronous and Synchronous modes of the EUSART. By default, the BRG operates in 8-bit mode; setting the BRG16 bit (BAUDCONx<3>) selects 16-bit mode.

The SPBRGHx:SPBRGx register pair controls the period of a free-running timer. In Asynchronous mode, bits, BRGH (TXSTAx<2>) and BRG16 (BAUDCONx<3>), also control the baud rate. In Synchronous mode, BRGH is ignored. Table 22-1 shows the formula for computation of the baud rate for different EUSART modes which only apply in Master mode (internally generated clock).

Given the desired baud rate and Fosc, the nearest integer value for the SPBRGHx:SPBRGx registers can be calculated using the formulas in Table 22-1. From this, the error in baud rate can be determined. An example calculation is shown in Example 22-1. Typical baud rates and error values for the various Asynchronous modes are shown in Table 22-2. It may be advantageous to use the high baud rate (BRGH = 1) or the 16-bit BRG to reduce the baud rate error, or achieve a slow baud rate for a fast oscillator frequency.

Writing a new value to the SPBRGHX:SPBRGX registers causes the BRG timer to be reset (or cleared). This ensures the BRG does not wait for a timer overflow before outputting the new baud rate. When operated in the Synchronous mode, SPBRGH:SPBRG values of 0000h and 0001h are not supported. In the Asynchronous mode, all BRG values may be used.

## 22.1.1 OPERATION IN POWER-MANAGED MODES

The device clock is used to generate the desired baud rate. When one of the power-managed modes is entered, the new clock source may be operating at a different frequency. This may require an adjustment to the value in the SPBRGx register pair.

## 22.1.2 SAMPLING

The data on the RXx pin (either RC7/RX1/DT1 or RG2/RX2/DT2/AN18/C3INA) is sampled three times by a majority detect circuit to determine if a high or a low level is present at the RXx pin.

C	onfiguration B	lits	BRG/EUSART Mode	Baud Rate Formula
SYNC	IC BRG16 BRGH		BRG/EUSART Mode	Baud Rate Formula
0	0	0	8-bit/Asynchronous	Fosc/[64 (n + 1)]
0	0	1	8-bit/Asynchronous	$E_{0000}/[16 (n + 1)]$
0	1	0	16-bit/Asynchronous	Fosc/[16 (n + 1)]
0	1	1	16-bit/Asynchronous	
1	0	x	8-bit/Synchronous	Fosc/[4 (n + 1)]
1	1 x		16-bit/Synchronous	

#### TABLE 22-1: BAUD RATE FORMULAS

**Legend:** x = Don't care, n = value of SPBRGHx:SPBRGx register pair

## EXAMPLE 22-1: CALCULATING BAUD RATE ERROR

For a device with FOSC of 10	6 MH	z, desired baud rate of 9600, Asynchronous mode, and 8-bit BRG:
Desired Baud Rate	=	Fosc/(64 ([SPBRGHx:SPBRGx] + 1))
Solving for SPBRGHx:	SPB1	RGx:
Х	=	((FOSC/Desired Baud Rate)/64) – 1
	=	((16000000/9600)/64) - 1
	=	[25.042] = 25
Calculated Baud Rate	=	16000000/(64 (25 + 1))
	=	9615
Error	=	(Calculated Baud Rate - Desired Baud Rate)/Desired Baud Rate
	=	(9615 - 9600)/9600 = 0.16%

## TABLE 22-2: REGISTERS ASSOCIATED WITH BAUD RATE GENERATOR

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0			
TXSTA1	CSRC	TX9	TXEN	SYNC	SENDB	BRGH	TRMT	TX9D			
RCSTA1	SPEN	RX9	SREN	CREN	ADDEN	FERR	OERR	RX9D			
BAUDCON1	ABDOVF	RCIDL	RXDTP	TXCKP	BRG16	_	WUE	ABDEN			
SPBRGH1	EUSART1 B	aud Rate Ge	nerator Regis	ster High Byte	9						
SPBRG1	EUSART1 B	EUSART1 Baud Rate Generator Register									
TXSTA2	CSRC	TX9	TXEN	SYNC	SENDB	BRGH	TRMT	TX9D			
RCSTA2	SPEN	RX9	SREN	CREN	ADDEN	FERR	OERR	RX9D			
BAUDCON2	ABDOVF	RCIDL	RXDTP	TXCKP	BRG16	_	WUE	ABDEN			
SPBRGH2	EUSART2 B	aud Rate Ge	nerator Regis	ster High Byte	;						
SPBRG2	EUSART2 B	EUSART2 Baud Rate Generator Register									
PMD0	CCP3MD	CCP2MD	CCP1MD	UART2MD	UART1MD	SSP2MD	SSP1MD	ADCMD			
Lonondi -	unimplement			II		<u>^</u>					

Legend: — = unimplemented, read as '0'. Shaded cells are not used by the BRG.

					SYNC	= 0, BRGH	l = 0, BRG	<b>616 =</b> 0					
BAUD	FOSC = 40.000 MHZ		) MHz	Fosc = 20.000 MHz			Fosc = 10.000 MHz			Fos	Fosc = 8.000 MHz		
RATE (K)	Actual Rate (K)	% Error	SPBRG value (decimal)	Actual Rate (K)	% Error	SPBRG value (decimal)	Actual Rate (K)	% Error	SPBRG value (decimal)	Actual Rate (K)	% Error	SPBRG value (decimal)	
0.3	_	_			_				_		_		
1.2	—	_	_	1.221	1.73	255	1.202	0.16	129	1.201	-0.16	103	
2.4	2.441	1.73	255	2.404	0.16	129	2.404	0.16	64	2.403	-0.16	51	
9.6	9.615	0.16	64	9.766	1.73	31	9.766	1.73	15	9.615	-0.16	12	
19.2	19.531	1.73	31	19.531	1.73	15	19.531	1.73	7	_	_	_	
57.6	56.818	-1.36	10	62.500	8.51	4	52.083	-9.58	2	—	_	_	
115.2	125.000	8.51	4	104.167	-9.58	2	78.125	-32.18	1	—	_	_	

TABI E 22-3	BAUD RATES FOR ASYNCHRONOUS MODES
IADLL 22-J.	

			S	YNC = 0, E	BRGH = (	, BRG16 =	0		
BAUD	Fos	c = 4.000	MHz	Fos	c = 2.000	MHz	Fos	c = 1.000	MHz
RATE (K)	Actual		SPBRG value (decimal)	Actual Rate (K)	% Error	SPBRG value (decimal)	Actual Rate (K)	% Error	SPBRG value (decimal)
0.3	0.300	0.16	207	0.300	-0.16	103	0.300	-0.16	51
1.2	1.202	0.16	51	1.201	-0.16	25	1.201	-0.16	12
2.4	2.404	0.16	25	2.403	-0.16	12	_	_	—
9.6	8.929	-6.99	6	—	_	_	_	_	_
19.2	20.833	8.51	2	—	_	_	—	_	_
57.6	62.500	8.51	0	—	_	_	—	_	_
115.2	62.500	-45.75	0	_		_	_		—

					SYNC	= 0, BRGH	<b>i</b> = 1, BRG	i <b>16 =</b> 0					
BAUD RATE	Fosc	= 40.000	) MHz	Fosc = 20.000 MHz			Fosc = 10.000 MHz			Fos	Fosc = 8.000 MHz		
(K)	Actual Rate (K)	% Error	SPBRG value (decimal)	Actual Rate (K)	Rate Frror		Actual Rate (K)	% Error	SPBRG value (decimal)	Actual Rate (K)	% Error	SPBRG value (decimal)	
0.3	—	_	_	_	_	_	_	_	_	_	_	_	
1.2	—	—	—	—	—	—	—		—	—	—	—	
2.4	—	_	_	—	_	_	2.441	1.73	255	2.403	-0.16	207	
9.6	9.766	1.73	255	9.615	0.16	129	9.615	0.16	64	9.615	-0.16	51	
19.2	19.231	0.16	129	19.231	0.16	64	19.531	1.73	31	19.230	-0.16	25	
57.6	58.140	0.94	42	56.818	-1.36	21	56.818	-1.36	10	55.555	3.55	8	
115.2	113.636	-1.36	21	113.636	-1.36	10	125.000	8.51	4	—	_	_	

		SYNC = 0, BRGH = 1, BRG16 = 0										
BAUD	Foso	c = 4.000	MHz	Fos	c = 2.000	MHz	Fos	c = 1.000	MHz			
(K)	Actual Rate (K)	% Error	SPBRG value (decimal)	Actual Rate (K)	% Error	SPBRG value (decimal)	Actual Rate (K)	% Error	SPBRG value (decimal)			
0.3	_		_		_	_	0.300	-0.16	207			
1.2	1.202	0.16	207	1.201	-0.16	103	1.201	-0.16	51			
2.4	2.404	0.16	103	2.403	-0.16	51	2.403	-0.16	25			
9.6	9.615	0.16	25	9.615	-0.16	12	_	_	_			
19.2	19.231	0.16	12	_	_	_	_	_	_			
57.6	62.500	8.51	3	—	_	_	—	_	_			
115.2	125.000	8.51	1		_	—	_	_	—			

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		SYNC = 0, BRGH = 0, BRG16 = 1											
BAUD	Fosc = 40.000 MHz			Fosc = 20.000 MHz			Fosc = 10.000 MHz			Fosc = 8.000 MHz			
(K)	A 641161		SPBRG value (decimal)	Actual Rate (K)	% Error	SPBRG value (decimal)	Actual Rate (K)	% Error	SPBRG value (decimal)	Actual Rate (K)	% Error	SPBRG value (decimal)	
0.3	0.300	0.00	8332	0.300	0.02	4165	0.300	0.02	2082	0.300	-0.04	1665	
1.2	1.200	0.02	2082	1.200	-0.03	1041	1.200	-0.03	520	1.201	-0.16	415	
2.4	2.402	0.06	1040	2.399	-0.03	520	2.404	0.16	259	2.403	-0.16	207	
9.6	9.615	0.16	259	9.615	0.16	129	9.615	0.16	64	9.615	-0.16	51	
19.2	19.231	0.16	129	19.231	0.16	64	19.531	1.73	31	19.230	-0.16	25	
57.6	58.140	0.94	42	56.818	-1.36	21	56.818	-1.36	10	55.555	3.55	8	
115.2	113.636	-1.36	21	113.636	-1.36	10	125.000	8.51	4	_	_	—	

## TABLE 22-3: BAUD RATES FOR ASYNCHRONOUS MODES (CONTINUED)

		SYNC = 0, BRGH = 0, BRG16 = 1													
BAUD	Foso	c = 4.000	MHz	Fos	c = 2.000	MHz	Fosc = 1.000 MHz								
(K)	Actual Rate (K)	% Error	SPBRG value (decimal)	Actual Rate (K)	% Error	SPBRG value (decimal)	Actual Rate (K)	% Error	SPBRG value (decimal)						
0.3	0.300	0.04	832	0.300	-0.16	415	0.300	-0.16	207						
1.2	1.202	0.16	207	1.201	-0.16	103	1.201	-0.16	51						
2.4	2.404	0.16	103	2.403	-0.16	51	2.403	-0.16	25						
9.6	9.615	0.16	25	9.615	-0.16	12	—	_	_						
19.2	19.231	0.16	12	—	_	_	—	_	_						
57.6	62.500	8.51	3	—	_	_	—	_	_						
115.2	125.000	8.51	1	_	_	—	_	_	_						

				SYNC = 0,	BRGH =	= 1, BRG16	5 = 1 or SY	NC = 1,	BRG16 = 1			
BAUD RATE	Fosc = 40.000 MHz			Fosc = 20.000 MHz			Fosc = 10.000 MHz			Fosc = 8.000 MHz		
(K)	Actual Rate (K)	% Error	SPBRG value (decimal)									
0.3	0.300	0.00	33332	0.300	0.00	16665	0.300	0.00	8332	0.300	-0.01	6665
1.2	1.200	0.00	8332	1.200	0.02	4165	1.200	0.02	2082	1.200	-0.04	1665
2.4	2.400	0.02	4165	2.400	0.02	2082	2.402	0.06	1040	2.400	-0.04	832
9.6	9.606	0.06	1040	9.596	-0.03	520	9.615	0.16	259	9.615	-0.16	207
19.2	19.193	-0.03	520	19.231	0.16	259	19.231	0.16	129	19.230	-0.16	103
57.6	57.803	0.35	172	57.471	-0.22	86	58.140	0.94	42	57.142	0.79	34
115.2	114.943	-0.22	86	116.279	0.94	42	113.636	-1.36	21	117.647	-2.12	16

		SYN	IC = 0, BR(	GH = 1, BI	RG16 = 1	or SYNC =	= 1, BRG1	6 = 1		
BAUD	Fost	c = 4.000	MHz	Fos	c = 2.000	MHz	Fosc = 1.000 MHz			
(K)	Actual Rate (K)	% Error	SPBRG value (decimal)	Actual Rate (K)	% Error	SPBRG value (decimal)	Actual Rate (K)	% Error	SPBRG value (decimal)	
0.3	0.300	0.01	3332	0.300	-0.04	1665	0.300	-0.04	832	
1.2	1.200	0.04	832	1.201	-0.16	415	1.201	-0.16	207	
2.4	2.404	0.16	415	2.403	-0.16	207	2.403	-0.16	103	
9.6	9.615	0.16	103	9.615	-0.16	51	9.615	-0.16	25	
19.2	19.231	0.16	51	19.230	-0.16	25	19.230	-0.16	12	
57.6	58.824	2.12	16	55.555	3.55	8	—	—	—	
115.2	111.111	-3.55	8	—	_	_	—	_	—	

## 22.1.3 AUTO-BAUD RATE DETECT

The Enhanced USART module supports the automatic detection and calibration of baud rate. This feature is active only in Asynchronous mode and while the WUE bit is clear.

The automatic baud rate measurement sequence (Figure 22-1) begins whenever a Start bit is received and the ABDEN bit is set. The calculation is self-averaging.

In the Auto-Baud Rate Detect (ABD) mode, the clock to the BRG is reversed. Rather than the BRG clocking the incoming RXx signal, the RXx signal is timing the BRG. In ABD mode, the internal Baud Rate Generator is used as a counter to time the bit period of the incoming serial byte stream.

Once the ABDEN bit is set, the state machine will clear the BRG and look for a Start bit. The Auto-Baud Rate Detect must receive a byte with the value, 55h (ASCII "U", which is also the LIN/J2602 bus Sync character), in order to calculate the proper bit rate. The measurement is taken over both a low and a high bit time in order to minimize any effects caused by asymmetry of the incoming signal. After a Start bit, the SPBRGx begins counting up, using the preselected clock source on the first rising edge of RXx. After eight bits on the RXx pin or the fifth rising edge, an accumulated value totalling the proper BRG period is left in the SPBRGHx:SPBRGx register pair. Once the 5th edge is seen (this should correspond to the Stop bit), the ABDEN bit is automatically cleared.

If a rollover of the BRG occurs (an overflow from FFFFh to 0000h), the event is trapped by the ABDOVF status bit (BAUDCONx<7>). It is set in hardware by BRG rollovers and can be set or cleared by the user in software. ABD mode remains active after rollover events and the ABDEN bit remains set (Figure 22-2).

While calibrating the baud rate period, the BRG registers are clocked at 1/8th the preconfigured clock rate. The BRG clock will be configured by the BRG16 and BRGH bits. The BRG16 bit must be set to use both SPBRG1 and SPBRGH1 as a 16-bit counter. This allows the user to verify that no carry occurred for 8-bit modes by checking for 00h in the SPBRGHx register. Refer to Table 22-4 for counter clock rates to the BRG.

While the ABD sequence takes place, the EUSART state machine is held in Idle. The RCxIF interrupt is set once the fifth rising edge on RXx is detected. The value in the RCREGx needs to be read to clear the RCxIF interrupt. The contents of RCREGx should be discarded.

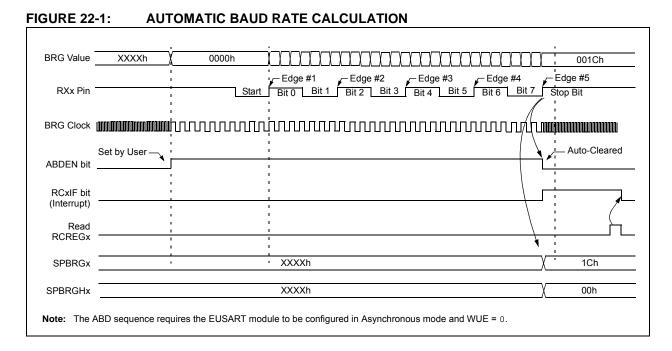
- Note 1: If the WUE bit is set with the ABDEN bit, Auto-Baud Rate Detection will occur on the byte *following* the Break character.
  - 2: It is up to the user to determine that the incoming character baud rate is within the range of the selected BRG clock source. Some combinations of oscillator frequency and EUSART baud rates are not possible due to bit error rates. Overall system timing and communication baud rates must be taken into consideration when using the Auto-Baud Rate Detection feature.
  - To maximize baud rate range, if that feature is used, it is recommended that the BRG16 bit (BAUDCONx<3>) be set.

## TABLE 22-4:BRG COUNTER<br/>CLOCK RATES

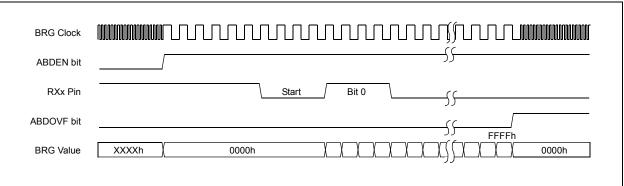
BRG16	BRGH	BRG Counter Clock
0	0	Fosc/512
0	1	Fosc/128
1	0	Fosc/128
1	1	Fosc/32

## 22.1.3.1 ABD and EUSART Transmission

Since the BRG clock is reversed during ABD acquisition, the EUSART transmitter cannot be used during ABD. This means that whenever the ABDEN bit is set, TXREGx cannot be written to. Users should also ensure that ABDEN does not become set during a transmit sequence. Failing to do this may result in unpredictable EUSART operation.



#### FIGURE 22-2: BRG OVERFLOW SEQUENCE



## 22.2 EUSART Asynchronous Mode

The Asynchronous mode of operation is selected by clearing the SYNC bit (TXSTAx<4>). In this mode, the EUSART uses standard Non-Return-to-Zero (NRZ) format (one Start bit, eight or nine data bits and one Stop bit). The most common data format is 8 bits. An on-chip, dedicated 8-bit/16-bit Baud Rate Generator can be used to derive standard baud rate frequencies from the oscillator.

The EUSART transmits and receives the LSb first. The EUSART's transmitter and receiver are functionally independent but use the same data format and baud rate. The Baud Rate Generator produces a clock, either x16 or x64 of the bit shift rate, depending on the BRGH and BRG16 bits (TXSTAx<2> and BAUDCONx<3>). Parity is not supported by the hardware but can be implemented in software and stored as the 9th data bit.

When operating in Asynchronous mode, the EUSART module consists of the following important elements:

- Baud Rate Generator
- Sampling Circuit
- Asynchronous Transmitter
- Asynchronous Receiver
- Auto-Wake-up on Sync Break Character
- 12-Bit Break Character Transmit
- Auto-Baud Rate Detection

#### 22.2.1 EUSART ASYNCHRONOUS TRANSMITTER

The EUSART transmitter block diagram is shown in Figure 22-3. The heart of the transmitter is the Transmit (Serial) Shift Register (TSR). The Shift register obtains its data from the Read/Write Transmit Buffer register, TXREGx. The TXREGx register is loaded with data in software. The TSR register is not loaded until the Stop bit has been transmitted from the previous load. As soon as the Stop bit is transmitted, the TSR is loaded with new data from the TXREGx register (if available).

Once the TXREGx register transfers the data to the TSR register (occurs in one TcY), the TXREGx register is empty and the TXxIF flag bit is set. This interrupt can be enabled or disabled by setting or clearing the interrupt enable bit, TXxIE. TXxIF will be set regardless of the state of TXxIE; it cannot be cleared in software. TXxIF is also not cleared immediately upon loading TXREGx, but becomes valid in the second instruction cycle following the load instruction. Polling TXxIF immediately following a load of TXREGx will return invalid results.

While TXxIF indicates the status of the TXREGx register; another bit, TRMT (TXSTAx<1>), shows the status of the TSR register. TRMT is a read-only bit which is set when the TSR register is empty. No interrupt logic is tied to this bit so the user has to poll this bit in order to determine if the TSR register is empty.

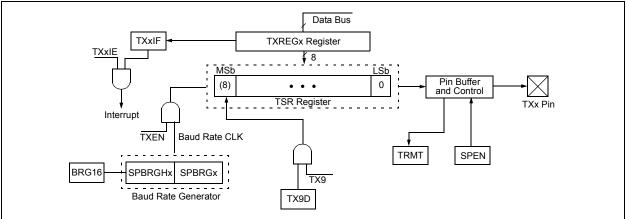
**Note 1:** The TSR register is not mapped in data memory, so it is not available to the user.

2: Flag bit, TXxIF, is set when enable bit, TXEN, is set.

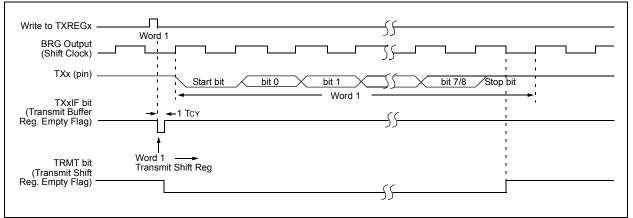
To set up an Asynchronous Transmission:

- 1. Initialize the SPBRGHx:SPBRGx registers for the appropriate baud rate. Set or clear the BRGH and BRG16 bits, as required, to achieve the desired baud rate.
- 2. Enable the asynchronous serial port by clearing bit, SYNC, and setting bit, SPEN.
- 3. If interrupts are desired, set enable bit, TXxIE.
- 4. If 9-bit transmission is desired, set transmit bit, TX9; can be used as an address/data bit.
- 5. Enable the transmission by setting bit, TXEN, which will also set bit, TXxIF.
- 6. If 9-bit transmission is selected, the ninth bit should be loaded in bit, TX9D.
- 7. Load data to the TXREGx register (starts transmission).
- If using interrupts, ensure that the GIE and PEIE bits in the INTCON register (INTCON<7:6>) are set.

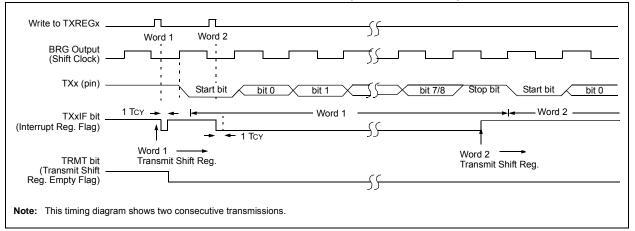
## FIGURE 22-3: EUSART TRANSMIT BLOCK DIAGRAM







## FIGURE 22-5: ASYNCHRONOUS TRANSMISSION (BACK-TO-BACK)



TADLE 22-J.													
Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0					
INTCON	GIE/GIEH	PEIE/GIEL	TMR0IE	INT0IE	RBIE	TMR0IF	INT0IF	RBIF					
PIR1	PSPIF	ADIF	RC1IF	TX1IF	SSP1IF	TMR1GIF	TMR2IF	TMR1IF					
PIE1	PSPIE	ADIE	RC1IE	TX1IE	SSP1IE	TMR1GIE	TMR2IE	TMR1IE					
IPR1	PSPIP	ADIP	RC1IP	TX1IP	SSP1IP	TMR1GIP	TMR2IP	TMR1IP					
PIR3	TMR5GIF	—	RC2IF	TX2IF	CTMUIF	CCP2IF	CCP1IF	RTCCIF					
PIE3	TMR5GIE	_	RC2IE	TX2IE	CTMUIE	CCP2IE	CCP1IE	RTCCIE					
IPR3	TMR5GIP	_	RC2IP	TX2IP	CTMUIP	CCP2IP	CCP1IP	RTCCIP					
RCSTA1	SPEN	RX9	SREN	CREN	ADDEN	FERR	OERR	RX9D					
TXREG1	EUSART1 T	ransmit Regi	ster										
TXSTA1	CSRC	TX9	TXEN	SYNC	SENDB	BRGH	TRMT	TX9D					
BAUDCON1	ABDOVF	RCIDL	RXDTP	TXCKP	BRG16		WUE	ABDEN					
SPBRGH1	EUSART1 E	aud Rate Ge	enerator Reg	ister High By	te								
SPBRG1	EUSART1 E	Baud Rate Ge	enerator Reg	ister									
RCSTA2	SPEN	RX9	SREN	CREN	ADDEN	FERR	OERR	RX9D					
TXREG2	EUSART2 T	ransmit Regi	ster										
TXSTA2	CSRC	TX9	TXEN	SYNC	SENDB	BRGH	TRMT	TX9D					
BAUDCON2	ABDOVF	RCIDL	RXDTP	TXCKP	BRG16	_	WUE	ABDEN					
SPBRGH2	EUSART2 Baud Rate Generator Register High Byte												
SPBRG2	EUSART2 Baud Rate Generator Register												
ODCON3	U2OD	U10D						CTMUDS					
PMD0	CCP3MD	CCP2MD	CCP1MD	UART2MD	UART1MD	SSP2MD	SSP1MD	ADCMD					
legend: _ =	unimplement	ed locations	read as '0' S	haded cells a	are not used	for asynchro	nous transm	ission					

## TABLE 22-5: REGISTERS ASSOCIATED WITH ASYNCHRONOUS TRANSMISSION

Legend: - = unimplemented locations read as '0'. Shaded cells are not used for asynchronous transmission.

#### 22.2.2 EUSART ASYNCHRONOUS RECEIVER

The receiver block diagram is shown in Figure 22-6. The data is received on the RXx pin and drives the data recovery block. The data recovery block is actually a high-speed shifter operating at x16 times the baud rate, whereas the main receive serial shifter operates at the bit rate or at Fosc. This mode would typically be used in RS-232 systems.

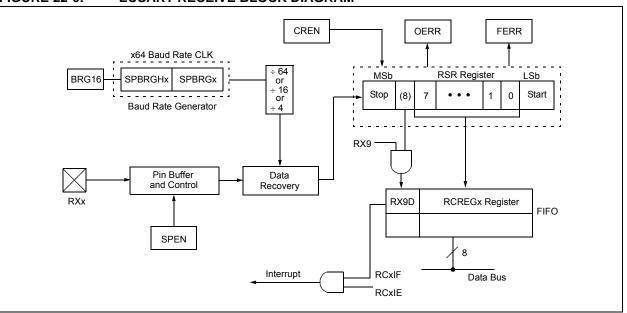
To set up an Asynchronous Reception:

- 1. Initialize the SPBRGHx:SPBRGx registers for the appropriate baud rate. Set or clear the BRGH and BRG16 bits, as required, to achieve the desired baud rate.
- 2. Enable the asynchronous serial port by clearing bit, SYNC, and setting bit, SPEN.
- 3. If interrupts are desired, set enable bit, RCxIE.
- 4. If 9-bit reception is desired, set bit, RX9.
- 5. Enable the reception by setting bit, CREN.
- Flag bit, RCxIF, will be set when reception is complete and an interrupt will be generated if enable bit, RCxIE, was set.
- 7. Read the RCSTAx register to get the 9th bit (if enabled) and determine if any error occurred during reception.
- 8. Read the 8-bit received data by reading the RCREGx register.
- 9. If any error occurred, clear the error by clearing enable bit, CREN.
- 10. If using interrupts, ensure that the GIE and PEIE bits in the INTCON register (INTCON<7:6>) are set.

## 22.2.3 SETTING UP 9-BIT MODE WITH ADDRESS DETECT

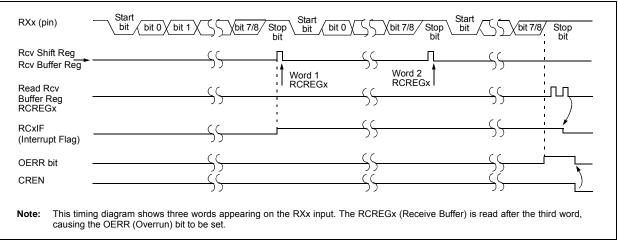
This mode would typically be used in RS-485 systems. To set up an Asynchronous Reception with Address Detect Enable:

- 1. Initialize the SPBRGHx:SPBRGx registers for the appropriate baud rate. Set or clear the BRGH and BRG16 bits, as required, to achieve the desired baud rate.
- 2. Enable the asynchronous serial port by clearing the SYNC bit and setting the SPEN bit.
- 3. If interrupts are required, set the RCEN bit and select the desired priority level with the RCxIP bit.
- 4. Set the RX9 bit to enable 9-bit reception.
- 5. Set the ADDEN bit to enable address detect.
- 6. Enable reception by setting the CREN bit.
- The RCxIF bit will be set when reception is complete. The interrupt will be Acknowledged if the RCxIE and GIE bits are set.
- 8. Read the RCSTAx register to determine if any error occurred during reception, as well as read bit 9 of data (if applicable).
- 9. Read RCREGx to determine if the device is being addressed.
- 10. If any error occurred, clear the CREN bit.
- 11. If the device has been addressed, clear the ADDEN bit to allow all received data into the receive buffer and interrupt the CPU.



## FIGURE 22-6: EUSART RECEIVE BLOCK DIAGRAM

## FIGURE 22-7: ASYNCHRONOUS RECEPTION



#### TABLE 22-6: REGISTERS ASSOCIATED WITH ASYNCHRONOUS RECEPTION

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0		
INTCON	GIE/GIEH	PEIE/GIEL	TMR0IE	INT0IE	RBIE	TMR0IF	INT0IF	RBIF		
PIR1	PSPIF	ADIF	RC1IF	TX1IF	SSP1IF	TMR1GIF	TMR2IF	TMR1IF		
PIE1	PSPIE	ADIE	RC1IE	TX1IE	SSP1IE	TMR1GIE	TMR2IE	TMR1IE		
IPR1	PSPIP	ADIP	RC1IP	TX1IP	SSP1IP	TMR1GIP	TMR2IP	TMR1IP		
PIR3	TMR5GIF	—	RC2IF	TX2IF	CTMUIF	CCP2IF	CCP1IF	RTCCIF		
PIE3	TMR5GIE		RC2IE	TX2IE	CTMUIE	CCP2IE	CCP1IE	RTCCIE		
IPR3	TMR5GIP	_	RC2IP	TX2IP	CTMUIP	CCP2IP	CCP1IP	RTCCIP		
RCSTA1	SPEN	RX9	SREN	CREN	ADDEN	FERR	OERR	RX9D		
RCREG1	EUSART1 F	Receive Regist	ter							
TXSTA1	CSRC	TX9	TXEN	SYNC	SENDB	BRGH	TRMT	TX9D		
BAUDCON1	ABDOVF	RCIDL	RXDTP	TXCKP	BRG16	—	WUE	ABDEN		
SPBRGH1	EUSART1 E	aud Rate Ger	nerator Regis	ster High Byte	9					
SPBRG1	EUSART1 E	aud Rate Ger	nerator Regis	ster						
RCSTA2	SPEN	RX9	SREN	CREN	ADDEN	FERR	OERR	RX9D		
RCREG2	EUSART2 F	Receive Regist	ter							
TXSTA2	CSRC	TX9	TXEN	SYNC	SENDB	BRGH	TRMT	TX9D		
BAUDCON2	ABDOVF	RCIDL	RXDTP	TXCKP	BRG16	_	WUE	ABDEN		
SPBRGH2	EUSART2 Baud Rate Generator Register High Byte									
SPBRG2	EUSART2 E	aud Rate Ger	nerator Regis	ster						
ODCON3	U2OD	U10D	—	—	_	_	—	CTMUDS		
PMD0	CCP3MD	CCP2MD	CCP1MD	UART2MD	UART1MD	SSP2MD	SSP1MD	ADCMD		

**Legend:** — = unimplemented locations read as '0'. Shaded cells are not used for asynchronous reception.

#### 22.2.4 AUTO-WAKE-UP ON SYNC BREAK CHARACTER

During Sleep mode, all clocks to the EUSART are suspended. Because of this, the Baud Rate Generator is inactive and a proper byte reception cannot be performed. The auto-wake-up feature allows the controller to wake up due to activity on the RXx/DTx line while the EUSART is operating in Asynchronous mode.

The auto-wake-up feature is enabled by setting the WUE bit (BAUDCONx<1>). Once set, the typical receive sequence on RXx/DTx is disabled and the EUSART remains in an Idle state, monitoring for a wake-up event independent of the CPU mode. A wake-up event consists of a high-to-low transition on the RXx/DTx line. (This coincides with the start of a Sync Break or a Wake-up Signal character for the LIN/J2602 protocol.)

Following a wake-up event, the module generates an RCxIF interrupt. The interrupt is generated synchronously to the Q clocks in normal operating modes (Figure 22-8) and asynchronously if the device is in Sleep mode (Figure 22-9). The interrupt condition is cleared by reading the RCREGx register.

The WUE bit is automatically cleared once a low-to-high transition is observed on the RXx line following the wake-up event. At this point, the EUSART module is in Idle mode and returns to normal operation. This signals to the user that the Sync Break event is over.

## 22.2.4.1 Special Considerations Using Auto-Wake-up

Since auto-wake-up functions by sensing rising edge transitions on RXx/DTx, information with any state changes before the Stop bit may signal a false End-of-Character (EOC) and cause data or framing errors. To work properly, therefore, the initial character in the transmission must be all '0's. This can be 00h (8 bits) for standard RS-232 devices or 000h (12 bits) for the LIN/J2602 bus.

Oscillator start-up time must also be considered, especially in applications using oscillators with longer start-up intervals (i.e., HS or HSPLL mode). The Sync Break (or Wake-up Signal) character must be of sufficient length and be followed by a sufficient interval to allow enough time for the selected oscillator to start and provide proper initialization of the EUSART.

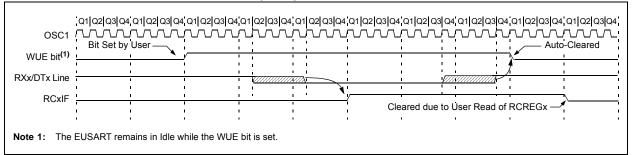
## 22.2.4.2 Special Considerations Using the WUE Bit

The timing of WUE and RCxIF events may cause some confusion when it comes to determining the validity of received data. As noted, setting the WUE bit places the EUSART in an Idle mode. The wake-up event causes a receive interrupt by setting the RCxIF bit. The WUE bit is cleared after this when a rising edge is seen on RXx/DTx. The interrupt condition is then cleared by reading the RCREGx register. Ordinarily, the data in RCREGx will be dummy data and should be discarded.

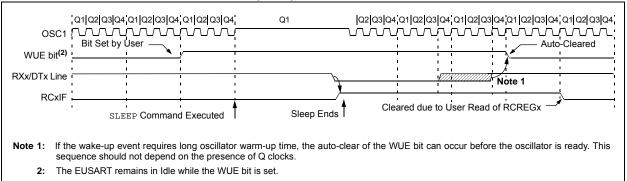
The fact that the WUE bit has been cleared (or is still set), and the RCxIF flag is set, should not be used as an indicator of the integrity of the data in RCREGx. Users should consider implementing a parallel method in firmware to verify received data integrity.

To assure that no actual data is lost, check the RCIDL bit to verify that a receive operation is not in process. If a receive operation is not occurring, the WUE bit may then be set just prior to entering Sleep mode.

## FIGURE 22-8: AUTO-WAKE-UP BIT (WUE) TIMINGS DURING NORMAL OPERATION



## FIGURE 22-9: AUTO-WAKE-UP BIT (WUE) TIMINGS DURING SLEEP



## 22.2.5 BREAK CHARACTER SEQUENCE

The EUSART module has the capability of sending the special Break character sequences that are required by the LIN/J2602 bus standard. The Break character transmit consists of a Start bit, followed by twelve '0' bits and a Stop bit. The Frame Break character is sent whenever the SENDB and TXEN bits (TXSTAx<3> and TXSTAx<5>, respectively) are set while the Transmit Shift Register is loaded with data. Note that the value of data written to TXREGx will be ignored and all '0's will be transmitted.

The SENDB bit is automatically reset by hardware after the corresponding Stop bit is sent. This allows the user to preload the transmit FIFO with the next transmit byte following the Break character (typically, the Sync character in the LIN/J2602 specification).

Note that the data value written to the TXREGx for the Break character is ignored. The write simply serves the purpose of initiating the proper sequence.

The TRMT bit indicates when the transmit operation is active or Idle, just as it does during normal transmission. See Figure 22-10 for the timing of the Break character sequence.

## 22.2.5.1 Break and Sync Transmit Sequence

The following sequence will send a message frame header made up of a Break, followed by an Auto-Baud Sync byte. This sequence is typical of a LIN/J2602 bus master.

- 1. Configure the EUSART for the desired mode.
- 2. Set the TXEN and SENDB bits to set up the Break character.
- 3. Load the TXREGx with a dummy character to initiate transmission (the value is ignored).
- 4. Write '55h' to TXREGx to load the Sync character into the transmit FIFO buffer.
- 5. After the Break has been sent, the SENDB bit is reset by hardware. The Sync character now transmits in the preconfigured mode.

When the TXREGx becomes empty, as indicated by the TXxIF, the next data byte can be written to TXREGx.

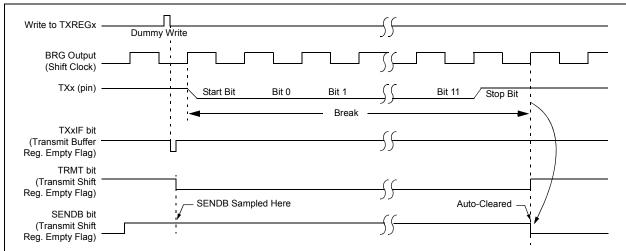
## 22.2.6 RECEIVING A BREAK CHARACTER

The Enhanced USART module can receive a Break character in two ways.

The first method forces configuration of the baud rate at a frequency of 9/13 the typical speed. This allows for the Stop bit transition to be at the correct sampling location (13 bits for Break versus Start bit and 8 data bits for typical data).

The second method uses the auto-wake-up feature described in Section 22.2.4 "Auto-Wake-up on Sync Break Character". By enabling this feature, the EUSART will sample the next two transitions on RXx/DTx, cause an RCxIF interrupt and receive the next data byte followed by another interrupt.

Note that following a Break character, the user will typically want to enable the Auto-Baud Rate Detect feature. For both methods, the user can set the ABDEN bit once the TXxIF interrupt is observed.



## FIGURE 22-10: SEND BREAK CHARACTER SEQUENCE

## 22.3 EUSART Synchronous Master Mode

The Synchronous Master mode is entered by setting the CSRC bit (TXSTAx<7>). In this mode, the data is transmitted in a half-duplex manner (i.e., transmission and reception do not occur at the same time). When transmitting data, the reception is inhibited and vice versa. Synchronous mode is entered by setting bit, SYNC (TXSTAx<4>). In addition, enable bit, SPEN (RCSTAx<7>), is set in order to configure the TXx and RXx pins to CKx (clock) and DTx (data) lines, respectively.

The Master mode indicates that the processor transmits the master clock on the CKx line. Clock polarity is selected with the TXCKP bit (BAUDCONx<4>). Setting TXCKP sets the Idle state on CKx as high, while clearing the bit sets the Idle state as low. This option is provided to support Microwire devices with this module.

#### 22.3.1 EUSART SYNCHRONOUS MASTER TRANSMISSION

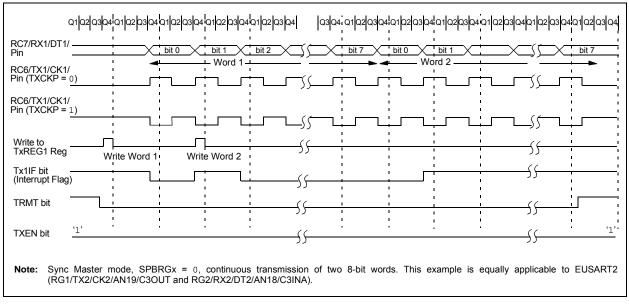
The EUSART transmitter block diagram is shown in Figure 22-3. The heart of the transmitter is the Transmit (Serial) Shift Register (TSR). The Shift register obtains its data from the Read/Write Transmit Buffer register, TXREGx. The TXREGx register is loaded with data in software. The TSR register is not loaded until the last bit has been transmitted from the previous load. As soon as the last bit is transmitted, the TSR is loaded with new data from the TXREGx (if available).

Once the TXREGx register transfers the data to the TSR register (occurs in one TcY), the TXREGx is empty and the TXxIF flag bit is set. The interrupt can be enabled or disabled by setting or clearing the interrupt enable bit, TXxIE. TXxIF is set regardless of the state of enable bit, TXxIE; it cannot be cleared in software. It will reset only when new data is loaded into the TXREGx register.

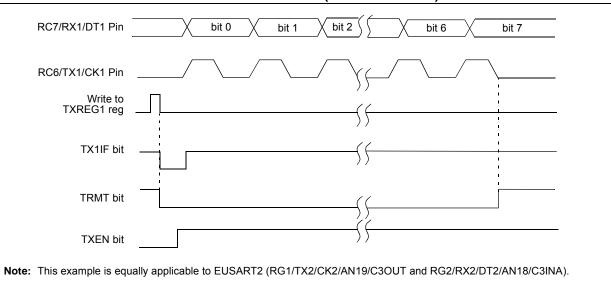
While flag bit, TXxIF, indicates the status of the TXREGx register, another bit, TRMT (TXSTAx<1>), shows the status of the TSR register. TRMT is a read-only bit which is set when the TSR is empty. No interrupt logic is tied to this bit, so the user must poll this bit in order to determine if the TSR register is empty. The TSR is not mapped in data memory so it is not available to the user.

To set up a Synchronous Master Transmission:

- Initialize the SPBRGHx:SPBRGx registers for the appropriate baud rate. Set or clear the BRG16 bit, as required, to achieve the desired baud rate.
- 2. Enable the synchronous master serial port by setting bits, SYNC, SPEN and CSRC.
- 3. If interrupts are desired, set enable bit, TXxIE.
- 4. If 9-bit transmission is desired, set bit, TX9.
- 5. Enable the transmission by setting bit, TXEN.
- 6. If 9-bit transmission is selected, the ninth bit should be loaded in bit, TX9D.
- 7. Start transmission by loading data to the TXREGx register.
- If using interrupts, ensure that the GIE and PEIE bits in the INTCON register (INTCON<7:6>) are set.



## FIGURE 22-11: SYNCHRONOUS TRANSMISSION



Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0			
INTCON	GIE/GIEH	PEIE/GIEL	TMR0IE	INT0IE	RBIE	TMR0IF	INT0IF	RBIF			
PIR1	PSPIF	ADIF	RC1IF	TX1IF	SSP1IF	TMR1GIF	TMR2IF	TMR1IF			
PIE1	PSPIE	ADIE	RC1IE	TX1IE	SSP1IE	TMR1GIE	TMR2IE	TMR1IE			
IPR1	PSPIP	ADIP	RC1IP	TX1IP	SSP1IP	TMR1GIP	TMR2IP	TMR1IP			
PIR3	TMR5GIF	_	RC2IF	TX2IF	CTMUIF	CCP2IF	CCP1IF	RTCCIF			
PIE3	TMR5GIE	_	RC2IE	TX2IE	CTMUIE	CCP2IE	CCP1IE	RTCCIE			
IPR3	TMR5GIP	_	RC2IP	TX2IP	CTMUIP	CCP2IP	CCP1IP	RTCCIP			
RCSTA1	SPEN	RX9	SREN	CREN	ADDEN	FERR	OERR	RX9D			
TXREG1	EUSART1 T	ransmit Regis	ster								
TXSTA1	CSRC	TX9	TXEN	SYNC	SENDB	BRGH	TRMT	TX9D			
BAUDCON1	ABDOVF	RCIDL	RXDTP	TXCKP	BRG16	_	WUE	ABDEN			
SPBRGH1	EUSART1 E	aud Rate Ge	nerator Regis	ster High Byt	e						
SPBRG1	EUSART1 E	aud Rate Ge	nerator Regis	ster							
RCSTA2	SPEN	RX9	SREN	CREN	ADDEN	FERR	OERR	RX9D			
TXREG2	EUSART2 T	ransmit Regis	ster								
TXSTA2	CSRC	TX9	TXEN	SYNC	SENDB	BRGH	TRMT	TX9D			
BAUDCON2	ABDOVF	RCIDL	RXDTP	TXCKP	BRG16	_	WUE	ABDEN			
SPBRGH2	EUSART2 Baud Rate Generator Register High Byte										
SPBRG2	EUSART2 Baud Rate Generator Register										
ODCON3	U2OD	U10D	_			_		CTMUDS			
PMD0	CCP3MD	CCP2MD	CCP1MD	UART2MD	UART1MD	SSP2MD	SSP1MD	ADCMD			

Legend: — = unimplemented, read as '0'. Shaded cells are not used for synchronous master transmission.

#### 22.3.2 EUSART SYNCHRONOUS MASTER RECEPTION

Once Synchronous mode is selected, reception is enabled by setting either the Single Receive Enable bit, SREN (RCSTAx<5>), or the Continuous Receive Enable bit, CREN (RCSTAx<4>). Data is sampled on the RXx pin on the falling edge of the clock.

If enable bit, SREN, is set, only a single word is received. If enable bit, CREN, is set, the reception is continuous until CREN is cleared. If both bits are set, then CREN takes precedence.

To set up a Synchronous Master Reception:

- 1. Initialize the SPBRGHx:SPBRGx registers for the appropriate baud rate. Set or clear the BRG16 bit, as required, to achieve the desired baud rate.
- 2. Enable the synchronous master serial port by setting bits, SYNC, SPEN and CSRC.

- 3. Ensure bits, CREN and SREN, are clear.
- 4. If interrupts are desired, set enable bit, RCxIE.
- 5. If 9-bit reception is desired, set bit, RX9.
- 6. If a single reception is required, set bit, SREN. For continuous reception, set bit, CREN.
- 7. Interrupt flag bit, RCxIF, will be set when reception is complete and an interrupt will be generated if the enable bit, RCxIE, was set.
- 8. Read the RCSTAx register to get the 9th bit (if enabled) and determine if any error occurred during reception.
- 9. Read the 8-bit received data by reading the RCREGx register.
- 10. If any error occurred, clear the error by clearing bit CREN.
- 11. If using interrupts, ensure that the GIE and PEIE bits in the INTCON register (INTCON<7:6>) are set.

## FIGURE 22-13: SYNCHRONOUS RECEPTION (MASTER MODE, SREN)

	bit 7	×:	bit 6	bit 5	bit 4	bit 3	bit 2	bit 1	bit 0		RC7/RX1/DT1 — Pin
	1 1 1	:	Ŀ÷Ĺ		÷∟						RC6/TX1/CK1 1 (TXCKP = 0)
											RC6/TX1/CK1 n (TXCKP = 1)
	1 1 1 1	1 1 1			• • •				1 1 1		Write to bit, SREN
	<u> </u>							:	i		SREN bit
'0	i									· ·	CREN bit
	÷			1 1 1	1 1 1			1 1 1	י י <u>י</u>		RC1IF bit (Interrupt)——
╧	i	•	1		<u> </u>					· ·	Read RCREG1 —

Note: Timing diagram demonstrates Sync Master mode with bit, SREN = 1, and bit, BRGH = 0. This example is equally applicable to EUSART2 (RG1/TX2/CK2/AN19/C3OUT and RG2/RX2/DT2/AN18/C3INA).

PIE1PSPIEADIERC1IETX1IESSP1IETMR1GIETMR2IETMR1IIIIPR1PSPIPADIPRC1IPTX1IPSSP1IPTMR1GIPTMR2IPTMR1IIIPIR3TMR5GIF—RC2IFTX2IFCTMUIFCCP2IFCCP1IFRTCCIIPIE3TMR5GIE—RC2IETX2IECTMUIFCCP2IFCCP1IPRTCCIIIPR3TMR5GIP—RC2IPTX2IPCTMUIPCCP2IPCCP1IPRTCCIIRCSTA1SPENRX9SRENCRENADDENFERROERRRX9DRCREG1EUSART1 Receive RegisterTX2NSYNCSENDBBRGHTRMTTX9DBAUDCON1ABDOVFRCIDLRXDTPTXCKPBRG16—WUEABDENSPBRG1EUSART1 Baud Rate Generator RegisterRCSTA2SPENRX9SRENCRENADDENFERROERRRX9DRCREG2EUSART2 Receive RegisterTXENSYNCSENDBBRGHTRMTTX9DBAUDCON2ABDOVFRCIDLRXDTPTXCKPBRG16—WUEABDENSPBRG12EUSART2 Baud Rate Generator RegisterTXCKPBRG16—WUEABDENSPBRG2EUSART2 Baud Rate Generator RegisterGOCON3U2ODU1OD—————CTMUD	TADLE 22-0.	REGISTERS ASSOCIATED WITH STNCHKONOUS MASTER RECEPTION												
PIR1PSPIFADIFRC1IFTX1IFSSP1IFTMR1GIFTMR2IFTMR1IIPIE1PSPIEADIERC1IETX1IESSP1IETMR1GIETMR2IFTMR1IIIPR1PSPIPADIPRC1IPTX1IPSSP1IPTMR1GIPTMR2IPTMR1IIPIR3TMR5GIF—RC2IFTX2IFCTMUIFCCP2IFCCP1IFRTCCIIPIR3TMR5GIE—RC2IETX2IFCTMUIFCCP2IFCCP1IFRTCCIIPR3TMR5GIP—RC2IPTX2IPCTMUIPCCP2IPCCP1IPRTCCIIRCSTA1SPENRX9SRENCRENADDENFERROERRRX9DRCREG1EUSART1 Receive RegisterTX2IPTXCKPBRG16—WUEABDENSPBRG11EUSART1 Baud Rate Generator RegisterHigh ByteSPBRG1EUSART1 Baud Rate Generator RegisterRX9DRCREG2EUSART2 Receive RegisterTX2INSYNCSENDBBRGHTRM1TX9DBAUDCON2ABDOVFRCIDLRXDTPTXCKPBRG16—WUEABDENSPBRG12EUSART2 Baud Rate Generator Register High ByteSPBRG16—WUEABDENSPBRG16—WUEABDENSPBRG2EUSART2 Baud Rate Generator Register High ByteSPBRG2EUSART2 Baud Rate Generator Register——————CTMUDPMD0CCP3MDCCP2MDCCP1MDUART2MDUART1MDSSP2MDSSP1MD <t< th=""><th>Name</th><th>Bit 7</th><th>Bit 6</th><th>Bit 5</th><th>Bit 4</th><th>Bit 3</th><th>Bit 2</th><th>Bit 1</th><th>Bit 0</th></t<>	Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0					
PIE1PSPIEADIERC1IETX1IESSP1IETMR1GIETMR2IETMR1IFIPR1PSPIPADIPRC1IPTX1IPSSP1IPTMR1GIPTMR2IPTMR1IFPIR3TMR5GIF—RC2IFTX2IFCTMUIFCCP2IFCCP1IFRTCCIIPIE3TMR5GIE—RC2IETX2IECTMUIFCCP2IFCCP1IFRTCCIIIPR3TMR5GIP—RC2IPTX2IPCTMUIPCCP2IPCCP1IPRTCCIIRCSTA1SPENRX9SRENCRENADDENFERROERRRX9DRCREG1EUSART1 Receive RegisterTXENSYNCSENDBBRGHTRMTTX9DBAUDCON1ABDOVFRCIDLRXDTPTXCKPBRG16—WUEABDENSPBRG1EUSART1 Baud Rate Generator RegisterRCREG2EUSART1 Baud Rate Generator RegisterRX9DRCREG2EUSART2 Receive RegisterTXENSYNCSENDBBRGHTRMTTX9DBAUDCON2ABDOVFRCIDLRXDTPTXCKPBRG16—WUEABDENSPBRG12EUSART2 Baud Rate Generator Register High ByteSPBRG16—WUEABDENSPBRG2EUSART2 Baud Rate Generator RegisterTXCKPBRG16—WUEABDENSPBRG2EUSART2 Baud Rate Generator RegisterGC0C03U2ODU1OD—————CTMUDPMD0CCP3MDCCP2MDCCP1MDUART1MDUART1MDSSP2MD <td>INTCON</td> <td>GIE/GIEH</td> <td>PEIE/GIEL</td> <td>TMR0IE</td> <td>INT0IE</td> <td>RBIE</td> <td>TMR0IF</td> <td>INT0IF</td> <td>RBIF</td>	INTCON	GIE/GIEH	PEIE/GIEL	TMR0IE	INT0IE	RBIE	TMR0IF	INT0IF	RBIF					
IPR1PSPIPADIPRC1IPTX1IPSSP1IPTMR1GIPTMR2IPTMR1IIPIR3TMR5GIF-RC2IFTX2IFCTMUIFCCP2IFCCP1IFRTCCIIPIE3TMR5GIE-RC2IETX2IECTMUIECCP2IECCP1IERTCCIIIPR3TMR5GIP-RC2IPTX2IPCTMUIPCCP2IPCCP1IPRTCCIIRCSTA1SPENRX9SRENCRENADDENFERROERRRX9DRCREG1EUSART1 Receive Register-TX2IPSENDBBRGHTRMTTX9DBAUDCON1ABDOVFRCIDLRXDTPTXCKPBRG16-WUEABDENSPBRG1EUSART1 Baud Rate Generator Register-CRENADDENFERROERRRX9DRCREG2EUSART1 Baud Rate Generator RegisterWUEABDENSPBRG1EUSART2 Receive RegisterTXSTA2CSRCTX9TXENSYNCSENDBBRGHTRMTTX9DBAUDCON2ABDOVFRCIDLRXDTPTXCKPBRG16-WUEABDENSPBRG12EUSART2 Baud Rate Generator Register High ByteSPBRG16-WUEABDENSPBRG2EUSART2 Baud Rate Generator RegisterCTMUDPMD0CCP3MDCCP2MDCCP1MDUART2MDUART1MDSSP2MDSSP1MDADCMID	PIR1	PSPIF	ADIF	RC1IF	TX1IF	SSP1IF	TMR1GIF	TMR2IF	TMR1IF					
PIR3TMR5GIF—RC2IFTX2IFCTMUIFCCP2IFCCP1IFRTCCIIPIE3TMR5GIE—RC2IETX2IECTMUIECCP2IECCP1IERTCCIIIPR3TMR5GIP—RC2IPTX2IPCTMUIPCCP2IPCCP1IPRTCCIIRCSTA1SPENRX9SRENCRENADDENFERROERRRX9DRCREG1EUSART1 Receive RegisterTXENSYNCSENDBBRGHTRMTTX9DBAUDCON1ABDOVFRCIDLRXDTPTXCKPBRG16—WUEABDENSPBRG11EUSART1 Baud Rate Generator RegisterHigh ByteSPBRG1EUSART1 Baud Rate Generator RegisterRCREG2EUSART2 Receive RegisterTXSTA2CSRCTX9TXENSYNCSENDBBRGHTRMTTX9DBAUDCON2ABDOVFRCIDLRXDTPTXCKPBRG16—WUEABDENSPBRG12EUSART2 Receive RegisterTXCKPBRG16—WUEABDENSPBRG12EUSART2 Baud Rate Generator Register High ByteSPBRG16—WUEABDENSPBRG2EUSART2 Baud Rate Generator RegisterSPBRG16—WUEABDENSPBRG2EUSART2 Baud Rate Generator RegisterG————ODCON3U2ODU1OD——————CTMUDPMD0CCP3MDCCP2MDCCP1MDUART2MDUART1MDSSP1MDADCMID	PIE1	PSPIE	ADIE	RC1IE	TX1IE	SSP1IE	TMR1GIE	TMR2IE	TMR1IE					
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IPR3TMR5GIP—RC2IPTX2IPCTMUIPCCP2IPCCP1IPRTCCIIRCSTA1SPENRX9SRENCRENADDENFERROERRRX9DRCREG1EUSART1 Receive RegisterTXSTA1CSRCTX9TXENSYNCSENDBBRGHTRMTTX9DBAUDCON1ABDOVFRCIDLRXDTPTXCKPBRG16—WUEABDENSPBRGH1EUSART1 Baud Rate Generator Register High ByteSPBRG1EUSART1 Baud Rate Generator RegisterRCSTA2SPENRX9SRENCRENADDENFERROERRRX9DRCREG2EUSART2 Receive RegisterTXSTA2CSRCTX9TXENSYNCSENDBBRGHTRMTTX9DBAUDCON2ABDOVFRCIDLRXDTPTXCKPBRG16—WUEABDENSPBRG12EUSART2 Baud Rate Generator Register High ByteSPBRG2EUSART2 Baud Rate Generator Register </td <td>PIR3</td> <td colspan="10">TMR5GIF — RC2IF TX2IF CTMUIF CCP2IF CCP1IF RTCCIF</td>	PIR3	TMR5GIF — RC2IF TX2IF CTMUIF CCP2IF CCP1IF RTCCIF												
RCSTA1SPENRX9SRENCRENADDENFERROERRRX9DRCREG1EUSART1 Receive RegisterTXSTA1CSRCTX9TXENSYNCSENDBBRGHTRMTTX9DBAUDCON1ABDOVFRCIDLRXDTPTXCKPBRG16—WUEABDENSPBRGH1EUSART1 Baud Rate Generator Register High ByteSPBRG1EUSART1 Baud Rate Generator RegisterRCSTA2SPENRX9SRENCRENADDENFERROERRRX9DRCREG2EUSART2 Receive RegisterTXENSYNCSENDBBRGHTRMTTX9DBAUDCON2ABDOVFRCIDLRXDTPTXCKPBRG16—WUEABDENSPBRGH2EUSART2 Baud Rate Generator Register High ByteSPBRG12EUSART2 Baud Rate Generator Register High ByteSPBRG2EUSART2 Baud Rate Generator RegisterODCON3U2ODU1OD—————CTMUDPMD0CCP3MDCCP2MDCCP1MDUART2MDUART1MDSSP2MDSSP1MDADCMI	PIE3	TMR5GIE         —         RC2IE         TX2IE         CTMUIE         CCP2IE         CCP1IE         RTCCIE												
RCREG1EUSART1 Receive RegisterTXSTA1CSRCTX9TXENSYNCSENDBBRGHTRMTTX9DBAUDCON1ABDOVFRCIDLRXDTPTXCKPBRG16—WUEABDENSPBRGH1EUSART1 Baud Rate Generator Register High ByteSPBRG1EUSART1 Baud Rate Generator RegisterRCSTA2SPENRX9SRENCRENADDENFERROERRRX9DRCREG2EUSART2 Receive RegisterTXSTA2CSRCTX9TXENSYNCSENDBBRGHTRMTTX9DBAUDCON2ABDOVFRCIDLRXDTPTXCKPBRG16—WUEABDENSPBRGH2EUSART2 Baud Rate Generator Register High ByteSPBRG2EUSART2 Baud Rate Generator Register </td <td>IPR3</td> <td colspan="13">3 TMR5GIP — RC2IP TX2IP CTMUIP CCP2IP CCP1IP RTCCIF</td>	IPR3	3 TMR5GIP — RC2IP TX2IP CTMUIP CCP2IP CCP1IP RTCCIF												
TXSTA1CSRCTX9TXENSYNCSENDBBRGHTRMTTX9DBAUDCON1ABDOVFRCIDLRXDTPTXCKPBRG16—WUEABDENSPBRG11EUSART1 Baud Rate Generator Register High ByteSPBRG1EUSART1 Baud Rate Generator RegisterRCSTA2SPENRX9SRENCRENADDENFERROERRRX9DRCREG2EUSART2 Receive RegisterTXSTA2CSRCTX9TXENSYNCSENDBBRGHTRMTTX9DBAUDCON2ABDOVFRCIDLRXDTPTXCKPBRG16—WUEABDENSPBRG12EUSART2 Baud Rate Generator Register High ByteSPBRG2EUSART2 Baud Rate Generator Register High ByteSPBRG2EUSART2 Baud Rate Generator RegisterODCON3U2ODU1OD————CTMUDPMD0CCP3MDCCP2MDCCP1MDUART1MDSSP2MDSSP1MDADCMI	RCSTA1	SPEN	RX9	SREN	CREN	ADDEN	FERR	OERR	RX9D					
BAUDCON1ABDOVFRCIDLRXDTPTXCKPBRG16—WUEABDENSPBRGH1EUSART1 Baud Rate Generator Register High ByteSPBRG1EUSART1 Baud Rate Generator RegisterRCSTA2SPENRX9SRENCRENADDENFERROERRRX9DRCREG2EUSART2 Receive RegisterTXSTA2CSRCTX9TXENSYNCSENDBBRGHTRMTTX9DBAUDCON2ABDOVFRCIDLRXDTPTXCKPBRG16—WUEABDENSPBRG12EUSART2 Baud Rate Generator Register High ByteSPBRG2EUSART2 Baud Rate Generator RegisterODCON3U2ODU1OD————CTMUDPMD0CCP3MDCCP1MDUART2MDUART1MDSSP2MDSSP1MDADCMI	RCREG1	EUSART1 R	eceive Regist	er										
SPBRGH1EUSART1 Baud Rate Generator Register High ByteSPBRG1EUSART1 Baud Rate Generator RegisterRCSTA2SPENRX9SRENCRENADDENFERROERRRX9DRCREG2EUSART2 Receive RegisterTXSTA2CSRCTX9TXENSYNCSENDBBRGHTRMTTX9DBAUDCON2ABDOVFRCIDLRXDTPTXCKPBRG16—WUEABDENSPBRG12EUSART2 Baud Rate Generator Register High ByteSPBRG2EUSART2 Baud Rate Generator RegisterODCON3U2ODU1OD————CTMUDPMD0CCP3MDCCP2MDCCP1MDUART2MDUART1MDSSP2MDSSP1MDADCMI	TXSTA1	CSRC	TX9	TXEN	SYNC	SENDB	BRGH	TRMT	TX9D					
SPBRG1       EUSART1 Baud Rate Generator Register         RCSTA2       SPEN       RX9       SREN       CREN       ADDEN       FERR       OERR       RX9D         RCREG2       EUSART2 Receive Register       TXEN       SYNC       SENDB       BRGH       TRMT       TX9D         BAUDCON2       ABDOVF       RCIDL       RXDTP       TXCKP       BRG16       —       WUE       ABDEN         SPBRGH2       EUSART2 Baud Rate Generator Register High Byte       SPBRG2       EUSART2 Baud Rate Generator Register         ODCON3       U2OD       U1OD       —       —       —       —       CTMUD         PMD0       CCP3MD       CCP2MD       CCP1MD       UART2MD       UART1MD       SSP2MD       SSP1MD       ADCMI	BAUDCON1	ABDOVF	RCIDL	RXDTP	TXCKP	BRG16	_	WUE	ABDEN					
RCSTA2SPENRX9SRENCRENADDENFERROERRRX9DRCREG2EUSART2 Receive RegisterTXSTA2CSRCTX9TXENSYNCSENDBBRGHTRMTTX9DBAUDCON2ABDOVFRCIDLRXDTPTXCKPBRG16—WUEABDENSPBRGH2EUSART2 Baud Rate Generator Register High ByteSPBRG2EUSART2 Baud Rate Generator RegisterODCON3U2ODU1OD————CTMUDPMD0CCP3MDCCP2MDCCP1MDUART2MDUART1MDSSP2MDSSP1MDADCMI	SPBRGH1	EUSART1 Ba	aud Rate Gen	erator Regi	ster High Byt	te								
RCREG2       EUSART2 Receive Register         TXSTA2       CSRC       TX9       TXEN       SYNC       SENDB       BRGH       TRMT       TX9D         BAUDCON2       ABDOVF       RCIDL       RXDTP       TXCKP       BRG16       —       WUE       ABDEN         SPBRGH2       EUSART2 Baud Rate Generator Register High Byte	SPBRG1	EUSART1 Ba	aud Rate Gen	erator Regi	ster									
TXSTA2       CSRC       TX9       TXEN       SYNC       SENDB       BRGH       TRMT       TX9D         BAUDCON2       ABDOVF       RCIDL       RXDTP       TXCKP       BRG16       —       WUE       ABDEN         SPBRGH2       EUSART2 Baud Rate Generator Register High Byte	RCSTA2	SPEN	RX9	SREN	CREN	ADDEN	FERR	OERR	RX9D					
BAUDCON2ABDOVFRCIDLRXDTPTXCKPBRG16—WUEABDENSPBRG12EUSART2 Baud Rate Generator Register High ByteSPBRG2EUSART2 Baud Rate Generator RegisterODCON3U2ODU1OD————CTMUDPMD0CCP3MDCCP2MDCCP1MDUART2MDUART1MDSSP2MDSSP1MDADCMI	RCREG2	EUSART2 R	eceive Regist	er										
SPBRGH2       EUSART2 Baud Rate Generator Register High Byte         SPBRG2       EUSART2 Baud Rate Generator Register         ODCON3       U2OD       U1OD       —       —       —       —       CTMUD         PMD0       CCP3MD       CCP2MD       CCP1MD       UART2MD       VART1MD       SSP2MD       SSP1MD       ADCMID	TXSTA2	CSRC	TX9	TXEN	SYNC	SENDB	BRGH	TRMT	TX9D					
SPBRG2       EUSART2 Baud Rate Generator Register         ODCON3       U2OD       U1OD       —       —       —       —       CTMUD         PMD0       CCP3MD       CCP2MD       CCP1MD       UART2MD       UART1MD       SSP2MD       SSP1MD       ADCMID	BAUDCON2	ABDOVF	RCIDL	RXDTP	TXCKP	BRG16	—	WUE	ABDEN					
ODCON3U2ODU1ODCTMUDPMD0CCP3MDCCP2MDCCP1MDUART2MDUART1MDSSP2MDSSP1MDADCMD	SPBRGH2	EUSART2 Baud Rate Generator Register High Byte												
PMD0 CCP3MD CCP2MD CCP1MD UART2MD UART1MD SSP2MD SSP1MD ADCMI	SPBRG2	EUSART2 Baud Rate Generator Register												
	ODCON3	U2OD	U10D	_	—	—	—	_	CTMUDS					
<b>Legend:</b> — = unimplemented, read as '0'. Shaded cells are not used for synchronous master reception.	PMD0	CCP3MD	CCP2MD	CCP1MD	UART2MD	UART1MD	SSP2MD	SSP1MD	ADCMD					
	Legend: -=	unimplement	ed, read as '0	'. Shaded c	ells are not u	ised for sync	hronous mas	ter reception						

## TABLE 22-8: REGISTERS ASSOCIATED WITH SYNCHRONOUS MASTER RECEPTION

## 22.4 EUSART Synchronous Slave Mode

Synchronous Slave mode is entered by clearing bit, CSRC (TXSTAx<7>). This mode differs from the Synchronous Master mode in that the shift clock is supplied externally at the CKx pin (instead of being supplied internally in Master mode). This allows the device to transfer or receive data while in any low-power mode.

#### 22.4.1 EUSART SYNCHRONOUS SLAVE TRANSMISSION

The operation of the Synchronous Master and Slave modes is identical, except in the case of Sleep mode.

If two words are written to the TXREGx and then the SLEEP instruction is executed, the following will occur:

- a) The first word will immediately transfer to the TSR register and transmit.
- b) The second word will remain in the TXREGx register.
- c) Flag bit, TXxIF, will not be set.
- d) When the first word has been shifted out of TSR, the TXREGx register will transfer the second word to the TSR and flag bit, TXxIF, will now be set.

e) If enable bit, TXxIE, is set, the interrupt will wake the chip from Sleep. If the global interrupt is enabled, the program will branch to the interrupt vector.

To set up a Synchronous Slave Transmission:

- 1. Enable the synchronous slave serial port by setting bits, SYNC and SPEN, and clearing bit, CSRC.
- 2. Clear bits, CREN and SREN.
- 3. If interrupts are desired, set enable bit, TXxIE.
- 4. If 9-bit transmission is desired, set bit, TX9.
- 5. Enable the transmission by setting enable bit, TXEN.
- 6. If 9-bit transmission is selected, the ninth bit should be loaded in bit, TX9D.
- 7. Start transmission by loading data to the TXREGx register.
- If using interrupts, ensure that the GIE and PEIE bits in the INTCON register (INTCON<7:6>) are set.

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
INTCON	GIE/GIEH	PEIE/GIEL	TMR0IE	INT0IE	RBIE	TMR0IF	INT0IF	RBIF
PIR1	PSPIF	ADIF	RC1IF	TX1IF	SSP1IF	TMR1GIF	TMR2IF	TMR1IF
PIE1	PSPIE	ADIE	RC1IE	TX1IE	SSP1IE	TMR1GIE	TMR2IE	TMR1IE
IPR1	PSPIP	ADIP	RC1IP	TX1IP	SSP1IP	TMR1GIP	TMR2IP	TMR1IP
PIR3	TMR5GIF	_	RC2IF	TX2IF	CTMUIF	CCP2IF	CCP1IF	RTCCIF
PIE3	TMR5GIE	_	RC2IE	TX2IE	CTMUIE	CCP2IE	CCP1IE	RTCCIE
IPR3	TMR5GIP	_	RC2IP	TX2IP	CTMUIP	CCP2IP	CCP1IP	RTCCIP
RCSTA1	SPEN	RX9	SREN	CREN	ADDEN	FERR	OERR	RX9D
TXREG1	EUSART1 Transmit Register							
TXSTA1	CSRC	TX9	TXEN	SYNC	SENDB	BRGH	TRMT	TX9D
BAUDCON1	ABDOVF	RCIDL	RXDTP	TXCKP	BRG16		WUE	ABDEN
SPBRGH1	EUSART1 B	aud Rate Gei	nerator Regi	ster High Byt	е			
SPBRG1	EUSART1 B	aud Rate Ger	nerator Regi	ster				
RCSTA2	SPEN	RX9	SREN	CREN	ADDEN	FERR	OERR	RX9D
TXREG2	EUSART2 T	ransmit Regis	ster					
TXSTA2	CSRC	TX9	TXEN	SYNC	SENDB	BRGH	TRMT	TX9D
BAUDCON2	ABDOVF	RCIDL	RXDTP	TXCKP	BRG16		WUE	ABDEN
SPBRGH2	EUSART2 B	aud Rate Ger	nerator Regi	ster High Byt	е			
SPBRG2	EUSART2 B	aud Rate Gei	nerator Regi	ster				
ODCON3	U2OD	U10D	—		_		_	CTMUDS
PMD0	CCP3MD	CCP2MD	CCP1MD	UART2MD	UART1MD	SSP2MD	SSP1MD	ADCMD

## TABLE 22-9: REGISTERS ASSOCIATED WITH SYNCHRONOUS SLAVE TRANSMISSION

Legend: — = unimplemented, read as '0'. Shaded cells are not used for synchronous slave transmission.

#### 22.4.2 EUSART SYNCHRONOUS SLAVE RECEPTION

The operation of the Synchronous Master and Slave modes is identical except in the case of Sleep, or any Idle mode, and bit, SREN, which is a "don't care" in Slave mode.

If receive is enabled by setting the CREN bit prior to entering Sleep or any Idle mode, then a word may be received while in this low-power mode. Once the word is received, the RSR register will transfer the data to the RCREGx register. If the RCxIE enable bit is set, the interrupt generated will wake the chip from the low-power mode. If the global interrupt is enabled, the program will branch to the interrupt vector. To set up a Synchronous Slave Reception:

- 1. Enable the synchronous master serial port by setting bits, SYNC and SPEN, and clearing bit, CSRC.
- 2. If interrupts are desired, set enable bit, RCxIE.
- 3. If 9-bit reception is desired, set bit, RX9.
- 4. To enable reception, set enable bit, CREN.
- 5. Flag bit, RCxIF, will be set when reception is complete. An interrupt will be generated if enable bit, RCxIE, was set.
- 6. Read the RCSTAx register to get the 9th bit (if enabled) and determine if any error occurred during reception.
- 7. Read the 8-bit received data by reading the RCREGx register.
- 8. If any error occurred, clear the error by clearing bit, CREN.
- If using interrupts, ensure that the GIE and PEIE bits in the INTCON register (INTCON<7:6>) are set.

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
INTCON	GIE/GIEH	PEIE/GIEL	TMR0IE	INT0IE	RBIE	TMR0IF	<b>INT0IF</b>	RBIF
PIR1	PSPIF	ADIF	RC1IF	TX1IF	SSP1IF	TMR1GIF	TMR2IF	TMR1IF
PIE1	PSPIE	ADIE	RC1IE	TX1IE	SSP1IE	TMR1GIE	TMR2IE	TMR1IE
IPR1	PSPIP	ADIP	RC1IP	TX1IP	SSP1IP	TMR1GIP	TMR2IP	TMR1IP
PIR3	TMR5GIF	_	RC2IF	TX2IF	CTMUIF	CCP2IF	CCP1IF	RTCCIF
PIE3	TMR5GIE	_	RC2IE	TX2IE	CTMUIE	CCP2IE	CCP1IE	RTCCIE
IPR3	TMR5GIP	_	RC2IP	TX2IP	CTMUIP	CCP2IP	CCP1IP	RTCCIP
RCSTA1	SPEN	RX9	SREN	CREN	ADDEN	FERR	OERR	RX9D
RCREG1	EUSART1 Receive Register							
TXSTA1	CSRC	TX9	TXEN	SYNC	SENDB	BRGH	TRMT	TX9D
BAUDCON1	ABDOVF	RCIDL	RXDTP	TXCKP	BRG16	_	WUE	ABDEN
SPBRGH1	EUSART1 E	Baud Rate Ge	nerator Regi	ster High Byt	е			
SPBRG1	EUSART1 E	Baud Rate Ge	nerator Regi	ster				
RCSTA2	SPEN	RX9	SREN	CREN	ADDEN	FERR	OERR	RX9D
RCREG2	EUSART2 F	Receive Regis	ter					
TXSTA2	CSRC	TX9	TXEN	SYNC	SENDB	BRGH	TRMT	TX9D
BAUDCON2	ABDOVF	RCIDL	RXDTP	TXCKP	BRG16	_	WUE	ABDEN
SPBRGH2	EUSART2 E	aud Rate Ge	nerator Regi	ster High Byt	е			
SPBRG2	EUSART2 E	Baud Rate Ge	nerator Regi	ster				
ODCON3	U2OD	U10D		_			_	CTMUDS
PMD0	CCP3MD	CCP2MD	CCP1MD	UART2MD	UART1MD	SSP2MD	SSP1MD	ADCMD

TABLE 22-10: REGISTERS ASSOCIATED WITH SYNCHRONOUS SLAVE RECEPTION

**Legend:** — = unimplemented, read as '0'. Shaded cells are not used for synchronous slave reception.

## 23.0 12-BIT ANALOG-TO-DIGITAL CONVERTER (A/D) MODULE

The Analog-to-Digital (A/D) Converter module in the PIC18F87K22 family of devices has 16 inputs for the 64-pin devices and 24 inputs for the 80-pin devices. This module allows conversion of an analog input signal to a corresponding 12-bit digital number.

The module has these registers:

- A/D Control Register 0 (ADCON0)
- A/D Control Register 1 (ADCON1)
- A/D Control Register 2 (ADCON2)
- A/D Port Configuration Register 0 (ANCON0)
- A/D Port Configuration Register 1 (ANCON1)
- A/D Port Configuration Register 2 (ANCON2)
- ADRESH (the upper, A/D Results register)
- ADRESL (the lower, A/D Results register)

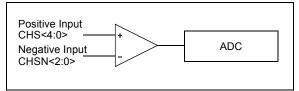
The ADCON0 register, shown in Register 23-1, controls the operation of the A/D module. The ADCON1 register, shown in Register 23-2, configures the voltage reference and special trigger selection. The ADCON2 register, shown in Register 23-3, configures the A/D clock source and programmed acquisition time and justification.

## 23.1 Differential A/D Converter

The converter in PIC18F87K22 family devices is implemented as a differential A/D where the differential voltage between two channels is measured and converted to digital values (see Figure 23-1).

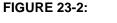
The converter can also be configured to measure a voltage from a single input by clearing the CHSN bits (ADCON1<2:0>). With this configuration, the negative channel input is connected internally to AVss (see Figure 23-2).

## FIGURE 23-1: DIFFERENTIAL CHANNEL MEASUREMENT

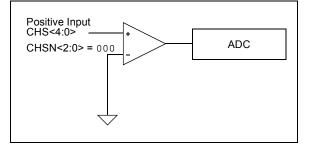


Differential conversion feeds the two input channels to a unity gain differential amplifier. The positive channel input is selected using the CHS bits (ADCON0<6:2>) and the negative channel input is selected using the CHSN bits (ADCON1<2:0>).

The output from the amplifier is fed to the A/D Converter, as shown in Figure 23-1. The 12-bit result is available on the ADRESH and ADRESL registers. An additional bit indicates if the 12-bit result is a positive or negative value.



#### SINGLE CHANNEL MEASUREMENT



In the Single Channel Measurement mode, the negative input is connected to Avss by clearing the CHSN bits (ADCON1<2:0>).

## 23.2 A/D Registers

23.2.1 A/D CONTROL REGISTERS

#### REGISTER 23-1: ADCON0: A/D CONTROL REGISTER 0

U-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0		
_	CHS4	CHS3	CHS2	CHS1	CHS0	GO/DONE	ADON		
bit 7 bit 0									
Legend:	Legend:								
R = Readable bit W = Writable bit			U = Unimplemented bit, read as '0'						
-n = Value at POR (1' = Bit is set			'0' = Bit is cleared x = Bit is unknown						

#### bit 7 Unimplemented: Read as '0'

DIT /	Unimplemented: Read as 10
bit 6-2	CHS<4:0>: Analog Channel Select bits
	00000 = Channel 00 (AN0) 10000 = Channel 16 (AN16)
	00001 = Channel 01 (AN1) 10001 = Channel 17 (AN17)
	00010 = Channel 02 (AN2) 10010 = Channel 18 (AN18)
	00011 = Channel 03 (AN3) 10011 = Channel 19 (AN19)
	00100 = Channel 04 (AN4) 10100 = Channel 20 (AN20) <sup>(1,2)</sup>
	00101 = Channel 05 (AN5) 10101 = Channel 21 (AN21) <sup>(1,2)</sup>
	00110 = Channel 06 (AN6) 10110 = Channel 22 (AN22) <sup>(1,2)</sup>
	00111 = Channel 07 (AN7) 10111 = Channel 23 (AN23) <sup>(1,2)</sup>
	01000 =  Channel 08 (AN8) $11000 = $ (Reserved) <sup>(2)</sup>
	01001 =  Channel 09 (AN9) $11001 = $ (Reserved) <sup>(2)</sup>
	01010 = Channel 10 (AN10 11010 = (Reserved) <sup>(2)</sup>
	01011 = Channel 11 (AN11) 11011 = (Reserved) <sup>(2)</sup>
	01100 = Channel 12 (AN12)(1.2) 11100 = Channel 28 (Reserved CTMU)
	01101 = Channel 13 (AN13)(1,20) 11101 = Channel 29 (Internal temperature diode)
	01110 = Channel 14 (AN14) $(1,2)$ 11110 = Channel 30 (VDDCORE)
	01111 = Channel 15 (AN15) <sup>(1,2)</sup> 11111 = Channel 31 (v1.024V band gap)
bit 1	GO/DONE: A/D Conversion Status bit
	1 = A/D (or calibration) cycle in progress. Setting this bit starts an A/D conversion cycle. The bit is
	cleared automatically by hardware when the A/D conversion is completed.
	0 = A/D conversion has completed or is not in progress
bit 0	ADON: A/D On bit
	1 = A/D Converter is operating
	0 = A/D conversion module is shut off and consuming no operating current
Note 1:	These channels are not implemented on 64-pin devices.
	Deferming a conversion on unimplemented channels will return random values

2: Performing a conversion on unimplemented channels will return random values.

R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0		
TRIGSEL1	TRIGSEL0	VCFG1	VCFG0	VNCFG	CHSN2	CHSN1	CHSN0		
bit 7	L.					·	bit 0		
Legend:									
R = Readabl	e bit	W = Writable	bit	U = Unimplem	nented bit, read	l as '0'			
-n = Value at	POR	'1' = Bit is set		'0' = Bit is clea	ared	x = Bit is unkr	nown		
bit 7-6			raar Calaat hita						
JIL 7-0			ger Select bits						
			ger from the R						
			ger from the Til ger from the C						
			ger from the E						
bit 5-4			- nfiguration bits						
	11 = Internal VREF+ (4.096V)								
		VREF+ (2.048V	)						
	01 = External	VREF+							
	00 <b>= AV</b> DD								
bit 3		/NCFG: A/D VREF- Configuration bit							
	1 = External VREF								
	0 = AVss								
bit 2-0		• •	ve Channel Se	lect bits					
	111 = Channe	、 ,							
	110 = Channe	、 ,							
	101 = Channe								
	100 = Channe 011 = Channe	• •							
	010 = Channe								
	001 = Channe								
	000 = Channe	. ,							

#### REGISTER 23-2: ADCON1: A/D CONTROL REGISTER 1

R/W-0	U-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0
ADFM	_	ACQT2	ACQT1	ACQT0	ADCS2	ADCS1	ADCS0
bit 7							bit
Legend:							
R = Reada		W = Writable		•	nented bit, read		
-n = Value	at POR	'1' = Bit is set		'0' = Bit is cle	ared	x = Bit is unkr	nown
bit 7	<b>ADFM:</b> A/D 1 = Right jus 0 = Left justi		elect bit				
bit 6	-	neu nted: Read as '	n'				
bit 5-3	-	: A/D Acquisition		hits			
	111 = 20  Tar $110 = 16  Tar$ $101 = 12  Tar$ $100 = 8  TaD$ $011 = 6  TaD$ $010 = 4  TaD$ $001 = 2  TaD$ $000 = 0  TaD$	5					
bit 2-0	111 = FRC (0 110 = Fosc/ 101 = Fosc/ 100 = Fosc/	16 4 clock derived fro 32 8	m A/D RC oso	cillator) <sup>(1)</sup>			

## REGISTER 23-3: ADCON2: A/D CONTROL REGISTER 2

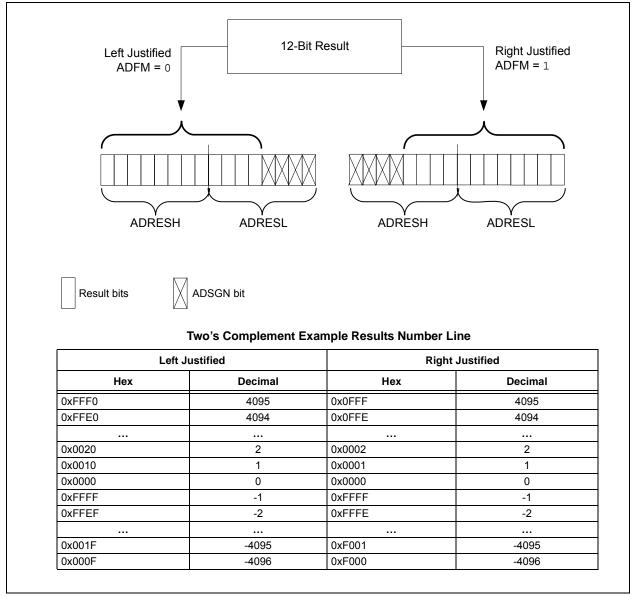
**Note 1:** If the A/D FRC clock source is selected, a delay of one TCY (instruction cycle) is added before the A/D clock starts. This allows the SLEEP instruction to be executed before starting a conversion.

## 23.2.2 A/D RESULT REGISTERS

The ADRESH:ADRESL register pair is where the 12-bit A/D result and extended sign bits (ADSGN) are loaded at the completion of a conversion. This register pair is 16 bits wide. The A/D module gives the flexibility of left or right justifying the 12-bit result in the 16-Bit Result register. The A/D Format Select bit (ADFM) controls this justification. Figure 23-3 shows the operation of the A/D result justification and location of the extended sign bits (ADSGN). The extended sign bits allow for easier 16-bit math to be performed on the result. The results are represented as a two's compliment binary value. This means that when sign bits and magnitude bits are considered together in right justification, the ADRESH and ADRESL registers can be read as a single signed integer value.

When the A/D Converter is disabled, these 8-bit registers can be used as two general purpose registers.

FIGURE 23-3: A/D RESULT JUSTIFICATION



#### **REGISTER 23-4:** ADRESH: A/D RESULT HIGH BYTE REGISTER, LEFT JUSTIFIED (ADFM = 0)

R/W-x	R/W-x	R/W-x	R/W-x	R/W-x	R/W-x	R/W-x	R/W-x
ADRES11	ADRES10	ADRES9	ADRES8	ADRES7	ADRES6	ADRES5	ADRES4
bit 7							bit 0
Legend:							

-09	ona.			
R =	Readable bit	W = Writable bit	U = Unimplemented bit, read	as '0'
-n =	Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown

bit 7-0 ADRES<11:4>: A/D Result High Byte bits

#### **REGISTER 23-5:** ADRESL: A/D RESULT HIGH BYTE REGISTER, LEFT JUSTIFIED (ADFM = 0)

R/W-x	R/W-x	R/W-x	R/W-x	R/W-x	R/W-x	R/W-x	R/W-x
ADRES3	ADRES2	ADRES1	ADRES0	ADSGN	ADSGN	ADSGN	ADSGN
bit 7							bit 0

Legend:			
R = Readable bit	W = Writable bit	U = Unimplemented bit,	read as '0'
-n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown

bit 7-4 ADRES<3:0>: A/D Result Low Byte bits

bit 3-0 ADSGN: A/D Result Sign bit

1 = A/D result is negative

0 = A/D result is positive

## REGISTER 23-6: ADRESH: A/D RESULT HIGH BYTE REGISTER, RIGHT JUSTIFIED (ADFM = 1)

R/W-x	R/W-x	R/W-x	R/W-x	R/W-x	R/W-x	R/W-x	R/W-x
ADSGN	ADSGN	ADSGN	ADSGN	ADRES11	ADRES10	ADRES9	ADRES8
bit 7							bit 0

Legend:			
R = Readable bit	W = Writable bit	U = Unimplemented bit, read	d as '0'
-n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown

bit 7-4	ADSGN: A/D Result Sign bit
	1 = A/D result is negative
	0 = A/D result is positive
bit 3-0	ADRESH<11:8>: A/D Result High Byte bits

## REGISTER 23-7: ADRESL: A/D RESULT LOW BYTE REGISTER, RIGHT JUSTIFIED (ADFM = 1)

| R/W-x  |
|--------|--------|--------|--------|--------|--------|--------|--------|
| ADRES7 | ADRES6 | ADRES5 | ADRES4 | ADRES3 | ADRES2 | ADRES1 | ADRES0 |
| bit 7  |        |        |        |        |        |        | bit 0  |

Legend:			
R = Readable bit	W = Writable bit	U = Unimplemented bit, read	as '0'
-n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown

bit 7-0 ADRES<7:0>: A/D Result Low Byte bits

The ANCONx registers are used to configure the operation of the I/O pin associated with each analog channel. Clearing an ANSELx bit configures the corresponding pin (ANx) to operate as a digital only I/O. Setting a bit configures the pin to operate as an analog input for either the A/D Converter or the comparator module, with all digital peripherals disabled and digital inputs read as '0'.

As a rule, I/O pins that are multiplexed with analog inputs default to analog operation on any device Reset.

## REGISTER 23-8: ANCON0: A/D PORT CONFIGURATION REGISTER 0

| R/W-1  |
|--------|--------|--------|--------|--------|--------|--------|--------|
| ANSEL7 | ANSEL6 | ANSEL5 | ANSEL4 | ANSEL3 | ANSEL2 | ANSEL1 | ANSEL0 |
| bit 7  |        |        |        |        |        |        | bit 0  |

Legend:			
R = Readable bit	W = Writable bit	U = Unimplemented bit, read	1 as '0'
-n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown

bit 7-0

#### **ANSEL<7:0>:** Analog Port Configuration bits (AN7 and AN0)

1 = Pin is configured as an analog channel; digital input is disabled and any inputs read as '0'

0 = Pin is configured as a digital port

#### REGISTER 23-9: ANCON1: A/D PORT CONFIGURATION REGISTER 1

R/W-1	R/W-1	R/W-1	R/W-1	R/W-1	R/W-1	R/W-1	R/W-1
ANSEL15 <sup>(1)</sup>	ANSEL14 <sup>(1)</sup>	ANSEL13 <sup>(1)</sup>	ANSEL12 <sup>(1)</sup>	ANSEL11	ANSEL10	ANSEL9	ANSEL8
bit 7							bit 0

Legend:			
R = Readable bit	W = Writable bit	U = Unimplemented bit	, read as '0'
-n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown

bit 7-0 ANSEL<15:8>: Analog Port Configuration bits (AN15 through AN8)<sup>(1)</sup>

1 = Pin is configured as an analog channel; digital input is disabled and any inputs read as '0'
 0 = Pin is configured as a digital port

**Note 1:** AN15 through AN12 and AN23 to AN20 are implemented only on 80-pin devices. For 64-pin devices, the corresponding ANSELx bits are still implemented for these channels, but have no effect.

#### REGISTER 23-10: ANCON2: A/D PORT CONFIGURATION REGISTER 2

R/W-1	R/W-1	R/W-1	R/W-1	R/W-1	R/W-1	R/W-1	R/W-1
ANSEL23 <sup>(1)</sup>	ANSEL22 <sup>(1)</sup>	ANSEL21 <sup>(1)</sup>	ANSEL20 <sup>(1)</sup>	ANSEL19	ANSEL18	ANSEL17	ANSEL16
bit 7							bit 0
Leaend:							

Legenu.				
R = Readable bit	W = Writable bit	t U = Unimplemented bit, read as '0'		
-n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown	

#### bit 7-0 ANSEL<23:16>: Analog Port Configuration bits (AN23 through AN16)<sup>(1)</sup>

- 1 = Pin is configured as an analog channel; digital input is disabled and any inputs read as '0'
   0 = Pin is configured as a digital port
- **Note 1:** AN15 through AN12 and AN23 through AN20 are implemented only on 80-pin devices. For 64-pin devices, the corresponding ANSELx bits are still implemented for these channels, but have no effect.

The analog reference voltage is software-selectable to either the device's positive and negative supply voltage (AVDD and AVSS) or the voltage level on the RA3/AN3/VREF+ and RA2/AN2/VREF- pins. VREF+ has two additional Internal Reference Voltage selections: 2.048V and 4.096V.

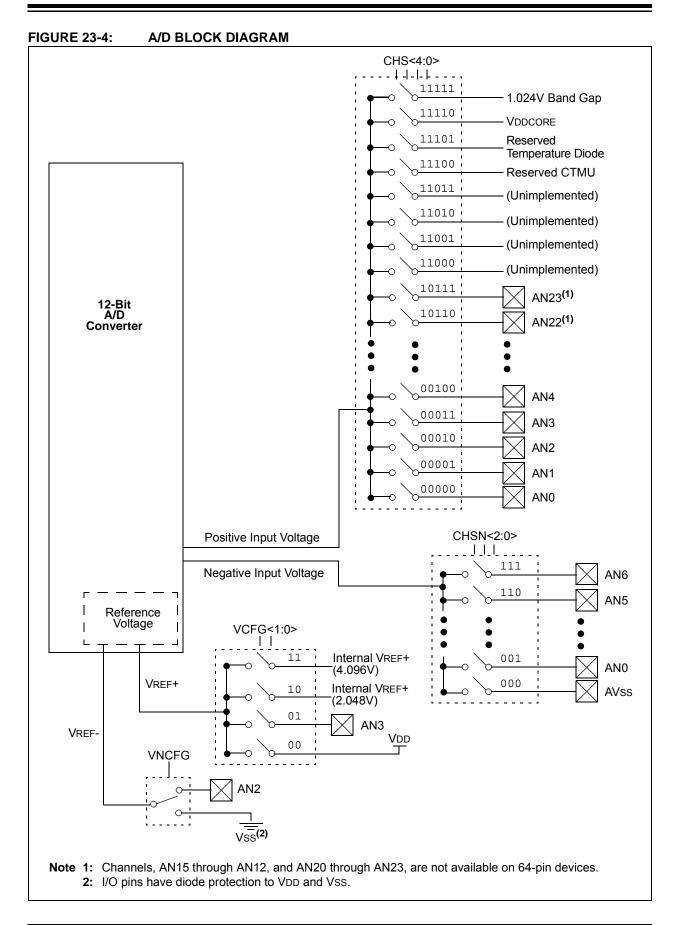
The A/D Converter can uniquely operate while the device is in Sleep mode. To operate in Sleep, the A/D conversion clock must be derived from the A/D Converter's internal RC oscillator.

The output of the Sample-and-Hold (S/H) is the input into the converter, which generates the result via successive approximation.

Each port pin associated with the A/D Converter can be configured as an analog input or a digital I/O. The ADRESH and ADRESL registers contain the result of the A/D conversion. When the A/D conversion is complete, the result is loaded into the ADRESH:ADRESL register pair, the GO/DONE bit (ADCON0<1>) is cleared and the A/D Interrupt Flag bit, ADIF (PIR1<6>), is set.

A device Reset forces all registers to their Reset state. This forces the A/D module to be turned off and any conversion in progress is aborted. The value in the ADRESH:ADRESL register pair is not modified for a Power-on Reset. These registers will contain unknown data after a Power-on Reset.

The block diagram of the A/D module is shown in Figure 23-4.



After the A/D module has been configured as desired, the selected channel must be acquired before the conversion can start. The analog input channels must have their corresponding TRIS bits selected as inputs. To determine acquisition time, see Section 23.3 "A/D Acquisition Requirements". After this acquisition time has elapsed, the A/D conversion can be started. An acquisition time can be programmed to occur between setting the GO/DONE bit and the actual start of the conversion.

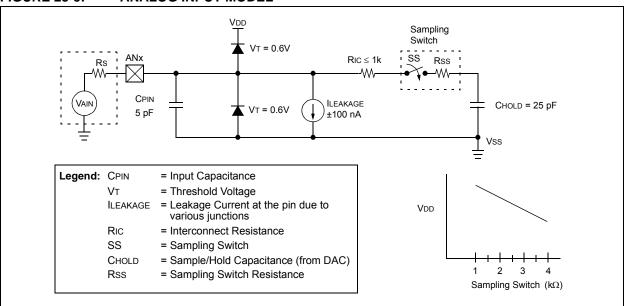
To do an A/D conversion, follow these steps:

- 1. Configure the A/D module:
  - Configure the required ADC pins as analog pins (ANCON0, ANCON1 and ANCON2)
  - Set the voltage reference (ADCON1)
  - Select the A/D positive and negative input channels (ADCON0 and ADCON1)
  - Select the A/D acquisition time (ADCON2)
  - Select the A/D conversion clock (ADCON2)
  - Turn on the A/D module (ADCON0)



- 2. Configure the A/D interrupt (if desired):
  - Clear the ADIF bit (PIR1<6>)
  - Set the ADIE bit (PIE1<6>)
  - Set the GIE bit (INTCON<7>)
- 3. Wait the required acquisition time (if required).
- 4. Start the conversion:
  - Set the GO/DONE bit (ADCON0<1>)
- 5. Wait for A/D conversion to complete, by either:
  Polling for the GO/DONE bit to be cleared OR
  - Waiting for the A/D interrupt
- 6. Read A/D Result registers (ADRESH:ADRESL) and, if required, clear bit, ADIF.
- 7. For the next conversion, begin with Step 1 or 2, as required.

The A/D conversion time per bit is defined as TAD. Before the next acquisition starts, a minimum Wait of 2 TAD is required.



## 23.3 A/D Acquisition Requirements

For the A/D Converter to meet its specified accuracy, the Charge Holding (CHOLD) capacitor must be allowed to fully charge to the input channel voltage level. The analog input model is shown in Figure 23-5. The source impedance (Rs) and the internal sampling switch (Rss) impedance directly affect the time required to charge the capacitor CHOLD. The sampling switch (Rss) impedance varies over the device voltage (VDD).

The source impedance affects the offset voltage at the analog input (due to pin leakage current). The maximum recommended impedance for analog sources is 2.5 k $\Omega$ . After the analog input channel is selected or changed, the channel must be sampled for at least the minimum acquisition time before starting a conversion.

**Note:** When the conversion is started, the holding capacitor is disconnected from the input pin.

To calculate the minimum acquisition time, Equation 23-1 can be used. This equation assumes that 1/2 LSb error is used (1,024 steps for the A/D). The 1/2 LSb error is the maximum error allowed for the A/D to meet its specified resolution.

Equation 23-3 shows the calculation of the minimum required acquisition time, TACQ. This calculation is based on the following application system assumptions:

• CHOLD =	25 pF
• Rs =	2.5 kΩ
• Conversion Error $\leq$	1/2 LSb
• VDD =	$3V \rightarrow Rss = 2 \ k\Omega$
Temperature =	85°C

#### EQUATION 23-1: ACQUISITION TIME

TACQ	=	Amplifier Settling Time + Holding Capacitor Charging Time + Temperature Coefficient
	=	TAMP + TC + TCOFF

#### EQUATION 23-2: A/D MINIMUM CHARGING TIME

VHOLD	=	$(\text{VREF} - (\text{VREF}/2048)) \bullet (1 - e^{(-\text{TC/CHOLD}(\text{Ric} + \text{Rss} + \text{Rs}))})$
or		
TC	=	-(CHOLD)(RIC + RSS + RS) $\ln(1/2048)$

#### EQUATION 23-3: CALCULATING THE MINIMUM REQUIRED ACQUISITION TIME

TACQ	=	TAMP + TC + TCOFF
TAMP	=	0.2 μs
TCOFF	=	(Temp – 25°C)(0.02 μs/°C) (85°C – 25°C)(0.02 μs/°C) 1.2 μs
Tempera	ture c	oefficient is only required for temperatures $> 25^{\circ}$ C. Below 25°C, TCOFF = 0 ms.
Тс	=	-(Chold)(Ric + Rss + Rs) $\ln(1/2048) \ \mu s$ -(25 pF) (1 k $\Omega$ + 2 k $\Omega$ + 2.5 k $\Omega$ ) ln(0.0004883) $\mu s$ 1.05 $\mu s$
TACQ	=	0.2 μs + 1.05 μs + 1.2 μs 2.45 μs

### 23.4 Selecting and Configuring Automatic Acquisition Time

The ADCON2 register allows the user to select an acquisition time that occurs each time the  $GO/\overline{DONE}$  bit is set.

When the GO/DONE bit is set, sampling is stopped and a conversion begins. The user is responsible for ensuring the required acquisition time has passed between selecting the desired input channel and setting the GO/DONE bit.

This occurs when the ACQT<2:0> bits (ADCON2<5:3>) remain in their Reset state ('000'), which is compatible with devices that do not offer programmable acquisition times.

If desired, the ACQTx bits can be set to select a programmable acquisition time for the A/D module. When the GO/DONE bit is set, the A/D module continues to sample the input for the selected acquisition time, then automatically begins a conversion. Since the acquisition time is programmed, there may be no need to wait for an acquisition time between selecting a channel and setting the GO/DONE bit.

In either case, when the conversion is completed, the GO/DONE bit is cleared, the ADIF flag is set and the A/D begins sampling the currently selected channel again. If an acquisition time is programmed, there is nothing to indicate if the acquisition time has ended or if the conversion has begun.

## 23.5 Selecting the A/D Conversion Clock

The A/D conversion time per bit is defined as TAD. The A/D conversion requires 14 TAD per 12-bit conversion. The source of the A/D conversion clock is software-selectable.

The possible options for TAD are:

- 2 Tosc
- 4 Tosc
- 8 Tosc
- 16 Tosc
- 32 Tosc
- 64 Tosc
- Using the internal RC Oscillator

For correct A/D conversions, the A/D conversion clock (TAD) must be as short as possible, but greater than the minimum TAD. (For more information, see Parameter 130 in Table 31-28.)

Table 23-1 shows the resultant TAD times derived from the device operating frequencies and the A/D clock source selected.

## TABLE 23-1: TAD vs. DEVICE OPERATING FREQUENCIES

AD Clock S	Maximum	
Operation	ADCS<2:0>	Device Frequency
2 Tosc	000	2.50 MHz
4 Tosc	100	5.00 MHz
8 Tosc	001	10.00 MHz
16 Tosc	101	20.00 MHz
32 Tosc	010	40.00 MHz
64 Tosc	110	64.00 MHz
RC <sup>(2)</sup>	x11	1.00 MHz <sup>(1)</sup>

Note 1: The RC source has a typical TAD time of  $4 \ \mu s$ .

2: For device frequencies above 1 MHz, the device must be in Sleep mode for the entire conversion or the A/D accuracy may be out of specification.

## 23.6 Configuring Analog Port Pins

The ANCON0, ANCON1, ANCON2, TRISA, TRISF, TRISG and TRISH registers control the operation of the A/D port pins. The port pins needed as analog inputs must have their corresponding TRISx bits set (input). If the TRISx bit is cleared (output), the digital output level (VOH or VOL) will be converted.

The A/D operation is independent of the state of the CHS<3:0> bits and the TRISx bits.

- Note 1: When reading the PORT register, all pins configured as analog input channels will read as cleared (a low level). Pins configured as digital inputs will convert an analog input. Analog levels on a digitally configured input will be accurately converted.
  - Analog levels on any pin defined as a digital input may cause the digital input buffer to consume current out of the device's specification limits.

### 23.7 A/D Conversions

Figure 23-6 shows the operation of the A/D Converter after the GO/DONE bit has been set and the ACQT<2:0> bits are cleared. A conversion is started after the following instruction to allow entry into Sleep mode before the conversion begins.

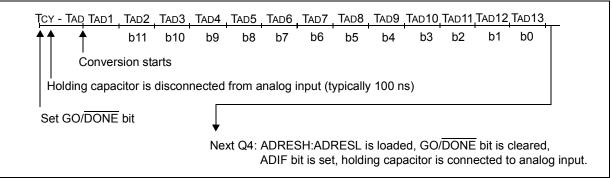
Figure 23-7 shows the operation of the A/D Converter after the GO/DONE bit has been set, the ACQT<2:0> bits set to '010' and a 4 TAD acquisition time selected.

Clearing the  $GO/\overline{DONE}$  bit during a conversion will abort the current conversion. The A/D Result register pair will NOT be updated with the partially completed A/D conversion sample. This means the ADRESH:ADRESL registers will continue to contain the value of the last completed conversion (or the last value written to the ADRESH:ADRESL registers).

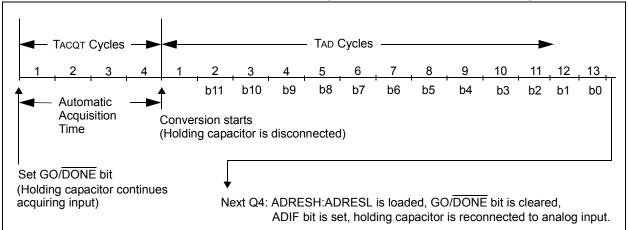
After the A/D conversion is completed or aborted, a 2 TAD Wait is required before the next acquisition can be started. After this Wait, acquisition on the selected channel is automatically started.

**Note:** The GO/DONE bit should **NOT** be set in the same instruction that turns on the A/D.

## FIGURE 23-6: A/D CONVERSION TAD CYCLES (ACQT<2:0> = 000, TACQ = 0)



#### FIGURE 23-7: A/D CONVERSION TAD CYCLES (ACQT<2:0> = 010, TACQ = 4 TAD)



## 23.8 Use of the Special Event Triggers

A/D conversion can be started by the Special Event Trigger of any of these modules:

- ECCP2 Requires CCP2M<3:0> bits (CCP2CON<3:0>) set at '1011'
- CTMU Requires the setting of the CTTRIG bit (CTMUCONH<0>)
- Timer1
- RTCC

To start an A/D conversion:

- The A/D module must be enabled (ADON = 1)
- The appropriate analog input channel is selected
- The minimum acquisition period is set one of these ways:
  - Timing provided by the user
  - Selection made of an appropriate TACQ time

With these conditions met, the trigger sets the GO/DONE bit and the A/D acquisition starts.

If the A/D module is not enabled (ADON = 0), the module ignores the Special Event Trigger.

Note: With an ECCP2 trigger, Timer1 or Timer 3 is cleared. The timers reset to automatically repeat the A/D acquisition period with minimal software overhead (moving ADRESH:ADRESL to the desired location). If the A/D module is not enabled, the Special Event Trigger is ignored by the module, but the timer's counter resets.

### 23.9 Operation in Power-Managed Modes

The selection of the automatic acquisition time and A/D conversion clock is determined, in part, by the clock source and frequency while in a power-managed mode.

If the A/D is expected to operate while the device is in a power-managed mode, the ACQT<2:0> and ADCS<2:0> bits in ADCON2 should be updated in accordance with the power-managed mode clock that will be used.

After the power-managed mode is entered (either of the power-managed Run modes), an A/D acquisition or conversion may be started. Once an acquisition or conversion is started, the device should continue to be clocked by the same power-managed mode clock source until the conversion has been completed. If desired, the device may be placed into the corresponding power-managed Idle mode during the conversion.

If the power-managed mode clock frequency is less than 1 MHz, the A/D RC clock source should be selected.

Operation in Sleep mode requires that the A/D RC clock be selected. If bits, ACQT<2:0>, are set to '000' and a conversion is started, the conversion will be delayed one instruction cycle to allow execution of the SLEEP instruction and entry into Sleep mode. The IDLEN and SCS<1:0> bits in the OSCCON register must have already been cleared prior to starting the conversion.

TADLE 23-2.	REGISTERS ASSOCIATED WITH A/D MODULE							
Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
INTCON	GIE/GIEH	PEIE/GIEL	TMR0IE	INT0IE	RBIE	TMR0IF	INT0IF	RBIF
PIR1	PSPIF	ADIF	RC1IF	TX1IF	SSP1IF	TMR1GIF	TMR2IF	TMR1IF
PIE1	PSPIE	ADIE	RC1IE	TX1IE	SSP1IE	TMR1GIE	TMR2IE	TMR1IE
IPR1	PSPIP	ADIP	RC1IP	TX1IP	SSP1IP	TMR1GIP	TMR2IP	TMR1IP
PIR3	TMR5GIF	_	RC2IF	TX2IF	CTMUIF	CCP2IF	CCP1IF	RTCCIF
PIE3	TMR5GIE	_	RC2IE	TX2IE	CTMUIE	CCP2IE	CCP1IE	RTCCIE
IPR3	TMR5GIP	_	RC2IP	TX2IP	CTMUIP	CCP2IP	CCP1IP	RTCCIP
ADRESH	A/D Result	Register Higl	n Byte					
ADRESL	A/D Result	Register Low	<sup>,</sup> Byte					
ADCON0	_	CHS4	CHS3	CHS2	CHS1	CHS0	GO/DONE	ADON
ADCON1	TRIGSEL1	TRIGSEL0	VCFG1	VCFG0	VNCFG	CHSN2	CHSN1	CHSN0
ADCON2	ADFM	_	ACQT2	ACQT1	ACQT0	ADCS2	ADCS1	ADCS0
ANCON0	ANSEL7	ANSEL6	ANSEL5	ANSEL4	ANSEL3	ANSEL2	ANSEL1	ANSEL0
ANCON1	ANSEL15	ANSEL14	ANSEL13	ANSEL12	ANSEL11	ANSEL10	ANSEL9	ANSEL8
ANCON2	ANSEL23	ANSEL22	ANSEL21	ANSEL20	ANSEL19	ANSEL18	ANSEL17	ANSEL16
CCP2CON	P2M1	P2M0	DC2B1	DC2B0	CCP2M3	CCP2M2	CCP2M1	CCP2M0
PORTA	RA7 <sup>(2)</sup>	RA6 <sup>(2)</sup>	RA5	RA4	RA3	RA2	RA1	RA0
TRISA	TRISA7 <sup>(2)</sup>	TRISA6 <sup>(2)</sup>	TRISA5	TRISA4	TRISA3	TRISA2	TRISA1	TRISA0
PORTF	RF7	RF6	RF5	RF4	RF3	RF2	RF1	—
TRISF	TRISF7	TRISF6	TRISF5	TRISF4	TRISF3	TRISF2	TRISF1	—
PORTG	_	_	RG5 <sup>(3)</sup>	RG4	RG3	RG2	RG1	RG0
TRISG				TRISG4	TRISG3	TRISG2	TRISG1	TRISG0
PORTH <sup>(1)</sup>	RH7	RH6	RH5	RH4	RH3	RH2	RH1	RH0
TRISH <sup>(1)</sup>	TRISH7	TRISH6	TRISH5	TRISH4	TRISH3	TRISH2	TRISH1	TRISH0
PMD0	CCP3MD	CCP2MD	CCP1MD	UART2MD	UART1MD	SSP2MD	SSP1MD	ADCMD
		wheel weed on						

#### TABLE 23-2: REGISTERS ASSOCIATED WITH A/D MODULE

**Legend:** — = unimplemented, read as '0'. Shaded cells are not used for A/D conversion.

Note 1: This register is not implemented on 64-pin devices.

**2:** These bits are available only in certain oscillator modes, when the FOSC2 Configuration bit = 0. If that Configuration bit is cleared, this signal is not implemented.

**3:** This bit is available when Master Clear is disabled (MCLRE = 0). When MCLRE is set, the bit is unimplemented.

## 24.0 COMPARATOR MODULE

The analog comparator module contains three comparators that can be independently configured in a variety of ways. The inputs can be selected from the analog inputs and two Internal Reference Voltages. The digital outputs are available at the pin level and can also be read through the control register. Multiple output and interrupt event generation are also available. A generic single comparator from the module is shown in Figure 24-1.

Key features of the module includes:

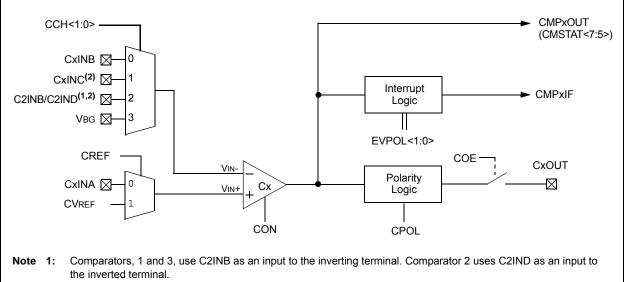
- · Independent comparator control
- · Programmable input configuration
- · Output to both pin and register levels
- · Programmable output polarity
- Independent interrupt generation for each comparator with configurable interrupt-on-change

#### 24.1 Registers

The CMxCON registers (CM1CON, CM2CON and CM3CON) select the input and output configuration for each comparator, as well as the settings for interrupt generation (see Register 24-1).

The CMSTAT register (Register 24-2) provides the output results of the comparators. The bits in this register are read-only.

#### FIGURE 24-1: COMPARATOR SIMPLIFIED BLOCK DIAGRAM



2: C1INC, C2INC and C2IND are all unavailable for 64-pin devices (PIC18F6XK22).

R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0			
CON	COE	CPOL	EVPOL1	EVPOL0	CREF	CCH1	CCH0			
bit 7						•	bit			
Legend:										
R = Readat	ole hit	W = Writable	hit	U = Unimplem	onted hit rea	nd as (0,				
-n = Value a		'1' = Bit is set		'0' = Bit is clea		x = Bit is unkr	own			
bit 7	CON: Compa	rator Enable b	it							
	1 = Comparat	tor is enabled								
	0 = Comparat	tor is disabled								
bit 6	COE: Compa	rator Output E	nable bit							
		1 = Comparator output is present on the CxOUT pin								
	0 = Comparat	tor output is inf	ernal only							
bit 5		CPOL: Comparator Output Polarity Select bit								
	1 = Comparator output is inverted									
	•	tor output is no								
bit 4-3		•	arity Select bits							
			any change of		المربوط المراجع					
				ow transition of t gh transition of t						
		t generation is			ne output					
bit 2	CREF: Comp	arator Referen	ce Select bit (r	on-inverting inp	out)					
	-		-	I CVREF voltage	-					
	0 = Non-inver	rting input conr	nects to CxINA	pin						
bit 1-0	CCH<1:0>: C	CCH<1:0>: Comparator Channel Select bits								
	11 = Inverting input of comparator connects to VBG									
				s to C2INB or C	2IND pin <sup>(2,3)</sup>					
			arator connect	•						
		y input of comp	arator connect							
	The CMPxIF bit is after the initial co	-	set any time th	iis mode is seled	cted and mus	t be cleared by t	he applicatio			
2: (	Comparators, 1 a	nd 3, use C2IN	B as an input	to the inverting	terminal. Con	nparator 2 uses	C2IND.			

#### REGISTER 24-1: CMxCON: COMPARATOR CONTROL x REGISTER

**3:** C1INC, C2INC and C2IND are all unavailable for 64-pin devices (PIC18F6XK22).

## REGISTER 24-2: CMSTAT: COMPARATOR STATUS REGISTER

R-x	R-x	R-x	U-0	U-0	U-0	U-0	U-0
CMP3OUT	CMP2OUT	CMP1OUT	—	—	—	—	—
bit 7							
Legend:							
R = Readable	e bit	W = Writable I	bit	U = Unimplem	ented bit, read	l as '0'	
-n = Value at	POR	'1' = Bit is set		'0' = Bit is cleared x = Bit is unknown			
bit 7-5 CMP<3:1>OUT: Comparator x Status bits							
<u>If CPOL (CMxCON&lt;5&gt;)= o (non-inverted polarity):</u>							
1 = Comparator x's VIN+ > VIN-							

0 = Comparator x's Vin+ < Vin-

If CPOL = 1 (inverted polarity):

1 = Comparator x's Vin+ < Vin-

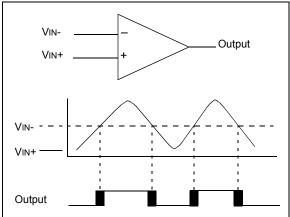
0 = Comparator x's Vin+ > Vin-

bit 4-0 Unimplemented: Read as '0'

### 24.2 Comparator Operation

A single comparator is shown in Figure 24-2, along with the relationship between the analog input levels and the digital output. When the analog input at VIN+ is less than the analog input, VIN-, the output of the comparator is a digital low level. When the analog input at VIN+ is greater than the analog input, VIN-, the output of the comparator is a digital high level. The shaded areas of the output of the comparator, in Figure 24-2, represent the uncertainty due to input offsets and response time.

FIGURE 24-2: SINGLE COMPARATOR



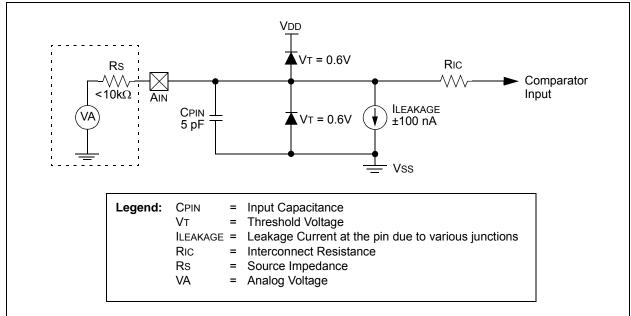
#### 24.3 Comparator Response Time

Response time is the minimum time, after selecting a new reference voltage or input source, before the comparator output has a valid level. The response time of the comparator differs from the settling time of the voltage reference. Therefore, both of these times must be considered when determining the total response to a comparator input change; otherwise, the maximum delay of the comparators should be used (see **Section 31.0 "Electrical Characteristics"**).

## 24.4 Analog Input Connection Considerations

A simplified circuit for an analog input is shown in Figure 24-3. Since the analog pins are connected to a digital output, they have reverse biased diodes to VDD and VSS. The analog input, therefore, must be between VSS and VDD. If the input voltage deviates from this range by more than 0.6V in either direction, one of the diodes is forward biased and a latch-up condition may occur.

A maximum source impedance of  $10 \text{ k}\Omega$  is recommended for the analog sources. Any external component connected to an analog input pin, such as a capacitor or a Zener diode, should have very little leakage current.



#### FIGURE 24-3: COMPARATOR ANALOG INPUT MODEL

## 24.5 Comparator Control and Configuration

Each comparator has up to eight possible combinations of inputs: up to four external analog inputs and one of two Internal Reference Voltages.

All of the comparators allow a selection of the signal from pin, CxINA, or the voltage from the comparator reference (CVREF) on the non-inverting channel. This is compared to either CxINB, CxINC, C2INB/C2IND or the microcontroller's fixed Internal Reference Voltage (VBG, 1.024V nominal) on the inverting channel. The comparator inputs and outputs are tied to fixed I/O pins, defined in Table 24-1. The available comparator configurations and their corresponding bit settings are shown in Figure 24-4.

TABLE 24-1:	COMPARATOR INPUTS AND
	OUTPUTS

Comparator	Input or Output	I/O Pin			
	C1INA (VIN+)	RF6			
	C1INB (VIN-)	RF5			
1	C1INC <sup>(1)</sup> (VIN-)	RH6			
	C2INB (VIN-)	RF3			
	C1OUT	RF2			
	C2INA (VIN+)	RF4			
	C2INB (VIN-)	RF3			
2	C2INC <sup>(1)</sup> (VIN-)	RH4			
	C2IND <sup>(1)</sup> (VIN-)	RH5			
	C2OUT	RF1			
	C3INA (VIN+)	RG2			
	C3INB (VIN-)	RG3			
3	C3INC (VIN-)	RG4			
	C2INB (VIN-)	RF3			
	C3OUT	RG1			

Note 1: C1INC, C2INC and C2IND are all unavailable for 64-pin devices (PIC18F6XK22).

## 24.5.1 COMPARATOR ENABLE AND INPUT SELECTION

Setting the CON bit of the CMxCON register (CMxCON<7>) enables the comparator for operation. Clearing the CON bit disables the comparator, resulting in minimum current consumption.

The CCH<1:0> bits in the CMxCON register (CMxCON<1:0>) direct either one of three analog input pins, or the Internal Reference Voltage (VBG), to the comparator, VIN-. Depending on the Comparator

operating mode, either an external or Internal Reference Voltage may be used. For external analog pins that are unavailable in 64-pin devices (C1INC, C2INC and C2IND), the corresponding configurations that use them as inputs are unavailable.

The analog signal present at VIN- is compared to the signal at VIN+ and the digital output of the comparator is adjusted accordingly.

The external reference is used when CREF = 0 (CMxCON<2>) and VIN+ is connected to the CxINA pin. When external reference voltages are used, the comparator module can be configured to have the reference sources externally. The reference signal must be between Vss and VDD, and can be applied to either pin of the comparator.

The comparator module also allows the selection of an internally generated reference voltage from the Comparator Voltage Reference (CVREF) module. This module is described in more detail in **Section 25.0 "Comparator Voltage Reference Module"**. The reference from the comparator reference voltage module is only available when CREF = 1. In this mode, the Internal Reference Voltage is applied to the comparator's VIN+ pin.

Note:	The comparator input pin, selected by					
	CCH<1:0>, must be configured as an input					
	by setting both the corresponding TRISF,					
	TRISG or TRISH bit and the corresponding					
	ANSELx bit in the ANCONx register.					

## 24.5.2 COMPARATOR ENABLE AND OUTPUT SELECTION

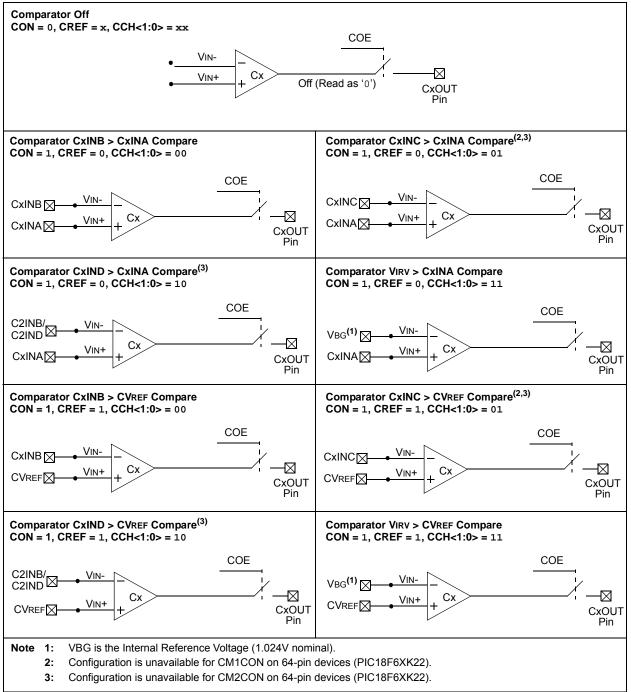
The comparator outputs are read through the CMSTAT register. The CMSTAT<5> bit reads the Comparator 1 output, CMSTAT<6> reads Comparator 2 output and CMSTAT<7> reads Comparator 3 output. These bits are read-only.

The comparator outputs may also be directly output to the RF2, RF1 and RG1 I/O pins by setting the COE bit (CMxCON<6>). When enabled, multiplexers in the output path of the pins switch to the output of the comparator. While in this mode, the TRISF<2:1> and TRISG<1> bits still function as the digital output enable bits for the RF2, RF1 and RG1 pins.

By default, the comparator's output is at logic high whenever the voltage on VIN+ is greater than on VIN-. The polarity of the comparator outputs can be inverted using the CPOL bit (CMxCON<5>).

The uncertainty of each of the comparators is related to the input offset voltage and the response time given in the specifications, as discussed in **Section 24.2 "Comparator Operation"**.

FIGURE 24-4: COMPARATOR CONFIGURATIONS



#### 24.6 Comparator Interrupts

The comparator interrupt flag is set whenever any of the following occurs:

- · Low-to-high transition of the comparator output
- High-to-low transition of the comparator output
- Any change in the comparator output

The comparator interrupt selection is done by the EVPOL<1:0> bits in the CMxCON register (CMxCON<4:3>).

In order to provide maximum flexibility, the output of the comparator may be inverted using the CPOL bit in the CMxCON register (CMxCON<5>). This is functionally identical to reversing the inverting and non-inverting inputs of the comparator for a particular mode.

An interrupt is generated on the low-to-high or high-tolow transition of the comparator output. This mode of interrupt generation is dependent on EVPOL<1:0> in the CMxCON register. When EVPOL<1:0> = 01 or 10, the interrupt is generated on a low-to-high or high-tolow transition of the comparator output. Once the interrupt is generated, it is required to clear the interrupt flag by software. When EVPOL<1:0> = 11, the comparator interrupt flag is set whenever there is a change in the output value of either comparator. Software will need to maintain information about the status of the output bits, as read from CMSTAT<7:5>, to determine the actual change that occurred.

The CMPxIF bits (PIR6<2:0>) are the Comparator Interrupt Flags. The CMPxIF bits must be reset by clearing them. Since it is also possible to write a '1' to this register, a simulated interrupt may be initiated. Table 24-2 shows the interrupt generation with respect to comparator input voltages and EVPOL bit settings.

Both the CMPxIE bits (PIE6<2:0>) and the PEIE bit (INTCON<6>) must be set to enable the interrupt. In addition, the GIE bit (INTCON<7>) must also be set. If any of these bits are clear, the interrupt is not enabled, though the CMPxIF bits will still be set if an interrupt condition occurs.

A simplified diagram of the interrupt section is shown in Figure 24-3.

Note: The CMPxIF bits will not be set when EVPOL<1:0> = 00.

CPOL	EVPOL<1:0>	Comparator Input Change	CxOUT Transition	Interrupt Generated
	0.0	VIN+ > VIN-	Low-to-High	No
	00	VIN+ < VIN-	High-to-Low	No
	01	VIN+ > VIN-	Low-to-High	Yes
0	ÛL	VIN+ < VIN-	High-to-Low	No
0	1.0	VIN+ > VIN-	Low-to-High	No
	10	VIN+ < VIN-	High-to-Low	Yes
		VIN+ > VIN-	Low-to-High	Yes
		Vin+ < Vin-	High-to-Low	Yes
	00	VIN+ > VIN-	High-to-Low	No
		VIN+ < VIN-	Low-to-High	No
		VIN+ > VIN-	High-to-Low	No
1		VIN+ < VIN-	Low-to-High	Yes
	10	VIN+ > VIN-	High-to-Low	Yes
	TO	VIN+ < VIN-	Low-to-High	No
	11	VIN+ > VIN-	High-to-Low	Yes
	11	VIN+ < VIN-	Low-to-High	Yes

 TABLE 24-2:
 COMPARATOR INTERRUPT GENERATION

### 24.7 Comparator Operation During Sleep

When a comparator is active and the device is placed in Sleep mode, the comparator remains active and the interrupt is functional, if enabled. This interrupt will wake up the device from Sleep mode, when enabled. Each operational comparator will consume additional current. To minimize power consumption while in Sleep mode, turn off the comparators (CON = 0) before entering Sleep. If the device wakes up from Sleep, the contents of the CMxCON register are not affected.

## 24.8 Effects of a Reset

A device Reset forces the CMxCON registers to their Reset state. This forces both comparators and the voltage reference to the OFF state.

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
INTCON	GIE/GIEH	PEIE/GIEL	TMR0IE	INT0IE	RBIE	TMR0IF	INT0IF	RBIF
PIR6	_	_	_	EEIF		CMP3IF	CMP2IF	CMP1IF
PIE6	—	_	_	EEIE	_	CMP3IE	CMP2IE	CMP1IE
IPR6	—	—	—	EEIP	—	CMP3IP	CMP2IP	CMP1IP
CM1CON	CON	COE	CPOL	EVPOL1	EVPOL0	CREF	CCH1	CCH0
CM2CON	CON	COE	CPOL	EVPOL1	EVPOL0	CREF	CCH1	CCH0
CM3CON	CON	COE	CPOL	EVPOL1	EVPOL0	CREF	CCH1	CCH0
CVRCON	CVREN	CVROE	CVRSS	CVR4	CVR3	CVR2	CVR1	CVR0
CMSTAT	CMP3OUT	CMP2OUT	CMP10UT	_	_		_	_
PORTF	RF7	RF6	RF5	RF4	RF3	RF2	RF1	—
LATF	LATF7	LATF6	LATF5	LATF4	LATF3	LATF2	LATF1	—
TRISF	TRISF7	TRISF6	TRISF5	TRISF4	TRISF3	TRISF2	TRISF1	—
PORTG	—	—	RG5 <sup>(1)</sup>	RG4	RG3	RG2	RG1	RG0
LATG	—	—	—	LATG4	LATG3	LATG2	LATG1	LATG0
TRISG	—	—	—	TRISG4	TRISG3	TRISG2	TRISG1	TRISG0
PORTH	RH7	RH6	RH5	RH4	RH3	RH2	RH1	RH0
LATH	LATH7	LATH6	LATH5	LATH4	LATH3	LATH2	LATH1	LATH0
TRISH	TRISH7	TRISH6	TRISH5	TRISH4	TRISH3	TRISH2	TRISH1	TRISH0
ANCON0	ANSEL7	ANSEL6	ANSEL5	ANSEL4	ANSEL3	ANSEL2	ANSEL1	ANSEL0
ANCON1	ANSEL15	ANSEL14	ANSEL13	ANSEL12	ANSEL11	ANSEL10	ANSEL9	ANSEL8
ANCON2	ANSEL23	ANSEL22	ANSEL21	ANSEL20	ANSEL19	ANSEL18	ANSEL17	ANSEL16
PMD0	CCP3MD	CCP2MD	CCP1MD	UART2MD	UART1MD	SSP2MD	SSP1MD	ADCMD

TABLE 24-3:	REGISTERS ASSOCIATED WITH COMPARATOR MODULE
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**Legend:** — = unimplemented, read as '0'.

Note 1: Bit is available when Master Clear is disabled (MCLRE = 0). When MCLRE is set, the bit is unimplemented.

## 25.0 COMPARATOR VOLTAGE REFERENCE MODULE

The comparator voltage reference is a 32-tap resistor ladder network that provides a selectable reference voltage. Although its primary purpose is to provide a reference for the analog comparators, it may also be used independently of them.

A block diagram of the module is shown in Figure 25-1. The resistor ladder is segmented to provide a range of CVREF values and has a power-down function to conserve power when the reference is not being used. The module's supply reference can be provided from either device VDD/VSS or an external voltage reference.

#### 25.1 Configuring the Comparator Voltage Reference

The comparator voltage reference module is controlled through the CVRCON register (Register 25-1). The comparator voltage reference provides a range of output voltage with 32 levels.

The CVR<4:0> selection bits (CVRCON<4:0>) offer a range of output voltages. Equation 25-1 shows the how the comparator voltage reference is computed.

## EQUATION 25-1:

If CVRSS = 1:

 $CVREF = (VREF-) + (CVR < 4:0 > /32) \cdot (VREF+ - VREF-)$ 

<u>If CVRSS = 0:</u>

 $CVREF = (AVSS) + (CVR < 4:0 > /32) \cdot (AVDD - AVSS)$ 

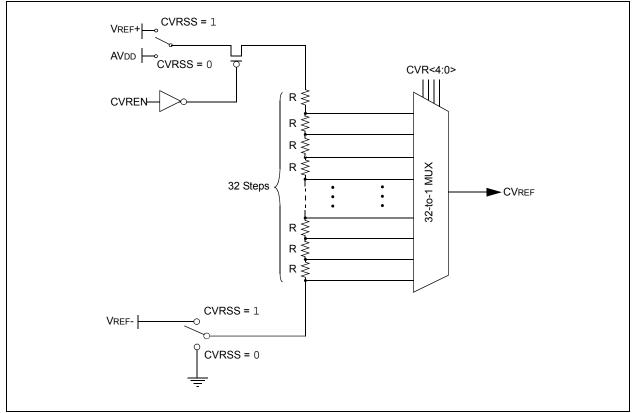
The comparator reference supply voltage can come from either VDD and VSS, or the external VREF+ and VREF- that are multiplexed with RA3 and RA2. The voltage source is selected by the CVRSS bit (CVRCON<5>).

The settling time of the comparator voltage reference must be considered when changing the CVREF output (see Table 31-2 in Section 31.0 "Electrical Characteristics").

## REGISTER 25-1: CVRCON: COMPARATOR VOLTAGE REFERENCE CONTROL REGISTER

R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0
	1	1 1		1		1	
CVREN	CVROE	CVRSS	CVR4	CVR3	CVR2	CVR1	CVR0
bit 7							bit 0
Legend:							
R = Readable	e bit	W = Writable b	pit	U = Unimplem	nented bit, rea	d as '0'	
-n = Value at	POR	'1' = Bit is set		'0' = Bit is clea	ared	x = Bit is unkr	nown
bit 7 bit 6	1 = CVREF ci 0 = CVREF ci <b>CVROE:</b> Com 1 = CVREF vo	nparator Voltage rcuit is powered rcuit is powered nparator VREF C oltage level is o oltage level is di	l on l down output Enable utput on CVRE	bit F pin			
bit 5 bit 4-0	1 = Compara 0 = Compara	ator reference so	ource, CVRSR ource, CVRSR	on bit c = VREF+ $-$ VRI c = AVDD $-$ AVs tion 0 $\leq$ CVR<4:	S		
	When CVRSS CVREF = (VRE When CVRSS	<u>5 = 1:</u> EF-) + (CVR<4:0	>/32) • (VREF	+ – Vref-)			





## 25.2 Voltage Reference Accuracy/Error

The full range of voltage reference cannot be realized due to the construction of the module. The transistors on the top and bottom of the resistor ladder network (Figure 25-1) keep CVREF from approaching the reference source rails. The voltage reference is derived from the reference source; therefore, the CVREF output changes with fluctuations in that source. The tested absolute accuracy of the voltage reference can be found in Section 31.0 "Electrical Characteristics".

## 25.3 Operation During Sleep

When the device wakes up from Sleep through an interrupt or a Watchdog Timer time-out, the contents of the CVRCON register are not affected. To minimize current consumption in Sleep mode, the voltage reference should be disabled.

## 25.4 Effects of a Reset

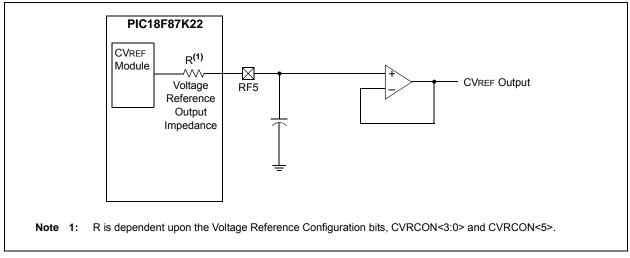
A device Reset disables the voltage reference by clearing bit, CVREN (CVRCON<7>). This Reset also disconnects the reference from the RF5 pin by clearing bit, CVROE (CVRCON<6>).

## 25.5 Connection Considerations

The voltage reference module operates independently of the comparator module. The output of the reference generator may be connected to the RF5 pin if the CVROE bit is set. Enabling the voltage reference output onto RF5 when it is configured as a digital input will increase current consumption. Connecting RF5 as a digital output, with CVRSS enabled, will also increase current consumption.

The RF5 pin can be used as a simple D/A output with limited drive capability. Due to the limited current drive capability, a buffer must be used on the voltage reference output for external connections to VREF. Figure 25-2 shows an example buffering technique.

#### FIGURE 25-2: COMPARATOR VOLTAGE REFERENCE OUTPUT BUFFER EXAMPLE



#### TABLE 25-1: REGISTERS ASSOCIATED WITH COMPARATOR VOLTAGE REFERENCE

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
CVRCON	CVREN	CVROE	CVRSS	CVR4	CVR3	CVR2	CVR1	CVR0
CM1CON	CON	COE	CPOL	EVPOL1	EVPOL0	CREF	CCH1	CCH0
CM2CON	CON	COE	CPOL	EVPOL1	EVPOL0	CREF	CCH1	CCH0
CM3CON	CON	COE	CPOL	EVPOL1	EVPOL0	CREF	CCH1	CCH0
TRISF	TRISF7	TRISF6	TRISF5	TRISF4	TRISF3	TRISF2	TRISF1	—
TRISA	TRISA7	TRISA6	TRISA5	TRISA4	TRISA3	TRISA2	TRISA1	TRISA0
ANCON0	ANSEL7	ANSEL6	ANSEL5	ANSEL4	ANSEL3	ANSEL2	ANSEL1	ANSEL0
ANCON1	ANSEL15	ANSEL14	ANSEL13	ANSEL12	ANSEL11	ANSEL10	ANSEL9	ANSEL8

**Legend:** — = unimplemented, read as '0'. Shaded cells are not used with the comparator voltage reference.

NOTES:

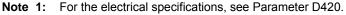
## 26.0 HIGH/LOW-VOLTAGE DETECT (HLVD)

The PIC18F87K22 family of devices has a High/Low-Voltage Detect module (HLVD). This is a programmable circuit that sets both a device voltage trip point and the direction of change from that point. If the device experiences an excursion past the trip point in that direction, an interrupt flag is set. If the interrupt is enabled, the program execution branches to the interrupt vector address and the software responds to the interrupt. The High/Low-Voltage Detect Control register (Register 26-1) completely controls the operation of the HLVD module. This allows the circuitry to be "turned off" by the user under software control, which minimizes the current consumption for the device.

The module's block diagram is shown in Figure 26-1.

## REGISTER 26-1: HLVDCON: HIGH/LOW-VOLTAGE DETECT CONTROL REGISTER

R/W-0	R-0	R-0	R/W-0	R/W-0	R/W-1	R/W-0	R/W-0
VDIRMAG	BGVST	IRVST	HLVDEN	HLVDL3 <sup>(1)</sup>	HLVDL2 <sup>(1)</sup>	HLVDL1 <sup>(1)</sup>	HLVDL0 <sup>(1)</sup>
bit 7							bit 0
Legend:							
R = Readable	bit	W = Writable	bit	U = Unimpler	nented bit, read	l as '0'	
-n = Value at F	POR	'1' = Bit is set		'0' = Bit is cle	ared	x = Bit is unkr	nown
bit 7	VDIRMAG: V	oltage Direction	n Magnitude S	Select bit			
					oint (HLVDL<3:0		
	0 = Event occ	curs when volta	ge equals or	falls below trip	point (HLVDL<	3:0>)	
bit 6		d Gap Referend	•		ag bit		
		and gap voltag					
		and gap voltag					
bit 5		al Reference V	•	•			
					e interrupt flag ate the interrup		
		d the HLVD int				t hay at the spe	ecilieu voltage
bit 4	•	gh/Low-Voltage					
	1 = HLVD is						
	0 = HLVD is	disabled					
bit 3-0	HLVDL<3:0>	: Voltage Deteo	ction Limit bits	. <mark>(1</mark> )			
		• •	it is used (inp	ut comes from	the HLVDIN pin	ı)	
	1110 <b>= Maxir</b>	num setting					
	•						
	•						
	0000 = Minim	num setting					
		na sifinationa a		D 400			



The module is enabled by setting the HLVDEN bit (HLVDCON<4>). Each time the HLVD module is enabled, the circuitry requires some time to stabilize. The IRVST bit (HLVDCON<5>) is a read-only bit used to indicate when the circuit is stable. The module can only generate an interrupt after the circuit is stable and IRVST is set.

The VDIRMAG bit (HLVDCON<7>) determines the overall operation of the module. When VDIRMAG is cleared, the module monitors for drops in VDD below a predetermined set point. When the bit is set, the module monitors for rises in VDD above the set point.

## 26.1 Operation

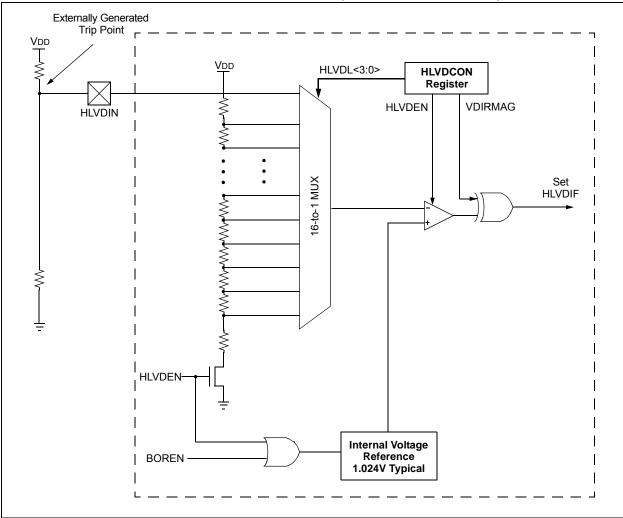
When the HLVD module is enabled, a comparator uses an internally generated voltage reference as the set point. The set point is compared with the trip point, where each node in the resistor divider represents a trip point voltage. The "trip point" voltage is the voltage level at which the device detects a high or low-voltage event, depending on the configuration of the module.

When the supply voltage is equal to the trip point, the voltage tapped off of the resistor array is equal to the internal reference voltage generated by the voltage reference module. The comparator then generates an interrupt signal by setting the HLVDIF bit.

The trip point voltage is software programmable to any of 16 values. The trip point is selected by programming the HLVDL<3:0> bits (HLVDCON<3:0>).

The HLVD module has an additional feature that allows the user to supply the trip voltage to the module from an external source. This mode is enabled when bits, HLVDL<3:0>, are set to '1111'. In this state, the comparator input is multiplexed from the external input pin, HLVDIN. This gives users the flexibility of configuring the High/Low-Voltage Detect interrupt to occur at any voltage in the valid operating range.

FIGURE 26-1: HLVD MODULE BLOCK DIAGRAM (WITH EXTERNAL INPUT)



## 26.2 HLVD Setup

To set up the HLVD module:

- 1. Select the desired HLVD trip point by writing the value to the HLVDL<3:0> bits.
- 2. Set the VDIRMAG bit to detect high voltage (VDIRMAG = 1) or low voltage (VDIRMAG = 0).
- 3. Enable the HLVD module by setting the HLVDEN bit.
- 4. Clear the HLVD interrupt flag (PIR2<2>), which may have been set from a previous interrupt.
- If interrupts are desired, enable the HLVD interrupt by setting the HLVDIE and GIE bits (PIE2<2> and INTCON<7>, respectively).

An interrupt will not be generated until the IRVST bit is set.

**Note:** Before changing any module settings (VDIRMAG, LVDL<3:0>), first disable the module (LVDEN = 0), make the changes and re-enable the module. This prevents the generation of false HLVD events.

## 26.3 Current Consumption

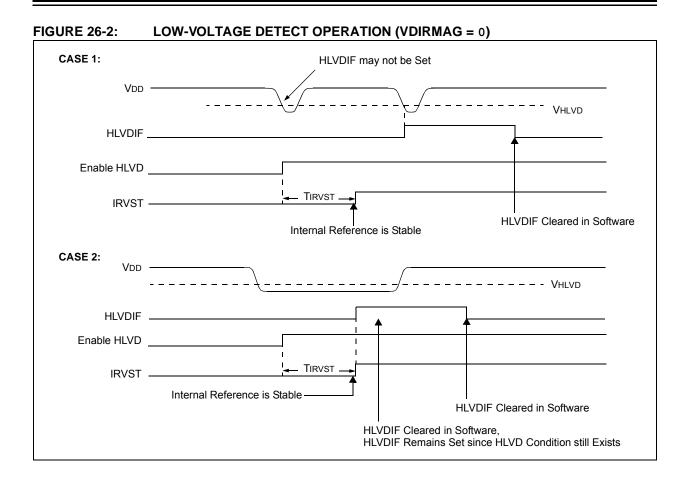
When the module is enabled, the HLVD comparator and voltage divider are enabled and consume static current. The total current consumption, when enabled, is specified in electrical specification Parameter D022B (Table 31-13).

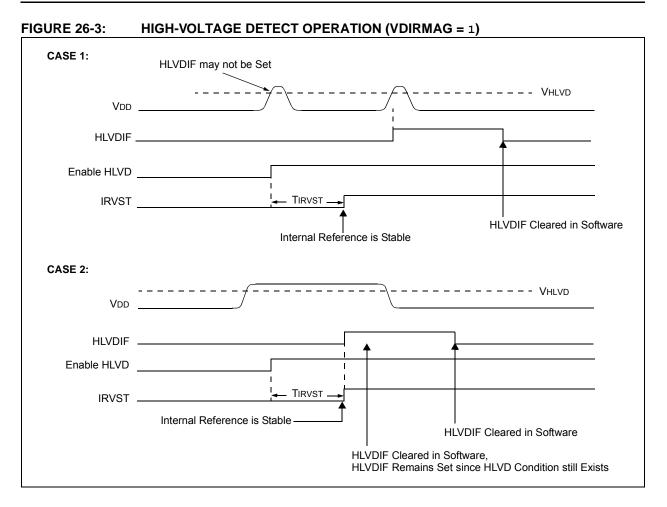
Depending on the application, the HLVD module does not need to operate constantly. To reduce current requirements, the HLVD circuitry may only need to be enabled for short periods where the voltage is checked. After such a check, the module could be disabled.

## 26.4 HLVD Start-up Time

The internal reference voltage of the HLVD module, specified in electrical specification Parameter 37 (Section 31.0 "Electrical Characteristics"), may be used by other internal circuitry, such as the programmable Brown-out Reset. If the HLVD or other circuits using the voltage reference are disabled to lower the device's current consumption, the reference voltage circuit will require time to become stable before a low or high-voltage condition can be reliably detected. This start-up time, TIRVST, is an interval that is independent of device clock speed. It is specified in electrical specification Parameter 37 (Table 31-13).

The HLVD interrupt flag is not enabled until TIRVST has expired and a stable reference voltage is reached. For this reason, brief excursions beyond the set point may not be detected during this interval (see Figure 26-2 or Figure 26-3).

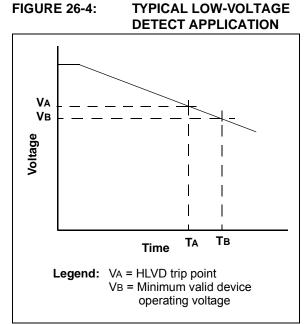




#### 26.5 Applications

In many applications, it is desirable to detect a drop below, or rise above, a particular voltage threshold. For example, the HLVD module could be periodically enabled to detect Universal Serial Bus (USB) attach or detach. This assumes the device is powered by a lower voltage source than the USB when detached. An attach would indicate a High-Voltage Detect from, for example, 3.3V to 5V (the voltage on USB) and vice versa for a detach. This feature could save a design a few extra components and an attach signal (input pin).

For general battery applications, Figure 26-4 shows a possible voltage curve. Over time, the device voltage decreases. When the device voltage reaches voltage, VA, the HLVD logic generates an interrupt at time, TA. The interrupt could cause the execution of an Interrupt Service Routine (ISR), which would allow the application to perform "housekeeping tasks" and a controlled shutdown before the device voltage exits the valid operating range at TB. This would give the application a time window, represented by the difference between TA and TB, to safely exit.



### 26.6 Operation During Sleep

When enabled, the HLVD circuitry continues to operate during Sleep. If the device voltage crosses the trip point, the HLVDIF bit will be set and the device will wake up from Sleep. Device execution will continue from the interrupt vector address if interrupts have been globally enabled.

### 26.7 Effects of a Reset

A device Reset forces all registers to their Reset state. This forces the HLVD module to be turned off.

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
HLVDCON	VDIRMAG	BGVST	IRVST	HLVDEN	HLVDL3	HLVDL2	HLVDL1	HLVDL0
INTCON	GIE/GIEH	PEIE/GIEL	TMR0IE	INT0IE	RBIE	TMR0IF	INT0IF	RBIF
PIR2	OSCFIF	—	SSP2IF	BLC2IF	BCL1IF	HLVDIF	TMR3IF	TMR3GIF
PIE2	OSCFIE	_	SSP2IE	BLC2IE	BCL1IE	HLVDIE	TMR3IE	TMR3GIE
IPR2	OSCFIP	_	SSP2IP	BLC2IP	BCL1IP	HLVDIP	TMR3IP	TMR3GIP
TRISA	TRISA7 <sup>(1)</sup>	TRISA6 <sup>(1)</sup>	TRISA5	TRISA4	TRISA3	TRISA2	TRISA1	TRISA0
ANCON0	ANSEL7	ANSEL6	ANSEL5	ANSEL4	ANSEL3	ANSEL2	ANSEL1	ANSEL0

TABLE 26-1: REGISTERS ASSOCIATED WITH HIGH/LOW-VOLTAGE DETECT MODULE

**Legend:** — = unimplemented, read as '0'. Shaded cells are unused by the HLVD module.

**Note 1:** PORTA<7:6> and their direction bits are individually configured as port pins based on various primary oscillator modes. When disabled, these bits read as '0'.

## 27.0 CHARGE TIME MEASUREMENT UNIT (CTMU)

The Charge Time Measurement Unit (CTMU) is a flexible analog module that provides accurate differential time measurement between pulse sources, as well as asynchronous pulse generation. By working with other on-chip analog modules, the CTMU can precisely measure time, capacitance and relative changes in capacitance or generate output pulses with a specific time delay. The CTMU is ideal for interfacing with capacitive-based sensors.

The module includes these key features:

- Up to 24 channels available for capacitive or time measurement input
- · On-chip precision current source
- Four-edge input trigger sources
- Polarity control for each edge source
- Control of edge sequence

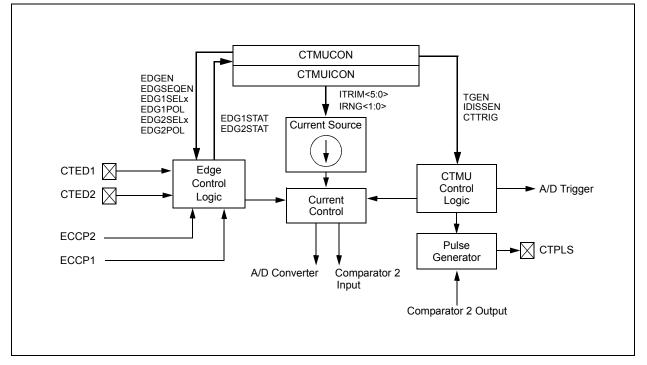
## FIGURE 27-1: CTMU BLOCK DIAGRAM

- Control of response to edges
- · Time measurement resolution of 1 nanosecond
- · High-precision time measurement
- Time delay of external or internal signal asynchronous to system clock
- Accurate current source suitable for capacitive measurement

The CTMU works in conjunction with the A/D Converter to provide up to 24 channels for time or charge measurement, depending on the specific device and the number of A/D channels available. When configured for time delay, the CTMU is connected to one of the analog comparators. The level-sensitive input edge sources can be selected from four sources: two external inputs or the ECCP1/ECCP2 Special Event Triggers.

The CTMU special event can trigger the Analog-to-Digital Converter module.

Figure 27-1 provides a block diagram of the CTMU.



## 27.1 CTMU Registers

The control registers for the CTMU are:

- CTMUCONH
- CTMUCONL
- CTMUICON

The CTMUCONH and CTMUCONL registers (Register 27-1 and Register 27-2) contain control bits for configuring the CTMU module edge source selection, edge source polarity selection, edge sequencing, A/D trigger, analog circuit capacitor discharge and enables. The CTMUICON register (Register 27-3) has bits for selecting the current source range and current source trim.

#### REGISTER 27-1: CTMUCONH: CTMU CONTROL HIGH REGISTER

R/W-0	U-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0
CTMUEN	—	CTMUSIDL	TGEN	EDGEN	EDGSEQEN	IDISSEN	CTTRIG
bit 7							bit 0

Legend:			
R = Readable bit	W = Writable bit	U = Unimplemented bit,	, read as '0'
-n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown

bit 7	CTMUEN: CTMU Enable bit
	<ul><li>1 = Module is enabled</li><li>0 = Module is disabled</li></ul>
bit 6	Unimplemented: Read as '0'
bit 5	CTMUSIDL: Stop in Idle Mode bit
	<ul><li>1 = Discontinue module operation when device enters Idle mode</li><li>0 = Continue module operation in Idle mode</li></ul>
bit 4	TGEN: Time Generation Enable bit
	<ul><li>1 = Enables edge delay generation</li><li>0 = Disables edge delay generation</li></ul>
bit 3	EDGEN: Edge Enable bit
	<ul><li>1 = Edges are not blocked</li><li>0 = Edges are blocked</li></ul>
bit 2	EDGSEQEN: Edge Sequence Enable bit
	<ul> <li>1 = Edge 1 event must occur before Edge 2 event can occur</li> <li>0 = No edge sequence is needed</li> </ul>
bit 1	IDISSEN: Analog Current Source Control bit
	<ul> <li>1 = Analog current source output is grounded</li> <li>0 = Analog current source output is not grounded</li> </ul>
bit 0	<b>CTTRIG:</b> Trigger Control bit 1 = Trigger output is enabled 0 = Trigger output is disabled

#### REGISTER 27-2: CTMUCONL: CTMU CONTROL LOW REGISTER

R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0				
EDG2POL	EDG2SEL1	EDG2SEL0	EDG1POL	EDG1SEL1	EDG1SEL0	EDG2STAT	EDG1STAT				
bit 7						•	bit 0				
Legend:											
R = Readabl	e bit	W = Writable b	bit	U = Unimplem	nented bit, read	l as '0'					
-n = Value at	POR	'1' = Bit is set		'0' = Bit is clea	ared	x = Bit is unkr	nown				
bit 7	EDG2POL: E	dge 2 Polarity S	Select bit								
	•	programmed f	•	•							
	0	programmed f	0	0							
bit 6-5		0>: Edge 2 Sou	urce Select bit	S							
	11 = CTED1 pin										
		10 = CTED2 pin 01 = ECCP1 Special Event Trigger									
		Special Event T									
bit 4		dge 1 Polarity S									
	1 = Edge 1 is	programmed f	or a positive e	dge response							
	•	programmed f	•	•							
bit 3-2	EDG1SEL<1:	<b>0&gt;:</b> Edge 1 Sou	urce Select bit	S							
	11 = CTED1	pin									
	10 = CTED2										
	01 = ECCP1 Special Event Trigger 00 = ECCP2 Special Event Trigger										
		•									
bit 1		EDG2STAT: Edge 2 Status bit									
		vent has occurr									
	•	vent has not oc									
bit 0		Edge 1 Status b									
		vent has occurr									
	v = reage real	vent has not oc	currea								

R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0					
ITRIM5	ITRIM4	ITRIM3	ITRIM2	ITRIM1		IRNG1	IRNG0					
bit 7						IKINGT	bit 0					
Legend:												
R = Readal	ole bit	W = Writable	bit	U = Unimplen	nented bit, read	d as '0'						
-n = Value a	at POR	'1' = Bit is set		'0' = Bit is clea	ared	x = Bit is unkr	nown					
bit 7-2	ITRIM<5:0>:	Current Source	e Trim bits									
	011111 = Maximum positive change from nominal current 011110											
	•											
	000001 = Mir	nimum positive	change from	nominal current								
			0	d by IRNG<1:0>								
	111111 = Minimum negative change from nominal current											
	100010											
		ximum negativ	e change from	n nominal currer	nt							
bit 1-0	IRNG<1:0>: (	Current Source	Range Select	bits								
	11 = 100 x Ba		-									
	10 <b>= 10 x Bas</b>	se Current										
	01 = Base Cu	01 = Base Current Level (0.55 μA nominal)										

#### REGISTER 27-3: CTMUICON: CTMU CURRENT CONTROL REGISTER

00 = Current Source Disabled

## 27.2 CTMU Operation

The CTMU works by using a fixed current source to charge a circuit. The type of circuit depends on the type of measurement being made.

In the case of charge measurement, the current is fixed and the amount of time the current is applied to the circuit is fixed. The amount of voltage read by the A/D becomes a measurement of the circuit's capacitance.

In the case of time measurement, the current, as well as the capacitance of the circuit, is fixed. In this case, the voltage read by the A/D is representative of the amount of time elapsed from the time the current source starts and stops charging the circuit.

If the CTMU is being used as a time delay, both capacitance and current source are fixed, as well as the voltage supplied to the comparator circuit. The delay of a signal is determined by the amount of time it takes the voltage to charge to the comparator threshold voltage.

### 27.2.1 THEORY OF OPERATION

The operation of the CTMU is based on the equation for charge:

$$C = I \bullet \frac{dV}{dT}$$

More simply, the amount of charge measured in coulombs in a circuit is defined as current in amperes (I) multiplied by the amount of time in seconds that the current flows (t). Charge is also defined as the capacitance in farads (C) multiplied by the voltage of the circuit (V). It follows that:

 $I \bullet t = C \bullet V$ 

The CTMU module provides a constant, known current source. The A/D Converter is used to measure (V) in the equation, leaving two unknowns: capacitance (C) and time (t). The above equation can be used to calculate capacitance or time, by either the relationship using the known fixed capacitance of the circuit:

 $t = (C \bullet V)/I$ 

or by:

$$C = (I \bullet t)/V$$

using a fixed time that the current source is applied to the circuit.

## 27.2.2 CURRENT SOURCE

At the heart of the CTMU is a precision current source, designed to provide a constant reference for measurements. The level of current is user-selectable across three ranges or a total of two orders of magnitude, with the ability to trim the output in  $\pm 2\%$  increments (nominal). The current range is selected by the IRNG<1:0> bits (CTMUICON<1:0>), with a value of '00' representing the lowest range.

Current trim is provided by the ITRIM<5:0> bits (CTMUICON<7:2>). These six bits allow trimming of the current source in steps of approximately 2% per step. Half of the range adjusts the current source positively and the other half reduces the current source. A value of '000000' is the neutral position (no change). A value of '100000' is the maximum negative adjustment (approximately -62%) and '011111' is the maximum positive adjustment (approximately +62%).

## 27.2.3 EDGE SELECTION AND CONTROL

CTMU measurements are controlled by edge events occurring on the module's two input channels. Each channel, referred to as Edge 1 and Edge 2, can be configured to receive input pulses from one of the edge input pins (CTED1 and CTED2) or CCPx Special Event Triggers. The input channels are level-sensitive, responding to the instantaneous level on the channel rather than a transition between levels. The inputs are selected using the EDG1SEL and EDG2SEL bit pairs (CTMUCONL<3:2, 6:5>).

In addition to source, each channel can be configured for event polarity using the EDGE2POL and EDGE1POL bits (CTMUCONL<7,4>). The input channels can also be filtered for an edge event sequence (Edge 1 occurring before Edge 2) by setting the EDGSEQEN bit (CTMUCONH<2>).

## 27.2.4 EDGE STATUS

The CTMUCON register also contains two status bits: EDG2STAT and EDG1STAT (CTMUCONL<1:0>). Their primary function is to show if an edge response has occurred on the corresponding channel. The CTMU automatically sets a particular bit when an edge response is detected on its channel. The level-sensitive nature of the input channels also means that the status bits become set immediately if the channel's configuration is changed and matches the channel's current state.

The module uses the edge status bits to control the current source output to external analog modules (such as the A/D Converter). Current is only supplied to external modules when only one (not both) of the status bits is set. Current is shut off when both bits are either set or cleared. This allows the CTMU to measure current only during the interval between edges. After both status bits are set, it is necessary to clear them before another measurement is taken. Both bits should be cleared simultaneously, if possible, to avoid re-enabling the CTMU current source.

In addition to being set by the CTMU hardware, the edge status bits can also be set by software. This permits a user application to manually enable or disable the current source. Setting either (but not both) of the bits enables the current source. Setting or clearing both bits at once disables the source.

## 27.2.5 INTERRUPTS

The CTMU sets its interrupt flag (PIR3<3>) whenever the current source is enabled, then disabled. An interrupt is generated only if the corresponding interrupt enable bit (PIE3<3>) is also set. If edge sequencing is not enabled (i.e., Edge 1 must occur before Edge 2), it is necessary to monitor the edge status bits and determine which edge occurred last and caused the interrupt.

## 27.3 CTMU Module Initialization

The following sequence is a general guideline used to initialize the CTMU module:

- 1. Select the current source range using the IRNGx bits (CTMUICON<1:0>).
- 2. Adjust the current source trim using the ITRIMx bits (CTMUICON<7:2>).
- Configure the edge input sources for Edge 1 and Edge 2 by setting the EDG1SEL and EDG2SEL bits (CTMUCONL<3:2> and <6:5>, respectively).
- Configure the input polarities for the edge inputs using the EDG2POL and EDG1POL bits (CTMUCONL<7,4>).

The default configuration is for negative edge polarity (high-to-low transitions).

5. Enable edge sequencing using the EDGSEQEN bit (CTMUCONH<2>).

By default, edge sequencing is disabled.

6. Select the operating mode (Measurement or Time Delay) with the TGEN bit.

The default mode is Time/Capacitance Measurement.

 Configure the module to automatically trigger an A/D conversion when the second edge event has occurred, using the CTTRIG bit (CTMUCONH<0>).

The conversion trigger is disabled by default.

- 8. Discharge the connected circuit by setting the IDISSEN bit (CTMUCONH<1>).
- 9. After waiting a sufficient time for the circuit to discharge, clear IDISSEN.
- 10. Disable the module by clearing the CTMUEN bit (CTMUCONH<7>).
- 11. Clear the Edge Status bits, EDG2STAT and EDG1STAT (CTMUCONL<1:0>).
- 12. Enable both edge inputs by setting the EDGEN bit (CTMUCONH<3>).
- 13. Enable the module by setting the CTMUEN bit.

Depending on the type of measurement or pulse generation being performed, one or more additional modules may also need to be initialized and configured with the CTMU module:

- Edge Source Generation: In addition to the external edge input pins, CCPx Special Event Triggers can be used as edge sources for the CTMU.
- Capacitance or Time Measurement: The CTMU module uses the A/D Converter to measure the voltage across a capacitor that is connected to one of the analog input channels.
- Pulse Generation: When generating system clock independent, output pulses, the CTMU module uses Comparator 2 and the associated comparator voltage reference.

## 27.4 Calibrating the CTMU Module

The CTMU requires calibration for precise measurements of capacitance and time, as well as for accurate time delay. If the application only requires measurement of a relative change in capacitance or time, calibration is usually not necessary. An example of a less precise application is a capacitive touch switch, in which the touch circuit has a baseline capacitance and the added capacitance of the human body changes the overall capacitance of a circuit.

If actual capacitance or time measurement is required, two hardware calibrations must take place:

- The current source needs calibration to set it to a precise current.
- The circuit being measured needs calibration to measure or nullify any capacitance other than that to be measured.

#### 27.4.1 CURRENT SOURCE CALIBRATION

The current source on board the CTMU module has a range of  $\pm 60\%$  nominal for each of three current ranges. For precise measurements, it is possible to measure and adjust this current source by placing a high-precision resistor, RCAL, onto an unused analog channel. An example circuit is shown in Figure 27-2.

To measure the current source:

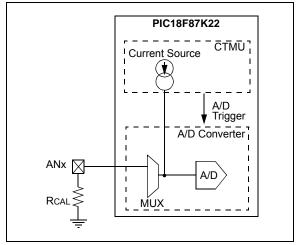
- 1. Initialize the A/D Converter.
- 2. Initialize the CTMU.
- Enable the current source by setting EDG1STAT (CTMUCONL<0>).
- 4. Issue the settling time delay.
- 5. Perform the A/D conversion.
- 6. Calculate the current source current using  $I = V/R_{CAL}$ , where  $R_{CAL}$  is a high-precision resistance and V is measured by performing an A/D conversion.

The CTMU current source may be trimmed with the trim bits in CTMUICON using an iterative process to get the exact current desired. Alternatively, the nominal value without adjustment may be used. That value may be stored by software for use in all subsequent capacitive or time measurements.

To calculate the value for  $\ensuremath{\mathrm{RCAL}}$  , the nominal current must be chosen. Then, the resistance can be calculated.

For example, if the A/D Converter reference voltage is 3.3V, use 70% of full scale (or 2.31V) as the desired approximate voltage to be read by the A/D Converter. If the range of the CTMU current source is selected to be 0.55  $\mu$ A, the resistor value needed is calculated as RCAL = 2.31V/0.55  $\mu$ A, for a value of 4.2 MΩ. Similarly, if the current source is chosen to be 5.5  $\mu$ A, RCAL would be 420,000Ω, and 42,000Ω if the current source is set to 55  $\mu$ A.

#### FIGURE 27-2: CTMU CURRENT SOURCE CALIBRATION CIRCUIT



A value of 70% of full-scale voltage is chosen to make sure that the A/D Converter was in a range that is well above the noise floor. If an exact current is chosen to incorporate the trimming bits from CTMUICON, the resistor value of RCAL may need to be adjusted accordingly. RCAL may also be adjusted to allow for available resistor values. RCAL should be of the highest precision available, in light of the precision needed for the circuit that the CTMU will be measuring. A recommended minimum would be 0.1% tolerance.

The following examples show a typical method for performing a CTMU current calibration.

• Example 27-1 demonstrates how to initialize the A/D Converter and the CTMU.

This routine is typical for applications using both modules.

• Example 27-2 demonstrates one method for the actual calibration routine.

This method manually triggers the A/D Converter to demonstrate the entire step-wise process. It is also possible to automatically trigger the conversion by setting the CTMU's CTTRIG bit (CTMUCONH<0>).

#### EXAMPLE 27-1: SETUP FOR CTMU CALIBRATION ROUTINES

```
#include "pl8cxxx.h"
void setup(void)
{ //CTMUCON - CTMU Control register
  CTMUCONH = 0 \times 00;
                         //make sure CTMU is disabled
  CTMUCONL = 0X90;
  //CTMU continues to run when emulator is stopped,CTMU continues
  //to run in idle mode, Time Generation mode disabled, Edges are blocked
  //No edge sequence order, Analog current source not grounded, trigger
  //output disabled, Edge2 polarity = positive level, Edge2 source =
  //source 0, Edgel polarity = positive level, Edgel source = source 0,
  // Set Edge status bits to zero
   //CTMUICON - CTMU Current Control Register
  CTMUICON = 0 \times 01;
                        //0.55uA, Nominal - No Adjustment
//Setup AD converter;
TRISA=0x04;
                         //set channel 2 as an input
  // Configured AN2 as an analog channel
  // ANCONO
  ANCON0 = 0 \times 04;
  // ANCON1
  ANCON1 = 0 XE0;
  // ADCON2
                     // Resulst format 1= Right justified
  ADCON2bits.ADFM=1;
                        // Acquition time 7 = 20TAD 2 = 4TAD 1=2TAD
  ADCON2bits.ACQT=1;
  ADCON2bits.ADCS=2;
                         // Clock conversion bits 6= FOSC/64 2=FOSC/32
  // ADCON0
                         // Vref+ = AVdd
  ADCON1bits.VCFG0 =0;
                         // Vref+ = AVdd
  ADCON0bits.VCFG1 =0;
  ADCON0bits.VCFG = 0;
                        // Vref- = AVss
  ADCON0bits.CHS=2;
                         // Select ADC channel
                         // Turn on ADC
  ADCON0bits.ADON=1;
}
```

#### EXAMPLE 27-2: CURRENT CALIBRATION ROUTINE

```
#include "pl8cxxx.h"
#define COUNT 500
                                         //@ 8MHz = 125uS.
#define DELAY for(i=0;i<COUNT;i++)</pre>
#define RCAL .027
                                         //R value is 4200000 (4.2M)
                                         //scaled so that result is in
                                         //1/100th of uA
#define ADSCALE 1023
                                         //for unsigned conversion 10 sig bits
#define ADREF 3.3
                                         //Vdd connected to A/D Vr+
int main(void)
{
   int i;
   int j = 0; //index for loop
   unsigned int Vread = 0;
   double VTot = 0;
   float Vavg=0, Vcal=0, CTMUISrc = 0; //float values stored for calcs
//assume CTMU and A/D have been setup correctly
//see Example 25-1 for CTMU & A/D setup
setup();
CTMUCONHbits.CTMUEN = 1;
                                         //Enable the CTMU
   for(j=0;j<10;j++)</pre>
   {
                                         //drain charge on the circuit
       CTMUCONHbits.IDISSEN = 1;
       DELAY;
                                         //wait 125us
       CTMUCONHbits.IDISSEN = 0;
                                         //end drain of circuit
       CTMUCONLbits.EDG1STAT = 1;
                                         //Begin charging the circuit
                                         //using CTMU current source
                                         //wait for 125us
       DELAY;
                                         //Stop charging circuit
       CTMUCONLbits.EDG1STAT = 0;
       PIR1bits.ADIF = 0;
                                         //make sure A/D Int not set
       ADCON0bits.GO=1;
                                         //and begin A/D conv.
       while(!PIR1bits.ADIF);
                                         //Wait for A/D convert complete
       Vread = ADRES;
                                         //Get the value from the A/D
       PIR1bits.ADIF = 0;
                                         //Clear A/D Interrupt Flag
       VTot += Vread;
                                         //Add the reading to the total
   }
   Vavg = (float)(VTot/10.000);
                                         //Average of 10 readings
   Vcal = (float)(Vavg/ADSCALE*ADREF);
   CTMUISrc = Vcal/RCAL;
                                         //CTMUISrc is in 1/100ths of uA
```

## 27.4.2 CAPACITANCE CALIBRATION

There is a small amount of capacitance from the internal A/D Converter sample capacitor, as well as stray capacitance from the circuit board traces and pads that affect the precision of capacitance measurements. A measurement of the stray capacitance can be taken by making sure the desired capacitance to be measured has been removed.

After removing the capacitance to be measured:

- 1. Initialize the A/D Converter and the CTMU.
- 2. Set EDG1STAT (= 1).
- 3. Wait for a fixed delay of time, t.
- 4. Clear EDG1STAT.
- 5. Perform an A/D conversion.
- 6. Calculate the stray and A/D sample capacitances:

$$COFFSET = CSTRAY + CAD = (I \cdot t)/V$$

Where:

- I is known from the current source measurement step
- · t is a fixed delay
- V is measured by performing an A/D conversion

This measured value is then stored and used for calculations of time measurement or subtracted for capacitance measurement. For calibration, it is expected that the capacitance of CSTRAY + CAD is approximately known; CAD is approximately 4 pF.

An iterative process may be required to adjust the time, t, that the circuit is charged to obtain a reasonable voltage reading from the A/D Converter. The value of t may be determined by setting COFFSET to a theoretical value and solving for t. For example, if CSTRAY is theoretically calculated to be 11 pF, and V is expected to be 70% of VDD or 2.31V, t would be:

$$(4 \text{ pF} + 11 \text{ pF}) \cdot 2.31 \text{V}/0.55 \text{ }\mu\text{A}$$

or 63 µs.

See Example 27-3 for a typical routine for CTMU capacitance calibration.

#### EXAMPLE 27-3: CAPACITANCE CALIBRATION ROUTINE

```
#include "pl8cxxx.h"
#define COUNT 25
                                          //@ 8MHz INTFRC = 62.5 us.
#define ETIME COUNT*2.5
                                          //time in uS
#define DELAY for(i=0;i<COUNT;i++)</pre>
#define ADSCALE 1023
                                         //for unsigned conversion 10 sig bits
#define ADREF 3.3
                                         //Vdd connected to A/D Vr+
#define RCAL .027
                                         //R value is 4200000 (4.2M)
                                          //scaled so that result is in
                                          //1/100th of uA
int main(void)
{
    int i;
   int j = 0;
                                          //index for loop
   unsigned int Vread = 0;
   float CTMUISrc, CTMUCap, Vavg, VTot, Vcal;
//assume CTMU and A/D have been setup correctly
//see Example 25-1 for CTMU & A/D setup
setup();
CTMUCONHbits.CTMUEN = 1;
                                         //Enable the CTMU
    for(j=0;j<10;j++)</pre>
    {
        CTMUCONHbits.IDISSEN = 1;
                                         //drain charge on the circuit
        DELAY;
                                         //wait 125us
        CTMUCONHbits.IDISSEN = 0;
                                         //end drain of circuit
        CTMUCONLbits.EDG1STAT = 1;
                                         //Begin charging the circuit
                                         //using CTMU current source
        DELAY;
                                          //wait for 125us
        CTMUCONLbits.EDG1STAT = 0;
                                         //Stop charging circuit
        PIR1bits.ADIF = 0;
                                         //make sure A/D Int not set
        ADCON0bits.GO=1;
                                         //and begin A/D conv.
        while(!PIR1bits.ADIF);
                                         //Wait for A/D convert complete
        Vread = ADRES;
                                         //Get the value from the A/D
        PIR1bits.ADIF = 0;
                                         //Clear A/D Interrupt Flag
        VTot += Vread;
                                         //Add the reading to the total
    }
   Vavg = (float)(VTot/10.000);
                                         //Average of 10 readings
   Vcal = (float)(Vavg/ADSCALE*ADREF);
                                          //CTMUISrc is in 1/100ths of uA
    CTMUISrc = Vcal/RCAL;
    CTMUCap = (CTMUISrc*ETIME/Vcal)/100;
}
```

## 27.5 Measuring Capacitance with the CTMU

There are two ways to measure capacitance with the CTMU. The absolute method measures the actual capacitance value. The relative method only measures for any change in the capacitance.

#### 27.5.1 ABSOLUTE CAPACITANCE MEASUREMENT

For absolute capacitance measurements, both the current and capacitance calibration steps found in **Section 27.4 "Calibrating the CTMU Module**" should be followed.

To perform these measurements:

- 1. Initialize the A/D Converter.
- 2. Initialize the CTMU.
- 3. Set EDG1STAT.
- 4. Wait for a fixed delay, T.
- 5. Clear EDG1STAT.
- 6. Perform an A/D conversion.
- 7. Calculate the total capacitance, CTOTAL = (I \* T)/V, where:
  - I is known from the current source measurement step (Section 27.4.1 "Current Source Calibration")
  - T is a fixed delay
  - V is measured by performing an A/D conversion
- Subtract the stray and A/D capacitance (COFFSET from Section 27.4.2 "Capacitance Calibration") from CTOTAL to determine the measured capacitance.

#### 27.5.2 RELATIVE CHARGE MEASUREMENT

Not all applications require precise capacitance measurements. When detecting a valid press of a capacitance-based switch, only a relative change of capacitance needs to be detected.

In such an application, when the switch is open (or not touched), the total capacitance is the capacitance of the combination of the board traces, the A/D Converter and other elements. A larger voltage will be measured by the A/D Converter. When the switch is closed (or touched), the total capacitance is larger due to the addition of the capacitance of the human body to the above listed capacitances and a smaller voltage will be measured by the A/D Converter.

To detect capacitance changes simply:

- 1. Initialize the A/D Converter and the CTMU.
- 2. Set EDG1STAT.
- 3. Wait for a fixed delay.
- 4. Clear EDG1STAT.
- 5. Perform an A/D conversion.

The voltage measured by performing the A/D conversion is an indication of the relative capacitance. In this case, no calibration of the current source or circuit capacitance measurement is needed. (For a sample software routine for a capacitive touch switch, see Example 27-4.)

#### EXAMPLE 27-4: ROUTINE FOR CAPACITIVE TOUCH SWITCH

```
#include "pl8cxxx.h"
#define COUNT 500
                                         //@ 8MHz = 125uS.
#define DELAY for(i=0;i<COUNT;i++)</pre>
#define OPENSW 1000
                                         //Un-pressed switch value
#define TRIP 300
                                         //Difference between pressed
                                         //and un-pressed switch
#define HYST 65
                                         //amount to change
                                         //from pressed to un-pressed
#define PRESSED 1
#define UNPRESSED 0
int main(void)
ł
   unsigned int Vread;
                                         //storage for reading
   unsigned int switchState;
   int i;
    //assume CTMU and A/D have been setup correctly
    //see Example 25-1 for CTMU & A/D setup
   setup();
   CTMUCONHbits.CTMUEN = 1;
                                         //Enable the CTMU
   CTMUCONHbits.IDISSEN = 1;
                                         //drain charge on the circuit
                                         //wait 125us
   DELAY;
   CTMUCONHbits.IDISSEN = 0;
                                         //end drain of circuit
   CTMUCONLbits.EDG1STAT = 1;
                                         //Begin charging the circuit
                                         //using CTMU current source
                                         //wait for 125us
   DELAY;
   CTMUCONLbits.EDG1STAT = 0;
                                         //Stop charging circuit
   PIR1bits.ADIF = 0;
                                         //make sure A/D Int not set
                                         //and begin A/D conv.
   ADCON0bits.GO=1;
   while(!PIR1bits.ADIF);
                                         //Wait for A/D convert complete
   Vread = ADRES;
                                         //Get the value from the A/D
    if(Vread < OPENSW - TRIP)
    {
       switchState = PRESSED;
   else if(Vread > OPENSW - TRIP + HYST)
    {
       switchState = UNPRESSED;
    }
```

### 27.6 Measuring Time with the CTMU Module

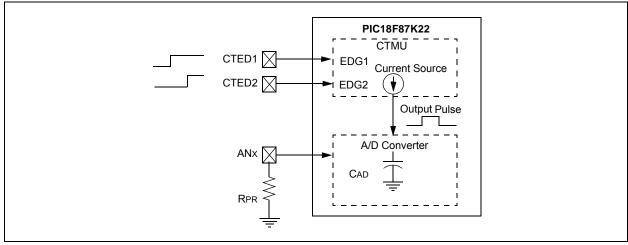
Time can be precisely measured after the ratio  $({\rm C/I})$  is measured from the current and capacitance calibration step. To do that:

- 1. Initialize the A/D Converter and the CTMU.
- 2. Set EDG1STAT.
- 3. Set EDG2STAT.
- 4. Perform an A/D conversion.
- 5. Calculate the time between edges as T = (C/I) \* V, where:
  - I is calculated in the current calibration step (Section 27.4.1 "Current Source Calibration")
  - C is calculated in the capacitance calibration step (Section 27.4.2 "Capacitance Calibration")
  - V is measured by performing the A/D conversion

It is assumed that the time measured is small enough that the capacitance, COFFSET, provides a valid voltage to the A/D Converter. For the smallest time measurement, always set the A/D Channel Select register (AD1CHS) to an unused A/D channel, the corresponding pin for which is not connected to any circuit board trace. This minimizes added stray capacitance, keeping the total circuit capacitance close to that of the A/D Converter itself (25 pF).

To measure longer time intervals, an external capacitor may be connected to an A/D channel and that channel selected whenever making a time measurement.

### FIGURE 27-3: TYPICAL CONNECTIONS AND INTERNAL CONFIGURATION FOR TIME MEASUREMENT



# 27.7 Creating a Delay with the CTMU Module

A unique feature on board the CTMU module is its ability to generate system clock independent output pulses, based on either an external voltage or an external capacitor value. When using an external voltage, this is accomplished using the CTDIN input pin as a trigger for the pulse delay. When using an external capacitor value, this is accomplished using the internal comparator voltage reference module and Comparator 2 input pin.The pulse is output onto the CTPLS pin. To enable this mode, set the TGEN bit.

See Figure 27-4 for an example circuit. When CTMUDS (ODCON3<0>) is cleared, the pulse delay is determined by the output of Comparator 2, and when it is set, the pulse delay is determined by the input of CTDIN. CDELAY is chosen by the user to determine the output pulse width on CTPLS. The pulse width is calculated by T = (CDELAY/I) \* V, where I is known from the current source measurement step (Section 27.4.1 "Current Source Calibration") and V is the Internal Reference Voltage (CVREF).

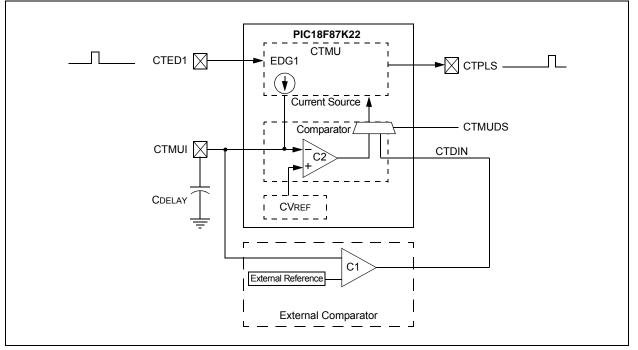
An example use of the external capacitor feature is interfacing with variable capacitive-based sensors, such as a humidity sensor. As the humidity varies, the pulse-width output on CTPLS will vary. An example use of the CTDIN feature is interfacing with a digital sensor. The CTPLS output pin can be connected to an input capture pin and the varying pulse width measured to determine the sensor's output in the application.

To use this feature:

- 1. If CTMUDS is cleared, initialize Comparator 2.
- 2. If CTMUDS is cleared, initialize the comparator voltage reference.
- 3. Initialize the CTMU and enable time delay generation by setting the TGEN bit.
- 4. Set EDG1STAT.

When CTMUDS is cleared, as soon as CDELAY charges to the value of the voltage reference trip point, an output pulse is generated on CTPLS. When CTMUDS is set, as soon as CTDIN is set, an output pulse is generated on CTPLS.

### FIGURE 27-4: TYPICAL CONNECTIONS AND INTERNAL CONFIGURATION FOR PULSE DELAY GENERATION



## 27.8 Measuring Temperature with the CTMU Module

The CTMU, along with an internal diode, can be used to measure the temperature. The ADC can be connected to the internal diode and the CTMU module can source the current to the diode. The ADC reading will reflect the temperature. With the increase, the ADC readings will go low. This can be used for low-cost temperature measurement applications.

#### EXAMPLE 27-5: ROUTINE FOR TEMPERATURE MEASUREMENT USING INTERNAL DIODE

<pre>// Initialize CTMU CTMUICON = 0x03; CTMUCONHbits.CTMUEN = 1; CTMUCONLbits.EDG1STAT = 1;</pre>	
// Initialize ADC	
$ADCON0 = 0 \times E5;$	// Enable ADC and connect to Internal diode
$ADCON1 = 0 \times 00;$	
$ADCON2 = 0 \times BE;$	// Right Justified
ADCON0bits.GO = 1;	// Start conversion
while(ADCON0bits.G0);	
Temp = ADRES;	// Read ADC results (inversely proportional to temperature)

Note: The temperature diode is not calibrated or standardized; the user must calibrate the diode to their application.

#### 27.9 Operation During Sleep/Idle Modes

#### 27.9.1 SLEEP MODE

When the device enters any Sleep mode, the CTMU module current source is always disabled. If the CTMU is performing an operation that depends on the current source when Sleep mode is invoked, the operation may not terminate correctly. Capacitance and time measurements may return erroneous values.

#### 27.9.2 IDLE MODE

The behavior of the CTMU in Idle mode is determined by the CTMUSIDL bit (CTMUCONH<5>). If CTMUSIDL is cleared, the module will continue to operate in Idle mode. If CTMUSIDL is set, the module's current source is disabled when the device enters Idle mode. In this case, if the module is performing an operation when Idle mode is invoked, the results will be similar to those with Sleep mode.

#### 27.10 Effects of a Reset on CTMU

Upon Reset, all registers of the CTMU are cleared. This disables the CTMU module, turns off its current source and returns all configuration options to their default settings. The module needs to be re-initialized following any Reset.

If the CTMU is in the process of taking a measurement at the time of Reset, the measurement will be lost. A partial charge may exist on the circuit that was being measured, which should be properly discharged before the CTMU makes subsequent attempts to make a measurement. The circuit is discharged by setting and clearing the IDISSEN bit (CTMUCONH<1>) while the A/D Converter is connected to the appropriate channel.

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
CTMUCONH	CTMUEN	_	CTMUSIDL	TGEN	EDGEN	EDGSEQEN	IDISSEN	CTTRIG
CTMUCONL	EDG2POL	EDG2SEL1	EDG2SEL0	EDG1POL	EDG1SEL1	EDG1SEL0	EDG2STAT	EDG1STAT
CTMUICON	ITRIM5	ITRIM4	ITRIM3	ITRIM2	ITRIM1	ITRIM0	IRNG1	IRNG0
PIR3	TMR5GIF	_	RC2IF	TX2IF	CTMUIF	CCP2IF	CCP1IF	RTCCIF
PIE3	TMR5GIE	_	RC2IE	TX2IE	CTMUIE	CCP2IE	CCP1IE	RTCCIE
IPR3	TMR5GIP		RC2IP	TX2IP	CTMUIP	CCP2IP	CCP1IP	RTCCIP

#### TABLE 27-1: REGISTERS ASSOCIATED WITH CTMU MODULE

**Legend:** — = unimplemented, read as '0'. Shaded cells are not used during ECCP operation.

# PIC18F87K22 FAMILY

NOTES:

#### 28.0 SPECIAL FEATURES OF THE CPU

The PIC18F87K22 family of devices includes several features intended to maximize reliability and minimize cost through elimination of external components. These include:

- Oscillator Selection
- Resets:
  - Power-on Reset (POR)
  - Power-up Timer (PWRT)
  - Oscillator Start-up Timer (OST)
- Brown-out Reset (BOR)
- Interrupts
- Watchdog Timer (WDT) and On-chip Regulator
- · Fail-Safe Clock Monitor
- Two-Speed Start-up
- Code Protection
- ID Locations
- In-Circuit Serial Programming<sup>™</sup> (ICSP<sup>™</sup>)

The oscillator can be configured for the application depending on frequency, power, accuracy and cost. All of the options are discussed in detail in **Section 3.0 "Oscillator Configurations"**.

A complete discussion of device Resets and interrupts is available in previous sections of this data sheet.

In addition to their Power-up and Oscillator Start-up Timers provided for Resets, the PIC18F87K22 family of devices has a Watchdog Timer, which is either permanently enabled via the Configuration bits or software controlled (if configured as disabled).

The inclusion of an internal RC oscillator (LF-INTOSC) also provides the additional benefits of a Fail-Safe Clock Monitor (FSCM) and Two-Speed Start-up. FSCM provides for background monitoring of the peripheral clock and automatic switchover in the event of its failure. Two-Speed Start-up enables code to be executed almost immediately on start-up, while the primary clock source completes its start-up delays.

All of these features are enabled and configured by setting the appropriate Configuration register bits.

#### 28.1 Configuration Bits

The Configuration bits can be programmed (read as '0') or left unprogrammed (read as '1') to select various device configurations. These bits are mapped starting at program memory location, 300000h.

The user will note that address, 300000h, is beyond the user program memory space. In fact, it belongs to the configuration memory space (300000h-3FFFFh), which can only be accessed using table reads and table writes.

Software programming of the Configuration registers is done in a manner similar to programming the Flash memory. The WR bit in the EECON1 register starts a self-timed write to the Configuration register. In normal Operation mode, a TBLWT instruction, with the TBLPTR pointing to the Configuration register, sets up the address and the data for the Configuration register write. Setting the WR bit starts a long write to the Configuration register. The Configuration registers are written a byte at a time. To write or erase a configuration cell, a TBLWT instruction can write a '1' or a '0' into the cell. For additional details on Flash programming, refer to Section 7.5 "Writing to Flash Program Memory".

File	Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Default/ Unprogrammed Value
300000h	CONFIG1L	_	XINST	_	SOSCSEL1	SOSCSEL0	INTOSCSEL	_	RETEN	-1-1 11
300001h	CONFIG1H	IESO	FCMEN	_	PLLCFG	FOSC3	FOSC2	FOSC1	FOSC0	00-0 1000
300002h	CONFIG2L	-	BORPWR1	BORWPR0	BORV1	BORV0	BOREN1	BOREN0	PWRTEN	-111 1111
300003h	CONFIG2H	_	WDTPS4	WDTPS3	WDTPS2	WDTPS1	WDTPS0	WDTEN1	WDTEN0	-111 1111
300004h	CONFIG3L	WAIT <sup>(2)</sup>	BW <sup>(2)</sup>	ABW1 <sup>(2)</sup>	ABW0 <sup>(2)</sup>	EASHFT <sup>(2)</sup>	_	_	RTCOSC	1
300005h	CONFIG3H	MCLRE	_	_	_	MSSPMSK	—	ECCPMX <sup>(2)</sup>	CCP2MX	1 1-11
300006h	CONFIG4L	DEBUG	-		BBSIZ0	—	—		STVREN	111
300008h	CONFIG5L	CP7 <sup>(1)</sup>	CP6 <sup>(1)</sup>	CP5 <sup>(1)</sup>	CP4 <sup>(1)</sup>	CP3	CP2	CP1	CP0	1111 1111
300009h	CONFIG5H	CPD	CPB	_	—	—	—	_	_	11
30000Ah	CONFIG6L	WRT7 <sup>(1)</sup>	WRT6 <sup>(1)</sup>	WRT5 <sup>(1)</sup>	WRT4 <sup>(1)</sup>	WRT3	WRT2	WRT1	WRT0	1111 1111
30000Bh	CONFIG6H	WRTD	WRTB	WRTC		—	—			111
30000Ch	CONFIG7L	EBTR7 <sup>(1)</sup>	EBTR6 <sup>(1)</sup>	EBTR5 <sup>(1)</sup>	EBTR4 <sup>(1)</sup>	EBTR3	EBTR2	EBTR1	EBTR0	1111 1111
30000Dh	CONFIG7H	—	EBTRB	_	_	_	_	_	_	-1
3FFFFEh	DEVID1 <sup>(3)</sup>	DEV2	DEV1	DEV0	REV4	REV3	REV2	REV1	REV0	XXXX XXXX
3FFFFFh	DEVID2 <sup>(3)</sup>	DEV10	DEV9	DEV8	DEV7	DEV6	DEV5	DEV4	DEV3	xxxx xxxx

#### TABLE 28-1: CONFIGURATION BITS AND DEVICE IDs

Legend:x = unknown, u = unchanged, - = unimplemented, q = value depends on condition. Shaded cells are unimplemented, read as '0'.Note1:Implemented only on the PIC18F67K22 and PIC18F87K22 devices.

2: Implemented only on the 80-pin devices (PIC18F8XK22).

3: See Register 28-14 for DEVID1 values. DEVID registers are read-only and cannot be programmed by the user.

#### REGISTER 28-1: CONFIG1L: CONFIGURATION REGISTER 1 LOW (BYTE ADDRESS 300000h)

U-0	R/P-1	U-0	R/P-1	R/P-1	R/P-1	U-0	R/P-1
—	XINST	_	SOSCSEL1	SOSCSEL0	INTOSCSEL	_	RETEN
bit 7							bit 0

Legend:	P = Programmable bit		
R = Readable bit	W = Writable bit	U = Unimplemented bit	, read as '0'
-n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown

bit 7	Unimplemented: Read as '0'
bit 6	XINST: Extended Instruction Set Enable bit
	<ul> <li>1 = Instruction set extension and Indexed Addressing mode are enabled</li> <li>0 = Instruction set extension and Indexed Addressing mode are disabled (Legacy mode)</li> </ul>
bit 5	Unimplemented: Read as '0'
bit 4-3	SOSCSEL<1:0>: SOSC Power Selection and Mode Configuration bits
	<ul> <li>11 = High-power SOSC circuit is selected</li> <li>10 = Digital (SCLKI) mode; I/O port functionality of RC0 and RC1 is enabled</li> <li>01 = Low-power SOSC circuit is selected</li> <li>00 = Reserved</li> </ul>
bit 2	INTOSCSEL: LF-INTOSC Low-power Enable bit
	<ul> <li>1 = LF-INTOSC is in High-Power mode during Sleep</li> <li>0 = LF-INTOSC is in Low-Power mode during Sleep</li> </ul>
bit 1	Unimplemented: Read as '0'
bit 0	RETEN: VREG Sleep Enable bit
	1 = Regulator power while in Sleep mode is controlled by VREGSLP (WDTCON<7>) 0 = Regulator power while in Sleep mode is controlled by SRETEN (WDTCON<4>). Ultra low-pow

= Regulator power while in Sleep mode is controlled by SRETEN (WDTCON<4>). Ultra low-power regulator is enabled.

# PIC18F87K22 FAMILY

					•					
R/P-0	R/P-0	U-0	U-0	R/P-1	R/P-0	R/P-0	R/P-0			
IESO	FCMEN	_	PLLCFG <sup>(1)</sup>	FOSC3 <sup>(2)</sup>	FOSC2 <sup>(2)</sup>	FOSC1 <sup>(2)</sup>	FOSC0 <sup>(2)</sup>			
bit 7							bit (			
Legend:		P = Progran	nmable bit							
R = Readabl	le bit	W = Writable	e bit	U = Unimpler	nented bit, read	1 as '0'				
-n = Value at	t POR	'1' = Bit is se	et	'0' = Bit is cle	ared	x = Bit is unkr	nown			
bit 7			cillator Switchov	ver bit						
		ed Start-up is ed Start-up is								
bit 6		I-Safe Clock M Clock Monito	lonitor Enable b r is enabled	bit						
		Clock Monito								
bit 5	-	nted: Read as								
bit 4	PLLCFG: 4x PLL Enable bit <sup>(1)</sup>									
	<ul> <li>1 = Oscillator is multiplied by 4</li> <li>0 = Oscillator is used directly</li> </ul>									
bit 3-0	FOSC<3:0>: Oscillator Selection bits <sup>(2)</sup>									
	1100 = EC1 1011 = EC2 1010 = EC2 1001 = INTIG 1000 = INTIG 0111 = RC, 0110 = RCIG 0101 = EC3 0100 = EC3 0011 = HS1	IO, EC oscillat , EC oscillator IO, EC oscillat O1, internal RC O2, internal RC external RC os O, external RC , EC oscillator IO, EC oscillator , HS oscillator		T function on F er, 160 kHz-16 T function on F CLKOUT funct CKLOUT funct MHz-64 MHz) T function on F er, 4 MHz-16 N	MHz) RA6 (medium p ition on RA6 ion on RA6 RA6 (high powe IHz)	oower, DC-160	kHz-16 MHz			
	0000 <b>=</b> LP o	scillator								
	lot valid for the II						o opoblod b			
Z: IN		se enabled on	iv ov me PLLEN		ENDZI UMEĽ P	r i modes cañ r	je enabled b			

#### REGISTER 28-2: CONFIG1H: CONFIGURATION REGISTER 1 HIGH (BYTE ADDRESS 300001h)

2: INTIO + PLL can be enabled only by the PLLEN bit (OSCTUNE<6>). Other PLL modes can be enabled by either the PLLEN bit or the PLLCFG (CONFIG1H<4>) bit.

#### REGISTER 28-3: CONFIG2L: CONFIGURATION REGISTER 2 LOW (BYTE ADDRESS 300002h)

U-0	R/P-1	R/P-1	R/P-1	R/P-1	R/P-1	R/P-1	R/P-1
_	BORPWR1 <sup>(1)</sup>	BORPWR0 <sup>(1)</sup>	BORV1 <sup>(1)</sup>	BORV0 <sup>(1)</sup>	BOREN1 <sup>(2)</sup>	BOREN0 <sup>(2)</sup>	PWRTEN <sup>(2)</sup>
bit 7							bit 0

Legend:	P = Programmable bit		
R = Readable bit	W = Writable bit	U = Unimplemented bit, rea	id as '0'
-n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown

bit 7	Unimplemented: Read as '0'
bit 6-5	BORPWR<1:0>: BORMV Power-Level bits <sup>(1)</sup>
	11 = ZPBORVMV instead of BORMV is selected
	10 = BORMV is set to a high-power level
	01 = BORMV is set to a medium power level 00 = BORMV is set to a low-power level
bit 4-3	BORV<1:0>: Brown-out Reset Voltage bits <sup>(1)</sup>
	11 = VBORMV is set to 1.8V
	10 = VBORMV is set to 2.0V
	01 = VBORMV is set to 2.7V
	00 = VBORMV is set to 3.0V
bit 2-1	BOREN<1:0>: Brown-out Reset Enable bits <sup>(2)</sup>
	11 = Brown-out Reset is enabled in hardware only (SBOREN is disabled)
	10 = Brown-out Reset is enabled in hardware only and disabled in Sleep mode (SBOREN is disabled)
	<ul> <li>01 = Brown-out Reset is enabled and controlled by software (SBOREN is enabled)</li> <li>00 = Brown-out Reset is disabled in hardware and software</li> </ul>
<b>h</b> # 0	
bit 0	PWRTEN: Power-up Timer Enable bit <sup>(2)</sup>
	1 = PWRT is disabled 0 = PWRT is enabled
Note 1:	For the specifications, see Section 31.1 "DC Characteristics: Supply Voltage PIC18F87K22 Family (Industrial/Extended)".

2: The Power-up Timer is decoupled from Brown-out Reset, allowing these features to be independently controlled.

# PIC18F87K22 FAMILY

U-0	R/P-1	R/P-1	R/P-1	R/P-1	R/P-1	R/P-1	R/P-1						
_	WDTPS4	WDTPS3	WDTPS2	WDTPS1	WDTPS0	WDTEN1	WDTEN0						
bit 7			·				bit (						
Legend:		P = Programn	nable bit										
R = Readab	le bit	W = Writable	bit	U = Unimpler	nented bit, read	d as '0'							
-n = Value at POR		'1' = Bit is set	:	'0' = Bit is cle	ared	x = Bit is unkr	nown						
bit 7	Unimplemer	ted: Read as '	0'										
bit 6-2	WDTPS<4:0>: Watchdog Timer Postscale Select bits												
	10101-1111	1 = Reserved											
	10100 = 1:1,	048,576 (4,194	l.304s)										
		24,288 (2,097.1											
		52,144 (1,048.5											
		10001 = 1:131,072 (524.288s)											
	10000 = 1:65,536 (262.144s) 01111 = 1:32,768 (131.072s)												
		01111 = 1:32,768 (131.072s) 01110 = 1:16,384 (65.536s)											
		01110 = 1.10,304 (05.3308) 01101 = 1.8,192 (32.768s)											
		00 = 1:4,096 (16.384s)											
		: 1:2,048 (8.192s)											
	01010 = 1:1,	024 (4.096s)											
	01001 = 1:5	• •											
	01000 = 1:25												
	00111 = 1:12												
	00110 = 1:64 00101 = 1:32	. ,											
	00100 = 1:16 (64 ms) 00011 = 1:8 (32 ms)												
	00010 = 1.4 (16  ms)												
	00001 = 1.2 (8  ms)												
	00000 = 1:1 (4  ms)												
bit 1-0	WDTEN<1:0	>: Watchdog Ti	mer Enable bi	ts									
	11 = WDT is	s enabled in ha	rdware; SWD	TEN bit is disat	bled								
		s controlled by											
			while the dev	rice is active a	nd disabled in	Sleep mode; S	WDTEN bit i						
	disable	ed											
		s disabled in ha											

### REGISTER 28-4: CONFIG2H: CONFIGURATION REGISTER 2 HIGH (BYTE ADDRESS 300003h)

#### REGISTER 28-5: CONFIG3L: CONFIGURATION REGISTER 3 LOW (BYTE ADDRESS 300004h)

U-1	U-1	U-1	U-1	U-1	U-0	U-0	R/P-1
WAIT <sup>(1)</sup>	BW <sup>(1)</sup>	ABW1 <sup>(1)</sup>	ABW0 <sup>(1)</sup>	EASHFT <sup>(1)</sup>	—	—	RTCOSC
bit 7							bit 0

Legend:	P = Programmable bit		
R = Readable bit	W = Writable bit	U = Unimplemented bit	t, read as '0'
-n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown

bit 7	WAIT: External Bus Wait Enable bit <sup>(1)</sup> 1 = Wait states on the external bus are disabled 0 = Wait states on the external bus are enabled and selected by MEMCON <5:4>
bit 6	<b>BW:</b> Data Bus Width Select bit <sup>(1)</sup> 1 = 16-Bit Data Width modes 0 = 8-Bit Data Width modes
bit 5-4	ABW<1:0>: External Memory Bus Configuration bits <sup>(1)</sup> 11 = 8-Bit Address mode (Microcontroller mode) 10 = 12-Bit Address mode 01 = 16-Bit Address mode 00 = 20-Bit Address mode
bit 3	<b>EASHFT:</b> External Address Bus Shift Enable bit <sup>(1)</sup> 1 = Address shifting is enabled; external address is shifted to start at 000000h 0 = Address shifting is disabled; external address bus reflects the PC value
bit 2-1	Unimplemented: Read as '0'
bit 0	<b>RTCOSC:</b> RTCC Reference Clock Select bit 1 = RTCC uses SOSC as the reference clock 0 = RTCC uses LF-INTOSC as the reference clock

Note 1: Unimplemented on 64-pin devices (PIC18F6XK22), read as '0'.

# PIC18F87K22 FAMILY

R/P-1	U-0	U-0	U-0	R/P-1	U-0	R/P-1	R/P-1
MCLRE	—	_	_	MSSPMSK		ECCPMX <sup>(1)</sup>	CCP2MX
bit 7							bit (
Logondy		D - Drogramm	abla bit				
Legend: R = Readable b	.i+	P = Programm W = Writable t			nonted hit rea	ad aa 'O'	
-n = Value at P		'1' = Bit is set	Л	U = Unimplen '0' = Bit is clea		x = Bit is unkn	0.00
	JK	I = BILIS SEL			areu		IOWI1
bit 7	MCLRE: MCI	R Pin Enable b	oit				
	$1 = \overline{\text{MCLR}}$ pir	is enabled; RC	65 input pin i	is disabled			
	0 = RG5 inpu	t pin is enabled	; MCLR is di	sabled			
bit 6-4	Unimplemen	ted: Read as '0	)'				
bit 3	MSSPMSK:	MSSP V3 7-Bit	Address Ma	sking Mode Ena	ble bit		
		ress Masking m					
	0 = 5-Bit Add	ress Masking m	ode is enab	led			
bit 2	•	ted: Read as '0	)'				
bit 1	ECCPMX: EC	CCP MUX bit <sup>(1)</sup>					
	1 =						
						RE6, and CCP7	
	- ECCP3 (	P3B/P3C) is mi	litiplexed on	to RE4 and RE3	3, CCP8 onto	RE4, and CCP9	onto RE3
	-	P1B/P1C) is mu	ultiplexed on	to RH7 and RH	6, CCP6 onto	RH7, and CCP7	<sup>(2)</sup> onto RH6
	- ECCP3 (	P3B/P3C) is mu	ultiplexed on	to RH5 and RH4	4, CCP8 onto	RH5, and CCP9	<sup>(2)</sup> onto RH4
bit 0	CCP2MX: EC	CP2 MUX bit					
		s multiplexed w					
		s multiplexed w licrocontroller m		Extended Microo	controller mod	de; ECCP2 is mu	ultiplexed wit
Note 1: Unin	plemented or	n 64-pin devices	(PIC18F6X	K22), read as '0	)'.		

#### REGISTER 28-6: CONFIG3H: CONFIGURATION REGISTER 3 HIGH (BYTE ADDRESS 300005h)

2: Not implemented on 32K devices (PIC18F65K22 and PIC18F85K22).

#### REGISTER 28-7: CONFIG4L: CONFIGURATION REGISTER 4 LOW (BYTE ADDRESS 300006h)

R/P-1	U-0	U-0	R/P-0	U-0	U-0	U-0	R/P-1
DEBUG		_	BBSIZ0	_	_	_	STVREN
bit 7							bit 0

Legend:         P = Programmable bit					
R = Readable bit	W = Writable bit	U = Unimplemented bit	t, read as '0'		
-n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown		

bit 7	<b>DEBUG:</b> Background Debugger Enable bit 1 = Background debugger is disabled, RB6 and RB7 are configured as general purpose I/O pins 0 = Background debugger is enabled, RB6 and RB7 are dedicated to In-Circuit Debug
bit 6-5	Unimplemented: Read as '0'
bit 4	BBSIZ0: Boot Block Size Select bit
	1 = 2 kW boot block size 0 = 1 kW boot block size
bit 3-1	Unimplemented: Read as '0'
bit 0	STVREN: Stack Full/Underflow Reset Enable bit
	1 = Stack full/underflow will cause Reset 0 = Stack full/underflow will not cause Reset

# PIC18F87K22 FAMILY

R/C-1	R/C-1	R/C-1	R/C-1	R/C-1	R/C-1	R/C-1	R/C-1
CP7 <sup>(1)</sup>	CP6 <sup>(1)</sup>	CP5 <sup>(1)</sup>	CP4 <sup>(1)</sup>	CP3	CP2	CP1	CP0
bit 7							bit
Legend:		C = Clearable	bit				
R = Readable	e bit	W = Writable	bit	U = Unimpler	nented bit, rea	ad as '0'	
-n = Value at	POR	'1' = Bit is set		'0' = Bit is cle	ared	x = Bit is unkr	nown
bit 7	CP7: Code F	Protection bit <sup>(1)</sup>					
		s not code-prote s code-protecte					
bit 6	CP6: Code F	Protection bit <sup>(1)</sup>					
		s not code-prote					
		s code-protecte	d <sup>(2)</sup>				
bit 5		Protection bit <sup>(1)</sup>	(0)				
	1 = Block 5 i	s not code-prote s code-protecte	ected <sup>(2)</sup>				
<b>b</b> :+ 4		Protection bit <sup>(1)</sup>	a,				
bit 4		s not code-prote	notod <sup>(2)</sup>				
		s code-protecte					
bit 3		Protection bit					
	1 = Block 3 i	s not code-prote	ected <sup>(2)</sup>				
	0 = Block 3 i	s code-protecte	d <sup>(2)</sup>				
bit 2	CP2: Code F	Protection bit					
		s not code-prote					
		s code-protecte	d( <u>-</u> )				
bit 1		Protection bit	+ ( <b>2</b> )				
		s not code-prote s code-protecte					
bit 0	CP0: Code F	•	-				
		s not code-prote	ected <sup>(2)</sup>				
		s code-protecte					

#### REGISTER 28-8: CONFIG5L: CONFIGURATION REGISTER 5 LOW (BYTE ADDRESS 300008h)

- This bit is available only on PIC18F67K22 and PIC18F87K22 devices. 1:
  - 2: For the memory size of the blocks, see Figure 28-6.

#### REGISTER 28-9: CONFIG5H: CONFIGURATION REGISTER 5 HIGH (BYTE ADDRESS 300009h)

R/C-1	R/C-1	U-0	U-0	U-0	U-0	U-0	U-0
CPD	CPB	—	_	—	—	—	—
bit 7				•			bit 0
Legend:		C = Clearable	bit				
R = Readabl	e bit	W = Writable	bit	U = Unimpler	mented bit, read	l as '0'	
-n = Value at POR '1' = Bit is set			'0' = Bit is cle	ared	x = Bit is unkr	nown	
bit 7	CPD: Data E	EPROM Code I	Protection bit				
		PROM is not co	•				
	∩ = Data FE	PROM is code_r	notactad				

- 0 = Data EEPROM is code-protected
- bit 6 **CPB:** Boot Block Code Protection bit 1 = Boot block is not code-protected<sup>(1)</sup> 0 = Boot block is code-protected<sup>(1)</sup>
- bit 5-0 Unimplemented: Read as '0'

Note 1: For the memory size of the blocks, see Figure 28-6. The boot block size changes with BBSIZO.

# PIC18F87K22 FAMILY

R/C-1	R/C-1	R/C-1	R/C-1	R/C-1	R/C-1	R/C-1	R/C-1
WRT7 <sup>(1)</sup>	WRT6 <sup>(1)</sup>	WRT5 <sup>(1)</sup>	WRT4 <sup>(1)</sup>	WRT3	WRT2	WRT1	WRT0
bit 7							bit
Logondy		C = Clearable	hit				
Legend: R = Readable	L:4	W = Writable			nanted bit was		
					nented bit, rea		
-n = Value at F	VOR	'1' = Bit is set		'0' = Bit is cle	ared	x = Bit is unkr	lown
bit 7	WRT7: Write	Protection bit <sup>(1</sup>	)				
		not write-prote					
	0 = Block 7 is	write-protecte	d <sup>(2)</sup>				
bit 6		Protection bit <sup>(1</sup>					
		not write-prote					
		write-protecte					
bit 5		Protection bit <sup>(1</sup>					
		not write-prote write-protecte					
bit 4		Protection bit <sup>(1</sup>					
DIL 4		not write-prote					
	1 = Block 4 is 0 = Block 4 is	write-protecte					
bit 3		Protection bit					
		not write-prote	ected <sup>(2)</sup>				
		write-protecte					
bit 2	WRT2: Write	Protection bit					
		not write-prote					
		write-protecte	d <sup>(2)</sup>				
bit 1	WRT1: Write						
		not write-prote					
		write-protecte	d( <u>~</u> )				
bit 0		Protection bit	(2)				
		not write-prote					
	0 = BIOCK U IS	write-protecte	uʻ '				

#### REGISTER 28-10: CONFIG6L: CONFIGURATION REGISTER 6 LOW (BYTE ADDRESS 30000Ah)

- Note 1: This bit is available only on PIC18F67K22 and PIC18F87K22 devices.
  - 2: For the memory size of the blocks, see Figure 28-6.

#### REGISTER 28-11: CONFIG6H: CONFIGURATION REGISTER 6 HIGH (BYTE ADDRESS 30000Bh)

					•		,
R/C-1	R/C-1	R-1	U-0	U-0	U-0	U-0	U-0
WRTD	WRTB	WRTC <sup>(1)</sup>	_	_	_	_	_
bit 7	•						bit 0
Legend:		C = Clearable	bit				
R = Readable bit W = Writable bit				U = Unimpler	mented bit, read	l as '0'	
-n = Value at POR '1' = Bit is set			'0' = Bit is cleared x = Bit is unk			nown	
bit 7	WRTD: Data	EEPROM Write	Protection b	it			
	1 = Data EEI	PROM is not wri	te-protected				
	0 <b>= Data EEI</b>	PROM is write-p	rotected				
bit 6	WRTB: Boot	Block Write Pro	tection bit				
	1 = Boot bloc	ck is not write-pr	otected <sup>(2)</sup>				
	0 = Boot bloc	ck is write-proted	cted <sup>(2)</sup>				
bit 5	WRTC: Conf	iguration Regist	er Write Prote	ection bit <sup>(1)</sup>			
		ation registers a	ro not writo n	rotoctod <sup>(2)</sup>			

- 1 = Configuration registers are not write-protected<sup>(2)</sup>
   0 = Configuration registers are write-protected<sup>(2)</sup>
- bit 4-0 **Unimplemented:** Read as '0'

**Note 1:** This bit is read-only in normal Execution mode; it can be written only in Program mode.

2: For the memory size of the blocks, see Figure 28-6.

					``		,
R/C-1	R/C-1	R/C-1	R/C-1	R/C-1	R/C-1	R/C-1	R/C-1
EBTR7 <sup>(1,3</sup>	<sup>3)</sup> EBTR6 <sup>(1,3)</sup>	EBTR5 <sup>(1,3)</sup>	EBTR4 <sup>(1,3)</sup>	EBTR3 <sup>(3)</sup>	EBTR2 <sup>(3)</sup>	EBTR1 <sup>(3)</sup>	EBTR0 <sup>(3)</sup>
bit 7							bit 0
Legend:		C = Clearable	bit				
R = Readal	ole bit	W = Writable	bit	U = Unimpler	nented bit, read	l as '0'	
-n = Value a	at POR	'1' = Bit is set		'0' = Bit is cle		x = Bit is unkn	own
			(4.2)				
bit 7		e Read Protect			(2)		
		not protected protected fron			other blocks <sup>(2)</sup> er blocks <sup>(2)</sup>		
bit 6	EBTR6: Table	e Read Protect	ion bit <sup>(1,3)</sup>				
					other blocks <sup>(2)</sup>		
		protected fron		executed in oth	er blocks <sup>(2)</sup>		
bit 5		e Read Protect			(2)		
		not protected protected fron			other blocks <sup>(2)</sup> er blocks <sup>(2)</sup>		
bit 4	EBTR4: Table	e Read Protect	ion bit <sup>(1,3)</sup>				
		not protected protected fron			other blocks <sup>(2)</sup> er blocks <sup>(2)</sup>		
bit 3		e Read Protect					
				ds executed in	other blocks <sup>(2)</sup>		
		protected fron		executed in oth	er blocks <sup>(2)</sup>		
bit 2	EBTR2: Table	e Read Protect	ion bit <sup>(3)</sup>				
		not protected protected fron			other blocks <sup>(2)</sup> er blocks <sup>(2)</sup>		
bit 1	EBTR1: Table	e Read Protect	ion bit <sup>(3)</sup>				
	1 = Block 1 is 0 = Block 1 is	not protected protected fron	from table rea n table reads e	ds executed in executed in	other blocks <sup>(2)</sup> er blocks <sup>(2)</sup>		
bit 0	EBTR0: Table	e Read Protect	ion bit <sup>(3)</sup>				
		s not protected s protected fror			n other blocks <sup>(2</sup> ner blocks <sup>(2)</sup>	)	
Note 1:	This bit is available	e only on PIC1	BF67K22 and	PIC18F87K22	devices.		
2: 1	For the memory si	ze of the block	s, see Figure 2	28-6			

#### REGISTER 28-12: CONFIG7L: CONFIGURATION REGISTER 7 LOW (BYTE ADDRESS 30000Ch)

- **2:** For the memory size of the blocks, see Figure 28-6
- 3: Enable the corresponding CPx bit to protect the block from external read operations.

#### REGISTER 28-13: CONFIG7H: CONFIGURATION REGISTER 7 HIGH (BYTE ADDRESS 30000Dh)

U-0	R/C-1	U-0	U-0	U-0	U-0	U-0	U-0
—	EBTRB <sup>(2)</sup>	—	—	—	—	—	—
bit 7							bit 0

Legend:	C = Clearable bit		
R = Readable bit	W = Writable bit	U = Unimplemented bit, rea	id as '0'
-n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown

bit 7 Unimplemented: Read as '0	0'
---------------------------------	----

bit 6 **EBTRB:** Boot Block Table Read Protection bit<sup>(2)</sup>

1 = Boot block is not protected from table reads executed in other blocks<sup>(1)</sup>

0 = Boot block is protected from table reads executed in other blocks<sup>(1)</sup>

bit 5-0 Unimplemented: Read as '0'

**Note 1:** For the memory size of the blocks, see Figure 28-6.

2: Enable the corresponding CPx bit to protect the block from external read operations.

#### REGISTER 28-14: DEVID1: DEVICE ID REGISTER 1 FOR THE PIC18F87K22 FAMILY

Р	R	Р	R	D	R	D	R	
R	1	R		R		R		
DEV2	DEV1	DEV0	REV4	REV3	REV2	REV1	REV0	
bit 7							bit 0	
Legend:								
R = Readable	bit	W = Writable	bit	U = Unimpler	nented bit, read	l as '0'		
-n = Value at F	-n = Value at POR '1' = Bit is set				'0' = Bit is cleared x = Bit is unknown			
bit 7-5 <b>DEV&lt;2:0&gt;:</b> Device ID bits <u>Devices with DEV&lt;10:3&gt; of '0101_0010</u> 010 = PIC18F65K22 000 = PIC18F66K22 101 = PIC18F85K22 011 = PIC18F86K22 <u>Devices with DEV&lt;10:3&gt; of '0101_0001</u> 000 = PIC18F67K22 010 = PIC18F87K22								
bit 4-0	REV<4:0>: Revision ID bits							

These bits are used to indicate the device revision.

#### REGISTER 28-15: DEVID2: DEVICE ID REGISTER 2 FOR THE PIC18F87K22 FAMILY

R	R	R	R	R	R	R	R	
DEV10 <sup>(1)</sup>	DEV9 <sup>(1)</sup>	DEV8 <sup>(1)</sup>	DEV7 <sup>(1)</sup>	DEV6 <sup>(1)</sup>	DEV5 <sup>(1)</sup>	DEV4 <sup>(1)</sup>	DEV3 <sup>(1)</sup>	
bit 7		•					bit 0	
Legend:								
R = Readable b	oit	W = Writable	bit	U = Unimpler	nented bit, read	as '0'		
-n = Value at P	-n = Value at POR '1' = Bit is set			'0' = Bit is cleared x = Bit is unknown				

bit 7-0 **DEV<10:3>:** Device ID bits<sup>(1)</sup> These bits are used with the DEV<2:0> bits in the Device ID Register 1 to identify the part number. 0101 0010 = PIC18F65K22, PIC18F66K22, PIC18F85K22 and PIC18F86K22 0101 0001 = PIC18F67K22 and PIC18F87K22

**Note 1:** These values for DEV<10:3> may be shared with other devices. The specific device is always identified by using the entire DEV<10:0> bit sequence.

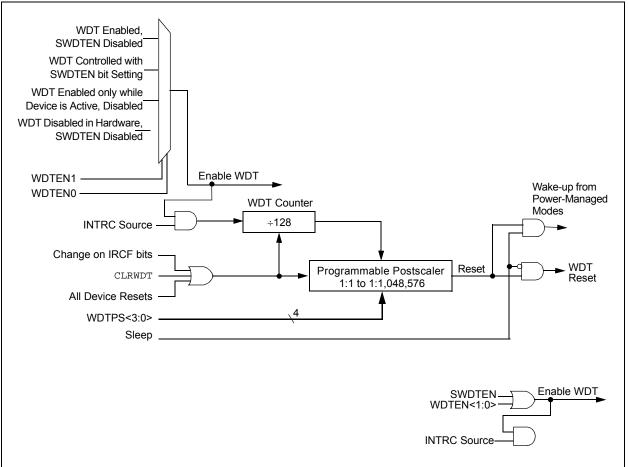
#### 28.2 Watchdog Timer (WDT)

For the PIC18F87K22 family of devices, the WDT is driven by the LF-INTOSC source. When the WDT is enabled, the clock source is also enabled. The nominal WDT period is 4 ms and has the same stability as the LF-INTOSC oscillator.

The 4 ms period of the WDT is multiplied by a 16-bit postscaler. Any output of the WDT postscaler is selected by a multiplexer, controlled by bits in Configuration Register 2H. Available periods range from 4 ms to 4,194 seconds (about one hour). The WDT and postscaler are cleared when any of the following events occur: a SLEEP or CLRWDT instruction is executed, the IRCF bits (OSCCON<6:4>) are changed or a clock failure has occurred.

The WDT can be operated in one of four modes as determined by the WDTEN<1:0> (CONFIG2H<1:0> bits. The four modes are:

- WDT Enabled
- · WDT Disabled
- WDT under Software Control, SWDTEN (WDTCON<0>)
- WDT
  - Enabled during normal operation
  - Disabled during Sleep
  - Note 1: The CLRWDT and SLEEP instructions clear the WDT and postscaler counts when executed.
    - Changing the setting of the IRCF bits (OSCCON<6:4>) clears the WDT and postscaler counts.
    - **3:** When a CLRWDT instruction is executed, the postscaler count will be cleared.



#### FIGURE 28-1: WDT BLOCK DIAGRAM

#### 28.2.1 CONTROL REGISTER

Register 28-16 shows the WDTCON register. This is a readable and writable register which contains a control bit that allows software to override the WDT Enable Configuration bit, but only if the Configuration bit has disabled the WDT.

#### REGISTER 28-16: WDTCON: WATCHDOG TIMER CONTROL REGISTER

R/W-0	U-0	R-x	R/W-0	U-0	R/W-0	R/W-0	R/W-0
REGSLP		ULPLVL	SRETEN <sup>(2)</sup>	—	ULPEN	ULPSINK	SWDTEN <sup>(1)</sup>
bit 7							bit (
Legend:							
R = Readabl	e bit	W = Writable	bit	U = Unimpler	mented bit, read	d as '0'	
-n = Value at	POR	'1' = Bit is set	t	'0' = Bit is cle	ared	x = Bit is unk	nown
bit 7		Regulator Voltage	-				
		or goes into Low or stays in norm					
bit 6	Unimpleme	ented: Read as '	0'				
bit 5	ULPLVL: U	Itra Low-Power	Wake-up Outpu	ut bit			
	Not valid un	Iless ULPEN = 1					
		e on RA0 pin > ~					
	•	e on RA0 pin < ~					
bit 4		Regulator Voltage	-				
	1 = If RETE mode ii	EN (CONFIG1L<	0>) = 0 and the	regulator is e	nabled, the dev	ice goes into Ul	tra Low-Powe
	0 = The re	gulator is on wl	nen the device	e's Sleep moo	le is enabled a	and the Low-P	ower mode is
bit 3	Unimpleme	ented: Read as '	0'				
bit 2	ULPEN: Ult	ra Low-Power V	/ake-up Modul	e Enable bit			
		ow-Power Wake ow-Power Wake			VL bit indicates	s the comparate	or output
bit 1		Ultra Low-Power	-	ent Sink Enab	ole bit		
	Not valid un	less ULPEN = 1					
		ow-Power Wake ow-Power Wake					
bit 0	SWDTEN: S	Software Control	lled Watchdog	Timer Enable	bit <sup>(1)</sup>		
		og Timer is on					
		og Timer is off					

**Note 1:** This bit has no effect if the Configuration bits, WDTEN<1:0>, are enabled.

**2:** This bit is available only when ENVREG = 1 and  $\overline{\text{RETEN}}$  = 0.

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
RCON	IPEN	SBOREN	СМ	RI	TO	PD	POR	BOR
WDTCON	REGSLP		ULPLVL	SRETEN	_	ULPEN	ULPSINK	SWDTEN

**Legend:** — = unimplemented, read as '0'. Shaded cells are not used by the Watchdog Timer.

#### 28.3 On-Chip Voltage Regulator

All of the PIC18F87K22 family devices power their core digital logic at a nominal 3.3V. For designs that are required to operate at a higher typical voltage, such as 5V, all family devices incorporate two on-chip regulators that allow the device to run its core logic from VDD. Those regulators are:

- Normal On-Chip Regulator
- Ultra Low-Power On-Chip Regulator

The hardware configuration of these regulators is the same and is explained in Section 28.3.1 "Regulator Enable/Disable By Hardware". The regulators' only differences relate to when the device enters Sleep, as explained in Section 28.3.2 "Operation of Regulator in Sleep".

#### 28.3.1 REGULATOR ENABLE/DISABLE BY HARDWARE

The regulator can be enabled or disabled only by hardware. The regulator is controlled by the ENVREG pin and the VDDCORE/VCAP pin.

#### 28.3.1.1 Regulator Enable Mode

Tying VDD to the pin enables the regulator, which in turn, provides power to the core from the other VDD pins.

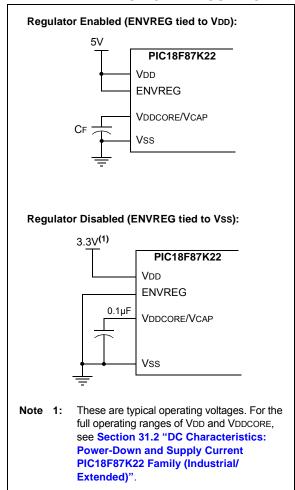
When the regulator is enabled, a low-ESR filter capacitor must be connected to the VDDCORE/VCAP pin (see Figure 28-2). This helps maintain the regulator's stability. The recommended value for the filter capacitor is given in Section 31.2 "DC Characteristics: Power-Down and Supply Current PIC18F87K22 Family (Industrial/Extended)".

#### 28.3.1.2 Regulator Disable Mode

If the regulator is disabled by connecting Vss to the ENVREG pin, the power to the core is supplied directly by VDD. The voltage levels for VDD must not exceed the specified VDDCORE levels. In Regulator Disabled mode, a 0.1  $\mu$ F capacitor should be connected to the VDDCORE/VCAP pin (see Figure 28-2).

When the regulator is being used, the overall voltage budget is very tight. The regulator should operate the device down to 1.8V. When VDD drops below 3.3V, the regulator no longer regulates, but the output voltage follows the input until VDD reaches 1.8V. Below this voltage, the output of the regulator output may drop to 0V.

### FIGURE 28-2: CONNECTIONS FOR THE ON-CHIP REGULATOR



#### 28.3.2 OPERATION OF REGULATOR IN SLEEP

The difference in the two regulators' operation arises with Sleep mode. The ultra low-power regulator gives the device the lowest current in the Regulator Enabled mode.

The on-chip regulator can go into a lower power mode, when the device goes to Sleep, by setting the REGSLP bit (WDTCON<7>). This puts the regulator in a Standby mode so that the device consumes much less current.

The on-chip regulator can also go into the Ultra Low-Power mode, which consumes the lowest current possible with the regulator enabled. This mode is controlled by the RETEN bit (CONFIG1L<0>) and SRETEN bit (WDTCON<4>). The various modes of regulator operation are shown in Table 28-3.

When the ultra low-power regulator is in Sleep mode, the internal reference voltages in the chip will be shut off and any interrupts referring to the internal reference will not wake up the device. If the BOR or LVD is enabled, the regulator will keep the internal references on and the lowest possible current will not be achieved.

When using the ultra low-power regulator in Sleep mode, the device will take about 250  $\mu$ s, typical, to start executing the code after it wakes up.

Power Mode	VREGSLP WDTCON<7>	SRETEN WDTCON<4>	RETEN CONFIG1L<0>					
Normal Operation (Sleep)	0	х	1					
Low-Power mode (Sleep)	1	x	1					
Normal Operation (Sleep)	0	0	х					
Low-Power mode (Sleep)	1	0	x					
Ultra Low-Power mode (Sleep)	x	1	0					
	Power Mode         Normal Operation (Sleep)         Low-Power mode (Sleep)         Normal Operation (Sleep)         Low-Power mode (Sleep)         Low-Power mode (Sleep)	Power ModeWDTCON<7>Normal Operation (Sleep)0Low-Power mode (Sleep)1Normal Operation (Sleep)0Low-Power mode (Sleep)1	Power ModeVREGSLP WDTCON<7>SRETEN WDTCON<4>Normal Operation (Sleep)0×Low-Power mode (Sleep)1×Normal Operation (Sleep)00Low-Power mode (Sleep)10					

#### TABLE 28-3: SLEEP MODE REGULATOR SETTINGS<sup>(1)</sup>

**Note 1:** x = Indicates that VIT status is invalid.

#### 28.4 Two-Speed Start-up

The Two-Speed Start-up feature helps to minimize the latency period, from oscillator start-up to code execution, by allowing the microcontroller to use the INTOSC (LF-INTOSC, MF-INTOSC, HF-INTOSC) oscillator as a clock source until the primary clock source is available; it is enabled by setting the IESO Configuration bit.

Two-Speed Start-up should be enabled only if the primary oscillator mode is LP, XT or HS (Crystal-Based modes). Other sources do not require an OST start-up delay; for these, Two-Speed Start-up should be disabled.

When enabled, Resets and wake-ups from Sleep mode cause the device to configure itself to run from the internal oscillator block as the clock source, following the time-out of the Power-up Timer after a Power-on Reset is enabled. This allows almost immediate code execution while the primary oscillator starts and the OST is running. Once the OST times out, the device automatically switches to PRI\_RUN mode.

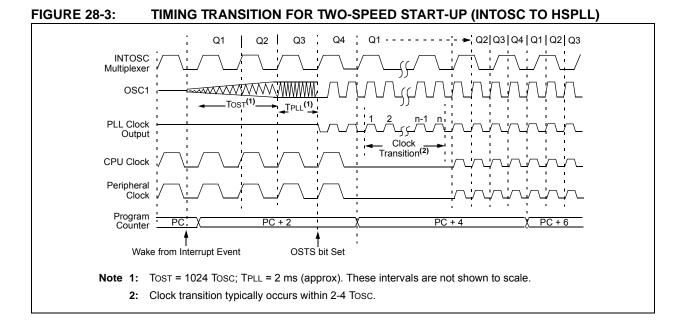
To use a higher clock speed on wake-up, the INTOSC or postscaler clock sources can be selected to provide a higher clock speed by setting bits, IRCF<2:0>, immediately after Reset. For wake-ups from Sleep, the INTOSC or postscaler clock sources can be selected by setting the IRCF<2:0> bits prior to entering Sleep mode.

In all other power-managed modes, Two-Speed Startup is not used. The device will be clocked by the currently selected clock source until the primary clock source becomes available. The setting of the IESO bit is ignored.

#### 28.4.1 SPECIAL CONSIDERATIONS FOR USING TWO-SPEED START-UP

While using the INTOSC oscillator in Two-Speed Startup, the device still obeys the normal command sequences for entering power-managed modes, including multiple SLEEP instructions (refer to Section 4.1.4 "Multiple Sleep Commands"). In practice, this means that user code can change the SCS<1:0> bit settings or issue SLEEP instructions before the OST times out. This would allow an application to briefly wake up, perform routine "housekeeping" tasks and return to Sleep before the device starts to operate from the primary oscillator.

User code can also check if the primary clock source is currently providing the device clocking by checking the status of the OSTS bit (OSCCON<3>). If the bit is set, the primary oscillator is providing the clock. Otherwise, the internal oscillator block is providing the clock during wake-up from Reset or Sleep mode.

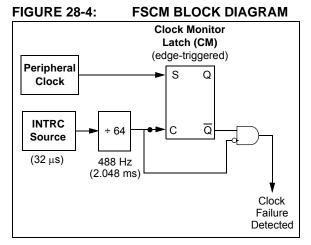


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#### 28.5 Fail-Safe Clock Monitor

The Fail-Safe Clock Monitor (FSCM) allows the microcontroller to continue operation in the event of an external oscillator failure by automatically switching the device clock to the internal oscillator block. The FSCM function is enabled by setting the FCMEN Configuration bit.

When FSCM is enabled, the LF-INTOSC oscillator runs at all times to monitor clocks to peripherals and provide a backup clock in the event of a clock failure. Clock monitoring (shown in Figure 28-4) is accomplished by creating a sample clock signal, which is the output from the LF-INTOSC, divided by 64. This allows ample time between FSCM sample clocks for a peripheral clock edge to occur. The peripheral device clock and the sample clock are presented as inputs to the Clock Monitor (CM) latch. The CM is set on the falling edge of the device clock source, but cleared on the rising edge of the sample clock.



Clock failure is tested for on the falling edge of the sample clock. If a sample clock falling edge occurs while CM is still set, a clock failure has been detected (Figure 28-5). This causes the following:

- The FSCM generates an oscillator fail interrupt by setting bit, OSCFIF (PIR2<7>)
- The device clock source switches to the internal oscillator block (OSCCON is not updated to show the current clock source – this is the Fail-Safe condition)
- · The WDT is reset

During switchover, the postscaler frequency from the internal oscillator block may not be sufficiently stable for timing-sensitive applications. In these cases, it may be desirable to select another clock configuration and enter an alternate power-managed mode. This can be done to attempt a partial recovery or execute a controlled shutdown. See Section 4.1.4 "Multiple Sleep Commands" and Section 28.4.1 "Special Considerations for Using Two-Speed Start-up" for more details.

To use a higher clock speed on wake-up, the INTOSC or postscaler clock sources can be selected to provide a higher clock speed by setting bits, IRCF<2:0>, immediately after Reset. For wake-ups from Sleep, the INTOSC or postscaler clock sources can be selected by setting the IRCF<2:0> bits prior to entering Sleep mode.

The FSCM will detect only failures of the primary or secondary clock sources. If the internal oscillator block fails, no failure would be detected nor would any action be possible.

#### 28.5.1 FSCM AND THE WATCHDOG TIMER

Both the FSCM and the WDT are clocked by the INTOSC oscillator. Since the WDT operates with a separate divider and counter, disabling the WDT has no effect on the operation of the INTOSC oscillator when the FSCM is enabled.

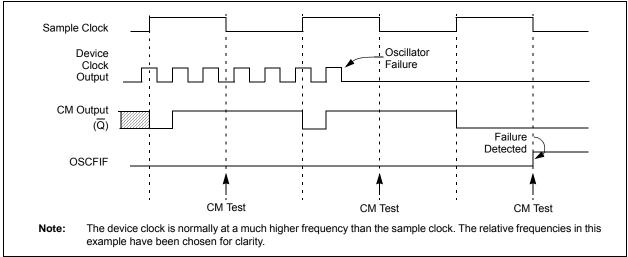
As already noted, the clock source is switched to the INTOSC clock when a clock failure is detected. Depending on the frequency selected by the IRCF<2:0> bits, this may mean a substantial change in the speed of code execution. If the WDT is enabled with a small prescale value, a decrease in clock speed allows a WDT time-out to occur and a subsequent device Reset. For this reason, Fail-Safe Clock events also reset the WDT and postscaler, allowing it to start timing from when execution speed was changed, and decreasing the likelihood of an erroneous time-out.

#### 28.5.2 EXITING FAIL-SAFE OPERATION

The Fail-Safe condition is terminated by either a device Reset or by entering a power-managed mode. On Reset, the controller starts the primary clock source specified in Configuration Register 1H (with any required start-up delays that are required for the Oscillator mode, such as the OST or PLL timer). The INTOSC multiplexer provides the device clock until the primary clock source becomes ready (similar to a Two-Speed Start-up). The clock source is then switched to the primary clock (indicated by the OSTS bit in the OSCCON register becoming set). The Fail-Safe Clock Monitor then resumes monitoring the peripheral clock.

The primary clock source may never become ready during start-up. In this case, operation is clocked by the INTOSC multiplexer. The OSCCON register will remain in its Reset state until a power-managed mode is entered.





#### 28.5.3 FSCM INTERRUPTS IN POWER-MANAGED MODES

By entering a power-managed mode, the clock multiplexer selects the clock source selected by the OSCCON register. Fail-Safe Clock Monitoring of the power-managed clock source resumes in the power-managed mode.

If an oscillator failure occurs during power-managed operation, the subsequent events depend on whether or not the Oscillator Failure Interrupt Flag bit is enabled (OSCFIF = 1). If enabled, code execution will be clocked by the INTOSC multiplexer. An automatic transition back to the failed clock source will not occur.

If the interrupt is disabled, subsequent interrupts while in Idle mode will cause the CPU to begin executing instructions while being clocked by the INTOSC source.

#### 28.5.4 POR OR WAKE FROM SLEEP

The FSCM is designed to detect oscillator failure at any point after the device has exited Power-on Reset (POR) or low-power Sleep mode. When the primary device clock is EC, RC or INTRC modes, monitoring can begin immediately following these events. For Oscillator modes involving a crystal or resonator (HS, HSPLL, LP or XT), the situation is somewhat different. Since the oscillator may require a start-up time considerably longer than the FCSM sample clock time, a false clock failure may be detected. To prevent this, the internal oscillator block is automatically configured as the device clock and functions until the primary clock is stable (when the OST and PLL timers have timed out).

This is identical to Two-Speed Start-up mode. Once the primary clock is stable, the INTOSC returns to its role as the FSCM source.

Note: The same logic that prevents false oscillator failure interrupts on POR, or wake from Sleep, also prevents the detection of the oscillator's failure to start at all following these events. This can be avoided by monitoring the OSTS bit and using a timing routine to determine if the oscillator is taking too long to start. Even so, no oscillator failure interrupt will be flagged.

As noted in Section 28.4.1 "Special Considerations for Using Two-Speed Start-up", it is also possible to select another clock configuration and enter an alternate power-managed mode while waiting for the primary clock to become stable. When the new powermanaged mode is selected, the primary clock is disabled.

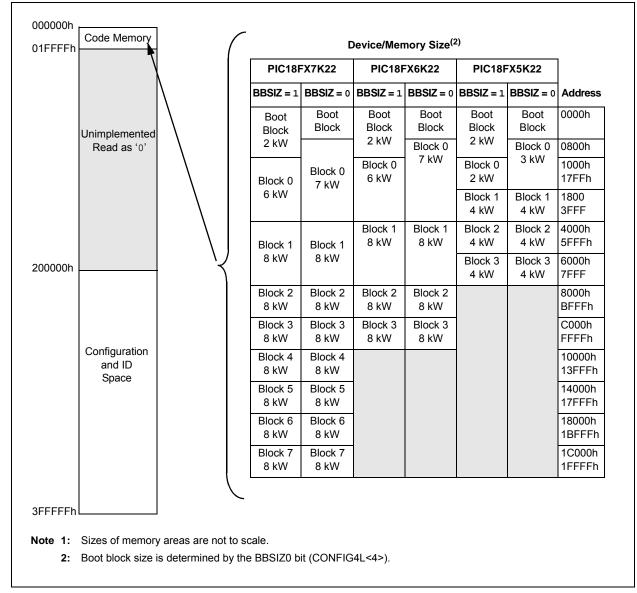
# 28.6 Program Verification and Code Protection

The user program memory is divided into four blocks for the PIC18FX5K22 and PIC18FX6K22 devices, and eight blocks for PIC18FX7K22 devices. One of these is a boot block of 1 or 2 Kbytes. The remainder of the memory is divided into blocks on binary boundaries. Each of the blocks has three code protection bits associated with them. They are:

- Code-Protect bit (CPx)
- Write-Protect bit (WRTx)
- External Block Table Read bit (EBTRx)

Figure 28-6 shows the program memory organization for 48, 64, 96 and 128 Kbyte devices and the specific code protection bit associated with each block. The actual locations of the bits are summarized in Table 28-4.

#### FIGURE 28-6: CODE-PROTECTED PROGRAM MEMORY FOR THE PIC18F87K22 FAMILY<sup>(1)</sup>



File	Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
300008h	CONFIG5L	CP7 <sup>(1)</sup>	CP6 <sup>(1)</sup>	CP5 <sup>(1)</sup>	CP4 <sup>(1)</sup>	CP3	CP2	CP1	CP0
300009h	CONFIG5H	CPD	CPB		_		_	—	_
30000Ah	CONFIG6L	WRT7 <sup>(1)</sup>	WRT6 <sup>(1)</sup>	WRT5 <sup>(1)</sup>	WRT4 <sup>(1)</sup>	WRT3	WRT2	WRT1	WRT0
30000Bh	CONFIG6H	WRTD	WRTB	WRTC		_	_	_	—
30000Ch	CONFIG7L	EBTR7 <sup>(1)</sup>	EBTR6 <sup>(1)</sup>	EBTR5 <sup>(1)</sup>	EBTR4 <sup>(1)</sup>	EBTR3	EBTR2	EBTR1	EBTR0
30000Dh	CONFIG7H	_	EBTRB		_		_	_	_

TABLE 28-4: SUMMARY OF CODE PROTECTION REGISTERS

**Legend:** Shaded cells are unimplemented.

Note 1: This bit is available only on the PIC18F67K22 and PIC18F87K22 devices.

#### 28.6.1 PROGRAM MEMORY CODE PROTECTION

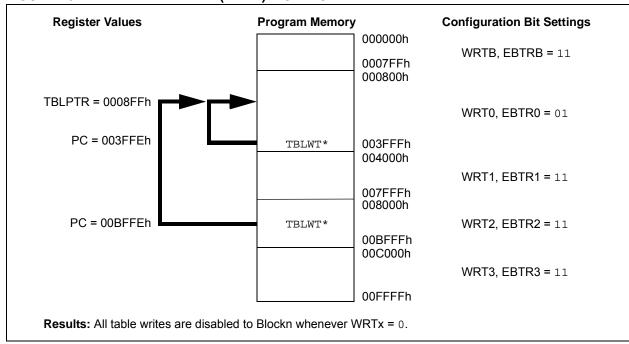
The program memory may be read to, or written from, any location using the table read and table write instructions. The Device ID may be read with table reads. The Configuration registers may be read and written with the table read and table write instructions.

In normal Execution mode, the CPx bits have no direct effect. CPx bits inhibit external reads and writes. A block of user memory may be protected from table writes if the WRTx Configuration bit is '0'.

The EBTRx bits control table reads. For a block of user memory, with the EBTRx bit set to '0', a table read instruction that executes from within that block is allowed

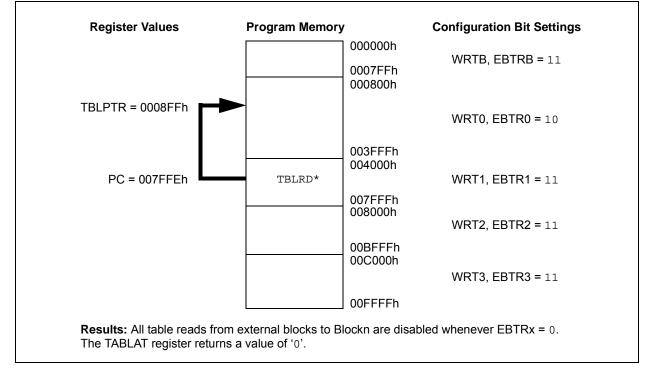
to read. A table read instruction that executes from a location outside of that block is not allowed to read and will result in reading '0's. Figures 28-7 through 28-9 illustrate table write and table read protection.

Note:	Code protection bits may only be written to a '0' from a '1' state. It is not possible to write a '1' to a bit in the '0' state. Code protection bits are only set to '1' by a full chip erase or block erase function. The full chip erase and block erase functions can only be initiated via ICSP or an external
	programmer. Refer to the device programming specification for more information.

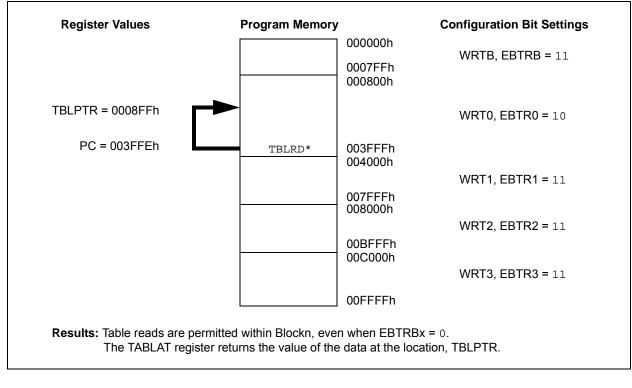


#### FIGURE 28-7: TABLE WRITE (WRTx) DISALLOWED

#### FIGURE 28-8: EXTERNAL BLOCK TABLE READ (EBTRx) DISALLOWED



#### FIGURE 28-9: EXTERNAL BLOCK TABLE READ (EBTRx) ALLOWED



#### 28.6.2 DATA EEPROM CODE PROTECTION

The entire data EEPROM is protected from external reads and writes by two bits: CPD and WRTD. CPD inhibits external reads and writes of data EEPROM. WRTD inhibits internal and external writes to data EEPROM. The CPU can always read data EEPROM under normal operation, regardless of the protection bit settings.

#### 28.6.3 CONFIGURATION REGISTER PROTECTION

The Configuration registers can be write-protected. The WRTC bit controls protection of the Configuration registers. In normal Execution mode, the WRTC bit is readable only. WRTC can only be written via ICSP or an external programmer.

#### 28.7 ID Locations

Eight memory locations (20000h-200007h) are designated as ID locations, where the user can store checksum or other code identification numbers. These locations are both readable and writable, during normal execution, through the TBLRD and TBLWT instructions or during program/verify. The ID locations can be read when the device is code-protected.

#### 28.8 In-Circuit Serial Programming

The PIC18F87K22 family of devices can be serially programmed while in the end application circuit. This is simply done with two lines for clock and data, and three other lines for power, ground and the programming voltage. This allows customers to manufacture boards with unprogrammed devices and then program the microcontroller just before shipping the product. This also allows the most recent firmware or a custom firmware to be programmed.

For the various Programming modes, see the device programming specification.

#### 28.9 In-Circuit Debugger

When the DEBUG Configuration bit is programmed to a '0', the In-Circuit Debugger (ICD) functionality is enabled. This function allows simple debugging functions when used with MPLAB<sup>®</sup> IDE. When the microcontroller has this feature enabled, some resources are not available for general use. Table 28-5 shows which resources are required by the background debugger.

I/O Pins:	RB6, RB7
Stack:	Two levels
Program Memory:	512 bytes
Data Memory:	10 bytes

To use the In-Circuit Debugger function of the microcontroller, the design must implement In-Circuit Serial Programming connections to MCLR/RG5/VPP, VDD, Vss, RB7 and RB6. This will interface to the In-Circuit Debugger module, available from Microchip or one of the third-party development tool companies.

# PIC18F87K22 FAMILY

NOTES:

#### 29.0 INSTRUCTION SET SUMMARY

The PIC18F87K22 family of devices incorporates the standard set of 75 PIC18 core instructions, as well as an extended set of 8 new instructions for the optimization of code that is recursive or that utilizes a software stack. The extended set is discussed later in this section.

#### 29.1 Standard Instruction Set

The standard PIC18 MCU instruction set adds many enhancements to the previous PIC<sup>®</sup> MCU instruction sets, while maintaining an easy migration from these PIC MCU instruction sets. Most instructions are a single program memory word (16 bits), but there are four instructions that require two program memory locations.

Each single-word instruction is a 16-bit word divided into an opcode, which specifies the instruction type and one or more operands, which further specify the operation of the instruction.

The instruction set is highly orthogonal and is grouped into four basic categories:

- Byte-oriented operations
- Bit-oriented operations
- · Literal operations
- Control operations

The PIC18 instruction set summary in Table 29-2 lists **byte-oriented**, **bit-oriented**, **literal** and **control** operations. Table 29-1 shows the opcode field descriptions.

Most byte-oriented instructions have three operands:

- 1. The file register (specified by 'f')
- 2. The destination of the result (specified by 'd')
- 3. The accessed memory (specified by 'a')

The file register designator, 'f', specifies which file register is to be used by the instruction. The destination designator, 'd', specifies where the result of the operation is to be placed. If 'd' is zero, the result is placed in the WREG register. If 'd' is one, the result is placed in the file register specified in the instruction.

All **bit-oriented** instructions have three operands:

- 1. The file register (specified by 'f')
- 2. The bit in the file register (specified by 'b')
- 3. The accessed memory (specified by 'a')

The bit field designator, 'b', selects the number of the bit affected by the operation, while the file register designator, 'f', represents the number of the file in which the bit is located.

The **literal** instructions may use some of the following operands:

- A literal value to be loaded into a file register (specified by 'k')
- The desired FSR register to load the literal value into (specified by 'f')
- No operand required (specified by '—')

The **control** instructions may use some of the following operands:

- A program memory address (specified by 'n')
- The mode of the CALL or RETURN instructions (specified by 's')
- The mode of the table read and table write instructions (specified by 'm')
- No operand required (specified by '—')

All instructions are a single word, except for four double-word instructions. These instructions were made double-word to contain the required information in 32 bits. In the second word, the 4 MSbs are '1's. If this second word is executed as an instruction (by itself), it will execute as a NOP.

All single-word instructions are executed in a single instruction cycle, unless a conditional test is true or the Program Counter is changed as a result of the instruction. In these cases, the execution takes two instruction cycles with the additional instruction cycle(s) executed as a NOP.

The double-word instructions execute in two instruction cycles.

One instruction cycle consists of four oscillator periods. Thus, for an oscillator frequency of 4 MHz, the normal instruction execution time is 1  $\mu$ s. If a conditional test is true, or the Program Counter is changed as a result of an instruction, the instruction execution time is 2  $\mu$ s. Two-word branch instructions (if true) would take 3  $\mu$ s.

Figure 29-1 shows the general formats that the instructions can have. All examples use the convention 'nnh' to represent a hexadecimal number.

The Instruction Set Summary, shown in Table 29-2, lists the standard instructions recognized by the Microchip MPASM<sup>™</sup> Assembler.

Section 29.1.1 "Standard Instruction Set" provides a description of each instruction.

#### TABLE 29-1: OPCODE FIELD DESCRIPTIONS

Field	Description
a	RAM access bit:
	a = 0: RAM location in Access RAM (BSR register is ignored)
	a = 1: RAM bank is specified by BSR register
bbb	Bit address within an 8-bit file register (0 to 7).
BSR	Bank Select Register. Used to select the current RAM bank.
C, DC, Z, OV, N	ALU Status bits: Carry, Digit Carry, Zero, Overflow, Negative.
d	Destination select bit: d = 0: store result in WREG
	d = 1: store result in file register f
dest	Destination: either the WREG register or the specified register file location.
f	8-bit register file address (00h to FFh), or 2-bit FSR designator (0h to 3h).
fs	12-bit register file address (000h to FFFh). This is the source address.
f <sub>d</sub>	12-bit register file address (000h to FFFh). This is the destination address.
GIE	Global Interrupt Enable bit.
k	Literal field, constant data or label (may be either an 8-bit, 12-bit or a 20-bit value).
label	Label name.
mm	The mode of the TBLPTR register for the table read and table write instructions.
	Only used with table read and table write instructions:
*	No Change to register (such as TBLPTR with table reads and writes)
*+	Post-Increment register (such as TBLPTR with table reads and writes)
* _	Post-Decrement register (such as TBLPTR with table reads and writes)
+*	Pre-Increment register (such as TBLPTR with table reads and writes)
n	The relative address (2's complement number) for relative branch instructions or the direct address for Call/Branch and Return instructions.
PC	Program Counter.
PCL	Program Counter Low Byte.
PCH	Program Counter High Byte.
PCLATH	Program Counter High Byte Latch.
PCLATU	Program Counter Upper Byte Latch.
PD	Power-Down bit.
PRODH	Product of Multiply High Byte.
PRODL	Product of Multiply Low Byte.
s	Fast Call/Return mode select bit:
	s = 0: do not update into/from shadow registers
	s = 1: certain registers loaded into/from shadow registers (Fast mode)
TBLPTR	21-bit Table Pointer (points to a Program Memory location).
TABLAT	8-bit Table Latch.
TO	Time-out bit.
TOS	Top-of-Stack.
u	Unused or Unchanged.
WDT	Watchdog Timer.
WREG	Working register (accumulator).
x	Don't care ('0' or '1'). The assembler will generate code with $x = 0$ . It is the recommended form of use for compatibility with all Microchip software tools.
Zs	7-bit offset value for Indirect Addressing of register files (source).
z <sub>d</sub>	7-bit offset value for Indirect Addressing of register files (destination).
{ }	Optional argument.
[text]	Indicates an Indexed Address.
(text)	The contents of text.
[expr] <n></n>	Specifies bit n of the register indicated by the pointer expr.
$\rightarrow$	Assigned to.
< >	Register bit field.
E	In the set of.
italics	User-defined term (font is Courier New).

Byte-oriented file register operations	Example Instruction
<u>15 10 9 8 7 0</u>	
OPCODE d a f (FILE #)	ADDWF MYREG, W, B
<ul> <li>d = 0 for result destination to be WREG register</li> <li>d = 1 for result destination to be file register (f)</li> <li>a = 0 to force Access Bank</li> <li>a = 1 for BSR to select bank</li> <li>f = 8-bit file register address</li> </ul>	
Byte to Byte move operations (2-word)	
<u>15 12 11 0</u>	
OPCODE f (Source FILE #)	MOVFF MYREG1, MYREG2
15 12 11 0	
1111 f (Destination FILE #)	
f = 12-bit file register address	
Bit-oriented file register operations	
<u>15 12 11 9 8 7 0</u>	
OPCODE b (BIT #) a f (FILE #)	BSF MYREG, bit, B
<ul> <li>b = 3-bit position of bit in file register (f)</li> <li>a = 0 to force Access Bank</li> <li>a = 1 for BSR to select bank</li> <li>f = 8-bit file register address</li> </ul>	
iteral operations	
15 8 7 O	
OPCODE k (literal)	MOVLW 7Fh
k = 8-bit immediate value	
Control operations	
CALL, GOTO and Branch operations	
15 8 7 0	
OPCODE n<7:0> (literal)	GOTO Label
15 12 11 0	
1111 n<19:8> (literal)	
n = 20-bit immediate value	
15 8 7 0	
OPCODE S n<7:0> (literal)	CALL MYFUNC
15 12 11 0	
1111 n<19:8> (literal)	
S = Fast bit	
15 11 10 0	
OPCODE n<10:0> (literal)	BRA MYFUNC
15 8 7 0	

#### TABLE 29-2: PIC18F87K22 FAMILY INSTRUCTION SET

Mnemo	onic,	Description	Cualas	16-E	Bit Instr	uction V	Vord	Status	Nataa
Opera	nds	Description	Cycles	MSb			LSb	Affected	Notes
BYTE-ORI	ENTED C	OPERATIONS							
ADDWF	f, d, a	Add WREG and f	1	0010	01da	ffff	ffff	C, DC, Z, OV, N	1, 2
ADDWFC	f, d, a	Add WREG and Carry bit to f	1	0010	00da	ffff	ffff	C, DC, Z, OV, N	1, 2
ANDWF	f, d, a	AND WREG with f	1	0001	01da	ffff	ffff	Z, N	1, 2
CLRF	f, a	Clear f	1	0110	101a	ffff	ffff	Z	2
COMF	f, d, a	Complement f	1	0001	11da	ffff	ffff	Z, N	1, 2
CPFSEQ	f, a	Compare f with WREG, Skip =	1 (2 or 3)	0110	001a	ffff	ffff	None	4
CPFSGT	f, a	Compare f with WREG, Skip >	1 (2 or 3)	0110	010a	ffff	ffff	None	4
CPFSLT	f, a	Compare f with WREG, Skip <	1 (2 or 3)	0110	000a	ffff	ffff	None	1, 2
DECF	f, d, a	Decrement f	1	0000	01da	ffff	ffff	C, DC, Z, OV, N	1, 2, 3, 4
DECFSZ	f, d, a	Decrement f, Skip if 0	1 (2 or 3)	0010	11da	ffff	ffff	None	1, 2, 3, 4
DCFSNZ	f, d, a	Decrement f, Skip if Not 0	1 (2 or 3)	0100	11da	ffff	ffff	None	1, 2
INCF	f, d, a	Increment f	1	0010	10da	ffff	ffff	C, DC, Z, OV, N	1, 2, 3, 4
INCFSZ	f, d, a	Increment f, Skip if 0	1 (2 or 3)	0011	11da	ffff	ffff	None	4
INFSNZ	f, d, a	Increment f, Skip if Not 0	1 (2 or 3)	0100	10da	ffff	ffff	None	1, 2
IORWF	f, d, a	Inclusive OR WREG with f	1	0001	00da	ffff	ffff	Z, N	1, 2
MOVF		Move f	1	0101	00da	ffff	ffff	Z, N	1
MOVFF	f <sub>s</sub> , f <sub>d</sub>	Move f <sub>s</sub> (source) to 1st word	2	1100	ffff	ffff	ffff	None	
	ũ ũ	f <sub>d</sub> (destination) 2nd word		1111	ffff	ffff	ffff		
MOVWF	f, a	Move WREG to f	1	0110	111a	ffff	ffff	None	
MULWF	f, a	Multiply WREG with f	1	0000	001a	ffff	ffff	None	1, 2
NEGF	f, a	Negate f	1	0110	110a	ffff	ffff	C, DC, Z, OV, N	
RLCF	f, d, a	Rotate Left f through Carry	1	0011	01da	ffff	ffff	C, Z, N	1, 2
RLNCF	f, d, a	Rotate Left f (No Carry)	1	0100	01da	ffff	ffff	Z, N	-
RRCF	f, d, a	Rotate Right f through Carry	1	0011	00da	ffff	ffff	C, Z, N	
RRNCF	f, d, a	Rotate Right f (No Carry)	1	0100	00da	ffff	ffff	Z, N	
SETF	f, a	Set f	1	0110	100a	ffff	ffff	None	1, 2
SUBFWB	f, d, a	Subtract f from WREG with	1	0101	01da	ffff	ffff	C, DC, Z, OV, N	-
		Borrow							
SUBWF	f, d, a	Subtract WREG from f	1	0101	11da	ffff	ffff	C, DC, Z, OV, N	1, 2
SUBWFB	f, d, a	Subtract WREG from f with	1	0101	10da	ffff		C, DC, Z, OV, N	
		Borrow							
SWAPF	f, d, a	Swap Nibbles in f	1	0011	10da	ffff	ffff	None	4
TSTFSZ	f, a	Test f, Skip if 0	1 (2 or 3)	0110	011a	ffff	ffff	None	1, 2
XORWF	f, d, a	Exclusive OR WREG with f	1	0001	10da	ffff	ffff	Z. N	

**Note 1:** When a PORT register is modified as a function of itself (e.g., MOVF PORTB, 1, 0), the value used will be that value present on the pins themselves. For example, if the data latch is '1' for a pin configured as input and is driven low by an external device, the data will be written back with a '0'.

2: If this instruction is executed on the TMR0 register (and, where applicable, d = 1), the prescaler will be cleared if assigned.

**3:** If the Program Counter (PC) is modified or a conditional test is true, the instruction requires two cycles. The second cycle is executed as a NOP.

4: Some instructions are two-word instructions. The second word of these instructions will be executed as a NOP unless the first word of the instruction retrieves the information embedded in these 16 bits. This ensures that all program memory locations have a valid instruction.

Mnem	onic,	Description	Ovelas	16-E	Bit Instr	uction V	Vord	Status	Note -
Opera		Description	Cycles	MSb			LSb	Affected	Notes
BIT-ORIE	NTED OP	ERATIONS							
BCF	f, b, a	Bit Clear f	1	1001	bbba	ffff	ffff	None	1, 2
BSF	f, b, a	Bit Set f	1	1000	bbba	ffff	ffff	None	1, 2
BTFSC	f, b, a	Bit Test f, Skip if Clear	1 (2 or 3)	1011	bbba	ffff	ffff	None	3, 4
BTFSS	f, b, a	Bit Test f, Skip if Set	1 (2 or 3)	1010	bbba	ffff	ffff	None	3, 4
BTG	f, b, a	Bit Toggle f	1	0111	bbba	ffff	ffff	None	<b>1</b> , <b>2</b>
CONTRO	OPERA	TIONS							
BC	n	Branch if Carry	1 (2)	1110	0010	nnnn	nnnn	None	
BN	n	Branch if Negative	1 (2)	1110	0110	nnnn	nnnn	None	
BNC	n	Branch if Not Carry	1 (2)	1110	0011	nnnn	nnnn	None	
BNN	n	Branch if Not Negative	1 (2)	1110	0111	nnnn	nnnn	None	
BNOV	n	Branch if Not Overflow	1 (2)	1110	0101	nnnn	nnnn	None	
BNZ	n	Branch if Not Zero	1 (2)	1110	0001	nnnn	nnnn	None	
BOV	n	Branch if Overflow	1 (2)	1110	0100	nnnn	nnnn	None	
BRA	n	Branch Unconditionally	2	1101	0nnn	nnnn	nnnn	None	
BZ	n	Branch if Zero	1 (2)	1110	0000	nnnn	nnnn	None	
CALL	n, s	Call Subroutine 1st word	2	1110	110s	kkkk	kkkk	None	
		2nd word		1111	kkkk	kkkk	kkkk		
CLRWDT	—	Clear Watchdog Timer	1	0000	0000	0000		TO, PD	
DAW	—	Decimal Adjust WREG	1	0000	0000	0000	0111	С	
GOTO	n	Go to Address 1st word	2	1110	1111	kkkk	kkkk	None	
		2nd word			kkkk	kkkk	kkkk		
NOP	—	No Operation	1	0000	0000	0000		None	
NOP	—	No Operation	1		XXXX	XXXX		None	4
POP	—	Pop Top of Return Stack (TOS)	1	0000	0000	0000		None	
PUSH	—	Push Top of Return Stack (TOS)		0000	0000	0000		None	
RCALL	n	Relative Call	2	1101	1nnn	nnnn		None	
RESET		Software Device Reset	1	0000	0000	1111		All	
RETFIE	S	Return from Interrupt Enable	2	0000	0000	0001	000s	GIE/GIEH, PEIE/GIEL	
RETLW	k	Return with Literal in WREG	2	0000	1100	kkkk	kkkk	None	
RETURN	S	Return from Subroutine	2	0000	0000	0001	001s	None	
SLEEP	—	Go into Standby mode	1	0000	0000	0000	0011	TO, PD	

#### TABLE 29-2: PIC18F87K22 FAMILY INSTRUCTION SET (CONTINUED)

**Note 1:** When a PORT register is modified as a function of itself (e.g., MOVF PORTB, 1, 0), the value used will be that value present on the pins themselves. For example, if the data latch is '1' for a pin configured as input and is driven low by an external device, the data will be written back with a '0'.

2: If this instruction is executed on the TMR0 register (and, where applicable, d = 1), the prescaler will be cleared if assigned.

**3:** If the Program Counter (PC) is modified or a conditional test is true, the instruction requires two cycles. The second cycle is executed as a NOP.

4: Some instructions are two-word instructions. The second word of these instructions will be executed as a NOP unless the first word of the instruction retrieves the information embedded in these 16 bits. This ensures that all program memory locations have a valid instruction.

TABLE 29	<u> </u>	IC18F8/K22 FAMILT INSTRU							
Mnemo	onic,	Description	Cycles	16-	-Bit Instruction Word Status Not		Notes		
Opera	nds	Description	Cycles	MSb			LSb	Affected	Noles
LITERAL C	PERAT	IONS							
ADDLW	k	Add Literal and WREG	1	0000	1111	kkkk	kkkk	C, DC, Z, OV, N	
ANDLW	k	AND Literal with WREG	1	0000	1011	kkkk	kkkk	Z, N	
IORLW	k	Inclusive OR Literal with WREG	1	0000	1001	kkkk	kkkk	Z, N	
LFSR	f, k	Move literal (12-bit) 2nd word	2	1110	1110	00ff	kkkk	None	
		to FSR(f) 1st word		1111	0000	kkkk	kkkk		
MOVLB	k	Move Literal to BSR<3:0>	1	0000	0001	0000	kkkk	None	
MOVLW	k	Move Literal to WREG	1	0000	1110	kkkk	kkkk	None	
MULLW	k	Multiply Literal with WREG	1	0000	1101	kkkk	kkkk	None	
RETLW	k	Return with Literal in WREG	2	0000	1100	kkkk	kkkk	None	
SUBLW	k	Subtract WREG from Literal	1	0000	1000	kkkk	kkkk	C, DC, Z, OV, N	
XORLW	k	Exclusive OR Literal with WREG	1	0000	1010	kkkk	kkkk	Z, N	
DATA MEN	IORY ↔	PROGRAM MEMORY OPERATI	ONS						
TBLRD*		Table Read	2	0000	0000	0000	1000	None	
TBLRD*+		Table Read with Post-Increment		0000	0000	0000	1001	None	
TBLRD*-		Table Read with Post-Decrement		0000	0000	0000	1010	None	
TBLRD+*		Table Read with Pre-Increment		0000	0000	0000	1011	None	
TBLWT*		Table Write	2	0000	0000	0000	1100	None	
TBLWT*+		Table Write with Post-Increment		0000	0000	0000	1101	None	
TBLWT*-		Table Write with Post-Decrement		0000	0000	0000	1110	None	
TBLWT+*		Table Write with Pre-Increment		0000	0000	0000	1111	None	

#### TABLE 29-2: PIC18F87K22 FAMILY INSTRUCTION SET (CONTINUED)

**Note 1:** When a PORT register is modified as a function of itself (e.g., MOVF PORTB, 1, 0), the value used will be that value present on the pins themselves. For example, if the data latch is '1' for a pin configured as input and is driven low by an external device, the data will be written back with a '0'.

2: If this instruction is executed on the TMR0 register (and, where applicable, d = 1), the prescaler will be cleared if assigned.

**3:** If the Program Counter (PC) is modified or a conditional test is true, the instruction requires two cycles. The second cycle is executed as a NOP.

4: Some instructions are two-word instructions. The second word of these instructions will be executed as a NOP unless the first word of the instruction retrieves the information embedded in these 16 bits. This ensures that all program memory locations have a valid instruction.

#### 29.1.1 STANDARD INSTRUCTION SET

ADDLW	ADD Litera	l to W		ADDWF	ADD W to	f	
Syntax:	ADDLW	k		Syntax:	ADDWF	f {,d {,a}}	
Operands:	$0 \le k \le 255$			Operands:	$0 \leq f \leq 255$		
Operation:	$(W) + k \to V$	N			d ∈ [0,1] a ∈ [0,1]		
Status Affected:	N, OV, C, D	)C, Z		Operation:	$(W) + (f) \rightarrow$	dest	
Encoding:	0000	1111 kkl	kk kkkk	Status Affected:	N, OV, C, E		
Description:		ts of W are ad		Encoding:	0010	01da ffi	f ffff
	8-bit literal ' W.	k' and the res	ult is placed in	Description:		egister 'f'. If 'd'	
Words:	1			Beconption	result is sto	ored in W. If 'd'	is '1', the
Cycles:	1					ored back in re	
Q Cycle Activity:						he Access Bar he BSR is use	
Q1	Q2	Q3	Q4		GPR bank.		
Decode	Read	Process	Write to		lf 'a' is '0' a	ind the extende	ed instruction
	literal 'k'	Data	W			led, this instruc	
<b>E</b> uropean les		-				Literal Offset A never f ≤ 95 (5l	
Example: Before Instruc		.5h			Section 29	.2.3 "Byte-Ori	iented and
W =	10h					ed Instruction set Mode" for	
After Instructio				Words:	1		dotano.
W =	25h			Cycles:	1		
				Q Cycle Activity:			
				Q1	Q2	Q3	Q4
				Decode	Read	Process	Write to
					register 'f'	Data	destination
				Example:	ADDWF		
						REG, 0, 0	
				Before Instrue W	= 17h		
				REG	= 0C2h		
				After Instructi W	on = 0D9h		
				REG	= 0C2h		

Note: All PIC18 instructions may take an optional label argument preceding the instruction mnemonic for use in symbolic addressing. If a label is used, the instruction format then becomes: {label} instruction argument(s).

ADDWFC	ADD W an	d Carry	bit to f	
Syntax:	ADDWFC	f {,d {,	a}}	
Operands:	$0 \le f \le 255$ $d \in [0,1]$ $a \in [0,1]$			
Operation:	(W) + (f) +	$(C) \rightarrow de$	st	
Status Affected:	N,OV, C, D	C, Z		
Encoding:	0010	00da	ffff	ffff
Description:	Add W, the location 'f'. placed in V placed in d	If 'd' is '0 V. If 'd' is	, the res '1', the re	ult is sult is
	If 'a' is '0', t If 'a' is '1', t GPR bank.	he BSR i		
	If 'a' is '0' a set is enab in Indexed mode when Section 29 Bit-Oriente Literal Off	led, this i Literal O never f ≤ 0.2.3 "By ed Instru	nstruction ffset Addr 95 (5Fh). te-Orient ctions in	operates essing See ed and Indexed
Words:	1			
Cycles:	1			
Q Cycle Activity:				
Q1	Q2	Q3		Q4
Decode	Read register 'f'	Proce Data		Vrite to stination
Example:	ADDWFC	REG,	0, 1	
Before Instruc Carry bit REG W After Instructio Carry bit REG W	= 1 = 02h = 4Dh			

ANDLW	AND Litera	al with W	1	
Syntax:	ANDLW	k		
Operands:	$0 \le k \le 255$	i		
Operation:	(W) .AND.	$k \rightarrow W$		
Status Affected:	N, Z			
Encoding:	0000	1011	kkk}	k kkkk
Description:				Ded with the placed in W
Words:	1			
Cycles:	1			
Q Cycle Activity:				
Q1	Q2	Q3	}	Q4
Decode	Read literal 'k'	Proce Data		Write to W
Example:	ANDLW	05Fh		
Before Instruc W After Instructio	= A3h			

ANDWF	AND W wit	th f		BC		Branch if C	Carry	
Syntax:	ANDWF	f {,d {,a}}		Synta	ax:	BC n		
Operands:	$0 \le f \le 255$			Oper	ands:	-128 ≤ n ≤ ′	127	
	$\begin{array}{l} d \in [0,1] \\ a \in [0,1] \end{array}$			Oper	ation:	if Carry bit i (PC) + 2 + 2	,	
Operation:	(W) .AND.	(f) $\rightarrow$ dest		Statu	s Affected:	None		
Status Affected:	N, Z			Enco	dina:	1110	0010 nn	nn nnnn
Encoding:	0001	01da ff:	ff ffff		ription:		bit is '1', then	
Description:	The conten	ts of W are AN	Ded with	Dest		will branch.		the program
	in W. If 'd' is in register ' If 'a' is '0', t		is stored back nk is selected.			added to th incremente instruction,	nplement num e PC. Since th d to fetch the the new addre n. This instruct	e PC will have next ess will be
	GPR bank.					two-cycle in		
	lf 'a' is '0' a	ind the extend	ed instruction	Word	s.	1		
		led, this instruc		Cycle		1(2)		
	mode wher	Literal Offset A never f ≤ 95 (5l <b>.2.3 "Byte-Or</b>	Fh). See	-	cle Activity:	(_)		
		ed Instruction			Q1	Q2	Q3	Q4
	Literal Off	set Mode" for	details.		Decode	Read literal	Process	Write to
Words:	1					ʻn'	Data	PC
Cycles:	1				No operation	No operation	No operation	No operation
Q Cycle Activity:				If No	Jump:	operation	operation	operation
Q1	Q2	Q3	Q4		Q1	Q2	Q3	Q4
Decode	Read	Process	Write to		Decode	Read literal	Process	No
	register 'f'	Data	destination			'n'	Data	operation
Example: Before Instruct		REG, 0, 0		Exan		HERE	BC 5	
W REG After Instructi W	= 17h = C2h				Before Instruc PC After Instructi If Carry	= ad on = 1;	dress (here	
REG	= C2h				PC If Carry PC	= 0;	dress (HERE dress (HERE	

BCF	Bit Clear f			BN		Branch if N	legative	
Syntax:	BCF f, b	{,a}		Synt	ax:	BN n		
Operands:	$0 \leq f \leq 255$			Oper	ands:	-128 ≤ n ≤ ′	127	
	$\begin{array}{l} 0 \leq b \leq 7 \\ a \in [0,1] \end{array}$			Oper	ation:	if Negative (PC) + 2 + 2	,	
Operation:	$0 \rightarrow f \le b >$			Statu	is Affected:	None		
Status Affected:	None			Enco	oding:	1110	0110 nn	nn nnnn
Encoding:	1001	bbba ff	ff ffff		cription:	If the Nega	tive bit is '1', th	nen the
Description:	Bit 'b' in reg	gister 'f' is clea	ared.			program wi		
	,		ink is selected. ed to select the			added to the incremente	d to fetch the	e PC will have next
	set is enabl		led instruction ction operates Addressing			,	the new addre n. This instruct istruction.	
		never f $\leq$ 95 (5		Word	ds:	1		
		.2.3 "Byte-Or	riented and	Cycle	es:	1(2)		
	Literal Offs	set Mode" for			ycle Activity:			
Words:	1				Q1	Q2	Q3	Q4
Cycles:	1				Decode	Read literal	Process	Write to
Q Cycle Activity:						ʻn'	Data	PC
Q1	Q2	Q3	Q4		No operation	No operation	No operation	No operation
Decode	Read register 'f'	Process Data	Write register 'f'	lf No	o Jump:	operation	operation	operation
	. egiotoi .	Data	regiotor i		Q1	Q2	Q3	Q4
Example:	BCF F	LAG_REG,	7, 0		Decode	Read literal	Process	No
Before Instruc	tion					'n'	Data	operation
FLAG_R	EG = C7h							
After Instructio				<u>Exar</u>	<u>nple:</u>	HERE	BN Jump	
FLAG_R	EG = 47h				Before Instruct PC After Instructi	= ad	dress (HERE)	
					If Negati PC If Negati	= ad	dress (Jump)	
					PC	= au	dress (HERE	+ 2)

BNC		Branch if N	lot Carry		BNN	
Synta	ax:	BNC n			Synta	
Oper	ands:	-128 ≤ n ≤ 1	127		Oper	
Oper	ation:	if Carry bit i (PC) + 2 + 2			Oper	
Statu	is Affected:	None			Statu	
Enco	oding:	1110 0011 nnnn nnn			Enco	
Desc	cription:	If the Carry will branch.	bit is '0', then	the program	Desc	
The 2's complement number '2n' is added to the PC. Since the PC will have incremented to fetch the next instruction, the new address will be PC + 2 + 2n. This instruction is then a two-cycle instruction.						
Word	ds:	1		Word		
Cycles: 1(2)				Cycle		
Q C If Ju	ycle Activity: imp:				Q C If Ju	
	Q1	Q2	Q3	Q4	_	
	Decode	Read literal 'n'	Process Data	Write to PC		
	No operation	No operation	No operation	No operation		
lf No	o Jump:				lf No	
	Q1	Q2	Q3	Q4	, ,	
	Decode	Read literal 'n'	Process Data	No operation		
<u>Exan</u>	Before Instruc PC	= ad	BNC Jump dress (HERE		Exan	
	After Instruction If Carry PC If Carry PC	= 0; = add = 1;	dress (Jump dress (HERE			

BNN		Branch if I	Not Nega	tive	
Synta	ax:	BNN n			
Oper	ands:	-128 ≤ n ≤	127		
Oper	ation:	if Negative (PC) + 2 +		;	
Statu	is Affected:	None			
Enco	oding:	1110	0111	nnnn	nnnn
Desc	cription:	If the Nega program wi			he
The 2's complement number '2n' i added to the PC. Since the PC will incremented to fetch the next instruction, the new address will b PC + 2 + 2n. This instruction is the two-cycle instruction.				C will have vill be	
Words:		1			
Cycles:		1(2)			
Q C If Ju	ycle Activity: imp:				
	Q1	Q2	Q3	•	Q4
	Decode	Read literal 'n'	Proce Data		Vrite to PC
	No	No	No		No
	operation	operation	operat	ion op	peration
If No	o Jump:				
	Q1	Q2	Q3		Q4
	Decode	Read literal 'n'	Proce Data		No peration
	L		Dala		
<u>Exan</u>	nple:	HERE	BNN	Jump	
	Before Instruct PC After Instruction	= ac	dress (1	HERE)	
	If Negativ PC If Negativ	= ac	dress (	Jump)	
	PC		dress (I	HERE + 2	2)

BNO	v	Branch if N	lot Overfle	ow						
Synta	ax:	BNOV n								
Oper	ands:	-128 ≤ n ≤ 1	127							
Oper	ation:		if Overflow bit is '0', (PC) + 2 + 2n $\rightarrow$ PC							
Statu	s Affected:	None	None							
Encoding:		1110	0101	nnnn	nnnn					
Description:			If the Overflow bit is '0', then the program will branch.							
		added to the incrementer instruction, PC + 2 + 2r	The 2's complement number '2n' is added to the PC. Since the PC will have incremented to fetch the next instruction, the new address will be PC + 2 + 2n. This instruction is then a two-cycle instruction.							
Words:		1								
Cycles:		1(2)								
Q C If Ju	ycle Activity: mp:									
	Q1	Q2	Q3		Q4					
	Decode	Read literal 'n'	Process Data	s V	Vrite to PC					
	No operation	No operation	No operatio	n op	No eration					
lf No	o Jump:									
	Q1	Q2	Q3		Q4					
	Decode	Read literal 'n'	Process Data	-	No eration					
<u>Exan</u>	nple:	HERE	BNOV Ju	ump						
	Before Instruc PC After Instructio	= ad	dress (HE	ERE)						
After Instruction If Overflow PC If Overflow PC										

Operands: $-128 \le n \le 127$ Operation:if Zero bit is '0', (PC) + 2 + 2n $\rightarrow$ PCStatus Affected:NoneEncoding: $1110$ $0001$ nmmDescription:If the Zero bit is '0', then the prograwill branch.The 2's complement number '2n' is added to the PC. Since the PC will brincremented to fetch the next instruction, the new address will be PC + 2 + 2n. This instruction is there two-cycle instruction.Words:1Cycles:1(2)Q Cycle Activity:If Jump:Q1Q2Q3Q4ProcessNoNoNoNoNoNoNoNoNoJump:Q1Q2Q3Q4	Syntax:	BNZ n						
Operation:       if Zero bit is '0', (PC) + 2 + 2n → PC         Status Affected:       None         Encoding:       1110       0001       nnnn       nn         Description:       If the Zero bit is '0', then the prograwill branch.         The 2's complement number '2n' is added to the PC. Since the PC will h incremented to fetch the next instruction, the new address will be PC + 2 + 2n. This instruction is then two-cycle instruction.         Words:       1         Cycles:       1(2)         Q Cycle Activity:       If Jump:         Q1       Q2       Q3       Q4         Decode       Read literal       Process       Write to 'n'         No       No       No       No         If No Jump:       Q1       Q2       Q3       Q4	5							
$(PC) + 2 + 2n \rightarrow PC$ Status Affected: None Encoding: <u>1110 0001 nnnn nn</u> Description: If the Zero bit is '0', then the progra will branch. The 2's complement number '2n' is added to the PC. Since the PC will h incremented to fetch the next instruction, the new address will be PC + 2 + 2n. This instruction is then two-cycle instruction. Words: 1 Cycles: 1(2) Q Cycle Activity: If Jump: <u>Q1 Q2 Q3 Q4</u> <u>Decode Read literal Process Write t 'n' Data PC No No No No No operation operation operation If No Jump: <u>Q1 Q2 Q3 Q4</u></u>	•							
Encoding:       1110       0001       nnnn       nn         Description:       If the Zero bit is '0', then the prograwill branch.         The 2's complement number '2n' is added to the PC. Since the PC will branch incremented to fetch the next instruction, the new address will be PC + 2 + 2n. This instruction is there two-cycle instruction.         Words:       1         Cycles:       1(2)         Q Cycle Activity:       If Jump:         Q1       Q2       Q3       Q4         No       No       No       No         No       No       No       No         If No Jump:       Q1       Q2       Q3       Q4	Operation:			;				
Description:       If the Zero bit is '0', then the prograwill branch.         The 2's complement number '2n' is added to the PC. Since the PC will here incremented to fetch the next instruction, the new address will be PC + 2 + 2n. This instruction is then two-cycle instruction.         Words:       1         Cycles:       1(2)         Q Cycle Activity:       If Jump:         Q1       Q2       Q3       Q4         Decode       Read literal       Process       Write to 'n'         No       No       No       No         If No Jump:       Q1       Q2       Q3       Q4	Status Affected:	None						
will branch.         The 2's complement number '2n' is added to the PC. Since the PC will h incremented to fetch the next instruction, the new address will be PC + 2 + 2n. This instruction is then two-cycle instruction.         Words:       1         Cycles:       1(2)         Q Cycle Activity:       If Jump:         Q1       Q2       Q3       Q4         Decode       Read literal       Process       Write t 'n'         No       No       No       No         If No Jump:       Q1       Q2       Q3       Q4	Encoding:	1110	0001	nnnr	n nnnn			
added to the PC. Since the PC will h       incremented to fetch the next       instruction, the new address will be       PC + 2 + 2n. This instruction is then       two-cycle instruction.       Words:     1       Cycles:     1(2)       Q Cycle Activity:       If Jump:       Q1     Q2       Q3     Q4       Decode     Read literal       Yrocess     Write t       'n'     Data       PC     No       operation     operation       If No Jump:       Q1     Q2       Q3     Q4	Description:		,	then the	e program			
Cycles: 1(2) Q Cycle Activity: If Jump: Q1 Q2 Q3 Q4 Decode Read literal Process Write t 'n' Data PC No No No No operation operation operation If No Jump: Q1 Q2 Q3 Q4		added to the incremented instruction, PC + 2 + 2r	e PC. Sir d to fetch the new n. This in	nce the n the ne addres structio	PC will have ext s will be			
Q Cycle Activity: If Jump: Q1 Q2 Q3 Q4 Decode Read literal Process Write t 'n' Data PC No No No No operation operation operation If No Jump: Q1 Q2 Q3 Q4	Words <sup>.</sup>	1	1					
If Jump: Q1 Q2 Q3 Q4 Decode Read literal Process Write t 'n' Data PC No No No No operation operation operation If No Jump: Q1 Q2 Q3 Q4								
DecodeRead literal (n'Process DataWrite t PCNoNoNoNooperationoperationoperationIf No Jump:Q1Q2Q3Q1Q2Q3		1(2)						
'n'         Data         PC           No         No         No         No           operation         operation         operation         operation           If No Jump:         Q1         Q2         Q3         Q4	Cycles: Q Cycle Activity:	1(2)						
If No Jump: Q1 Q2 Q3 Q4	Cycles: Q Cycle Activity: If Jump:	. ,	Q3	5	Q4			
If No Jump: Q1 Q2 Q3 Q4	Cycles: Q Cycle Activity: If Jump: Q1	Q2 Read literal	Proce	SS	Write to			
Q1 Q2 Q3 Q4	Cycles: Q Cycle Activity: If Jump: Q1 Decode	Q2 Read literal 'n'	Proce Data	ess a	Write to PC			
	Cycles: Q Cycle Activity: If Jump: Q1 Decode No	Q2 Read literal 'n' No	Proce Data No	ess a	Write to PC			
Decode Read literal Process No	Cycles: Q Cycle Activity: If Jump: Q1 Decode No operation	Q2 Read literal 'n' No	Proce Data No	ess a	Write to PC No			
	Cycles: Q Cycle Activity: If Jump: Q1 Decode No operation If No Jump:	Q2 Read literal 'n' No operation	Proce Data No operat	ess a ion	Write to PC No operation			
'n' Data operatio	Cycles: Q Cycle Activity: If Jump: Q1 Decode No operation If No Jump:	Q2 Read literal 'n' No operation	Proce Data No operat	ion	Write to PC No operation			
	Cycles: Q Cycle Activity: If Jump: Q1 Decode No operation If No Jump: Q1	Q2 Read literal 'n' No operation Q2 Read literal	Proce Data No operat Q3 Proce	ion	Write to PC No operation Q4			
Example: HERE BNZ Jump	Cycles: Q Cycle Activity: If Jump: Q1 Decode No operation If No Jump: Q1 Decode	Q2 Read literal 'n' No operation Q2 Read literal 'n'	Proce Data No operat Q3 Proce Data	ion	Write to PC No operation Q4 No			

PC After Instruction	=	address (HERE)
If Zero PC If Zero PC	= = =	0; address (Jump) 1; address (HERE + 2)

BRA		Unconditio	Unconditional Branch					
Synta	ax:	BRA n	BRA n					
Oper	ands:	-1024 ≤ n ≤	1023					
Oper	ation:	(PC) + 2 + 2	$2n \rightarrow PC$					
Statu	s Affected:	None						
Enco	ding:	1101	0nnn	nnn	ın	nnnn		
Desc	ription:	Add the 2's the PC. Sin incremente instruction, PC + 2 + 2r two-cycle ir	ce the Po d to fetch the new n. This in:	C will the n addre structi	have iext ss w	e vill be		
Word	ls:	1	1					
Cycle	es:	2	2					
QC	ycle Activity:							
	Q1	Q2	Q3		Q4			
	Decode	Read literal 'n'	Proce: Data		V	/rite to PC		
	No operation	No operation	No operati	ion	ор	No eration		
	n <u>ple:</u> Before Instruc PC After Instructio PC	= ad		Jump IERE ) Jump )				

BSF	Bit Set f			
Syntax:	BSF f, b {	,a}		
Operands:	$0 \le f \le 255$ $0 \le b \le 7$ $a \in [0,1]$			
Operation:	$1 \rightarrow \text{f}$			
Status Affected:	None			
Encoding:	1000	bbba	ffff	ffff
Description:	Bit 'b' in reg	gister 'f' is	s set.	
	If 'a' is '0', t If 'a' is '1', t GPR bank.			
	set is enabl in Indexed mode wher Section 29 Bit-Oriente Literal Offs	Literal Of lever f ≤ 9 .2.3 "Byted Instru	fset Addre 95 (5Fh). te-Oriente ctions in	essing See ed and Indexed
Words:	1			
Cycles:	1			
Q Cycle Activity:				
Q1	Q2	Q3		Q4
Decode	Read register 'f'	Proce Data		Write gister 'f'
Example:	BSF F	LAG_RE	G, 7, 1	
Before Instruc FLAG_R After Instructic FLAG_R	EG = 0A on			

BTFSC	Bit Test File	, Skip if Clea		BTFS	s	Bit Test File	e, Skip if Set		
Syntax:	BTFSC f, b	{,a}		Synta:	x:	BTFSS f, b {,a}			
Operands:	0 ≤ f ≤ 255 0 ≤ b ≤ 7 a ∈ [0,1]			Opera	ands:	0 ≤ f ≤ 255 0 ≤ b < 7 a ∈ [0,1]			
Operation:	skip if (f <b>)</b>	= 0		Opera	ation:	skip if (f <b>)</b>	= 1		
Status Affected:	None			Status	Affected:	None			
Encoding:	1011	bbba ff	ff ffff	Encod	ding:	1010 bbba ffff fff			
Description: If bit 'b' in register 'f' is '0', then the next instruction is skipped. If bit 'b' is '0', then the next instruction fetched during the current instruction execution is discarded and a NOP is executed instead, making this a two-cycle instruction. If 'a' is '0', the Access Bank is selected. If		Descr	iption:	instruction is the next inst current instr and a NOP is	gister 'f' is '1', f skipped. If bit ruction fetched uction executio s executed inst- rcle instruction.	'b' is '1', then I during the on is discarded ead, making			
		e Access Banl BSR is used to				,	e Access Bank BSR is used to	k is selected. If a select the	
	is enabled, t Indexed Lite whenever f ≤ Section 29.2 Bit-Oriented	d the extended his instruction ral Offset Addr ≤ 95 (5Fh). See 2.3 "Byte-Orie d Instructions et Mode" for d	essing mode ented and in Indexed			set is enable Indexed Lite whenever f Section 29. Bit-Orientee	d the extended d, this instructi ral Offset Addr ≤ 95 (5Fh). See 2.3 "Byte-Orie d Instructions et Mode" for d	on operates in essing mode ented and in Indexed	
Words:	1			Words	S:	1			
Cycles:		cles if skip and 2-word instrue		Cycles	s:		ycles if skip an a 2-word instru		
Q Cycle Activity:	- , -			Q Cv	cle Activity:	- ,			
Q1	Q2	Q3	Q4		Q1	Q2	Q3	Q4	
Decode	Read	Process	No	Γ	Decode	Read	Process	No	
	register 'f'	Data	operation			register 'f'	Data	operation	
If skip:		~~~	<u>.</u>	lf skip		00	~~~	~ ~ ~	
Q1 No	Q2 No	Q3 No	Q4 No	Г	Q1 No	Q2 No	Q3 No	Q4 No	
operation	operation	operation	operation		operation	operation	operation	operation	
If skip and followe				lf skip		d by 2-word ins			
Q1	Q2	Q3	Q4		Q1	Q2	Q3	Q4	
No	No	No	No		No	No	No	No	
operation	operation	operation	operation	-	operation	operation	operation	operation	
No operation	No operation	No operation	No operation		No operation	No operation	No operation	No operation	
Example:	HERE B FALSE : TRUE :	IFSC FLAG	s, 1, 0	<u>Exam</u>	<u>ple:</u>	HERE B' FALSE : TRUE :	IFSS FLAG	, 1, O	
Before Instruc PC After Instructio If FLAG<	= add on	ress (HERE)			Before Instruc PC After Instructio If FLAG<	= add	ress (HERE)		
If FLAG PC If FLAG PC	= add :1> = 1;	ress (TRUE)	)		If FLAG< PC If FLAG< PC	= add 1> = 1;	ress (FALSE	)	

BTG		Bit Toggle	f		BOV		Branch if C	Verflow		
Syntax	x:	BTG f, b {,a	}		Synta	ax:	BOV n			
Opera	inds:	$0 \leq f \leq 255$			Oper	ands:	-128 ≤ n ≤ 1	$-128 \le n \le 127$		
		0 ≤ b < 7 a ∈ [0,1]			Oper	ation:	if Overflow bit is '1', (PC) + 2 + 2n $\rightarrow$ PC			
Opera	ition:	$(\overline{f} < b >) \to f <$	b>		Statu	s Affected:	None			
Status	Affected:	None	None			ding:	1110	0100 nn	nn nnnn	
Encod	ling:	0111	bbba ff	ff ffff	Desc	ription:	If the Overfl	ow bit is '1', t	hen the	
Descr	iption:	Bit 'b' in dat inverted.	a memory loc	ation, 'f', is			program wil	l branch.		
	If 'a' is '0', the Access Bank is selected. If 'a' is '1', the BSR is used to select the GPR bank.					added to the incremented instruction,	The 2's complement number '2n' is added to the PC. Since the PC will ha incremented to fetch the next instruction, the new address will be			
	If 'a' is '0' and the extended instruction set is enabled, this instruction operates						PC + 2 + 2n. This instruction is then a two-cycle instruction.			
	set is enabled, this instruction operates in Indexed Literal Offset Addressing mode whenever $f \le 95$ (5Fh). See			Word	s:	1				
				Cycle	es:	1(2)				
		<b>Bit-Oriente</b>	.2.3 "Byte-Or d Instruction set Mode" for	s in Indexed		ycle Activity:				
Words	8:	1				Q1	Q2	Q3	Q4	
Cycles	S:	1				Decode	Read literal 'n'	Process Data	Write to PC	
Q Cy	cle Activity:					No	No	No	No	
г	Q1	Q2	Q3	Q4		operation	operation	operation	operation	
	Decode	Read	Process	Write	lf No	o Jump:				
		register 'f'	Data	register 'f'		Q1	Q2	Q3	Q4	
Exam	nlo <sup>.</sup>	BTG P	ORTC, 4, (			Decode	Read literal 'n'	Process Data	No operation	
			JRIC, 4, 0	)				Dulu	oporation	
E	Before Instruct PORTC		0101 <b>[75h]</b>		Exan	<u>iple:</u>	HERE	BOV Jump	)	
After Instruction: PORTC = 0110 0101 [65h]					Before Instruction PC = address (HERE) After Instruction					
						If Overflo PC If Overflo PC	= ade ow = 0;	dress (Jump dress (HERE		

ΒZ		Branch if Z	lero					
Synta	ax:	BZ n						
Oper	rands:	-128 ≤ n ≤ 1	$-128 \le n \le 127$					
Oper	ration:		if Zero bit is '1', (PC) + 2 + 2n $\rightarrow$ PC					
Statu	is Affected:	None						
Enco	oding:	1110	0000	nnnn	nnnn			
Desc	cription:	If the Zero I will branch.	oit is '1',	then the	program			
		The 2's con added to the incrementer instruction, PC + 2 + 2r two-cycle in	e PC. Sin d to fetch the new n. This in	nce the Pen the next address struction	C will have : will be			
Word	ds:	1						
Cycle	es:	1(2)	1(2)					
	ycle Activity:							
	Q1	Q2	Q3	8	Q4			
	Decode	Read literal	Proce		Write to			
	No	ʻn' No	Data	-	PC No			
	operation	operation	No operat		peration			
lf No	o Jump:							
	Q1	Q2	Q3	5	Q4			
	Decode	Read literal	Proce		No			
		'n'	Data	a o	peration			
Exar	<u>nple:</u>	HERE	ΒZ	Jump				
	Before Instruct PC	= ad	dress (1	HERE)				
	After Instruction	011						

Ormstann							
Syntax:	•	CALL k {,s}					
Operands:	0 ≤ k ≤ 104 s ∈ [0,1]	0070					
Operation:	(PC) + 4 $\rightarrow$	TOS,					
	$k \rightarrow PC < 20$	):1>;					
	if s = 1 (W) $\rightarrow$ WS,						
	(STATUS) -	$\rightarrow$ STATL	JSS,				
	$(BSR) \rightarrow B$	SRS					
Status Affected:	None						
Encoding:	1110	110					
1st word (k<7:0>) 2nd word(k<19:8>)	1110 1111	110s k <sub>19</sub> kkk	k <sub>7</sub> kl kkk		kkkk <sub>0</sub> kkkk <sub>8</sub>		
	registers ar	•					
Words: Cycles:	registers ar respective s STATUSS a update occi is loaded in two-cycle ir 2	shadow r and BSR urs. Ther ito PC<20	egiste S. If 's n, the 2 0:1>. (	ers, V s' = 0 20-bi	their VS, ), no it value 'l		
	respective s STATUSS a update occi is loaded in two-cycle ir 2 2	shadow r and BSR urs. Ther ito PC<20	egiste S. If 's n, the 2 0:1>. (	ers, V s' = 0 20-bi	their VS, ), no it value 'l		
Cycles: Q Cycle Activity: Q1	respective s STATUSS a update occi is loaded in two-cycle ir 2 2 Q2	shadow r and BSR urs. Ther to PC<2 nstruction	registe S. If 's n, the 2 0:1>. ( n.	ers, V 5' = 0 20-bi CALI	their VS, ), no it value 'l : is a Q4		
Cycles: Q Cycle Activity:	respective s STATUSS a update occi is loaded in two-cycle ir 2 2	shadow r and BSR urs. Ther to PC<20 nstructior	registe S. If 's n, the 2 0:1>. ( n. C to	ers, V S' = 0 20-bi CALI	their VS, ), no it value 'i L is a Q4		
Cycles: Q Cycle Activity: Q1	respective s STATUSS a update occl is loaded in two-cycle ir 2 2 2 Q2 Read literal	shadow r and BSR urs. Ther to PC<20 hstruction Q3 Push P	registe S. If 's n, the 2 0:1>. ( n. C to	Rea 'k'	their NS, o, no it value 'l c is a Q4 ad literal <19:8>,		
Cycles: Q Cycle Activity: Q1 Decode No	respective s STATUSS a update occi is loaded in two-cycle ir 2 2 Q2 Read literal 'k'<7:0>,	shadow r and BSR urs. Ther to PC<2 nstruction Q3 Push P stac No	egiste S. If 's n, the 2 0:1>. 0 n. C to k	ers, V s' = 0 20-bi CALI CALI K' K' Wri	their NS, o, no it value 'l : is a Q4 ad literal <19:8>, te to PC No		
Cycles: Q Cycle Activity: Q1 Decode	respective s STATUSS a update occi is loaded in two-cycle ir 2 2 Q2 Read literal 'k'<7:0>,	shadow r and BSR urs. Ther to PC<2 nstruction Q3 Push P stac	egiste S. If 's n, the 2 0:1>. 0 n. C to k	ers, V s' = 0 20-bi CALI CALI (CALI	their NS, o, no it value 'l : is a Q4 ad literal <19:8>, te to PC		
Cycles: Q Cycle Activity: Q1 Decode No	respective s STATUSS a update occi is loaded in two-cycle ir 2 2 Q2 Read literal 'k'<7:0>,	shadow r and BSR urs. Ther to PC<2 nstruction Q3 Push P stac No	egiste S. If 's n, the 2 0:1>. 0 n. C to k	ers, V 3' = 0 20-bi 20-bi CALI (CALI (K'- (K'- Wri op	their NS, o, no it value ' ' is a Q4 ad literal <19:8>, te to PC No eration		
Cycles: Q Cycle Activity: Q1 Decode No operation Example: Before Instruct	respective s STATUSS a update occi is loaded in two-cycle ir 2 2 Q2 Read literal 'k'<7:0>, No operation HERE tion	shadow r and BSR urs. Ther to PC<20 sstruction Q3 Push P stac No operat	registe S. If 's I, the 2 0:1>. 0 C to k ion	ers, V 3' = 0 20-bi 20-bi CALI (CALI (K'- (K'- Wri op	their NS, o, no it value ' ' is a Q4 ad literal <19:8>, te to PC No eration		
Cycles: Q Cycle Activity: Q1 Decode No operation Example:	respective s STATUSS a update occi is loaded in two-cycle ir 2 2 Q2 Read literal 'k'<7:0>, No operation HERE tion = address	shadow r and BSR urs. Ther to PC<2/ nstruction Q3 Push P stac No operat	registe S. If 's I, the 2 0:1>. 0 C to k ion	ers, V 3' = 0 20-bi 20-bi CALI (CALI (K'- (K'- Wri op	their NS, o, no it value ' ' is a Q4 ad literal <19:8>, te to PC No eration		

CLRF		Clear f			CLR	WDT	Clear Watc	hdog Timer	
Syntax	x:	CLRF f{,;	a}		Synt	ax:	CLRWDT		
Opera	nds:	$0 \leq f \leq 255$			Ope	rands:	None		
		a ∈ [0,1]			Ope	ration:	$000h \rightarrow WE$	DT,	
Opera	tion:	$\begin{array}{l} 000h \rightarrow f, \\ 1 \rightarrow Z \end{array}$					$1 \rightarrow \overline{\text{TO}},$	OT postscaler,	
Status	Affected:	Z			<b>.</b>		$1 \rightarrow \overline{PD}$		
Encod	ling:	0110	101a fff	f ffff		us Affected:	TO, PD		
Descri	iption:	Clears the or register.	contents of the	specified		oding: cription:	0000 CLRWDT ins	0000 00	
		,	he Access Bar he BSR is used				•	e WDT. Status	esets <u>the</u> post- s bits, TO and
		lf 'a' is '0' a	nd the extende	ed instruction	Wor	ds:	1		
			ed, this instruc	•	Cycl	es:	1		
			Literal Offset A iever f ≤ 95 (5F	•	QC	Cycle Activity:			
			.2.3 "Byte-Ori	,		Q1	Q2	Q3	Q4
			ed Instructions set Mode" for			Decode	No operation	Process Data	No operation
Words	61	1							
Cycles	6:	1			Exa	<u>mple:</u>	CLRWDT		
Q Cy	cle Activity:					Before Instruc		•	
•	Q1	Q2	Q3	Q4		WDT Cor After Instruction		?	
	Decode	Read register 'f'	Process Data	Write register 'f'		WDT Cor <u>WD</u> T Pos	unter = stscaler =	<b>00h</b> 0	
<u>Exam</u>	<u>ple:</u>	CLRF	FLAG_REG,	1		TO PD	=	1 1	
	Before Instruc FLAG_R After Instructio	EG = 5A	h						
-	FLAG_R		h						

COMF	Compleme	ent f		CPFSEQ	Compare	f with W, Skip	if f = W		
Syntax:	COMF f	{,d {,a}}		Syntax:	CPFSEQ	f {,a}			
Operands:	$0 \le f \le 255$			Operands:	$0 \leq f \leq 255$				
•	d ∈ [0,1]				a ∈ [0,1]				
	a ∈ [0,1]			Operation:	(f) – (W),				
Operation:	$\overline{f} \rightarrow dest$				• • • • •	skip if (f) = (W) (unsigned comparison)			
Status Affected:	N, Z			Status Affected		None			
Encoding:	0001	11da ff:	ff ffff						
Description:	The conten	ts of register 'f	" are	Encoding: Description:					
		ted. If 'd' is '0'		Description.		to the contents of	f data memory		
		. If 'd' is '1', th	e result is			an unsigned s			
stored back in register		in register 'f'.			If 'f' = W, th	nen the fetched	d instruction is		
		he Access Ba				and a NOP is e			
	GPR bank.	he BSR is use	d to select the		instead, mainstruction.	aking this a two	o-cycle		
		nd the extende					nk is selected.		
	in Indexed	ed, this instruc Literal Offset A	ddressing		lf 'a' is '1', ' GPR bank		ed to select the		
		ever $f \le 95$ (5)	,		lf 'a' is '0' a	and the extend	ed instruction		
		.2.3 "Byte-Or d Instruction				led, this instru			
		set Mode" for				Literal Offset /	•		
Words:	1					never f ≤ 95 (5 <b>).2.3 "Byte-Or</b>	,		
Cycles:	1					ed Instruction			
Q Cycle Activity:	·					set Mode" for			
Q Cycle Activity. Q1	Q2	Q3	Q4	Words:	1				
Decode	Read	Process	Write to	Cycles:	1(2)				
	register 'f'	Data	destination			Note: 3 cycles if skip and followed by a 2-word instruction.			
					-	a 2-word instru	ction.		
Example:	COMF	REG, 0, 0		Q Cycle Activ	-	02	04		
Before Instruc	tion			Q1 Decod	Q2 e Read	Q3 Process	Q4 No		
REG	= 13h			Decou	register 'f'	Data	operation		
After Instruction				If skip:	-5				
REG W	= 13h = ECh			Q1	Q2	Q3	Q4		
				No	No	No	No		
				operation operation	on operation owed by 2-word ir	operation	operation		
				li skip and ioi Q1	Q2	Q3	Q4		
				No	No	No	No		
				operatio	-	operation	operation		
				No	No	No	No		
				operatio	on operation	operation	operation		
				Example:	HERE	CPFSEQ REC	G, O		
					NEQUAL	:			
					EQUAL	:			
				Before In					
				PC / W	Address = HI = ?	CRE			
				REG					

REG	-	<i>!</i>	
After Instruction			
If REG	=	W;	
PC	=		(EQUAL)
If REG	≠	W;	
PC	=	Address	(NEQUAL)

CPFSGT	Compare f	with W, Skip	if f > W	CPF	SLT	Compare f	with W, Skip	if f < W		
Syntax:	CPFSGT	f {,a}		Synta	ax:	CPFSLT 1	<sup>;</sup> {,a}			
Operands:	$\begin{array}{l} 0 \leq f \leq 255 \\ a \in [0,1] \end{array}$			Oper	ands:	0 ≤ f ≤ 255 a ∈ [0,1]				
Operation:	(f) – (W), skip if (f) > (unsigned c	(W) comparison)		Oper	ation:	(f) – (W), skip if (f) <	(f) – (W), skip if (f) < (W) (unsigned comparison)			
Status Affected:	None			Statu	s Affected:	None				
Encoding:	0110	010a ff	ff ffff			0110	ff ffff			
Description:	Description: Compares the contents of data memory location 'f' to the contents of the W by performing an unsigned subtraction. If the contents of 'f' are greater than the			Enco Desc	ription:	Compares t location 'f' t	o the contents	f data memory of W by		
	contents of instruction i executed in two-cycle ir	WREG, then is discarded a istead, making istruction.	the fetched nd a NOP is g this a				an unsigned s nts of 'f' are le W, then the fe s discarded a stead, making istruction.	ess than the etched nd a NOP is		
	lf 'a' is '1', t GPR bank.	he BSR is use	nk is selected. ed to select the					nk is selected. d to select the		
	If 'a' is '0' and the extended set is enabled, this instructi in Indexed Literal Offset Ad mode whenever f ≤ 95 (5Fh Section 29.2.3 "Byte-Oriel Bit-Oriented Instructions		ction operates Addressing Fh). See <b>iented and</b>	Words: Cycles:		1 1(2) <b>Note:</b> 3 cycles if skip and followed by a 2-word instruction.				
	Literal Offs	set Mode" for	details.	QC	ycle Activity:	Q2	Q3	Q4		
Words:	1				Q1 Decode	Read	Process	No		
Cycles:		cycles if skip a		lf sk		register 'f'	Data	operation		
	by	a 2-word instr	ruction.		Q1	Q2	Q3	Q4		
Q Cycle Activity: Q1	Q2	Q3	Q4		No	No	No	No		
Decode	Read	Process	No		operation	operation	operation	operation		
200040	register 'f'	Data	operation	lf sk		d by 2-word in				
lf skip:					Q1	Q2	Q3	Q4		
Q1	Q2	Q3	Q4		No operation	No	No	No		
No	No	No	No		No	operation No	operation No	operation No		
operation If skip and followe	operation	operation	operation		operation	operation	operation	operation		
Q1	Q2	Q3	Q4							
No	No	No	No	Exan	nole:	HERE (	CPFSLT REG	. 1		
operation	operation	operation	operation		<u></u>		:			
No	No	No	No			LESS	:			
operation	operation	operation	operation		Before Instruc	tion				
Example:	HERE NGREATER GREATER	CPFSGT RI : :	EG, 0		PC W After Instruction If REG	= Ad = ?	dress (here	)		
Before Instruc PC	tion = Ad	dress (here	:)		PC If REG PC	= Ad ≥ W;	dress (LESS			
W After Instructio If REG PC If REG PC	> W; = Ad ≤ W;	dress (GREA					(1.20			

DAW	Decimal A	djust W Regis	ter	DECF	Decremen	tf	
Syntax:	DAW		Syntax:	DECF f{,	d {,a}}		
Operands:	None		Operands:	$0 \leq f \leq 255$	$0 \leq f \leq 255$		
Operation:		> 9] or [DC = 1 6 → W<3:0>;	], then		d ∈ [0,1] a ∈ [0,1]		
	else			Operation:	(f) – 1 $\rightarrow$ de	est	
	(W<3:0>) –	→ W<3:0>		Status Affected:	C, DC, N, 0	DV, Z	
	lf [W<7:4>	> 9] or [C = 1],	then	Encoding:	0000	01da ff	ff ffff
	· ,	$6 \rightarrow W < 7:4>;$		Description:	Decrement	register, 'f'. If	'd' is '0', the
	C = 1; else					red in W. If 'd	
	eise (W<7:4>) –	→ W<7:4>				red back in re	•
Status Affected:	C						ink is selected. ed to select the
Encoding:	0000	0000 000	00 0111		GPR bank.		
Description:					lf 'a' is '0' a	nd the extend	led instruction
2 000110100	DAW adjusts the 8-bit value in W, resulting from the earlier addition of two					,	ction operates
	```	ach in packed	,			Literal Offset . iever f ≤ 95 (5	•
	and produc result.	es a correct pa	acked BCD			.2.3 "Byte-O	,
Words:	1				Bit-Oriente	ed Instruction	ns in Indexed
Cycles:	1					set Mode" for	details.
-	I			Words:	1		
Q Cycle Activity: Q1	Q2	Q3	Q4	Cycles:	1		
Decode	Read	Process	Write	Q Cycle Activity:			
	register W	Data	W	Q1	Q2	Q3	Q4
				Decode	Read register 'f'	Process Data	Write to destination
Example 1:	DAW					Data	destination
Before Instru W	ction = A5h			Example:	DECF	CNT, 1, 0	)
С	= 0			Before Instru		- , , -	
DC After Instructi	= 0			CNT	= 01h		
W	= 05h			Z After Instructi	= 0		
C	= 1			CNT	= 00h		
DC	= 0			Z	= 1		
Example 2:							
Before Instru W	ction = CEh						
С	= 0						
DC After leastruct	= 0						
After Instructi W	on = 34h						
С	= 1						
DC	= 0						

DECFSZ	Decrement	t f, Skip if 0		DCF	SNZ	Decremen	t f, Skip if Not	t <b>O</b>
Syntax:	DECFSZ	f {,d {,a}}		Syn	tax:	DCFSNZ f {,d {,a}}		
Operands:	$0 \le f \le 255$ $d \in [0,1]$ $a \in [0,1]$	d ∈ [0,1]		Оре	rands:	0 ≤ f ≤ 255 d ∈ [0,1] a ∈ [0,1]		
Operation:	$(f) - 1 \rightarrow dest,$ skip if result = 0		Оре	peration: $(f) - 1 \rightarrow dest,$ skip if result $\neq 0$				
Status Affected:	None		Stat	us Affected:	None			
Encoding: Description:	ng: 0010 11da ffff ffff			oding: cription:	0100 11da ffff ffff The contents of register 'f' are decremented. If 'd' is '0', the result is placed in W. If 'd' is '1', the result is placed back in register 'f'.			
	which is alr and a NOP i it a two-cyc	le instruction.	is discarded stead, making			instruction discarded a	is not '0', the which is alread and a NOP is ea aking it a two-c	dy fetched is xecuted
	lf 'a' is '1', ti GPR bank.	he BSR is use	nk is selected. In to select the				he Access Bar he BSR is use	
If 'a' is '0' and the extended i set is enabled, this instruction in Indexed Literal Offset Add mode whenever f ≤ 95 (5Fh) Section 29.2.3 "Byte-Orien Bit-Oriented Instructions in Literal Offset Mode" for def		ction operates Addressing Fh). See <b>riented and</b> <b>hs in Indexed</b>			set is enab in Indexed mode wher Section 29 Bit-Oriente	nd the extend led, this instruct Literal Offset A never f ≤ 95 (5 0.2.3 "Byte-Or ed Instruction	ction operates Addressing Fh). See <b>iented and</b> is in Indexed	
Words:	1						set Mode" for	details.
Cycles:	1(2)			Wor		1		
		cles if skip ar a 2-word instru		Сус	les:		cycles if skip ar a 2-word instr	
Q Cycle Activity: Q1	Q2	Q3	Q4	Q	Cycle Activity:			
Decode	Read	Process	Write to		Q1	Q2	Q3	Q4
	register 'f'	Data	destination		Decode	Read	Process	Write to destination
If skip:	00	00	04	lfs	kip:	register 'f'	Data	destination
Q1 No	Q2 No	Q3 No	Q4 No		Q1	Q2	Q3	Q4
operation	operation	operation	operation		No	No	No	No
If skip and followe		struction:			operation	operation	operation	operation
Q1	Q2	Q3	Q4	lf s	kip and followe	d by 2-word in	struction:	
No	No	No	No		Q1	Q2	Q3	Q4
operation	operation	operation	operation		No	No	No	No
No operation	No operation	No operation	No operation		operation No	operation No	operation No	operation No
operation	operation	operation	operation		operation	operation	operation	operation
Example:	HERE	DECFSZ GOTO	CNT, 1, 1 LOOP	<u>Exa</u>	mple:	ZERO	DCFSNZ TEN : :	MP, 1, 0
Before Instruc					Before Instruc			
PC After Instructio		S (HERE)			TEMP	=	?	
CNT	= CNT – <sup>-</sup>	1			After Instruction	on		
If CNT PC	= 0; = Address	S (CONTINUE	7)		TEMP If TEMP	=	TEMP – 1, 0;	
If CNT	≠ 0;				PC	=	Address (	ZERO)
PC	= Address	S (HERE + 2	2)		If TEMP PC	≠ =	0; Address (1	NZERO)
								- ,

GOTO Unconditional Branch					
Syntax:	GOTO k				
Operands:	$0 \le k \le 104$	8575			
Operation:	$k \rightarrow PC<20$	D:1>			
Status Affected:	None				
Encoding: 1st word (k<7:0>) 2nd word(k<19:8>)	1110 1111	1111 k <sub>19</sub> kkk	k <sub>7</sub> kkk kkkk		kkkk <sub>0</sub> kkkk <sub>8</sub>
Description: GOTO allows an unconditional anywhere within entire 2-Mby range. The 20-bit value 'k' is PC<20:1>. GOTO is always a instruction.					e memory aded into
Words:	2				
Cycles:	2				
Q Cycle Activity:					
Q1	Q2	Q3			Q4
Decode	Read literal 'k'<7:0>,	No operat	ion	'k'·	ad literal <19:8>, te to PC
No operation	No operation	No operat	ion	ор	No eration
Example: After Instruction PC =	GOTO THE n Address (T				

INCF	Increment	f				
Syntax:	INCF f{,	d {,a}}				
Operands:	$0 \le f \le 255$ $d \in [0,1]$ $a \in [0,1]$					
Operation:	(f) + 1 $\rightarrow$ C	lest				
Status Affected:	C, DC, N,	OV, Z				
Encoding:	0010	10da	ffff	ffff		
Description:	The conter incrementer placed in V placed bac	ed. If 'd' is V. If 'd' is	'0', the '1', the	result is		
	,	the BSR i		is selected. to select the		
	set is enabling in Indexed mode whe Section 28	bled, this i Literal Of never f ≤ 9 9.2.3 "By ed Instru	nstruction fset Ad 95 (5Fh te-Orien ctions	). See nted and in Indexed		
Words:	1					
Cycles:	1					
Q Cycle Activity:						
Q1	Q2	Q3		Q4		
Decode	Read register 'f'	Proce Data		Write to destination		
Example:	INCF	CNT,	1, 0			
	tion					

INCFSZ	Increment	f, Skip if 0		INFSNZ	Inc	rement	f, Skip if Not	0
Syntax:	INCFSZ f	{,d {,a}}		Syntax:	INF	SNZ f	{,d {,a}}	
Operands:	$0 \le f \le 255$			Operands:		f ≤ 255		
	d ∈ [0,1]					[0,1] [0,1]		
	a ∈ [0,1]			Operation:		$1 \rightarrow de$	set	
Operation:	(f) + 1 $\rightarrow$ de	-		Operation.	.,	$r \perp \rightarrow ue$		
Status Affected:	skip if result = 0 None		Status Affected	: Nor	ne			
		11do ff	F.F. F.F.F.F.F.F.F.F.F.F.F.F.F.F.F.F.F.	Encoding:	C	100	10da ff	ff ffff
Encoding:	0011	11da ff		Description:			ts of register "	
Description:		ts of register '' d. If 'd' is '0', t					d. If 'd' is '0', t	
		/. If 'd' is '1', th			•		′. If 'd' is '1', th ‹ in register 'f'	
	placed bacl	k in register 'f'					is not '0', the	
		is '0', the nex					which is alread	
		eady fetched in	s discarded stead, making				nd a NOP is e	
		le instruction.	stead, making			ead, ma	king it a two-o	cycle
	lf 'a' is '0', t	he Access Ba	nk is selected.				he Access Ba	nk is selected
		he BSR is use	d to select the			-		ed to select the
	GPR bank.				GP	R bank.		
		nd the extend	ed instruction ction operates					ed instruction
		Literal Offset A	•				ed, this instru _iteral Offset /	ction operates
	mode when	ever f $\leq$ 95 (5	Fh). See				ever f $\leq$ 95 (5	0
		.2.3 "Byte-Or					.2.3 "Byte-Or	,
		ed Instruction set Mode" for						ns in Indexed
Words:	1			\A/anda.		eral Offs	et Mode" for	detalls.
Cycles:	1(2)			Words:	1	<b>`</b>		
Cycles.		cycles if skip a	nd followed	Cycles:	1(2 <b>No</b> t		cles if skip ar	nd followed
		a 2-word instr					2-word instru	
Q Cycle Activity:				Q Cycle Activit	iy:			
Q1	Q2	Q3	Q4	Q1		ຸ 22	Q3	Q4
Decode	Read	Process	Write to	Decode		ead	Process	Write to
lf skip:	register 'f'	Data	destination	lf skip:	regis	ster 'f'	Data	destination
Q1	Q2	Q3	Q4	II SKIP. Q1		22	Q3	Q4
No	No	No	No	No	1	10	No	No
operation	operation	operation	operation	operatio		ration	operation	operation
If skip and followe	2			If skip and follo	owed by 2	-word in	struction:	
Q1	Q2	Q3	Q4	Q1		ຸ 22	Q3	Q4
No operation	No operation	No operation	No operation	No		NO Notion	No operation	No
No	No	No	No	operatio No		ration No	No	operation No
operation	operation	operation	operation	operatio		ration	operation	operation
								•
Example:	NZERO	:	VT, 1, 0	Example:	HEF ZEF	20	INFSNZ REC	G, 1, 0
		:			NZI	ERO		
Before Instruc PC		(HERE)		Before Ins PC		Address	(HERE)	
After Instructi		) (REKE)		After Instr		, iaa 638	, (1111/12)	
CNT	= CNT + 7	1		REG	=	REG +	1	
IF CNT PC	= 0; = Address	S (ZERO)		lf RE PC	=	0; Address	(NZERO)	
If CNT	≠ 0;	G (NZERO)		If RE PC	G =	0;	(ZERO)	
PC								

IORLW	Inclusive C	OR Litera	al with W		
Syntax:	IORLW k				
Operands:	$0 \le k \le 255$				
Operation:	(W) .OR. k	$\rightarrow W$			
Status Affected:	N, Z				
Encoding:	0000	1001	kkkk	kkkk	
Description: The contents of W are ORed with the eight-bit literal 'k'. The result is placed in W.					
Words:	1				
Cycles:	1				
Q Cycle Activity:					
Q1	Q2	Q3		Q4	
Decode	Read literal 'k'	Proce Data		Vrite to W	
Example:	IORLW	35h			
Before Instruc W	tion = 9Ah				

BFh

=

After Instruction W

R W with f				
{,d {,a}}				
$\rightarrow$ dest				
00da fi	ff	ffff		
Inclusive OR W with register 'f'. If 'd' is '0', the result is placed in W. If 'd' is '1', the result is placed back in register 'f'.				
If 'a' is '0', the Access Bank is selected. If 'a' is '1', the BSR is used to select the GPR bank.				
If 'a' is '0' and the extended instruction set is enabled, this instruction operate in Indexed Literal Offset Addressing mode whenever f ≤ 95 (5Fh). See Section 29.2.3 "Byte-Oriented and Bit-Oriented Instructions in Indexed Literal Offset Mode" for details.				
Q3	(	Q4		
Process Data		te to nation		
	Data			

ample:	IOI	۲WF
Before Instructi	on	
RESULT	=	13h
W	=	91h
After Instruction	n	
RESULT	=	13h
W	=	93h

LFSF	ર	Load FSR					
Synta	ax:	LFSR f, k	[				
Oper	ands:	$\begin{array}{l} 0 \leq f \leq 2 \\ 0 \leq k \leq 408 \end{array}$	$\begin{array}{l} 0 \leq f \leq 2 \\ 0 \leq k \leq 4095 \end{array}$				
Oper	ation:	$k \to FSRf$					
Status Affected:		None					
Encoding:		1110 1111	1110 0000	00ff k <sub>7</sub> kkk	k <sub>11</sub> kkk kkkk		
Desc	ription:	The 12-bit file select					
Word	ls:	2					
Cycles:		2					
QC	ycle Activity:						
	Q1	Q2	Q3		Q4		
	Decode	Read literal 'k' MSB	Proce Data	a l	Write iteral 'k' MSB to FSRfH		
	Decode	Read literal	Proce		rite literal		
		ʻk' LSB	Data	a 'k'	to FSRfL		
<u>Exan</u>	n <u>ple:</u> After Instructio FSR2H		3ABh <b>3h</b>				
	FSR2L		Bh				

моу	′F	Move f					
Synta	ax:	MOVF f	{,d {,a}}				
Oper	ands:	0 ≤ f ≤ 255 d ∈ [0,1] a ∈ [0,1]	$\begin{array}{l} 0 \leq f \leq 255 \\ d \in [0,1] \end{array}$				
Oper	ration:	$f \to \text{dest}$					
Statu	is Affected:	N, Z					
Enco	oding:	0101	00da	ffff	ffff		
Desc	ription:	a destinations status of 'd placed in V placed bac	The contents of register 'f' are moved to a destination dependent upon the status of 'd'. If 'd' is '0', the result is placed in W. If 'd' is '1', the result is placed back in register 'f'. Location 'f' can be anywhere in the 256 byte bank				
		If 'a' is '0', If 'a' is '1', GPR bank	the BSR				
If 'a' is '0' and the extended instru- set is enabled, this instruction ope in Indexed Literal Offset Addressir mode whenever f ≤ 95 (5Fh). See Section 29.2.3 "Byte-Oriented an Bit-Oriented Instructions in Inde Literal Offset Mode" for details.					operates essing See ed and Indexed		
Word	ds:	1					
Cycle	es:	1					
QC	ycle Activity:						
	Q1	Q2	Q3	3	Q4		
	Decode	Read register 'f'	Proce Data		Write W		
Exan	•		EG, 0,	0			
	Before Instruct REG W	= 22 = Ff	2h Fh				
	After Instructic REG W	on = 22 = 22					

MOVFF		Move f to	F				
Syntax:		MOVFF f	s,f <sub>d</sub>				
Operands:		$\begin{array}{l} 0 \leq f_s \leq 4095 \\ 0 \leq f_d \leq 4095 \end{array}$					
Operation:		$(f_{s}) \rightarrow f_{d}$					
Status Affected	d:	None					
Encoding: 1st word (sour 2nd word (des	,	1100 1111	3				
Description:		moved to c Location of in the 4096 FFFh) and	The contents of source register, ' $f_s$ ', are moved to destination register ' $f_d$ '. Location of source ' $f_s$ ' can be anywhere in the 4096-byte data space (000h to FFFh) and location of destination ' $f_d$ ' can also be anywhere from 000h to				
		Either source or destination can be W (a useful special situation).					
		MOVFF is particularly useful for transferring a data memory location to a peripheral register (such as the transmit buffer or an I/O port).					
		The MOVFF instruction cannot use the PCL, TOSU, TOSH or TOSL as the destination register					
Words:		2					
Cycles:		2					
Q Cycle Activ	ity:						
Q1		Q2	Q3	5	Q4		
Decod	e	Read register 'f' (src)	Proce Data		No operation		
Decod		No operation No dummy read	No operat		Write register 'f' (dest)		
Example:		MOVFF	REG1, F	EG2			
Before In	structi						
REC	51 52	= 33 = 11					
After Inst REC REC	G1	n = 33 = 33					

MOVLB Move Literal to Low Nibble in BSF						
MOVLB k						
$0 \le k \le 255$	$0 \le k \le 255$					
$k \to BSR$						
None						
Encoding: 0000 0001 kkkk			k kkkk			
Bank Selec of BSR<7:4 regardless	t Registe > always	er (BSR s remai	t). The value ns '0'			
•						
1						
Q2	Q3		Q4			
Read literal 'k'			Write literal 'k' to BSR			
MOVLB	5					
	$\begin{array}{l} \text{MOVLB}  k \\ 0 \leq k \leq 255 \\ k \rightarrow \text{BSR} \\ \text{None} \\ \hline 0000 \\ \hline \text{The eight-b} \\ \text{Bank Selec} \\ \text{of BSR<7:4} \\ \text{regardless} \\ 1 \\ 1 \\ \hline \\ Q2 \\ \hline \\ \text{Read} \\ \text{literal 'k'} \\ \\ \text{MOVLB} \end{array}$	MOVLBk $0 \le k \le 255$ $k \rightarrow BSR$ None00000001The eight-bit literal ' Bank Select Register of BSR<7:4> always regardless of the variable11Q2Q3ReadProce Dataliteral 'k'Data	MOVLBk $0 \le k \le 255$ $k \rightarrow BSR$ None00000001kkkłThe eight-bit literal 'k' is loaBank Select Register (BSR of BSR<7:4> always remai regardless of the value of k11Q2Q3ReadProcess Data			

BSR Register = 02h After Instruction BSR Register = 05h

MOVLW		Move Lite	Move Literal to W					
Synta	ax:	MOVLW	k					
Oper	ands:	$0 \le k \le 25$	5					
Oper	ation:	$k\toW$						
Status Affected:		None						
Enco	oding:	0000	1110	kkkl	k	kkkk		
Desc	ription:	The eight-	The eight-bit literal 'k' is loaded into W.					
Word	ls:	1	1					
Cycle	es:	1	1					
QC	ycle Activity:							
	Q1	Q2	Q3	5	Q4			
	Decode	Read literal 'k'	Proce Data		W	rite to W		
Example:		MOVLW	5Ah					
	After Instructic W	n = 5Ah						

MOVWF	Move W to	f				
Syntax:	MOVWF	f {,a}				
Operands:	0 ≤ f ≤ 255 a ∈ [0,1]					
Operation:	$(W) \to f$					
Status Affected:	None					
Encoding:	0110	111a	fff	f ffff		
Description:	Move data Location 'f' 256-byte ba	can be a	0			
	,			to selected.		
	set is enabl in Indexed mode wher Section 29 Bit-Oriente	If 'a' is '0' and the extended instruction set is enabled, this instruction operates in Indexed Literal Offset Addressing mode whenever f ≤ 95 (5Fh). See Section 29.2.3 "Byte-Oriented and Bit-Oriented Instructions in Indexed Literal Offset Mode" for details.				
Words:	1					
Cycles:	1					
Q Cycle Activity:						
Q1	Q2	Q3		Q4		
Decode	Read	Proce		Write		
	register 'f'	Data	a	register 'f'		
Example:	MOVWF	REG, 0				
Before Instruc W REG After Instructio	= 4Fh = FFh					
W REG	= 4Fh = 4Fh					

MULLW	Multiply Li	iteral with W		MULWF	Multiply W w	/ith f	
Syntax:	MULLW	k		Syntax:	MULWF f{	,a}	
Operands:	$0 \le k \le 255$	5		Operands:	$0 \leq f \leq 255$		
Operation:	(W) x k $\rightarrow$	PRODH:PRO	DL		a ∈ [0,1]		
Status Affected:	None			Operation:	(W) x (f) $\rightarrow$ P	RODH:PROD	L
Encoding:	0000	1101 kk	kk kkkk	Status Affected:	None		
Description:	An unsigne	ed multiplication	n is carried	Encoding:	0000	001a ffi	ff ffff
	out betwee 8-bit literal placed in th	n the contents 'k'. The 16-bit	s of W and the result is RODL register	Description:	between the register file lo stored in the	PRODH:PRO	and the 16-bit result is DL register
	W is uncha	inged.			pair. PRODH W and 'f' are		nigh byte. Both
		e Status flags				Status flags ar	e affected.
	possible in		w nor Carry is . A Zero result ed.				nor Carry is A Zero result is
Words:	1				•		k is selected. If
Cycles: Q Cycle Activity:	1					BSR is used to	
Q1	Q2	Q3	Q4		If 'a' is '0' and	I the extended	instruction set
Decode	Read literal 'k'	Process Data	Write registers PRODH: PRODL		is enabled, th Indexed Liter whenever f ≤ Section 29.2 Bit-Oriented	al Offset Addr 95 (5Fh). See .3 "Byte-Orie Instructions t Mode" for d	operates in essing mode e inted and in Indexed
Example:	MULLW	0C4h		Words:	1		
Before Instruct W	ion = E2	2h		Cycles:	1		
PRODH	= ?	-11		Q Cycle Activity:			
PRODL After Instruction	= ? n			Q1	Q2	Q3	Q4
W PRODH PRODL	= E2 = AI = 08	Dh		Decode	Read register 'f'	Process Data	Write registers PRODH: PRODL
				Example:	MULWF	REG, 1	

Before Instruction		
W	=	C4h
REG	=	B5h
PRODH	=	?
PRODI	=	?
FRODE	-	:
After Instruction		
W	=	C4h
REG	=	B5h
PRODH	=	8Ah
PRODI		0.4
PRODL	=	94n

NEGF	Negate f				
Syntax:	NEGF f	{,a}			
Operands:	$\begin{array}{l} 0 \leq f \leq 255 \\ a \in [0,1] \end{array}$	5			
Operation:	$(\overline{f}) + 1 \rightarrow f$				
Status Affected:	N, OV, C,	DC, Z			
Encoding:	0110 110a ffff ffff				
Description:	Location 'f' is negated using two's complement. The result is placed in the data memory location 'f'.				
	If 'a' is '0', the Access Bank is selected. If 'a' is '1', the BSR is used to select the GPR bank.				
	If 'a' is '0' and the extended instruction set is enabled, this instruction operates in Indexed Literal Offset Addressing mode whenever f ≤ 95 (5Fh). See Section 29.2.3 "Byte-Oriented and Bit-Oriented Instructions in Indexed Literal Offset Mode" for details.				
Words:	1				
Cycles:	1				
Q Cycle Activity:					
Q1	Q2	Q3	3	Q4	

Q1	Q2	Q3	Q4
Decode	Read	Process	Write
	register 'f'	Data	register 'f'

Example:	NEGF	REG,	1
----------	------	------	---

Before Instru	ction			
REG	=	0011	1010	[3Ah]
After Instruct	ion			
REG	=	1100	0110	[C6h]

NOP		No Operation					
Synta	ax:	NOP					
Oper	ands:	None	None				
Oper	ation:	No operati	No operation				
Statu	s Affected:	None					
Encoding:		0000 1111	0000 xxxx			0000 xxxx	
Desc	ription:	No operation.					
Word	ls:	1	1				
Cycle	es:	1	1				
QC	ycle Activity:						
	Q1	Q2	Q	3		Q4	
	Decode	No operation	No operat		ор	No eration	

Example:

None.

POP		Pop Top o	f Return	Stack	I.	
Synta	ax:	POP				
Oper	ands:	None				
Oper	ation:	$(TOS) \rightarrow b$	it bucket			
Statu	s Affected:	None				
Enco	ding:	0000	0000	000	0	0110
Desc	ription:	The TOS v stack and is then becom was pushed This instruct the user to stack to inc	s discard nes the p d onto the ction is pr properly	ed. Th reviou e retui rovide mana	ne TO is va in sta d to e ge th	DS value lue that ack. enable ne return
Word	ls:	1				
Cycle	es:	1				
QC	ycle Activity:					
	Q1	Q2	Q3	3		Q4
	Decode	No operation	POP T valu		ор	No eration
Exan	nple:	POP GOTO	NEW			
	Before Instruction TOS Stack (1 level down) After Instruction TOS PC			)031A: )14332		
				)14332 NEW	2h	

PUSH	Push Top o	of Retur	n Stac	k	
Syntax:	PUSH				
Operands:	None				
Operation:	$(PC + 2) \rightarrow$	TOS			
Status Affected:	None				
Encoding:	0000	0000	000	0	0101
Description:	The PC + 2 the return s value is pus This instruc software sta then pushin	tack. Th shed dov tion allo ack by m	e previ vn on t ws imp iodifyir	ious he s olem ng To	TOS stack. enting a OS and
Words:	1				
Cycles:	1				
Q Cycle Activity:					
Q1	Q2	Q	3		Q4
Decode	PUSH PC + 2 onto return stack	No operat		ор	No eration
Example:	PUSH				
Example: Before Instru TOS PC			345Ah )124h		

RCA	LL	Relative C	all				
Synta	ax:	RCALL n					
Oper	ands:	-1024 ≤ n ≤	≤ 1023				
Oper	ation:	· · ·	$(PC) + 2 \rightarrow TOS,$ (PC) + 2 + 2n $\rightarrow$ PC				
Statu	s Affected:	None					
Enco	ding:	1101	1nnn	nnnn	nnnn		
Desc	ription:	from the cu address (P stack. Then number '2r have increa instruction, PC + 2 + 2	Subroutine call with a jump up to 1K from the current location. First, return address (PC + 2) is pushed onto the stack. Then, add the 2's complement number '2n' to the PC. Since the PC will have incremented to fetch the next instruction, the new address will be PC + 2 + 2n. This instruction is a two-cycle instruction.				
Word	ls:	1	1				
Cycle	es:	2					
QC	ycle Activity:						
	Q1	Q2	Q3	5	Q4		
	Decode	Read literal 'n' PUSH PC	Proce Data		rite to PC		
		to stack					
	No	No	No		No		

RES	ET	Reset				
Synta	ax:	RESET				
Oper	ands:	None				
Oper	ation:	Reset all re affected by	·		•	at are
Statu	s Affected:	All				
Enco	ding:	0000	0000	111	.1	1111
Desc	ription:	This instruction execute a				•
Word	ls:	1				
Cycle	es:	1				
QC	ycle Activity:					
	Q1	Q2	Q3	8		Q4
	Decode	Start	No			No
		reset	operat	ion	ор	eration

Example:

•	Instri	iction	

After Instruction	
Registers =	Reset Value
Flags* =	Reset Value

RESET

Example: HERE RCALL Jump

operation

operation

operation

Before Instruction

operation

PC = Address (HERE) After Instruction PC = TOS = Address (Jump) Address (HERE + 2)

RET	FIE	Return from	m Interrupt		
Synta	ax:	RETFIE {s	;}		
Oper	Operands: $s \in [0,1]$				
Operation: $(TOS) \rightarrow PC,$ $1 \rightarrow GIE/GIEH \text{ or PEIE/GIEL};$ if $s = 1,$ $(WS) \rightarrow W,$ $(STATUSS) \rightarrow STATUS,$ $(BSRS) \rightarrow BSR,$ PCLATU, PCLATH are unchanged					
Statu	s Affected:	GIE/GIEH,	PEIE/GIEL.		
Enco	ding:	0000	0000 00	01 000s	
Desc	ription:	and Top-of- the PC. Inte setting eithe Global Inter contents of STATUSS a their corres STATUS an	n interrupt. Sta Stack (TOS) is errupts are ena- er the high or l rupt Enable bi the shadow re- and BSRS are ponding regist of BSR. If 's' = gisters occurs.	s loaded into abled by low-priority it. If 's' = 1, the egisters WS, loaded into ters W, = 0, no update	
Word	ls:	1	•		
Cycle	es:	2			
Q C	ycle Activity:				
	Q1	Q2	Q3	Q4	
	Decode	No operation	No operation	POP PC from stack Set GIEH or GIEL	
	No	No	No	No	
	operation	operation	operation	operation	
Exan	After Interrupt PC W BSR STATUS	RETFIE :	= TOS = WS = BSRS = STATL = 1	JSS	

RETL	_W	Return Lite	eral to W				
Synta	ax:	RETLW k	RETLW k				
Opera	ands:	$0 \le k \le 255$	$0 \le k \le 255$				
Opera	ation:	$k \rightarrow W$ , (TOS) $\rightarrow P$ PCLATU, P		re un	chang	ed	
Status	s Affected:	None					
Enco	ding:	0000	1100	kkk	k	kkkk	
Desc	ription:	W is loaded Program Co of the stack high addres unchanged	ounter is lo (the retui is latch (P	oadeo rn ad	d from dress)	the top . The	
Word	s:	1					
Cycle	es:	2					
Q C	cle Activity:						
	Q1	Q2	Q3		C	Q4	
Г	Decode	Read	Proces	2	POF	P PC	
	Decode	literal 'k'	Data	.0		stack, to W	
	No		Data	.0	write	-	
·		literal 'k'			write N	to W	
Exam	No operation	literal 'k' No operation	No operatio	on	write N	to W lo	
	No operation	literal 'k' No operation ; W contai	No operatio	on	write N	to W lo	
	No operation	<pre>iteral 'k' No operation ; W contai ; offset v</pre>	No operation ns tabl ralue	on	write N	to W lo	
	No operation	<pre>iteral 'k' No operation ; W contai ; offset v</pre>	No operation ns tabl value as	on	write N	to W lo	
:	No operation nple: CALL TABLE	<pre>iteral 'k' No operation ; W contai ; offset v ; W now hat</pre>	No operation ns tabl value as	on	write N	to W lo	
( : TABL	No operation nple: CALL TABLE	<pre>iteral 'k' No operation ; W contai ; offset v ; W now ha ; table va </pre>	No operation ns table value as alue	on	write N	to W lo	
: TABL	No operation nple: CALL TABLE	<pre>iteral 'k' No operation ; W contai; ; offset v; ; W now ha ; table va ; W = offs</pre>	No operation ns table value as alue set	on	write N	to W lo	
: TABL J	No operation nple: CALL TABLE E ADDWF PCL	<pre>iteral 'k' No operation ; W contai ; offset v ; W now ha ; table va ; W = offs</pre>	No operation ns table value as alue set	on	write N	to W lo	

Before Instruction

Belore instru	CUON	
W	=	07h
After Instruct	ion	
W	=	value of kn

RET	RETURN Return from Subroutine						
Synta	ax:	RETURN	{s}				
Oper	ands:	s ∈ [0,1]					
Oper	ation:	$(TOS) \rightarrow PC;$ if s = 1, $(WS) \rightarrow W,$ $(STATUSS) \rightarrow STATUS,$ $(BSRS) \rightarrow BSR,$ PCLATU, PCLATH are unchanged					
Statu	is Affected:	None					
Enco	oding:	0000	0000	0001	001s		
Desc	rription:	otion: Return from subroutine. The stack is popped and the top of the stack (TOS) is loaded into the Program Counter. If 's'= 1, the contents of the shadow registers WS, STATUSS and BSRS are loaded into their corresponding registers W, STATUS and BSR. If 's' = 0, no update of these registers occurs.			ack (TOS) ounter. If adow BSRS are ng SR. If		
Word	ls:	1	1				
Cycle	es:	2					
QC	ycle Activity:						
	Q1	Q2	Q	3	Q4		
	Decode	No operation	Proce Data		POP PC om stack		
	No operation	No operation	No operat		No operation		
<u>Exan</u>	nple:	RETURN					

After Instruction: PC = TOS

$\begin{array}{l} 0 \leq f \leq 255\\ d \in [0,1]\\ a \in [0,1]\\ (f < n >) \rightarrow de\\ (f < 7 >) \rightarrow C,\\ (C) \rightarrow dest < \\ C, N, Z\\ \hline 0011\\ \hline The content\\ one bit to the is '0', the rest''.\\ If 'a' is '0', the rest''.\\ If 'a' is '1', the rest''.\\ If 'a' is '0', the is '1', the GPR bank.\\ If 'a' is '0' are transformed by the transformation of transformation of transformation of the transformation of transfor$	01da s of registe e left throu he result is sult is store he Access he BSR is i	ffff er 'f' are gh the C placed i ed back i Bank is :	arry flag. n W. If 'd' n register selected.
$\begin{array}{l} d \in [0,1] \\ a \in [0,1] \\ (f < n >) \rightarrow de \\ (f < 7 >) \rightarrow C, \\ (C) \rightarrow dest < \\ C, N, Z \\ \hline \\ \hline \\ 0011 \\ \hline \\ The content \\ one bit to the \\ If 'd' is '0', th \\ is '1', the rest \\ 'f. \\ If 'a' is '0', th \\ If 'a' is '0', th \\ If 'a' is '0', th \\ If 'a' is '0' and \\ I$	01da s of registe e left throu he result is sult is store he Access he BSR is i	ffff er 'f' are gh the C placed i ed back i Bank is :	rotated arry flag. n W. If 'd' n register selected.
$a \in [0,1]$ $(f < n >) \rightarrow de$ $(f < 7 >) \rightarrow C,$ $(C) \rightarrow dest < C, N, Z$ $\boxed{0011}$ The content one bit to the bit to the fid' is '0', the result of the fid' is '0', the result of the fid' is '0', the fid' is '1', the GPR bank. If 'a' is '0' ar	01da s of registe e left throu he result is sult is store he Access he BSR is i	ffff er 'f' are gh the C placed i ed back i Bank is :	rotated carry flag. n W. If 'd' n register selected.
$(f < n >) \rightarrow de$ $(f < 7 >) \rightarrow C$ , $(C) \rightarrow dest < C$ , N, Z 0011 The content one bit to the If 'd' is '0', the rest 'f'. If 'a' is '0', the If 'a' is '1', the GPR bank. If 'a' is '0' ar	01da s of registe e left throu he result is sult is store he Access he BSR is i	ffff er 'f' are gh the C placed i ed back i Bank is :	rotated carry flag. n W. If 'd' n register selected.
$(f<7>) \rightarrow C,$ $(C) \rightarrow dest<$ C, N, Z 0011 The content one bit to the If 'd' is '0', the rest 'f'. If 'a' is '0' ar 'f'. If 'a' is '0' ar 'f'.	01da s of registe e left throu he result is sult is store he Access he BSR is i	ffff er 'f' are gh the C placed i ed back i Bank is :	rotated carry flag. n W. If 'd' n register selected.
0011 The content one bit to the If 'd' is '0', th is '1', the res 'f'. If 'a' is '0', th If 'a' is '1', th GPR bank. If 'a' is '0' ar	is of registe e left throu he result is sult is store he Access he BSR is i	er 'f' are igh the C placed i ed back i Bank is	rotated carry flag. n W. If 'd' n register selected.
The content one bit to the If 'd' is '0', th is '1', the res 'f'. If 'a' is '0', th If 'a' is '0', th If 'a' is '1', th GPR bank. If 'a' is '0' ar	is of registe e left throu he result is sult is store he Access he BSR is i	er 'f' are igh the C placed i ed back i Bank is	rotated carry flag. n W. If 'd' n register selected.
one bit to the If 'd' is '0', th is '1', the res 'f'. If 'a' is '0', th If 'a' is '1', th GPR bank. If 'a' is '0' an	e left throu ne result is sult is store ne Access ne BSR is i	gh the C placed i ed back i Bank is	arry flag. n W. If 'd' n register selected.
If 'a' is '1', th GPR bank. If 'a' is '0' ar	ne BSR is i		
in Indexed L mode when Section 29. Bit-Oriente	ever f ≤ 95 2.3 "Byte- d Instruct et Mode"	et Addre (5Fh). S Oriente ions in I for detail	ssing See d and ndexed
	J <u>L</u>		
1			
1			
			Q4
register 'f'	Data	-	Vrite to stination
RLCF	REG,	0, 0	
	0110		
n			
	0110 1100		
	Section 29. Bit-Oriente Literal Offs C 1 1 1 Q2 Read register 'f' RLCF tion = 1110 n = 1110	Section 29.2.3 "Byte- Bit-Oriented Instructi Literal Offset Mode" C reg 1 1 2 Q2 Q3 Read Proces register 'f' Data RLCF REG, tion = 1110 0110 n = 1110 0110	1 1 Q2 Q3 Read Process V register 'f' Data des RLCF REG, 0, 0 tion = 1110 0110 = 0 n = 1110 0110

RLNCF	Rotate Lef	t f (No Carry	)	
Syntax:	RLNCF	f {,d {,a}}		
Operands:	0 ≤ f ≤ 255 d ∈ [0,1] a ∈ [0,1]			
Operation:	$(f \le n >) \rightarrow d$ $(f \le 7 >) \rightarrow d$	est <n +="" 1="">, est&lt;0&gt;</n>		
Status Affected:	N, Z			
Encoding:	0100	01da ff	ff ffff	
Description:	The contents of register 'f' are rotated one bit to the left. If 'd' is '0', the result is placed in W. If 'd' is '1', the result is stored back in register 'f'.			
	,	he BSR is us	ank is selected. ed to select the	
	in Indexed mode when Section 29 Bit-Oriente	Literal Offset never f ≤ 95 ( .2.3 "Byte-C	5Fh). See Priented and Ins in Indexed	
	_	register	f 🖣	
Words:	1			
Cycles:	1			
Q Cycle Activity:				
Q1	Q2	Q3	Q4	
Decode	Read register 'f'	Process Data	Write to destination	
Example:	RLNCF	REG, 1,	0	
Before Instruct REG After Instruction	= 1010 1	011		
REG	-	111		

RRCF	Rotate Rigi	ht f through	
Syntax:	RRCF f{,	d {,a}}	
Operands:	$\begin{array}{l} 0 \leq f \leq 255 \\ d  \in  [0,1] \\ a  \in  [0,1] \end{array}$		
Operation:	$(f < n >) \rightarrow de$ $(f < 0 >) \rightarrow C$ , $(C) \rightarrow dest <$	,	
Status Affected:	C, N, Z		
Encoding:	0011	00da ff	ff ffff
Description:	one bit to th flag. If 'd' is	e right throug	is placed in W
			ank is selected ed to select the
	set is enabl		led instruction iction operates Addressing
	set is enabling in Indexed I mode when Section 29. Bit-Oriente Literal Offs	ed, this instru Literal Offset ever $f \le 95$ (5 2.3 "Byte-O d Instruction set Mode" for	Addressing SFh). See riented and ns in Indexed r details.
	set is enable in Indexed I mode when Section 29 Bit-Oriente	ed, this instru Literal Offset ever f ≤ 95 (5 .2.3 "Byte-O ed Instruction	Addressing SFh). See riented and ns in Indexed r details.
Words:	set is enabling in Indexed I mode when Section 29. Bit-Oriente Literal Offs	ed, this instru Literal Offset ever $f \le 95$ (5 2.3 "Byte-O d Instruction set Mode" for	Addressing SFh). See riented and ns in Indexed r details.
Words: Cycles:	set is enablin Indexed I mode when Section 29. Bit-Oriente Literal Offs	ed, this instru Literal Offset ever $f \le 95$ (5 2.3 "Byte-O d Instruction set Mode" for	Addressing SFh). See riented and ns in Indexed r details.
	set is enablin Indexed I mode when Section 29. Bit-Oriente Literal Offs	ed, this instru Literal Offset ever $f \le 95$ (5 2.3 "Byte-O d Instruction set Mode" for	Addressing SFh). See riented and ns in Indexed r details.
Cycles: Q Cycle Activity: Q1	set is enablin Indexed I mode when Section 29. Bit-Oriente Literal Offs 1 1 2 2	ed, this instru Literal Offset ever f ≤ 95 (5 2.3 "Byte-O d Instruction set Mode" for registe	Q4
Cycles: Q Cycle Activity:	set is enable in Indexed I mode when Section 29. Bit-Oriente Literal Offs 1 1 1 2 Q2 Read	ed, this instru Literal Offset ever f ≤ 95 (5 .2.3 "Byte-O d Instruction set Mode" for registe	Q4
Cycles: Q Cycle Activity: Q1	set is enablin Indexed I mode when Section 29. Bit-Oriente Literal Offs 1 1 2 2	ed, this instru Literal Offset ever f ≤ 95 (5 2.3 "Byte-O d Instruction set Mode" for registe Q3 Process	Q4
Cycles: Q Cycle Activity: Q1	set is enable in Indexed I mode when Section 29. Bit-Oriente Literal Offs 1 1 1 2 Q2 Read	ed, this instru Literal Offset ever f ≤ 95 (5 2.3 "Byte-O d Instruction set Mode" for registe Q3 Process	Q4 Write to destination
Cycles: Q Cycle Activity: Q1 Decode Example: Before Instruc	set is enable in Indexed I mode when Section 29. Bit-Oriente Literal Offs I 1 1 1 Q2 Read register 'f' RRCF tion	ed, this instru Literal Offset. iever f ≤ 95 (5 .2.3 "Byte-O d Instruction set Mode" for registe Q3 Process Data REG, 0,	Q4 Write to destination
Cycles: Q Cycle Activity: Q1 Decode Example: Before Instruc REG	set is enable in Indexed I mode when Section 29. Bit-Oriente Literal Offs I 1 1 2 Q2 Read register 'f' RRCF tion = 1110 0	ed, this instru Literal Offset. iever f ≤ 95 (5 .2.3 "Byte-O d Instruction set Mode" for registe Q3 Process Data REG, 0,	Q4 Write to destination
Cycles: Q Cycle Activity: Q1 Decode Example: Before Instruc	set is enable in Indexed I mode when Section 29. Bit-Oriente Literal Offs C 1 1 1 2 Read register 'f' RRCF tion = 1110 0	ed, this instru Literal Offset. iever f ≤ 95 (5 .2.3 "Byte-O d Instruction set Mode" for registe Q3 Process Data REG, 0,	Q4 Write to destination
Cycles: Q Cycle Activity: Q1 Decode Example: Before Instruc REG C	set is enable in Indexed I mode when Section 29, Bit-Oriente Literal Offs C 1 1 1 2 Read register 'f' RRCF tion = 1110 0 on = 1110 0	ed, this instru Literal Offset. iever f ≤ 95 (5 .2.3 "Byte-O d Instruction set Mode" for registe Q3 Process Data REG, 0,	Q4 Write to destination

Rotate R	ight f (No Carry)		
RRNCF	f {,d {,a}}		
$0 \le f \le 25$ $d \in [0,1]$ $a \in [0,1]$	5		
N, Z			
0100	00da ffff	ffff	
The contents of register 'f' are rotated one bit to the right. If 'd' is '0', the result is placed in W. If 'd' is '1', the result is placed back in register 'f'.			
If 'a' is '0', the Access Bank will be selected, overriding the BSR value. is '1', then the bank will be selected per the BSR value.			
set is ena in Indexe mode whe Section 2 Bit-Orien	abled, this instruction o d Literal Offset Addres enever f ≤ 95 (5Fh). Se 29.2.3 "Byte-Oriented ited Instructions in In	perates sing ee and dexed	
Г	► register f	>1	
1			
1			
Q2	Q3 (	Q4	
Read		te to	
register 'f'	Data desti	nation	
RRNCF	REG, 1, 0		
= 1101	0111		
	1011		
RRNCF	REG, 0, 0		
ction = ?			
= ?	0111		
(	RRNCF $0 \le f \le 25$ $d \in [0,1]$ $a \in [0,1]$ $a \in [0,1]$ $(f < 0^{>}) \rightarrow$ N, Z 0100 The conte one bit to is placed ba If 'a' is '0' selected, is '1', then per the B If 'a' is '0' set is ena in Indexe mode wh Section 2 Bit-Orient Literal O 1 1 Q2 Read register 'f' RRNCF ction = 1101 Dn = 1110	$0 \le f \le 255$ $d \in [0,1]$ $a \in [0,1]$ $(f(n>) \rightarrow dest < n - 1>, (f(0>) \rightarrow dest < 7>$ N, Z $\boxed{0100  00da  ffff}$ The contents of register 'f' are re one bit to the right. If 'd' is '0', the is placed in W. If 'd' is '1', the re placed back in register 'f'. If 'a' is '0', the Access Bank will selected, overriding the BSR value is '1', then the bank will be selected, overriding the BSR value. If 'a' is '0' and the extended inst set is enabled, this instruction o in Indexed Literal Offset Address mode whenever f ≤ 95 (5Fh). Se Section 29.2.3 "Byte-Oriented Bit-Oriented Instructions in In Literal Offset Mode" for details $\boxed{Pocess  Wri}$ $1$ $1$ $1$ $Q2  Q3  Q3$ $\boxed{Read  Process  Wri}$ $RRNCF  REG, 1, 0$ etion $= 1101  0111$	

SETF	Set f			
Syntax:	SETF f{,a	a}		
Operands:	0 ≤ f ≤ 255 a ∈ [0,1]			
Operation:	$FFh\tof$			
Status Affected:	None			
Encoding:	0110	100a -	ffff	ffff
Description:	The conten are set to F		ecified I	register
	lf 'a' is '0', t If 'a' is '1', t GPR bank.			
	If 'a' is '0' a set is enabl in Indexed mode wher Section 29 Bit-Oriente Literal Offs	ed, this ins Literal Offse ever f ≤ 95 .2.3 "Byte- d Instruction	truction et Addre (5Fh). Oriente ons in	operates essing See ed and Indexed
Words:	1			
Cycles:	1			
Q Cycle Activity:				
Q1	Q2	Q3		Q4
Decode	Read register 'f'	Process Data		Write gister 'f'
Example: Before Instruc REG After Instructio REG	= 5A		1	

SLEEP	Enter Slee	Enter Sleep Mode					
Syntax:	SLEEP	SLEEP					
Operands:	None	None					
Operation:							
Status Affected:	TO, PD	TO, PD					
Encoding:	0000	0000 0000 0000 0					
Description:	cleared. T is set. The	The Power-Down status bit (PD) is cleared. The Time-out status bit (TO) is set. The Watchdog Timer and its postscaler are cleared.					
		The processor is put into Sleep mode with the oscillator stopped.					
Words:	1	1					
Cycles:	1	1					
Q Cycle Activity:							
Q1	Q2	Q3	Q4				
Decode	No operation	Process Data	Go to Sleep				
Example:	SLEEP						
Before Instruct TO = PD = After Instruction TO = TO =	? ? on 1 †						
PD = 0 † If WDT causes wake-up, this bit is cleared.							

SUBFWB f	{,d {,a}}			
0 ≤ f ≤ 255				
d ∈ [0,1]				
0101 01da ffff ffff				
Subtract register 'f' and Carry flag (borrow) from W (2's complement method). If 'd' is '0', the result is stored in W. If 'd' is '1', the result is stored in register 'f'.				
If 'a' is '0', the Access Bank is selected. If 'a' is '1', the BSR is used to select the GPR bank.				
set is enabled Indexed Liter whenever f ≤ Section 29.2 Bit-Oriented	d, this instruction al Offset Addre 95 (5Fh). See .3 "Byte-Orient Instructions	on operates in essing mode nted and in Indexed		
1				
1				
Q2	Q3	Q4		
Read	Process	Write to		
register 'f'	Data	Write to destination		
register 'f' SUBFWB				
register 'f' SUBFWB	Data			
register 'f' SUBFWB	Data			
register 'f' SUBFWB ction = 3	Data			
register 'f' SUBFWB stion = 3 = 2 = 1 on = FF	Data			
register 'f' SUBFWB stion = 3 = 2 = 1 on = FF	Data			
register 'f' SUBFWB stion = 3 = 2 = 1 on = FF = 2 = 0 = 0	Data REG, 1, 0	destination		
register 'f' SUBFWB stion = 3 = 2 = 1 on = FF = 2 = 0 = 0 = 1 ; r	Data REG, 1, 0	destination		
register 'f' SUBFWB Stion = 3 = 2 = 1 on = FF = 2 = 0 = 0 = 0 = 1 ; r SUBFWB	Data REG, 1, 0	destination		
register 'f' SUBFWB Stion = 3 = 2 = 1 SUBFWB SUBFWB Stion	Data REG, 1, 0	destination		
register 'f SUBFWB stion = 3 = 2 = 1 on = FF = 2 = 0 = 0 = 1; r SUBFWB stion = 2 = 5	Data REG, 1, 0	destination		
register 'f SUBFWB stion = 3 = 2 = 1 on = FF = 2 = 0 = 0 = 1; r SUBFWB stion = 2 = 5 = 1	Data REG, 1, 0	destination		
register 'f' SUBFWB stion = 3 = 2 = 1 con = FF = 2 = 0 = 0 = 1 ; r SUBFWB stion = 2 = 5 = 1 con	Data REG, 1, 0	destination		
register 'f' SUBFWB stion = 3 = 2 = 1 on = FF = 2 = 0 = 0 = 0 = 1 ; r SUBFWB stion = 2 = 1 on = 1 = 1 on = 1 = 1 on = 2 = 1 on = 0 = 0 = 1 ; r SUBFWB stion = 2 = 1 on = 2 = 0 = 1 ; r SUBFWB stion = 2 = 1 on = 2 = 1 on = 1 ; r SUBFWB stion = 2 = 1 ; r SUBFWB stion = 2 = 1 ; r SUBFWB = 3 = 1 ; r = 3 ; r	Data REG, 1, 0	destination		
register 'f' SUBFWB stion = 3 = 2 = 1 con = FF = 2 = 0 = 0 = 1 ; r SUBFWB stion = 2 = 5 = 1 con	Data REG, 1, 0	destination		
register 'f' SUBFWB stion = 3 = 2 = 1 on = FF = 2 = 0 = 0 = 0 = 1; r SUBFWB stion = 2 = 1 on = 2 = 3 = 1 = 0 = 1 = 0; r = 2 = 1 = 0; r = 1 = 1 = 1 = 0; r = 1 = 1 = 1 = 0; r = 1 = 1 = 1 = 0 = 1 = 1 = 1 = 0 = 1 = 1 = 1 = 0 = 1 = 1 = 1 = 1 = 1 = 0 = 1 = 1 = 1 = 1 = 1 = 1 = 1 = 1	Data REG, 1, 0	destination /e		
register 'f' SUBFWB Stion = 3 = 2 = 1 on = FF = 2 = 0 = 0 = 1; r SUBFWB Stion = 2 = 1 on = 2 = 1 ; r SUBFWB SUBFWB	Data REG, 1, 0 esult is negativ REG, 0, 0	destination /e		
register 'f' SUBFWB Stion = 3 = 2 = 1 on = FF = 2 = 0 = 0 = 1 ; r SUBFWB SUBFWB Stion = 2 = 1 = 1 on = 2 = 1 ; r SUBFWB Stion = 0 = 0 = 0 = 0 = 1 ; r SUBFWB Stion = 2 = 0 = 0 = 0 = 0 = 0 = 0 = 0 = 1 ; r SUBFWB Stion = 2 = 0 = 0 = 0 = 0 = 1 ; r SUBFWB Stion = 2 = 0 = 0 = 1 ; r SUBFWB Stion = 2 = 0 = 0 = 1 ; r SUBFWB Stion = 2 = 0 = 0 = 0 = 0 = 1 ; r SUBFWB Stion = 2 = 0 = 0 = 0 = 0 = 0 = 1 ; r SUBFWB Stion = 2 = 0 = 0 = 0 = 0 = 0 = 0 = 1 ; r SUBFWB Stion = 2 = 0 = 0 = 0 = 0 = 0 = 0 = 0 = 0	Data REG, 1, 0 esult is negativ REG, 0, 0	destination /e		
register 'f' SUBFWB Stion = 3 = 2 = 1 Dn = FF = 2 = 0 = 0 = 1; r SUBFWB SUBFWB SUBFWB SUBFWB SUBFWB SUBFWB	Data REG, 1, 0 esult is negativ REG, 0, 0	destination /e		
register 'f' SUBFWB Stion = 3 = 2 = 1 on = FF = 2 = 0 = 0 = 1; r SUBFWB SUBFWB on = 2 = 5 = 1 on = 0; r SUBFWB SUBFWB	Data REG, 1, 0 esult is negativ REG, 0, 0	destination /e		
register 'f' SUBFWB Stion = 3 = 2 = 1 Dn = FF = 2 = 0 = 0 = 1; r SUBFWB SUBFWB SUBFWB SUBFWB SUBFWB SUBFWB	Data REG, 1, 0 esult is negativ REG, 0, 0	destination /e		
register 'f' SUBFWB Stion = 3 = 2 = 1 on = FF = 2 = 0 = 0 = 1 ; r SUBFWB SUBFWB Stion = 2 = 3 = 1 = 0 = 0 ; r SUBFWB Stion = 1 = 2 = 0 on = 0 = 2 = 0 on = 2 = 0 on = 0 = 2 = 0 on = 2 = 0 on = 0 = 0 = 0 = 0 = 0 = 0 = 0 = 0 = 0 = 0	Data REG, 1, 0 esult is negativ REG, 0, 0	destination /e		
register 'f' SUBFWB Stion = 3 = 2 = 1 SUBFWB SUBFWB SUBFWB SUBFWB SUBFWB SUBFWB SUBFWB SUBFWB SUBFWB = 2 = 3 = 1 0 = 0; r SUBFWB	Data REG, 1, 0 esult is negativ REG, 0, 0	destination /e		
	$\begin{array}{l} 0 \leq f \leq 255\\ d \in [0,1]\\ a \in [0,1]\\ (W) - (f) - (\overline{C}\\ N, OV, C, DC\\ \hline 0101 \\ \\ \\ \\ \\ \\ \\ \\ \\ \\ \\ \\ \\ \\ \\ \\ \\ $	$0 \le f \le 255$ $d \in [0,1]$ $a \in [0,1]$ $(W) - (f) - (\overline{C}) \rightarrow dest$ N, OV, C, DC, Z $\boxed{0101  01da  ffff:}$ Subtract register 'f' and Car (borrow) from W (2's complemented). If 'd' is '0', the result is seregister 'f'. If 'd' is '1', the result is seregister 'f'. If 'a' is '0', the Access Bank 'a' is '1', the BSR is used to GPR bank. If 'a' is '0' and the extended set is enabled, this instruction Indexed Literal Offset Addrewhenever f \le 95 (5Fh). See Section 29.2.3 "Byte-Oriented Instructions Literal Offset Mode" for det 1 1		

Subtract f from W with Borrow

SUBFWB

SUBLW	:	Subtract W from Literal					
Syntax:		SUBLW	k	(			
Operands:		$0 \le k \le 255$					
Operation:	I	$k - (W) \rightarrow W$					
Status Affected:	I	N, OV, C, DC, Z					
Encoding:		0000 1000 kkkk kkkk				kkkk	
Description:				acted from			
Words:		1					
Cycles:		1					
Q Cycle Activity:							
Q1		Q2		Q3			Q4
Decode		Read teral 'k'		Proce: Data		۷	Vrite to W
Example 1:		SUBLW	0	2h	-		
Before Instruc	tion						
W C	=	01h ?					
After Instructio		041					
W C	=	01h 1	; I	result is p	ositiv	/e	
Z N	=	0 0					
Example 2:		SUBLW	0	2h			
Before Instruc	Before Instruction						
W C	=	02h ?					
After Instructio	n	•					
W C	=	<b>00h</b> 1		result is z	zoro		
Z	=	1	, '		2010		
N Example 2:	=	0	~	Ob			
Example 3:		SUBLW	U	2h			
Before Instruc W	uon =	03h					
Ċ	=	?					
After Instructio W	on =	FFh		(2's com	oleme	ent)	
Ċ	=	0	; I	result is r	hegati	ve	
Ž N	=	0 1					

SUBWF	Subtract	W from f			
Syntax:	SUBWF f {,d {,a}}				
Operands:	0 ≤ f ≤ 255				
·	d ∈ [0,1]				
	a ∈ [0,1]				
Operation:	$(f) - (W) \rightarrow dest$				
Status Affected:	N, OV, C, DC, Z				
Encoding:	0101 11da ffff ffff				
Description:	Subtract W from register 'f' (2's complement method). If 'd' is '0', the result is stored in W. If 'd' is '1', the result is ctored back in granter 'f'				
	is stored back in register 'f'.				
	If 'a' is '0', the Access Bank is selected. If 'a' is '1', the BSR is used to select the GPR bank.				
	If 'a' is '0' and the extended instruction set is enabled, this instruction operates in Indexed Literal Offset Addressing mode whenever f ≤ 95 (5Fh). See Section 29.2.3 "Byte-Oriented and Bit-Oriented Instructions in Indexed Literal Offset Mode" for details.				
Words:	1				
Cycles:	1				
Q Cycle Activity:					
Q1	Q2	Q3	Q4		
Decode	Read	Process	Write to		
	register 'f	' Data	destination		
Example 1:	SUBWF	REG, 1, 0			
Before Instruc	tion				
REG W	= 3 = 2 = 2				
Č	- 2				
	= ?				
After Instruction	•				
REG	on				
	= 1 = 2 = 1	; result is positiv	e		
REG W	on = 1 = 2 = 1 = 0	; result is positiv	e		
REG W C Z N	on = 1 = 2 = 1 = 0 = 0		e		
REG W	on = 1 = 2 = 1 = 0 = 0 SUBWF	; result is positiv REG, 0, 0	e		
REG W C Z N <u>Example 2:</u> Before Instruc REG	on = 1 = 2 = 1 = 0 = 0 SUBWF		e		
REG W C Z N Example 2: Before Instruct REG W	on = 1 = 2 = 1 = 0 = 0 SUBWF tion = 2 = 2		e		
REG W C Z N <u>Example 2:</u> Before Instruc REG	on = 1 = 2 = 1 = 0 SUBWF ttion = 2 = 2 = ?		e		
REG W C Z N Example 2: Before Instruct REG W C After Instruction REG	$ \begin{array}{rcl}         & & & \\             & = & 1 \\             & = & 2 \\             & = & 0 \\             & = & 0 \\             & = & 0 \\             & & \\             & & \\         $		e		
REG W C Z N Example 2: Before Instruct REG W C After Instructio REG W	$ \begin{array}{rcl}  & & & \\  & = & 1 \\  & = & 2 \\  & = & 1 \\  & = & 0 \\  & = & 0 \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  $	REG, 0, 0	e		
REG W C Z N Example 2: Before Instruct REG W C After Instruction REG	$ \begin{array}{rcl}  & & & \\  & = & 1 \\  & = & 2 \\  & = & 1 \\  & = & 0 \\  & = & 0 \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  & & & \\  $		e		
REG W C Z N N Example 2: Before Instruct REG W C After Instruction REG W C Z N	$ \begin{array}{rcl} \text{Dn} & = & 1 \\       = & 2 \\       = & 1 \\       = & 0 \\       = & 0 \\       & \text{SUBWF} \\ \begin{array}{r} \text{SUBWF} \\       = & 2 \\       = & 2 \\       = & 2 \\       = & 2 \\       = & 2 \\       = & 2 \\       = & 2 \\       = & 2 \\       = & 2 \\       = & 2 \\       = & 2 \\       = & 2 \\       = & 2 \\       = & 2 \\       = & 2 \\       = & 2 \\       = & 2 \\       = & 2 \\       = & 2 \\       = & 2 \\       = & 2 \\       = & 2 \\       = & 2 \\       = & 2 \\       = & 2 \\       = & 2 \\       = & 2 \\       = & 2 \\       = & 2 \\       = & 2 \\       = & 2 \\       = & 2 \\       = & 2 \\       = & 2 \\       = & 2 \\       = & 2 \\       = & 2 \\       = & 2 \\       = & 2 \\       = & 2 \\       = & 2 \\       = & 2 \\       = & 2 \\       = & 2 \\       = & 2 \\       = & 2 \\       = & 2 \\       = & 2 \\       = & 2 \\       = & 2 \\       = & 2 \\       = & 2 \\       = & 2 \\       = & 2 \\       = & 2 \\       = & 2 \\       = & 2 \\       = & 2 \\       = & 2 \\       = & 2 \\       = & 2 \\       = & 2 \\       = & 2 \\       = & 2 \\       = & 2 \\       = & 2 \\       = & 2 \\       = & 2 \\       = & 2 \\       = & 2 \\       = & 2 \\       = & 2 \\       = & 2 \\       = & 2 \\       = & 2 \\       = & 2 \\       = & 2 \\       = & 2 \\       = & 2 \\       = & 2 \\       = & 2 \\       = & 2 \\       = & 2 \\       = & 2 \\       = & 2 \\       = & 2 \\       = & 2 \\       = & 2 \\       = & 2 \\       = & 2 \\       = & 2 \\       = & 2 \\       = & 2 \\       = & 2 \\       = & 2 \\       = & 2 \\       = & 2 \\       = & 2 \\       = & 2 \\       = & 2 \\       = & 2 \\       = & 2 \\       = & 2 \\       = & 2 \\       = & 2 \\       = & 2 \\       = & 2 \\       = & 2 \\       = & 2 \\       = & 2 \\       = & 2 \\       = & 2 \\       = & 2 \\       = & 2 \\       = & 2 \\       = & 2 \\       = & 2 \\       = & 2 \\       = & 2 \\       = & 2 \\       = & 2 \\       = & 2 \\       = & 2 \\       = & 2 \\       = & 2 \\       = & 2 \\       = & 2 \\       = & 2 \\       = & 2 \\       = & 2 \\       = & 2 \\       = & 2 \\       = & 2 \\       = & 2 \\       = & 2 \\       = & 2 \\       = & 2 \\       = & 2$	REG, 0, 0	e		
REG W C Z N Before Instruct REG W C After Instructio REG W C Z N N Example 3:	on = 1 = 2 = 1 = 0 SUBWF stion = 2 = 2 = ? on = 2 = ? on = 1 = 0 SUBWF	REG, 0, 0	e		
REG W C Z N Before Instruct REG W C After Instruction REG W C Z N N Example 3: Before Instruct	$ \begin{array}{rcl}         = & 1 \\             = & 2 \\             = & 1 \\             = & 0 \\             = & 0 \\             SUBWF         \\             stion         \\             = & 2 \\             = & 2 \\             = & 2 \\             = & 2 \\             = & 2 \\             = & 2 \\             = & 2 \\             = & 2 \\             = & 2 \\             = & 2 \\             = & 2 \\             = & 2 \\             = & 2 \\             = & 2 \\             = & 2 \\             = & 2 \\             = & 2 \\             = & 2 \\             = & 2 \\             = & 2 \\             = & 2 \\             = & 2 \\             = & 2 \\             = & 2 \\             = & 2 \\             = & 2 \\             = & 2 \\             = & 2 \\             = & 2 \\             = & 2 \\             = & 2 \\             = & 2 \\             = & 2 \\             = & 2 \\             = & 2 \\             = & 2 \\             = & 2 \\             = & 2 \\             = & 1 \\             = & 0 \\             SUBWF         $ stion	REG, 0, 0	e		
REG W C Z N Before Instruct REG W C After Instruction REG W C Example 3: Before Instruct REG W	$ \begin{array}{rcl}         & & \\                         $	REG, 0, 0	e		
REG W C Z N Example 2: Before Instruct REG W C After Instruction REG W C Example 3: Before Instruct REG W C C	$\begin{array}{c} & & \\ & = & 1 \\ & = & 2 \\ & = & 1 \\ & = & 0 \\ & = & 0 \\ & \\ & \\ & \\ & \\ & \\ & \\ & \\ & \\ &$	REG, 0, 0	e		
REG W C Z N Before Instruct REG W C After Instruction REG W C Example 3: Before Instruct REG W	$\begin{array}{c} & & \\ & = & 1 \\ & = & 2 \\ & = & 1 \\ & = & 0 \\ & = & 0 \\ & \\ & \\ & \\ & \\ & \\ & \\ & \\ & \\ &$	REG, 0, 0 ; result is zero REG, 1, 0			
REG W C Z N Example 2: Before Instruct REG W C After Instruction REG W C Example 3: Before Instruct REG W C After Instruction REG W C After Instruction REG After Instruction REG M C After Instruction REG M C After Instruction REG M C After Instruction REG M C After Instruction REG M C After Instruction REG M C	$\begin{array}{c} & & \\ & = & 1 \\ & = & 2 \\ & = & 1 \\ & = & 0 \\ & = & 0 \\ & \\ & \\ & \\ & \\ & \\ & \\ & \\ & \\ &$	REG, 0, 0	nt)		

SUBWFB	Subtract V	/ from f with Bor	row		
Syntax:	SUBWFB	f {,d {,a}}			
Operands:	$0 \leq f \leq 255$				
	d ∈ [0,1] a ∈ [0,1]				
Operations		$\overline{C}$ , deat			
Operation:	(f) - (W) - (W) - (f) = (f) - (f)				
Status Affected:	N, OV, C, DC, Z				
Encoding:		0101 10da ffff ffff			
Description:	Subtract W and the Carry flag (borrow) from register 'f' (2's complement method). If 'd' is '0', the result is stored in W. If 'd' is '1', the result is stored back in register 'f'.				
	If 'a' is '0', the Access Bank is selected. If 'a' is '1', the BSR is used to select the GPR bank.				
	set is enablin Indexed mode wher Section 29 Bit-Oriente	nd the extended in ed, this instruction Literal Offset Addu ever $f \le 95$ (5Fh). .2.3 "Byte-Orient of Instructions in set Mode" for det	n operates ressing See red and Indexed		
Vords:	1				
ycles:	1				
Q Cycle Activity:					
Q1	Q2	Q3	Q4		
Decode	Read register 'f'		Write to estination		
ample 1:	SUBWFB	REG, 1, 0	countation		
Before Instruc		REG, I, U			
REG W C	= 19h = 0Dh = 1	(0001 1001) (0000 1101)			
After Instructio	_				
REG W C	= 0Ch = 0Dh = 1	(0000 1011) (0000 1101)			
Z N	= 0 = 0	; result is posit	ivo		
xample 2:	0	REG, 0, 0			
Before Instruc					
REG W C	= 1Bh = 1Ah = 0	(0001 1011) (0001 1010)			
After Instructio REG W	on = 1Bh = 00h	(0001 1011)			
C Z N	= 1 = 1 = 0	; result is zero			
xample 3:	- 0 SUBWFB	REG, 1, 0			
Before Instruc					
REG W C	= 03h = 0Eh = 1	(0000 0011) (0000 1101)			
After Instruction					
REG	= F5h	(1111 0100) ; <b>[2's comp]</b>			
W	= 0Eh	(0000 1101)			
C Z N	= 0 = 0 = 1	; result is nega	tive		

Swap f					
SWAPF f	SWAPF f {,d {,a}}				
$\begin{array}{l} 0 \leq f \leq 255 \\ d  \in  [0,1] \\ a  \in  [0,1] \end{array}$	d ∈ [0,1]				
	(f<3:0>) → dest<7:4>, (f<7:4>) → dest<3:0>				
None	None				
0011	0011 10da ffff ffff				
'f' are exch is placed in	The upper and lower nibbles of register 'f' are exchanged. If 'd' is '0', the result is placed in W. If 'd' is '1', the result is placed in register 'f'.				
,	If 'a' is '0', the Access Bank is selected. If 'a' is '1', the BSR is used to select the GPR bank.				
set is enab in Indexed mode wher Section 29 Bit-Oriente	If 'a' is '0' and the extended instruction set is enabled, this instruction operates in Indexed Literal Offset Addressing mode whenever f ≤ 95 (5Fh). See Section 29.2.3 "Byte-Oriented and Bit-Oriented Instructions in Indexed Literal Offset Mode" for details.				
1					
1					
Q2	Q3		Q4		
Read register 'f'			Vrite to stination		
ction = 53h ion	REG, 1,	0			
	SWAPF f $0 \le f \le 255$ $d \in [0,1]$ $a \in [0,1]$ $a \in [0,1]$ $(f<3:0>) \rightarrow$ $(f<7:4>) \rightarrow$ None 0011 The upper f' are exch is placed in re If 'a' is '0', t If 'a' is '1', t GPR bank. If 'a' is '0' a set is enab in Indexed mode wher Section 29 Bit-Oriente Literal Offs 1 1 Q2 Read register 'f' SWAPF F ction = 53h ion	SWAPFf {,d {,a}} $0 \le f \le 255$ $d \in [0,1]$ $a \in [0,1]$ $a \in [0,1]$ $(f<3:0>) \rightarrow dest<7:4:$ $(f<7:4>) \rightarrow dest<3:0:$ None $0011$ $10da$ The upper and lower'f are exchanged. If 'is placed in W. If 'd' iplaced in W. If 'd' iplaced in W. If 'd' iplaced in W. If 'd' is 'o', the AccesIf 'a' is '0', the AscesIf 'a' is '0' and the exset is enabled, this irin Indexed Literal Offmode whenever $f \le 0$ Section 29.2.3 "ByteBit-Oriented InstructLiteral Offset Mode1Q2Q3ReadProcesregister 'f'DataSWAPFREG, 1,ction= 53hion	SWAPF f {.d {.a}} $0 \le f \le 255$ $d \in [0,1]$ $a \in [0,1]$ $a \in [0,1]$ $(f<3:0>) \rightarrow dest<7:4>,$ $(f<7:4>) \rightarrow dest<3:0>$ None $0011$ $10da$ $ffff$ The upper and lower nibbles off' are exchanged. If 'd' is '0', theplaced in W. If 'd' is '1', theplaced in register 'f'.If 'a' is '0', the Access Bank isIf 'a' is '0' and the extended inset is enabled, this instructionin Indexed Literal Offset Addresmode whenever $f \le 95$ (5Fh).Section 29.2.3 "Byte-OrientedBit-Oriented Instructions inLiteral Offset Mode" for deta11Q2Q3ReadProcessSWAPFREG, 1, 0ction=53hion		

Table Read (Continued)

TBLRD

TBL	RD	Table Read					
Synta	ax:	TBLRD (*;	*+; *	-; +*)			
Oper	ands:	None					
Oper	ation:	if TBLRD *, (Prog Mem (TBLPTR)) $\rightarrow$ TABLAT; TBLPTR – No Change if TBLRD *+, (Prog Mem (TBLPTR)) $\rightarrow$ TABLAT; (TBLPTR) + 1 $\rightarrow$ TBLPTR if TBLRD *-, (Prog Mem (TBLPTR)) $\rightarrow$ TABLAT; (TBLPTR) – 1 $\rightarrow$ TBLPTR if TBLRD +*, (TBLPTR) + 1 $\rightarrow$ TBLPTR; (Prog Mem (TBLPTR)) $\rightarrow$ TABLAT					
Statu	s Affected:	None					
Enco	ding:	0000	0	000	000	00	10nn nn=0 * =1 *+ =2 *- =3 +*
Desc	ription:	This instruct of Program program me Pointer (TBI The TBLPT	Men emor LPTI	nory (F y, a po R) is u	P.M.). Dinter o sed.	To ad called	dress the Table
		each byte in has a 2-Mby	the	progra	am me	emory	
		TBLPTR<	0> =				nt Byte of mory Word
		TBLPTR<	0> =				t Byte of nory Word
		The TBLRD of TBLPTR				nodify	the value
		<ul> <li>no chang</li> </ul>	е				
		<ul> <li>post-increase</li> </ul>	emei	nt			
		<ul> <li>post-decr</li> </ul>					
		pre-incre	men	t			
Words: 1							
Cycle	es:	2					
QC	ycle Activity:			-			<u>.</u>
	Q1 Decode	Q2 No operation		C N opera	-	on	Q4 No eration

TBLRD	*+	;	
on			
		=	55h
(004356h)	`		00A356h 34h
. ,	,	-	3411
•		=	34h
		=	00A357h
TBLRD	+*	;	
on			
		=	AAh
(01 A 2576)		=	01A357h 12h
		=	34h
) 1			
		=	34h
		=	01A358h
	(00A356h) TBLRD on (01A357h) (01A358h)	(00A356h) TBLRD +* on (01A357h) (01A358h)	DON = = = = = = = = = = = = = = = = = = =

No

operation

No operation

(Read Program

Memory)

No

operation

No operation (Write TABLAT)

TBLWT	Table Wri	te				
Syntax:	TBLWT ( '	*; *+; *-; +*	*)			
Operands:	None					
Operation:	if TBLWT*, (TABLAT) $\rightarrow$ Holding Register; TBLPTR – No Change if TBLWT*+, (TABLAT) $\rightarrow$ Holding Register; (TBLPTR) + 1 $\rightarrow$ TBLPTR if TBLWT*-, (TABLAT) $\rightarrow$ Holding Register; (TBLPTR) – 1 $\rightarrow$ TBLPTR if TBLWT+*, (TBLPTR) + 1 $\rightarrow$ TBLPTR; (TABLAT) $\rightarrow$ Holding Register					
Status Affected:	None		0 0			
Encoding:	0000	0000	0000	11nn nn=0 * =1 *+ =2 *- =3 +*		
Description:	This instruction uses the 3 LSBs of TBLPTR to determine which of the 8 holding registers the TABLAT is written to. The holding registers are used to program the contents of Program Memory (P.M.). (Refer to Section 6.0 "Memory Organization" for additional details on programming Flash memory.) The TBLPTR (a 21-bit pointer) points to each byte in the program memory.					
	TBLPTR I The LSb c byte of the access.	of the TBL e program	PTR select memory le	ts which		
	IDLFI	R[U] – 0.L		n Memory		
	TBLPT	R[0] = 1:N	lost Signif Program I Word	icant Byte of Memory		
	The TBLW value of T	BLPTR as		odify the		
	•	crement crement				
Words:	1					
Cycles: Q Cycle Activity:	2					
	Q1	Q2	Q3	Q4		
	Decode	No	No	No		
			operation	operation		
	No operation	No operation (Read TABLAT)	No operation	No operation (Write to Holding Register)		

Register)

#### TBLWT Table Write (Continued)

Example 1:	TBLWT	*+

Example 1:	TBLWT *+;		
Before Instru	uction		
TABLA TBLPTI HOLDII		=	55h 00A356h
(00A35		=	FFh
After Instruct	tions (table write	comp	letion)
TABLA	Г	=	55h
TBLPT		=	00A357h
HOLDII (00A35	NG REGISTER 6h)	=	55h
Example 2:	TBLWT +*;		
Before Instru	uction		
TABLA	Г	=	34h
TBLPT		=	01389Ah
(01389)	NG REGISTER Ah) NG REGISTER	=	FFh
(01389)		=	FFh
After Instruct	tion (table write c	omple	etion)
TABLA	•	=	34h
TBLPT	R	=	01389Bh
(01389)	NG REGISTER Ah) NG REGISTER	=	FFh
(01389)		=	34h

TSTR	⁼sz	Test f, Skip	Test f, Skip if 0				
Synta	ax:	TSTFSZ f {	,a}				
Oper	ands:	0 ≤ f ≤ 255 a ∈ [0,1]					
Oper	ation:	skip if f = 0					
Statu	s Affected:	None					
Enco	ding:	0110	011a fff	f ffff			
Desc	ription:	during the o	If 'f' = 0, the next instruction fetched during the current instruction execution is discarded and a NOP is executed, making this a two-cycle instruction.				
			he Access Bar he BSR is used				
If 'a' is '0' and the extended instruction set is enabled, this instruction operate in Indexed Literal Offset Addressing mode whenever f ≤ 95 (5Fh). See Section 29.2.3 "Byte-Oriented and Bit-Oriented Instructions in Indexe Literal Offset Mode" for details.							
Words: 1							
Cycle		•	vcles if skip and a 2-word instru				
QC	ycle Activity:	00	00	04			
	Q1 Decode	Q2 Read	Q3 Process	Q4 No			
	Decoue	register 'f'	Data	operation			
lf sk	ip:						
	Q1	Q2	Q3	Q4			
	No	No	No	No			
	operation	operation	operation	operation			
IT SK	ip and followe	-		04			
	Q1 No	Q2 No	Q3 No	Q4 No			
	operation	operation	operation	operation			
	No	No	No	No			
	operation	operation	operation	operation			
Example: HERE TSTFSZ CNT, 1 NZERO : ZERO :							
	Before Instruc	tion					
	PC		dress (HERE	)			
	After Instructio		h				
	If CNT PC	= 00 = Ad	n, dress (ZERO	)			
	If CNT PC	≠ 00 = Ad	h, dress (NZERG	)			

XORLW	Exclusive	Exclusive OR Literal with W					
Syntax:	XORLW	k					
Operands:	$0 \le k \le 25$	5					
Operation:	(W) .XOR	$k \rightarrow W$					
Status Affected:	N, Z						
Encoding:	0000	1010	kkkk	kkkk			
Description:		The contents of W are XORed with the 8-bit literal 'k'. The result is placed in W.					
Words:	1	1					
Cycles:	1						
Q Cycle Activity:							
Q1	Q2	Q3		Q4			
Decode	Read literal 'k'	Proces Data		Write to W			
Example:	XORLW	0AFh					
Before Instruc W After Instructio W	= B5h						

XORWF	Exclusive	OR W with	f			
Syntax:	XORWF	f {,d {,a}}				
Operands:	$\begin{array}{l} 0 \leq f \leq 255 \\ d  \in  [0,1] \\ a  \in  [0,1] \end{array}$	d ∈ [0,1]				
Operation:	(W) .XOR. (	(f) $\rightarrow$ dest				
Status Affected:	N, Z					
Encoding:	0001	10da i	Efff	ffff		
Description:	Exclusive C register 'f'. I in W. If 'd' is in the regist	f 'd' is '0', th s '1', the res	e resul	t is stored		
	lf 'a' is '0', ti lf 'a' is '1', ti GPR bank.					
	If 'a' is '0' and the extended instruction set is enabled, this instruction operates in Indexed Literal Offset Addressing mode whenever f ≤ 95 (5Fh). See Section 29.2.3 "Byte-Oriented and Bit-Oriented Instructions in Indexed Literal Offset Mode" for details.					
Words:	1					
Cycles:	1					
Q Cycle Activity:						
Q1	Q2	Q3		Q4		
Decode	Read register 'f'	Process Data	-	Vrite to stination		
Example: XORWF REG, 1, 0 Before Instruction						
Ŵ	REG = AFh W = B5h					
After Instructio REG W	on = 1Ah = B5h					

### 29.2 Extended Instruction Set

In addition to the standard 75 instructions of the PIC18 instruction set, the PIC18F87K22 family of devices also provides an optional extension to the core CPU functionality. The added features include eight additional instructions that augment Indirect and Indexed Addressing operations and the implementation of Indexed Literal Offset Addressing for many of the standard PIC18 instructions.

The additional features of the extended instruction set are enabled by default on unprogrammed devices. Users must properly set or clear the XINST Configuration bit during programming to enable or disable these features.

The instructions in the extended set can all be classified as literal operations, which either manipulate the File Select Registers, or use them for Indexed Addressing. Two of the instructions, ADDFSR and SUBFSR, each have an additional special instantiation for using FSR2. These versions (ADDULNK and SUBULNK) allow for automatic return after execution.

The extended instructions are specifically implemented to optimize re-entrant program code (that is, code that is recursive or that uses a software stack) written in high-level languages, particularly C. Among other things, they allow users working in high-level languages to perform certain operations on data structures more efficiently. These include:

- Dynamic allocation and deallocation of software stack space when entering and leaving subroutines
- Function Pointer invocation
- Software Stack Pointer manipulation
- Manipulation of variables located in a software stack

A summary of the instructions in the extended instruction set is provided in Table 29-3. Detailed descriptions are provided in Section 29.2.2 "Extended Instruction Set". The opcode field descriptions in Table 29-1 (page 432) apply to both the standard and extended PIC18 instruction sets.

Note: The instruction set extension and the Indexed Literal Offset Addressing mode were designed for optimizing applications written in C; the user may likely never use these instructions directly in assembler. The syntax for these commands is provided as a reference for users who may be reviewing code that has been generated by a compiler.

### 29.2.1 EXTENDED INSTRUCTION SYNTAX

Most of the extended instructions use indexed arguments, using one of the File Select Registers and some offset to specify a source or destination register. When an argument for an instruction serves as part of Indexed Addressing, it is enclosed in square brackets ("[]"). This is done to indicate that the argument is used as an index or offset. The MPASM<sup>™</sup> Assembler will flag an error if it determines that an index or offset value is not bracketed.

When the extended instruction set is enabled, brackets are also used to indicate index arguments in byte-oriented and bit-oriented instructions. This is in addition to other changes in their syntax. For more details, see Section 29.2.3.1 "Extended Instruction Syntax with Standard PIC18 Commands".

Note: In the past, square brackets have been used to denote optional arguments in the PIC18 and earlier instruction sets. In this text and going forward, optional arguments are denoted by braces ("{ }").

Mnemonic, Operands		Description	Cuolos	16-Bit Instruction Word				Status
		Description	Cycles	MSb			LSb	Affected
ADDFSR	f, k	Add Literal to FSR	1	1110	1000	ffkk	kkkk	None
ADDULNK	k	Add Literal to FSR2 and Return	2	1110	1000	11kk	kkkk	None
CALLW		Call Subroutine using WREG	2	0000	0000	0001	0100	None
MOVSF	z <sub>s</sub> , f <sub>d</sub>	Move z <sub>s</sub> (source) to 1st word	2	1110	1011	0zzz	ZZZZ	None
		f <sub>d</sub> (destination) 2nd word		1111	ffff	ffff	ffff	
MOVSS	z <sub>s</sub> , z <sub>d</sub>	Move z <sub>s</sub> (source) to 1st word	2	1110	1011	lzzz	zzzz	None
		z <sub>d</sub> (destination) 2nd word		1111	xxxx	XZZZ	ZZZZ	
PUSHL	k	Store Literal at FSR2,	1	1110	1010	kkkk	kkkk	None
		Decrement FSR2						
SUBFSR	f, k	Subtract Literal from FSR	1	1110	1001	ffkk	kkkk	None
SUBULNK	k	Subtract Literal from FSR2 and	2	1110	1001	11kk	kkkk	None
		return						

### TABLE 29-3: EXTENSIONS TO THE PIC18 INSTRUCTION SET

### 29.2.2 EXTENDED INSTRUCTION SET

ADD	FSR	Add Liter	Add Literal to FSR				
Synta	ax:	ADDFSR	ADDFSR f, k				
Oper	ands:	$0 \le k \le 63$					
		f ∈ [ 0, 1, 2 ]					
Operation: $FSR(f) + k \rightarrow FSR(f)$							
Statu	s Affected:	None	None				
Encoding: 1110 1000 ffkk kkkk					kkkk		
Desc	ription:	The 6-bit	The 6-bit literal 'k' is added to the				
		contents of	contents of the FSR specified by 'f'.				
Word	ls:	1	1				
Cycle	es:	1					
QC	ycle Activity:						
	Q1	Q2	Q3		Q4		
	Decode	Read	Proces	ss	Write to		
		literal 'k'	Data		FSR		

Example: ADDFSR 2, 23h

Before Instruction					
FSR2	=	03FFh			
After Instruct					
FSR2	=	0422h			

ADDULNK Add Literal to FSR2 and Return					d Return		
Synta	ax:	ADDULN	K k				
Oper	ands:	$0 \le k \le 63$					
Oper	ation:	FSR2 + k	$\rightarrow$ FSR2,				
		$(TOS) \rightarrow F$	PC				
Statu	s Affected:	None					
Enco	ding:	1110	1000	11k]	c kkkk		
Desc	ription:	contents o	The 6-bit literal 'k' is added to the contents of FSR2. A RETURN is then executed by loading the PC with the TOS.				
	The instruction takes two cycles to execute; a NOP is performed during the second cycle.						
		case of the	e ADDFSF 3 (binary	e instr	s a special uction, it operates		
Word	ls:	1	1				
Cycle	es:	2	2				
QC	ycle Activity:						
	Q1	Q2	Q3		Q4		
	Decode	Read literal 'k'	Proces Data	ss	Write to FSR		
	No	No	No		No		
	Operation	Operation Operation Operation					
<u>Exan</u>	nple:	ADDULNK 2	23h				

ampie:	AI	DDULNK	2
Before Instruc	tion		
FSR2	=	03FFh	
PC	=	0100h	
After Instruction	n		
FSR2	=	0422h	
PC	=	(TOS)	

**Note:** All PIC18 instructions may take an optional label argument preceding the instruction mnemonic for use in symbolic addressing. If a label is used, the instruction format then becomes: {label} instruction argument(s).

CALI	LW	Subroutine	Subroutine Call Using WREG				
Synta	ax:	CALLW					
Opera	ands:	None	None				
Opera	ation:	$(W) \rightarrow PCL$ $(PCLATH)$	$(PC + 2) \rightarrow TOS,$ $(W) \rightarrow PCL,$ $(PCLATH) \rightarrow PCH,$ $(PCLATU) \rightarrow PCU$				
Statu	s Affected:	None					
Enco	ding:	0000	0000 000	01 0100			
Desc	ription	pushed ont contents of existing val contents of latched into respectively executed a	First, the return address (PC + 2) is pushed onto the return stack. Next, the contents of W are written to PCL; the existing value is discarded. Then, the contents of PCLATH and PCLATU are latched into PCH and PCU, respectively. The second cycle is executed as a NOP instruction while the new next instruction is fetched.				
			⊥, there is no o STATUS or BS				
Word	ls:	1					
Cycle	es:	2					
QC	ycle Activity:						
	Q1	Q2	Q3	Q4			
	Decode	Read WREG	Push PC to stack	No operation			
	No	No	No	No			
	operation	operation	operation	operation			
	nple: PC PCLATH PCLATU W After Instructio PC TOS PCLATH PCLATU W	= address = 10h = 00h = 06h m = 001006 = address = 10h	h	)			

		Move Indexed to f				
Syntax:	MOVSF [z <sub>s</sub> ], f <sub>d</sub>					
Operands:	$\begin{array}{l} 0 \leq z_s \leq 127 \\ 0 \leq f_d \leq 4095 \end{array}$					
Operation:	$((FSR2) + z_s) \rightarrow f_d$					
Status Affected:	None					
Encoding: 1st word (source) 2nd word (destin.)	1110 1111	1011 ffff	0zz fff	ъ		
Description:	The contents of the source register are moved to destination register 'f <sub>d</sub> '. The actual address of the source register is determined by adding the 7-bit literal offset ' $z_s$ ', in the first word, to the value of FSR2. The address of the destination register is specified by the 12-bit literal 'f <sub>d</sub> ' in the second word. Both addresses can be anywhere in the 4096-byte data					
	space (000 The MOVSF PCL, TOSL destination	instructic J, TOSH d	n car	not use the SL as the		
	If the result an Indirect value returr	Addressir	ng reg			
Words:	2					
Cycles:	2					
Q Cycle Activity:						
Q1	Q2	Q3		Q4		
Decode	Determine source addr	Determi source a	-	Read source reg		
Decode	No operation No dummy read	No No operation operatio No dummy		Write register 'f' (dest)		
Example: MOVSF [05h], REG2						
Before Instruction FSR2 = 80h						
Contents of 85h REG2						
After Instructio FSR2	n					
Contents of 85h						

MOVSS	Move Inde	Move Indexed to Indexed					
Syntax:	MOVSS [	MOVSS [z <sub>s</sub> ], [z <sub>d</sub> ]					
Operands:	$0 \le z_s \le 12$ $0 \le z_d \le 12$						
Operation:	((FSR2) +	$z_s) \rightarrow ((F$	SR2) + z <sub>d</sub>	)			
Status Affected:	None						
Encoding: 1st word (source) 2nd word (dest.)	1110 1111	1011 xxxx	lzzz xzzz	zzzz <sub>s</sub> zzzz <sub>d</sub>			
Description	The contents of the source register are moved to the destination register. The addresses of the source and destination registers are determined by adding the 7-bit literal offsets, 'z <sub>s</sub> ' or 'z <sub>d</sub> ', respectively, to the value of FSR2. Both registers can be located anywhere in the 4096-byte data memory space (000h to FFFh). The MOVSS instruction cannot use the PCL, TOSU, TOSH or TOSL as the destination register.						
	If the resultant source address points to an Indirect Addressing register, the value returned will be 00h. If the resultant destination address points to an Indirect Addressing register, the instruction will execute as a NOP.						
Words:	2						
Cycles:	2						
Q Cycle Activity:							

ycles:	2	
Q Cycle Activity:		
Q1	Q2	Q3

Q1	Q2	Q3	Q4
Decode	Determine	Determine	Read
	source addr	source addr	source reg
Decode	Determine dest addr	Determine dest addr	Write to dest reg

Example:	MOVSS	[05h],	[06h]

Before Instruction		
FSR2	=	80h
Contents of 85h Contents	=	33h
of 86h	=	11h
After Instruction		
FSR2	=	80h
Contents of 85h Contents	=	33h
of 86h	=	33h

Syntax:	PUSHL k							
o .			PUSHL k					
Operands:	$0 \le k \le 255$							
Operation:	$k \rightarrow (FSR2)$ FSR2 – 1 –							
Status Affected:	None							
Encoding:	1110	1010	kkkł	kkkk				
	memory address specified by FSR2. FSR2 is decremented by 1 after the operation. This instruction allows users to push							
	values onto	a softwa	re stack	κ.				
Words:	1							
Cycles:	1							
Q Cycle Activity:								
Q1	Q2	G	3	Q4				
Decode	Read 'k'	Proc da		Write to destination				
Example:	PUSHL 0	8h						

Before Instruction FSR2H:FSR2L Memory (01ECh)	= =	01ECh 00h
After Instruction FSR2H:FSR2L Memory (01ECh)	= =	01EBh 08h

SUBULNK k

 $FSR2 - k \rightarrow FSR2$ ,

 $0 \leq k \leq 63$ 

Subtract Literal from FSR2 and Return

SUBFSR	Subtract	Subtract Literal from FSR				
Syntax:	SUBFSR	SUBFSR f, k				
Operands:	$0 \le k \le 63$	1				
	$f \in [0, 1,$	2]				
Operation:	FSRf – k	$\rightarrow$ FSRf				
Status Affected:	None					
Encoding:	1110	1110 1001 ffkk kkkk				
Description:		The 6-bit literal 'k' is subtracted from the contents of the FSR specified by 'f'.				
Words:	1					
Cycles:	1					
Q Cycle Activity:						
Q1	Q2	Q3		Q4		
Decode	Read register 'f'	Proce Data		Write to destination		
Example: Before Instruc	SUBFSR 1	2, 23h				

oper		$(TOS) \rightarrow PC$				
Statu	is Affected:	None				
Enco	oding:	1110	1001	11kk	kkkk	
Desc		The 6-bit literal 'k' is subtracted from the contents of the FSR2. A RETURN is then executed by loading the PC with the TOS.				
		The instruction execute; a second cye	NOP is per			
ı	This may be thought of as a special case of the SUBFSR instruction, where f = 3 (binary '11'); it operates only on FSR2.					
Word	ls:	1				
Cycle	es:	2				
QC	ycle Activity:					
	Q1	Q2	C	23	Q4	
	Decode	Read register		cess ata	Write to destination	
	No	No	N	lo	No	
	Operation	Operatio	n Oper	ation	Operation	

Before Instruction 03FFh FSR2 = After Instruction 03DCh FSR2 =

Example:

SUBULNK

Operands:

Operation:

Syntax:

ample:	SU	JBULNK	23h
Before Instruction	on		
FSR2 :	=	03FFh	
PC :	-	0100h	
After Instruction	ı		
FSR2 :	=	03DCh	
PC :	=	(TOS)	

### 29.2.3 BYTE-ORIENTED AND BIT-ORIENTED INSTRUCTIONS IN INDEXED LITERAL OFFSET MODE

Note:	Enabling the PIC18 instruction set exten-
	sion may cause legacy applications to
	behave erratically or fail entirely.

In addition to eight new commands in the extended set, enabling the extended instruction set also enables Indexed Literal Offset Addressing (Section 6.6.1 "Indexed Addressing with Literal Offset"). This has a significant impact on the way that many commands of the standard PIC18 instruction set are interpreted.

When the extended set is disabled, addresses embedded in opcodes are treated as literal memory locations: either as a location in the Access Bank (a = 0) or in a GPR bank designated by the BSR (a = 1). When the extended instruction set is enabled and a = 0, however, a file register argument of 5Fh or less is interpreted as an offset from the pointer value in FSR2 and not as a literal address. For practical purposes, this means that all instructions that use the Access RAM bit as an argument – that is, all byte-oriented and bit-oriented instructions, or almost half of the core PIC18 instructions – may behave differently when the extended instruction set is enabled.

When the content of FSR2 is 00h, the boundaries of the Access RAM are essentially remapped to their original values. This may be useful in creating backward compatible code. If this technique is used, it may be necessary to save the value of FSR2 and restore it when moving back and forth between C and assembly routines in order to preserve the Stack Pointer. Users must also keep in mind the syntax requirements of the extended instruction set (see Section 29.2.3.1 "Extended Instruction Syntax with Standard PIC18 Commands").

Although the Indexed Literal Offset mode can be very useful for dynamic stack and pointer manipulation, it can also be very annoying if a simple arithmetic operation is carried out on the wrong register. Users who are accustomed to the PIC18 programming must keep in mind, that when the extended instruction set is enabled, register addresses of 5Fh or less are used for Indexed Literal Offset Addressing.

Representative examples of typical byte-oriented and bit-oriented instructions in the Indexed Literal Offset mode are provided on the following page to show how execution is affected. The operand conditions shown in the examples are applicable to all instructions of these types.

### 29.2.3.1 Extended Instruction Syntax with Standard PIC18 Commands

When the extended instruction set is enabled, the file register argument 'f' in the standard byte-oriented and bit-oriented commands is replaced with the literal offset value 'k'. As already noted, this occurs only when 'f' is less than or equal to 5Fh. When an offset value is used, it must be indicated by square brackets ("[]"). As with the extended instructions, the use of brackets indicates to the compiler that the value is to be interpreted as an index or an offset. Omitting the brackets, or using a value greater than 5Fh within the brackets, will generate an error in the MPASM<sup>™</sup> Assembler.

If the index argument is properly bracketed for Indexed Literal Offset Addressing, the Access RAM argument is never specified; it will automatically be assumed to be '0'. This is in contrast to standard operation (extended instruction set disabled), when 'a' is set on the basis of the target address. Declaring the Access RAM bit in this mode will also generate an error in the MPASM Assembler.

The destination argument, 'd', functions as before.

In the latest versions of the MPASM Assembler, language support for the extended instruction set must be explicitly invoked. This is done with either the command line option,  $/_{Y}$ , or the PE directive in the source listing.

### 29.2.4 CONSIDERATIONS WHEN ENABLING THE EXTENDED INSTRUCTION SET

It is important to note that the extensions to the instruction set may not be beneficial to all users. In particular, users who are not writing code that uses a software stack may not benefit from using the extensions to the instruction set.

Additionally, the Indexed Literal Offset Addressing mode may create issues with legacy applications written to the PIC18 assembler. This is because instructions in the legacy code may attempt to address registers in the Access Bank below 5Fh. Since these addresses are interpreted as literal offsets to FSR2 when the instruction set extension is enabled, the application may read or write to the wrong data addresses.

When porting an application to the PIC18F87K22 family, it is very important to consider the type of code. A large, re-entrant application that is written in C and would benefit from efficient compilation will do well when using the instruction set extensions. Legacy applications that heavily use the Access Bank will most likely not benefit from using the extended instruction set.

ADDWF	ADD W to (Indexed	ie)		
Syntax:	ADDWF	[k] {,d}		
Operands:	$\begin{array}{l} 0 \leq k \leq 95 \\ d  \in  [0,1] \end{array}$			
Operation:	(W) + ((FS	SR2) + k) -	$\rightarrow$ dest	
Status Affected:	N, OV, C,	DC, Z		
Encoding:	0010	01d0	kkkk	kkkk
Description:	The conte contents o FSR2, offs	f the regis	ster indica	ated by
	If 'd' is '0', is '1', the r register 'f'.	esult is st		in W. lf 'd' k in
Words:	1			
Cycles:	1			
Q Cycle Activity:				
Q1	Q2	Q3	ľ	Q4
Decode	Read 'k'	Proce Data		Write to estination
Example:	ADDWF	[OFST]	,0	
Before Instructio W OFST FSR2 Contents of 0A2Ch After Instruction W Contents of 0A2Ch	= = =	2Ch 0A00r 20h 37h	1	

BSF	Bit Set Ind (Indexed L	exed iteral Offse	t mode	e)				
Syntax:	BSF [k], b							
Operands:	$\begin{array}{l} 0 \leq f \leq 95 \\ 0 \leq b \leq 7 \end{array}$							
Operation:	$1 \rightarrow ((FSR))$	2) + k) <b></b>						
Status Affected:	None							
Encoding:	1000	bbb0 k	kkk	kkkk				
Description:		e register ind e value 'k', i		by FSR2,				
Words:	1							
Cycles:	1							
Q Cycle Activity:								
Q1	Q2	Q3		Q4				
Decode	Read register 'f'	Process Data		/rite to stination				
Example:		FLAG_OFS	r], 7					
Before Instructi FLAG_OF FSR2		0Ah 0A00h						
Contents of 0A0Ah After Instructior	=	55h						
Contents of 0A0Ah	=	D5h						
SETF	Set Indexe (Indexed L	d iteral Offse	t mode	e)				
SETF Syntax:			t mode	e)				
	(Indexed L		t mode	e)				
Syntax:	(Indexed L SETF [k]	iteral Offse	t mode	e)				
Syntax: Operands:	(Indexed L SETF [k] $0 \le k \le 95$	iteral Offse	t mode	ə)				
Syntax: Operands: Operation:	(Indexed L SETF [k] $0 \le k \le 95$ FFh $\rightarrow$ ((FS	iteral Offse SR2) + k)	t mod	<b>e)</b> kkkk				
Syntax: Operands: Operation: Status Affected:	(Indexed L SETF [k] $0 \le k \le 95$ FFh $\rightarrow$ ((FS None 0110 The conten	iteral Offse SR2) + k)	ckkk	kkkk dicated by				
Syntax: Operands: Operation: Status Affected: Encoding:	(Indexed L SETF [k] $0 \le k \le 95$ FFh $\rightarrow$ ((FS None 0110 The conten	Iteral Offse       SR2) + k)       1000       Its of the region	ckkk	kkkk dicated by				
Syntax: Operands: Operation: Status Affected: Encoding: Description: Words:	(Indexed L SETF [k] $0 \le k \le 95$ FFh $\rightarrow$ ((FS None 0110 The conten FSR2, offset	Iteral Offse       SR2) + k)       1000       Its of the region	ckkk	kkkk dicated by				
Syntax: Operands: Operation: Status Affected: Encoding: Description: Words: Cycles: Q Cycle Activity:	(Indexed L SETF [k] $0 \le k \le 95$ FFh $\rightarrow$ ((FS None 0110 The conten FSR2, offse 1 1	iteral Offse         SR2) + k)         1000         its of the regiset by 'k', are	ckkk	kkkk dicated by FFh.				
Syntax: Operands: Operation: Status Affected: Encoding: Description: Words: Cycles: Q Cycle Activity: Q1	(Indexed L SETF [k] $0 \le k \le 95$ FFh $\rightarrow$ ((FS None 0110 The conten FSR2, offset 1 1 Q2	Iteral Offse       SR2) + k)       1000       H       ts of the registry 'k', are       Q3	ster ind set to	kkkk dicated by FFh. Q4				
Syntax: Operands: Operation: Status Affected: Encoding: Description: Words: Cycles: Q Cycle Activity:	(Indexed L SETF [k] $0 \le k \le 95$ FFh $\rightarrow$ ((FS None 0110 The conten FSR2, offse 1 1	iteral Offse         SR2) + k)         1000         its of the regiset by 'k', are	ster ind set to	kkkk dicated by FFh.				
Syntax: Operands: Operation: Status Affected: Encoding: Description: Words: Cycles: Q Cycle Activity: Q1 Decode	(Indexed L SETF [k] $0 \le k \le 95$ FFh $\rightarrow$ ((FS None 0110 The conten FSR2, offse 1 1 Q2 Read 'k'	GR2) + k) 1000 ts of the regination of the re	ster ind set to	kkkk dicated by FFh. Q4 Write				
Syntax: Operands: Operation: Status Affected: Encoding: Description: Words: Cycles: Q Cycle Activity: Q1 Decode Example:	(Indexed L SETF [k] $0 \le k \le 95$ FFh $\rightarrow$ ((FS None 0110 The conten FSR2, offse 1 1 2 Read 'k' SETF	iteral Offse         SR2) + k)         1000       H         ts of the registry 'k', are         Q3         Process	ster ind set to	kkkk dicated by FFh. Q4 Write				
Syntax: Operands: Operation: Status Affected: Encoding: Description: Words: Cycles: Q Cycle Activity: Q1 Decode Example: Before Instructi OFST	(Indexed L SETF [k] $0 \le k \le 95$ FFh $\rightarrow$ ((FS None 0110 The conten FSR2, offse 1 1 Q2 Read 'k' SETF [ on = 20	Coffse Coffse Coffse Coffse Coffse Coffse Coffse Coffse Coffse Coffse Coffse Coffse Coffse Coffse Coffse Coffse Coffse Coffse Coffse Coffse Coffse Coffse Coffse Coffse Coffse Coffse Coffse Coffse Coffse Coffse Coffse Coffse Coffse Coffse Coffse Coffse Coffse Coffse Coffse Coffse Coffse Coffse Coffse Coffse Coffse Coffse Coffse Coffse Coffse Coffse Coffse Coffse Coffse Coffse Coffse Coffse Coffse Coffse Coffse Coffse Coffse Coffse Coffse Coffse Coffse Coffse Coffse Coffse Coffse Coffse Coffse Coffse Coffse Coffse Coffse Coffse Coffse Coffse Coffse Coffse Coffse Coffse Coffse Coffse Coffse Coffse Coffse Coffse Coffse Coffse Coffse Coffse Coffse Coffse Coffse Coffse Coffse Coffse Coffse Coffse Coffse Coffse Coffse Coffse Coffse Coffse Coffse Coffse Coffse Coffse Coffse Coffse Coffse Coffse Coffse Coffse Coffse Coffse Coffse Coffse Coffse Coffse Coffse Coffse Coffse Coffse Coffse Coffse Coffse Coffse Coffse Coffse Coffse Coffse Coffse Coffse Coffse Coffse Coffse Coffse Coffse Coffse Coffse Coffse Coffse Coffse Coffse Coffse Coffse Coffse Coffse Coffse Coffse Coffse Coffse Coffse Coffse Coffse Coffse Coffse Coffse Coffse Coffse Coffse Coffse Coffse Coffse Coffse Coffse Coffse Coffse Coffse Coffse Coffse Coffse Coffse Coffse Coffse Coffse Coffse Coffse Coffse Coffse Coffse Coffse Coffse Coffse Coffse Coffse Coffse Coffse Coffse Coffse Coffse Coffse Coffse Coffse Coffse Coffse Coffse Coffse Coffse Coffse Coffse Coffse Coffse Coffse Coffse Coffse Coffse Coffse Coffse Coffse Coffse Coffse Coffse Coffse Coffse Coffse Coffse Coffse Coffse Coffse Coffse Coffse Coffse Coffse Coffse Coffse Coffse Coffse Coffse Coffse Coffse Coffse Coffse Coffse Coffse Coffse Coffse Coffse Coffse Coffse Coffse Coffse Coffse Coffse Coffse Coffse Coffse Coffse Coffse Coffse Coffse Coffse Coffse	ster ind set to	kkkk dicated by FFh. Q4 Write				
Syntax: Operands: Operation: Status Affected: Encoding: Description: Words: Cycles: Q Cycle Activity: Q1 Decode Example: Before Instructi	(Indexed L SETF [k] $0 \le k \le 95$ FFh $\rightarrow$ ((FS None 0110 The conten FSR2, offse 1 1 Q2 Read 'k' SETF [ on = 20	iteral Offse         SR2) + k)         1000         1000         its of the registry         et by 'k', are         Q3         Process         Data	ster ind set to	kkkk dicated by FFh. Q4 Write				
Syntax: Operands: Operation: Status Affected: Encoding: Description: Words: Cycles: Q Cycle Activity: Q1 Decode Example: Before Instructi OFST FSR2 Contents of 0A2Ch	(Indexed L SETF [k] $0 \le k \le 95$ FFh $\rightarrow$ ((FS None 0110 The conten FSR2, offset 1 1 Q2 Read 'k' SETF [ on = 20 = 00	iteral Offse         SR2) + k)         1000       J         ts of the registion of the registion of the register by 'k', are         Q3         Q3         Process         Data         CoFST]         Ch         OOh	ster ind set to	kkkk dicated by FFh. Q4 Write				
Syntax: Operands: Operation: Status Affected: Encoding: Description: Words: Cycles: Q Cycle Activity: Q1 Decode Example: Before Instructi OFST FSR2 Contents	(Indexed L SETF [k] $0 \le k \le 95$ FFh $\rightarrow$ ((FS None 0110 The conten FSR2, offset 1 1 Q2 Read 'k' SETF [ on = 20 = 00	iteral Offse         SR2) + k)         1000       Iteration         ts of the registry         et by 'k', are         Q3         Process         Data         OFST]         Ch         000h         h	ster ind set to	kkkk dicated by FFh. Q4 Write				

## 29.2.5 SPECIAL CONSIDERATIONS WITH MICROCHIP MPLAB<sup>®</sup> IDE TOOLS

The latest versions of Microchip's software tools have been designed to fully support the extended instruction set for the PIC18F87K22 family. This includes the MPLAB C18 C Compiler, MPASM assembly language and MPLAB Integrated Development Environment (IDE).

When selecting a target device for software development, MPLAB IDE will automatically set default Configuration bits for that device. The default setting for the XINST Configuration bit is '0', disabling the extended instruction set and Indexed Literal Offset Addressing. For proper execution of applications developed to take advantage of the extended instruction set, XINST must be set during programming.

To develop software for the extended instruction set, the user must enable support for the instructions and the Indexed Addressing mode in their language tool(s). Depending on the environment being used, this may be done in several ways:

- A menu option or dialog box within the environment that allows the user to configure the language tool and its settings for the project
- A command line option
- · A directive in the source code

These options vary between different compilers, assemblers and development environments. Users are encouraged to review the documentation accompanying their development systems for the appropriate information.

## 30.0 DEVELOPMENT SUPPORT

The PIC<sup>®</sup> microcontrollers and dsPIC<sup>®</sup> digital signal controllers are supported with a full range of software and hardware development tools:

- Integrated Development Environment
- MPLAB<sup>®</sup> IDE Software
- Compilers/Assemblers/Linkers
  - MPLAB C Compiler for Various Device Families
  - HI-TECH C for Various Device Families
  - MPASM<sup>™</sup> Assembler
  - MPLINK<sup>™</sup> Object Linker/ MPLIB<sup>™</sup> Object Librarian
  - MPLAB Assembler/Linker/Librarian for Various Device Families
- Simulators
  - MPLAB SIM Software Simulator
- Emulators
  - MPLAB REAL ICE™ In-Circuit Emulator
- In-Circuit Debuggers
  - MPLAB ICD 3
  - PICkit™ 3 Debug Express
- Device Programmers
  - PICkit<sup>™</sup> 2 Programmer
  - MPLAB PM3 Device Programmer
- Low-Cost Demonstration/Development Boards, Evaluation Kits, and Starter Kits

### 30.1 MPLAB Integrated Development Environment Software

The MPLAB IDE software brings an ease of software development previously unseen in the 8/16/32-bit microcontroller market. The MPLAB IDE is a Windows<sup>®</sup> operating system-based application that contains:

- A single graphical interface to all debugging tools
  - Simulator
  - Programmer (sold separately)
  - In-Circuit Emulator (sold separately)
  - In-Circuit Debugger (sold separately)
- A full-featured editor with color-coded context
- A multiple project manager
- Customizable data windows with direct edit of contents
- · High-level source code debugging
- · Mouse over variable inspection
- Drag and drop variables from source to watch windows
- · Extensive on-line help
- Integration of select third party tools, such as IAR C Compilers

The MPLAB IDE allows you to:

- Edit your source files (either C or assembly)
- One-touch compile or assemble, and download to emulator and simulator tools (automatically updates all project information)
- Debug using:
  - Source files (C or assembly)
  - Mixed C and assembly
  - Machine code

MPLAB IDE supports multiple debugging tools in a single development paradigm, from the cost-effective simulators, through low-cost in-circuit debuggers, to full-featured emulators. This eliminates the learning curve when upgrading to tools with increased flexibility and power.

### 30.2 MPLAB C Compilers for Various Device Families

The MPLAB C Compiler code development systems are complete ANSI C compilers for Microchip's PIC18, PIC24 and PIC32 families of microcontrollers and the dsPIC30 and dsPIC33 families of digital signal controllers. These compilers provide powerful integration capabilities, superior code optimization and ease of use.

For easy source level debugging, the compilers provide symbol information that is optimized to the MPLAB IDE debugger.

### 30.3 HI-TECH C for Various Device Families

The HI-TECH C Compiler code development systems are complete ANSI C compilers for Microchip's PIC family of microcontrollers and the dsPIC family of digital signal controllers. These compilers provide powerful integration capabilities, omniscient code generation and ease of use.

For easy source level debugging, the compilers provide symbol information that is optimized to the MPLAB IDE debugger.

The compilers include a macro assembler, linker, preprocessor, and one-step driver, and can run on multiple platforms.

### 30.4 MPASM Assembler

The MPASM Assembler is a full-featured, universal macro assembler for PIC10/12/16/18 MCUs.

The MPASM Assembler generates relocatable object files for the MPLINK Object Linker, Intel<sup>®</sup> standard HEX files, MAP files to detail memory usage and symbol reference, absolute LST files that contain source lines and generated machine code and COFF files for debugging.

The MPASM Assembler features include:

- · Integration into MPLAB IDE projects
- User-defined macros to streamline assembly code
- Conditional assembly for multi-purpose source files
- Directives that allow complete control over the assembly process

### 30.5 MPLINK Object Linker/ MPLIB Object Librarian

The MPLINK Object Linker combines relocatable objects created by the MPASM Assembler and the MPLAB C18 C Compiler. It can link relocatable objects from precompiled libraries, using directives from a linker script.

The MPLIB Object Librarian manages the creation and modification of library files of precompiled code. When a routine from a library is called from a source file, only the modules that contain that routine will be linked in with the application. This allows large libraries to be used efficiently in many different applications.

The object linker/library features include:

- Efficient linking of single libraries instead of many smaller files
- Enhanced code maintainability by grouping related modules together
- Flexible creation of libraries with easy module listing, replacement, deletion and extraction

### 30.6 MPLAB Assembler, Linker and Librarian for Various Device Families

MPLAB Assembler produces relocatable machine code from symbolic assembly language for PIC24, PIC32 and dsPIC devices. MPLAB C Compiler uses the assembler to produce its object file. The assembler generates relocatable object files that can then be archived or linked with other relocatable object files and archives to create an executable file. Notable features of the assembler include:

- · Support for the entire device instruction set
- · Support for fixed-point and floating-point data
- Command line interface
- · Rich directive set
- Flexible macro language
- · MPLAB IDE compatibility

### 30.7 MPLAB SIM Software Simulator

The MPLAB SIM Software Simulator allows code development in a PC-hosted environment by simulating the PIC MCUs and dsPIC<sup>®</sup> DSCs on an instruction level. On any given instruction, the data areas can be examined or modified and stimuli can be applied from a comprehensive stimulus controller. Registers can be logged to files for further run-time analysis. The trace buffer and logic analyzer display extend the power of the simulator to record and track program execution, actions on I/O, most peripherals and internal registers.

The MPLAB SIM Software Simulator fully supports symbolic debugging using the MPLAB C Compilers, and the MPASM and MPLAB Assemblers. The software simulator offers the flexibility to develop and debug code outside of the hardware laboratory environment, making it an excellent, economical software development tool.

### 30.8 MPLAB REAL ICE In-Circuit Emulator System

MPLAB REAL ICE In-Circuit Emulator System is Microchip's next generation high-speed emulator for Microchip Flash DSC and MCU devices. It debugs and programs PIC<sup>®</sup> Flash MCUs and dsPIC<sup>®</sup> Flash DSCs with the easy-to-use, powerful graphical user interface of the MPLAB Integrated Development Environment (IDE), included with each kit.

The emulator is connected to the design engineer's PC using a high-speed USB 2.0 interface and is connected to the target with either a connector compatible with incircuit debugger systems (RJ11) or with the new high-speed, noise tolerant, Low-Voltage Differential Signal (LVDS) interconnection (CAT5).

The emulator is field upgradable through future firmware downloads in MPLAB IDE. In upcoming releases of MPLAB IDE, new devices will be supported, and new features will be added. MPLAB REAL ICE offers significant advantages over competitive emulators including low-cost, full-speed emulation, run-time variable watches, trace analysis, complex breakpoints, a ruggedized probe interface and long (up to three meters) interconnection cables.

### 30.9 MPLAB ICD 3 In-Circuit Debugger System

MPLAB ICD 3 In-Circuit Debugger System is Microchip's most cost effective high-speed hardware debugger/programmer for Microchip Flash Digital Signal Controller (DSC) and microcontroller (MCU) devices. It debugs and programs PIC<sup>®</sup> Flash microcontrollers and dsPIC<sup>®</sup> DSCs with the powerful, yet easyto-use graphical user interface of MPLAB Integrated Development Environment (IDE).

The MPLAB ICD 3 In-Circuit Debugger probe is connected to the design engineer's PC using a high-speed USB 2.0 interface and is connected to the target with a connector compatible with the MPLAB ICD 2 or MPLAB REAL ICE systems (RJ-11). MPLAB ICD 3 supports all MPLAB ICD 2 headers.

### 30.10 PICkit 3 In-Circuit Debugger/ Programmer and PICkit 3 Debug Express

The MPLAB PICkit 3 allows debugging and programming of PIC<sup>®</sup> and dsPIC<sup>®</sup> Flash microcontrollers at a most affordable price point using the powerful graphical user interface of the MPLAB Integrated Development Environment (IDE). The MPLAB PICkit 3 is connected to the design engineer's PC using a full speed USB interface and can be connected to the target via an Microchip debug (RJ-11) connector (compatible with MPLAB ICD 3 and MPLAB REAL ICE). The connector uses two device I/O pins and the reset line to implement in-circuit debugging and In-Circuit Serial Programming<sup>™</sup>.

The PICkit 3 Debug Express include the PICkit 3, demo board and microcontroller, hookup cables and CDROM with user's guide, lessons, tutorial, compiler and MPLAB IDE software.

### 30.11 PICkit 2 Development Programmer/Debugger and PICkit 2 Debug Express

The PICkit<sup>™</sup> 2 Development Programmer/Debugger is a low-cost development tool with an easy to use interface for programming and debugging Microchip's Flash families of microcontrollers. The full featured Windows<sup>®</sup> programming interface supports baseline (PIC10F, PIC12F5xx, PIC16F5xx), midrange (PIC12F6xx, PIC16F), PIC18F, PIC24, dsPIC30, dsPIC33, and PIC32 families of 8-bit, 16-bit, and 32-bit microcontrollers, and many Microchip Serial EEPROM products. With Microchip's powerful MPLAB Integrated Development Environment (IDE) the PICkit<sup>™</sup> 2 enables in-circuit debugging on most PIC<sup>®</sup> microcontrollers. In-Circuit-Debugging runs, halts and single steps the program while the PIC microcontroller is embedded in the application. When halted at a breakpoint, the file registers can be examined and modified.

The PICkit 2 Debug Express include the PICkit 2, demo board and microcontroller, hookup cables and CDROM with user's guide, lessons, tutorial, compiler and MPLAB IDE software.

### 30.12 MPLAB PM3 Device Programmer

The MPLAB PM3 Device Programmer is a universal, CE compliant device programmer with programmable voltage verification at VDDMIN and VDDMAX for maximum reliability. It features a large LCD display (128 x 64) for menus and error messages and a modular, detachable socket assembly to support various package types. The ICSP™ cable assembly is included as a standard item. In Stand-Alone mode, the MPLAB PM3 Device Programmer can read, verify and program PIC devices without a PC connection. It can also set code protection in this mode. The MPLAB PM3 connects to the host PC via an RS-232 or USB cable. The MPLAB PM3 has high-speed communications and optimized algorithms for quick programming of large memory devices and incorporates an MMC card for file storage and data applications.

### 30.13 Demonstration/Development Boards, Evaluation Kits, and Starter Kits

A wide variety of demonstration, development and evaluation boards for various PIC MCUs and dsPIC DSCs allows quick application development on fully functional systems. Most boards include prototyping areas for adding custom circuitry and provide application firmware and source code for examination and modification.

The boards support a variety of features, including LEDs, temperature sensors, switches, speakers, RS-232 interfaces, LCD displays, potentiometers and additional EEPROM memory.

The demonstration and development boards can be used in teaching environments, for prototyping custom circuits and for learning about various microcontroller applications.

In addition to the PICDEM<sup>™</sup> and dsPICDEM<sup>™</sup> demonstration/development board series of circuits, Microchip has a line of evaluation kits and demonstration software for analog filter design, KEELOQ<sup>®</sup> security ICs, CAN, IrDA<sup>®</sup>, PowerSmart battery management, SEEVAL<sup>®</sup> evaluation system, Sigma-Delta ADC, flow rate sensing, plus many more.

Also available are starter kits that contain everything needed to experience the specified device. This usually includes a single application and debug capability, all on one board.

Check the Microchip web page (www.microchip.com) for the complete list of demonstration, development and evaluation kits.

## 31.0 ELECTRICAL CHARACTERISTICS

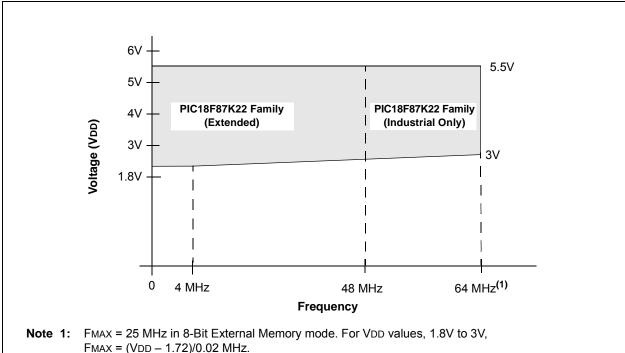
### Absolute Maximum Ratings<sup>(†)</sup>

Ambient temperature under bias	40°C to +125°C
Storage temperature	65°C to +150°C
Voltage on any digital only I/O pin with respect to Vss (except VDD)	0.3V to 7.5V
Voltage on MCLR with respect to Vss	0.3V to 9.0V
Voltage on any combined digital and analog pin with respect to Vss (except VDD and MCLR)	-0.3V to (VDD + 0.3V)
Voltage on VDD with respect to Vss (with regulator enabled)	0.3V to 5.5V
Voltage on VDD with respect to Vss (with regulator disabled)	0.3V to 3.6V
Total power dissipation (Note 1)	1W
Maximum current out of Vss pin	300 mA
Maximum current into VDD pin	250 mA
Input clamp current, liк (Vi < 0 or Vi > VDD)	±20 mA
Output clamp current, IOK (VO < 0 or VO > VDD)	±20 mA
Maximum output current sunk by PORTA<7:6> and any PORTB and PORTC I/O pins	25 mA
Maximum output current sunk by any PORTD, PORTE and PORTJ I/O pins	8 mA
Maximum output current sunk by PORTA<5:0> and any PORTF, PORTG and PORTH I/O pins	2 mA
Maximum output current sourced by PORTA<7:6> and any PORTB and PORTC I/O pins	25 mA
Maximum output current sourced by any PORTD, PORTE and PORTJ I/O pins	8 mA
Maximum output current sourced by PORTA<5:0> and any PORTF, PORTG and PORTH I/O pins	s2 mA
Maximum current sunk by all ports combined	200 mA

Note 1: Power dissipation is calculated as follows: Pdis = VDD x {IDD  $-\Sigma$  IOH} +  $\Sigma$  {(VDD - VOH) x IOH} +  $\Sigma$ (VOL x IOL)

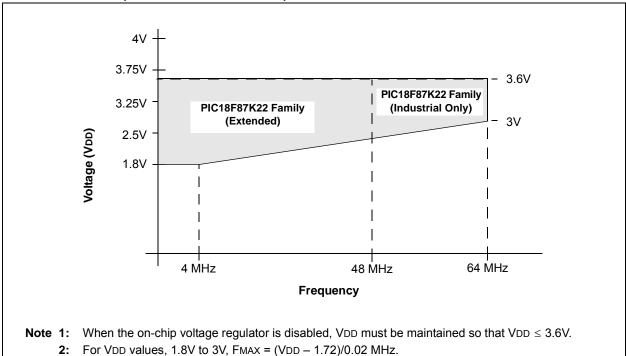
**† NOTICE:** Stresses above those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. This is a stress rating only and functional operation of the device at those or any other conditions above those indicated in the operation listings of this specification is not implied. Exposure to maximum rating conditions for extended periods may affect device reliability.





2: FMAX = 64 MHz in all other modes. For VDD values, 1.8V to 3V, FMAX = (VDD - 1.72)/0.02 MHz.

## FIGURE 31-2: VOLTAGE-FREQUENCY GRAPH, REGULATOR DISABLED (INDUSTRIAL/EXTENDED)<sup>(1,2)</sup>



### 31.1 DC Characteristics: Supply Voltage PIC18F87K22 Family (Industrial/Extended)

PIC18F87K22 Family (Industrial/Extended)		$\begin{array}{ll} \mbox{Standard Operating Conditions (unless otherwise state} \\ \mbox{Operating temperature} & -40^\circ C \leq TA \leq +85^\circ C \mbox{ for industria} \\ -40^\circ C \leq TA \leq +125^\circ C \mbox{ for extend} \end{array}$					
Param No.	Symbol	Characteristic	Min	Тур	Max	Units	Conditions
D001	Vdd	Supply Voltage	1.8 1.8	_	3.6 5.5	V V	ENVREG tied to Vss ENVREG tied to VDD
D001C	AVdd	Analog Supply Voltage	Vdd - 0.3	_	VDD + 0.3	V	
D001D	AVss	Analog Ground Potential	Vss – 0.3		Vss + 0.3	V	
D002	Vdr	RAM Data Retention Voltage <sup>(1)</sup>	1.5	_	—	V	
D003	VPOR	VDD Start Voltage to Ensure Internal Power-on Reset Signal	—	_	0.7	V	See Section 5.3 "Power-on Reset (POR)" for details
D004	Svdd	<b>VDD Rise Rate</b> to Ensure Internal Power-on Reset Signal	0.05	_	-	V/ms	See Section 5.3 "Power-on Reset (POR)" for details
D005	Bvdd	Brown-out Reset Voltage (High/Medium/ Low-Power mode) <sup>(2)</sup> BORV<1:0> = 11 <sup>(3)</sup> BORV<1:0> = 10 BORV<1:0> = 01 BORV<1:0> = 00	1.69 1.88 2.53 2.82	1.8 2.0 2.7 3.0	1.91 2.12 2.86 3.18		

Note 1: This is the limit to which VDD can be lowered in Sleep mode, or during a device Reset, without losing RAM data.

**2:** The following values are taken in HP-BOR mode.

3: The device will operate normally until Brown-out Reset occurs, even though VDD may be below VDDMIN.

	7K22 Family strial/Extended)											
Param Device		Тур	Max	Units	Conditio	ons						
	Power-Down Current (IPD) <sup>(1)</sup>											
	All devices	10	500	nA	-40°C							
		20	500	nA	+25°C	VDD = 1.8V <sup>(4)</sup>						
		120	600	nA	+60°C	(Sleep mode)						
		630	1800	nA	+85°C	Regulator Disabled						
		4	9	μA	+125°C							
	All devices	50	700	nA	-40°C							
		60	700	nA	+25°C	VDD = 3.3V <sup>(4)</sup>						
		170	800	nA	+60°C	(Sleep mode)						
		700	2700	nA	+85°C	Regulator Disabled						
		5	11	μA	+125°C							
	All devices	350	1300	nA	-40°C							
		400	1400	nA	+25°C	VDD = 5V <sup>(5)</sup>						
		550	1500	nA	+60°C	(Sleep mode)						
		1350	4000	nA	+85°C	Regulator Enabled						
		6	12	μA	+125°C							

**Note 1:** The power-down current in Sleep mode does not depend on the oscillator type. Power-down current is measured with the part in Sleep mode, with all I/O pins in a high-impedance state and tied to VDD or VSS, and all features that add delta current are disabled (such as WDT, SOSC oscillator, BOR, etc.).

2: The supply current is mainly a function of operating voltage, frequency and mode. Other factors, such as I/O pin loading and switching rate, oscillator type and circuit, internal code execution pattern and temperature, also have an impact on the current consumption.

The test conditions for all IDD measurements in Active Operation mode are:

OSC1 = External square wave, from rail-to-rail; all I/O pins tri-stated, pulled to VDD;

- MCLR = VDD; WDT enabled/disabled as specified.
- **3:** Standard, low-cost 32 kHz crystals have an operating temperature range of -10°C to +70°C. Extended temperature crystals are available at a much higher cost.

4: Voltage regulator disabled (ENVREG = 0, tied to Vss, RETEN (CONFIG1L<0>) = 1).

- 5: Voltage regulator enabled (ENVREG = 1, tied to VDD, SRETEN (WDTCON<4>) = 1 and RETEN (CONFIG1L<0>) = 0).
- 6: 48 MHz, maximum frequency at +125°C.

PIC18F87K22 Family (Industrial/Extended)		$ \begin{array}{ll} \mbox{Standard Operating Conditions (unless otherwise stated)} \\ \mbox{Operating temperature} & -40^{\circ}C \leq TA \leq +85^{\circ}C \mbox{ for industrial} \\ -40^{\circ}C \leq TA \leq +125^{\circ}C \mbox{ for extended} \end{array} $							
Param No.	Device	Тур	Max	Units		Conditions			
	Supply Current (IDD) <sup>(2,3)</sup>								
	All devices	5.3	10	μA	-40°C				
		5.5	10	μA	+25°C	VDD = 1.8V <sup>(4)</sup>			
		5.5	10	μA	+85°C	Regulator Disabled			
		12	24	μA	+125°C				
	All devices	10	15	μA	-40°C				
		10	16	μA	+25°C	VDD = 3.3V <sup>(4)</sup>	Fosc = 31 kHz ( <b>RC_RUN</b> mode,		
		11	17	μA	+85°C	Regulator Disabled	LF-INTOSC)		
		15	35	μA	+125°C		,		
	All devices	70	180	μA	-40°C				
		80	185	μA	+25°C	VDD = 5V <sup>(5)</sup>			
		90	190	μA	+85°C	Regulator Enabled			
		200	500	μA	+125°C				
	All devices	410	850	μA	-40°C				
		410	800	μA	+25°C	VDD = 1.8V <sup>(4)</sup>			
		410	830	μA	+85°C	Regulator Disabled			
		700	1500	μA	+125°C				
	All devices	680	990	μA	-40°C		Fosc = 1 MHz		
		680	960	μA	+25°C	VDD = 3.3V <sup>(4)</sup>	( <b>RC_RUN</b> mode, HF-INTOSC)		
		670	950	μA	+85°C	Regulator Disabled			
		800	1700	μA	+125°C		,		
	All devices	760	1400	μA	-40°C				
		780	1400	μA	+25°C	VDD = 5V <sup>(5)</sup>			
		800	1500	μA	+85°C	Regulator Enabled			
		1200	2400	μA	+125°C	<u> </u>			
	All devices	760	1300	μA	-40°C				
		760	1400	μA	+25°C	VDD = 1.8V <sup>(4)</sup>			
		770	1500	μA	+85°C	Regulator Disabled			
		800	1700	μA	+125°C				
	All devices	1.4	2.5	mA	-40°C		Fosc = 4 MHz		
		1.4	2.5	mA	+25°C	VDD = 3.3V <sup>(4)</sup>	( <b>RC_RUN</b> mode,		
		1.4	2.5	mA	+85°C	Regulator Disabled	HF-INTOSC)		
		1.5	3.0	mA	+125°C				
	All devices	1.5	2.7	mA	-40°C				
		1.5	2.7	mA	+25°C	VDD = 5V <sup>(5)</sup>			
		1.5	2.7	mA	+85°C	Regulator Enabled			
		1.6	3.3	mA	+125°C				

Note 1: The power-down current in Sleep mode does not depend on the oscillator type. Power-down current is measured with the part in Sleep mode, with all I/O pins in a high-impedance state and tied to VDD or Vss, and all features that add delta current are disabled (such as WDT, SOSC oscillator, BOR, etc.).

2: The supply current is mainly a function of operating voltage, frequency and mode. Other factors, such as I/O pin loading and switching rate, oscillator type and circuit, internal code execution pattern and temperature, also have an impact on the current consumption.

The test conditions for all IDD measurements in Active Operation mode are:

OSC1 = External square wave, from rail-to-rail; all I/O pins tri-stated, pulled to VDD;

MCLR = VDD; WDT enabled/disabled as specified.

3: Standard, low-cost 32 kHz crystals have an operating temperature range of -10°C to +70°C. Extended temperature crystals are available at a much higher cost.

4: Voltage regulator disabled (ENVREG = 0, tied to Vss, RETEN (CONFIG1L<0>) = 1).

5: Voltage regulator enabled (ENVREG = 1, tied to VDD, SRETEN (WDTCON<4>) = 1 and RETEN (CONFIG1L<0>) = 0).

**6:** 48 MHz, maximum frequency at +125°C.

PIC18F87K22 Family (Industrial/Extended)		Standard ( Operating f	•	otherwise stated) +85°C for industrial +125°C for extended			
Param No.	Device	Тур	Max	Units		Conditions	5
	Supply Current (IDD) Cont.	(2,3)					
	All devices	2.1	5.5	μA	-40°C		
		2.1	5.7	μA	+25°C	VDD = 1.8V <sup>(4)</sup>	
		2.2	6.0	μA	+85°C	Regulator Disabled	
		10	20	μA	+125°C		
	All devices	3.7	7.5	μA	-40°C		
		3.9	7.8	μA	+25°C	VDD = 3.3V <sup>(4)</sup>	Fosc = 31 kHz ( <b>RC_IDLE</b> mode,
		3.9	8.5	μA	+85°C	Regulator Disabled	LF-INTOSC)
		12	24	μA	+125°C		2
	All devices	70	180	μA	-40°C		
		80	190	μA	+25°C	VDD = 5V <sup>(5)</sup>	
		80	200	μA	+85°C	Regulator Enabled	
		200	420	μA	+125°C		
	All devices	330	650	μA	-40°C		
		330	640	μA	+25°C	VDD = 1.8V <sup>(4)</sup>	
		330	630	μA	+85°C	Regulator Disabled	
		500	850	μA	+125°C		
	All devices	520	850	μA	-40°C		Fosc = 1 MHz ( <b>RC_IDLE</b> mode, HF-INTOSC)
		520	900	μA	+25°C	VDD = 3.3V <sup>(4)</sup>	
		520	850	μA	+85°C	Regulator Disabled	
		800	1200	μA	+125°C		,
	All devices	590	940	μA	-40°C		
		600	960	μA	+25°C	VDD = 5V <sup>(5)</sup>	
		620	990	μA	+85°C	Regulator Enabled	
		1000	1400	μA	+125°C		
	All devices	470	770	μA	-40°C		
		470	770	μA	+25°C	VDD = 1.8V <sup>(4)</sup>	
		460	760	μA	+85°C	Regulator Disabled	
		700	1000	μA	+125°C		
	All devices	800	1400	μA	-40°C		Fosc = 4 MHz
		800	1350	μA	+25°C	VDD = 3.3V <sup>(4)</sup>	( <b>RC_IDLE</b> mode,
		790	1300	μA	+85°C	Regulator Disabled	internal HF-INTOSC)
		1100	1400	μA	+125°C		)
	All devices	880	1600	μA	-40°C		
		890	1700	μA	+25°C	VDD = 5V <sup>(5)</sup>	
		910	1800	μA	+85°C	Regulator Enabled	
		1200	2200	μA	+125°C		

Note 1: The power-down current in Sleep mode does not depend on the oscillator type. Power-down current is measured with the part in Sleep mode, with all I/O pins in a high-impedance state and tied to VDD or Vss, and all features that add delta current are disabled (such as WDT, SOSC oscillator, BOR, etc.).

2: The supply current is mainly a function of operating voltage, frequency and mode. Other factors, such as I/O pin loading and switching rate, oscillator type and circuit, internal code execution pattern and temperature, also have an impact on the current consumption.

The test conditions for all IDD measurements in Active Operation mode are:

OSC1 = External square wave, from rail-to-rail; all I/O pins tri-stated, pulled to VDD;

MCLR = VDD; WDT enabled/disabled as specified.

3: Standard, low-cost 32 kHz crystals have an operating temperature range of -10°C to +70°C. Extended temperature crystals are available at a much higher cost.

4: Voltage regulator disabled (ENVREG = 0, tied to Vss, RETEN (CONFIG1L<0>) = 1).

5: Voltage regulator enabled (ENVREG = 1, tied to VDD, SRETEN (WDTCON<4>) = 1 and RETEN (CONFIG1L<0>) = 0).

**6:** 48 MHz, maximum frequency at +125°C.

PIC18F87K22 Family (Industrial/Extended)		$\begin{array}{llllllllllllllllllllllllllllllllllll$							
Param No.	Device	Тур	Max	Units		Conditions			
	Supply Current (IDD) Cont	(2,3)							
	All devices		390	μA	-40°C				
		130	390	μA	+25°C	VDD = 1.8V <sup>(4)</sup>			
		130	390	μA	+85°C	Regulator Disabled			
		250	500	μA	+125°C	1			
	All devices	270	790	μA	-40°C				
		270	790	μA	+25°C	VDD = 3.3V <sup>(4)</sup>	Fosc = 1 MHz		
		270	790	μA	+85°C	Regulator Disabled	( <b>PRI_RUN</b> mode, EC oscillator)		
		400	900	μA	+125°C				
	All devices	430	990	μA	-40°C				
		450	980	μA	+25°C	Vdd = 5V <sup>(5)</sup>			
		460	980	μA	+85°C	Regulator Enabled			
		600	1300	μA	+125°C	]			
	All devices	430	860	μA	-40°C				
		530	900	μA	+25°C	VDD = 1.8V <sup>(4)</sup>			
		490	880	μA	+85°C	Regulator Disabled			
		750	1600	μA	+125°C				
	All devices	850	1750	μA	-40°C				
		850	1700	μA	+25°C	VDD = 3.3V <sup>(4)</sup>	Fosc = 4 MHz ( <b>PRI_RUN</b> mode,		
		850	1800	μA	+85°C	Regulator Disabled	EC oscillator)		
		1150	2400	μA	+125°C				
	All devices	1.1	2.7	mA	-40°C				
		1.1	2.6	mA	+25°C	VDD = 5V <sup>(5)</sup>			
		1.1	2.6	mA	+85°C	Regulator Enabled			
		2.0	4.0	mA	+125°C				
	All devices	12	19	mA	-40°C	_			
		12	19	mA	+25°C	VDD = 3.3V <sup>(4)</sup>			
		12	19	mA	+85°C	Regulator Disabled			
		13	22	mA	+125°C <sup>(6)</sup>		Fosc = 64 MHz ( <b>PRI_RUN</b> mode,		
	All devices	13	20	mA	-40°C		EC oscillator)		
		13	20	mA	+25°C	VDD = 5V <sup>(4)</sup>			
		13	20	mA	+85°C	Regulator Enabled			
		14	23	mA	+125°C <sup>(6)</sup>				

Note 1: The power-down current in Sleep mode does not depend on the oscillator type. Power-down current is measured with the part in Sleep mode, with all I/O pins in a high-impedance state and tied to VDD or Vss, and all features that add delta current are disabled (such as WDT, SOSC oscillator, BOR, etc.).

2: The supply current is mainly a function of operating voltage, frequency and mode. Other factors, such as I/O pin loading and switching rate, oscillator type and circuit, internal code execution pattern and temperature, also have an impact on the current consumption.

The test conditions for all IDD measurements in Active Operation mode are:

<u>OSC1</u> = External square wave, from rail-to-rail; all I/O pins tri-stated, pulled to VDD;

MCLR = VDD; WDT enabled/disabled as specified.

- **3:** Standard, low-cost 32 kHz crystals have an operating temperature range of -10°C to +70°C. Extended temperature crystals are available at a much higher cost.
- 4: Voltage regulator disabled (ENVREG = 0, tied to Vss, RETEN (CONFIG1L<0>) = 1).
- 5: Voltage regulator enabled (ENVREG = 1, tied to VDD, SRETEN (WDTCON<4>) = 1 and RETEN (CONFIG1L<0>) = 0).
- 6: 48 MHz, maximum frequency at +125°C.

31.2

### DC Characteristics: Power-Down and Supply Current PIC18F87K22 Family (Industrial/Extended) (Continued)

PIC18F87K22 Family (Industrial/Extended)		$ \begin{array}{ll} \mbox{Standard Operating Conditions (unless otherwise stated)} \\ \mbox{Operating temperature} & -40^{\circ}C \leq TA \leq +85^{\circ}C \mbox{ for industrial} \\ -40^{\circ}C \leq TA \leq +125^{\circ}C \mbox{ for extended} \end{array} $								
Param No.	Device		Max	Units	Conditions					
	Supply Current (IDD) Cont	(2,3)								
	All devices	3.3	5.6	mA	-40°C					
		3.3	5.5	mA	+25°C	VDD = 3.3V <sup>(4)</sup>				
		3.3	5.5	mA	+85°C	Regulator Disabled	Fosc = 16 MHz, ( <b>PRI_RUN</b> mode, 4 MHz EC oscillator with PLL)			
		3.6	6.0	mA	+125°C					
	All devices	3.5	5.9	mA	-40°C					
		3.5	5.8	mA	+25°C	Vdd = 5V <sup>(5)</sup>				
		3.5	5.8	mA	+85°C	Regulator Enabled				
		3.8	7.0	mA	+125°C					
	All devices	12	18	mA	-40°C					
		12	18	mA	+25°C	VDD = 3.3V <sup>(4)</sup>				
		12	18	mA	+85°C	Regulator Disabled				
		13	22	mA	+125°C <sup>(6)</sup>	V <sub>DD</sub> = 5V <sup>(5)</sup> Regulator Enabled	Fosc = 64 MHz,			
	All devices	13	20	mA	-40°C		<ul> <li>(PRI_RUN mode, 16 MHz EC oscillator with PLL)</li> </ul>			
		13	20	mA	+25°C					
		13	20	mA	+85°C					
		14	24	mA	+125°C <sup>(6)</sup>					

Note 1: The power-down current in Sleep mode does not depend on the oscillator type. Power-down current is measured with the part in Sleep mode, with all I/O pins in a high-impedance state and tied to VDD or Vss, and all features that add delta current are disabled (such as WDT, SOSC oscillator, BOR, etc.).

2: The supply current is mainly a function of operating voltage, frequency and mode. Other factors, such as I/O pin loading and switching rate, oscillator type and circuit, internal code execution pattern and temperature, also have an impact on the current consumption.

The test conditions for all IDD measurements in Active Operation mode are:

OSC1 = External square wave, from rail-to-rail; all I/O pins tri-stated, pulled to VDD;

MCLR = VDD; WDT enabled/disabled as specified.

- **3:** Standard, low-cost 32 kHz crystals have an operating temperature range of -10°C to +70°C. Extended temperature crystals are available at a much higher cost.
- 4: Voltage regulator disabled (ENVREG = 0, tied to Vss, RETEN (CONFIG1L<0>) = 1).
- 5: Voltage regulator enabled (ENVREG = 1, tied to VDD, SRETEN (WDTCON<4>) = 1 and RETEN (CONFIG1L<0>) = 0).
- 6: 48 MHz, maximum frequency at +125°C.

PIC18F87K22 Family (Industrial/Extended)		$\begin{array}{llllllllllllllllllllllllllllllllllll$							
Param No.	Device	Тур	Max	Units	Conditions				
	Supply Current (IDD) Cont.	(2,3)							
	All devices	42	73	μA	-40°C				
		42	73	μA	+25°C	VDD = 1.8V <sup>(4)</sup>			
		43	74	μA	+85°C	Regulator Disabled			
		53	100	μA	+125°C				
	All devices	110	190	μA	-40°C				
		110	195	μA	+25°C	VDD = 3.3V <sup>(4)</sup>	Fosc = 1 MHz		
		110	195	μA	+85°C	Regulator Disabled	(PRI_IDLE mode, EC oscillator)		
		130	250	μA	+125°C				
	All devices	280	450	μA	-40°C				
		290	440	μA	+25°C	Vdd = 5V <sup>(5)</sup>			
		300	460	μA	+85°C	Regulator Enabled			
		330	500	μA	+125°C				
	All devices	160	360	μA	-40°C				
		160	360	μA	+25°C	VDD = 1.8V <sup>(4)</sup>			
		170	370	μA	+85°C	Regulator Disabled			
		200	400	μA	+125°C				
	All devices	330	650	μA	-40°C				
		340	660	μA	+25°C	VDD = 3.3V <sup>(4)</sup>	Fosc = 4 MHz ( <b>PRI IDLE</b> mode,		
		340	660	μA	+85°C	Regulator Disabled	EC oscillator)		
		370	700	μA	+125°C				
	All devices	510	900	μA	-40°C				
		520	950	μA	+25°C	VDD = 5V <sup>(5)</sup>			
		540	990	μA	+85°C	Regulator Enabled			
		600	1200	μA	+125°C				
	All devices	4.7	9	mA	-40°C				
		4.8	9	mA	+25°C	VDD = 3.3V <sup>(4)</sup>			
		4.8	10	mA	+85°C	Regulator Disabled	Fosc = 64 MHz		
		5.2	12	mA	+125°C <sup>(6)</sup>		( <b>PRI_IDLE</b> mode,		
	All devices	5.1	11	mA	-40°C		EC oscillator)		
		5.1	11	mA	+25°C	VDD = 5V <sup>(5)</sup>			
		5.2	12	mA	+85°C	Regulator Enabled			
		5.7	14	mA	+125⁰C <sup>(6)</sup>				

Note 1: The power-down current in Sleep mode does not depend on the oscillator type. Power-down current is measured with the part in Sleep mode, with all I/O pins in a high-impedance state and tied to VDD or Vss, and all features that add delta current are disabled (such as WDT, SOSC oscillator, BOR, etc.).

2: The supply current is mainly a function of operating voltage, frequency and mode. Other factors, such as I/O pin loading and switching rate, oscillator type and circuit, internal code execution pattern and temperature, also have an impact on the current consumption.

The test conditions for all IDD measurements in Active Operation mode are:

OSC1 = External square wave, from rail-to-rail; all I/O pins tri-stated, pulled to VDD;

MCLR = VDD; WDT enabled/disabled as specified.

3: Standard, low-cost 32 kHz crystals have an operating temperature range of -10°C to +70°C. Extended temperature crystals are available at a much higher cost.

4: Voltage regulator disabled (ENVREG = 0, tied to Vss, RETEN (CONFIG1L<0>) = 1).

5: Voltage regulator enabled (ENVREG = 1, tied to VDD, SRETEN (WDTCON<4>) = 1 and RETEN (CONFIG1L<0>) = 0).

6: 48 MHz, maximum frequency at +125°C.

	7K22 Family strial/Extended)	$\begin{array}{llllllllllllllllllllllllllllllllllll$								
Param No.	Device	Тур	Max	Units		Conditions				
	Supply Current (IDD) Cont	(2,3)								
	All devices	3.7	8.5	μA	-40°C					
		5.4	10	μA	+25°C	VDD = 1.8V <sup>(4)</sup>				
		6.6	13	μA	+85°C	Regulator Disabled				
		13	30	μA	+125°C					
	All devices	8.7	18	μA	-40°C		Fosc = 32 kHz <sup>(3)</sup> ( <b>SEC_RUN</b> mode, SOSCSEL = 01)			
		10	20	μA	+25°C	VDD = 3.3V <sup>(4)</sup>				
		12	23	μA	+85°C	Regulator Disabled				
		25	60	μA	+125°C					
	All devices	60	160	μA	-40°C					
		90	190	μA	+25°C	VDD = 5V <sup>(4)</sup>				
		100	240	μA	+85°C	Regulator Enabled				
		200	450	μA	+125°C					
	All devices		4	μA	-40°C					
		1.7	5	μA	+25°C	VDD = 1.8V <sup>(4)</sup>				
		2.6	6	μA	+85°C	Regulator Disabled				
		9	20	μA	+125°C					
	All devices	-	7	μA	-40°C		Fosc = 32 kHz <sup>(3)</sup>			
		2.8	9	μA	+25°C	VDD = 3.3V <sup>(4)</sup>	(SEC IDLE mode,			
		4.1	10	μA	+85°C	Regulator Disabled	SOSCSEL = 01)			
		17	40	μA	+125°C		/			
	All devices		150	μA	-40°C					
		80	180	μA	+25°C	Vdd = 5V <sup>(5)</sup>				
		100	240	μA	+85°C	Regulator Enabled				
		180	440	μA	+125°C					

Note 1: The power-down current in Sleep mode does not depend on the oscillator type. Power-down current is measured with the part in Sleep mode, with all I/O pins in a high-impedance state and tied to VDD or Vss, and all features that add delta current are disabled (such as WDT, SOSC oscillator, BOR, etc.).

2: The supply current is mainly a function of operating voltage, frequency and mode. Other factors, such as I/O pin loading and switching rate, oscillator type and circuit, internal code execution pattern and temperature, also have an impact on the current consumption.

The test conditions for all IDD measurements in Active Operation mode are:

<u>OSC1</u> = External square wave, from rail-to-rail; all I/O pins tri-stated, pulled to VDD;

MCLR = VDD; WDT enabled/disabled as specified.

3: Standard, low-cost 32 kHz crystals have an operating temperature range of -10°C to +70°C. Extended temperature crystals are available at a much higher cost.

4: Voltage regulator disabled (ENVREG = 0, tied to Vss, RETEN (CONFIG1L<0>) = 1).

5: Voltage regulator enabled (ENVREG = 1, tied to VDD, SRETEN (WDTCON<4>) = 1 and RETEN (CONFIG1L<0>) = 0).

**6:** 48 MHz, maximum frequency at +125°C.

	7K22 Family strial/Extended)	Standard ( Operating t	•	•	tions (unless otherwise stated) -40°C $\leq$ TA $\leq$ +85°C for industrial -40°C $\leq$ TA $\leq$ +125°C for extended					
Param No.	Device	Тур	Max	Units		Conditions				
	Module Differential Currents (ΔΙωστ, ΔΙΒΟR, ΔΙΗLVD, ΔΙΟSCΒ, ΔΙΑΟ)									
	Watchdog Timer									
D022	All devices	0.3	1	μA	-40°C					
(∆IWDT)		0.3	1	μA	+25°C	VDD = 1.8V <sup>(4)</sup>				
D022		0.3	1	μA	+85°C	Regulator Disabled				
(∆IWDT)		0.5	2	μA	+125°C					
	All devices	0.6	2	μA	-40°C					
		0.6	2	μA	+25°C	VDD = 3.3V <sup>(4)</sup>				
		0.7	2	μA	+85°C	Regulator Disabled				
		1	3	μA	+125°C					
	All Devices	0.6	2	μA	-40°C					
		0.6	2	μA	+25°C	VDD = 5V <sup>(5)</sup>				
		0.7	2	μA	+85°C	Regulator Enabled				
		1.5	4	μA	+125°C					
D022A	Brown-out Reset		•							
( $\Delta$ IBOR)	All devices	4.6	19	μA	-40°C					
( $\Delta$ IBOR)		4.5	20	μA	+25°C	VDD = 3.3V <sup>(4)</sup>	Linh Dower DOD			
		4.7	20	μA	+85°C	Regulator Disabled	High-Power BOR			
		18	40	μA	+125°C					
	All devices	4.2	20	μA	-40°C					
		4.3	20	μA	+25°C	VDD = 5V <sup>(5)</sup>	Linh Dower DOD			
		4.4	20	μA	+85°C	Regulator Enabled	High-Power BOR			
		20	40	μA	+125°C					
D022B	High/Low-Voltage Detect		•							
(∆Ihlvd)	All devices	3.8	9	μA	-40°C					
		4.2	9	μA	+25°C	VDD = 1.8V <sup>(4)</sup>				
		4.3	10	μA	+85°C	Regulator Disabled				
		4.5	12	μA	+125°C					
	All devices	4.5	11	μA	-40°C					
		4.8	12	μA	+25°C	VDD = 3.3V <sup>(4)</sup>				
		4.8	12	μA	+85°C	Regulator Disabled				
		5.0	14	μA	+125°C	7				
	All devices	4.9	13	μA	-40°C					
		4.9	13	μA	+25°C	VDD = 5√ <sup>(5)</sup>				
		4.9	13	μA	+85°C	Regulator Enabled				
		5.3	15	μA	+125°C					

Note 1: The power-down current in Sleep mode does not depend on the oscillator type. Power-down current is measured with the part in Sleep mode, with all I/O pins in a high-impedance state and tied to VDD or Vss, and all features that add delta current are disabled (such as WDT, SOSC oscillator, BOR, etc.).

2: The supply current is mainly a function of operating voltage, frequency and mode. Other factors, such as I/O pin loading and switching rate, oscillator type and circuit, internal code execution pattern and temperature, also have an impact on the current consumption.

The test conditions for all IDD measurements in Active Operation mode are:

OSC1 = External square wave, from rail-to-rail; all I/O pins tri-stated, pulled to VDD;

MCLR = VDD; WDT enabled/disabled as specified.

- **3:** Standard, low-cost 32 kHz crystals have an operating temperature range of -10°C to +70°C. Extended temperature crystals are available at a much higher cost.
- 4: Voltage regulator disabled (ENVREG = 0, tied to Vss, RETEN (CONFIG1L<0>) = 1).
- 5: Voltage regulator enabled (ENVREG = 1, tied to VDD, SRETEN (WDTCON<4>) = 1 and RETEN (CONFIG1L<0>) = 0).
- **6:** 48 MHz, maximum frequency at +125°C.

PIC18F87K22 Family (Industrial/Extended)		$\begin{array}{llllllllllllllllllllllllllllllllllll$								
Param No.	Device	Тур	Max	Units	Conditions					
D025	Real-Time Clock/Calendar	with SOSC	C Oscilla	itor						
$(\Delta   RTCC)$	All devices	0.7	2.7	μA	-40°C					
		0.7	2.8	μA	+25°C	$V_{DD} = 1.8 \sqrt{(4)}$	00 700 11 1-			
		1.1	2.8	μA	+60°C	Regulator Disabled				
		1.1	2.9	μA	+85°C					
		2.2	4.4	μA	+125°C					
	All devices	1.2	2.9	μA	-40°C					
		1.1	2.8	μA	+25°C	VDD = 3.3V <sup>(4)</sup>				
		2	4.6	μA	+60°C	Regulator Disabled	32.768 kHz, SOSCRUN = 1			
		2	4.8	μA	+85°C					
		4	6.5	μA	+125°C					
	All devices	1.5	4.4	μA	-40°C					
		1.5	4.4	μA	+25°C	VDD = 5V <sup>(5)</sup>				
		1.7	4.7	μA	+60°C	Regulator Enabled				
		1.7	4.7	μA	+85°C					
		3.5	6.9	μA	+125°C					

**Note 1:** The power-down current in Sleep mode does not depend on the oscillator type. Power-down current is measured with the part in Sleep mode, with all I/O pins in a high-impedance state and tied to VDD or VSS, and all features that add delta current are disabled (such as WDT, SOSC oscillator, BOR, etc.).

2: The supply current is mainly a function of operating voltage, frequency and mode. Other factors, such as I/O pin loading and switching rate, oscillator type and circuit, internal code execution pattern and temperature, also have an impact on the current consumption.

The test conditions for all IDD measurements in Active Operation mode are:

OSC1 = External square wave, from rail-to-rail; all I/O pins tri-stated, pulled to VDD;

- MCLR = VDD; WDT enabled/disabled as specified.
- 3: Standard, low-cost 32 kHz crystals have an operating temperature range of -10°C to +70°C. Extended temperature crystals are available at a much higher cost.
- 4: Voltage regulator disabled (ENVREG = 0, tied to Vss, RETEN (CONFIG1L<0>) = 1).
- 5: Voltage regulator enabled (ENVREG = 1, tied to VDD, SRETEN (WDTCON<4>) = 1 and RETEN (CONFIG1L<0>) = 0).
- 6: 48 MHz, maximum frequency at +125°C.

### 31.3 DC Characteristics: PIC18F87K22 Family (Industrial/Extended)

DC CHA	RACTE	RISTICS		perature -4	l0°C ≤ 1	s (unless otherwise stated) $\Gamma A \le +85^{\circ}C$ for industrial $\Gamma A \le +125^{\circ}C$ for extended
Param No.	Symbol	Characteristic	Min	Max	Units	Conditions
	VIL	Input Low Voltage				
		All I/O Ports:				
D030		with TTL Buffer	Vss	0.15 VDD	V	VDD < 4.5V
D031		with Schmitt Trigger Buffer	_	0.8	V	$4.5V \le VDD \le 5.5V$
		RC3, RC4	Vss	0.2 VDD	V	VDD < 4.5
		RD5, RD6	Vss	1.5	V	$4.5V \le VDD \le 5.5V$
D031A		RC3, RC4	Vss	0.3 VDD	V	I <sup>2</sup> C™ enabled
D031B		RD5, RD6	Vss	0.8	V	SMBus enabled
D032		MCLR	Vss	0.2 Vdd	V	
D033		OSC1	Vss	0.2 Vdd	V	LP, XT, HS modes
D033A		OSC1	Vss	0.2 Vdd	V	EC modes
D034		SOSCI	Vss	0.3 Vdd	V	
	VIH	Input High Voltage				
		I/O Ports with Analog Functions:				
D040		with TTL Buffer	0.25 VDD	Vdd	V	VDD < 4.5V
			2.0	Vdd	V	$4.5V \le VDD \le 5.5V$
D041		with Schmitt Trigger Buffer	0.8	Vdd	V	VDD < 4.5V
		RC3, RC4	0.7 Vdd	Vdd	V	VDD < 4.5V
		RD5, RD6	3V	5.5	V	$4.5V \le VDD \le 5.5V$
D041A		RC3, RC4	0.7 Vdd	Vdd	V	I <sup>2</sup> C enabled
D041B		RD5, RD6	2.1	Vdd	V	SMBus enabled
D042		MCLR	0.8 VDD	Vdd	V	
D043		OSC1	0.9 Vdd	Vdd	V	RC mode
D043A		OSC1	0.7 Vdd	Vdd	V	HS mode
D044		SOSCI	0.7 Vdd	Vdd	V	
	lı∟	Input Leakage Current <sup>(1)</sup>				
D060		I/O Ports	±50	±200	nA	$\label{eq:VSS} \begin{split} &V{\sf SS} \leq V{\sf PIN} \leq V{\sf DD}, \\ &P{\sf in} \mbox{ at high-impedance} \end{split}$
D061		MCLR	—	±5	μA	$Vss \le VPIN \le VDD, +85^{\circ}C$
D063		OSC1	_	±5	μA	$Vss \leq V PIN \leq V DD$
	IPU	Weak Pull-up Current				
D070	IPURB	PORTB Weak Pull-up Current	50	400	μA	VDD = 3.3V, VPIN = VSS

Note 1: Negative current is defined as current sourced by the pin.

### 31.3 DC Characteristics: PIC18F87K22 Family (Industrial/Extended) (Continued)

DC CHA	DC CHARACTERISTICS			$\begin{array}{l} \mbox{Standard Operating Conditions (unless otherwise stated)}\\ \mbox{Operating temperature} & -40^{\circ}C \leq TA \leq +85^{\circ}C \mbox{ for industrial}\\ & -40^{\circ}C \leq TA \leq +125^{\circ}C \mbox{ for extended} \end{array}$				
Param No.	Symbol	Characteristic	Min	Max	Units	Conditions		
	Vol	Output Low Voltage						
D080		I/O Ports:						
		PORTA, PORTB, PORTC	—	0.6	V	IOL = 8.5 mA, VDD = 4.5V, -40°C to +125°C		
		PORTD, PORTE, PORTF, PORTG, PORTH, PORTJ	—	0.6	V	IOL = 3.5 mA, VDD = 4.5V, -40°C to +125°C		
D083		OSC2/CLKO (EC modes)	—	0.6	V	IOL = 1.6 mA. VDD = 5.5V, -40°C to +125°C		
	Vон	Output High Voltage <sup>(1)</sup>						
D090		I/O Ports:			V			
		PORTA, PORTB, PORTC	Vdd - 0.7	—	V	IOH = -3 mA, VDD = 4.5V, -40°С to +125°С		
		PORTD, PORTE, PORTF, PORTG, PORTH, PORTJ	Vdd - 0.7	—	V	IOH = -2 mA, VDD = 4.5V, -40°C to +125°C		
D092		OSC2/CLKO (INTOSC, EC modes)	Vdd - 0.7	—	V	IOH = -1 mA, VDD = 5.5V, -40°C to +125°C		
		Capacitive Loading Specs on Output Pins						
D100	COSC2	OSC2 Pin		20	pF	In HS mode when external clock is used to drive OSC1		
D101	Сю	All I/O Pins and OSC2	_	50	pF	To meet the AC Timing Specifications		
D102	Св	SCLx, SDAx	—	400	pF	I <sup>2</sup> C <sup>™</sup> Specification		

**Note 1:** Negative current is defined as current sourced by the pin.

### 31.4 DC Characteristics: CTMU Current Source Specifications

DC CH	ARACT	ERISTICS	$\begin{array}{ll} \mbox{Standard Operating Conditions: 1.8V to 5.5V (unless otherwise stated)} \\ \mbox{Operating temperature} & -40^\circ C \leq TA \leq +85^\circ C \mbox{ for industrial} \\ -40^\circ C \leq TA \leq +125^\circ C \mbox{ for extended} \end{array}$					
Param No.SymCharacteristicMinTyp(1)MaxUn		Units	Conditions					
	IOUT1	CTMU Current Source, Base Range	_	550	_	nA	CTMUICON<1:0> = 01	
	IOUT2	CTMU Current Source, 10x Range	—	5.5	—	μA	CTMUICON<1:0> = 10	
	Ιουτ3	CTMU Current Source, 100x Range	—	55		μA	CTMUICON<1:0> = 11	

**Note 1:** Nominal value at center point of current trim range (CTMUICON<7:2> = 000000).

DC CH/				$\begin{array}{l} \textbf{Standard Operating Conditions (unless otherwise stated)}\\ \textbf{Operating temperature } -40^{\circ}\text{C} \leq \text{TA} \leq +85^{\circ}\text{C} \text{ for industrial}\\ -40^{\circ}\text{C} \leq \text{TA} \leq +125^{\circ}\text{C} \text{ for extended} \end{array}$					
Param No. Sym		Characteristic	Min	Тур†	Max	Units	Conditions		
		Internal Program Memory Programming Specifications <sup>(1)</sup>							
D110	VPP	Voltage on MCLR/VPP/RE5 pin	VDD + 1.5	_	10	V	(Note 3, Note 4)		
D113	IDDP	Supply Current during Programming	—	—	10	mA			
		Data EEPROM Memory					(Note 2)		
D120	ED	Byte Endurance	100K	_	—	E/W	-40°C to +125°C		
D121	Vdrw	VDD for Read/Write	1.8	—	5.5	V	Using EECON to read/ write, ENVREG tied to VDD		
			1.8	—	3.6	V	Using EECON to read/ write, ENVREG tied to Vss		
D122	TDEW	Erase/Write Cycle Time	—	4	_	ms			
D123	TRETD	Characteristic Retention	40	—	—	Year	Provided no other specifications are violated		
D124	TREF	Number of Total Erase/Write Cycles before Refresh <sup>(2)</sup>	1M	10M	—	E/W	-40°C to +125°C		
		Program Flash Memory							
D130	Eр	Cell Endurance	10K	_	_	E/W	-40°C to +125°C		
D131	VPR	VDD for Read	1.8	—	5.5	V	ENVREG tied to VDD		
			1.8	—	3.6	V	ENVREG tied to Vss		
D132B	VPEW	Voltage for Self-Timed Erase or Write Operations							
_		VDD	1.8		5.5	V	ENVREG tied to VDD		
D133A	Tiw	Self-Timed Write Cycle Time	—	2	—	ms			
D134	TRETD	Characteristic Retention	40	_	_	Year	Provided no other specifications are violated		
D135	IDDP	Supply Current during Programming	_	_	10	mA			
D140	TWE	Writes per Erase Cycle	—	_	1		For each physical address		

### TABLE 31-1: MEMORY PROGRAMMING REQUIREMENTS

† Data in "Typ" column is at 3.3V, 25°C unless otherwise stated. These parameters are for design guidance only and are not tested.

**Note 1:** These specifications are for programming the on-chip program memory through the use of table write instructions.

2: Refer to Section 9.8 "Using the Data EEPROM" for a more detailed discussion on data EEPROM endurance.

**3:** Required only if Single-Supply Programming is disabled.

4: The MPLAB<sup>®</sup> ICD 2 does not support variable VPP output. Circuitry to limit the MPLAB ICD 2 VPP voltage must be placed between the MPLAB ICD 2 and the target system when programming or debugging with the MPLAB ICD 2.

### TABLE 31-2: COMPARATOR SPECIFICATIONS

Operating	<b>Operating Conditions:</b> $1.8V \le VDD \le 5V$ , $-40^{\circ}C \le TA \le +125^{\circ}C$ (unless otherwise stated)							
Param No.	Sym	Characteristics	Min	Тур	Мах	Units	Comments	
D300	VIOFF	Input Offset Voltage	_	±5.0	40	mV		
D301	VICM	Input Common Mode Voltage	_		AVDD - 1.5	V		
D302	CMRR	Common Mode Rejection Ratio	55	_	—	dB		
D303	TRESP	Response Time <sup>(1)</sup>	—	150	400	ns		
D304	TMC2OV	Comparator Mode Change to Output Valid*	—	—	10	μS		

**Note 1:** Response time measured with one comparator input at (AVDD – 1.5)/2, while the other input transitions from Vss to VDD.

### TABLE 31-3: VOLTAGE REFERENCE SPECIFICATIONS

Operating	<b>Operating Conditions:</b> $1.8V \le VDD \le 5V$ , $-40^{\circ}C \le TA \le +125^{\circ}C$ (unless otherwise stated)								
Param No.	Sym	Characteristics	Min	Тур	Max	Units	Comments		
D310	VRES	Resolution	—	VDD/32		LSb			
D311	VRAA	Absolute Accuracy	_	—	1/2	LSb			
D312	VRur	Unit Resistor Value (R)	_	2k	_	Ω			
D313	TSET	Settling Time <sup>(1)</sup>	—	—	10	μS			

**Note 1:** Settling time measured while CVRR = 1 and CVR<3:0> transitions from '0000' to '1111'.

### TABLE 31-4: INTERNAL VOLTAGE REGULATOR SPECIFICATIONS

Operatir	<b>Operating Conditions:</b> -40°C $\leq$ TA $\leq$ +125°C (unless otherwise stated)							
Param No.	Sym	Characteristics	Min	Тур	Max	Units	Comments	
	VRGOUT	Regulator Output Voltage		3.3		V		
	CEFC	External Filter Capacitor Value	4.7	10		μF	Capacitor must be low-ESR, a low series resistance (< 5Ω)	

### 31.5 AC (Timing) Characteristics

### 31.5.1 TIMING PARAMETER SYMBOLOGY

The timing parameter symbols have been created following one of the following formats:

1. TppS2ppS		3. TCC:ST	(I <sup>2</sup> C specifications only)
2. TppS		4. Ts	(I <sup>2</sup> C specifications only)
Т			
F	Frequency	Т	Time
Lowercase le	etters (pp) and their meanings:		
рр			
сс	CCP1	osc	OSC1
ck	CLKO	rd	RD
cs	CS	rw	RD or WR
di	SDI	sc	SCK
do	SDO	SS	SS
dt	Data in	tO	TOCKI
io	I/O port	t1	T1CKI
mc	MCLR	wr	WR
Uppercase le	etters and their meanings:		
S			
F	Fall	Р	Period
Н	High	R	Rise
I	Invalid (High-impedance)	V	Valid
L	Low	Z	High-impedance
I <sup>2</sup> C only			
AA	output access	High	High
BUF	Bus free	Low	Low
Tcc:s⊤ (I <sup>2</sup> C s	pecifications only)		
CC			
HD	Hold	SU	Setup
ST			
DAT	DATA input hold	STO	Stop condition
STA	Start condition		

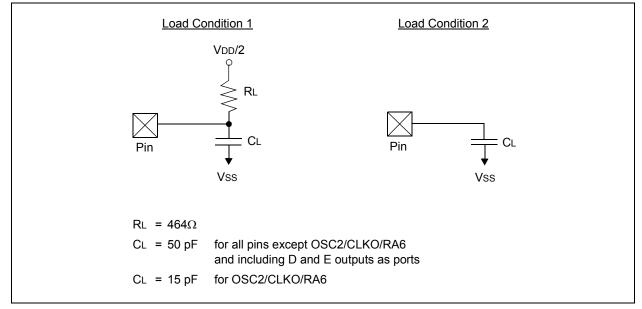
### 31.5.2 TIMING CONDITIONS

The temperature and voltages specified in Table 31-5 apply to all timing specifications unless otherwise noted. Figure 31-3 specifies the load conditions for the timing specifications.

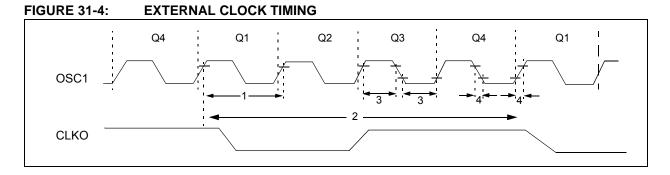
### TABLE 31-5: TEMPERATURE AND VOLTAGE SPECIFICATIONS – AC

	Standard Operating Conditions (unless otherwise stated)						
AC CHARACTERISTICS	Operating temperature $-40^{\circ}C \le TA \le +85^{\circ}C$ for industrial						
AC CHARACTERISTICS	$-40^{\circ}C \le TA \le +125^{\circ}C$ for extended						
	Operating voltage VDD range as described in Section 31.1 and Section 31.3.						

#### FIGURE 31-3: LOAD CONDITIONS FOR DEVICE TIMING SPECIFICATIONS



## 31.5.3 TIMING DIAGRAMS AND SPECIFICATIONS



### TABLE 31-6: EXTERNAL CLOCK TIMING REQUIREMENTS

Param. No.	Symbol	Characteristic	Min	Max	Units	Conditions
1A	Fosc	External CLKIN Frequency <sup>(1)</sup>	DC	64	MHz	EC, ECIO Oscillator mode -40°C ≤ TA ≤ +85°C
			DC	48	MHz	-40°C ≤ TA ≤ +125°C
		Oscillator Frequency <sup>(1)</sup>	DC	4	MHz	RC Oscillator mode
			0.1	4	MHz	XT Oscillator mode
			4	16	MHz	HS Oscillator mode
			4	16	MHz	HS + PLL Oscillator mode
			5	33	kHz	LP Oscillator mode
1	Tosc	External CLKIN Period <sup>(1)</sup> Oscillator Period <sup>(1)</sup>	15.6	—	ns	EC, ECIO Oscillator mode
			250	—	ns	RC Oscillator mode
			250	10,000	ns	XT Oscillator mode
			40 62.5	250 250	ns ns	HS Oscillator mode HS + PLL Oscillator mode
			5	200	μs	LP Oscillator mode
2	Тсү	Instruction Cycle Time <sup>(1)</sup>	62.5	_	ns	Tcy = 4/Fosc
3	TosL,	External Clock in (OSC1)	30	_	ns	XT Oscillator mode
	TosH	High or Low Time	2.5	_	μs	LP Oscillator mode
			10	_	ns	HS Oscillator mode
4	TosR, TosF	External Clock in (OSC1) Rise or Fall Time	_	20	ns	XT Oscillator mode
			—	50	ns	LP Oscillator mode
			—	7.5	ns	HS Oscillator mode

**Note 1:** Instruction cycle period (TCY) equals four times the input oscillator time base period for all configurations except PLL. All specified values are based on characterization data for that particular oscillator type under standard operating conditions with the device executing code. Exceeding these specified limits may result in an unstable oscillator operation and/or higher than expected current consumption. All devices are tested to operate at "min." values with an external clock applied to the OSC1/CLKIN pin. When an external clock input is used, the "max." cycle time limit is "DC" (no clock) for all devices.

Param No.	Sym	Characteristic	Min	Тур†	Max	Units	Conditions
F10	Fosc	Oscillator Frequency Range	4		5	MHz	VDD = 1.8-5.5V
			4	—	16	MHz	VDD = 3.0-5.5V, -40°C to +85°C
			4	—	12	MHz	VDD = 3.0-5.5V, -40°C to +125°C
F11	Fsys	On-Chip VCO System Frequency	16	—	20	MHz	VDD = 1.8-5.5V
			16	—	64	MHz	VDD = 3.0-5.5V, -40°C to +85°C
			16	—	48	MHz	VDD = 3.0-5.5V, -40°C to +125°C
F12	t <sub>rc</sub>	PLL Start-up Time (Lock Time)	_		2	ms	
F13	$\Delta \text{CLK}$	CLKOUT Stability (Jitter)	-2		+2	%	

### TABLE 31-7: PLL CLOCK TIMING SPECIFICATIONS (VDD = 1.8V TO 5.5V)

† Data in "Typ" column is at 3V, 25°C unless otherwise stated. These parameters are for design guidance only and are not tested.

### TABLE 31-8: INTERNAL RC ACCURACY (INTOSC)

PIC18F87K22 Family		Standard Operating Conditions (unless otherwise stated)Operating temperature $-40^{\circ}C \le TA \le +125^{\circ}C$							
Param No.		Min	Тур	Max	Units	Conditions			
OA1	HFINTOSC/MFINTOSC Accuracy @ Freq = 16 MHz, 8 MHz, 4 MHz, 2 MHz, 1 MHz, 500 kHz, 250 kHz								
		-2	—	2	%	+25°C	VDD = 3.0-5.5V		
		-5	—	5	%				
		-5	—	5	%	-40°C to +85°C	VDD = 1.8-5.5V		
		-10	—	10	%	-40°C to +125°C	VDD = 1.8-5.5V		
OA2	2 LFINTOSC Accuracy @ Freq = 31 kHz								
		-15	—	15	%	-40°C to +125°C	VDD = 1.8-5.5V		

Note 1: Frequency calibrated at 25°C. OSCTUNE register can be used to compensate for temperature drift.

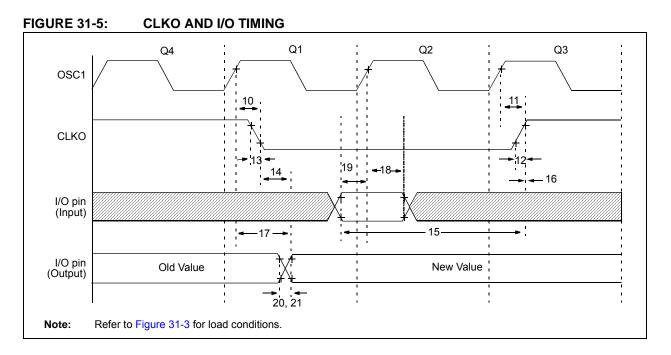


TABLE 31-9: CLKO AND I/O TIMING REQUIREMENTS
----------------------------------------------

Param No.	Symbol	Characteristic	Min	Тур	Max	Units	Conditions
10	TosH2ckL	OSC1 ↑ to CLKO ↓		75	200	ns	(Note 1)
11	TosH2ckH	OSC1 ↑ to CLKO ↑	—	75	200	ns	(Note 1)
12	TCKR	CLKO Rise Time	—	15	30	ns	(Note 1)
13	ТскF	CLKO Fall Time	—	15	30	ns	(Note 1)
14	TckL2IoV	CLKO $\downarrow$ to Port Out Valid			0.5 Tcy + 20	ns	
15	TioV2скH	Port In Valid before CLKO ↑	0.25 Tcy + 25			ns	
16	TckH2iol	Port In Hold after CLKO ↑	0			ns	
17	TosH2IoV	OSC1 ↑ (Q1 cycle) to Port Out Valid		50	150	ns	
18	TosH2ıol	OSC1 ↑ (Q2 cycle) to Port Input Invalid (I/O in hold time)	100	—		ns	
19	TioV2osH	Port Input Valid to OSC1 ↑ (I/O in setup time)	0	—		ns	
20	TIOR	Port Output Rise Time		10	25	ns	
21	TIOF	Port Output Fall Time		10	25	ns	
22†	Tinp	INTx pin High or Low Time	20		—	ns	
23†	Trbp	RB<7:4> Change INTx High or Low Time	Тсү		—	ns	

† These parameters are asynchronous events not related to any internal clock edges.

**Note 1:** Measurements are taken in EC mode, where CLKO output is 4 x Tosc.

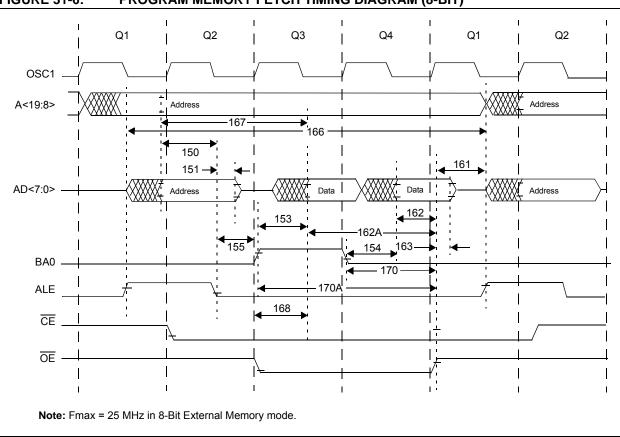
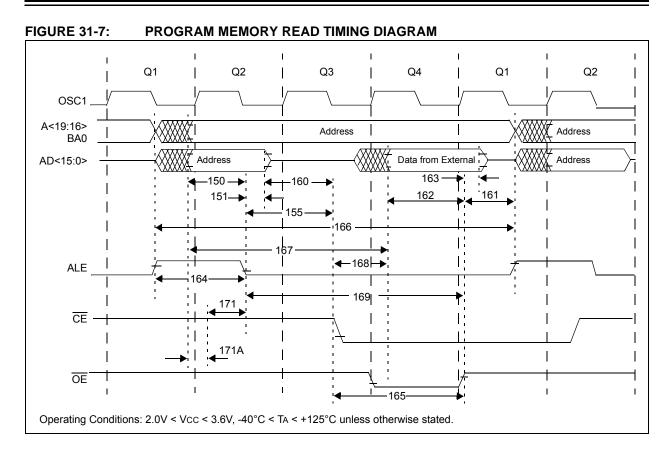


FIGURE 31-6:	PROGRAM MEMORY FETCH TIMING DIAGRAM (8-BIT)
	FROGRAM MEMORITETCH HMING DIAGRAM (0-DH)

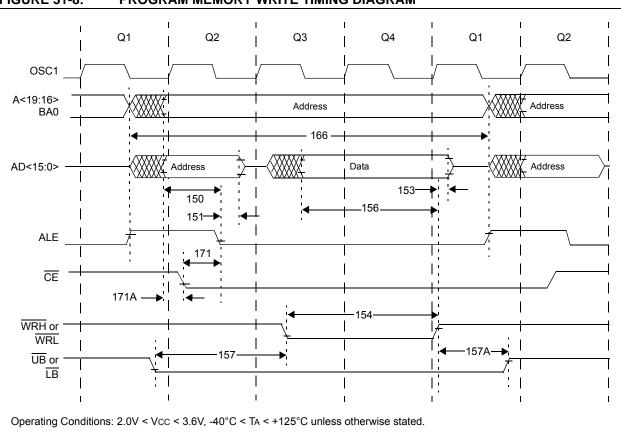
TABLE 31-10:	PROGRAM MEMORY FETCH TIMING REQUIREMENTS (8-BIT)
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Param No	Symbol	Characteristics	Min	Тур	Мах	Units
150	TadV2alL	Address Out Valid to ALE $\downarrow$ (address setup time)	0.25 Tcy – 10		—	ns
151	TalL2adl	ALE $\downarrow$ to Address Out Invalid (address hold time)	5		—	ns
153	BA01	BA0 $\uparrow$ to Most Significant Data Valid	0.125 TCY		—	ns
154	BA02	BA0 $\downarrow$ to Least Significant Data Valid	0.125 TCY	_	—	ns
155	TalL2oeL	ALE $\downarrow$ to $\overline{OE} \downarrow$	0.125 TCY		—	ns
161	ToeH2adD	OE ↑ to A/D Driven	0.125 Tcy – 5	_	—	ns
162	TadV2oeH	Least Significant Data Valid Before $\overline{OE}$ $\uparrow$ (data setup time)	20	_	—	ns
162A	TadV2oeH	Most Significant Data Valid Before OE ↑ (data setup time)	0.25 Tcy + 20	—	_	ns
163	ToeH2adI	OE ↑ to Data in Invalid (data hold time)	0		_	ns
166	TalH2alH	ALE $\uparrow$ to ALE $\uparrow$ (cycle time)	—	TCY	—	ns
167	TACC	Address Valid to Data Valid	0.5 Tcy – 10		—	ns
168	Тое	OE ↓ to Data Valid		_	0.125 Tcy + 5	ns
170	TubH2oeH	BA0 = 0 Valid Before OE ↑	0.25 TCY		—	ns
170A	TubL2oeH	BA0 = 1 Valid Before OE ↑	0.5 TCY		_	ns



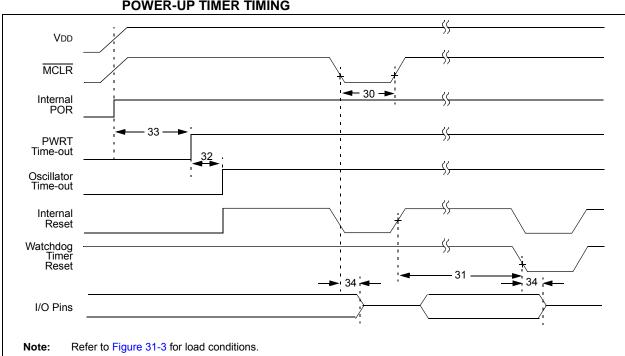
Param. No	Symbol	Characteristics	Min	Тур	Мах	Units
150	TadV2alL	Address Out Valid to ALE $\downarrow$ (address setup time)	0.25 TCY – 10	_	_	ns
151	TalL2adl	ALE $\downarrow$ to Address Out Invalid (address hold time)	5	_	—	ns
155	TalL2oeL	ALE $\downarrow$ to $\overline{OE} \downarrow$	10	0.125 TCY	—	ns
160	TadZ2oeL	AD High-Z to $\overline{OE} \downarrow$ (bus release to $\overline{OE}$ )	0	_	—	ns
161	ToeH2adD	OE ↑ to AD Driven	0.125 Tcy – 5	_	—	ns
162	TadV2oeH	LS Data Valid before $\overline{OE}$ $\uparrow$ (data setup time)	20	_	—	ns
163	ToeH2adl	OE ↑ to Data In Invalid (data hold time)	0	_	—	ns
164	TalH2alL	ALE Pulse Width	—	0.25 TCY	—	ns
165	ToeL2oeH	OE Pulse Width	0.5 Tcy – 5	0.5 TCY	_	ns
166	TalH2alH	ALE $\uparrow$ to ALE $\uparrow$ (cycle time)	—	Тсү	—	ns
167	Tacc	Address Valid to Data Valid	0.75 Tcy – 25		—	ns
168	Тое	$\overline{OE}\downarrow$ to Data Valid			0.5 TCY – 25	ns
169	TalL2oeH	ALE ↓ to OE ↑	0.625 Tcy – 10	_	0.625 Tcy + 10	ns
171	TalH2csL	Chip Enable Active to ALE $\downarrow$	0.25 Tcy – 20	_	—	ns
171A	TubL2oeH	AD Valid to Chip Enable Active			10	ns

TABLE 31-11:	<b>CLKO AND I/O TIMING REQUIREMENTS</b>



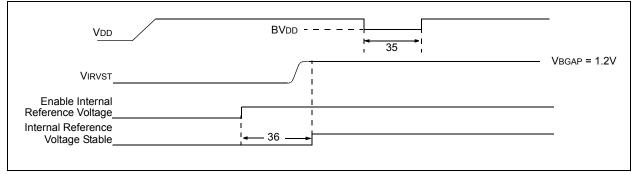
#### FIGURE 31-8: PROGRAM MEMORY WRITE TIMING DIAGRAM

Param. No	Symbol	Characteristics	Min	Тур	Max	Units
150	TadV2alL	Address Out Valid to ALE $\downarrow$ (address setup time)	0.25 Tcy – 10	—	_	ns
151	TalL2adl	ALE $\downarrow$ to Address Out Invalid (address hold time)	5	—		ns
153	TwrH2adl	$\overline{WRn}$ $\uparrow$ to Data Out Invalid (data hold time)	5	_	_	ns
154	TwrL	WRn Pulse Width	0.5 TCY – 5	0.5 TCY	_	ns
156	TadV2wrH	Data Valid before $\overline{WRn}$ $\uparrow$ (data setup time)	0.5 Tcy – 10	—	_	ns
157	TbsV2wrL	Byte Select Valid before WRn ↓ (byte select setup time)	0.25 TCY	—	_	ns
157A	TwrH2bsl	WRn ↑ to Byte Select Invalid (byte select hold time)	0.125 Tcy – 5	_	_	ns
166	TalH2alH	ALE $\uparrow$ to ALE $\uparrow$ (cycle time)	—	Тсү	_	ns
171	TalH2csL	Chip Enable Active to ALE $\downarrow$	0.25 TCY – 20	—	_	ns
171A	TubL2oeH	AD Valid to Chip Enable Active		_	10	ns



### FIGURE 31-9: RESET, WATCHDOG TIMER, OSCILLATOR START-UP TIMER AND POWER-UP TIMER TIMING

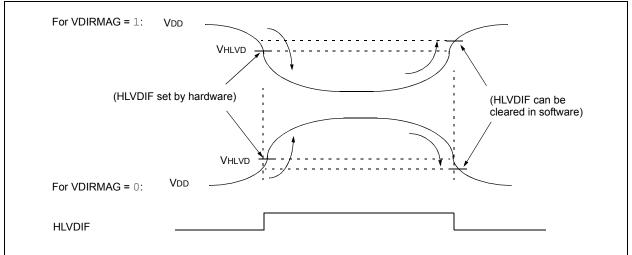
#### FIGURE 31-10: BROWN-OUT RESET TIMING



### TABLE 31-13: RESET, WATCHDOG TIMER, OSCILLATOR START-UP TIMER, POWER-UP TIMER AND BROWN-OUT RESET REQUIREMENTS

Param. No.	Symbol	Characteristic	Min	Тур	Мах	Units	Conditions
30	TmcL	MCLR Pulse Width (low)	2			μS	
31	Twdt	Watchdog Timer Time-out Period (no postscaler)	—	4.00	—	ms	
32	Tost	Oscillation Start-up Timer Period	1024 Tosc		1024 Tosc	_	Tosc = OSC1 period
33	TPWRT	Power-up Timer Period	—	65.5	140	ms	
34	Tioz	I/O High-Impedance from MCLR Low or Watchdog Timer Reset	_	2	—	μS	
35	TBOR	Brown-out Reset Pulse Width	200		_	μS	$VDD \le BVDD$ (see D005)
36	TIRVST	Time for Internal Reference Voltage to become Stable	—	25	—	μS	
37	Thlvd	High/Low-Voltage Detect Pulse Width	200	_		μS	$VDD \leq VHLVD$
38	TCSD	CPU Start-up Time	5	_	10	μS	
39	TIOBST	Time for INTOSC to Stabilize	_	1		μS	

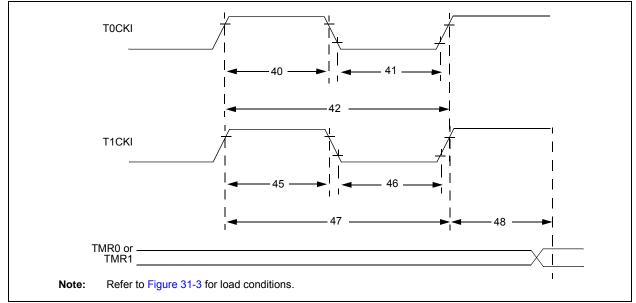




#### TABLE 31-14: HIGH/LOW-VOLTAGE DETECT CHARACTERISTICS

Standard Operating Conditions (unless otherwise stated)Operating temperature $-40^{\circ}C \le TA \le +85^{\circ}C$ for industrial $-40^{\circ}C \le TA \le +125^{\circ}C$ for extended								
Param No.	Sym	Charact	eristic	Min	Тур	Max	Units	Conditions
D420		HLVD Voltage on VDD		1.80	1.86	1.90	V	
		Transition High-to-Low	HLVDL<3:0> = 0001	2.03	2.12	2.13	V	
			HLVDL<3:0> = 0010	2.24	2.33	2.35	V	
			HLVDL<3:0> = 0011	2.40	2.49	2.53	V	
			HLVDL<3:0> = 0100	2.50	2.59	2.62	V	
			HLVDL<3:0> = 0101	2.70	2.75	2.84	V	
			HLVDL<3:0> = 0110	2.82	2.93	2.97	V	
			HLVDL<3:0> = 0111	2.95	3.07	3.10	V	
			HLVDL<3:0> = 1000	3.24	3.30	3.41	V	
			HLVDL<3:0> = 1001	3.42	3.48	3.59	V	
			HLVDL<3:0> = 1010	3.61	3.67	3.79	V	
			HLVDL<3:0> = 1011	3.82	3.87	4.01	V	
			HLVDL<3:0> = 1100	4.06	4.21	4.26	V	
			HLVDL<3:0> = 1101	4.33	4.42	4.55	V	
			HLVDL<3:0> = 1110	4.64	4.77	4.87	V	

#### FIGURE 31-12: TIMER0 AND TIMER1 EXTERNAL CLOCK TIMINGS



Param No.	Symbol		Characteristic	;	Min	Max	Units	Conditions
40	T⊤0H	T0CKI High F	ulse Width	ulse Width No prescaler		—	ns	
				With prescaler	10	_	ns	
41	T⊤0L	T0CKI Low P	ulse Width	No prescaler	0.5 Tcy + 20	—	ns	
				With prescaler	10	_	ns	
42	T⊤0P	T0CKI Period		No prescaler	Tcy + 10	_	ns	
				With prescaler	Greater of: 20 ns or (Tcy + 40)/N	_	ns	N = prescale value (1, 2, 4,, 256)
45	T⊤1H	T1CKI High	Synchronous, no prescaler		0.5 Tcy + 20	—	ns	
		Time	Synchronous, with prescaler		10	_	ns	
			Asynchronous		30	—	ns	
46	T⊤1L	T1CKI Low	Synchronous, n	o prescaler	0.5 Tcy + 5	_	ns	
		Time	Synchronous, with prescaler		10	_	ns	
			Asynchronous		30	_	ns	
47	T⊤1P	T1CKI Input Period	Synchronous		Greater of: 20 ns or (Tcy + 40)/N	_	ns	N = prescale value (1, 2, 4, 8)
			Asynchronous	Asynchronous		—	ns	
	F⊤1	T1CKI Oscilla	tor Input Freque	ncy Range	DC	50	kHz	
48	TCKE2TMRI	Delay from Ex Timer Increm	kternal T1CKI Clo ent	ock Edge to	2 Tosc	7 Tosc	—	

#### FIGURE 31-13: CAPTURE/COMPARE/PWM TIMINGS (ECCP1, ECCP2 MODULES)

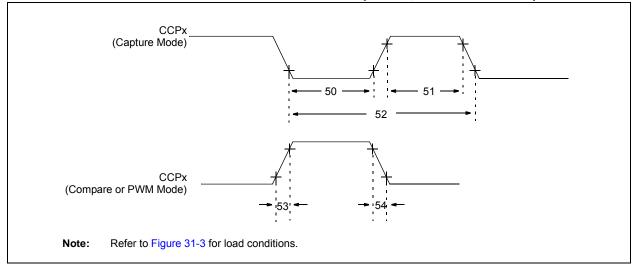
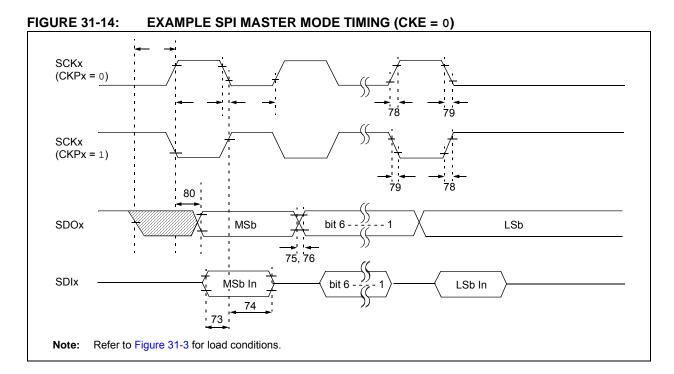
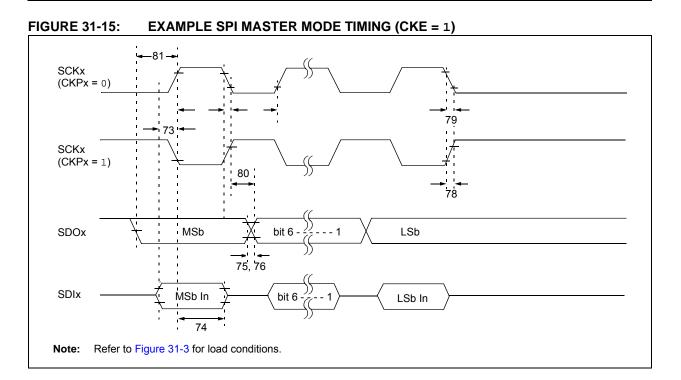


TABLE 31-16: CAPTURE/COMPARE/PWM REQUIREMENTS (ECCP1, ECCP2 MODULES)

Param No.	Symbol	Characteristic		Min	Max	Units	Conditions
50	TccL	CCPx Input Low	No prescaler	0.5 TCY + 20		ns	
	Time		With prescaler	10	_	ns	
51	ТссН	CCPx Input	No prescaler	0.5 TCY + 20	_	ns	
		High Time With prescaler		10	_	ns	
52	TCCP	CCPx Input Perio	CCPx Input Period		_	ns	N = prescale value (1, 4 or 16)
53	TCCR	CCPx Output Fal	CCPx Output Fall Time		25	ns	
54	TCCF	CCPx Output Fal	l Time	—	25	ns	

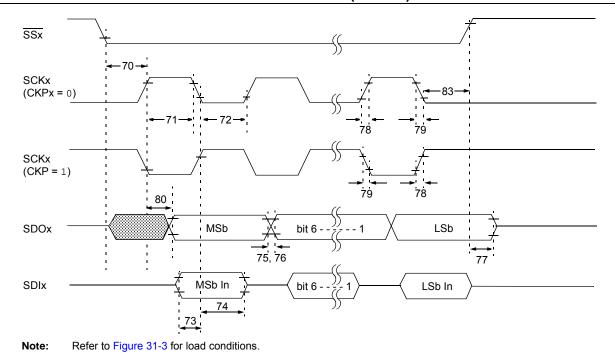


Param No.	Symbol	Characteristic	Min	Max	Units	Conditions
73	TDIV2scH, TDIV2scL	Setup Time of SDIx Data Input to SCKx Edge	20	-	ns	
73A	Тв2в	Last Clock Edge of Byte 1 to the 1st Clock Edge of Byte 2	1.5 Tcy + 40	_	ns	
74	TscH2dlL, TscL2dlL	Hold Time of SDIx Data Input to SCKx Edge	40		ns	
75	TDOR	SDOx Data Output Rise Time	—	25	ns	
76	TDOF	SDOx Data Output Fall Time	—	25	ns	
78	TscR	SCKx Output Rise Time (Master mode)	—	25	ns	
79	TscF	SCKx Output Fall Time (Master mode)	—	25	ns	
80	TscH2doV, TscL2doV	SDOx Data Output Valid after SCKx Edge	—	50	ns	



Param. No.	Symbol	Characteristic	Min	Max	Units	Conditions
73	TDIV2scH, TDIV2scL	Setup Time of SDIx Data Input to SCKx Edge	20	—	ns	
73A	Тв2в	Last Clock Edge of Byte 1 to the 1st Clock Edge of Byte 2	1.5 Tcy + 40	_	ns	
74	TscH2diL, TscL2diL	Hold Time of SDIx Data Input to SCKx Edge	40	_	ns	
75	TDOR	SDOx Data Output Rise Time	—	25	ns	
76	TDOF	SDOx Data Output Fall Time	—	25	ns	
78	TscR	SCKx Output Rise Time (Master mode)	—	25	ns	
79	TscF	SCKx Output Fall Time (Master mode)	—	25	ns	
80	TscH2doV, TscL2doV	SDOx Data Output Valid after SCKx Edge	—	50	ns	
81	TDOV2SCH, TDOV2SCL	SDOx Data Output Setup to SCKx Edge	Тсү	_	ns	

#### TABLE 31-18: EXAMPLE SPI MODE REQUIREMENTS (MASTER MODE, CKE = 1)



#### FIGURE 31-16: EXAMPLE SPI SLAVE MODE TIMING (CKE = 0)

Param No.	Symbol	Characteristic	Characteristic		Max	Units	Conditions
70	TssL2scH, TssL2scL	$\overline{\mathrm{SSx}}\downarrow$ to SCKx $\downarrow$ or SCKx $\uparrow$ Input	r SCKx ↑ Input		_	ns	
70A	TssL2WB	SSx to Write to SSPBUF		3 TCY		ns	
71	TscH	SCKx Input High Time	Continuous	1.25 Tcy + 30		ns	
71A		(Slave mode)	Single Byte	40	_	ns	(Note 1)
72	TscL	SCKx Input Low Time	Continuous	1.25 Tcy + 30	_	ns	
72A		(Slave mode)	Single Byte	40	_	ns	(Note 1)
73	TDIV2scH, TDIV2scL	Setup Time of SDIx Data Input to SCKx Edge		20	_	ns	
73A	Тв2в	Last Clock Edge of Byte 1 to the First Clo	1.5 Tcy + 40		ns	(Note 2)	
74	TscH2DIL, TscL2DIL	Hold Time of SDIx Data Input to SCKx	Edge	40	_	ns	
75	TDOR	SDOx Data Output Rise Time		—	25	ns	
76	TdoF	SDOx Data Output Fall Time		—	25	ns	
77	TssH2doZ	SSx ↑ to SDOx Output High-Impedance	е	10	50	ns	
78	TscR	SCKx Output Rise Time (Master mode)	—	25	ns		
79	TscF	SCKx Output Fall Time (Master mode)	—	25	ns		
80	TscH2doV, TscL2doV	SDOx Data Output Valid after SCKx Edge		—	50	ns	
83	TscH2ssH, TscL2ssH	SSx ↑ after SCKx Edge		1.5 Tcy + 40		ns	

#### TABLE 31-19: EXAMPLE SPI MODE REQUIREMENTS (SLAVE MODE TIMING, CKE = 0)

**Note 1:** Requires the use of Parameter #73A.

2: Only if Parameter #71A and #72A are used.

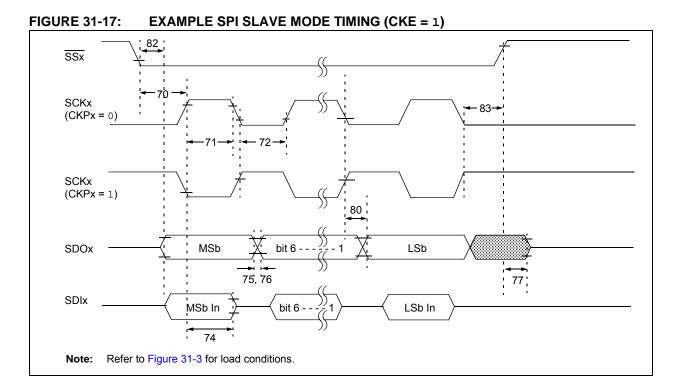


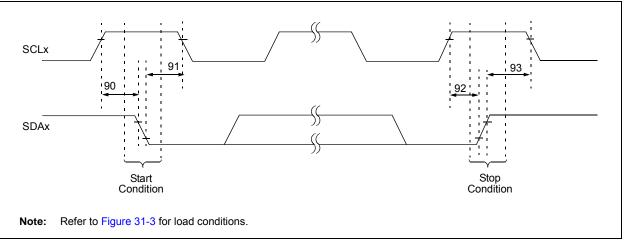
TABLE 31-20: EXAMPLE SPI SLAVE MODE REQUIREMENTS (CKE = 1)
------------------------------------------------------------

Param No.	Symbol	Characteristic	Min	Max	Units	Conditions	
70	TssL2scH, TssL2scL	$\overline{\mathrm{SSx}}\downarrow$ to SCKx $\downarrow$ or SCKx $\uparrow$ Input	3 Тсү	—	ns		
70A	TssL2WB	SSx to Write to SSPBUF		3 Tcy	_	ns	
71	TscH	SCKx Input High Time	Continuous	1.25 Tcy + 30	—	ns	
71A		(Slave mode)	Single Byte	40	_	ns	(Note 1)
72	TscL	SCKx Input Low Time	Continuous	1.25 Tcy + 30	—	ns	
72A		(Slave mode)	ave mode) Single Byte		—	ns	(Note 1)
73A	Тв2в	Last Clock Edge of Byte 1 to the First	1.5 Tcy + 40		ns	(Note 2)	
74	TscH2diL, TscL2diL	Hold Time of SDIx Data Input to SC	40	—	ns		
75	TDOR	SDOx Data Output Rise Time		—	25	ns	
76	TDOF	SDOx Data Output Fall Time		—	25	ns	
77	TssH2doZ	SSx ↑ to SDOx Output High-Impeda	ince	10	50	ns	
78	TscR	SCKx Output Rise Time (Master mo	de)	—	25	ns	
79	TscF	SCKx Output Fall Time (Master mod		25	ns		
80	TscH2doV, TscL2doV	SDOx Data Output Valid after SCKx	—	50	ns		
82	TssL2DoV	SDOx Data Output Valid after $\overline{\text{SSx}}$ $\downarrow$	Edge	_	50	ns	
83	TscH2ssH, TscL2ssH	SSx ↑ after SCKx Edge		1.5 Tcy + 40	_	ns	

**Note 1:** Requires the use of Parameter #73A.

2: Only if Parameter #71A and #72A are used.

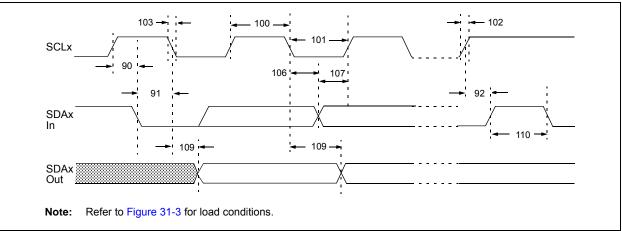




### TABLE 31-21: I<sup>2</sup>C<sup>™</sup> BUS START/STOP BITS REQUIREMENTS (SLAVE MODE)

Param. No.	Symbol	Characte	ristic	Min	Мах	Units	Conditions	
90	TSU:STA	Start Condition	100 kHz mode	4700		ns	Only relevant for Repeated	
		Setup Time	400 kHz mode	600	—		Start condition	
91	THD:STA	Start Condition	100 kHz mode	4000	—	ns	After this period, the first	
		Hold Time	400 kHz mode	600	_		clock pulse is generated	
92	Tsu:sto	Stop Condition	100 kHz mode	4700	_	ns		
		Setup Time	400 kHz mode	600	_			
93	THD:STO	Stop Condition	100 kHz mode	4000	_	ns		
		Hold Time	400 kHz mode	600				





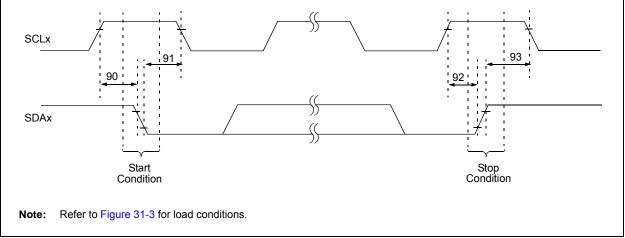
Param. No.	Symbol	Characteris	tic	Min	Max	Units	Conditions
100	Тнідн	Clock High Time	100 kHz mode	4.0	_	μS	
			400 kHz mode	0.6	—	μS	
			MSSP module	1.5 TCY	—		
101	TLOW	Clock Low Time	100 kHz mode	4.7	—	μS	
			400 kHz mode	1.3	—	μS	
			MSSP module	1.5 TCY	—		
102	TR	SDAx and SCLx Rise Time	100 kHz mode	—	1000	ns	
			400 kHz mode	20 + 0.1 CB	300	ns	CB is specified to be from 10 to 400 pF
103	TF	SDAx and SCLx Fall Time	100 kHz mode	_	300	ns	
			400 kHz mode	20 + 0.1 CB	300	ns	CB is specified to be from 10 to 400 pF
90	TSU:STA	Start Condition Setup Time	100 kHz mode	4.7	—	μS	Only relevant for Repeated
			400 kHz mode	0.6	—	μS	Start condition
91	THD:STA	Start Condition Hold Time	100 kHz mode	4.0	—	μS	After this period, the first clock
			400 kHz mode	0.6	—	μS	pulse is generated
106	THD:DAT	Data Input Hold Time	100 kHz mode	0	—	ns	
			400 kHz mode	0	0.9	μS	
107	TSU:DAT	Data Input Setup Time	100 kHz mode	250	—	ns	(Note 2)
			400 kHz mode	100	—	ns	
92	Tsu:sto	Stop Condition Setup Time	100 kHz mode	4.7	—	μS	
			400 kHz mode	0.6	—	μS	
109	ΤΑΑ	Output Valid from Clock	100 kHz mode	—	3500	ns	(Note 1)
			400 kHz mode	—	_	ns	
110	TBUF	Bus Free Time	100 kHz mode	4.7	—	μS	Time the bus must be free before
			400 kHz mode	1.3	—	μS	a new transmission can start
D102	Св	Bus Capacitive Loading		—	400	pF	

#### TABLE 31-22: I<sup>2</sup>C<sup>™</sup> BUS DATA REQUIREMENTS (SLAVE MODE)

Note 1: As a transmitter, the device must provide this internal minimum delay time to bridge the undefined region (min. 300 ns) of the falling edge of SCLx to avoid unintended generation of Start or Stop conditions.

2: A Fast mode I<sup>2</sup>C<sup>™</sup> bus device can be used in a Standard mode I<sup>2</sup>C bus system, but the requirement, TSU:DAT ≥ 250 ns, must then be met. This will automatically be the case if the device does not stretch the LOW period of the SCLx signal. If such a device does stretch the LOW period of the SCLx signal, it must output the next data bit to the SDAx line, TR max. + TSU:DAT = 1000 + 250 = 1250 ns (according to the Standard mode I<sup>2</sup>C bus specification), before the SCLx line is released.

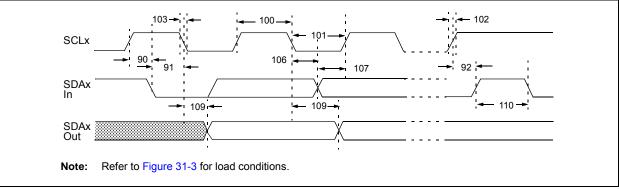




Param. No.	Symbol	Characte	eristic	Min	Max	Units	Conditions
90	TSU:STA	Start Condition	100 kHz mode	2(Tosc)(BRG + 1)	_	ns	Only relevant for
		Setup Time	400 kHz mode	2(Tosc)(BRG + 1)	_		Repeated Start
			1 MHz mode <sup>(1)</sup>	2(Tosc)(BRG + 1)	_		condition
91	THD:STA	Start Condition	100 kHz mode	2(Tosc)(BRG + 1)	_	ns	After this period, the
		Hold Time	400 kHz mode	2(Tosc)(BRG + 1)	_		first clock pulse is
			1 MHz mode <sup>(1)</sup>	2(Tosc)(BRG + 1)	_		generated
92	Tsu:sto	Stop Condition	100 kHz mode	2(Tosc)(BRG + 1)	_	ns	
		Setup Time	400 kHz mode	2(Tosc)(BRG + 1)	_		
			1 MHz mode <sup>(1)</sup>	2(Tosc)(BRG + 1)	_		
93	THD:STO	Stop Condition	100 kHz mode	2(Tosc)(BRG + 1)	_	ns	
		Hold Time	400 kHz mode	2(Tosc)(BRG + 1)	_	1	
			1 MHz mode <sup>(1)</sup>	2(Tosc)(BRG + 1)	_	]	

**Note 1:** Maximum pin capacitance = 10 pF for all  $I^2C^{TM}$  pins.



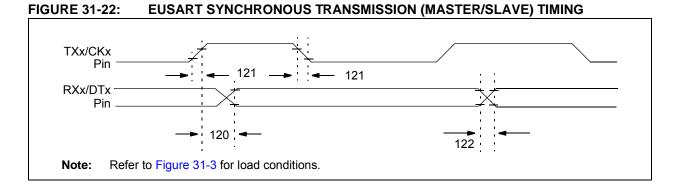


Param. No.	Symbol	Charac	teristic	Min	Max	Units	Conditions
100	Тнідн	Clock High	100 kHz mode	2(Tosc)(BRG + 1)		_	
		Time	400 kHz mode	2(Tosc)(BRG + 1)		_	
			1 MHz mode <sup>(1)</sup>	2(Tosc)(BRG + 1)		_	
101	TLOW	Clock Low Time	100 kHz mode	2(Tosc)(BRG + 1)	—		
			400 kHz mode	2(Tosc)(BRG + 1)		_	
			1 MHz mode <sup>(1)</sup>	2(Tosc)(BRG + 1)		_	
102	Tr	SDAx and	100 kHz mode	—	1000	ns	CB is specified to be from
		SCLx Rise	400 kHz mode	20 + 0.1 Св	300	ns	10 to 400 pF
		Time	1 MHz mode <sup>(1)</sup>		300	ns	
103	TF	SDAx and	100 kHz mode	—	300	ns	CB is specified to be from
		SCLx Fall Time	400 kHz mode	20 + 0.1 Св	300	ns	10 to 400 pF
			1 MHz mode <sup>(1)</sup>	_	100	ns	
90	TSU:STA	Start Condition	100 kHz mode	2(Tosc)(BRG + 1)			Only relevant for Repeated
		Setup Time	400 kHz mode	2(Tosc)(BRG + 1)		_	Start condition
			1 MHz mode <sup>(1)</sup>	2(Tosc)(BRG + 1)		_	
91	THD:STA	A Start Condition Hold Time	100 kHz mode	2(Tosc)(BRG + 1)			After this period, the first
			400 kHz mode	2(Tosc)(BRG + 1)		_	clock pulse is generated
			1 MHz mode <sup>(1)</sup>	2(Tosc)(BRG + 1)		_	
106	THD:DAT	Data Input	100 kHz mode	0		ns	
		Hold Time	400 kHz mode	0	0.9	μs	
			1 MHz mode <sup>(1)</sup>	_		ns	
107	TSU:DAT	Data Input	100 kHz mode	250		ns	(Note 2)
		Setup Time	400 kHz mode	100		ns	
			1 MHz mode <sup>(1)</sup>	_		ns	
92	Tsu:sto	Stop Condition	100 kHz mode	2(Tosc)(BRG + 1)			
		Setup Time	400 kHz mode	2(Tosc)(BRG + 1)		_	
			1 MHz mode <sup>(1)</sup>	2(Tosc)(BRG + 1)			
109	ΤΑΑ	Output Valid	100 kHz mode	—	3500	ns	
		from Clock	400 kHz mode	_	1000	ns	
			1 MHz mode <sup>(1)</sup>	_		ns	
110	TBUF	Bus Free Time	100 kHz mode	4.7	—	μS	Time the bus must be free
			400 kHz mode	1.3	—	μS	before a new transmission
			1 MHz mode <sup>(1)</sup>	—	—	μS	can start
D102	Св	Bus Capacitive L	oading	_	400	pF	

TABLE 31-24: M	ISSP I <sup>2</sup> C™ BUS DATA	REQUIREMENTS
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**Note 1:** Maximum pin capacitance = 10 pF for all  $I^2C^{TM}$  pins.

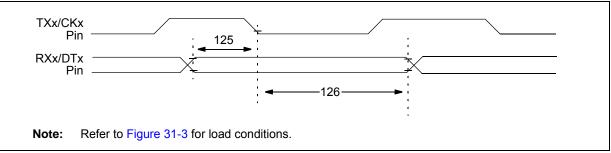
2: A Fast mode I<sup>2</sup>C bus device can be used in a Standard mode I<sup>2</sup>C bus system, but Parameter #107 ≥ 250 ns must then be met. This will automatically be the case if the device does not stretch the LOW period of the SCLx signal. If such a device does stretch the LOW period of the SCLx signal, it must output the next data bit to the SDAx line, Parameter #102 + Parameter #107 = 1000 + 250 = 1250 ns (for 100 kHz mode), before the SCLx line is released.



#### TABLE 31-25: EUSART/AUSART SYNCHRONOUS TRANSMISSION REQUIREMENTS

Param No.	Symbol	Characteristic		Max	Units	Conditions
120	TCKH2DTV	<u>SYNC XMIT (MASTER and SLAVE)</u> Clock High to Data Out Valid	_	40	ns	
121	TCKRF	Clock Out Rise Time and Fall Time (Master mode)	—	20	ns	
122	TDTRF	Data Out Rise Time and Fall Time		20	ns	

#### FIGURE 31-23: EUSART/AUSART SYNCHRONOUS RECEIVE (MASTER/SLAVE) TIMING



#### TABLE 31-26: EUSART/AUSART SYNCHRONOUS RECEIVE REQUIREMENTS

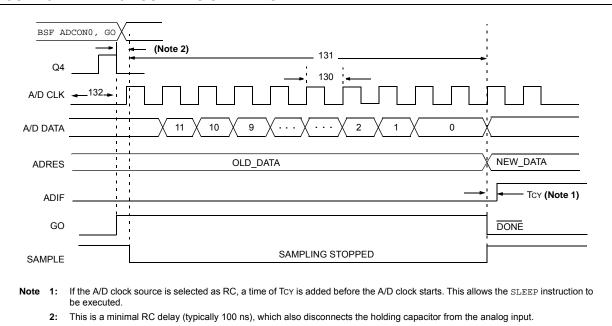
Param. No.	Symbol	Characteristic	Min	Max	Units	Conditions
125	TDTV2CKL	SYNC RCV (MASTER and SLAVE)				
		Data Hold before CKx $\downarrow$ (DTx hold time)	10	—	ns	
126	TCKL2DTL	Data Hold after CKx $\downarrow$ (DTx hold time)	15	_	ns	

Param No.	Sym	Characteristic	Min	Тур	Max	Units	Conditions
A01	NR	Resolution	—	—	12	bit	$\Delta VREF \ge 5.0V$
A03	EIL	Integral Linearity Error	—	±1	±6.0	LSB	ΔVREF = 5.0V
A04	Edl	Differential Linearity Error	_	±1	+3.0/-1.0	LSB	ΔVREF = 5.0V
A06	EOFF	Offset Error	—	±1	±9.0	LSB	ΔVREF = 5.0V
A07	Egn	Gain Error	_	±1	±8.0	LSB	ΔVREF = 5.0V
A10	_	Monotonicity <sup>(1)</sup>	_	_		_	$VSS \le VAIN \le VREF$
A20	$\Delta VREF$	Reference Voltage Range (VREFH – VREFL)	3	—	Vdd - Vss	V	
A21	Vrefh	Reference Voltage High	Vss + 3.0V	—	VDD + 0.3V	V	
A22	Vrefl	Reference Voltage Low	Vss – 0.3V	_	VDD - 3.0V	V	
A25	VAIN	Analog Input Voltage	VREFL	_	VREFH	V	
A30	Zain	Recommended Impedance of Analog Voltage Source	_	_	2.5	kΩ	
A50	IREF	VREF Input Current <sup>(2)</sup>	—	_	5 150	μΑ μΑ	During VAIN acquisition During A/D conversion cycle

#### TABLE 31-27: A/D CONVERTER CHARACTERISTICS: PIC18F87K22 FAMILY (INDUSTRIAL)

Note 1: The A/D conversion result doesn't decrease with an increase in the input voltage.

2: VREFH current is from the RA3/AN3/VREF+ pin or VDD, whichever is selected as the VREFH source. VREFL current is from the RA2/AN2/VREF- pin or VSs, whichever is selected as the VREFL source.



#### FIGURE 31-24: A/D CONVERSION TIMING

Param No.	Symbol	Characteristic	Min	Мах	Units	Conditions
130	Tad	A/D Clock Period	0.8	12.5 <sup>(1)</sup>	μs	Tosc-based, VREF $\geq$ 3.0V
			1.4	25 <sup>(1)</sup>	μs	VDD = 3.0V; TOSC-based, VREF full range
			_	1	μs	A/D RC mode
			_	3	μs	VDD = 3.0V; A/D RC mode
131	TCNV	Conversion Time (not including acquisition time) <sup>(2)</sup>	14	15	Tad	
132	TACQ	Acquisition Time <sup>(3)</sup>	1.4		μs	-40°C to +125°C
135	Tswc	Switching Time from Convert $\rightarrow$ Sample	_	(Note 4)		
137	TDIS	Discharge Time	0.2	_	μs	-40°C to +125°C

#### TABLE 31-28: A/D CONVERSION REQUIREMENTS

Note 1: The time of the A/D clock period is dependent on the device frequency and the TAD clock divider.

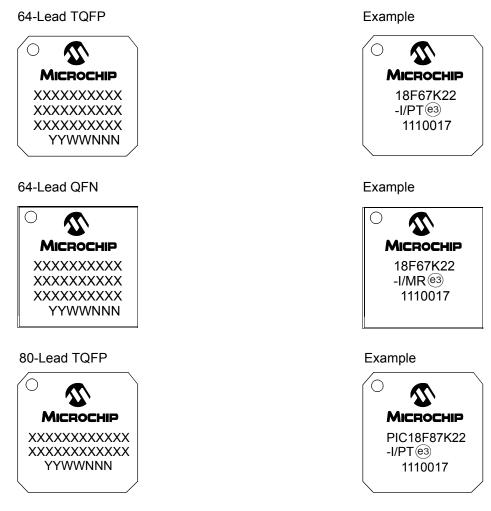
2: ADRES registers may be read on the following TCY cycle.

**3:** The time for the holding capacitor to acquire the "New" input voltage when the voltage changes full scale after the conversion (VDD to Vss or Vss to VDD). The source impedance (Rs) on the input channels is 50 $\Omega$ .

4: On the following cycle of the device clock.

### 32.0 PACKAGING INFORMATION

#### 32.1 Package Marking Information



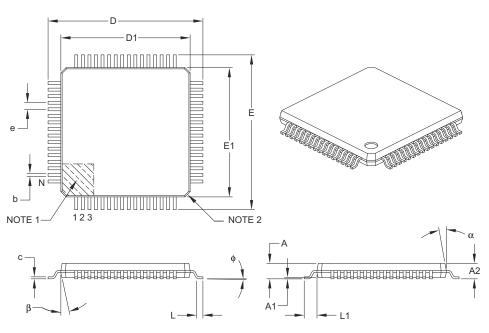
Legend	: XXX Y YY WW NNN @3 *	Customer-specific information Year code (last digit of calendar year) Year code (last 2 digits of calendar year) Week code (week of January 1 is week '01') Alphanumeric traceability code Pb-free JEDEC designator for Matte Tin (Sn) This package is Pb-free. The Pb-free JEDEC designator (e3) can be found on the outer packaging for this package.
Note:	be carried	nt the full Microchip part number cannot be marked on one line, it will d over to the next line, thus limiting the number of available s for customer-specific information.

#### 32.2 Package Details

The following sections give the technical details of the packages.

### 64-Lead Plastic Thin Quad Flatpack (PT) – 10x10x1 mm Body, 2.00 mm [TQFP]

**Note:** For the most current package drawings, please see the Microchip Packaging Specification located at http://www.microchip.com/packaging



	Units	MILLIMETERS		
Dimens	ion Limits	MIN	NOM	MAX
Number of Leads	Ν		64	
Lead Pitch	е		0.50 BSC	
Overall Height	А	-	-	1.20
Molded Package Thickness	A2	0.95	1.00	1.05
Standoff	A1	0.05	-	0.15
Foot Length	L	0.45	0.60	0.75
Footprint	L1		1.00 REF	
Foot Angle	φ	0°	3.5°	7°
Overall Width	E		12.00 BSC	
Overall Length	D		12.00 BSC	
Molded Package Width	E1		10.00 BSC	
Molded Package Length	D1		10.00 BSC	
Lead Thickness	С	0.09	-	0.20
Lead Width	b	0.17	0.22	0.27
Mold Draft Angle Top	α	11°	12°	13°
Mold Draft Angle Bottom	β	11°	12°	13°

#### Notes:

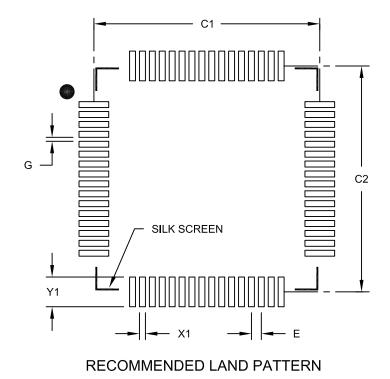
- 1. Pin 1 visual index feature may vary, but must be located within the hatched area.
- 2. Chamfers at corners are optional; size may vary.
- 3. Dimensions D1 and E1 do not include mold flash or protrusions. Mold flash or protrusions shall not exceed 0.25 mm per side.
- 4. Dimensioning and tolerancing per ASME Y14.5M.
  - BSC: Basic Dimension. Theoretically exact value shown without tolerances.

REF: Reference Dimension, usually without tolerance, for information purposes only.

Microchip Technology Drawing C04-085B

64-Lead Plastic Thin Quad Flatpack (PT) 10x10x1 mm Body, 2.00 mm Footprint [TQFP]

**Note:** For the most current package drawings, please see the Microchip Packaging Specification located at http://www.microchip.com/packaging



	Units			S
Dimensio	n Limits	MIN	NOM	MAX
Contact Pitch	E	0.50 BSC		
Contact Pad Spacing	C1		11.40	
Contact Pad Spacing	C2		11.40	
Contact Pad Width (X64)	X1			0.30
Contact Pad Length (X64)	Y1			1.50
Distance Between Pads	G	0.20		

Notes:

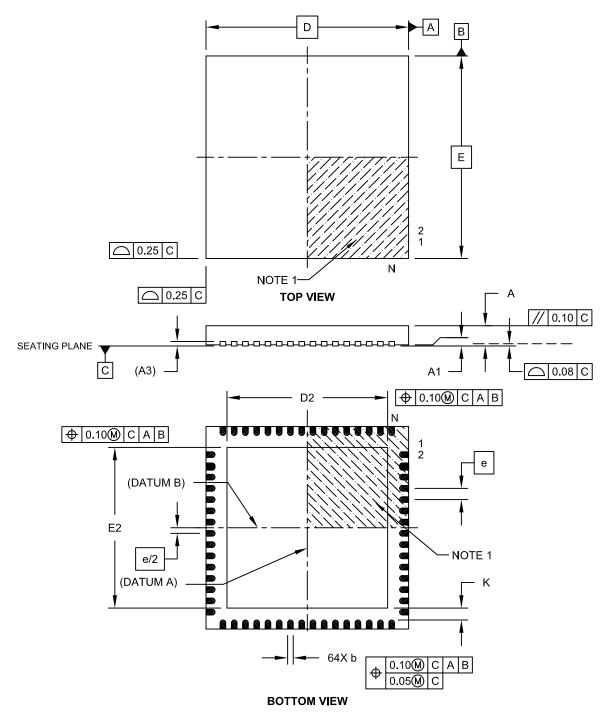
1. Dimensioning and tolerancing per ASME Y14.5M

BSC: Basic Dimension. Theoretically exact value shown without tolerances.

Microchip Technology Drawing No. C04-2085B

### 64-Lead Plastic Quad Flat, No Lead Package (MR) – 9x9x0.9 mm Body [QFN]

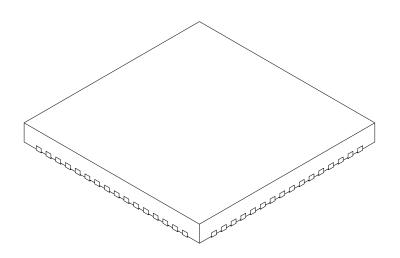
**Note:** For the most current package drawings, please see the Microchip Packaging Specification located at http://www.microchip.com/packaging



Microchip Technology Drawing C04-149B Sheet 1 of 2

#### 64-Lead Plastic Quad Flat, No Lead Package (MR) – 9x9x0.9 mm Body [QFN]

**Note:** For the most current package drawings, please see the Microchip Packaging Specification located at http://www.microchip.com/packaging



	Units			S
Dimensi	MIN	NOM	MAX	
Number of Pins	N		64	
Pitch	е		0.50 BSC	
Overall Height	A	0.80	0.90	1.00
Standoff	A1	0.00	0.02	0.05
Contact Thickness	A3	0.20 REF		
Overall Width	E		9.00 BSC	
Exposed Pad Width	E2	7.05	7.15	7.50
Overall Length	D		9.00 BSC	
Exposed Pad Length	D2	7.05	7.15	7.50
Contact Width	b	0.18	0.25	0.30
Contact Length	L	0.30	0.40	0.50
Contact-to-Exposed Pad	K	0.20	-	-

Notes:

1. Pin 1 visual index feature may vary, but must be located within the hatched area.

2. Package is saw singulated

3. Dimensioning and tolerancing per ASME Y14.5M.

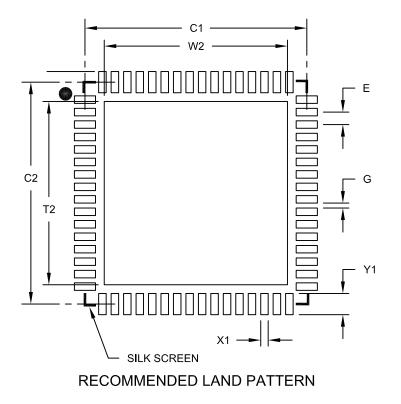
BSC: Basic Dimension. Theoretically exact value shown without tolerances.

REF: Reference Dimension, usually without tolerance, for information purposes only.

Microchip Technology Drawing C04-149B Sheet 2 of 2

### 64-Lead Plastic Quad Flat, No Lead Package (MR) – 9x9x0.9 mm Body [QFN] With 0.40 mm Contact Length

**Note:** For the most current package drawings, please see the Microchip Packaging Specification located at http://www.microchip.com/packaging



	MILLIMETERS				
Dimensior	n Limits	MIN	NOM	MAX	
Contact Pitch	Е	0.50 BSC			
Optional Center Pad Width	W2			7.35	
Optional Center Pad Length	T2			7.35	
Contact Pad Spacing	C1		8.90		
Contact Pad Spacing	C2		8.90		
Contact Pad Width (X64)	X1			0.30	
Contact Pad Length (X64)	Y1			0.85	
Distance Between Pads	G	0.20			

#### Notes:

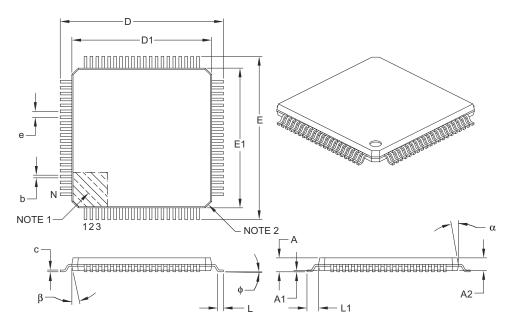
1. Dimensioning and tolerancing per ASME Y14.5M

BSC: Basic Dimension. Theoretically exact value shown without tolerances.

Microchip Technology Drawing No. C04-2149A

#### 80-Lead Plastic Thin Quad Flatpack (PT) – 12x12x1 mm Body, 2.00 mm [TQFP]

**Note:** For the most current package drawings, please see the Microchip Packaging Specification located at http://www.microchip.com/packaging



	Units	MILLIMETERS			
	Dimension Limits	MIN	NOM	MAX	
Number of Leads	N		80		
Lead Pitch	e		0.50 BSC		
Overall Height	A	-	-	1.20	
Molded Package Thickness	A2	0.95	1.00	1.05	
Standoff	A1	0.05	-	0.15	
Foot Length	L	0.45	0.60	0.75	
Footprint	L1	1.00 REF			
Foot Angle	ф	0°	3.5°	7°	
Overall Width	E		14.00 BSC		
Overall Length	D		14.00 BSC		
Molded Package Width	E1		12.00 BSC		
Molded Package Length	D1		12.00 BSC		
Lead Thickness	С	0.09	-	0.20	
Lead Width	b	0.17	0.22	0.27	
Mold Draft Angle Top	α	11°	12°	13°	
Mold Draft Angle Bottom	β	11°	12°	13°	

#### Notes:

1. Pin 1 visual index feature may vary, but must be located within the hatched area.

2. Chamfers at corners are optional; size may vary.

3. Dimensions D1 and E1 do not include mold flash or protrusions. Mold flash or protrusions shall not exceed 0.25 mm per side.

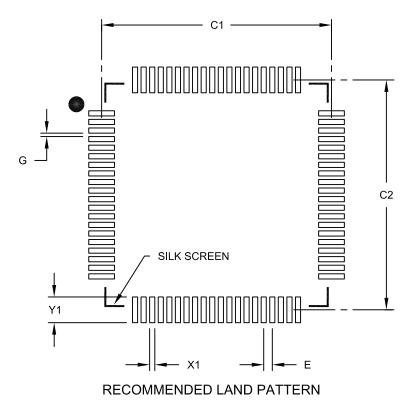
- 4. Dimensioning and tolerancing per ASME Y14.5M.
  - BSC: Basic Dimension. Theoretically exact value shown without tolerances.

REF: Reference Dimension, usually without tolerance, for information purposes only.

Microchip Technology Drawing C04-092B

80-Lead Plastic Thin Quad Flatpack (PT)-12x12x1mm Body, 2.00 mm Footprint [TQFP]

**Note:** For the most current package drawings, please see the Microchip Packaging Specification located at http://www.microchip.com/packaging



Units		l	MILLIMETER	S
Dimensi	on Limits	MIN	NOM	MAX
Contact Pitch	E		0.50 BSC	
Contact Pad Spacing	C1		13.40	
Contact Pad Spacing	C2		13.40	
Contact Pad Width (X80)	X1			0.30
Contact Pad Length (X80)	Y1			1.50
Distance Between Pads	G	0.20		

Notes:

1. Dimensioning and tolerancing per ASME Y14.5M

BSC: Basic Dimension. Theoretically exact value shown without tolerances.

Microchip Technology Drawing No. C04-2092B

### APPENDIX A: REVISION HISTORY

#### **Revision A (November 2009)**

Original data sheet for PIC18F87K22 family devices.

#### Revision B (May 2010)

Minor edits to text throughout document. Replaced all TBDs with the correct value.

#### **Revision C (March 2011)**

Section 2.4 "Voltage Regulator Pins (ENVREG and VCAP/VDDCORE)" has been replaced with a new and more detailed description and the 80-pin TQFP diagrams have been updated. Minor text edits throughout document.

#### Revision D (June 2011)

Updated Section 31.2 "DC Characteristics: Power-Down and Supply Current PIC18F87K22 Family (Industrial/Extended)" and Table 31-8 with new electrical specification information. The extended temperature information has been included in this revision.

#### APPENDIX B: MIGRATION FROM PIC18F87J11 AND PIC18F8722 TO PIC18F87K22

Devices in the PIC18F87K22, PIC18F87J11 and PIC18F8722 families are similar in their functions and features. Code can be migrated from the other families to the PIC18F87K22 without many changes. The differences between the device families are listed in Table B-1.

### TABLE B-1:NOTABLE DIFFERENCES BETWEEN PIC18F87K22, PIC18F87J11 AND PIC18F8722<br/>FAMILIES

Characteristic	PIC18F87K22 Family	18F87J11 Family	PIC18F8722 Family
Max Operating Frequency	64 MHz	48 MHz	40 MHz
Max Program Memory	128 Kbytes	128 Kbytes	128 Kbytes
Data Memory	3,862 bytes	3,930 bytes	3,930 bytes
Program Memory Endurance	10,000 Write/Erase (minimum)	10,000 Write/Erase (minimum)	10,000 Write/Erase (minimum)
Single Word Write for Flash	No	Yes	No
Oscillator Options	PLL can be used with INTOSC	PLL can be used with INTOSC	PLL can be used with INTOSC
СТМИ	Yes	No	No
RTCC	Yes	No	No
SOSC Oscillator Options TICKI Clock	Low-Power Oscillator Option for SOSC T1CKI can be used as a Clock without Enabling the SOSC Oscillator	No	No
INTOSC	Up to 16 MHz	Up to 8 MHz	Up to 8 MHz
SPI/I <sup>2</sup> C™	2	2	2
Timers	11	5	5
ECCP	3	3	
ССР	7	2	2
Data EEPROM	Yes	No	Yes
Programmable BOR	Multiple Level BOR	One Level BOR	Multiple Level BOR
WDT Prescale Options	22	16	16
5V Operation	Yes	No (3.3V)	Yes
nanoWatt XLP	Yes	No	No
Regulator	Yes	Yes	No
Low-Power BOR	Yes	No	No
A/D Converter	12-Bit Resolution, 24 Input Channels, Differential	10-Bit Resolution, 15 Input Channels, Non-Differential	10-Bit Resolution, 16 Input Channels, Non-Differential
Internal Temperature Sensor	Yes	No	No
Programmable HLVD	Yes	No	Yes
EUSART	2 EUSARTs	2 EUSARTs	2 EUSARTs
Comparators	3	2	2
Oscillator options	14 Options by Fosc<3:0>	8 Options by Fosc<3:0>	12 Options by Fosc<3:0>
Ultra-Low-Power Wake-up (ULPW)	Yes	No	No
Power-up Timer	Yes	Yes	Yes
MCLR Pin as Input Port	Yes	No	Yes

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4.	What additions to the document do you think would enha	nce the structure and subject?
5.	What deletions from the document could be made withou	t affecting the overall usefulness?
6.	Is there any incorrect or misleading information (what and	where)?
7.	How would you improve this document?	

### **PRODUCT IDENTIFICATION SYSTEM**

To order or obtain information, e.g., on pricing or delivery, refer to the factory or the listed sales office.

PART NO.	X /XX XXX T Temperature Package Pattern Range	<ul> <li>Examples:</li> <li>a) PIC18F87K22-I/PT 301 = Industrial temperature, TQFP package, QTP pattern #301.</li> <li>b) PIC18F87K22T-I/PT = Tape and reel, Industrial temperature, TQFP package</li> </ul>
Device <sup>(1,2)</sup>	PIC18F65K22, PIC18F65K22T PIC18F66K22, PIC18F66K22T PIC18F67K22, PIC18F67K22T PIC18F85K22, PIC18F85K22T PIC18F86K22, PIC18F86K22T PIC18F87K22, PIC18F87K22T	<ul> <li>PIC18F87K22T-E/PT = Tape and reel, Extended temperature, TQFP package</li> </ul>
Temperature Range	I = -40°C to +85°C (Industrial) E = -40°C to +125°C (Extended)	
Package	PT = TQFP (Plastic Thin Quad Flatpack) MR = QFN (Plastic Quad Flat)	Note 1: F = Standard Voltage Range 2: T = In tape and reel PLCC and
Pattern	QTP, SQTP, Code or Special Requirements (blank otherwise)	TQFP packages only 3: RSL = Silicon Revision A3



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