

Synchronous Buck-Boost DC/DC Converter with 2 A Switches ($V_{IN} = 2.0 \text{ V}$ to 5.5 V, 1ch)

BD83070GWL

General Description

The BD83070GWL is a synchronous buck-boost DC/DC convertor providing 3.3 V or 2.5 V output from single-cell Li-ion battery or other input between 2.0 V and 5.5 V. It has the capability to support up to 1 A output over input voltage range of 2.7 V to 5.5 V. It seamlessly changes between buck and boost operations depending on the input voltage.

It is based on pulse width modulation (PWM) and provides high efficiency for heavy load. While in PWM operation, internal FETs switch at fixed frequency 1.5 MHz typical. It automatically changes over control system to hysteresis pulse frequency modulation (PFM) to suppress switching loss and current consumption during light load. Battery drain fall down to only 2.8 µA typical at no load current. It is possible to disable auto-PFM/PWM mode by the MODE pin for suppressing output ripple and fixed frequency switching.

The device is packaged in a 1.2 mm x 1.6 mm WLCSP package.

Features

- Synchronous Buck-Boost DC/DC Converter
- Automatic PFM/PWM Transition
- Output Current: Up To 1 A ($V_{IN} > 2.7 \text{ V}$, $V_{OUT} = 3.3 \text{ V}$)
- Selectable Output Voltage: 2.5 V or 3.3 V
- Efficiency: Up To 95 %
- UVLO Detection: 1.61 V(Max)
- Built-in Thermal, Over Voltage, And Over Current Protection

Key Specifications

Input Voltage Range:
 Output Voltage:
 Output Current:
 Switching Frequency:
 Quiescent VIN Current:
 Operating Temperature Range:
 2.0 V to 5.5 V
 2.5 V or 3.3 V
 1 A(Max)
 SMHz(Typ)
 2.8 µA(Typ)
 -40 °C to +85 °C

Package

W(Typ) x **D(Typ)** x **H(Max)** 1.20 mm × 1.60 mm × 0.57 mm



Applications

- Single Cell Li-ion or 3 Cell NiMH Battery-Powered Portable Products
- Tablet Terminal Device
- Smartphone

Typical Application Circuit

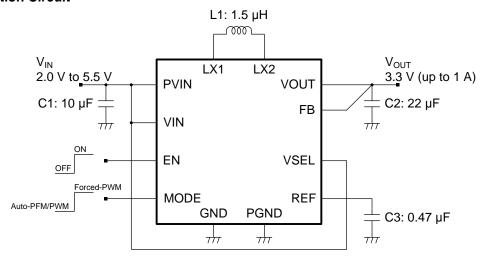
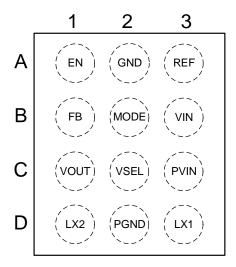


Figure 1. Typical Application Circuit

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Pin Configuration



Top View

Figure 2. Pin Configuration

Pin Descriptions

Description	113	
Pin No.	Pin Name	Function
A1	EN	Enable pin of the DC/DC converter. Inputting high (≥ 1.2 V) the EN pin turns on the regulator. Inputting low (≤ 0.4 V) or open the EN pin turns off the regulator.
A2	GND	Ground for sensing
А3	REF	Linear regulator output for power supply of internal circuits. Connect ceramic capacitor (0.47 µF) between this pin and the GND pin for output stability. Do not connect the other devices.
B1	FB	Voltage feedback pin. Connect this pin to the VOUT pin.
B2	MODE	Mode selection pin. Low (≤ 0.4 V): Auto-PFM/PWM mode. High (≥ 1.2 V): Forced-PWM mode. Do not leave this pin floating.
В3	VIN	Power supply input of controller. Connect this pin to the PVIN pin.
C1	VOUT	Output pin of the DC/DC converter. Connect ceramic capacitor (22 µF recommended) between this pin and the PGND pin for output stability.
C2	VSEL	Output voltage selection pin. High (V _{IN}): 3.3 V. Low (GND): 2.5 V. Connect this pin to either the VIN pin or the GND pin.
СЗ	PVIN	Power supply input of the DC/DC converter and LX1 side gate drivers. Connect ceramic capacitor (≥ 10 µF) between this pin and the PGND pin for power supply noise reduction.
D1	LX2	Inductor connection pin. Connect inductor (1.5 μH) between this pin and the LX1 pin.
D2	PGND	Ground of power FET, discharge, and gate drivers.
D3	LX1	Inductor connection pin. Connect inductor (1.5 µH) between this pin and the LX2 pin.

Block Diagram

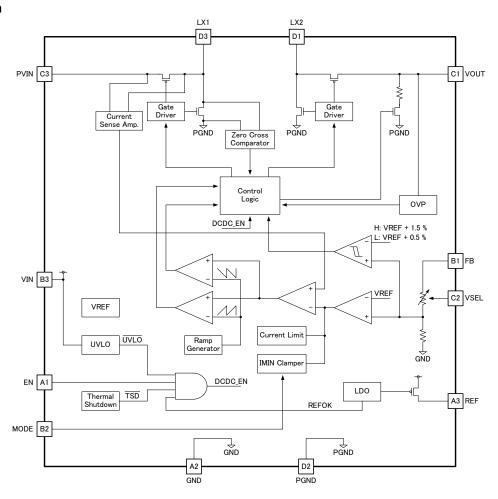


Figure 3. Block Diagram

Absolute Maximum Ratings (Ta=25 °C)

Parameter	Symbol	Ratings	Unit
Voltage Range in Pins: VIN, PVIN, VOUT, FB, EN, MODE	V _{MAXVIN} , V _{MAXPVIN} , V _{MAXVOUT} , V _{MAXFB} , V _{MAXEN} , V _{MAXMODE}	-0.3 to +6.0	V
Voltage Range in Pins: LX1, LX2 (Note 1)	V _{MAXLX1} , V _{MAXLX2}	-1.0 to +7.0	V
Voltage Range in Pin: REF	V _{MAXREF}	-0.3 to +2.1	V
Voltage Range in Pin: PGND	V _{MAXPGND}	-0.3 to +0.3	V
Maximum Junction Temperature	Tjmax	150	°C
Storage Temperature Range	Tstg	-55 to +125	°C

⁽Note 1) Voltage transients on the LX1 or the LX2 pins beyond the DC limits specified in the absolute maximum ratings are non-disruptive to normal operation when using good layout practices as described elsewhere in the data sheet and application notes and as seen on the product demo board.

Caution 1: Operating the IC over the absolute maximum ratings may damage the IC. The damage can either be a short circuit between pins or an open circuit between pins and the internal circuitry. Therefore, it is important to consider circuit protection measures, such as adding a fuse, in case the IC is operated over the absolute maximum ratings.
 Caution 2: Should by any chance the maximum junction temperature rating be exceeded the rise in temperature of the chip may result in deterioration of the

Caution 2: Should by any chance the maximum junction temperature rating be exceeded the rise in temperature of the chip may result in deterioration of the properties of the chip. In case of exceeding this absolute maximum rating, design a PCB with thermal resistance taken into consideration by increasing board size and copper area so as not to exceed the maximum junction temperature rating.

Thermal Resistance (Note 2)

Parameter	Symbol	Thermal Res	Unit		
Farameter		1s ^(Note 3)	2s2p ^(Note 3)	Offic	
UCSP50L1C					
Junction to Ambient	θ_{JA}	-	186.6	°C/W	

(Note 2) Based on JESD51-2A(Still-Air). (Note 3) Using a PCB board based on JESD51-9.

Layer Number of Measurement Board	Material	Board Size
Single	FR-4	114.5 mm x 101.5 mm x 1.6 mmt

Тор	
Copper Pattern	Thickness
Footprints and Traces	70 µm

Layer Number of Measurement Board	Material	Board Size
4 Layers	FR-4	114.5 mm x 101.5 mm x 1.6 mmt

Тор		2 Internal Layers Botton			
Copper Pattern	Thickness	Copper Pattern Thickness		Copper Pattern	Thickness
Footprints and Traces	70 µm	99.5 mm x 99.5 mm	35 µm	99.5 mm x 99.5 mm	70 µm

Recommended Operating Conditions

commende operating commission					
Parameter	Symbol	Min	Тур	Max	Unit
Power Supply Voltage	V _{IN}	2.0	3.6	5.5	V
Operating Temperature	Topr	-40	+25	+85	°C
REF Connection Capacitor (Note 1)	C_REF	0.22	0.47	1.00	μF

⁽Note 1) The minimum value capacitance must be met this specification over full operating condition. Ceramic capacitors are recommended for input/output

Electrical Characteristics (Unless otherwise specified VIN=PVIN=EN=VSEL=3.6 V, C_{REF}=0.47 μF, Ta=25 °C)

Parameter	Symbol	Limit			Unit	Conditions
i arameter	Gyllibol	Min	Тур	Max	Offic	Conditions
DC/DC Converter						
Switching Frequency during PWM	f _{SW}	1.35	1.50	1.65	MHz	MODE=V _{IN}
Maximum Duty	D _{MAX}	80	87	95	%	MODE=V _{IN}
LX1 High Side FET ON Resistance	R _{ON1H}	-	50	-	mΩ	
LX1 Low Side FET ON Resistance	R _{ON1L}	-	60	-	mΩ	
LX2 High Side FET ON Resistance	R _{ON2H}	-	55	-	mΩ	V _{OUT} =3.3 V
LX2 Low Side FET ON Resistance	R _{ON2L}	-	65	-	mΩ	V _{OUT} =3.3 V
Over Current Protection	I _{OCP}	2.0	-	-	Α	PVIN=3.6 V
Output Voltage 1	V _{OUT1}	3.267	3.300	3.333	V	MODE=V _{IN} , VSEL=V _{IN} , No Load
Output Voltage 2	V _{OUT2}	2.468	2.500	2.532	V	MODE=V _{IN} , VSEL=0 V, No Load
Load Regulation	V_{LR}	-	0.5	-	mV	MODE=V _{IN} , VSEL=V _{IN} I _{OUT} =0 mA to 1000 mA

Electrical Characteristics - continued (Unless otherwise specified VIN=PVIN=EN=VSEL=3.6 V, C_{REF} =0.47 μ F, Ta=25 °C)

Parameter		Symbol	Limit		Unit	Conditions	
		- - - - - - - - - -	Min	Тур	Max		Containe
DC/DC Converter - c	ontinued						
Over Voltage	Detect	V _{OVPDET}	5.3	5.5	5.7	V	V _{OUT} voltage increasing
Threshold	Release	V _{OVPRST}	5.2	5.4	5.6	V	V _{OUT} voltage decreasing
Startup Delay Time		t _{ST}	-	4.9	-	ms	From EN=High to V _{OUT} =100 mV
Startup Slew Rate		SR _{ST}	2.5	5.0	10.0	mV/μs	
Discharge Resistance	•	R _{DCG}	40	85	200	Ω	EN=0 V
Short Circuit	Detect	V _{SCPDET}	1.3	1.4	1.5	V	FB voltage decreasing
Threshold	Release	V _{SCPRST}	1.51	1.61	1.71	V	FB voltage increasing
Main Controller							
Under Voltage	Up	V _{UVLOUP}	-	1.740	1.990	V	V _{IN} voltage increasing
Lockout Threshold	Down	V _{UVLODN}	1.51	1.56	1.61	V	V _{IN} voltage decreasing
EN Pin Control	ON	V _{ENH}	1.2	-	5.5	V	
Voltage	OFF	V _{ENL}	-0.3	-	+0.4	V	
EN Pin Input Current		I _{EN}	-	200	500	nA	
MODE Pin Control	High	V _{MODEH}	1.2	-	5.5	V	
Voltage	Low	V _{MODEL}	-0.3	-	+0.4	V	
VSEL Pin Control	High	V _{VSELH}	V _{IN} -0.3	-	V _{IN} +0.3	V	
Voltage	Low	V _{VSELL}	-0.3	-	+0.3	V	
REF Output Voltage		V_{REF}	1.45	1.50	1.55	V	I _{REF} =-100 μA
Whole Device							
Quiescent VIN Currer	nt	I _{VIN}	-	2.8	5.6	μΑ	MODE=0 V, FB=3.5 V
Quiescent FB Curren	t	I _{FB}	-	0.2	0.4	μΑ	MODE=0 V, FB=3.5 V
Shutdown VIN Currer	nt	I _{SHD}	-	0.1	1.0	μΑ	EN=0 V
			1		1	1	1

Detailed Descriptions

1. Startup and Shutdown Control

When the EN pin goes from low (input under 0.4 V or open) to high (input over 1.2 V), the BD83070GWL turns on internal LDO, REF, and the DC/DC converter. There is typically 4.9 ms delay from the EN high edge to startup of DC/DC converter (t_{ST}). It has a soft-start structure and ramps up the output voltage in 5 mV/ μ s typically (SR_{ST}). On the other hand, when the EN pin goes from high to low, it disables the internal LDO and DC/DC immediately. While in shutdown, it turns discharge switch on to pull V_{OUT} to ground through 85 Ω typically. If the EN pin goes again from low to high during discharge sequence, the output voltage is ramped up from remaining voltage to target voltage in 5 mV/ μ s (Typ).

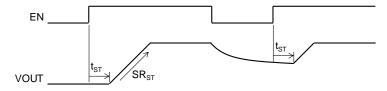


Figure 4. Startup Sequence

2. MODE Pin

In the case of the MODE pin is pulled high (over 1.2 V), the BD83070GWL operates in forced PWM mode and uses fixed frequency 1.5 MHz regardless its loads. If the MODE pin is pulled low (under 0.4 V), it operates in automatic PFM-PWM mode and automatically changes over form PWM to hysteresis PFM operation depending on its loads. Do not leave this pin floating because it is neither pulled down nor up, internally.

3. Output Voltage Setting

The BD83070GWL has internal feedback resistors. It is possible to select target output voltage from either 2.5 V or 3.3 V by the VSEL pin. If the VSEL pin is connected to ground, the nominal output voltage is 2.5 V. On the other hand when the VSEL is connected to V_{IN} , the nominal output voltage is 3.3 V. It is not recommended to change while in EN is logic high.

4. Maximum Load Current

The maximum load current varies depending on PVIN voltage and output voltage setting. When using the recommended application, the maximum load current becomes as follows.

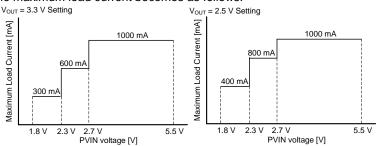


Figure 5. Maximum Load Current

5. Current Limit Protection

The BD83070GWL has a current limit protection circuit to prevent excessive electric stress on itself and external inductor at overload condition.

6. Short Circuit Protection

If FB voltage drops less than 1.4 V(Typ), the current limit value is reduced to about half of the normal that. The current limit value returns to the normal that when the FB voltage exceeds 1.61 V(Typ).

7. Over Voltage Protection

The BD83070GWL has an over voltage comparator. When the FB pin becomes open, the output voltage rises beyond target voltage. If the VOUT pin reaches 5.5 V(Typ), it stops switching to prevent over voltage stress on its power FETs. If the VOUT pin voltage falls lower than 5.4 V(Typ), it restarts switching.

8. Under Voltage Lockout (UVLO)

The BD83070GWL has a UVLO comparator to turn the device off and prevent malfunction when the input voltage is too low. As same as UVLO, it has a REFOK comparator to monitor REF voltage, internal LDO output, and turns the device off when the REF voltage is too low.

9. Thermal Shutdown

The BD83070GWL has a Thermal Shutdown Circuit (TSD Circuit). When the temperature of its chip is higher than 175 °C typical, the TSD circuit turns off the DC/DC converter. There is the hysteresis width of 20 °C between the detection point and release point to prevent malfunctions from temperature fluctuations. Because TSD Circuit is only designed for protecting the device from thermal over load, it is not recommended to design the application as TSD working in normal condition.

Typical Performance Curves

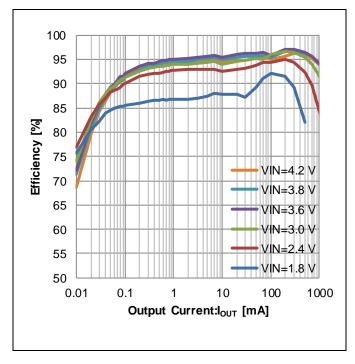


Figure 6. Efficiency vs Output Current (VSEL=High, MODE=Low: Auto-PFM/PWM)

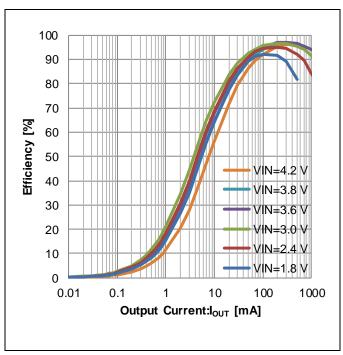


Figure 7. Efficiency vs Output Current (VSEL=High, MODE=High: Forced-PWM)

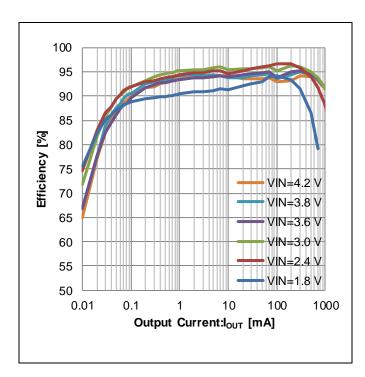


Figure 8. Efficiency vs Output Current (VSEL=Low, MODE=Low: Auto-PFM/PWM)

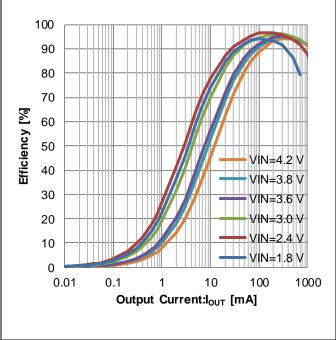


Figure 9. Efficiency vs Output Current (VSEL=Low, MODE=High: Forced-PWM)

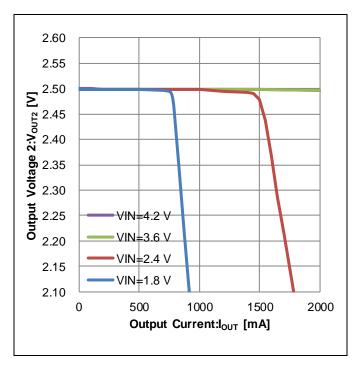


Figure 10. Output Voltage 2 vs Output Current ("Load Regulation", VSEL=Low, MODE=High: Forced-PWM)

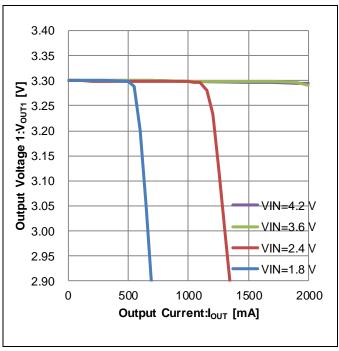


Figure 11. Output Voltage 1 vs Output Current ("Load Regulation", VSEL=High, MODE=High: Forced-PWM)

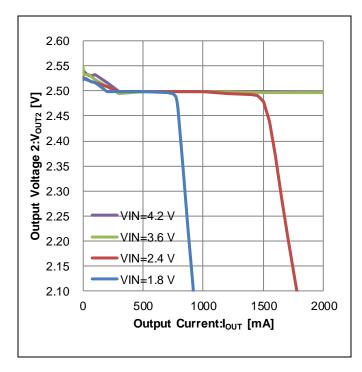


Figure 12. Output Voltage 2 vs Output Current ("Load Regulation", VSEL=Low, MODE=Low: Auto-PFM/PWM)

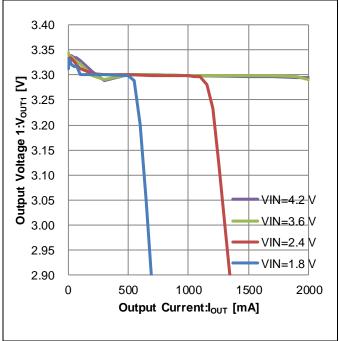


Figure 13. Output Voltage 1 vs Output Current ("Load Regulation", VSEL=High, MODE=Low: Auto-PFM/PWM)

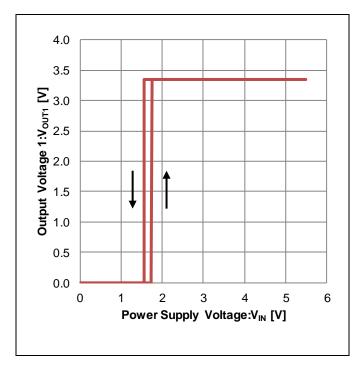


Figure 14. Output Voltage 1 vs Power Supply Voltage ("Line Regulation", EN=VSEL=High, MODE=Low: Auto-PFM/PWM, 3.3 kΩ resistive load)

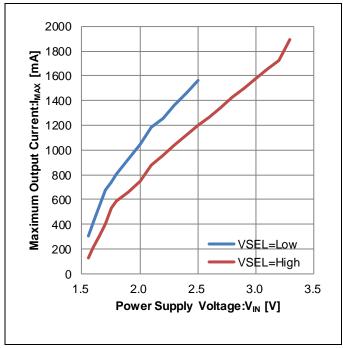


Figure 15. Maximum Output Current vs Power Supply Voltage

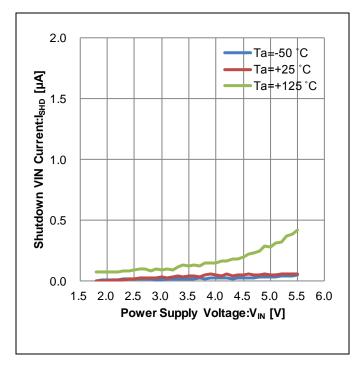


Figure 16. Shutdown VIN Current vs Power Supply Voltage (EN=MODE=Low, No load)

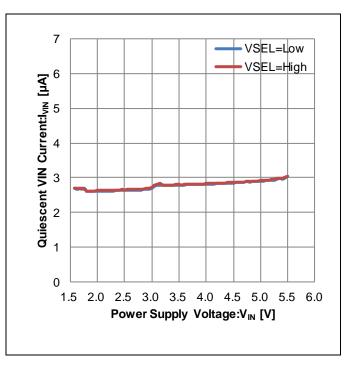


Figure 17. Quiescent VIN Current vs Power Supply Voltage (MODE=Low: Auto-PFM/PWM, FB=3.5 V, No load)

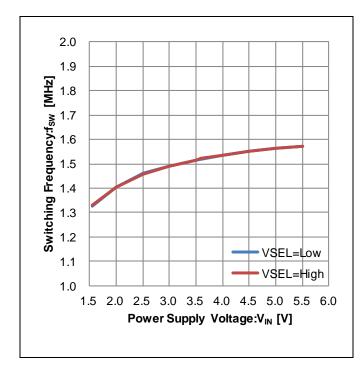


Figure 18. Switching Frequency vs Power Supply Voltage (MODE=High: Forced-PWM, No load)

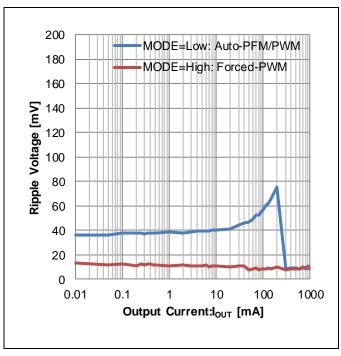


Figure 19. Ripple Voltage vs Output Current (VIN=3.6 V, VSEL=High)

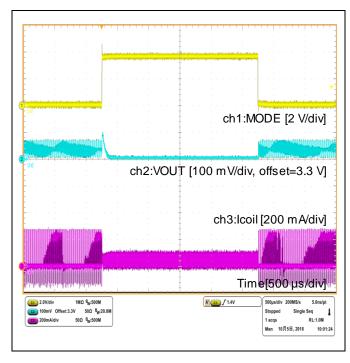


Figure 20. Transient Response ("Mode Change", VIN=3.6 V, VSEL=High, MODE=Low<->High, Output current 50 mA)

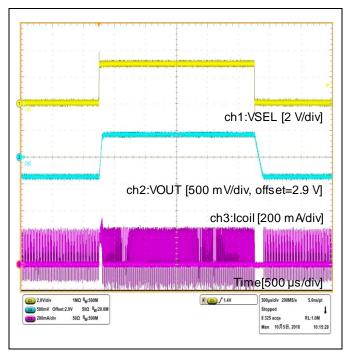


Figure 21. Transient Response
("Output Voltage Change", VIN=2.9 V, VSEL=Low<->High,
MODE=Low: Auto-PFM/PWM, Output current 50 mA)

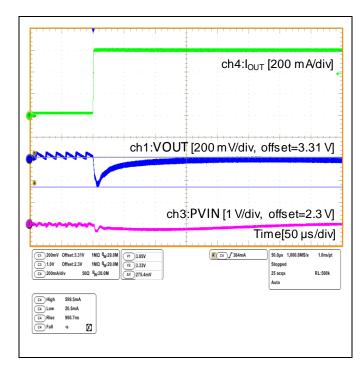


Figure 22. Transient Response (VIN=2.3 V, VSEL=High, MODE=Low: Auto-PFM/PWM, Output current 20 mA->600 mA)

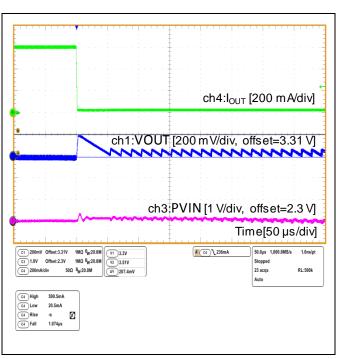


Figure 23. Transient Response (VIN=2.3 V, VSEL=High, MODE=Low: Auto-PFM/PWM, Output current 600 mA->20 mA)

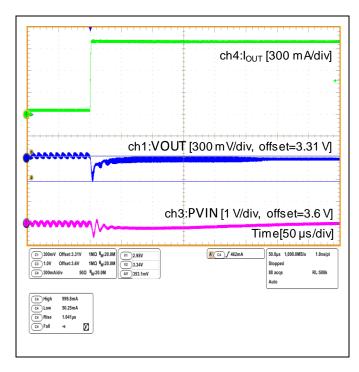


Figure 24. Transient Response (VIN=3.6 V, VSEL=High, MODE=Low: Auto-PFM/PWM, Output current 50 mA->1000 mA)

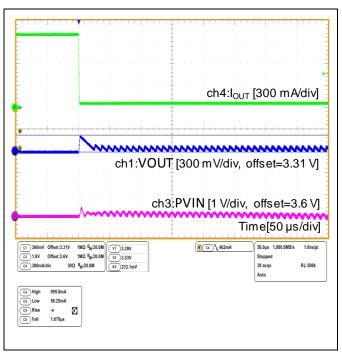
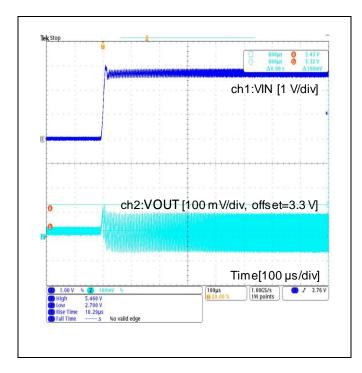


Figure 25. Transient Response (VIN=3.6 V, VSEL=High, MODE=Low: Auto-PFM/PWM, Output current 1000 mA->50 mA)

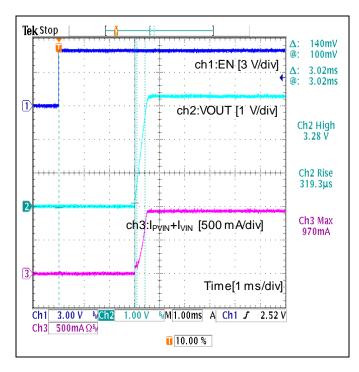


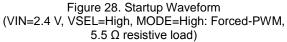
Time[100 μs/div]

| Ch2:VOUT [100 mV/div, offset=3.3 V]
| Ch2:VOUT [100 mV/div, offset=3.3 V]
| Ch2:VOUT [100 mV/div, offset=3.5 V]
| Ch2:VOUT [100 mV/div, offset=3.6 V]
| Ch2:VOUT [100 mV/div, offset=3.76 V]
| C

Figure 26. Transient Response (VIN=2.7 V->5.5 V, VSEL=High, MODE=Low: Auto-PFM/PWM, Output current 300 mA)

Figure 27. Transient Response (VIN=5.5 V->2.7 V, VSEL=High, MODE=Low: Auto-PFM/PWM, Output current 300 mA)





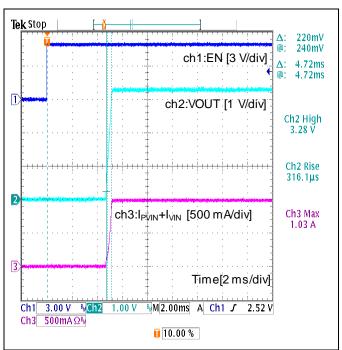
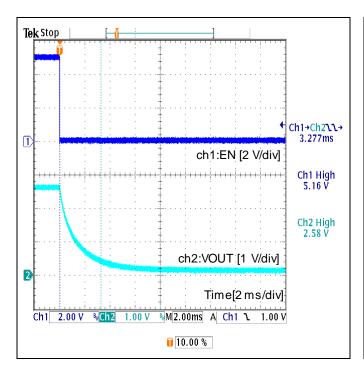


Figure 29. Startup Waveform (VIN=3.6 V, VSEL=High, MODE=High: Forced-PWM, 3.3 Ω resistive load)



Ch1+Ch2\lambda+3.262ms

ch1:EN [2 V/div]
Ch1 High
5.00 V

ch2:VOUT [1 V/div]
Time[2 ms/div]
Ch1 2.00 V & Ch2 1.00 V M 2.00ms A Ch1 \lambda 1.00 V

Figure 30. Shutdown Waveform (VIN=3.6 V, VSEL=Low, MODE=Low: Auto-PFM/PWM, No load)

Figure 31. Shutdown Waveform (VIN=3.6 V, VSEL=High, MODE=Low: Auto-PFM/PWM, No load)

Application Examples

VSEL = V_{IN} (V_{OUT} = 3.3 V setting)

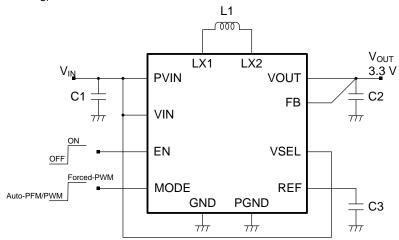


Figure 32. 3.3V Output Application Circuit

VSEL = GND (V_{OUT} = 2.5 V setting)

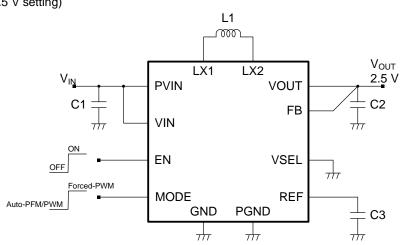
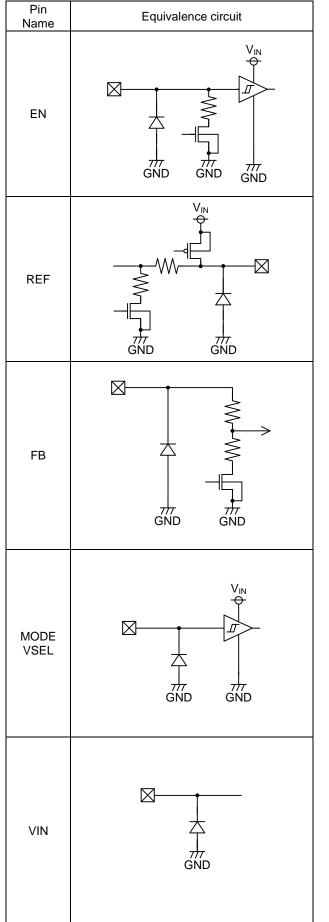


Figure 33. 2.5 V Output Application Circuit

Parts Number	Description	Supplier
L1	1239AS-H-1R5M (1.5 μH, 2.5 mm x 2.0 mm x 1.2 mm)	muRata
C1	EMK212ABJ106KD (10 μF, 16 V, X5R, 0805)	Taiyo Yuden
C2 ^(Note 1)	JMK107BBJ226MA (22 μF, 6.3 V, X5R, 0603)	Taiyo Yuden
C3	EMK105ABJ474KV-F (0.47 μF, 16 V, X5R, 0402)	Taiyo Yuden

(Note 1) The effective load capacitance value considering accuracy, temperature characteristic and DC bias characteristic of output capacitors should not be less than 22 µF. The amount of output capacitance will have a significant effect on the output ripple voltage.

I/O Equivalence Circuits



Pin	
Name	Equivalence circuit
VOUT	LX2 R _{DCG} RDCG RDCG RDCG
PVIN	PGND
LX2	V _{OUT} N
PGND	PVIN V _{OUT}
LX1	PVIN

Operational Notes

1. Reverse Connection of Power Supply

Connecting the power supply in reverse polarity can damage the IC. Take precautions against reverse polarity when connecting the power supply, such as mounting an external diode between the power supply and the IC's power supply pins.

2. Power Supply Lines

Design the PCB layout pattern to provide low impedance supply lines. Furthermore, connect a capacitor to ground at all power supply pins. Consider the effect of temperature and aging on the capacitance value when using electrolytic capacitors.

3. Ground Voltage

Ensure that no pins are at a voltage below that of the ground pin at any time, even during transient condition.

4. Ground Wiring Pattern

When using both small-signal and large-current ground traces, the two ground traces should be routed separately but connected to a single ground at the reference point of the application board to avoid fluctuations in the small-signal ground caused by large currents. Also ensure that the ground traces of external components do not cause variations on the ground voltage. The ground lines must be as short and thick as possible to reduce line impedance.

5. Recommended Operating Conditions

The function and operation of the IC are guaranteed within the range specified by the recommended operating conditions. The characteristic values are guaranteed only under the conditions of each item specified by the electrical characteristics.

6. Inrush Current

When power is first supplied to the IC, it is possible that the internal logic may be unstable and inrush current may flow instantaneously due to the internal powering sequence and delays, especially if the IC has more than one power supply. Therefore, give special consideration to power coupling capacitance, power wiring, width of ground wiring, and routing of connections.

7. Testing on Application Boards

When testing the IC on an application board, connecting a capacitor directly to a low-impedance output pin may subject the IC to stress. Always discharge capacitors completely after each process or step. The IC's power supply should always be turned off completely before connecting or removing it from the test setup during the inspection process. To prevent damage from static discharge, ground the IC during assembly and use similar precautions during transport and storage.

8. Inter-pin Short and Mounting Errors

Ensure that the direction and position are correct when mounting the IC on the PCB. Incorrect mounting may result in damaging the IC. Avoid nearby pins being shorted to each other especially to ground, power supply and output pin. Inter-pin shorts could be due to many reasons such as metal particles, water droplets (in very humid environment) and unintentional solder bridge deposited in between pins during assembly to name a few.

9. Unused Input Pins

Input pins of an IC are often connected to the gate of a MOS transistor. The gate has extremely high impedance and extremely low capacitance. If left unconnected, the electric field from the outside can easily charge it. The small charge acquired in this way is enough to produce a significant effect on the conduction through the transistor and cause unexpected operation of the IC. So unless otherwise specified, unused input pins should be connected to the power supply or ground line.

Operational Notes - continued

10. Regarding the Input Pin of the IC

This monolithic IC contains P+ isolation and P substrate layers between adjacent elements in order to keep them isolated. P-N junctions are formed at the intersection of the P layers with the N layers of other elements, creating a parasitic diode or transistor. For example (refer to figure below):

When GND > Pin A and GND > Pin B, the P-N junction operates as a parasitic diode. When GND > Pin B, the P-N junction operates as a parasitic transistor.

Parasitic diodes inevitably occur in the structure of the IC. The operation of parasitic diodes can result in mutual interference among circuits, operational faults, or physical damage. Therefore, conditions that cause these diodes to operate, such as applying a voltage lower than the GND voltage to an input pin (and thus to the P substrate) should be avoided.

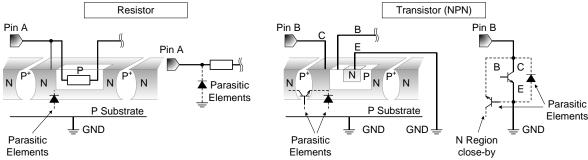


Figure 34. Example of Monolithic IC Structure

11. Ceramic Capacitor

When using a ceramic capacitor, determine a capacitance value considering the change of capacitance with temperature and the decrease in nominal capacitance due to DC bias and others.

12. Thermal Shutdown Circuit (TSD)

This IC has a built-in thermal shutdown circuit that prevents heat damage to the IC. Normal operation should always be within the IC's maximum junction temperature rating. If however the rating is exceeded for a continued period, the junction temperature (Tj) will rise which will activate the TSD circuit that will turn OFF power output pins. When the Tj falls below the TSD threshold, the circuits are automatically restored to normal operation.

Note that the TSD circuit operates in a situation that exceeds the absolute maximum ratings and therefore, under no circumstances, should the TSD circuit be used in a set design or for any purpose other than protecting the IC from heat damage.

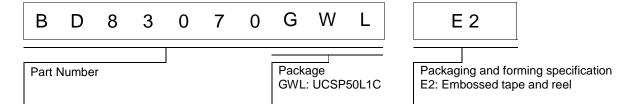
13. Over Current Protection Circuit (OCP)

This IC incorporates an integrated overcurrent protection circuit that is activated when the load is shorted. This protection circuit is effective in preventing damage due to sudden and unexpected incidents. However, the IC should not be used in applications characterized by continuous operation or transitioning of the protection circuit.

14. Disturbance Light

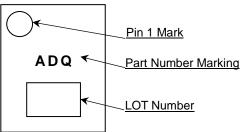
In a device where a portion of silicon is exposed to light such as in a WL-CSP and chip products, IC characteristics may be affected due to photoelectric effect. For this reason, it is recommended to come up with countermeasures that will prevent the chip from being exposed to light.

Ordering Information

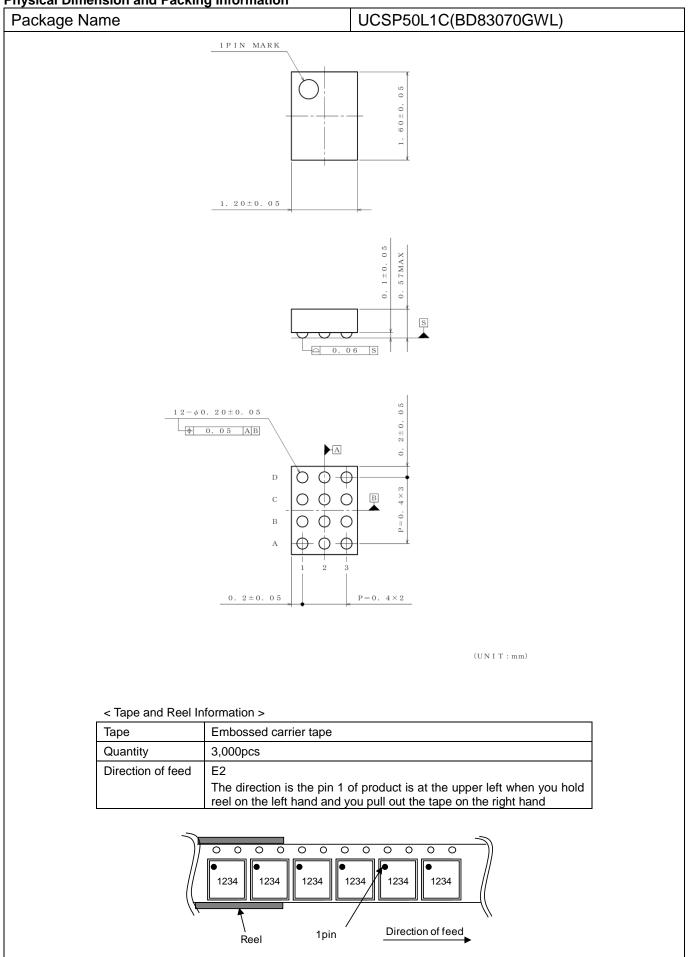


Marking Diagram

TOP VIEW UCSP50L1C (BD83070GWL)



Physical Dimension and Packing Information



Revision History

Date	Revision	Changes
09.Oct.2018	001	New Release

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JAPAN	USA	EU	CHINA
CLASSⅢ	CLASSⅢ	CLASS II b	СГУССШ
CLASSIV		CLASSⅢ	CLASSIII

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 - [g] Use of our Products without cleaning residue of flux (Exclude cases where no-clean type fluxes is used. However, recommend sufficiently about the residue.); or Washing our Products by using water or water-soluble cleaning agents for cleaning residue after soldering
 - [h] Use of the Products in places subject to dew condensation
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- 6. In particular, if a transient load (a large amount of load applied in a short period of time, such as pulse. is applied, confirmation of performance characteristics after on-board mounting is strongly recommended. Avoid applying power exceeding normal rated power; exceeding the power rating under steady-state loading condition may negatively affect product performance and reliability.
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- 8. Confirm that operation temperature is within the specified range described in the product specification.
- 9. ROHM shall not be in any way responsible or liable for failure induced under deviant condition from what is defined in this document.

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For details, please refer to ROHM Mounting specification

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 - [b] the temperature or humidity exceeds those recommended by ROHM
 - [c] the Products are exposed to direct sunshine or condensation
 - [d] the Products are exposed to high Electrostatic
- 2. Even under ROHM recommended storage condition, solderability of products out of recommended storage time period may be degraded. It is strongly recommended to confirm solderability before using Products of which storage time is exceeding the recommended storage time period.
- 3. Store / transport cartons in the correct direction, which is indicated on a carton with a symbol. Otherwise bent leads may occur due to excessive stress applied when dropping of a carton.
- 4. Use Products within the specified time after opening a humidity barrier bag. Baking is required before using Products of which storage time is exceeding the recommended storage time period.

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