

PCA85233

Universal LCD driver for low multiplex rates

Rev. 1 — 17 September 2013

Product data sheet

1. General description

The PCA85233 is a peripheral device which interfaces to almost any Liquid Crystal Display (LCD)¹ with low multiplex rates. It generates the drive signals for any static or multiplexed LCD containing up to four backplanes and up to 80 segments and can easily be cascaded for larger LCD applications. The PCA85233 is compatible with most microcontrollers and communicates via the two-line bidirectional I²C-bus. Communication overheads are minimized by a display RAM with auto-incremental addressing, by hardware subaddressing, and by display memory switching (static and duplex drive modes).

For a selection of NXP LCD segment drivers, see Table 29 on page 48.

2. Features and benefits

- AEC-Q100 compliant for automotive applications
- Single-chip LCD controller and driver
- Selectable backplane drive configuration: static, 2, 3, or 4 backplane multiplexing
- Selectable display bias configuration: static, ½, or ⅓
- Selectable frame frequency: 150 Hz or 220 Hz
- Internal LCD bias generation with voltage-follower buffers
- 80 segment drives:
 - Up to 40 7-segment alphanumeric characters
 - ◆ Up to 20 14-segment alphanumeric characters
 - Any graphics of up to 320 elements
- 80 × 4 bit RAM for display data storage
- Auto-incremental display data loading across device subaddress boundaries
- Display memory bank switching in static and duplex drive modes
- Versatile blinking modes
- Independent supplies possible for LCD and logic voltages
- Wide power supply range: from 1.8 V to 5.5 V
- Wide logic LCD supply range:
 - From 2.5 V for low-threshold LCDs
 - ◆ Up to 8.0 V for guest-host LCDs and high-threshold twisted nematic LCDs
- Low power consumption
- 400 kHz I²C-bus interface
- Extended temperature range up to 105 °C
- May be cascaded for large LCD applications (up to 5120 segments possible)

^{1.} The definition of the abbreviations and acronyms used in this data sheet can be found in Section 20.



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- No external components needed
- Compatible with Chip-On-Glass (COG) technology
- Manufactured using silicon gate CMOS process

3. Ordering information

Table 1. Ordering information

Type number	Package								
	Name	Description	Version						
PCA85233UG	bare die	110 bumps	PCA85233						

3.1 Ordering options

Table 2. Ordering options

Product type number	Sales item (12NC)	Orderable part number	IC revision	Delivery form
PCA85233UG/2DA/Q1	935302508033	PCA85233UG/2DA/Q1Z	1	chip with hard bumps in tray[1]

^[1] Bump hardness see Table 26.

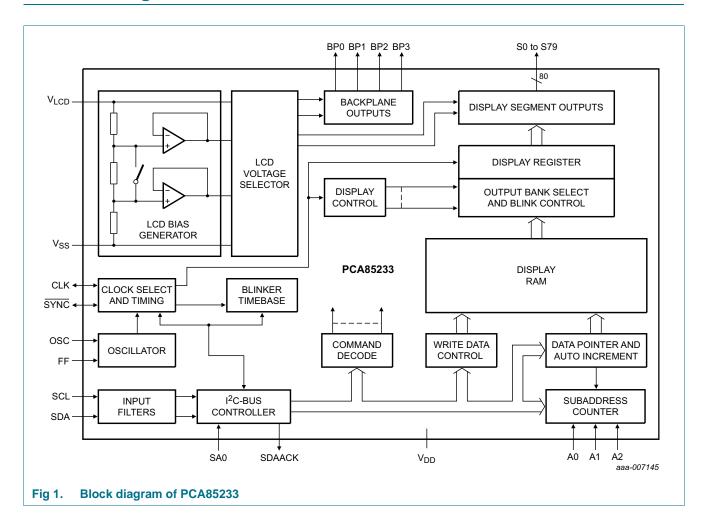
4. Marking

Table 3. Marking codes

Type number	Marking code
PCA85233UG/2DA/Q1	PC85233-1

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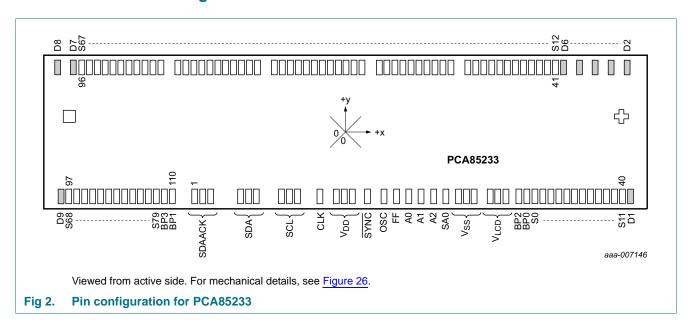
5. Block diagram



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6. Pinning information

6.1 Pinning



6.2 Pin description

Table 4. Pin description overview

Input or input/output pins must always be at a defined level (V_{SS} or V_{DD}) unless otherwise specified.

Symbol	Pin	Description
SDAACK	1 to 3	I ² C-bus acknowledge output
SDA	4 to 6	I ² C-bus serial data input
SCL	7 to 9	I ² C-bus serial clock input
CLK	10	clock input and output
V _{DD}	11 to 13	supply voltage
SYNC	14	cascade synchronization input and output
OSC	15	oscillator select
FF	16	frame frequency select
A0, A1, and A2	17 to 19	subaddress input
SA0	20	I ² C-bus slave address input
V _{SS} [1]	21 to 23	ground supply voltage
V _{LCD}	24 to 26	LCD supply voltage
BP2, BP0, BP3, and BP1	27, 28, 109 and 110	LCD backplane output
S0 to S79	29 to 108	LCD segment output
D1 to D9	-	dummy pins

^[1] The substrate (rear side of the die) is at V_{SS} potential and should be electrically isolated.

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7. Functional description

The PCA85233 is a versatile peripheral device designed to interface between any microcontroller to a wide variety of LCD segment or dot matrix displays (see <u>Figure 3</u>). It can directly drive any static or multiplexed LCD containing up to four backplanes and up to 80 segments.

The display configurations possible with the PCA85233 depend on the required number of active backplane outputs. A selection of display configurations is given in <u>Table 5</u>.

All of the display configurations given in $\underline{\text{Table 5}}$ can be implemented in a typical system as shown in $\underline{\text{Figure 4}}$.

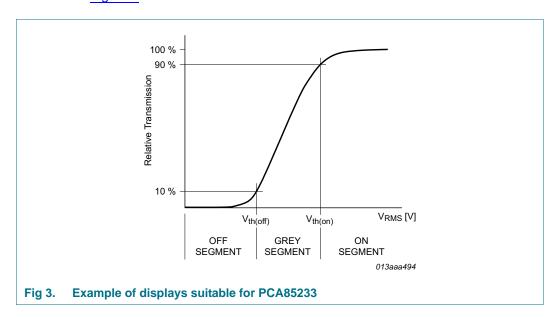


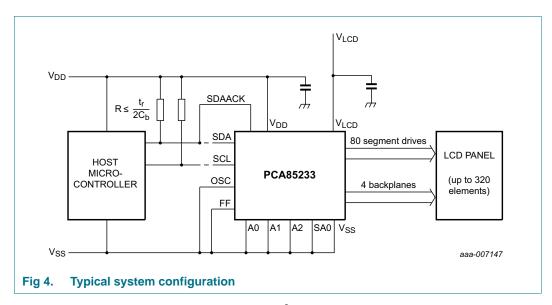
Table 5. Selection of possible display configurations

Number of										
Backplanes	Icons	Digits/Characte	Digits/Characters							
		7-segment[1]	14-segment[2]	Elements						
4	320	40	20	320 (4 × 80)						
3	240	30	15	240 (3 × 80)						
2	160	20	10	160 (2 × 80)						
1	80	10	5	80 (1 × 80)						

^{[1] 7} segment display has 8 elements including the decimal point.

^{[2] 14} segment display has 16 elements including decimal point and accent dot.

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The host microcontroller maintains the 2-line I²C-bus communication channel with the PCA85233. The internal oscillator is enabled by connecting pin OSC to pin V_{SS} . The appropriate biasing voltages for the multiplexed LCD waveforms are generated internally. The only other connections required to complete the system are the power supplies (V_{DD} , V_{SS} , and V_{LCD}) and the LCD panel chosen for the application.

7.1 Initialization

At power-on the status of the I²C-bus and the registers of the PCA85233 is undefined. Therefore the PCA85233 should be initialized as quickly as possible after power-on to ensure a proper bus communication and to avoid display artifacts. The following instructions should be accomplished for initialization:

- I²C-bus (see <u>Section 8</u>) initialization (only necessary if no communication with another I²C device on the bus has already taken place)
 - generating a START condition
 - sending 0h and ignoring the acknowledge
 - generating a STOP condition
- Mode-set command (see Table 12), setting
 - bit E = 0
 - bit B to the required LCD bias configuration
 - bits M[1:0] to the required LCD drive mode
- Load-data-pointer command (see Table 13), setting
 - bits P[5:0] to 0h (or any other required address)
- Device-select command (see <u>Table 14</u>), setting
 - bits A[2:0] to the required hardware subaddress (for example, 0h)
- Bank-select command (see <u>Table 15</u>), setting
 - bit I to 0
 - bit O to 0

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- Blink-select command (see Table 16), setting
 - bit AB to 0 or 1
 - bits BF[1:0] to 00 (or to a desired blinking mode)
- writing meaningful information (for example, a logo) into the display RAM

After the initialization, the display can be switched on by setting bit E = 1 with the mode-set command.

7.2 LCD bias generator

Fractional LCD biasing voltages are obtained from an internal voltage divider of three impedances connected between pins V_{LCD} and V_{SS} . The center impedance is bypassed by switch if the $\frac{1}{2}$ bias voltage level for the 1:2 multiplex drive mode configuration is selected.

7.3 LCD voltage selector

The LCD voltage selector coordinates the multiplexing of the LCD in accordance with the selected LCD drive configuration. The operation of the voltage selector is controlled by the mode-set command from the command decoder. The biasing configurations that apply to the preferred modes of operation, together with the biasing characteristics as functions of V_{LCD} and the resulting discrimination ratios (D) are given in Table 6.

Discrimination is a term which is defined as the ratio of the on and off RMS voltage across a segment. It can be thought of as a measurement of contrast.

Table 6. Biasing characteristics

LCD drive	Number of:		LCD bias	$V_{off(RMS)}$	$V_{on(RMS)}$	$D = V_{on(RMS)}$	
mode	Backplanes	Levels	configuration	V _{LCD}	V _{LCD}	$D = \frac{on(RMS)}{V_{off(RMS)}}$	
static	1	2	static	0	1	∞	
1:2 multiplex	2	3	1/2	0.354	0.791	2.236	
1:2 multiplex	2	4	1/3	0.333	0.745	2.236	
1:3 multiplex	3	4	1/3	0.333	0.638	1.915	
1:4 multiplex	4	4	1/3	0.333	0.577	1.732	

A practical value for V_{LCD} is determined by equating $V_{off(RMS)}$ with a defined LCD threshold voltage ($V_{th(off)}$), typically when the LCD exhibits approximately 10 % contrast. In the static drive mode a suitable choice is $V_{LCD} > 3V_{th(off)}$.

Multiplex drive modes of 1:3 and 1:4 with $\frac{1}{2}$ bias are possible but the discrimination and hence the contrast ratios are smaller.

Bias is calculated by $\frac{1}{1+a}$, where the values for a are

$$a = 1$$
 for $\frac{1}{2}$ bias

$$a = 2$$
 for $\frac{1}{3}$ bias

The RMS on-state voltage (V_{on(RMS)}) for the LCD is calculated with Equation 1:

$$V_{on(RMS)} = V_{LCD} \sqrt{\frac{a^2 + 2a + n}{n \times (1 + a)^2}}$$
 (1)

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where the values for n are

n = 1 for static drive mode

n = 2 for 1:2 multiplex drive mode

n = 3 for 1:3 multiplex drive mode

n = 4 for 1:4 multiplex drive mode

The RMS off-state voltage (Voff(RMS)) for the LCD is calculated with Equation 2:

$$V_{off(RMS)} = V_{LCD} \sqrt{\frac{a^2 - 2a + n}{n \times (1 + a)^2}}$$
 (2)

Discrimination is the ratio of $V_{on(RMS)}$ to $V_{off(RMS)}$ and is determined from Equation 3:

$$D = \frac{V_{on(RMS)}}{V_{off(RMS)}} = \sqrt{\frac{a^2 + 2a + n}{a^2 - 2a + n}}$$
 (3)

Using Equation 3, the discrimination for an LCD drive mode of 1:3 multiplex with 1/2 bias is $\sqrt{3} = 1.732$ and the discrimination for an LCD drive mode of 1:4 multiplex with 1/2 bias is $\frac{\sqrt{21}}{3} = 1.528$.

The advantage of these LCD drive modes is a reduction of the LCD full scale voltage V_{LCD} as follows:

- 1:3 multiplex ($\frac{1}{2}$ bias): $V_{LCD} = \sqrt{6} \times V_{off(RMS)} = 2.449 V_{off(RMS)}$
- 1:4 multiplex (½ bias): $V_{LCD} = \left[\frac{(4 \times \sqrt{3})}{3}\right] = 2.309 V_{off(RMS)}$

These compare with $V_{LCD} = 3V_{off(RMS)}$ when $\frac{1}{3}$ bias is used.

V_{LCD} is sometimes referred as the LCD operating voltage.

7.3.1 Electro-optical performance

Suitable values for $V_{on(RMS)}$ and $V_{off(RMS)}$ are dependent on the LCD liquid used. The RMS voltage, at which a pixel will be switched on or off, determine the transmissibility of the pixel.

For any given liquid, there are two threshold values defined. One point is at 10 % relative transmission (at $V_{th(off)}$) and the other at 90 % relative transmission (at $V_{th(on)}$), see Figure 5. For a good contrast performance, the following rules should be followed:

$$V_{on(RMS)} \ge V_{th(on)} \tag{4}$$

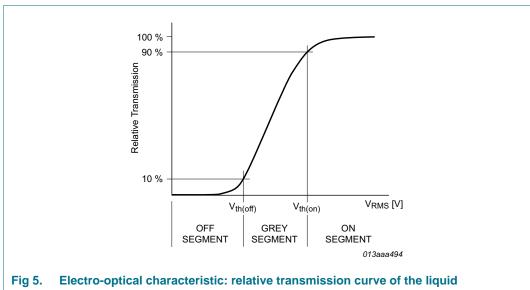
$$V_{off(RMS)} \le V_{th(off)}$$
 (5)

 $V_{on(RMS)}$ and $V_{off(RMS)}$ are properties of the display driver and are affected by the selection of a, n (see <u>Equation 1</u> to <u>Equation 3</u>) and the V_{LCD} voltage.

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 $V_{th(off)}$ and $V_{th(on)}$ are properties of the LCD liquid and can be provided by the module manufacturer. $V_{th(off)}$ is sometimes named V_{th} . $V_{th(on)}$ is sometimes named saturation voltage V_{sat} .

It is important to match the module properties to those of the driver in order to achieve optimum performance.

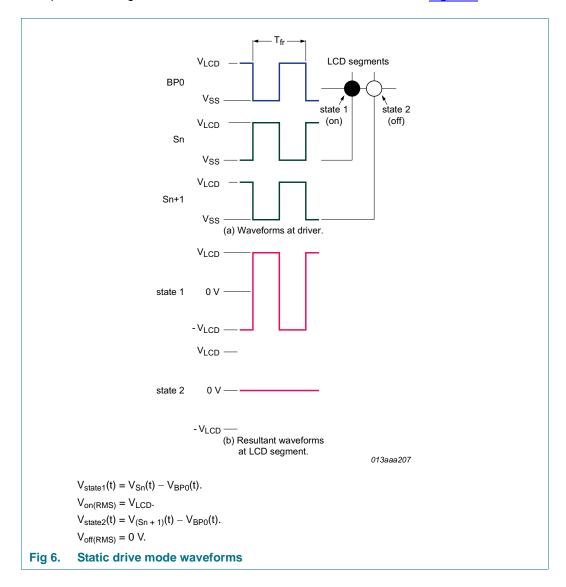


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7.4 LCD drive mode waveforms

7.4.1 Static drive mode

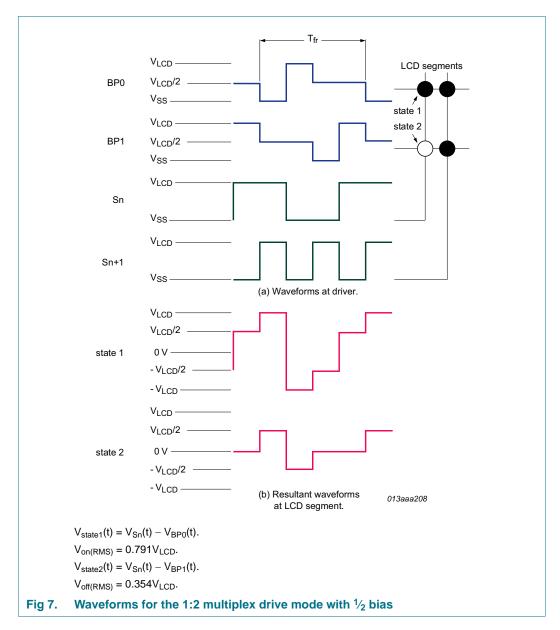
The static LCD drive mode is used when a single backplane is provided in the LCD. Backplane and segment drive waveforms for this mode are shown in Figure 6.



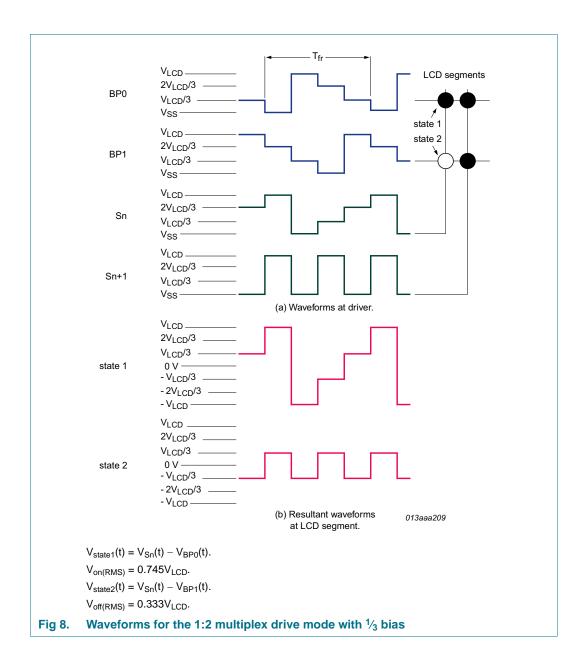
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7.4.2 1:2 Multiplex drive mode

When two backplanes are provided in the LCD, the 1:2 multiplex mode applies. The PCA85233 allows the use of $\frac{1}{2}$ bias or $\frac{1}{3}$ bias in this mode as shown in Figure 7 and Figure 8.



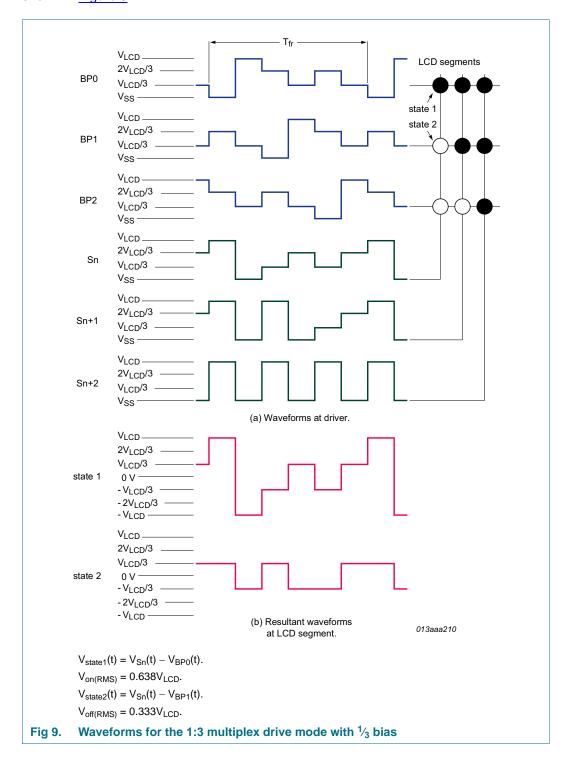
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7.4.3 1:3 Multiplex drive mode

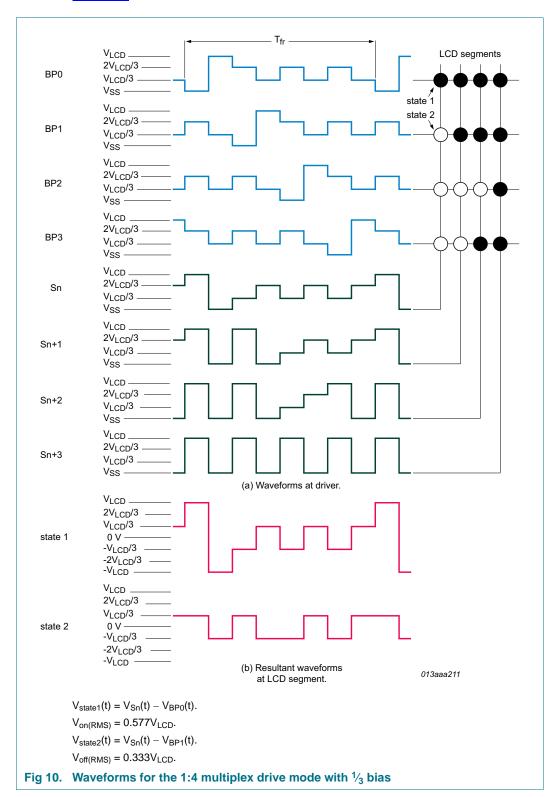
When three backplanes are provided in the LCD, the 1:3 multiplex drive mode applies, as shown in Figure 9.



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7.4.4 1:4 Multiplex drive mode

When four backplanes are provided in the LCD, the 1:4 multiplex drive mode applies, as shown in Figure 10.



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7.5 Oscillator

The internal logic and the LCD drive signals of the PCA85233 are timed by a frequency f_{clk} which either is derived from the built-in oscillator frequency f_{osc}:

$$f_{clk} = \frac{f_{osc}}{64} \tag{6}$$

or equals an external clock frequency fclk(ext):

$$f_{clk} = f_{clk(ext)} \tag{7}$$

7.5.1 Internal clock

The internal oscillator is enabled by connecting pin OSC to V_{SS} . In this case the output from pin CLK provides the clock signal for any cascaded PCA85233 in the system.

7.5.2 External clock

Connecting pin OSC to V_{DD} enables an external clock source. Pin CLK then becomes the external clock input.

Remark: A clock signal must always be supplied to the device; removing the clock may freeze the LCD in a DC state, which is not suitable for the liquid crystal.

7.6 Timing and frame frequency

The clock frequency f_{clk} determines the LCD frame frequency f_{fr} and is calculated as follows:

$$f_{fr} = \frac{f_{clk}}{24} \tag{8}$$

The internal clock frequency f_{clk} can be selected using pin FF. As a result 2 frame frequencies are available: 150 Hz or 220 Hz (typical), see <u>Table 7</u>.

Table 7. LCD frame frequencies

Pin FF tied to	Typical clock frequency (Hz)	LCD frame frequency (Hz)
V_{DD}	3600	150
V _{SS}	5280	220

The timing of the PCA85233 organizes the internal data flow of the device. This includes the transfer of display data from the display RAM to the display segment outputs. In cascaded applications, the synchronization signal (SYNC) maintains the correct timing relationship between all the PCA85233 in the system.

7.7 Display register

The display register holds the display data while the corresponding multiplex signals are generated.

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7.8 Segment outputs

The LCD drive section includes 80 segment outputs (S0 to S79) which must be connected directly to the LCD. The segment output signals are generated in accordance with the multiplexed backplane signals and with data residing in the display register. When less than 80 segment outputs are required the unused segment outputs must be left open-circuit.

7.9 Backplane outputs

The LCD drive section includes four backplane outputs: BP0 to BP3. The backplane output signals are generated in accordance with the selected LCD drive mode.

• In the 1:4 multiplex drive mode BP0 to BP3 must be connected directly to the LCD.

If less than four backplane outputs are required the unused outputs can be left open-circuit.

- In 1:3 multiplex drive mode: BP3 carries the same signal as BP1; therefore, these two
 adjacent outputs can be tied together to give enhanced drive capabilities.
- In 1:2 multiplex drive mode: BP0 and BP2, respectively, BP1 and BP3 carry the same signals and can also be paired to increase the drive capabilities.
- In static drive mode: The same signal is carried by all four backplane outputs; and they can be connected in parallel for very high drive requirements.

7.10 Display RAM

The display RAM is a static 80 × 4 bit RAM which stores LCD data.

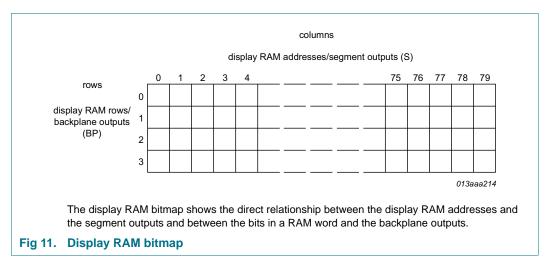
There is a one-to-one correspondence between

- the bits in the RAM bitmap and the LCD elements
- the RAM columns and the segment outputs
- the RAM rows and the backplane outputs.

A logic 1 in the RAM bitmap indicates the on-state of the corresponding LCD element; similarly, a logic 0 indicates the off-state.

The display RAM bit map, <u>Figure 11</u>, shows rows 0 to 3 which correspond with the backplane outputs BP0 to BP3, and columns 0 to 79 which correspond with the segment outputs S0 to S79. In multiplexed LCD applications the segment data of the first, second, third and fourth row of the display RAM are time-multiplexed with BP0, BP1, BP2, and BP3 respectively.

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When display data is transmitted to the PCA85233, the received display bytes are stored in the display RAM in accordance with the selected LCD drive mode. The data is stored as it arrives and depending on the current multiplex drive mode the bits are stored singularly, in pairs, triples or quadruples. To illustrate the filling order, an example of a 7-segment display showing all drive modes is given in Figure 12; the RAM filling organization depicted applies equally to other LCD types.

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Static Sm-2	drive mode	LCD segments	LCD backplanes			(display F	RAM fillir	ng order					transmitted di	splay byte
Static Shr4		S _{n+2} — a				dis		M addre	ss/segm	·					
Sintant Sint	statio		BP0	diamin. DAM			n + 2		n + 4	n + 5					
Sn+6 -	Static	$A = A = S_n$		rows/backplane							l -			l	
1:2				2	x	l .		х		х	l	х		c b a f g	e d DP
display RAM address/segment outputs (s) byte1 byte2 multiplex Sn+1		3n+6 - <u>a</u> <u>v</u> O DP		3	Х	×	Х	X	×	Х	×	×			
multiplex Sn+1		S _n a	BP0			dis b	play RA								
multiplex Sn+2 - e c columns outputs (BP) 1 Sn+2 - f c columns outputs (BP) 2 Sn+3 - d c columns display RAM address/segment outputs (s) byte1 byte2 byte3 MSB LSB Tows display RAM address/segment outputs (s) byte1 byte2 byte3 Tows display RAM address/segment outputs (s) byte1 byte2 byte3 DP c a d g f e DP c a d g f e MSB LSB DP c a d g f e MSB LSB DP c a d g f e MSB LSB DP c a d g f e DP d e Outputs (BP) 2 3 x x x x x x MSB LSB DP c a d g f e	1:2	S _{n+1} — f b		rows	n	n + 1	n + 2	n + 3				;			
Sn+2 - e	multiplex	g		rows/backplane	-	· ·									
The second secon		I 🗀 🎮	L BP1	outputs (BP)			-					į		a b f g e	c d DP
display RAM address/segment outputs (s) byte1 byte2 byte3 multiplex S_{n+1}		S _{n+3} – d DP		3	Х	х	х	х	L			i			
1:3 S _{n+2} - f		S _{n+1} a	BP0 T				play RA	M addre	ss/segm	ent outp		3			
multiplex Mode Mod	1:3			rows	n	n + 1	n + 2			-		 !	! 	Med	LCD
outputs (BP) 1 DP	multiploy			display RAM 0			f								
display RAM address/segment outputs (s) byte1 byte2 byte3 byte4 byte5 rows display RAM or nows/backplane outputs (BP) 2 b g MSB LSB a c b DP f e g d	muniplex	e c	BP1 BP2	outputs (BP)						i		i		b DP c a	d g f e
display RAM address/segment outputs (s) byte1 byte2 byte3 byte4 byte5 Sn		d OP		3	х		х	L				;	!		
1:4 multiplex BP2 rows display RAM rows/backplane 1 outputs (BP) 2 b g MSB LSB								M addre	ss/segm						
multiplex	1:4	S _n — a b	BP2			_	byte	; - :	·		4 · - ·	oyte5 			
multiplex BP1 BP3 rows/backplane 1 c e outputs (BP) 2 b g a c b DP f e g d		f	BPO					‡							
	multiplex		BP1 DP2	rows/backplane 1	С			1				i		l	
		S _{n+1} ODP		2				į			į		i i	a c b DP	f e g d
001aaj646							L	· - '- - ·							

x = data bit unchanged

Fig 12. Relationships between LCD layout, drive mode, display RAM filling order, and display data transmitted over the I²C-bus

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The following applies to Figure 12:

- In static drive mode the eight transmitted data bits are placed into row 0 as one byte.
- In 1:2 multiplex drive mode the eight transmitted data bits are placed in pairs into row 0 and 1 as two successive 4-bit RAM words.
- In 1:3 multiplex drive mode the eight bits are placed in triples into row 0, 1, and 2 as three successive 3-bit RAM words, with bit 3 of the third address left unchanged. It is not recommended to use this bit in a display because of the difficult addressing. This last bit may, if necessary, be controlled by an additional transfer to this address, but care should be taken to avoid overwriting adjacent data because always full bytes are transmitted (see Section 7.10.3).
- In 1:4 multiplex drive mode, the eight transmitted data bits are placed in quadruples into row 0, 1, 2, and 3 as two successive 4-bit RAM words.

7.10.1 Data pointer

The addressing mechanism for the display RAM is realized using a data pointer. This allows the loading of an individual display data byte, or a series of display data bytes, into any location of the display RAM. The sequence commences with the initialization of the data pointer by the load-data-pointer command (see <u>Table 13</u>). Following this command, an arriving data byte is stored at the display RAM address indicated by the data pointer. The filling order is shown in <u>Figure 12</u>. After each byte is stored, the content of the data pointer is automatically incremented by a value dependent on the selected LCD drive mode:

- In static drive mode by eight
- In 1:2 multiplex drive mode by four
- In 1:3 multiplex drive mode by three
- In 1:4 multiplex drive mode by two

If an I²C-bus data access is terminated early then the state of the data pointer is unknown. Consequently, the data pointer must be rewritten prior to further RAM accesses.

7.10.2 Subaddress counter

The storage of display data is determined by the content of the subaddress counter. Storage is allowed only when the content of the subaddress counter match with the hardware subaddress applied to A0, A1, and A2. The subaddress counter value is defined by the device-select command (see <u>Table 14</u>). If the content of the subaddress counter and the hardware subaddress do not match, then data storage is inhibited but the data pointer is incremented as if data storage had taken place. The subaddress counter is also incremented when the data pointer overflows.

The storage arrangements described lead to extremely efficient data loading in cascaded applications. When a series of display bytes are sent to the display RAM, automatic wrap-over to the next PCA85233 occurs when the last RAM address is exceeded. Subaddressing across device boundaries is successful even if the change to the next device in the cascade occurs within a transmitted character.

The hardware subaddress must not be changed whilst the device is being accessed on the I²C-bus interface.

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7.10.3 RAM writing in 1:3 multiplex drive mode

In 1:3 multiplex drive mode, the RAM is written as shown in <u>Table 8</u> (see <u>Figure 12</u> as well).

Table 8. Standard RAM filling in 1:3 multiplex drive mode

Assumption: BP2/S2, BP2/S5, BP2/S8 etc. are not connected to any elements on the display.

Display RAM bits (rows)/ backplane outputs (BPn)	Disp	Display RAM addresses (columns)/segment outputs (Sn)										
	0	1	2	3	4	5	6	7	8	9	:	
0	a7	a4	a1	b7	b4	b1	с7	c4	c1	d7	:	
1	a6	а3	a0	b6	b3	b0	c6	сЗ	c0	d6	:	
2	a5	a2	-	b5	b2	-	c5	c2	-	d5	:	
3	-	-	-	-	-	-	-	-	-	-	:	

If the bit at position BP2/S2 would be written by a second byte transmitted, then the mapping of the segment bits would change as illustrated in Table 9.

Table 9. Entire RAM filling by rewriting in 1:3 multiplex drive mode

Assumption: BP2/S2, BP2/S5, BP2/S8 etc. are connected to elements on the display.

Display RAM	Displa	Display RAM addresses (columns)/segment outputs (Sn)											
bits (rows)/ backplane outputs (BPn)	0	1	2	3	4	5	6	7	8	9	:		
0	a7	a4	a1/b7	b4	b1/c7	c4	c1/d7	d4	d1/e7	e4	:		
1	a6	a3	a0/b6	b3	b0/c6	сЗ	c0/d6	d3	d0/e6	е3	:		
2	a5	a2	b5	b2	c5	c2	d5	d2	e5	e2	:		
3	-	-	-	-	-	-	-	-	-	-	:		

In the case described in <u>Table 9</u> the RAM has to be written entirely and BP2/S2, BP2/S5, BP2/S8 etc. have to be connected to elements on the display. This can be achieved by a combination of writing and rewriting the RAM like follows:

- In the first write to the RAM, bits a7 to a0 are written.
- In the second write, bits b7 to b0 are written, overwriting bits a1 and a0 with bits b7 and b6.
- In the third write, bits c7 to c0 are written, overwriting bits b1 and b0 with bits c7 and c6.

Depending on the method of writing to the RAM (standard or entire filling by rewriting), some elements remain unused or can be used, but it has to be considered in the module layout process as well as in the driver software design.

7.10.4 Writing over the RAM address boundary

In all multiplex drive modes, depending on the setting of the data pointer, it is possible to fill the RAM over the RAM address boundary. If the PCA85233 is part of a cascade the additional bits fall into the next device that also generates the acknowledge signal. If the PCA85233 is a single device or the last device in a cascade the additional bits will be discarded and no acknowledge signal will be generated.

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7.10.5 Output bank selector

The output bank selector (see <u>Table 15</u>) selects one of the four rows per display RAM address for transfer to the display register. The actual row selected depends on the selected LCD drive mode in operation and on the instant in the multiplex sequence.

- In 1:4 multiplex mode, all RAM addresses of row 0 are selected, these are followed by the contents of row 1, 2, and then 3
- In 1:3 multiplex mode, rows 0, 1, and 2 are selected sequentially
- In 1:2 multiplex mode, rows 0 and 1 are selected
- In static mode, row 0 is selected

The PCA85233 includes a RAM bank switching feature in the static and 1:2 multiplex drive modes. In the static drive mode, the bank-select command may request the contents of row 2 to be selected for display instead of the contents of row 0. In the 1:2 multiplex mode, the contents of rows 2 and 3 may be selected instead of rows 0 and 1. This gives the provision for preparing display information in an alternative bank and to be able to switch to it once it is assembled.

7.10.6 Input bank selector

The input bank selector loads display data into the display RAM in accordance with the selected LCD drive configuration. Display data can be loaded in row 2 in static drive mode or in rows 2 and 3 in 1:2 multiplex drive mode by using the bank-select command (see Table 15). The input bank selector functions independently to the output bank selector.

7.11 Blinking

The display blink capabilities of the PCA85233 are very versatile. The whole display can blink at frequencies selected by the blink-select command (see <u>Table 16</u>). The blink frequencies are derived from the clock frequency. The ratios between the clock and blink frequencies depend on the blink mode selected (see <u>Table 10</u>).

Table 10. Blink frequencies

Blink mode	Operating mode ratio	Blink frequency with	respect to f _{clk} (typical)	Unit
		f _{clk} = 3.600 kHz	f _{clk} = 5.280 kHz	
off	-	blinking off	blinking off	Hz
1	$\frac{f_{clk}}{768}$	4.7	6.9	Hz
2	$\frac{f_{clk}}{1536}$	2.3	3.4	Hz
3	$\frac{f_{clk}}{3072}$	1.2	1.7	Hz

An additional feature is for an arbitrary selection of LCD segments to blink. This applies to the static and 1:2 multiplex drive modes and can be implemented without any communication overheads. By means of the output bank selector, the displayed RAM banks are exchanged with alternate RAM banks at the blink frequency. This mode can also be specified by the blink-select command.

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In the 1:3 and 1:4 multiplex modes, where no alternate RAM bank is available, groups of LCD segments can blink by selectively changing the display RAM data at fixed time intervals.

If the entire display can blink at a frequency other then the typical blink frequency. This can be effectively performed by resetting and setting the display enable bit E at the required rate using the mode-set command (see Table 12).

7.12 Command decoder

The command decoder identifies command bytes that arrive on the I²C-bus. The commands available to the PCA85233 are defined in <u>Table 11</u>.

Table 11. Definition of commands

Command	Ope	ration o	Reference						
Bit	7	6	5	4	3	2	1		
mode-set	1	1	1 0 0 E B M[1:0]						
load-data-pointer	0	P[6:0	0]		Table 13				
device-select	1	1	1	0	0	A[2:0]		Table 14
bank-select	1	1	1	1	1	0	I	Table 15	
blink-select	1	1	1	1	0	AB	BF[1:	Table 16	

Table 12. Mode-set command bit description

Bit	Symbol	Value	Description
7 to 4	-	1100	fixed value
3	Е		display status ^[1]
		0	disabled (blank) ^[2]
		1	enabled
2	В		LCD bias configuration[3]
		0	$\frac{1}{3}$ bias
		1	$\frac{1}{2}$ bias
1 to 0	M[1:0]		LCD drive mode selection
		01	static; 1 backplane
		10	1:2 multiplex; 2 backplanes
		11	1:3 multiplex; 3 backplanes
		00	1:4 multiplex; 4 backplanes

^[1] The possibility to disable the display allows implementation of blinking under external control.

^[2] The display is disabled by setting all backplane and segment outputs to V_{LCD} .

^[3] Not applicable for static drive mode.

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Table 13. Load-data-pointer command bit description See Section 7.10.1.

Bit	Symbol	Value	Description
7	-	0	fixed value
6 to 0	[data pointer
		1001111	7-bit binary value of 0 to 79, transferred to the data pointer to define one of 80 display RAM addresses

Table 14. Device-select command bit description See Section 7.10.2.

Bit	Symbol	Value	Description
7 to 3	-	11100	fixed value
2 to 0	A[2:0]	000 to 111	device selection
			3-bit binary value of 0 to 7, transferred to the subaddress counter to define one of 8 hardware subaddresses

Table 15. Bank-select command bit description See Section 7.10.5 and Section 7.10.6.

Bit	Symbol	Value	Description		
			Static	1:2 multiplex	
7 to 2	-	111110	fixed value		
1	I		input bank selection: storage of a	arriving display data	
		0	RAM row 0	RAM rows 0 and 1	
		1	RAM row 2	RAM rows 2 and 3	
0	0		output bank selection: retrieval o	f LCD display data	
		0	RAM row 0	RAM rows 0 and 1	
		1	RAM row 2	RAM rows 2 and 3	

^[1] The bank-select command has no effect in 1:3 or 1:4 multiplex drive modes.

Table 16. Blink-select command bit description See Section 7.11.

<u> </u>				
Bit	Symbol	Value	Description	
7 to 3	-	11110	fixed value	
2	AB		blink mode selection[1]	
		0	normal blinking	
		1	blinking by alternating display RAM banks	
1 to 0	BF[1:0]		blink frequency selection[2]	
		00	off	
		01	1	
		10	2	
		11	3	

^[1] Normal blinking can only be selected in multiplex drive mode 1:3 or 1:4.

^[2] For the blink frequencies, see <u>Table 10</u>.

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7.13 Display controller

The display controller executes the commands identified by the command decoder. It contains the status registers and coordinates their effects. The display controller also loads the display data into the display RAM as required by the storage order.

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8. Characteristics of the I²C-bus

The I²C-bus is for bidirectional, two-line communication between different ICs or modules. The two lines are a Serial DAta line (SDA) and a Serial CLock line (SCL). Both lines must be connected to a positive supply via a pull-up resistor when connected to the output stages of a device. Data transfer may be initiated only when the bus is not busy.

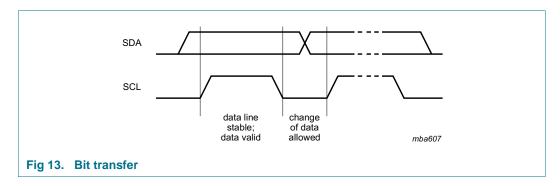
By connecting pin SDAACK to pin SDA on the PCA85233, the SDA line becomes fully I²C-bus compatible. In COG applications where the track resistance from the SDAACK pin to the system SDA line can be significant, possibly a voltage divider is generated by the bus pull-up resistor and the Indium Tin Oxide (ITO) track resistance. As a consequence it may be possible that the acknowledge generated by the PCA85233 can't be interpreted as logic 0 by the master. In COG applications where the acknowledge cycle is required, it is therefore necessary to minimize the track resistance from the SDAACK pin to the system SDA line to guarantee a valid LOW level.

By separating the acknowledge output from the serial data line (having the SDAACK open circuit) design efforts to generate a valid acknowledge level can be avoided. However, in that case the I²C-bus master has to be set up in such a way that it ignores the acknowledge cycle.²

The following definition assumes SDA and SDAACK are connected and refers to the pair as SDA.

8.1 Bit transfer

One data bit is transferred during each clock pulse. The data on the SDA line must remain stable during the HIGH period of the clock pulse as changes in the data line at this time will be interpreted as a control signal (see <u>Figure 13</u>).



8.2 START and STOP conditions

Both data and clock lines remain HIGH when the bus is not busy.

A HIGH-to-LOW change of the data line, while the clock is HIGH, is defined as the START condition (S).

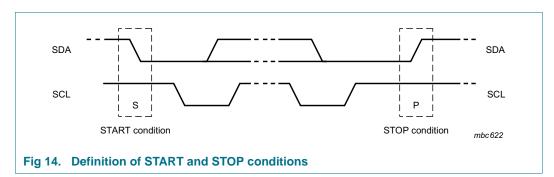
PCA85233

^{2.} For further information, please consider the NXP application note: Ref. 1 "AN10170".

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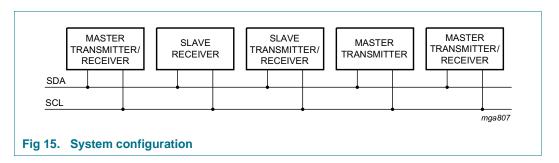
A LOW-to-HIGH change of the data line, while the clock is HIGH, is defined as the STOP condition (P).

The START and STOP conditions are shown in Figure 14.



8.3 System configuration

A device generating a message is a transmitter, a device receiving a message is the receiver. The device that controls the message is the master; and the devices which are controlled by the master are the slaves. The system configuration is shown in Figure 15.



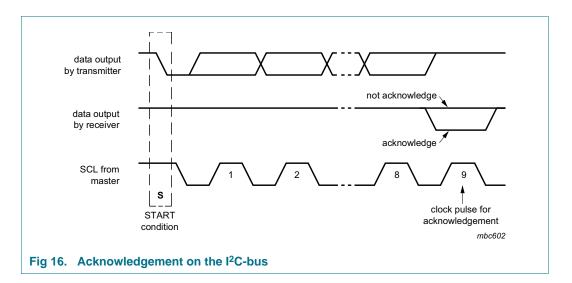
8.4 Acknowledge

The number of data bytes transferred between the START and STOP conditions from transmitter to receiver is unlimited. Each byte of eight bits is followed by an acknowledge cycle.

- A slave receiver, which is addressed, must generate an acknowledge after the reception of each byte.
- A master receiver must generate an acknowledge after the reception of each byte that has been clocked out of the slave transmitter.
- The device that acknowledges must pull-down the SDA line during the acknowledge clock pulse, so that the SDA line is stable LOW during the HIGH period of the acknowledge related clock pulse (set-up and hold times must be taken into consideration).
- A master receiver must signal an end of data to the transmitter by not generating an
 acknowledge on the last byte that has been clocked out of the slave. In this event, the
 transmitter must leave the data line HIGH to enable the master to generate a STOP
 condition.

Acknowledgement on the I²C-bus is shown in Figure 16.

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8.5 l²C-bus controller

The PCA85233 acts as an I^2 C-bus slave receiver. It does not initiate I^2 C-bus transfers or transmit data to an I^2 C-bus master receiver. The only data output from the PCA85233 are the acknowledge signals from the selected devices. Device selection depends on the I^2 C-bus slave address, on the transferred command data, and on the hardware subaddress.

In single device applications, the hardware subaddress inputs A0, A1, and A2 are normally tied to V_{SS} which defines the hardware subaddress 0. In multiple device applications A0, A1, and A2 are tied to V_{SS} or V_{DD} using a binary coding scheme, so that no two devices with a common I^2C -bus slave address have the same hardware subaddress.

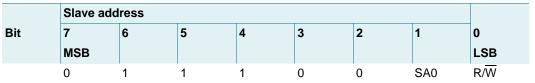
8.6 Input filters

To enhance noise immunity in electrically adverse environments, RC low-pass filters are provided on the SDA and SCL lines.

8.7 I²C-bus protocol

Two I^2C -bus slave addresses (0111 000 and 0111 001) are used to address the PCA85233. The entire I^2C -bus slave address byte is shown in <u>Table 17</u>.

Table 17. I²C slave address byte



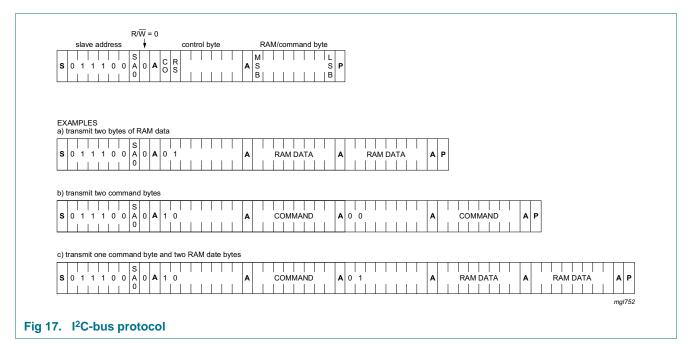
The PCA85233 is a write-only device and will not respond to a read access, therefore bit 0 should always be logic 0. Bit 1 of the slave address byte that a PCA85233 will respond to, is defined by the level tied to its SA0 input (V_{SS} for logic 0 and V_{DD} for logic 1).

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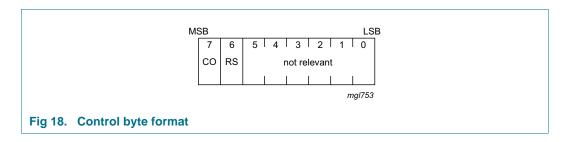
Having two reserved slave addresses allows the following on the same I²C-bus:

- Up to 16 PCA85233 on the same I²C-bus for very large LCD applications
- The use of two types of LCD multiplex drive modes on the same I²C-bus

The I^2C -bus protocol is shown in <u>Figure 17</u>. The sequence is initiated with a START condition (S) from the I^2C -bus master which is followed by one of the available PCA85233 slave addresses. All PCA85233 with the same SA0 level acknowledge in parallel to the slave address. All PCA85233 with the alternative SA0 level ignore the whole I^2C -bus transfer.



After acknowledgement, the control byte is sent, defining if the next byte is a RAM or command information. The control byte also defines if the next byte is a control byte or further RAM or command data (see Figure 18 and <a href=Table 18). In this way it is possible to configure the device and then fill the display RAM with little overhead.



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Table 18. Control byte descrip

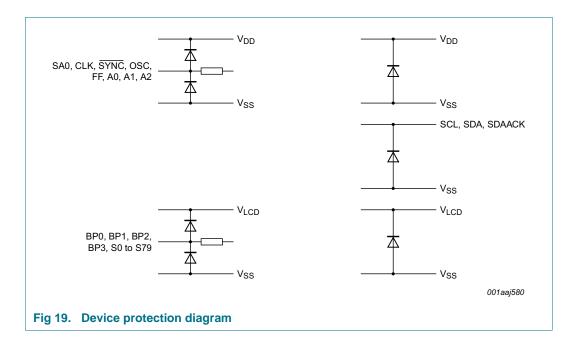
Bit	Symbol	Value	Description
7	CO		continue bit
		0	last control byte
		1	control bytes continue
6	RS		register selection
		0	command register
		1	data register
5 to 0	-		not relevant

The command bytes and control bytes are also acknowledged by all addressed PCA85233 connected to the bus.

The display bytes are stored in the display RAM at the address specified by the data pointer and the subaddress counter. Both data pointer and subaddress counter are automatically updated.

The acknowledgement after each byte is made only by the (A0, A1, and A2) addressed PCA85233. After the last display byte, the I²C-bus master issues a STOP condition (P). Alternatively a START may be asserted to RESTART an I²C-bus access.

9. Internal circuitry



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10. Safety notes

CAUTION



This device is sensitive to ElectroStatic Discharge (ESD). Observe precautions for handling electrostatic sensitive devices.

Such precautions are described in the ANSI/ESD S20.20, IEC/ST 61340-5, JESD625-A or equivalent standards.

CAUTION



Static voltages across the liquid crystal display can build up when the LCD supply voltage (V_{LCD}) is on while the IC supply voltage (V_{DD}) is off, or vice versa. This may cause unwanted display artifacts. To avoid such artifacts, V_{LCD} and V_{DD} must be applied or removed together.

CAUTION



Semiconductors are light sensitive. Exposure to light sources can cause the IC to malfunction. The IC must be protected against light. The protection must be applied to all sides of the IC.

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11. Limiting values

Table 19. Limiting values

In accordance with the Absolute Maximum Rating System (IEC 60134).[1]

Symbol	Parameter	Conditions		Min	Max	Unit
V_{DD}	supply voltage			-0.5	+6.5	V
V_{LCD}	LCD supply voltage			-0.5	+9.0	V
V _{i(n)}	voltage on any input	V _{DD} related inputs		-0.5	+6.5	V
$V_{o(n)}$	voltage on any output	V _{LCD} related outputs		-0.5	+9.0	V
l _l	input current			–10	+10	mA
Io	output current			–10	+10	mA
I_{DD}	supply current			-50	+50	mA
I _{SS}	ground supply current			-50	+50	mA
I _{DD(LCD)}	LCD supply current			-50	+50	mA
P _{tot}	total power dissipation			-	400	mW
P/out	power dissipation per output			-	100	mW
V _{ESD}	electrostatic discharge voltage	НВМ	<u>[2]</u>	-	±4000	V
		MM	[3]	-	±250	V
I _{lu}	latch-up current		[4]	-	100	mA
T _{stg}	storage temperature		<u>[5]</u>	–65	+150	°C
T _{amb}	ambient temperature	operating device		-40	+105	°C

^[1] Stresses above these values listed may cause permanent damage to the device.

^[2] Pass level; Human Body Model (HBM) according to Ref. 7 "JESD22-A114".

^[3] Pass level; Machine Model (MM), according to Ref. 8 "JESD22-A115".

^[4] Pass level; latch-up testing, according to Ref. 9 "JESD78" at maximum ambient temperature (T_{amb(max)}).

^[5] According to the store and transport requirements (see Ref. 12 "UM10569") the devices have to be stored at a temperature of +8 °C to +45 °C and a humidity of 25 % to 75 %.

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12. Static characteristics

Table 20. Static characteristics

 $V_{DD} = 1.8 \text{ V to } 5.5 \text{ V}; V_{SS} = 0 \text{ V}; V_{LCD} = 2.5 \text{ V to } 8.0 \text{ V}; T_{amb} = -40 ^{\circ}\text{C} \text{ to } +105 ^{\circ}\text{C}; unless otherwise specified.}$

Symbol	Parameter	Conditions		Min	Тур	Max	Unit
Supplies							
V_{DD}	supply voltage	$V_{LCD} \le 6.5 \text{ V}$		1.8	-	5.5	V
		V _{LCD} > 6.5 V		2.5	-	5.5	V
V_{LCD}	LCD supply voltage	V _{DD} < 2.5 V		2.5	-	6.5	V
		$V_{DD} \ge 2.5 \text{ V}$		2.5	-	8.0	V
I _{DD}	supply current	f _{clk(ext)} = 1536 Hz; see <u>Figure 20</u>	<u>[1]</u>	-	3	6	μΑ
I _{DD(LCD)}	LCD supply current	f _{clk(ext)} = 1536 Hz; see <u>Figure 20</u>	<u>[1]</u>	-	22	45	μΑ
Logic							
V_{I}	input voltage			$V_{SS}-0.5$	-	$V_{DD} + 0.5$	V
V _{IH}	HIGH-level input voltage	on pins CLK, SYNC, OSC, A0 to A2, SA0, FF		0.7V _{DD}	-	V_{DD}	V
V_{IL}	LOW-level input voltage	on pins CLK, SYNC, OSC, A0 to A2, SA0, FF		V_{SS}	-	$0.3V_{DD}$	V
V _{OH}	HIGH-level output voltage			0.8V _{DD}	-	-	V
V _{OL}	LOW-level output voltage			-	-	0.2V _{DD}	V
I _{OH}	HIGH-level output current	on pin CLK; $V_{OH} = 4.6 \text{ V}$; $V_{DD} = 5 \text{ V}$		+1	-	-	mA
I _{OL}	LOW-level output current	on pin CLK, $\overline{\text{SYNC}}$; $V_{\text{OL}} = 0.4 \text{ V}$; $V_{\text{DD}} = 5 \text{ V}$		-	-	-1	mA
IL	leakage current	on pins OSC, CLK, SCL, SDA, A0 to A2, SA0, FF; $V_I = V_{DD}$ or V_{SS}		-1	-	+1	μΑ
Cı	input capacitance		[3]	-	-	7	pF
I ² C-bus[2]							
Input on pi	ns SDA and SCL						
VI	input voltage			$V_{SS} - 0.5$	-	5.5	V
V _{IH}	HIGH-level input voltage			0.7V _{DD}	-	5.5	V
V_{IL}	LOW-level input voltage			V_{SS}	-	$0.3V_{DD}$	V
Cı	input capacitance		[3]	-	-	7	pF
I _{OL(SDA)}	LOW-level output current on pin SDA	$V_{OL} = 0.4 \text{ V}; V_{DD} = 5 \text{ V}$		+3	-	-	mA

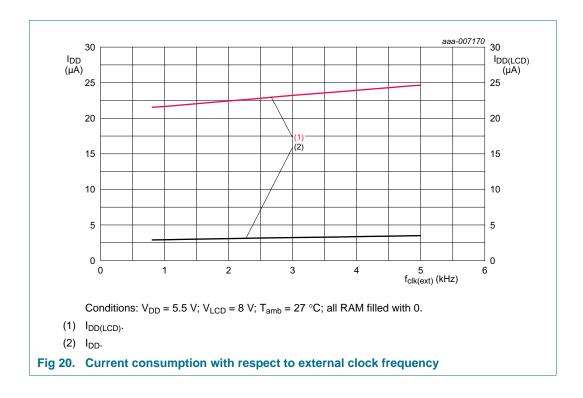
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Table 20. Static characteristics ... continued

 V_{DD} = 1.8 V to 5.5 V; V_{SS} = 0 V; V_{LCD} = 2.5 V to 8.0 V; T_{amb} = -40 °C to +105 °C; unless otherwise specified.

Symbol	Parameter	Conditions	Min	Тур	Max	Unit
LCD outpu	ıts					
ΔV_{O}	output voltage	on pins BP0 to BP3; $C_{bpl} = 35 \text{ nF}$	-100	-	+100	mV
	variation	on pins S0 to S79; $C_{sgm} = 5 \text{ nF}$	-100	-	+100	mV
Ro	output resistance	V _{LCD} = 5 V				
		on pins BP0 to BP3	[4] _	1.5	10	kΩ
		on pins S0 to S79	<u>[4]</u> _	6.0	13.5	kΩ

- [1] LCD outputs are open-circuit; inputs at V_{SS} or V_{DD} ; external clock with 50 % duty factor; I^2C -bus inactive.
- [2] The I²C-bus interface of PCA85233 is 5 V tolerant.
- [3] Not tested, design specification only.
- [4] Outputs measured individually and sequentially.



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13. Dynamic characteristics

Table 21. Dynamic characteristics

 $V_{DD} = 1.8 \text{ V to } 5.5 \text{ V}; V_{SS} = 0 \text{ V}; V_{LCD} = 2.5 \text{ V to } 8.0 \text{ V}; T_{amb} = -40 ^{\circ}\text{C} \text{ to } +105 ^{\circ}\text{C}; unless otherwise specified.}$

Symbol	Parameter	Conditions		Min	Тур	Max	Unit
Clock							
Internal: ou	tput pin CLK						
f _{clk}	clock frequency	FF = V _{SS}	[1][2]	2630	3600	4680	Hz
		$FF = V_{DD}$	[1][2]	3855	5280	6865	Hz
f _{fr}	frame frequency	FF = V _{SS}		-	150	-	Hz
		$FF = V_{DD}$		-	220	-	Hz
Δf_{fr}	frame frequency	FF = V _{SS} ; see <u>Figure 21</u>		110	150	195	Hz
	variation	FF = V _{DD} ; see <u>Figure 21</u>		161	220	286	Hz
External: in	put pin CLK						
f _{clk(ext)}	external clock frequency		[2]	800	-	7000	Hz
t _{clk(H)}	HIGH-level clock time			90	-	-	μS
t _{clk(L)}	LOW-level clock time			90	-	-	μS
Synchronia	zation: input pin SYNC						
t _{PD(SYNC_N)}	SYNC propagation delay			-	30	-	ns
t _{SYNC_NL}	SYNC LOW time			1	-	-	μS
Outputs: p	ins BP0 to BP3 and S0) to S79					
t _{PD(drv)}	driver propagation delay	$V_{LCD} = 5 \text{ V}$		-	-	30	μS
I ² C-bus: tir	ning[<u>3]</u>						
Pin SCL							
f _{SCL}	SCL clock frequency			-	-	400	kHz
t _{HIGH}	HIGH period of the SCL clock			0.6	-	-	μS
t_{LOW}	LOW period of the SCL clock			1.3	-	-	μS
Pin SDA							
t _{SU;DAT}	data set-up time			100	-	-	ns
t _{HD;DAT}	data hold time			0	-	-	ns

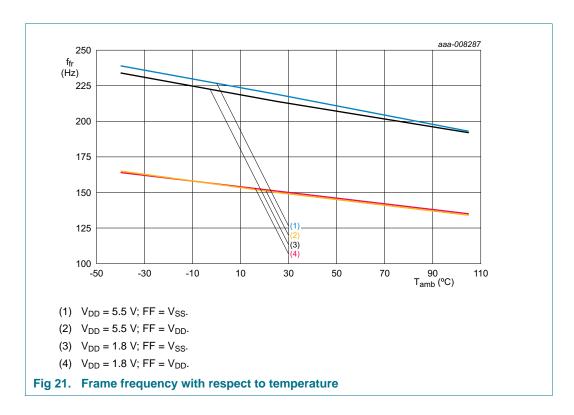
Universal LCD driver for low multiplex rates

Table 21. Dynamic characteristics ...continued $V_{DD} = 1.8 \text{ V to } 5.5 \text{ V; } V_{SS} = 0 \text{ V; } V_{LCD} = 2.5 \text{ V to } 8.0 \text{ V; } T_{amb} = -40 \text{ °C to } +105 \text{ °C; } unless \text{ otherwise specified.}$

Symbol	Parameter	Conditions	Min	Тур	Max	Unit
Pins SCL	and SDA					
t _{BUF}	bus free time between a STOP and START condition		1.3	-	-	μS
t _{SU;STO}	set-up time for STOP condition		0.6	-	-	μS
t _{HD;STA}	hold time (repeated) START condition		0.6	-	-	μS
t _{SU;STA}	set-up time for a repeated START condition		0.6	-	-	μS
t _r	rise time of both SDA	f _{SCL} = 400 kHz	-	-	0.3	μS
	and SCL signals	f _{SCL} < 125 kHz	-	-	1.0	μS
t _f	fall time of both SDA and SCL signals		-	-	0.3	μS
C _b	capacitive load for each bus line		-	-	400	pF
t _{w(spike)}	spike pulse width	on bus	-	-	50	ns

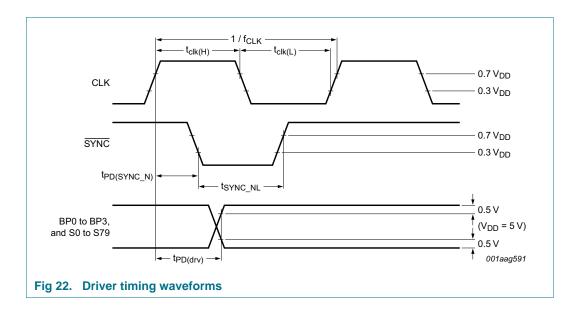
^[1] Typical output duty cycle of 50 %.

^[3] All timing values are valid within the operating supply voltage and ambient temperature range and are referenced to V_{IL} and V_{IH} with an input voltage swing of V_{SS} to V_{DD}. For I²C-bus timings see Figure 23.



^[2] The corresponding frame frequency is $f_{fr} = \frac{f_{clk}}{24}$.

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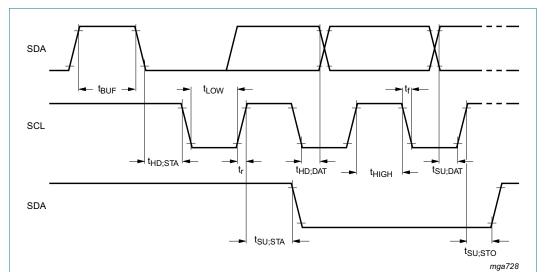


Fig 23. I²C-bus timing waveforms

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14. Application information

14.1 Cascaded operation

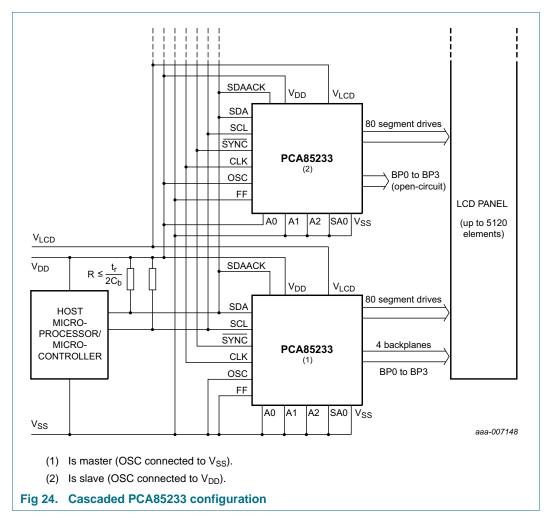
In large display configurations up to 16 PCA85233 can be recognized on the same I²C-bus by using the 3-bit hardware subaddress (A0, A1, and A2) and the programmable I²C-bus slave address (SA0).

Table 22. Addressing cascaded PCA85233

Cluster	Bit SA0	Pin A2	Pin A1	Pin A0	Device
1	0	0	0	0	0
		0	0	1	1
		0	1	0	2
		0	1	1	3
		1	0	0	4
		1	0	1	5
		1	1	0	6
		1	1	1	7
2	1	0	0	0	8
		0	0	1	9
		0	1	0	10
		0	1	1	11
		1	0	0	12
		1	0	1	13
		1	1	0	14
		1	1	1	15

When cascaded PCA85233 are synchronized, they can share the backplane signals from one of the devices in the cascade. Such an arrangement is cost-effective in large LCD applications since the backplane outputs of only one device need to be through-plated to the backplane electrodes of the display. The other PCA85233 of the cascade contribute additional segment outputs. The backplanes can either be connected together to enhance the drive capability or some can be left open-circuit (such as the ones from the slave in Figure 24) or just some of the master and some of the slave will be taken to facilitate the layout of the display.

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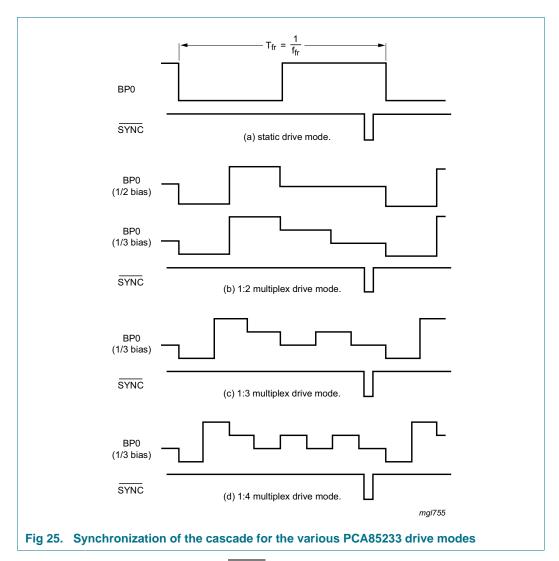
For display sizes that are not multiple of 320 elements, a mixed cascaded system can be considered containing only devices like PCA85233 and PCA85132. Depending on the application, one must take care of the software command and pin connection compatibility.

Only one master but multiple slaves are allowed in a cascade. All devices in the cascade have to use the same clock whether it is supplied externally or provided by the master.

The SYNC line is provided to maintain the correct synchronization between all cascaded PCA85233. The only time that SYNC is likely to be needed is if synchronization is accidentally lost (e.g. by noise in adverse electrical environments or by the definition of a multiplex drive mode when PCA85233 with different SA0 levels are cascaded).

SYNC is organized as an input/output pin; The output selection is realized as an open-drain driver with an internal pull-up resistor. A PCA85233 asserts the SYNC line at the onset of its last active backplane signal and monitors the SYNC line at all other times. If synchronization in the cascade is lost, it is restored by the first PCA85233 to assert SYNC. The timing relationships between the backplane waveforms and the SYNC signal for the various drive modes of the PCA85233 are shown in Figure 25.

Universal LCD driver for low multiplex rates



The contact resistance between the SYNC pins of cascaded devices must be controlled. If the resistance is too high, then the device will not be able to synchronize properly. This is particularly applicable to COG applications. <u>Table 23</u> shows the limiting values for contact resistance.

Table 23. SYNC contact resistance

Number of devices	Maximum contact resistance
2	6000 Ω
3 to 5	2200 Ω
6 to 10	1200 Ω
11 to 16	700 Ω

Universal LCD driver for low multiplex rates

15. Test information

15.1 Quality information

This product has been qualified in accordance with the Automotive Electronics Council (AEC) standard *Q100 - Failure mechanism based stress test qualification for integrated circuits*, and is suitable for use in automotive applications.

PCA85233 **NXP Semiconductors**

Universal LCD driver for low multiplex rates

16. Bare die outline

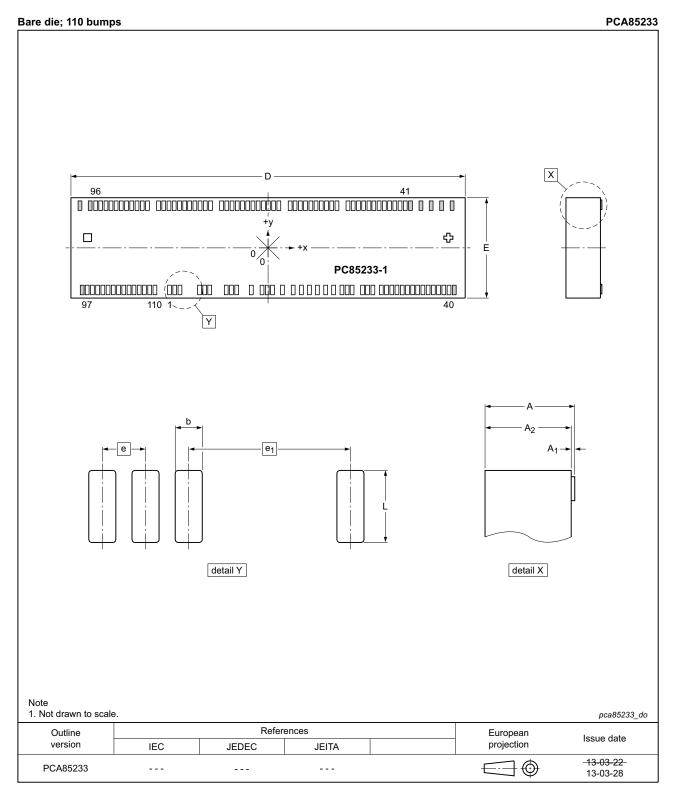


Fig 26. Bare die outline of PCA85233

Product data sheet

Universal LCD driver for low multiplex rates

Table 24. Dimensions of PCA85233U

Original dimensions are in mm.

Unit (mm)	Α	A_1	A_2	b	D	E	е	e ₁	L
max	-	0.018	-	-	-	-	-	-	-
nom	0.40	0.015	0.38	0.03	4.16	1.07	0.054	0.203	0.09
min	-	0.012	-	-	-	-	-	-	-

Table 25. Bump locations

All x/y coordinates represent the position of the center of each bump with respect to the center (x/y = 0) of the chip; see Figure 26.

Symbol	Bump	X (μm)	Υ (μm)		Description
SDAACK	1	-1022.67	-436.5	<u>[1]</u>	I ² C-bus acknowledge output
SDAACK	2	-968.67	-436.5		
SDAACK	3	-914.67	-436.5		
SDA	4	-712.17	-436.5	<u>[1]</u>	I ² C-bus serial data input
SDA	5	-658.17	-436.5		
SDA	6	-604.17	-436.5		
SCL	7	-433.17	-436.5		I ² C-bus serial clock input
SCL	8	-379.17	-436.5		
SCL	9	-325.17	-436.5		
CLK	10	-173.52	-436.5		clock input/output
V_{DD}	11	-61.47	-436.5		supply voltage
V_{DD}	12	-7.47	-436.5		
V_{DD}	13	46.53	-436.5		
SYNC	14	149.58	-436.5		cascade synchronization input/output
osc	15	262.08	-436.5		oscillator select
FF	16	345.78	-436.5		frame frequency select
A0	17	429.48	-436.5		subaddress input
A1	18	513.18	-436.5		
A2	19	596.88	-436.5		
SA0	20	680.58	-436.5		I ² C-bus slave address input; bit 0
V_{SS}	21	765.63	-436.5		ground supply voltage
V_{SS}	22	819.63	-436.5		
V_{SS}	23	873.63	-436.5		
V_{LCD}	24	979.83	-436.5		LCD supply voltage
V_{LCD}	25	1033.83	-436.5		
V_{LCD}	26	1087.83	-436.5		
BP2	27	1176.03	-436.5		LCD backplane output
BP0	28	1230.03	-436.5		
S0	29	1284.03	-436.5		LCD segment output
S1	30	1338.03	-436.5		
S2	31	1392.03	-436.5		
S3	32	1446.03	-436.5		

Universal LCD driver for low multiplex rates

Table 25. Bump locations

All x/y coordinates represent the position of the center of each bump with respect to the center (x/y = 0) of the chip; see <u>Figure 26</u>.

Symbol	Bump	X (μm)	Υ (μm)	Description
S4	33	1500.03	-436.5	LCD segment output
S5	34	1554.03	-436.5	
S6	35	1608.03	-436.5	
S7	36	1662.03	-436.5	
S8	37	1716.03	-436.5	
S9	38	1770.03	-436.5	
S10	39	1824.03	-436.5	
S11	40	1878.03	-436.5	
S12	41	1423.53	436.5	
S13	42	1369.53	436.5	
S14	43	1315.53	436.5	
S15	44	1261.53	436.5	
S16	45	1207.53	436.5	
S17	46	1153.53	436.5	
S18	47	1099.53	436.5	
S19	48	1045.53	436.5	
S20	49	991.53	436.5	
S21	50	937.53	436.5	
S22	51	883.53	436.5	
S23	52	829.53	436.5	
S24	53	714.06	436.5	
S25	54	660.06	436.5	
S26	55	606.06	436.5	
S27	56	552.06	436.5	
S28	57	498.06	436.5	
S29	58	444.06	436.5	
S30	59	390.06	436.5	
S31	60	336.06	436.5	
S32	61	282.06	436.5	
S33	62	228.06	436.5	
S34	63	112.59	436.5	
S35	64	58.59	436.5	
S36	65	4.59	436.5	
S37	66	-49.41	436.5	
S38	67	-103.41	436.5	
S39	68	-157.41	436.5	
S40	69	-211.41	436.5	

Universal LCD driver for low multiplex rates

Table 25. Bump locations

All x/y coordinates represent the position of the center of each bump with respect to the center (x/y = 0) of the chip; see <u>Figure 26</u>.

(x/y = 0) of the chip; see <u>Figure 26</u> .											
Symbol	Bump	X (μm)	Υ (μm)	Description							
S41	70	-265.41	436.5	LCD segment output							
S42	71	-319.41	436.5								
S43	72	-373.41	436.5								
S44	73	-427.41	436.5								
S45	74	-481.41	436.5								
S46	75	-596.88	436.5								
S47	76	-650.88	436.5								
S48	77	-704.88	436.5								
S49	78	-758.88	436.5								
S50	79	-812.88	436.5								
S51	80	-866.88	436.5								
S52	81	-920.88	436.5								
S53	82	-974.88	436.5								
S54	83	-1028.88	436.5								
S55	84	-1082.88	436.5								
S56	85	-1136.88	436.5								
S57	86	-1252.35	436.5								
S58	87	-1306.35	436.5								
S59	88	-1360.35	436.5								
S60	89	-1414.35	436.5								
S61	90	-1468.35	436.5								
S62	91	-1522.35	436.5								
S63	92	-1576.35	436.5								
S64	93	-1630.35	436.5								
S65	94	-1684.35	436.5								
S66	95	-1738.35	436.5								
S67	96	-1792.35	436.5								
S68	97	-1876.05	-436.5								
S69	98	-1822.05	-436.5								
S70	99	-1768.05	-436.5								
S71	100	-1714.05	-436.5								
S72	101	-1660.05	-436.5								
S73	102	-1606.05	-436.5								
S74	103	-1552.05	-436.5								
S75	104	-1498.05	-436.5								
S76	105	-1444.05	-436.5								
S77	106	-1390.05	-436.5								
S78	107	-1336.05	-436.5								
S79	108	-1282.05	-436.5								

Universal LCD driver for low multiplex rates

Table 25. Bump locations

All x/y coordinates represent the position of the center of each bump with respect to the center (x/y = 0) of the chip; see Figure 26.

Symbol	Bump	X (μm)	Υ (μm)		Description
BP3	109	-1228.05	-436.5		LCD backplane output
BP1	110	-1174.05	-436.5		
D1	-	1932.03	-436.5	[2]	dummy pad
D2	-	1909.53	436.5		
D3	-	1801.53	436.5		
D4	-	1693.53	436.5		
D5	-	1585.53	436.5		
D6	-	1477.53	436.5		
D7	-	-1846.35	436.5		dummy pad
D8	-	-1953	436.5		
D9	-	-1930.05	-436.5		

^[1] For most applications SDA and SDAACK are shorted together; see Section 8.

Table 26. Gold bump hardness

Type number	Min	Max	Unit ^[1]
PCA85233UG/2DA/Q1	60	120	HV

^[1] Pressure of diamond head: 10 g to 50 g.

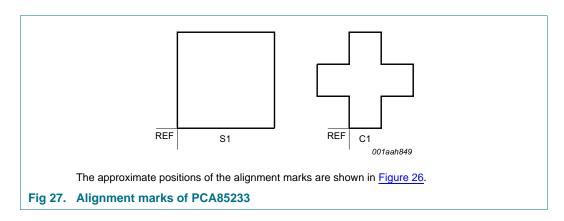


Table 27. Alignment mark locations

All x/y coordinates represent the position of the REF point (see <u>Figure 27</u>) with respect to the center (x/y = 0) of the chip; see <u>Figure 26</u>.

Symbol	Size (μm)	Χ (μm)	Υ (μm)
S1	81 × 81	-1916.1	45
C1	81 × 81	1855.8	45

^[2] The dummy pads are connected to V_{SS} but are not tested.

Universal LCD driver for low multiplex rates

17. Handling information

All input and output pins are protected against ElectroStatic Discharge (ESD) under normal handling. When handling Metal-Oxide Semiconductor (MOS) devices ensure that all normal precautions are taken as described in *JESD625-A*, *IEC 61340-5* or equivalent standards.

18. Packing information

18.1 Packing information on the tray

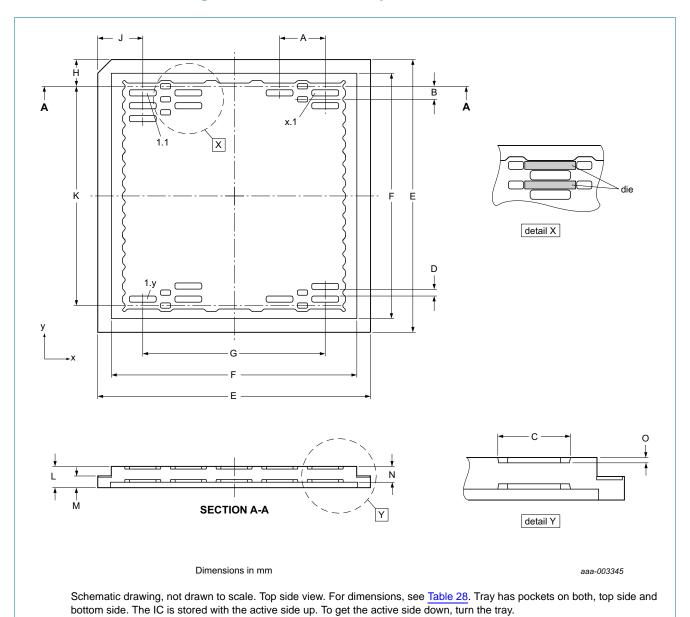


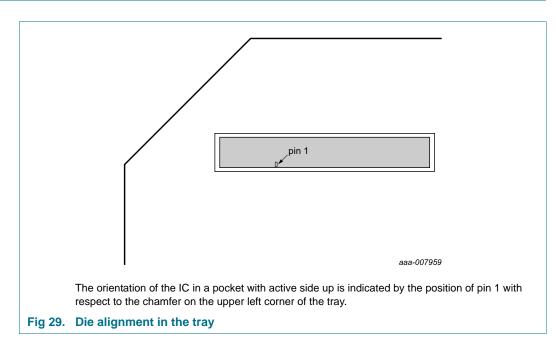
Fig 28. Tray details of PCA85233UG

Universal LCD driver for low multiplex rates

Table 28. Specification of 3 inch tray details

Tray details are shown in Figure 28. Nominal values without production tolerances.

Tray	details													
Dime	nsions													
Α	В	С	D	Е	F	G	Н	J	K	L	M	N	О	Unit
6.0	2.5	4.26	1.17	76.0	68.0	60.0	6.75	8.0	62.5	4.2	2.6	3.2	0.48	mm
Numl	ber of po	ockets												
x direction								y direction						
11								26						



Universal LCD driver for low multiplex rates

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19.1 LCD segment driver selection

Table 29. Selection of LCD segment drivers

Type name	Nun	nber (of ele	ment	s at I	MUX		V _{DD} (V) V _{LCD} (V)	V _{LCD} (V)	V _{LCD} (V)	V _{LCD} (V)	f _{fr} (Hz)	T _{amb} (°C)	Interface	Package	AEC-
	1:1	1:2	1:3	1:4	1:6	1:8	1:9			charge pump	temperature compensat.					Q10
PCF8566TS	24	48	72	96	-	-	-	2.5 to 6	2.5 to 6	N	N	69	-40 to 85	I ² C	VSO40	Ν
PCF85162T	32	64	96	128	-	-	-	1.8 to 5.5	2.5 to 6.5	N	N	82	-40 to 85	I ² C	TSSOP48	N
PCA85162T	32	64	96	128	-	-	-	1.8 to 5.5	2.5 to 8	N	N	110	-40 to 95	I ² C	TSSOP48	Υ
PCF8551ATT ⁶	36	72	108	144	-	-	-	1.8 to 5.5	1.8 to 5.5	N	N	32, 64, 96, 128 ^[1]	-40 to 85	I ² C	TSSOP48	N
PCF8551BTT ^[6]	36	72	108	144	-	-	-	1.8 to 5.5	1.8 to 5.5	N	N	32, 64, 96, 128 <mark>11</mark>	-40 to 85	SPI	TSSOP48	N
PCF85176T	40	80	120	160	-	-	-	1.8 to 5.5	2.5 to 6.5	N	N	82	-40 to 85	I ² C	TSSOP56	N
PCA85176T	40	80	120	160	-	-	-	1.8 to 5.5	2.5 to 8	N	N	110	-40 to 95	I ² C	TSSOP56	Υ
PCF85176H	40	80	120	160	-	-	-	1.8 to 5.5	2.5 to 6.5	N	N	82	-40 to 85	I ² C	TQFP64	Ν
PCA85176H	40	80	120	160	-	-	-	1.8 to 5.5	2.5 to 8	N	N	82	-40 to 95	I ² C	TQFP64	Υ
PCA8546ATT ⁵	-	-	-	176	-	-	-	1.8 to 5.5	2.5 to 9	N	N	60 to 300[1]	-40 to 95	I ² C	TSSOP56	Υ
PCA8546BTT[5]	-	-	-	176	-	-	-	1.8 to 5.5	2.5 to 9	N	N	60 to 300[1]	-40 to 95	SPI	TSSOP56	Υ
PCA8547AHT[3][5]	44	88		176	-	-	-	1.8 to 5.5	2.5 to 9	Υ	Υ	60 to 300[1]	-40 to 85	I ² C	TQFP64	Υ
PCA8547BHT[3][5]	44	88		176	-	-	-	1.8 to 5.5	2.5 to 9	Υ	Υ	60 to 300[1]	-40 to 95	SPI	TQFP64	Υ
PCF85134HL	60	120	180	240	-	-	-	1.8 to 5.5	2.5 to 6.5	N	N	82	-40 to 85	I ² C	LQFP80	Ν
PCA85134H	60	120	180	240	-	-	-	1.8 to 5.5	2.5 to 8	N	N	82	-40 to 95	I ² C	LQFP80	Υ
PCF8545ATT[5]	-	-	-	176	252	320	-	1.8 to 5.5	2.5 to 5.5	N	N	60 to 300[1]	-40 to 85	I ² C	TSSOP56	Ν
PCF8545BTT ^[5]	-	-	-	176	252	320	-	1.8 to 5.5	2.5 to 5.5	N	N	60 to 300[1]	-40 to 85	SPI	TSSOP56	N
PCF8536AT[4]	-	-	-	176	252	320	-	1.8 to 5.5	2.5 to 9	N	N	60 to 300[1]	-40 to 85	I ² C	TSSOP56	N
PCF8536BT[4]	-	-	-	176	252	320	-	1.8 to 5.5	2.5 to 9	N	N	60 to 300[1]	-40 to 85	SPI	TSSOP56	N
PCA8536AT[4]	-	-	-	176	252	320	-	1.8 to 5.5	2.5 to 9	N	N	60 to 300[1]	-40 to 95	I ² C	TSSOP56	Υ
PCA8536BT[4]	-	-	-	176	252	320	-	1.8 to 5.5	2.5 to 9	N	N	60 to 300[1]	-40 to 95	SPI	TSSOP56	Υ
PCF8537AH3	44	88	-	176	276	352	-	1.8 to 5.5	2.5 to 9	Υ	Υ	60 to 300[1]	-40 to 85	I ² C	TQFP64	N
PCF8537BH3	44	88	-	176	276	352	-	1.8 to 5.5	2.5 to 9	Υ	Υ	60 to 300[1]	-40 to 85	SPI	TQFP64	Ν
PCA8537AH[3]	44	88	-	176	276	352	-	1.8 to 5.5	2.5 to 9	Υ	Υ	60 to 300[1]	-40 to 95	I ² C	TQFP64	Υ

Product data sheet

Table 29.	Selection	of LCD segment dri	verscontinued

Type name	Number of elements at MUX						V _{DD} (V)	V _{LCD} (V)	V _{LCD} (V)	V _{LCD} (V)	f _{fr} (Hz)	T _{amb} (°C)	Interface	Package	AEC-	
	1:1	1:2	1:3	1:4	1:6	1:8	1:9		charge pump	_	temperature compensat.					Q100
PCA8537BH[3]	44	88	-	176	276	352	-	1.8 to 5.5	2.5 to 9	Υ	Υ	60 to 300[1]	-40 to 95	SPI	TQFP64	Υ
PCA9620H[3]	60	120	-	240	320	480	-	2.5 to 5.5	2.5 to 9	Υ	Υ	60 to 300[1]	-40 to 105	I ² C	LQFP80	Υ
PCA9620U[3]	60	120	-	240	320	480	-	2.5 to 5.5	2.5 to 9	Υ	Υ	60 to 300[1]	-40 to 105	I ² C	bare die	Υ
PCF8552DUG[6]	36	72	108	144	-	-	-	1.8 to 5.5	1.8 to 5.5	N	N	32, 64, 96, 128 <mark>1</mark> 1	-40 to 85	I ² C	bare die	N
PCF8576DU	40	80	120	160	-	-	-	1.8 to 5.5	2.5 to 6.5	N	N	77	-40 to 85	I ² C	bare die	N
PCF8576EUG	40	80	120	160	-	-	-	1.8 to 5.5	2.5 to 6.5	N	N	77	-40 to 85	I ² C	bare die	N
PCA8576FUG ^[5]	40	80	120	160	-	-	-	1.8 to 5.5	2.5 to 8	N	N	200	-40 to 105	I ² C	bare die	Υ
PCF85133U	80	160	240	320	-	-	-	1.8 to 5.5	2.5 to 6.5	N	N	82, 110[2]	-40 to 85	I ² C	bare die	N
PCA85133U	80	160	240	320	-	-	-	1.8 to 5.5	2.5 to 8	N	N	82, 110[2]	-40 to 95	I ² C	bare die	Υ
PCA85233U	80	160	240	320	-	-	-	1.8 to 5.5	2.5 to 8	N	N	150,220[2]	-40 to 105	I ² C	bare die	Υ
PCF85132U	160	320	480	640	-	-	-	1.8 to 5.5	1.8 to 8	N	N	60 to 90[1]	-40 to 85	I ² C	bare die	N
PCA85132U	160	320	480	640	-	-	-	1.8 to 5.5	1.8 to 8	N	N	60 to 90[1]	-40 to 95	I ² C	bare die	Υ
PCA85232U	160	320	480	640	-	-	-	1.8 to 5.5	1.8 to 8	N	N	117 to 176[1]	-40 to 95	I ² C	bare die	Υ
PCF8538UG[5][3]	102	204	-	408	612	816	918	2.5 to 5.5	4 to 12	Υ	Υ	45 to 300[1]	-40 to 85	I ² C, SPI[2]	bare die	N
PCA8538UG[3]	102	204	-	408	612	816	918	2.5 to 5.5	4 to 12	Υ	Υ	45 to 300[1]	-40 to 105	I ² C, SPI ²	bare die	Υ

^[1] Can be selected by command.

^[2] Can be selected by pin configuration.

^[3] Extra feature: Temperature sensor.

^[4] Extra feature: 6 PWM channels.

^[5] In development. Expected product release end of 2013.

^[6] In development. Expected product release mid of 2014.

Universal LCD driver for low multiplex rates

20. Abbreviations

Table 30. Abbreviations

Acronym	Description					
CMOS	Complementary Metal-Oxide Semiconductor					
COG	Chip-On-Glass					
DC	Direct Current					
НВМ	Human Body Model					
I ² C	Inter-Integrated Circuit					
IC	Integrated Circuit					
ITO	Indium Tin Oxide					
LCD	Liquid Crystal Display					
MM	Machine Model					
RAM	Random Access Memory					
RC	Resistance-Capacitance					
RMS	Root Mean Square					

21. References

- [1] AN10170 Design guidelines for COG modules with NXP monochrome LCD drivers
- [2] AN10706 Handling bare die
- [3] AN10853 ESD and EMC sensitivity of IC
- [4] AN11267 EMC and system level ESD design guidelines for LCD drivers
- [5] IEC 60134 Rating systems for electronic tubes and valves and analogous semiconductor devices
- [6] IEC 61340-5 Protection of electronic devices from electrostatic phenomena
- [7] JESD22-A114 Electrostatic Discharge (ESD) Sensitivity Testing Human Body Model (HBM)
- [8] JESD22-A115 Electrostatic Discharge (ESD) Sensitivity Testing Machine Model (MM)
- [9] JESD78 IC Latch-Up Test
- [10] JESD625-A Requirements for Handling Electrostatic-Discharge-Sensitive (ESDS) Devices
- [11] UM10204 I²C-bus specification and user manual
- [12] UM10569 Store and transport requirements

Universal LCD driver for low multiplex rates

22. Revision history

Table 31. Revision history

Document ID	Release date	Data sheet status	Change notice	Supersedes
PCA85233 v.1	20130917	Product data sheet	-	-

Universal LCD driver for low multiplex rates

23. Legal information

23.1 Data sheet status

Document status[1][2]	Product status[3]	Definition
Objective [short] data sheet	Development	This document contains data from the objective specification for product development.
Preliminary [short] data sheet	Qualification	This document contains data from the preliminary specification.
Product [short] data sheet	Production	This document contains the product specification.

- [1] Please consult the most recently issued document before initiating or completing a design.
- [2] The term 'short data sheet' is explained in section "Definitions"
- [3] The product status of device(s) described in this document may have changed since this document was published and may differ in case of multiple devices. The latest product status information is available on the Internet at URL http://www.nxp.com.

23.2 Definitions

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Short data sheet — A short data sheet is an extract from a full data sheet with the same product type number(s) and title. A short data sheet is intended for quick reference only and should not be relied upon to contain detailed and full information. For detailed and full information see the relevant full data sheet, which is available on request via the local NXP Semiconductors sales office. In case of any inconsistency or conflict with the short data sheet, the full data sheet shall prevail.

Product specification — The information and data provided in a Product data sheet shall define the specification of the product as agreed between NXP Semiconductors and its customer, unless NXP Semiconductors and customer have explicitly agreed otherwise in writing. In no event however, shall an agreement be valid in which the NXP Semiconductors product is deemed to offer functions and qualities beyond those described in the Product data sheet.

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