## FEATURES

DVI interface
Supports high-bandwidth digital content protection
RGB to YCbCr 2-way color conversion
1.8 V/3.3 V power supply

100-lead, Pb-free LQFP
RGB and YCbCr output formats
Digital video interface
DVI 1.0
150 MHz DVI receiver
Supports high-bandwidth digital content protection (HDCP 1.1)

## APPLICATIONS

Advanced TVs
HDTVs
Projectors
LCD monitors


Figure 1.

## GENERAL DESCRIPTION

The AD9397 is a digital visual interface (DVI) receiver integrated on a single chip. Also included is support for high bandwidth digital content protection (HDCP) with internal key storage.

The AD9397 contains a DVI 1.0-compatible receiver and supports all HDTV formats (up to 1080p and 720p) and display resolutions up to SXGA ( $1280 \times 1024 @ 80 \mathrm{~Hz}$ ). The receiver features an intrapair skew tolerance of up to one full clock cycle. With the inclusion of HDCP, displays can receive encrypted
video content. The AD9397 allows for authentication of a video receiver, decryption of encoded data at the receiver, and renewability of that authentication during transmission as specified by the HDCP 1.1 protocol.

Fabricated in an advanced CMOS process, the AD9397 is provided in a space-saving, 100-lead, surface-mount, Pb -free plastic LQFP and is specified over the $0^{\circ} \mathrm{C}$ to $70^{\circ} \mathrm{C}$ temperature range.

## Rev. 0

## AD9397

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## REVISION HISTORY

## 10/05-Revision 0: Initial Version

## SPECIFICATIONS

## ELECTRICAL CHARACTERISTICS

$\mathrm{V}_{\mathrm{DD}}, \mathrm{V}_{\mathrm{D}}=3.3 \mathrm{~V}, \mathrm{DV} \mathrm{DD}=\mathrm{P} \mathrm{V}_{\mathrm{DD}}=1.8 \mathrm{~V}, \mathrm{ADC}$ clock $=$ maximum.
Table 1.


[^0]
## AD9397

## DIGITAL INTERFACE ELECTRICAL CHARACTERISTICS

$\mathrm{V}_{\mathrm{DD}}=\mathrm{V}_{\mathrm{D}}=3.3 \mathrm{~V}, \mathrm{DV} \mathrm{VD}_{\mathrm{DD}}=\mathrm{PV} \mathrm{DD}_{\mathrm{DD}}=1.8 \mathrm{~V}, \mathrm{ADC}$ clock $=$ maximum.
Table 2.

| Parameter | Test <br> Level | Conditions | AD9397KSTZ-100 |  |  | AD9397KSTZ-150 |  |  | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  |  | Min | Typ | Max | Min | Typ | Max |  |
| RESOLUTION |  |  |  | 8 |  |  | 8 |  | Bit |
| DC DIGITAL I/O SPECIFICATIONS <br> High Level Input Voltage, ( $\mathrm{V}_{\mathbf{H}}$ ) <br> Low Level Input Voltage, (VIL) <br> High Level Output Voltage, (V $\mathrm{V}_{\text {H }}$ ) <br> Low Level Output Voltage, (VoL) | $\begin{aligned} & \mathrm{VI} \\ & \mathrm{VI} \\ & \mathrm{VI} \\ & \mathrm{VI} \end{aligned}$ |  | $\begin{aligned} & 2.5 \\ & V_{D D}-0.1 \\ & V_{D D}-0.1 \end{aligned}$ |  | $\begin{aligned} & 0.8 \\ & 0.1 \end{aligned}$ | 2.5 |  | 0.8 0.1 | $\begin{aligned} & \mathrm{V} \\ & \mathrm{~V} \\ & \mathrm{~V} \\ & \mathrm{~V} \end{aligned}$ |
| DC SPECIFICATIONS <br> Output High Level <br> $\mathrm{IOHD}_{\text {, }}\left(\mathrm{V}_{\text {OUT }}=\mathrm{V}_{\text {OH }}\right)$ <br> Output Low Level <br> lold, $\left(V_{\text {out }}=V_{\text {oL }}\right)$ <br> DATACK High Level <br> $\mathrm{V}_{\text {OHC }}\left(\mathrm{V}_{\text {OUT }}=\mathrm{V}_{\text {OH }}\right)$ <br> DATACK Low Level <br> Volc, $\left(\right.$ Vout $\left.=V_{\text {ol }}\right)$ <br> Differential Input Voltage, Single-Ended Amplitude | $\begin{aligned} & \text { IV } \\ & \text { IV } \\ & \text { IV } \\ & \text { IV } \\ & \text { IV } \\ & \text { IV } \\ & \text { IV } \end{aligned}$ IV IV | Output drive $=$ high <br> Output drive = low <br> Output drive $=$ high <br> Output drive = low <br> Output drive $=$ high <br> Output drive = low <br> Output drive = high <br> Output drive = low | $75$ | $\begin{aligned} & 36 \\ & 24 \\ & 12 \\ & 8 \\ & 40 \\ & 20 \\ & 30 \\ & 15 \end{aligned}$ | 700 | 75 | $\begin{aligned} & 36 \\ & 24 \\ & 12 \\ & 8 \\ & 40 \\ & 20 \\ & 30 \\ & 15 \end{aligned}$ | 700 | mA <br> mA <br> mA <br> mA <br> mA <br> mA <br> mA <br> mA <br> mV |
| POWER SUPPLY <br> $V_{D}$ Supply Voltage <br> VDD Supply Voltage <br> DV ${ }_{\text {DD }}$ Supply Voltage <br> PV VD Supply Voltage <br> Ivo Supply Current (Typical Pattern) ${ }^{1}$ <br> Ivod Supply Current (Typical Pattern) ${ }^{2}$ <br> Iovod Supply Current (Typical Pattern) ${ }^{1,4}$ <br> Ipvod Supply Current (Typical Pattern) ${ }^{1}$ <br> Power-Down Supply Current (Ipo) | $\begin{aligned} & \mathrm{IV} \\ & \mathrm{IV} \\ & \mathrm{IV} \\ & \mathrm{IV} \\ & \mathrm{~V} \\ & \mathrm{~V} \\ & \mathrm{~V} \\ & \mathrm{~V} \\ & \mathrm{VI} \\ & \hline \end{aligned}$ |  | $\begin{aligned} & 3.15 \\ & 1.7 \\ & 1.7 \\ & 1.7 \end{aligned}$ | $\begin{aligned} & 3.3 \\ & 3.3 \\ & 1.8 \\ & 1.8 \\ & 80 \\ & 40 \\ & 88 \\ & 26 \\ & 130 \\ & \hline \end{aligned}$ | $\begin{aligned} & 3.47 \\ & 347 \\ & 1.9 \\ & 1.9 \\ & 100 \\ & 100^{3} \\ & 110 \\ & 35 \end{aligned}$ | $\begin{aligned} & 3.15 \\ & 1.7 \\ & 1.7 \\ & 1.7 \end{aligned}$ | $\begin{aligned} & 3.3 \\ & 3.3 \\ & 1.8 \\ & 1.8 \\ & 80 \\ & 55 \\ & 110 \\ & 30 \\ & 130 \end{aligned}$ | $\begin{aligned} & 3.47 \\ & 347 \\ & 1.9 \\ & 1.9 \\ & 110 \\ & 175^{3} \\ & 145 \\ & 40 \end{aligned}$ | V <br> V <br> V <br> V <br> mA <br> mA <br> mA <br> mA <br> mA |


| Parameter | Test <br> Level | Conditions | AD9397KSTZ-100 |  |  | AD9397KSTZ-150 |  |  | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  |  | Min | Typ | Max | Min | Typ | Max |  |
| AC SPECIFICATIONS |  |  |  |  |  |  |  |  |  |
| Intrapair (+ to -) Differential Input Skew (TDPS) | IV |  |  |  |  |  |  | 360 | ps |
| Channel to Channel Differential Input Skew (Tccs) | IV |  |  |  |  |  |  | 6 | Clock Period |
| Low-to-High Transition Time for Data and Controls ( $\mathrm{D}_{\text {Lнт }}$ ) | IV | $\begin{aligned} & \text { Output drive = high; } \\ & C_{L}=10 \mathrm{pF} \end{aligned}$ |  |  |  |  |  | 900 | ps |
|  | IV | $\begin{aligned} & \text { Output drive = low; } \\ & \mathrm{C}_{\mathrm{L}}=5 \mathrm{pF} \end{aligned}$ |  |  |  |  |  | 1300 | ps |
| Low-to-High Transition Time for DATACK (DLнT) | IV | $\begin{aligned} & \text { Output drive = high; } \\ & \mathrm{C}_{\mathrm{L}}=10 \mathrm{pF} \end{aligned}$ |  |  |  |  |  | 650 | ps |
|  | IV | $\begin{aligned} & \text { Output drive = low; } \\ & C_{L}=5 \mathrm{pF} \end{aligned}$ |  |  |  |  |  | 1200 | ps |
| High-to-Low Transition Time for Data and Controls (Dніт) | IV | $\begin{aligned} & \text { Output drive }=\text { high; } \\ & \mathrm{C}_{\mathrm{L}}=10 \mathrm{pF} \end{aligned}$ |  |  |  |  |  | 850 | ps |
|  | IV | $\begin{aligned} & \text { Output drive = low; } \\ & \mathrm{C}_{\mathrm{L}}=5 \mathrm{pF} \end{aligned}$ |  |  |  |  |  | 1250 | ps |
| High-to-Low Transition Time for DATACK (DHIT) | IV | $\begin{aligned} & \text { Output drive = high; } \\ & \mathrm{C}_{\mathrm{L}}=10 \mathrm{pF} \end{aligned}$ |  |  |  |  |  | 800 | ps |
|  | IV | $\begin{aligned} & \text { Output drive = low; } \\ & \mathrm{C}_{\mathrm{L}}=5 \mathrm{pF} \end{aligned}$ |  |  |  |  |  | 1200 | ps |
| Clock to Data Skew ${ }^{5}$ (Tskew) | IV |  | -0.5 |  | +2.0 | -0.5 |  | +2.0 | ns |
| Duty Cycle, DATACK ${ }^{5}$ | IV |  | 45 | 50 |  |  |  | 55 | \% |
| DATACK Frequency ( $\mathrm{F}_{\text {cIP }}$ ) | VI |  | 20 |  |  |  |  | 150 | MHz |

[^1]
## ABSOLUTE MAXIMUM RATINGS

Table 3.

| Parameter | Rating |
| :--- | :--- |
| $\mathrm{V}_{\mathrm{D}}$ | 3.6 V |
| $\mathrm{~V}_{\mathrm{DD}}$ | 3.6 V |
| $\mathrm{D} V_{D D}$ | 1.98 V |
| $\mathrm{P} V_{D D}$ | 1.98 V |
| Analog Inputs | $\mathrm{V}_{\mathrm{D}}$ to 0.0 V |
| Digital Inputs | 5 V to 0.0 V |
| Digital Output Current | 20 mA |
| Operating Temperature Range | $-25^{\circ} \mathrm{C}$ to $+85^{\circ} \mathrm{C}$ |
| Storage Temperature Range | $-65^{\circ} \mathrm{C}$ to $+150^{\circ} \mathrm{C}$ |
| Maximum Junction Temperature | $150^{\circ} \mathrm{C}$ |
| Maximum Case Temperature | $150^{\circ} \mathrm{C}$ |

Stresses above those listed under Absolute Maximum Ratings may cause permanent damage to the device. This is a stress rating only; functional operation of the device at these or any other conditions above those indicated in the operational section of this specification is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

## EXPLANATION OF TEST LEVELS

Table 4.

| Level | Test |
| :--- | :--- |
| I | $100 \%$ production tested. |
| II | $100 \%$ production tested at $25^{\circ} \mathrm{C}$ and sample tested at <br> specified temperatures. |
| III | Sample tested only. <br> IV <br> Parameter is guaranteed by design and <br> characterization testing. |
| V | Parameter is a typical value only. <br> $100 \%$ production tested at $25^{\circ} \mathrm{C} ;$ guaranteed by design <br> and characterization testing. |

## ESD CAUTION

ESD (electrostatic discharge) sensitive device. Electrostatic charges as high as 4000 V readily accumulate on the human body and test equipment and can discharge without detection. Although this product features proprietary ESD protection circuitry, permanent damage may occur on devices subjected to high energy electrostatic discharges. Therefore, proper ESD precautions are recommended to avoid performance degradation or loss of functionality.


## PIN CONFIGURATION AND FUNCTION DESCRIPTIONS



Figure 2. Pin Configuration
Table 5. Complete Pinout List

| Pin Type | Pin No. | Mnemonic | Function | Value |
| :---: | :---: | :---: | :---: | :---: |
| INPUTS | 81 | PWRDN | Power-Down Control | 3.3 V CMOS |
| DIGITAL VIDEO DATA INPUTS | $\begin{aligned} & 35 \\ & 34 \\ & 38 \\ & 37 \\ & 41 \\ & 40 \end{aligned}$ | $\begin{aligned} & \text { Rx0+ } \\ & \text { Rx0- } \\ & R \times 1+ \\ & R \times 1- \\ & R \times 2+ \\ & R \times 2- \end{aligned}$ | Digital Input Channel 0 True <br> Digital Input Channel 0 Complement <br> Digital Input Channel 1 True <br> Digital Input Channel 1 Complement <br> Digital Input Channel 2 True <br> Digital Input Channel 2 Complement | TMDS <br> TMDS <br> TMDS <br> TMDS <br> TMDS <br> TMDS |
| DIGITAL VIDEO CLOCK INPUTS | $\begin{aligned} & 43 \\ & 44 \end{aligned}$ | $\begin{aligned} & \mathrm{RxC}+ \\ & \mathrm{RxC}- \end{aligned}$ | Digital Data Clock True Digital Data Clock Complement | TMDS TMDS |
| OUTPUTS | $\begin{aligned} & \hline 92 \text { to } 99 \\ & 2 \text { to } 9 \\ & 12 \text { to } 19 \\ & 89 \\ & 87 \\ & 85 \\ & 84 \\ & 27,26,25,24 \end{aligned}$ | RED [7:0] <br> GREEN [7:0] <br> BLUE [7:0] <br> DATACK <br> HSOUT <br> VSOUT <br> O/E FIELD <br> CTL(0 to 3) | Outputs of Red Converter, Bit 7 is MSB <br> Outputs of Green Converter, Bit 7 is MSB <br> Outputs of Blue Converter, Bit 7 is MSB <br> Data Output Clock <br> HSYNC Output Clock (Phase-Aligned with DATACK) VSYNC Output Clock (Phase-Aligned with DATACK) <br> Odd/Even Field Output <br> Control 0, 1, 2, 3 | $V_{D D}$ <br> $V_{D D}$ <br> $V_{D D}$ <br> $V_{D D}$ <br> $V_{D D}$ <br> $V_{D D}$ <br> $V_{D D}$ <br> $V_{D D}$ |

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| Pin Type | Pin No. | Mnemonic | Function | Value |
| :---: | :---: | :---: | :---: | :---: |
| POWER SUPPLY | $\begin{aligned} & 80,76,72,67, \\ & 45,33 \\ & 100,90,10 \\ & 59,56,54 \\ & 48,32,30 \end{aligned}$ | V <br> $V_{D D}$ <br> PV $V_{D}$ <br> DVD <br> GND | Analog Power Supply and DVI Terminators <br> Output Power Supply <br> PLL Power Supply <br> Digital Logic Power Supply <br> Ground | $\begin{aligned} & \hline 3.3 \mathrm{~V} \\ & \\ & 1.8 \mathrm{~V} \text { to } 3.3 \mathrm{~V} \\ & 1.8 \mathrm{~V} \\ & 1.8 \mathrm{~V} \\ & 0 \mathrm{~V} \\ & \hline \end{aligned}$ |
| CONTROL | $\begin{aligned} & \hline 83 \\ & 82 \end{aligned}$ | $\begin{aligned} & \hline \text { SDA } \\ & \text { SCL } \end{aligned}$ | Serial Port Data I/O <br> Serial Port Data Clock | 3.3 V CMOS <br> 3.3 V CMOS |
| HDCP | $\begin{aligned} & 49 \\ & 50 \\ & 51 \\ & 52 \end{aligned}$ | $\begin{aligned} & \text { DDCSCL } \\ & \text { DDCSDA } \\ & \text { MCL } \\ & \text { MDA } \end{aligned}$ | HDCP Slave Serial Port Data Clock HDCP Slave Serial Port Data I/O HDCP Master Serial Port Data Clock HDCP Master Serial Port Data I/O | 3.3 V CMOS <br> 3.3 V CMOS <br> 3.3 V CMOS <br> 3.3 V CMOS |
| DATA ENABLE | 88 | DE | Data Enable | 3.3 V CMOS |
| RTERM | 46 | RTERM | Sets Internal Termination Resistance | $500 \Omega$ |

Table 6. Pin Function Descriptions

| Pin | Description |
| :---: | :---: |
| INPUTS |  |
| Rx0+ | Digital Input Channel 0 True. |
| Rx0- | Digital Input Channel 0 Complement. |
| Rx1+ | Digital Input Channel 1 True. |
| Rx1- | Digital Input Channel 1 Complement. |
| R×2+ | Digital Input Channel 2 True. |
| R×2- | Digital input Channel 2 Complement. |
|  | These six pins receive three pairs of transition minimized differential signaling (TMDS ) pixel data (at $10 \times$ the pixel rate) from a digital graphics transmitter. |
| RxC+ | Digital Data Clock True. |
| RxC- | Digital Data Clock Complement. |
|  | This clock pair receives a TMDS clock at $1 \times$ pixel data rate. |
| PWRDN | Power-Down Control/Three-State Control. |
|  | The function of this pin is programmable via Register 0x26 [2:1]. |
| RTERM | RTERM is the termination resistor used to drive the AD9397 internally to a precise $50 \Omega$ termination for TMDS lines. This should be a $500 \Omega 1 \%$ tolerance resistor. |
| OUTPUTS |  |
| HSOUT | Horizontal Sync Output. |
|  | A reconstructed and phase-aligned version of the HSYNC input. Both the polarity and duration of this output can be programmed via serial bus registers. By maintaining alignment with DATACK and Data, data timing with respect to horizontal sync can always be determined. |
| VSOUT | Vertical Sync Output. |
|  | The separated VSYNC from a composite signal or a direct pass through of the VSYNC signal. The polarity of this output can be controlled via the serial bus bit (Register 0x24 [6]). |
| FIELD | Odd/Even Field Bit for Interlaced Video. This output identifies whether the current field (in an interlaced signal) is odd or even. The polarity of this signal is programmable via Register 0x24[4]. |
| DE | Data Enable that defines valid video. Can be received in the signal or generated by the AD9397. |
| CTL(3-0) | Control 3, Control 2, Control 1, and Control 0 are output from the DVI stream. Refer to the DVI 1.0 specification for explanation. |
| SERIAL PORT |  |
| SDA | Serial Port Data I/O for Programming AD9397 Registers-12C Address is 0x98. |
| SCL | Serial Port Data Clock for Programming AD9397 Registers. |
| DDCSDA | Serial Port Data I/O for HDCP Communications to Transmitter-I2C Address is $0 \times 74$ or $0 \times 76$. |
| DDCSCL | Serial Port Data Clock for HDCP Communications to Transmitter. |
| MDA | Serial Port Data I/O to EEPROM with HDCP Keys-12C Address is $0 \times A 0$. |
| MCL | Serial Port Data Clock to EEPROM with HDCP Keys. |


| Pin | Description |
| :---: | :---: |
| DATA OUTPUTS RED [7:0] GREEN [7:0] BLUE [7:0] | Data Output, Red Channel. <br> Data Output, Green Channel. <br> Data Output, Blue Channel. <br> The main data outputs. Bit 7 is the MSB. The delay from pixel sampling time to output is fixed, but is different if the color space converter is used. When the sampling time is changed by adjusting the phase register, the output timing is shifted as well. The DATACK and HSOUT outputs are also moved, so the timing relationship among the signals is maintained. |
| DATA CLOCK OUTPUT DATACK | Data Clock Output. <br> This is the main clock output signal used to strobe the output data and HSOUT into external logic. Four possible output clocks can be selected with Register $0 \times 25$ [7:6]. These are related to the pixel clock ( $1 / 2 \times$ pixel clock, $1 \times$ pixel clock, $2 \times$ frequency pixel clock, and a $90^{\circ}$ phase shifted pixel clock). They are produced either by the internal PLL clock generator or EXTCLK and are synchronous with the pixel sampling clock. The polarity of DATACK can also be inverted via Register $0 \times 24$ [ 0 ]. The sampling time of the internal pixel clock can be changed by adjusting the phase register. When this is changed, the pixelrelated DATACK timing is shifted as well. The DATA, DATACK, and HSOUT outputs are all moved, so the timing relationship among the signals is maintained. |
| POWER SUPPLY ${ }^{1}$ |  |
| $\mathrm{V}_{\mathrm{D}}(3.3 \mathrm{~V})$ | Analog Power Supply. |
| VDD (1.8 V to 3.3 V ) | These pins supply power to the ADCs and terminators. They should be as quiet and filtered as possible. Digital Output Power Supply. |
| PV $\mathrm{DD}^{\text {(1.8 }} \mathrm{V}$ ) | A large number of output pins (up to 27) switching at high speed (up to 150 MHz ) generates many power supply transients (noise). These supply pins are identified separately from the $V_{D}$ pins, so output noise transferred into the sensitive analog circuitry can be minimized. If the AD9397 is interfacing with lower voltage logic, $\mathrm{V}_{\mathrm{DD}}$ may be connected to a lower supply voltage (as low as 1.8 V ) for compatibility. Clock Generator Power Supply. |
|  | The most sensitive portion of the AD9397 is the clock generation circuitry. These pins provide power to the clock PLL and help the user design for optimal performance. The designer should provide quiet, noise-free power to these pins. |
| DV $\mathrm{DD}^{(1.8 \mathrm{~V}}$ ) | Digital Input Power Supply. |
|  | This supplies power to the digital logic. |
| GND | Ground. |
|  | The ground return for all circuitry on chip. It is recommended that the AD9397 be assembled on a single solid ground plane, with careful attention to ground current paths. |

[^2]
## DESIGN GUIDE

## GENERAL DESCRIPTION

The AD9397 is a fully integrated digital visual interface (DVI) for receiving RGB or YUV signals for display on flat panel monitors, projectors or PDPs. This interface is capable of decoding HDCP-encrypted signals through connection to an external EEPROM. The circuit is ideal for providing an interface for HDTV monitors or as the front-end to high performance video scan converters.

Implemented in a high performance CMOS process, the interface can capture signals with pixel rates of up to 150 MHz .

The AD9397 includes all necessary input buffering, signal dc restoration (clamping), offset and gain (brightness and contrast) adjustment, pixel clock generation, sampling phase control, and output data formatting. Included in the output formatting is a color space converter (CSC), which accommodates any input color space and can output any color space. All controls are programmable via a 2-wire serial interface. Full integration of these sensitive analog functions makes system design straightforward and less sensitive to the physical and electrical environments.

## DIGITAL INPUTS

All digital control inputs (HSYNC, VSYNC, $\mathrm{I}^{2} \mathrm{C}$ ) on the AD9397 operate to 3.3 V CMOS levels. In addition, all digital inputs except the TMDS (DVI) inputs are 5 V tolerant. (Applying 5 V to them does not cause any damage.) TMDS inputs ( $\mathrm{Rx} 0+/ \mathrm{Rx} 0-$, $\mathrm{Rx} 1+/ \mathrm{Rx} 1-, \mathrm{Rx} 2+/ \mathrm{Rx} 2-$, and $\mathrm{RxC}+/ \mathrm{RxC}-$ ) must maintain a $100 \Omega$ differential impedance (through proper PCB layout) from the connector to the input where they are internally terminated ( $50 \Omega$ to 3.3 V ). If additional ESD protection is desired, use of a California Micro Devices (CMD) CM1213 (among others) series low capacitance ESD protection offers 8 kV of protection to the HDMI TMDS lines.

## SERIAL CONTROL PORT

The serial control port is designed for 3.3 V logic. However, it is tolerant of 5 V logic signals.

## OUTPUT SIGNAL HANDLING

The digital outputs operate from 1.8 V to $3.3 \mathrm{~V}\left(\mathrm{~V}_{\mathrm{DD}}\right)$.

## POWER MANAGEMENT

The AD9397 uses the activity detect circuits, the active interface bits in the serial bus, the active interface override bits, the power-down bit, and the power-down pin to determine the correct power state. There are four power states: full-power, seek mode, auto power-down, and power-down. Table 7 summarizes how the AD9397 determines which power mode to be in and which circuitry is powered on/off in each of these modes. The power-down command has priority and then the automatic circuitry. The power-down pin (Pin 81 -polarity set by Register 0x26[3]) can drive the chip into four power-down options. Bit 2 and Bit 1 of Register 0x26 control these four options. Bit 0 controls whether the chip is powered down or the outputs are placed in high impedance mode (with the exception of SOG). Bit 7 to Bit 4 of Register 0x26 control whether the outputs, SOG, Sony Philips digital interface (S/PDIF), or InterIC sound bus ( $\mathrm{I}^{2}$ S or IIS) outputs are in high impedance mode or not. See the 2-Wire Serial Control Register Detail section for more details.

Table 7. Power-Down Mode Descriptions

| Mode | Inputs |  |  |  |
| :--- | :--- | :--- | :--- | :--- |
|  | Sync Detect $^{\mathbf{2}}$ | Auto PD Enable $^{\mathbf{3}}$ | Power-On or Comments |  |
| Full Power | 1 | 1 | X | Everything |
| Seek Mode | 1 | 0 | 0 | Everything |
| Seek Mode | 1 | 0 | 1 | Serial bus, sync activity detect, SOG, band gap reference |
| Power-Down | 0 | X | Serial bus, sync activity detect, SOG, band gap reference |  |

[^3]
## TIMING

The output data clock signal is created so that its rising edge always occurs between data transitions and can be used to latch the output data externally.

Figure 3 shows the timing operation of the AD9397.


## HSYNC TIMING

Horizontal sync (HSYNC) is processed in the AD9397 to eliminate ambiguity in the timing of the leading edge with respect to the phase-delayed pixel clock and data.

The HSYNC input is used as a reference to generate the pixel sampling clock. The sampling phase can be adjusted, with respect to HSYNC, through a full $360^{\circ}$ in 32 steps via the phase adjust register (to optimize the pixel sampling time). Display systems use HSYNC to align memory and display write cycles, so it is important to have a stable timing relationship between the HSYNC output (HSOUT) and data clock (DATACK).

## VSYNC FILTER AND ODD/EVEN FIELDS

The VSYNC filter is used to eliminate spurious VSYNCs, maintain a consistent timing relationship between the VSYNC and HSYNC output signals, and generate the odd/even field output.

The filter works by examining the placement of VSYNC with respect to HSYNC and, if necessary, slightly shifting it in time at the VSOUT output. The goal is to keep the VSYNC and HSYNC leading edges from switching at the same time, eliminating confusion as to when the first line of a frame occurs. Enabling the VSYNC filter is done with Register 0x21[5]. Use of the VSYNC filter is recommended for all cases, including interlaced video, and is required when using the HSYNC per VSYNC counter. Figure 4 and Figure 5 illustrate even/odd field determination in two situations.


Figure 5. VSYNC Filter—Odd/Even

## DVI RECEIVER

The DVI receiver section of the AD9397 allows the reception of a digital video stream compatible with DVI 1.0. Embedded in this data stream are HSYNCs, VSYNCs, and display enable (DE) signals. DVI restricts the received format to RGB, but the inclusion of a programmable color space converter (CSC) allows the output to be tailored to any format necessary. With this, the scaler following the AD9397 can specify that it always wishes to receive a particular format-for instance, 4:2:2 $\mathrm{YCrCb}-$ regardless of the transmitted mode. If RGB is sent, the CSC can easily convert that to $4: 2: 2 \mathrm{YCrCb}$ while relieving the scaler of this task.

## DE GENERATOR

The AD9397 has an onboard generator for DE, for start of active video (SAV), and for end of active video (EAV), all of which are necessary for describing the complete data stream for a BT656-compatible output. In addition to this particular output, it is possible to generate the DE for cases in which a scaler is not used. This signal alerts the following circuitry as to which are displayable video pixels.

## AD9397

## 4:4:4 TO 4:2:2 FILTER

The AD9397 contains a filter that allows it to convert a signal from $\mathrm{YCrCb} 4: 4: 4$ to $\mathrm{YCrCb} 4: 2: 2$ while maintaining the maximum accuracy and fidelity of the original signal.

## Input Color Space to Output Color Space

The AD9397 can support a wide variety of output formats, such as:

- RGB 24-bit
- 4:4:4 YCrCb 8-bit
- 4:2:2 YCrCb 8-bit, 10 -bit, and 12-bit
- Dual 4:2:2 YCrCb 8-bit


## Color Space Conversion (CSC) Matrix

The CSC matrix in the AD9397 consists of three identical processing channels. In each channel, three input values are multiplied by three separate coefficients. Also included are an offset value for each row of the matrix and a scaling multiple for all values. Each value has a 13-bit, twos complement resolution to ensure the signal integrity is maintained. The CSC is designed to run at speeds up to 150 MHz supporting resolutions up to 1080 p at 60 Hz . With any-to-any color space support, formats such as RGB, YUV, YCbCr, and others are supported by the CSC.

The main inputs, $\mathrm{R}_{\mathrm{IN}}, \mathrm{GiN}_{\mathrm{IN}}$, and $\mathrm{B}_{\mathrm{IN}}$, come from the 8 -bit to 12 -bit inputs from each channel. These inputs are based on the input format detailed in Table 9. The mapping of these inputs to the CSC inputs is shown in Table 8.

Table 8. CSC Port Mapping

| Input Channel | CSC Input Channel |
| :--- | :--- |
| $R / C R$ | $\mathrm{R}_{\mathbb{N}}$ |
| $\mathrm{Gr} / \mathrm{Y}$ | $\mathrm{G}_{\mathbb{N}}$ |
| $\mathrm{B} / \mathrm{CB}$ | $\mathrm{B}_{\mathbb{N}}$ |

One of the three channels is represented in Figure 6. In each processing channel, the three inputs are multiplied by three separate coefficients marked a1, a2, and a3. These coefficients are divided by 4096 to obtain nominal values ranging from -0.9998 to +0.9998 . The variable labeled a4 is used as an offset control. The CSC_Mode setting is the same for all three processing channels. This multiplies all coefficients and offsets by a factor of $2^{\text {CSC_Mode }}$.

The functional diagram for a single channel of the CSC, as shown in Figure 6, is repeated for the remaining G and B channels. The coefficients for these channels are b1, b2, b3, b4, $\mathrm{c} 1, \mathrm{c} 2, \mathrm{c} 3$, and c4.


Figure 6. Single CSC Channel
A programming example and register settings for several common conversions are listed in the Color Space Converter (CSC) Common Settings section.

For a detailed functional description and more programming examples, refer to the Application Note AN-795, AD9880 Color Space Converter User's Guide.

## OUTPUT DATA FORMATS

The AD9398 supports 4:4:4, 4:2:2, double data-rate (DDR), and BT656 output formats. Register 0x25[3:0] controls the output mode. These modes and the pin mapping are illustrated in Table 8.

Table 9.

| Port | Red |  |  |  |  |  |  |  | Green |  |  |  |  |  |  |  |  | Blue |  |  |  |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Bit | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 | 7 | 6 | 5 | 4 | 3 | 2 |  | 1 | 0 | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| 4:4:4 | Red/Cr [7:0] |  |  |  |  |  |  |  | Green/Y [7:0] |  |  |  |  |  |  |  |  | Blue/Cb [7:0] |  |  |  |  |  |  |  |
| 4:2:2 | CbCr [7:0] |  |  |  |  |  |  |  | Y [7:0] |  |  |  |  |  |  |  |  | DDR 4:2:2 $\uparrow \mathrm{CbCr} \downarrow \mathrm{Y}, \mathrm{Y}$ |  |  |  |  |  |  |  |
| 4:4:4 DDR | DDR $\uparrow{ }^{1} \mathrm{G}$ [3:0] |  |  |  | DDR $\uparrow$ B [7:4] |  |  |  | DDR $\uparrow$ B [3:0] |  |  |  | DDR 4:2:2 $\uparrow$ CbCr [11:0] |  |  |  |  |  |  |  |  |  |  |  |  |
|  | DDR $\downarrow \mathrm{R}$ [7:0] |  |  |  |  |  |  |  | DDR $\downarrow \mathrm{G}$ [7:4] |  |  |  | DDR 4:2:2 $\downarrow$ Y,Y [11:0] |  |  |  |  |  |  |  |  |  |  |  |  |
| 4:2:2 to 12 | CbCr[11:0] |  |  |  |  |  |  |  |  |  |  |  | Y [11:0] |  |  |  |  |  |  |  |  |  |  |  |  |

[^4]
## 2-WIRE SERIAL REGISTER MAP

The AD9397 is initialized and controlled by a set of registers that determines the operating modes. An external controller is employed to write and read the control registers through the 2-wire serial interface port.
Table 10. Control Register Map

| Hex <br> Address | Read/Write or Read Only | Bits | Default Value | Register Name | Description |
| :---: | :---: | :---: | :---: | :---: | :---: |
| 0x00 | Read | [7:0] | 00000000 | Chip Revision | Chip revision ID. |
| 0x11 | Read/Write | [7] <br> [6] | 0****** | HSYNC Source | $\begin{aligned} & 0=\text { HSYNC. } \\ & 1=\text { SOG. } . \end{aligned}$ |
|  |  |  | * $0^{* * * * * * ~}$ | HSYNC Source Override | $0=$ auto HSYNC source. |
|  |  | [5] | ${ }^{* *} 0^{* * * * *}$ | VSYNC Source | $\begin{aligned} & 0=\text { VSYNC. } \\ & 1=\text { VSYNC from SOG. } \end{aligned}$ |
|  |  | [4] | ${ }^{* * *} 0^{* * * *}$ | VSYNC Source Override | $0=$ auto HSYNC source. |
|  |  | [3] | ****0*** | Channel Select | $\begin{aligned} & 1=\text { manual HSYNC source. } \\ & 0=\text { Channel } 0 . \\ & 1=\text { Channel } 1 . \end{aligned}$ |
|  |  | [2] | *****0** | Channel Select Override | $0=$ autochannel select. |
|  |  | [1] | ******0* | Interface Select | $\begin{aligned} & 1=\text { manual channel select. } \\ & 0=\text { analog interface. } \\ & 1=\text { digital interface. } \end{aligned}$ |
|  |  | [0] | *******0 | Interface Override | $\begin{aligned} & 0=\text { auto-interface select. } \\ & 1=\text { manual interface select. } \end{aligned}$ |
| 0x12 | Read/Write |  | $1^{* * * * * * *}$ | Input HSYNC Polarity | $\begin{aligned} & 0 \text { = active low. } \\ & 1 \text { = active high. } \end{aligned}$ |
|  |  | [6] | *0****** | HSYNC Polarity Override | $0=$ auto HSYNC polarity. |
|  |  |  |  |  | 1 = manual HSYNC polarity. |
|  |  | [5] | ${ }^{* *} 1^{* * * * *}$ | Input VSYNC Polarity | $\begin{aligned} & 0=\text { active low. } \\ & 1=\text { active high. } \end{aligned}$ |
|  |  | [4] | ***0**** | VSYNC Polarity Override | 0 = auto VSYNC polarity. |
|  |  |  |  |  | 1 = manual VSYNC polarity. |
| 0x17 | Read | [3:0] | ****0000 | HSYNCs per VSYNC MSB | MSB of HSYNCs per VSYNC. |
| 0x18 | Read | [7:0] | 00000000 | HSYNCs per VSYNC | HSYNCs per VSYNC count. |
| 0x22 | Read/Write | [7:0] | 4 | VSYNC Duration | VSYNC duration. |
| 0x23 | Read/Write | [7:0] | 32 | HSYNC Duration | HSYNC duration. Sets the duration of the output HSYNC in pixel clocks. |
| 0x24 | Read/Write | [7] | $1^{* * * * * * *}$ | HSYNC Output Polarity | Output HSYNC polarity. <br> 0 = active low out. <br> 1 = active high out. |
|  |  | [6] | *1****** | VSYNC Output Polarity | Output VSYNC polarity. |
|  |  |  |  |  | $\begin{aligned} & 0=\text { active low out. } \\ & 1=\text { active high out. } \end{aligned}$ |
|  |  | [5] | ** $1^{* * * * * ~}$ | DE Output Polarity | Output DE polarity. <br> $0=$ active low out. <br> 1 = active high out. |

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| Hex Address | Read/Write or Read Only | Bits | Default Value | Register Name | Description |
| :---: | :---: | :---: | :---: | :---: | :---: |
|  |  | [4] [0] | $* * * 1 * * * *$ $* * * * * * * 0$ | Field Output Polarity Output CLK Invert | Output field polarity. <br> $0=$ active low out. <br> 1 = active high out. <br> $0=$ Don't invert clock out. <br> 1 = Invert clock out. |
| 0x25 | Read/Write | [7:6] | 01****** | Output CLK Select | Selects which clock to use on output pin. $1 \times$ CLK is divided down from TMDS clock input when pixel repetition is in use. $\begin{aligned} & 00=1 / 2 \times \text { CLK. } \\ & 01=1 \times \text { CLK. } \\ & 10=2 \times \text { CLK. } \\ & 11=90^{\circ} \text { phase } 1 \times \text { CLK. } \end{aligned}$ |
|  |  | [5:4] | ${ }^{* *} 11^{* * * *}$ | Output Drive Strength | Sets the drive strength of the outputs. 00 = lowest, 11 = highest. |
|  |  | [3:2] | ****00** | Output Mode | Selects which pins the data comes out on. $00=4: 4: 4$ mode (normal). <br> $01=4: 2: 2+$ DDR 4:2:2 on blue. <br> $10=$ DDR 4:4:4 + DDR 4:2:2 on blue. |
|  |  | [1] | ****** ${ }^{*}$ | Primary Output Enable | Enables primary output. |
|  |  | [0] | *******0 | Secondary Output Enable | Enables secondary output (DDR 4:2:2 in Output Mode 1 and Mode 2). |
| 0x26 | Read/Write | [7] <br> [5] <br> [4] <br> [3] | $\begin{aligned} & 0^{* * * * * * *} \\ & * * 0^{* * * * *} \\ & * * 0^{* * * *} \\ & * * *{ }^{*}{ }^{* * * *} \end{aligned}$ | Output Three-State SPDIF Three-State $I^{2}$ S Three-State Power-Down Pin Polarity | Three-state the outputs. <br> Three-state the SPDIF output. <br> Three-state the I2S output and the MCLK out. <br> Sets polarity of power-down pin. $\begin{aligned} & 0=\text { active low. } \\ & 1=\text { active high. } \end{aligned}$ |
|  |  |  |  |  |  |
|  |  |  |  |  |  |
|  |  |  | ${ }^{* * * *} 1^{* * *}$ |  |  |
|  |  |  |  |  |  |
|  |  |  |  |  |  |
|  |  | [2:1] | *****00* | Power-Down Pin Function | Selects the function of the power-down pin. |
|  |  |  |  |  | 00 = power-down. |
|  |  |  |  |  | 01 = power-down and three-state SOG. |
|  |  |  |  |  | $10=$ three-state outputs only. |
|  |  |  |  |  | 11 = three-state outputs and SOG. |
|  |  | [0] | *******0 | Power-Down | $0=$ normal. |
|  |  |  |  |  | 1 = power-down. |
| $0 \times 27$ | Read/Write | [7] | $1^{* * * * * * *}$ | Auto Power-Down Enable | 0 = disable auto low power state.1 = enable auto low power state. |
|  |  |  |  |  |  |
|  |  | [6] | *0****** | HDCP AO | Sets the LSB of the address of the $\operatorname{HDCP} 1^{2} C$. Set to 1 only for a second receiver in a dual-link configuration. $\begin{aligned} & 0=\text { use internally generated MCLK. } \\ & 1=\text { use external MCLK input. } \end{aligned}$ |
|  |  | [5] | ${ }^{* *} 0^{* * * * *}$ | MCLK External Enable | If an external MCLK is used, it must be locked to the video clock according to the CTS and N available in the $\mathrm{I}^{2} \mathrm{C}$. Any mismatch between the internal MCLK and the input MCLK results in dropped or repeated audio samples. |
|  |  | $[4]$$[3]$$[2: 0]$ | ***0**** | BT656 EN | Enables EAV/SAV codes to be inserted into the video output data. |
|  |  |  | $\begin{aligned} & * * * * 0^{* * *} \\ & * * * * * 000 \end{aligned}$ | Force DE Generation Interlace Offset | Allows use of the internal DE generator in DVI mode. Sets the difference (in HSYNCs) in field length between Field 0 and Field 1. |
|  |  | [2:0] |  |  |  |


| Hex <br> Address | Read/Write or Read Only | Bits | Default Value | Register Name | Description |
| :---: | :---: | :---: | :---: | :---: | :---: |
| 0x28 | Read/Write | $\begin{aligned} & {[7: 2]} \\ & {[1: 0]} \\ & \hline \end{aligned}$ | $\begin{aligned} & 011000^{* *} \\ & * * * * * * 01 \end{aligned}$ | VS Delay <br> HS Delay MSB | Sets the delay (in lines) from the VSYNC leading edge to the start of active video. MSB, Register 0x29. |
| 0x29 | Read/Write | [7:0] | 00000100 | HS Delay | Sets the delay (in pixels) from the HSYNC leading edge to the start of active video. |
| 0x2A | Read/Write | [3:0] | ****0101 | Line Width MSB | MSB, Register 0x2B. |
| 0x2B | Read/Write | [7:0] | 00000000 | Line Width | Sets the width of the active video line in pixels. |
| 0x2C | Read/Write | [3:0] | ****0010 | Screen Height MSB | MSB, Register 0x2D. |
| 0x2D | Read/Write | [7:0] | 11010000 | Screen Height | Sets the height of the active screen in lines. |
| 0x2E | Read/Write | [7] | 0******* | Test 1 | Must be written to 1 for proper operation. |
| 0x2F | Read | [6] <br> [5] <br> [3] <br> [2:0] | $\begin{aligned} & * 0^{* * * * * *} \\ & * * 0^{* * * * *} \\ & * * * 0^{* * *} \\ & * * * * * 000 \end{aligned}$ | TMDS Sync Detect TMDS Active HDCP Keys Read DVI Quality | Detects a TMDS DE. <br> Detects a TMDS clock. <br> Returns 1 when read of EEPROM keys is successful. <br> Returns quality number based on DE edges. |
| 0x30 | Read | [6] <br> [5] <br> [4] | $\begin{aligned} & \hline{ }^{*} 0^{* * * * * *} \\ & \\ & * * 0^{* * * * *} \\ & * * * 0^{* * * *} \end{aligned}$ | DVI Content Encrypted <br> DVI HSYNC Polarity DVI VSYNC Polarity | This bit is high when HDCP decryption is in use (content is protected). The signal goes low when HDCP is not being used. Customers can use this bit to determine whether to allow copying of the content. The bit should be sampled at regular intervals because it can change on a frame-by-frame basis. <br> Returns DVI HSYNC polarity. <br> Returns DVI VSYNC polarity. |
| 0x31 | Read/Write | $\begin{aligned} & {[7: 4]} \\ & {[3: 0]} \end{aligned}$ | $\begin{aligned} & 1001^{* * * *} \\ & * * * 0110 \end{aligned}$ | MV Pulse Max <br> MV Pulse Min | Sets the maximum pseudo sync pulse width for Macrovision detection. <br> Sets the minimum pseudo sync pulse width for Macrovision ${ }^{\ominus}$ detection. |
| 0x32 | Read/Write | [7] <br> [6] <br> [5:0] | $\begin{aligned} & 0^{* * * * * *} \\ & * 0^{* * * * * *} \\ & * * 001101 \end{aligned}$ | MV Oversample En <br> MV Pal En <br> MV Line Count Start | Tells the Macrovision detection engine whether we are oversampling or not. <br> Tells the Macrovision detection engine to enter PAL mode. Sets the start line for Macrovision detection. |
| 0x33 | Read/Write | [7] <br> [6] <br> [5:0] | $\begin{aligned} & 1 * * * * * * \\ & * 0^{* * * * * *} \\ & { }^{* *} 010101 \end{aligned}$ | MV Detect Mode <br> MV Settings Override <br> MV Line Count End | 0 = standard definition. <br> 1 = progressive scan mode. <br> $0=$ use hard-coded settings for line counts and pulse widths. <br> $1=$ use $I^{2} \mathrm{C}$ values for these settings. <br> Sets the end line for Macrovision detection. |
| 0x34 | Read/Write | [7:6] <br> [5] <br> [4] <br> [3] <br> [2] <br> [1] | $\begin{aligned} & 10^{* * * * * *} \\ & * * 0^{* * * * *} \\ & * * * 0^{* * * *} \\ & * * * * 0^{* * *} \\ & * * * * * 0^{* *} \\ & * * * * * 0^{*} \end{aligned}$ | MV Pulse Limit Set <br> Low Freq Mode <br> Low Freq Override <br> Up Conversion Mode <br> CrCb Filter Enable CSC_Enable | Sets the number of pulses required in the last 3 lines (SD mode only). <br> Sets audio PLL to low frequency mode. Low frequency mode should only be set for pixel clocks $<80 \mathrm{MHz}$. <br> Allows the previous bit to be used to set low frequency mode rather than the internal auto-detect. <br> $0=$ repeat Cr and Cb values. <br> $1=$ interpolate Cr and Cb values. <br> Enables the FIR filter for 4:2:2 CrCb output. <br> Enables the color space converter (CSC). The default settings for the CSC provide HDTV-to-RGB conversion. <br> Sets the fixed point position of the CSC coefficients, including the A4, B4, and C4 offsets. |
| 0x35 | Read/Write | $\begin{aligned} & \hline \text { [6:5] } \\ & {[4: 0]} \end{aligned}$ | ${ }^{*} 01^{* * * * *}$ <br> ***01100 | CSC_Mode <br> CSC_Coeff_A1 MSB | $\begin{aligned} & 00= \pm 1.0,-4096 \text { to }+4095 . \\ & 01= \pm 2.0,-8192 \text { to }+8190 . \\ & 1 \times= \pm 4.0,-16384 \text { to }+16380 . \end{aligned}$ <br> MSB, Register 0x36. |

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| Hex Address | Read/Write or Read Only | Bits | Default Value | Register Name | Description |
| :---: | :---: | :---: | :---: | :---: | :---: |
| 0x36 | Read/Write | [7:0] | 01010010 | CSC_Coeff_A1 LSB | Color space converter (CSC) coefficient for equation: |
| $\begin{aligned} & \hline 0 \times 37 \\ & 0 \times 38 \end{aligned}$ | Read/Write Read/Write | $\begin{aligned} & \hline[4: 0] \\ & {[7: 0]} \end{aligned}$ | $\begin{aligned} & \hline * * * 01000 \\ & 00000000 \end{aligned}$ | $\begin{aligned} & \hline \text { CSC_Coeff_A2 MSB } \\ & \text { CSC_Coeff_A2 LSB } \end{aligned}$ | MSB, Register 0x38. <br> CSC coefficient for equation: $\begin{aligned} & \text { Rout }=\left(\mathrm{A} 1 \times \mathrm{R}_{\text {IN }}+\left(\mathrm{A} 2 \times \mathrm{G}_{\text {IN }}\right)+\left(\mathrm{A} 3 \times \mathrm{B}_{\text {IN }}\right)+\mathrm{A} 4\right. \\ & \mathrm{G}_{\text {out }}=\left(\mathrm{B} 1 \times \mathrm{R}_{\text {IN }}\right)+\left(\mathrm{B} 2 \times \mathrm{G}_{\text {IN }}\right)+\left(\mathrm{B} 3 \times \mathrm{BiN}_{\text {N }}\right)+\mathrm{B} 4 \\ & \text { Bout }\left(\mathrm{C} 1 \times \mathrm{R}_{\text {IN }}\right)+\left(\mathrm{C} 2 \times \mathrm{G}_{\text {IN }}\right)+\left(\mathrm{C} 3 \times \mathrm{B}_{\text {IN }}\right)+\mathrm{C} \end{aligned}$ |
| 0x39 | Read/Write | [4:0] | ***00000 | CSC_Coeff_A3 MSB | MSB, Register 0x3A. |
| 0x3A | Read/Write | [7:0] | 00000000 | CSC_Coeff_A3 LSB | CSC coefficient for equation: |
| 0x3B | Read/Write | [4:0] | ***11001 | CSC_Coeff_A4 MSB | MSB, Register 0x3C. |
| 0x3C | Read/Write | [7:0] | 11010111 | CSC_Coeff_A4 LSB | CSC coefficient for equation: |
| 0x3D | Read/Write | [4:0] | ***11100 | CSC_Coeff_B1 MSB | MSB, Register 0x3E. |
| 0x3E | Read/Write | [7:0] | 01010100 | CSC_Coeff_B1 LSB | CSC coefficient for equation: |
| 0x3F | Read/Write | [4:0] | ***01000 | CSC_Coeff_B2 MSB | MSB, Register 0x40. |
| 0x40 | Read/Write | [7:0] | 00000000 | CSC_Coeff_B2 LSB | CSC coefficient for equation: |
| 0x41 | Read/Write | [4:0] | ***11110 | CSC_Coeff_B3 MSB | MSB, Register 0x42. |
| 0x42 | Read/Write | [7:0] | 10001001 | CSC_Coeff_B3 | CSC coefficient for equation: |
| 0x43 | Read/Write | [4:0] | ***00010 | CSC_Coeff_B4 MSB | MSB, Register 0x44. |
| 0x44 | Read/Write | [7:0] | 10010010 | CSC_Coeff_B4 LSB | CSC coefficient for equation: |
| 0x45 | Read/Write | [4:0] | ***00000 | CSC_Coeff_C1 MSB | MSB, Register 0x46. |
| 0x46 | Read/Write | [7:0] | 00000000 | CSC_Coeff_C1 LSB | CSC coefficient for equation: |
| 0x47 | Read/Write | [4:0] | ***01000 | CSC_Coeff_C2 MSB | MSB, Register 0x48. |
| 0x48 | Read/Write | [7:0] | 00000000 | CSC_Coeff_C2 LSB | CSC coefficient for equation: |


| Hex Address | Read/Write or Read Only | Bits | Default Value | Register Name | Description |
| :---: | :---: | :---: | :---: | :---: | :---: |
| $\begin{aligned} & 0 \times 49 \\ & 0 \times 4 \mathrm{~A} \end{aligned}$ | Read/Write Read/Write | $\begin{aligned} & {[4: 0]} \\ & {[7: 0]} \end{aligned}$ | $\begin{aligned} & \hline * * * 01110 \\ & 10000111 \end{aligned}$ | CSC_Coeff_C3 MSB CSC_Coeff_C3 LSB | MSB, Register 0x4A. <br> CSC coefficient for equation: $\begin{aligned} & \text { Rout }=\left(\mathrm{A} 1 \times \mathrm{R}_{\text {IN }}\right)+\left(\mathrm{A} 2 \times \mathrm{G}_{\text {IN }}\right)+\left(\mathrm{A} 3 \times \mathrm{B}_{\text {IN }}\right)+\mathrm{A} 4 \\ & \mathrm{G}_{\text {out }}=\left(\mathrm{B} 1 \times \mathrm{R}_{\text {IN }}\right)+\left(\mathrm{G} \times \mathrm{G}_{\text {IN }}\right)+\left(\mathrm{B} \times \mathrm{B}_{\text {IN }}\right)+\mathrm{B} \\ & \text { Bout }\left(\mathrm{C} 1 \times \mathrm{R}_{\text {IN }}\right)+\left(\mathrm{C} 2 \times \mathrm{G}_{\text {IN }}\right)+\left(\mathrm{C} 3 \times \mathrm{B}_{\text {IN }}\right)+\mathrm{C} \end{aligned}$ |
| 0x4B | Read/Write | [4:0] | ***11000 | CSC_Coeff_C4 MSB | MSB, Register 0x4C. |
| 0x4C | Read/Write | [7:0] | 10111101 | CSC_Coeff_C4 LSB | CSC coefficient for equation: |
| 0x50 | Read/Write | [7:0] | 00100000 | Test | Must be written to $0 \times 20$ for proper operation. |
| 0x56 | Read/Write | [7:0] | 00001111 | Test | Must be written to default of 0x0F for proper operation. |
| 0x59 | Read/Write | $\begin{aligned} & {[6]} \\ & {[5]} \\ & {[4]} \\ & {[0]} \end{aligned}$ |  | MDA/MCL PU CLK Term O/R Manual CLK Term MDA/MCL ThreeState | This disables the MDA/MCL pull-ups. <br> Clock termination power-down override: $0=$ auto, $1=$ manual. <br> Clock termination: $0=$ normal, $1=$ disconnected. <br> This bit three-states the MDA/MCL lines. |

## AD9397

## 2-WIRE SERIAL CONTROL REGISTER DETAILS

## CHIP IDENTIFICATION

0x00—Bit[7:0] Chip Revision
An 8-bit value that reflects the current chip revision.

## 0x11—Bit[7] HSYNC Source

$0=$ HSYNC, $1=$ SOG. The power-up default is 0 . These selections are ignored if Register 0x11, Bit $6=0$.

## 0x11-Bit[6] HSYNC Source Override

$0=$ auto HSYNC source, $1=$ manual HSYNC source. Manual HSYNC source is defined in Register 0x11, Bit 7. The power-up default is 0 .

## 0x11-Bit[5] VSYNC Source

$0=$ VSYNC, $1=$ VSYNC from SOG. The power-up default is 0 . These selections are ignored if Register 0x11, Bit $4=0$.

## 0x11—Bit[4] VSYNC Source Override

$0=$ auto VSYNC source, $1=$ manual VSYNC source. Manual VSYNC source is defined in Register 0x11, Bit 5. The power-up default is 0 .

## 0x11-Bit[3] Channel Select

$0=$ Channel $0,1=$ Channel 1 . The power-up default is 0 . These selections are ignored if Register 0x11, Bit $2=0$.

## 0x11-Bit[2] Channel Select Override

$0=$ auto channel select, $1=$ manual channel select. Manual channel select is defined in Register 0x11, Bit 3. The power-up default is 0 .

## Ox11—Bit[1] Interface Select

$0=$ analog interface, $1=$ digital interface. The power-up default is 0 . These selections are ignored if Register $0 \times 11$, Bit $0=0$.

## 0x11-Bit[0] Interface Select Override

$0=$ auto interface select, $1=$ manual interface select. Manual interface select is defined in Register 0x11, Bit 1. The power-up default is 0 .

## Ox12—Bit[7] Input HSYNC Polarity

$0=$ active low, $1=$ active high. The power-up default is 1 . These selections are ignored if Register 10x2, Bit $6=0$.

## 0x12-Bit[6] HSYNC Polarity Override

0 = auto HSYNC polarity, $1=$ manual HSYNC polarity. Manual HSYNC polarity is defined in Register 0x11, Bit 7. The powerup default is 0 .

## 0x12—Bit[5] Input VSYNC Polarity

$0=$ active low, $1=$ active high. The power-up default is 1 . These selections are ignored if Register $0 \times 11$, Bit $4=0$.

## 0x12-Bit[4] VSYNC Polarity Override

$0=$ auto VSYNC polarity, $1=$ manual VSYNC polarity. Manual VSYNC polarity is defined in Register 0x11, Bit 5. The powerup default is 0 .

## 0x17—Bits[3:0] HSYNCs per VSYNC MSBs

The 4 MSBs of the 12-bit counter that reports the number of HSYNCs/VSYNC on the active input. This is useful in determining the mode and aids in setting the PLL divide ratio.

## 0x18—Bit[7:0] HSYNCs per VSYNC LSBs

The 8 LSBs of the 12-bit counter that reports the number of HSYNCs/VSYNC on the active input.

## 0x21—Bit[5] VSYNC Filter Enable

The purpose of the VSYNC filter is to guarantee the position of the VSYNC edge with respect to the HSYNC edge and to generate a field signal. The filter works by examining the placement of VSYNC and regenerating a correctly placed VSYNC one line later. The VSYNC is first checked to see whether it occurs in the Field 0 position or the Field 1 position. This is done by checking the leading edge position against the sync separator threshold and the HSYNC position. The HSYNC width is divided into four quadrants with Quadrant 1 starting at the HSYNC leading edge plus a sync separator threshold. If the VSYNC leading edge occurs in Quadrant 1 or Quadrant 4, the field is set to 0 and the output VSYNC is placed coincident with the HSYNC leading edge. If the VSYNC leading edge occurs in Quadrant 2 or Quadrant 3, the field is set to 1 and the output VSYNC leading edge is placed in the center of the line. In this way, the VSYNC filter creates a predictable relative position between HSYNC and VSYNC edges at the output.

If the VSYNC occurs near the HSYNC edge, this guarantees that the VSYNC edge follows the HSYNC edge. This performs filtering also in that it requires a minimum of 64 lines between VSYNCs. The VSYNC filter cleans up extraneous pulses that might occur on the VSYNC. This should be enabled whenever the HSYNC/VSYNC count is used. Setting this bit to 0 disables the VSYNC filter. Setting this bit to 1 enables the VSYNC filter. Power-up default is 0 .

## 0x21—Bit[4] VSYNC Duration Enable

This enables the VSYNC duration block which is designed to be used with the VSYNC filter. Setting the bit to 0 leaves the VSYNC output duration unchanged; setting the bit to 1 sets the VSYNC output duration based on Register 0x22. The power-up default is 0 .

## 0x22—Bits[7:0] VSYNC Duration

This is used to set the output duration of the VSYNC, and is designed to be used with the VSYNC filter. This is valid only if Register $0 \times 21$, Bit 4 is set to 1 . Power-up default is 4 .

## 0x23-Bit[7:0] HSYNC Duration

An 8-bit register that sets the duration of the HSYNC output pulse. The leading edge of the HSYNC output is triggered by the internally generated, phase-adjusted PLL feedback clock. The AD9397 then counts a number of pixel clocks equal to the value in this register. This triggers the trailing edge of the HSYNC output, which is also phase-adjusted. The power-up default is 32 .

## 0x24—Bit[7] HSYNC Output Polarity

This bit sets the polarity of the HSYNC output. Setting this bit to 0 sets the HSYNC output to active low. Setting this bit to 1 sets the HSYNC output to active high. Power-up default setting is 1 .

## 0x24—Bit[6] VSYNC Output Polarity

This bit sets the polarity of the VSYNC output (both DVI and analog). Setting this bit to 0 sets the VSYNC output to active low. Setting this bit to 1 sets the VSYNC output to active high. Power-up default is 1 .

## 0x24—Bit[5] Display Enable Output Polarity

This bit sets the polarity of the display enable (DE) for both DVI and analog. $0=\mathrm{DE}$ output polarity is negative. $1=\mathrm{DE}$ output polarity is positive. The power-up default is 1 .

## 0x24—Bit[4] Field Output Polarity

This bit sets the polarity of the field output signal (both DVI and analog) on Pin $21.0=$ active low out = even field; active high = odd field. $1=$ active high out = odd field; active high = even field. The power-up default is 1 .

## 0x24—Bit[0] Output Clock Invert

This bit allows inversion of the output clock as specified by Register $0 \times 25$, Bits 7 to $6.0=$ noninverted clock. $1=$ inverted clock. The power-up default setting is 0 .

## 0x25—Bits[7:6] Output Clock Select

These bits select the clock output on the DATACLK pin. They include $1 / 2 \times$ clock, a $2 \times$ clock, a $90^{\circ}$ phase shifted clock, or the normal pixel clock. The power-up default setting is 01 .

Table 11. Output Clock Select

| Select | Result |
| :--- | :--- |
| 00 | $1 / 2 \times$ pixel clock |
| 01 | $1 \times$ pixel clock |
| 10 | $2 \times$ pixel clock |
| 11 | $90^{\circ}$ phase $1 \times$ pixel clock |

## 0x25—Bit[5:4] Output Drive Strength

These two bits select the drive strength for all the high speed digital outputs (except VSOUT, A0 and O/E field). Higher drive strength results in faster rise/fall times and in general makes it easier to capture data. Lower drive strength results in slower rise/fall times and helps to reduce EMI and digitally generated power supply noise. The power-up default setting is 11 .
Table 12. Output Drive Strength

| Output Drive | Result |
| :--- | :--- |
| 00 | Low output drive strength |
| 01 | Medium low output drive strength |
| 10 | Medium high output drive strength |
| 11 | High output drive strength |

## 0x25—Bits[3:2] Output Mode

These bits choose between four options for the output mode, one of which is exclusive to an HDMI input. 4:4:4 mode is standard RGB; 4:2:2 mode is YCrCb , which reduces the number of active output pins from 24 to $16 ; 4: 4: 4$ is double data rate (DDR) output mode; and the data is RGB mode, but changes on every clock edge. The power-up default setting is 00 .
Table 13. Output Mode

| Output Mode | Result |
| :--- | :--- |
| 00 | $4: 4: 4$ RGB mode <br> 01 |
| 4:2:2 YCrCb mode + DDR 4:2:2 on blue <br> (secondary) |  |
| 10 | DDR 4:4:4: DDR mode + DDR 4:2:2 on blue <br> (secondary) <br> 12-bit 4:2:2 (HDMI option only) |
| 11 |  |

## 0x25—Bit[1] Primary Output Enable

This bit places the primary output in active or high impedance mode. The primary output is designated when using either 4:2:2 or DDR 4:4:4. In these modes, the data on the red and green output channels is the primary output, while the output data on the blue channel (DDR YCrCb ) is the secondary output. $0=$ primary output is in high impedance mode. $1=$ primary output is enabled. The power-up default setting is 1 .

## 0x25-Bit[0] Secondary Output Enable

This bit places the secondary output in active or high impedance mode. The secondary output is designated when using either 4:2:2 or DDR 4:4:4. In these modes, the data on the blue output channel is the secondary output while the output data on the red and green channels is the primary output. Secondary output is always a DDR YCrCb data mode. The power-up default setting is $0.0=$ secondary output is in high impedance mode. $1=$ secondary output is enabled.

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## 0x26—Bit[7] Output Three-State

When enabled, this bit puts all outputs (except SOGOUT) in a high impedance state. $0=$ normal outputs. $1=$ all outputs (except SOGOUT) in high impedance mode. The power-up default setting is 0 .

## 0x26—Bit[3] Power-Down Polarity

This bit defines the polarity of the input power-down pin. $0=$ power-down pin is active low. $1=$ power-down pin is active high. The power-up default setting is 1 .

## 0x26—Bits[2:1] Power-Down Pin Function

These bits define the different operational modes of the powerdown pin. These bits are functional only when the power-down pin is active; when it is not active, the part is powered up and functioning. $0 \mathrm{x}=$ the chip is powered down and all outputs are in high impedance mode. $1 \mathrm{x}=$ the chip remains powered up, but all outputs are in high impedance mode. The power-up default setting is 00 .

## 0x26—Bit[0] Power-Down

This bit is used to put the chip in power-down mode. In this mode, the power dissipation is reduced to a fraction of the typical power (see Table 1 for exact power dissipation). When in power-down, the HSOUT, VSOUT, DATACK, and all 30 of the data outputs are put into a high impedance state. Note that the SOGOUT output is not put into high impedance. Circuit blocks that continue to be active during power-down include the voltage references, sync processing, sync detection, and the serial register. These blocks facilitate a fast start-up from powerdown. $0=$ normal operation. $1=$ power-down. The power-up default setting is 0 .

## 0x27—Bit[7] Auto Power-Down Enable

This bit enables the chip to go into low power mode, or seek mode if no sync inputs are detected. $0=$ auto power-down disabled. $1=$ chip powers down if no sync inputs present. The power-up default setting is 1 .

## 0x27-Bit[6] HDCP AO Address

This bit sets the LSB of the address of the HDCP $\mathrm{I}^{2} \mathrm{C}$. This should be set to 1 only for a second receiver in a dual-link configuration. The power-up default is 0 .

## BT656 GENERATION

## 0x27—Bit[4] BT656 Enable

This bit enables the output to be BT656 compatible with the defined start of active video (SAV) and end of active video (EAV) controls to be inserted. These require specification of the number of active lines, active pixels per line, and delays to place these markers. $0=$ disable BT656 video mode. $1=$ enable BT656 video mode. The power-up default setting is 0 .

## 0x27—Bit[3] Force DE Generation

This bit allows the use of the internal DE generator in DVI mode. $0=$ internal DE generation disabled. $1=$ force DE generation via programmed registers. The power-up default setting is 0 .

## 0x27—Bits[2:0] Interlace Offset

These bits define the offset in HSYNCs from Field 0 to Field 1. The power-up default setting is 000 .

## 0x28—Bits[7:2] VSYNC Delay

These bits set the delay (in lines) from the leading edge of VSYNC to active video. The power-up default setting is 24 .

## 0x28—Bit[1:0] HSYNC Delay MSBs

These 2 bits and the following 8 bits set the delay (in pixels) from the HSYNC leading edge to the start of active video. The power-up default setting is $0 \times 104$.

## 0x29—Bits[7:0] HSYNC Delay LSBs

See the HSYNC Delay MSBs section.

## 0x2A—Bits[3:0] Line Width MSBs

These 4 bits and the following 8 bits set the width of the active video line (in pixels). The power-up default setting is $0 \times 500$.

## 0x2B—Bits[7:0] Line Width LSBs

See the line width MSBs section.

## 0x2C—Bits[3:0] Screen Height MSBs

These 4 bits and the following 8 bits set the height of the active screen (in lines). The power-up default setting is $0 \times 2 \mathrm{D} 0$.

## 0x2D—Bits[7:0] Screen Height LSBs

See the Screen Height MSBs section.

## 0x2F—Bit[6] TMDS Sync Detect

This read-only bit indicates the presence of a TMDS DE. $0=$ no TMDS DE present. $1=$ TMDS DE detected.

## 0x2F—Bit[5] TMDS Active

This read-only bit indicates the presence of a TMDS clock. $0=$ no TMDS clock present. $1=$ TMDS clock detected.

## 0x2F—Bit[3] HDCP Keys Read

This read-only bit reports if the HDCP keys were read successfully. $0=$ failure to read HDCP keys. $1=$ HDCP keys read.

## 0x2F—Bits[2:0] DVI Quality

These read-only bits indicate a level of DVI quality based on the DE edges. A larger number indicates a higher quality.

## 0x30—Bit[6] DVI Content Encrypted

This read-only bit is high when HDCP decryption is in use (content is protected). The signal goes low when HDCP is not being used. Customers can use this bit to allow copying of the content. The bit should be sampled at regular intervals because it can change on a frame-by-frame basis. $0=$ HDCP not in use.
$1=$ HDCP decryption in use.

## 0x30—Bit[5] DVI HSYNC Polarity

This read-only bit indicates the polarity of the DVI HSYNC. $0=$ DVI HSYNC polarity is low active. $1=$ DVI HSYNC polarity is high active.

## 0x30—Bit[4] DVI VSYNC Polarity

This read-only bit indicates the polarity of the DVI VSYNC. $0=$ DVI VSYNC polarity is low active. $1=$ DVI VSYNC polarity is high active.

## MACROVISION

## 0x31—Bits[7:4] Macrovision Pulse Max

These bits set the pseudo sync pulse width maximum for Macrovision detection in pixel clocks. This is functional for 13.5 MHz SDTV or 27 MHz progressive scan. Power-up default is 9 .

## 0x31—Bits[3:0] Macrovision Pulse Min

These bits set the pseudo sync pulse width maximum for Macrovision detection in pixel clocks. This is functional for 13.5 MHz SDTV or 27 MHz progressive scan. Power up default is 6 .

## 0x32—Bit[7] Macrovision Oversample Enable

Tells the Macrovision detection engine whether oversampling is used. This accommodates 27 MHz sampling for SDTV and 54 MHz sampling for progressive scan and is used as a correction factor for clock counts. Power-up default is 0 .

## 0x32-Bit[6] Macrovision PAL Enable

Tells the Macrovision detection engine to enter PAL mode when set to 1 . Default is 0 for NTSC mode.

## 0x32—Bits[5:0] Macrovision Line Count Start

Set the start line for Macrovision detection. Along with Register 0x33, Bits [5:0], they define the region where MV pulses are expected to occur. The power-up default is Line 13.

## 0x33—Bit[7] Macrovision Detect Mode

$0=$ standard definition. $1=$ progressive scan mode

## 0x33—Bit[6] Macrovision Settings Override

This defines whether preset values are used for the MV line counts and pulse widths or the values stored in $\mathrm{I}^{2} \mathrm{C}$ registers are used. $0=$ use hard-coded settings for line counts and pulse widths. $1=$ use $I^{2} \mathrm{C}$ values for these settings.

## 0x33—Bits[5:0] Macrovision Line Count End

Set the end line for Macrovision detection. Along with Register $0 \times 32$, Bits [5:0], they define the region where MV pulses are expected to occur. The power-up default is Line 21.

## 0x34—Bits[7:6] Macrovision Pulse Limit Select

Set the number of pulses required in the last three lines (SD mode only). If there is not at least this number of MV pulses, the engine stops. These two bits define these pulse counts:

$$
\begin{aligned}
& 00=6 . \\
& 01=4 . \\
& 10=5 \text { (default). } \\
& 11=7 .
\end{aligned}
$$

## Ox34—Bit[5] Low Frequency Mode

Sets whether the audio PLL is in low frequency mode or not. Low frequency mode should only be set for pixel clocks $<80 \mathrm{MHz}$.

## 0x34—Bit[4] Low Frequency Override

Allows the previous bit to be used to set low frequency mode rather than the internal autodetect.

## 0x34—Bit[3] Upconversion Mode

$0=$ repeat $\mathrm{Cb} / \mathrm{Cr}$ values. $1=$ interpolate $\mathrm{Cb} / \mathrm{Cr}$ values.

## 0x34-Bit[2] CbCr Filter Enable

Enables the FIR filter for 4:2:2 CbCr output.

## COLOR SPACE CONVERSION

The default power-up values for the color space converter coefficients (R0x35 through R0x4C) are set for ATSC RGB-toYCbCr conversion. They are completely programmable for other conversions.

## 0x34—Bit[1] Color Space Converter Enable

This bit enables the color space converter. $0=$ disable color space converter. 1 = enable color space converter. The power-up default setting is 0 .

## 0x35—Bits[6:5] Color Space Converter Mode

These two bits set the fixed-point position of the CSC coefficients, including the A4, B4, and C4 offsets. Default $=01$.

Table 14. CSC Fixed Point Converter Mode

| Select | Result |
| :--- | :--- |
| 00 | $\pm 1.0,-4096$ to +4095 |
| 01 | $\pm 2.0,-8192$ to +8190 |
| $1 \times$ | $\pm 4.0,-16384$ to +16380 |

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## 0x35—Bits[4:0] Color Space Conversion Coefficient A1 MSBs

These 5 bits form the 5 MSBs of the Color Space Conversion Coefficient A1. This combined with the 8 LSBs of the following register form a 13-bit, twos complement coefficient which is user programmable. The equation takes the form of:

$$
\begin{aligned}
& \text { Rout }=\left(\mathrm{A} 1 \times \mathrm{R}_{\text {IN }}\right)+\left(\mathrm{A} 2 \times \mathrm{G}_{\text {IN }}\right)+\left(\mathrm{A} 3 \times \mathrm{B}_{\text {IN }}\right)+\mathrm{A} 4 \\
& \mathrm{G}_{\text {out }}=\left(\mathrm{B} 1 \times \mathrm{R}_{\text {IN }}\right)+\left(\mathrm{B} 2 \times \mathrm{G}_{\text {IN }}\right)+\left(\mathrm{B} 3 \times \mathrm{B}_{\text {IN }}\right)+\mathrm{B} 4 \\
& \text { Bout }=\left(\mathrm{C} 1 \times \mathrm{R}_{\text {IN }}\right)+\left(\mathrm{C} 2 \times \mathrm{G}_{\text {IN }}\right)+\left(\mathrm{C} 3 \times \mathrm{B}_{\text {IN }}\right)+\mathrm{C} 4
\end{aligned}
$$

The default value for the 13-bit A1 coefficient is $0 \times 0 \mathrm{C} 52$.

## 0x36—Bits[7:0] Color Space Conversion Coefficient A1 LSBs

See the Register 0x35 section.

## 0x37—Bits[4:0] CSC A2 MSBs

These five bits form the 5 MSBs of the Color Space Conversion Coefficient A2. Combined with the 8 LSBs of the following register they form a 13-bit, twos complement coefficient that is user programmable. The equation takes the form of:

$$
\begin{aligned}
& \text { Rout }=\left(\mathrm{A} 1 \times \mathrm{R}_{\text {IN }}\right)+\left(\mathrm{A} 2 \times \mathrm{G}_{\text {IN }}\right)+\left(\mathrm{A} 3 \times \mathrm{B}_{\text {IN }}\right)+\mathrm{A} 4 \\
& \mathrm{G}_{\text {out }}=\left(\mathrm{B} 1 \times \mathrm{R}_{\text {IN }}\right)+\left(\mathrm{B} 2 \times \mathrm{G}_{\text {IN }}\right)+\left(\mathrm{B} 3 \times \mathrm{B}_{\text {IN }}\right)+\mathrm{B} 4 \\
& \text { Bout }=\left(\mathrm{C} 1 \times \mathrm{R}_{\text {IN }}\right)+\left(\mathrm{C} 2 \times \mathrm{G}_{\text {IN }}\right)+\left(\mathrm{C} 3 \times \mathrm{B}_{\text {IN }}\right)+\mathrm{C} 4
\end{aligned}
$$

The default value for the 13 -bit A2 coefficient is $0 \times 0800$.

## 0x38-Bits[7:0] CSC A2 LSBs

See the Register 0x37 section.

## 0x39—Bits[4:0] CSC A3 MSBs

The default value for the 13 -bit A3 is $0 x 0000$.

## 0x3A—Bits[7:0] CSC A3 LSBs <br> 0x3B—Bits[4:0] CSC A4 MSBs

The default value for the 13-bit A4 is 0x19D7.

## 0x3C—Bits[7:0] CSC A4 LSBs

0x3D—Bits[4:0] CSC B1 MSBs
The default value for the 13 -bit B 1 is $0 \times 1 \mathrm{C} 54$.

## 0x3E—Bits[7:0] CSC B1 LSBs

0x3F—Bits[4:0] CSC B2 MSBs
The default value for the 13 -bit B 2 is $0 \times 0800$.

## 0x40—Bits[7:0] CSC B2 LSBs <br> 0x41—Bits[4:0] CSC B3 MSBs

The default value for the 13 -bit B3 is $0 \times 1$ E89.

## 0x42—Bit[7:0] CSC B3 LSBs <br> 0x43-Bit[4:0] CSC B4 MSBs

The default value for the 13 -bit B4 is $0 \times 0291$.

## 0x44-Bits[7:0] CSC B4 LSBs <br> 0x45—Bits[4:0] CSC C1 MSBs

The default value for the 13 -bit C 1 is 0 x 0000 .

## 0x46—Bits[7:0] CSC C1 LSBs <br> 0x47—Bits[4:0] CSC C2 MSBs

The default value for the 13 -bit C2 is $0 \times 0800$.

```
0x48-Bits[7:0] CSC C2 LSBs
0x49—Bits[4:0] CSC C3 MSBs
```

The default value for the 13 -bit C 3 is 0 x 0 E 87 .

## 0x4A—Bits[7:0] CSC C3 LSBs <br> 0x4B—Bits[4:0] CSC C4 MSBs

The default value for the 13 -bit C 4 is 0 x 18 BD .

## 0x4C—Bits[7:0] CSC C4 LSBs <br> 0x59—Bit[6] MDA/MCL PU Disable

This bit disables the inter-MDA/MCL pull-ups.

## 0x59—Bit[5] CLK Term O/R

This bit allows for overriding during power down.
0 = auto, 1 = manual.

## 0x59—Bit[4] Manual CLK Term

This bit allows normal clock termination or disconnects this. 0 $=$ normal, $1=$ disconnected.

## 0x59—Bit[0] MDA/MCL Three-State

This bit three-states the MDA/MCL lines to allow in-circuit programming of the EEPROM.

## 2-WIRE SERIAL CONTROL PORT

A 2-wire serial interface control interface is provided in the AD9397. Up to two AD9397 devices can be connected to the 2-wire serial interface, with a unique address for each device.

The 2-wire serial interface comprises a clock (SCL) and a bidirectional data (SDA) pin. The analog flat panel interface acts as a slave for receiving and transmitting data over the serial interface. When the serial interface is not active, the logic levels on SCL and SDA are pulled high by external pull-up resistors.

Data received or transmitted on the SDA line must be stable for the duration of the positive-going SCL pulse. Data on SDA must change only when SCL is low. If SDA changes state while SCL is high, the serial interface interprets that action as a start or stop sequence.
There are six components to serial bus operation:

- Start signal
- Slave address byte
- Base register address byte
- Data byte to read or write
- Stop signal
- Acknowledge (Ack)

When the serial interface is inactive (SCL and SDA are high), communications are initiated by sending a start signal. The start signal is a high-to-low transition on SDA while SCL is high. This signal alerts all slave devices that a data transfer sequence is coming.

The first 8 bits of data transferred after a start signal comprise a 7 -bit slave address (the first 7 bits) and a single $\mathrm{R} / \overline{\mathrm{W}}$ bit (the eighth bit). The $\mathrm{R} / \overline{\mathrm{W}}$ bit indicates the direction of data transfer, read from (1) or write to (0) the slave device. If the transmitted slave address matches the address of the device (set by the state of the SA0 input pin as shown in Table 15), the AD9397 acknowledges by bringing SDA low on the 9th SCL pulse. If the addresses do not match, the AD9397 does not acknowledge.
Table 15. Serial Port Addresses


## DATA TRANSFER VIA SERIAL INTERFACE

For each byte of data read or written, the MSB is the first bit of the sequence.

If the AD9397 does not acknowledge the master device during a write sequence, the SDA remains high so the master can generate a stop signal. If the master device does not acknowledge the AD9397 during a read sequence, the AD9397 interprets this as the end of data. The SDA remains high, so the master can generate a stop signal.

To write data to specific control registers of the AD9397, the 8 -bit address of the control register of interest must be written after the slave address has been established. This control register address is the base address for subsequent write operations. The base address auto-increments by 1 for each byte of data written after the data byte intended for the base address. If more bytes are transferred than there are available addresses, the address does not increment and remains at its maximum value. Any base address higher than the maximum value does not produce an acknowledge signal.

Data are read from the control registers of the AD9397 in a similar manner. Reading requires two data transfer operations:

- The base address must be written with the $\mathrm{R} / \overline{\mathrm{W}}$ bit of the slave address byte low to set up a sequential read operation.
- Reading (the $\mathrm{R} / \overline{\mathrm{W}}$ bit of the slave address byte high) begins at the previously established base address. The address of the read register auto-increments after each byte is transferred.
To terminate a read/write sequence to the AD9397, a stop signal must be sent. A stop signal comprises a low-to-high transition of SDA while SCL is high.

A repeated start signal occurs when the master device driving the serial interface generates a start signal without first generating a stop signal to terminate the current communication. This is used to change the mode of communication (read, write) between the slave and master without releasing the serial interface lines.


Figure 7. Serial Port Read/Write Timing

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## SERIAL INTERFACE READ/WRITE EXAMPLES

Write to one control register:

- Start signal
- $\quad$ Slave address byte $(\mathrm{R} / \overline{\mathrm{W}}$ bit $=$ low $)$
- Base address byte
- Data byte to base address
- Stop signal

Write to four consecutive control registers:

- Start signal
- $\quad$ Slave address byte $(\mathrm{R} / \overline{\mathrm{W}}$ bit $=$ low $)$
- Base address byte
- Data byte to base address
- Data byte to (base address +1 )
- Data byte to (base address +2 )
- Data byte to (base address +3 )
- Stop signal

Read from one control register:

- Start signal
- $\quad$ Slave address byte $(\mathrm{R} / \overline{\mathrm{W}}$ bit $=$ low $)$
- Base address byte
- Start signal
- $\quad$ Slave address byte $(\mathrm{R} / \overline{\mathrm{W}}$ bit $=$ high $)$
- Data byte from base address
- Stop signal

Read from four consecutive control registers:

- Start signal
- $\quad$ Slave address byte $(\mathrm{R} / \overline{\mathrm{W}}$ bit $=$ low $)$
- Base address byte
- Start signal
- $\quad$ Slave address byte $(\mathrm{R} / \overline{\mathrm{W}}$ bit $=$ high $)$
- Data byte from base address
- Data byte from (base address +1 )
- Data byte from (base address +2 )
- Data byte from (base address +3 )
- Stop signal


Figure 8. Serial Interface—Typical Byte Transfer

## PCB LAYOUT RECOMMENDATIONS

The AD9397 is a high precision, high speed digital device. To achieve the maximum performance from the part, it is important to have a well laid-out board. The following is a guide for designing a board using the AD9397.

## POWER SUPPLY BYPASSING

It is recommended to bypass each power supply pin with a $0.1 \mu \mathrm{~F}$ capacitor. The exception is in the case where two or more supply pins are adjacent to each other. For these groupings of powers/grounds, it is only necessary to have one bypass capacitor. The fundamental idea is to have a bypass capacitor within about 0.5 cm of each power pin. Also, avoid placing the capacitor on the opposite side of the PC board from the AD9397, because that interposes resistive vias in the path.

The bypass capacitors should be physically located between the power plane and the power pin. Current should flow from the power plane to the capacitor to the power pin. Do not make the power connection between the capacitor and the power pin. Placing a via underneath the capacitor pads down to the power plane is generally the best approach.

It is particularly important to maintain low noise and good stability of $\mathrm{PV} \mathrm{VD}_{\mathrm{DD}}$ (the clock generator supply). Abrupt changes in $\mathrm{PV}_{\text {DD }}$ can result in similarly abrupt changes in sampling clock phase and frequency. This can be avoided by careful attention to regulation, filtering, and bypassing. It is highly desirable to provide separate regulated supplies for each of the analog circuitry groups ( $\mathrm{V}_{\mathrm{D}}$ and $\mathrm{PV} \mathrm{VD}_{\mathrm{D}}$ ).

Some graphic controllers use substantially different levels of power when active (during active picture time) and when idle (during HSYNC and VSYNC periods). This can result in a measurable change in the voltage supplied to the analog supply regulator, which can in turn produce changes in the regulated analog supply voltage. This can be mitigated by regulating the analog supply, or at least $P V_{D D}$, from a different, cleaner power source (for example, from a 12 V supply).

It is recommended to use a single ground plane for the entire board. Experience has repeatedly shown that the noise performance is the same or better with a single ground plane. Using multiple ground planes can be detrimental because each separate ground plane is smaller and long ground loops can result.

In some cases, using separate ground planes is unavoidable, so it is recommend to place a single ground plane under the AD9397. The location of the split should be at the receiver of the digital outputs. In this case, it is even more important to place components wisely because the current loops are much longer (current takes the path of least resistance). An example of a current loop is: power plane to AD9397 to digital output trace to digital data receiver to digital ground plane.

## OUTPUTS (BOTH DATA AND CLOCKS)

Try to minimize the trace length that the digital outputs have to drive. Longer traces have higher capacitance, which require more current that causes more internal digital noise.

Shorter traces reduce the possibility of reflections.
Adding a $50 \Omega$ to $200 \Omega$ series resistor can suppress reflections, reduce EMI, and reduce the current spikes inside the AD9397. If series resistors are used, place them as close as possible to the AD9397 pins (although try not to add vias or extra length to the output trace to move the resistors closer).

If possible, limit the capacitance that each of the digital outputs drives to less than 10 pF . This can be accomplished easily by keeping traces short and by connecting the outputs to only one device. Loading the outputs with excessive capacitance increases the current transients inside of the AD9397 and creates more digital noise on its power supplies.

## DIGITAL INPUTS

The digital inputs on the AD9397 were designed to work with 3.3 V signals, but are tolerant of 5.0 V signals. Therefore, no extra components need to be added if using 5.0 V logic.

Any noise that enters the HSYNC input trace can add jitter to the system. Therefore, minimize the trace length and do not run any digital or other high frequency traces near it.

## AD9397

## COLOR SPACE CONVERTER (CSC) COMMON SETTINGS

Table 16. HDTV YCrCb (0 to 255) to RGB (0 to 255) (Default Setting for AD9397)

| Register | Red/Cr Coeff 1 |  | Red/Cr Coeff 2 |  | Red/Cr Coeff 3 |  | Red/Cr Offset |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Address | $0 \times 35$ | 0x36 | $0 \times 37$ | 0x38 | 0x39 | 0x3A | 0x3B | 0x3C |
| Value | 0×2C | 0x52 | $0 \times 08$ | 0x00 | 0x00 | 0x00 | 0x19 | 0xD7 |
| Register | Green/Y Coeff 1 |  | Green/Y Coeff 2 |  | Green/Y Coeff 3 |  | Green/Y Offset |  |
| Address | 0x3D | 0x3E | 0x3F | 0x40 | 0x41 | 0x42 | 0x43 | 0x44 |
| Value | $0 \times 1 \mathrm{C}$ | 0x54 | $0 \times 08$ | 0x00 | 0x3E | 0x89 | 0x02 | 0x91 |
| Register | Blue/Cb Coeff 1 |  | Blue/Cb Coeff 2 |  | Blue/Cb Coeff 3 |  | Blue/Cb Offset |  |
| Address | 0x45 | 0x46 | 0x47 | 0x48 | 0x49 | 0x4A | 0x4B | 0x4C |
| Value | $0 \times 00$ | 0x00 | $0 \times 08$ | 0x00 | 0x0E | 0x87 | 0x18 | 0xBD |

Table 17. HDTV YCrCb (16 to 235) to RGB (0 to 255)

| Register | Red/Cr Coeff 1 |  | Red/Cr Coeff 2 |  | Red/Cr Coeff 3 |  | Red/Cr Offset |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Address | 0x35 | 0x36 | 0x37 | 0x38 | 0x39 | 0x3A | 0x3B | 0x3C |
| Value | $0 \times 47$ | 0x2C | 0x04 | 0xA8 | $0 \times 00$ | 0x00 | $0 \times 1 \mathrm{C}$ | 0x1F |
| Register | Green/Y Coeff 1 |  | Green/Y Coeff 2 |  | Green/Y Coeff 3 |  | Green/Y Offset |  |
| Address | 0x3D | 0x3E | 0x3F | 0x40 | 0x41 | 0x42 | 0x43 | 0x44 |
| Value | $0 \times 1 \mathrm{D}$ | 0xDD | 0x04 | 0xA8 | $0 \times 1 \mathrm{~F}$ | 0x26 | $0 \times 01$ | 0x34 |
| Register | Blue/Cb Coeff 1 |  | Blue/Cb Coeff 2 |  | Blue/Cb Coeff 3 |  | Blue/Cb Offset |  |
| Address | 0x45 | 0x46 | 0x47 | 0x48 | 0x49 | 0x4A | 0x4B | 0x4C |
| Value | $0 \times 00$ | 0x00 | 0x04 | 0xA8 | 0x08 | 0x 75 | 0x1B | 0x7B |

Table 18. SDTV YCrCb (0 to 255) to RGB (0 to 255)

| Register | Red/Cr Coeff 1 |  | Red/Cr Coeff 2 |  | Red/Cr Coeff 3 |  | Red/Cr Offset |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Address | 0x35 | 0x36 | 0x37 | 0x38 | 0x39 | 0x3A | 0x3B | 0x3C |
| Value | $0 \times 2 \mathrm{~A}$ | 0xF8 | 0x08 | 0x00 | 0x00 | 0x00 | $0 \times 1 \mathrm{~A}$ | 0x84 |
| Register | Green/Y Coeff 1 |  | Green/Y Coeff 2 |  | Green/Y Coeff 3 |  | Green/Y Offset |  |
| Address | 0x3D | 0x3E | 0x3F | 0x40 | 0x41 | 0x42 | 0x43 | 0x44 |
| Value | $0 \times 1 \mathrm{~A}$ | 0x6A | 0x08 | 0x00 | 0x1D | 0x50 | 0x04 | 0x23 |
| Register | Blue/Cb Coeff. 1 |  | Blue/Cb Coeff 2 |  | Blue/Cb Coeff 3 |  | Blue/Cb Offset |  |
| Address | 0x45 | 0x46 | 0x47 | 0x48 | 0x49 | 0x4A | 0x4B | 0x4C |
| Value | $0 \times 00$ | 0x00 | 0x08 | 0x00 | 0x0D | 0xDB | 0x19 | 0x12 |

Table 19. SDTV YCrCb (16 to 235) to RGB (0 to 255)

| Register | Red/Cr Coeff 1 |  | Red/Cr Coeff 2 |  | Red/Cr Coeff 3 |  | Red/Cr Offset |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Address | 0x35 | 0x36 | 0x37 | 0x38 | 0x39 | 0x3A | 0x3B | 0x3C |
| Value | 0x46 | 0x63 | 0x04 | 0xA8 | 0x00 | 0x00 | $0 \times 1 \mathrm{C}$ | 0x84 |
| Register | Green/Y Coeff 1 |  | Green/Y Coeff 2 |  | Green/Y Coeff 3 |  | Green/Y Offset |  |
| Address | 0x3D | 0x3E | 0x3F | 0x40 | 0x41 | 0x42 | 0x43 | 0x44 |
| Value | $0 \times 1 \mathrm{C}$ | 0xC0 | 0x04 | 0xA8 | 0x1E | 0x6F | $0 \times 02$ | 0x1E |
| Register | Blue/Cb Coeff 1 |  | Blue/Cb Coeff 2 |  | Blue/Cb Coeff 3 |  | Blue/Cb Offset |  |
| Address | 0x45 | 0x46 | 0x47 | 0x48 | 0x49 | $0 \times 4 \mathrm{~A}$ | 0x4B | 0x4C |
| Value | 0x00 | 0x00 | 0x04 | 0xA8 | 0x08 | $0 \times 11$ | 0x1B | 0xAD |

Table 20. RGB (0 to 255) to HDTV YCrCb (0 to 255)

| Register | Red/Cr Coeff 1 |  | Red/Cr Coeff 2 |  | Red/Cr Coeff |  | Red/Cr Offset |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Address | 0x35 | 0x36 | 0x37 | 0x38 | 0x39 | 0x3A | 0x3B | 0x3C |
| Value | 0x08 | 0x2D | 0x18 | 0x93 | 0x1F | 0x3F | 0x08 | 0x00 |
| Register | Green/Y Coeff 1 |  | Green/Y Coeff 2 |  | Green/Y Coeff 3 |  | Green/Y Offset |  |
| Address | 0x3D | 0x3E | 0x3F | 0x40 | 0x41 | 0x42 | 0x43 | 0x44 |
| Value | $0 \times 03$ | 0x68 | 0x0B | 0x71 | 0x01 | 0x27 | 0x00 | 0x00 |
| Register | Blue/Cb Coeff 1 |  | Blue/Cb Coeff 2 |  | Blue/Cb Coeff 3 |  | Blue/Cb Offset |  |
| Address | 0x45 | 0x46 | 0x47 | 0x48 | 0x49 | 0x4A | 0x4B | 0x4C |
| Value | 0x1E | 0x21 | 0x19 | 0xB2 | 0x08 | 0x2D | 0x08 | 0x00 |

Table 21. RGB (0 to 255) to HDTV YCrCb (16 to 235)

| Register | Red/Cr Coeff 1 |  | Red/Cr Coeff 2 |  | Red/Cr Coeff 3 |  | Red/Cr Offset |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Address | 0x35 | 0x36 | 0x37 | 0x38 | 0x39 | 0x3A | 0x3B | 0x3C |
| Value | $0 \times 07$ | 0x06 | 0x19 | 0xA0 | $0 \times 1 \mathrm{~F}$ | 0x5B | 0x08 | 0x00 |
| Register | Green/Y Coeff 1 |  | Green/Y Coeff 2 |  | Green/Y Coeff 3 |  | Green/Y Offset |  |
| Address | 0x3D | 0x3E | 0x3F | 0x40 | 0x41 | 0x42 | 0x43 | 0x44 |
| Value | $0 \times 02$ | 0xED | 0x09 | 0xD3 | 0x00 | 0xFD | 0x01 | 0x00 |
| Register | Blue/Cb Coeff 1 |  | Blue/Cb Coeff 2 |  | Blue/Cb Coeff 3 |  | Blue/Cb Offset |  |
| Address | 0x45 | 0x46 | 0x47 | 0x48 | 0x49 | 0x4A | 0x4B | 0x4C |
| Value | 0x1E | 0x64 | $0 \times 1 \mathrm{~A}$ | 0x96 | 0x07 | 0x06 | 0x08 | 0x00 |

Table 22. RGB (0 to 255) to SDTV YCrCb (0 to 255)

| Register | Red/Cr Coeff 1 |  | Red/Cr Coeff 2 |  | Red/Cr Coeff 3 |  | Red/Cr Offset |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Address | $0 \times 35$ | 0x36 | 0x37 | 0x38 | 0x39 | 0x3A | 0x3B | 0x3C |
| Value | $0 \times 08$ | 0x2D | 0x19 | 0x27 | 0x1E | 0xAC | $0 \times 08$ | 0x00 |
| Register | Green/Y Coeff 1 |  | Green/Y Coeff 2 |  | Green/Y Coeff 3 |  | Green/Y Offset |  |
| Address | 0x3D | 0x3E | 0x3F | 0x40 | 0x41 | 0x42 | 0x43 | 0x44 |
| Value | $0 \times 04$ | 0xC9 | 0x09 | 0x64 | 0x01 | 0xD3 | 0x00 | 0x00 |
| Register | Blue/Cb Coeff 1 |  | Blue/Cb Coeff 2 |  | Blue/Cb Coeff 3 |  | Blue/Cb Offset |  |
| Address | 0x45 | 0x46 | 0x47 | 0x48 | 0x49 | 0x4A | 0x4B | 0x4C |
| Value | 0x1D | 0x3F | $0 \times 1 \mathrm{~A}$ | $0 \times 93$ | 0x08 | 0x2D | 0x08 | 0x00 |

Table 23. RGB (0 to 255) to SDTV YCrCb ( 16 to 235)

| Register | Red/Cr Coeff 1 |  | Red/Cr Coeff 2 |  | Red/Cr Coeff 3 |  | Red/Cr Offset |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Address | 0x35 | 0x36 | 0x37 | 0x38 | 0x39 | 0x3A | 0x3B | 0x3C |
| Value | $0 \times 07$ | 0x06 | $0 \times 1 \mathrm{~A}$ | 0x1E | 0x1E | 0xDC | 0x08 | 0x00 |
| Register | Green/Y Coeff 1 |  | Green/Y Coeff 2 |  | Green/Y Coeff 3 |  | Green/Y Offset |  |
| Address | 0x3D | 0x3E | 0x3F | 0x40 | 0x41 | 0x42 | 0x43 | 0x44 |
| Value | $0 \times 04$ | 0x1C | 0x08 | 0x11 | 0x01 | 0x91 | $0 \times 01$ | 0x00 |
| Register | Blue/Cb Coeff 1 |  | Blue/Cb Coeff 2 |  | Blue/Cb Coeff 3 |  | Blue/Cb Offset |  |
| Address | 0x45 | 0x46 | 0x47 | 0x48 | 0x49 | 0x4A | 0x4B | 0x4C |
| Value | 0x1D | 0xA3 | $0 \times 1 \mathrm{~B}$ | 0x57 | 0x07 | 0x06 | 0x08 | 0x00 |

## AD9397

## OUTLINE DIMENSIONS



COMPLIANT TO JEDEC STANDARDS MS-026-BED
Figure 9. 100-Lead Low Profile Quad Flat Package [LQFP] (ST-100)
Dimensions shown in millimeters

ORDERING GUIDE

| Model | Max Speeds (MHz) |  | Temperature <br> Range | Package Description | Package Option |
| :---: | :---: | :---: | :---: | :---: | :---: |
|  | Analog | Digital |  |  |  |
| AD9397KSTZ-100 ${ }^{1}$ | 100 | 100 | $0^{\circ} \mathrm{C}$ to $70^{\circ} \mathrm{C}$ | 100-Lead Low Profile Quad Flat Package (LQFP) | ST-100 |
| AD9397KSTZ-150¹ | 150 | 150 | $0^{\circ} \mathrm{C}$ to $70^{\circ} \mathrm{C}$ | 100-Lead Low Profile Quad Flat Package (LQFP) | ST-100 |
| AD9397/PCB |  |  |  | Evaluation Board |  |

${ }^{1} \mathrm{Z}=\mathrm{Pb}$-free part.


[^0]:    DATACK load $=15 \mathrm{pF}$, data load $=5 \mathrm{pF}$.
    ${ }^{2}$ Specified current and power values with a worst-case pattern (on/off).

[^1]:    ${ }^{1}$ The typical pattern contains a gray scale area, output drive $=$ high. Worst-case pattern is alternating black and white pixels.
    ${ }^{2}$ The typical pattern contains a gray scale area, output drive $=$ high.
    ${ }^{3}$ Specified current and power values with a worst-case pattern (on/off)
    ${ }^{4}$ DATACK load $=10 \mathrm{pF}$, data load $=5 \mathrm{pF}$.
    ${ }^{5}$ Drive strength $=$ high.

[^2]:    ${ }^{1}$ The supplies should be sequenced such that $V_{D}$ and $V_{D D}$ are never less than 300 mV below $\mathrm{DV}_{\mathrm{DD}}$. At no time should $\mathrm{DV} \mathrm{V}_{\mathrm{DD}}$ be more than 300 mV greater than $\mathrm{V}_{D}$ or $\mathrm{V}_{\mathrm{DD}}$.

[^3]:    ${ }^{1}$ Power-down is controlled via Bit 0 in Serial Bus Register 0x26.
    ${ }^{2}$ Sync detect is determined by OR'ing Bit 7 to Bit 2 in Serial Bus Register 0x15.
    ${ }^{3}$ Auto power-down is controlled via Bit 7 in Serial Bus Register 0x27.

[^4]:    ${ }^{1}$ Arrows in the table indicate clock edge. Rising edge of clock $=\uparrow$, falling edge $=\downarrow$.

