

# PCA85162 32 x 4 automotive LCD driver for low multiplex rates Rev. 4 – 9 April 2015 Product

**Product data sheet** 

## 1. General description

The PCA85162 is a peripheral device which interfaces to almost any Liquid Crystal Display (LCD)<sup>1</sup> with low multiplex rates. It generates the drive signals for any static or multiplexed LCD containing up to four backplanes and up to 32 segments. It can be easily cascaded for larger LCD applications. The PCA85162 is compatible with most microcontrollers and communicates via the two-line bidirectional I<sup>2</sup>C-bus. Communication overheads are minimized by a display RAM with auto-incremented addressing, by hardware subaddressing, and by display memory switching (static and duplex drive modes).

For a selection of NXP LCD segment drivers, see <u>Table 25 on page 48</u>.

## 2. Features and benefits

- AEC-Q100 compliant for automotive applications
- Single chip LCD controller and driver
- Selectable backplane drive configuration: static, 2, 3, or 4 backplane multiplexing
- Selectable display bias configuration: static, <sup>1</sup>/<sub>2</sub>, or <sup>1</sup>/<sub>3</sub>
- Internal LCD bias generation with voltage-follower buffers
- 32 segment drives:
  - Up to 16 7-segment numeric characters
  - Up to 8 14-segment alphanumeric characters
  - Any graphics of up to 128 segments/elements
- 32 × 4-bit RAM for display data storage
- Display memory bank switching in static and duplex drive modes
- Versatile blinking modes
- Independent supplies possible for LCD and logic voltages
- Wide power supply range: from 1.8 V to 5.5 V
- Wide logic LCD supply range:
  - From 2.5 V for low-threshold LCDs
  - ◆ Up to 8.0 V for guest-host LCDs and high-threshold twisted nematic LCDs
- Low power consumption
- Extended temperature range up to 95 °C
- 400 kHz I<sup>2</sup>C-bus interface
- No external components required
- Manufactured in silicon gate CMOS process

<sup>1.</sup> The definition of the abbreviations and acronyms used in this data sheet can be found in Section 22.



## 3. Ordering information

Table 1. Ordering info	Table 1. Ordering information							
Type number								
	Name	Description	Version					
PCA85162T		plastic thin shrink small outline package; 48 leads; body width 6.1 mm	SOT362-1					

### 3.1 Ordering options

#### Table 2. Ordering options

Product type number	Orderable part number	Sales item (12NC)	Delivery form	IC revision
PCA85162T/Q900/1	PCA85162T/Q900/1,1	935291388118	tape and reel, 13 inch	1

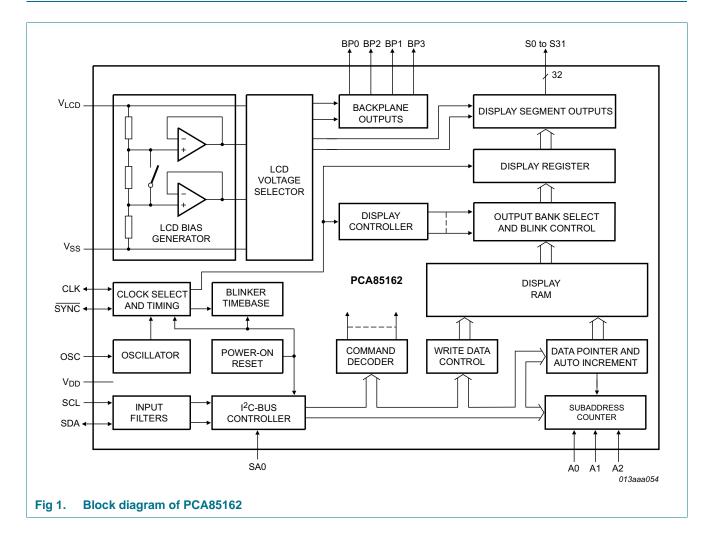
## 4. Marking

#### Table 3. Marking codes

Product type number	Marking code
PCA85162T/Q900/1	PCA85162T

32 x 4 automotive LCD driver for low multiplex rates

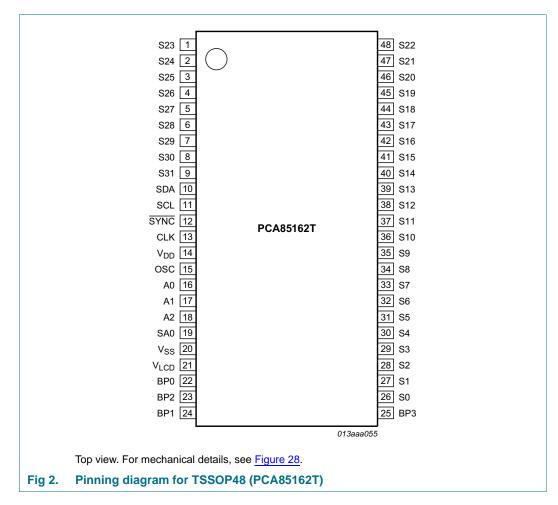
### 5. Block diagram



32 x 4 automotive LCD driver for low multiplex rates

### 6. Pinning information

### 6.1 Pinning



### 6.2 Pin description

#### Table 4. Pin description

Input or input/output pins must always be at a defined level (V<sub>SS</sub> or V<sub>DD</sub>) unless otherwise specified.

Symbol	Pin	Туре	Description
SDA	10	input/output	I <sup>2</sup> C-bus serial data line
SCL	11	input	I <sup>2</sup> C-bus serial clock
SYNC	12	input/output	cascade synchronization input or output; if not used it must be left open
CLK	13	input/output	clock line
V <sub>DD</sub>	14	supply	supply voltage
OSC	15	input	internal oscillator enable
A0 to A2	16 to 18	input	subaddress inputs
SA0	19	input	I <sup>2</sup> C-bus address input
V <sub>SS</sub>	20	supply	ground supply voltage
V <sub>LCD</sub>	21	supply	LCD supply voltage
BP0 to BP3	22 to 25	output	LCD backplane outputs
S0 to S22, S23 to S31	26 to 48, 1 to 9	output	LCD segment outputs

#### **Functional description** 7.

The PCA85162 is a versatile peripheral device designed to interface between any microcontroller to a wide variety of LCD segment or dot matrix displays. It can directly drive any static or multiplexed LCD containing up to four backplanes and up to 32 segments.

### 7.1 Commands of PCA85162

The commands available to the PCA85162 are defined in Table 5.

#### Table 5. **Definition of PCA85162 commands**

Bit position labeled as - is not used.

Command	Opera	Operation code							Reference
Bit	7	6	5	4	3	2	1	0	-
mode-set	С	1	0	-	E	В	M[1:0]		Table 7
load-data-pointer	С	0	0	P[4:0] <u>Table 8</u>			Table 8		
device-select	С	1	1	0 0 A[2:0] <u>Table 9</u>			Table 9		
bank-select	С	1	1	1	1	0	I	0	Table 10
blink-select	С	1	1	1	0	AB	BF[1:0	]	Table 11

All available commands carry a continuation bit C in their most significant bit position as shown in Figure 21. When this bit is set logic 1, it indicates that the next byte of the transfer to arrive will also represent a command. If this bit is set logic 0, it indicates that the command byte is the last in the transfer. Further bytes will be regarded as display data (see Table 6).

#### Table 6. **C** bit description

Bit	Symbol	Value	Description
7	С		continue bit
		0	last control byte in the transfer; next byte will be regarded as display data
		1	control bytes continue; next byte will be a command too

PCA85162 **Product data sheet** 

#### 7.1.1 Command: mode-set

The mode-set command allows configuring the multiplex mode, the bias levels and enabling or disabling the display.

Mode-set	Mode-set command bit description				
Symbol	Value	Description			
С	0, 1	see Table 6			
-	10	fixed value			
-	-	unused			
E		display status <sup>[1]</sup>			
	0[2]	disabled (blank) <sup>[3]</sup>			
	1	enabled			
2 B		LCD bias configuration <sup>[4]</sup>			
	0[2]	1⁄3 bias			
	1	$\frac{1}{2}$ bias			
M[1:0]		LCD drive mode selection			
	01	static; BP0			
	10	1:2 multiplex; BP0, BP1			
	11	1:3 multiplex; BP0, BP1, BP2			
	00 <mark>[2]</mark>	1:4 multiplex; BP0, BP1, BP2, BP3			
	Symbol C - E B	Symbol         Value           C         0, 1           -         10           -         -           E         0[2]           1         0[2]           1         0[2]           1         0[2]           1         0[2]           1         0[2]           1         1			

Table 7.	Mode-set	command	bit	descri	ntion
	Mode-Set	command	DIL	ueschi	puon

[1] The possibility to disable the display allows implementation of blinking under external control.

- [2] Default value.
- [3] The display is disabled by setting all backplane and segment outputs to  $V_{LCD}$ .
- [4] Not applicable for static drive mode.

#### 7.1.2 Command: load-data-pointer

The load-data-pointer command defines the display RAM address where the following display data will be sent to.

Table 8.Load-data-pointer command bit descriptionSee Section 7.6.1.

Bit	Symbol	Value	Description
7	С	0, 1	see <u>Table 6</u>
6 to 5	-	00	fixed value
4 to 0	P[4:0]	00000[ <u>1]</u> to 11111	5 bit binary value, 0 to 31; transferred to the data pointer to define one of 32 display RAM addresses

[1] Default value.

PCA85162 Product data sheet

#### 7.1.3 Command: device-select

The device-select command allows defining the subaddress counter value.

Table 9.	Device-select command bit description
See Section	n 7.6.2.

Bit	Symbol	Value	Description
7	С	0, 1	see <u>Table 6</u>
6 to 3	-	1100	fixed value
2 to 0	A[2:0]	000 <u><sup>[1]</sup></u> to 111	3 bit binary value, 0 to 7; transferred to the subaddress counter to define one of eight hardware subaddresses

[1] Default value.

#### 7.1.4 Command: bank-select

The bank-select command controls where data is written to RAM and where it is displayed from.

# Table 10. Bank-select command bit description See Section 7.6.5.

Bit	Symbol	Value	Description			
			Static	1:2 multiplex <sup>[1]</sup>		
7	С	0, 1	see <u>Table 6</u>	see Table 6		
6 to 2	-	11110	fixed value	fixed value		
1	I		input bank selection; storage	input bank selection; storage of arriving display data		
		0[2]	RAM row 0	RAM rows 0 and 1		
		1	RAM row 2	RAM rows 2 and 3		
0	0		output bank selection; retrieval of LCD display data			
		0 <u>[2]</u>	RAM row 0	RAM rows 0 and 1		
l		1	RAM row 2	RAM rows 2 and 3		

[1] The bank-select command has no effect in 1:3 and 1:4 multiplex drive modes.

[2] Default value.

#### 7.1.5 Command: blink-select

The blink-select command allows configuring the blink mode and the blink frequency.

See <u>Section</u>	<u>n 7.1.5.1</u> .		
Bit	Symbol	Value	Description
7	С	0, 1	see <u>Table 6</u>
6 to 3	-	1110	fixed value
2	AB		blink mode selection
		0[1]	normal blinking <sup>[2]</sup>
		1	alternate RAM bank blinking <sup>[3]</sup>
1 to 0	BF[1:0]		blink frequency selection
		00 <mark>[1]</mark>	off
		01	1
		10	2
		11	3

Table 11.	<b>Blink-select</b>	command	bit	description
See Section	7151			

- [1] Default value.
- [2] Normal blinking is assumed when the LCD multiplex drive modes 1:3 or 1:4 are selected.
- [3] Alternate RAM bank blinking does not apply in 1:3 and 1:4 multiplex drive modes.

#### 7.1.5.1 Blinking

The display blinking capabilities of the PCA85162 are very versatile. The whole display can blink at frequencies selected by the blink-select command (see <u>Table 11</u>). The blink frequencies are derived from the clock frequency. The ratio between the clock and blink frequencies depends on the blink mode selected (see <u>Table 12</u>).

An additional feature is for an arbitrary selection of LCD segments/elements to blink. This applies to the static and 1:2 multiplex drive modes and can be implemented without any communication overheads. By means of the output bank selector, the displayed RAM banks are exchanged with alternate RAM banks at the blink frequency. This mode can also be specified by the blink-select command.

In the 1:3 and 1:4 multiplex modes, where no alternative RAM bank is available, groups of LCD segments/elements can blink by selectively changing the display RAM data at fixed time intervals.

The entire display can blink at a frequency other than the nominal blink frequency. This can be effectively performed by resetting and setting the display enable bit E at the required rate using the mode-set command (see <u>Table 7</u>).

Table	12.	Blink	frequen	cies
-------	-----	-------	---------	------

Blink mode	Blink frequency equation <sup>[1]</sup>
off	-
1	$f_{blink} = \frac{f_{clk}}{768}$
2	$f_{blink} = \frac{f_{clk}}{1536}$
3	$f_{blink} = \frac{f_{clk}}{3072}$

[1] The blink frequency is proportional to the clock frequency ( $f_{clk}$ ). For the range of the clock frequency see <u>Table 20</u>.

### 7.2 Power-On Reset (POR)

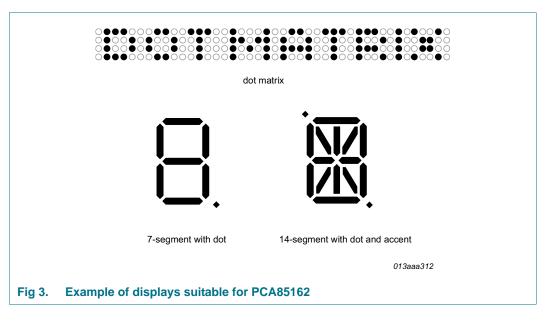
At power-on the PCA85162 resets to the following starting conditions:

- All backplane and segment outputs are set to V<sub>LCD</sub>
- The selected drive mode is: 1:4 multiplex with  $\frac{1}{3}$  bias
- Blinking is switched off
- Input and output bank selectors are reset
- The I<sup>2</sup>C-bus interface is initialized
- The data pointer and the subaddress counter are cleared (set to logic 0)
- The display is disabled (bit E = 0, see <u>Table 7</u>)

**Remark:** Do not transfer data on the  $I^2C$ -bus for at least 1 ms after a power-on to allow the reset action to complete.

#### 7.3 Possible display configurations

The possible display configurations of the PCA85162 depend on the number of active backplane outputs required. A selection of display configurations is shown in <u>Table 13</u>. All of these configurations can be implemented in the typical system shown in <u>Figure 4</u>.



#### Table 13. Selection of possible display configurations

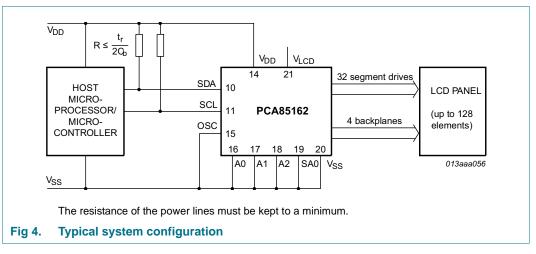
Number of				
Backplanes	Icons	Digits/Characte	rs	Dot matrix:
		7-segment <sup>[1]</sup>	14-segment <sup>[2]</sup>	segments/ elements
4	128	16	8	128 dots (4 × 32)
3	96	12	6	96 dots (3 × 32)
2	64	8	4	64 dots (2 × 32)
1	32	4	2	32 dots (1 × 32)

[1] 7 segment display has 8 segments/elements including the decimal point.

[2] 14 segment display has 16 segments/elements including decimal point and accent dot.

11 of 56

32 x 4 automotive LCD driver for low multiplex rates



The host microcontroller maintains the 2-line I<sup>2</sup>C-bus communication channel with the PCA85162. The internal oscillator is enabled by connecting pin OSC to pin V<sub>SS</sub>. The appropriate biasing voltages for the multiplexed LCD waveforms are generated internally. The only other connections required to complete the system are the power supplies (V<sub>DD</sub>, V<sub>SS</sub>, and V<sub>LCD</sub>) and the LCD panel chosen for the application.

#### 7.3.1 LCD bias generator

Fractional LCD biasing voltages are obtained from an internal voltage divider consisting of three impedances connected in series between V<sub>LCD</sub> and V<sub>SS</sub>. The center impedance is bypassed by switch if the  $\frac{1}{2}$  bias voltage level for the 1:2 multiplex drive mode configuration is selected. The LCD voltage can be temperature compensated externally, using the supply to pin V<sub>LCD</sub>.

#### 7.3.2 Display register

The display register holds the display data while the corresponding multiplex signals are generated.

#### 7.3.3 LCD voltage selector

The LCD voltage selector coordinates the multiplexing of the LCD in accordance with the selected LCD drive configuration. The operation of the voltage selector is controlled by the mode-set command from the command decoder. The biasing configurations that apply to the preferred modes of operation, together with the biasing characteristics as functions of V<sub>LCD</sub> and the resulting discrimination ratios (D) are given in Table 14.

Discrimination is a term which is defined as the ratio of the on and off RMS voltage across a segment. It can be thought of as a measurement of contrast.

12 of 56

Table 14.Biasing characteristics

LCD drive mode static 1:2 multiplex 1:2 multiplex	Number of:		LCD bias	V <sub>off(RMS)</sub>	V <sub>on(RMS)</sub>	$V_{on(RMS)}$	
mode	Backplanes	Levels	configuration	V <sub>LCD</sub>	V <sub>LCD</sub>	$D = \frac{on(RMS)}{V_{off(RMS)}}$	
static	1	2	static	0	1	x	
1:2 multiplex	2	3	1/2	0.354	0.791	2.236	
1:2 multiplex	2	4	1/3	0.333	0.745	2.236	
1:3 multiplex	3	4	1/3	0.333	0.638	1.915	
1:4 multiplex	4	4	1/3	0.333	0.577	1.732	

A practical value for V<sub>LCD</sub> is determined by equating V<sub>off(RMS)</sub> with a defined LCD threshold voltage (V<sub>th(off)</sub>), typically when the LCD exhibits approximately 10 % contrast. In the static drive mode a suitable choice is V<sub>LCD</sub> >  $3V_{th(off)}$ .

Multiplex drive modes of 1:3 and 1:4 with  $\frac{1}{2}$  bias are possible but the discrimination and hence the contrast ratios are smaller.

Bias is calculated by  $\frac{1}{1+a}$ , where the values for a are

a = 1 for  $\frac{1}{2}$  bias

a = 2 for  $\frac{1}{3}$  bias

The RMS on-state voltage (Von(RMS)) for the LCD is calculated with Equation 1:

$$V_{on(RMS)} = v_{LCD} \sqrt{\frac{a^2 + 2a + n}{n \times (1 + a)^2}}$$
(1)

where the values for n are

n = 1 for static drive mode

n = 2 for 1:2 multiplex drive mode

n = 3 for 1:3 multiplex drive mode

n = 4 for 1:4 multiplex drive mode

The RMS off-state voltage (Voff(RMS)) for the LCD is calculated with Equation 2:

$$V_{off(RMS)} = V_{LCD} \sqrt{\frac{a^2 - 2a + n}{n \times (1 + a)^2}}$$
 (2)

Discrimination is the ratio of V<sub>on(RMS)</sub> to V<sub>off(RMS)</sub> and is determined from Equation 3:

$$D = \frac{V_{on(RMS)}}{V_{off(RMS)}} = \sqrt{\frac{a^2 + 2a + n}{a^2 - 2a + n}}$$
(3)

Using Equation 3, the discrimination for an LCD drive mode of 1:3 multiplex with  $\frac{1}{2}$  bias is  $\sqrt{3} = 1.732$  and the discrimination for an LCD drive mode of 1:4 multiplex with  $\frac{1}{2}$  bias is  $\frac{\sqrt{21}}{3} = 1.528$ .

PCA85162

#### 32 x 4 automotive LCD driver for low multiplex rates

The advantage of these LCD drive modes is a reduction of the LCD full scale voltage V<sub>LCD</sub> as follows:

- 1:3 multiplex ( $\frac{1}{2}$  bias):  $V_{LCD} = \sqrt{6} \times V_{off(RMS)} = 2.449 V_{off(RMS)}$
- 1:4 multiplex (1/<sub>2</sub> bias):  $V_{LCD} = \left[\frac{(4 \times \sqrt{3})}{3}\right] = 2.309 V_{off(RMS)}$

These compare with  $V_{LCD} = 3V_{off(RMS)}$  when  $\frac{1}{3}$  bias is used.

It should be noted that  $V_{LCD}$  is sometimes referred as the LCD operating voltage.

#### 7.3.3.1 **Electro-optical performance**

Suitable values for Von(RMS) and Voff(RMS) are dependent on the LCD liquid used. The RMS voltage, at which a pixel will be switched on or off, determine the transmissibility of the pixel.

For any given liquid, there are two threshold values defined. One point is at 10 % relative transmission (at V<sub>th(off)</sub>) and the other at 90 % relative transmission (at V<sub>th(on)</sub>), see Figure 5. For a good contrast performance, the following rules should be followed:

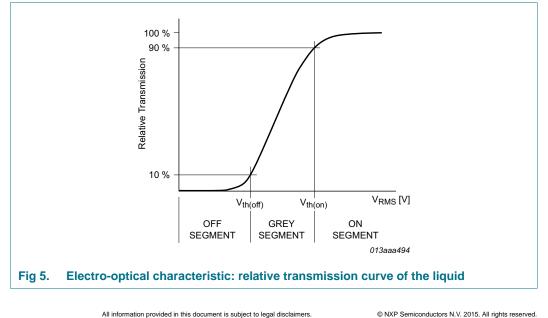
$$V_{on(RMS)} \ge V_{th(on)} \tag{4}$$

$$V_{off(RMS)} \le V_{th(off)} \tag{5}$$

Von(RMS) and Voff(RMS) are properties of the display driver and are affected by the selection of a, n (see Equation 1 to Equation 3) and the  $V_{LCD}$  voltage.

 $V_{th(off)}$  and  $V_{th(on)}$  are properties of the LCD liquid and can be provided by the module manufacturer. V<sub>th(off)</sub> is sometimes just named V<sub>th</sub>. V<sub>th(on)</sub> is sometimes named saturation voltage V<sub>sat</sub>.

It is important to match the module properties to those of the driver in order to achieve optimum performance.



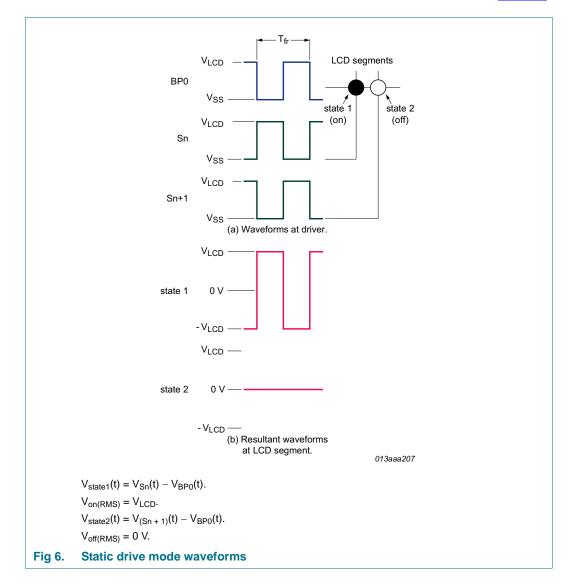
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PCA85162

#### 7.3.4 LCD drive mode waveforms

#### 7.3.4.1 Static drive mode

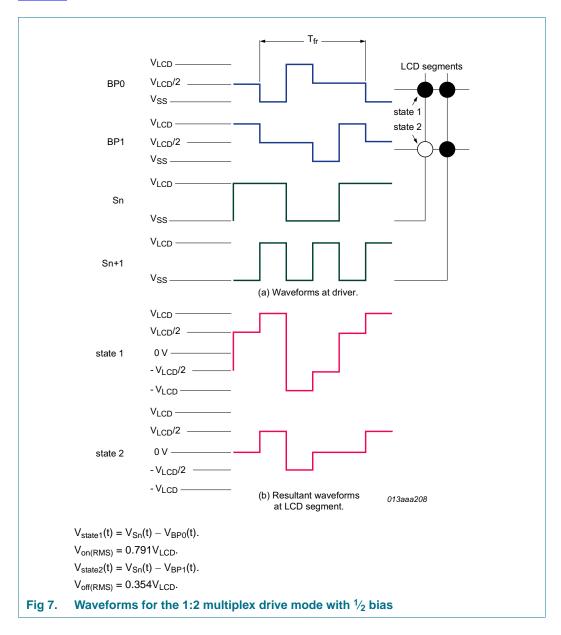
The static LCD drive mode is used when a single backplane is provided in the LCD. The backplane (BPn) and segment (Sn) drive waveforms for this mode are shown in Figure 6.



#### 32 x 4 automotive LCD driver for low multiplex rates

#### 7.3.4.2 1:2 Multiplex drive mode

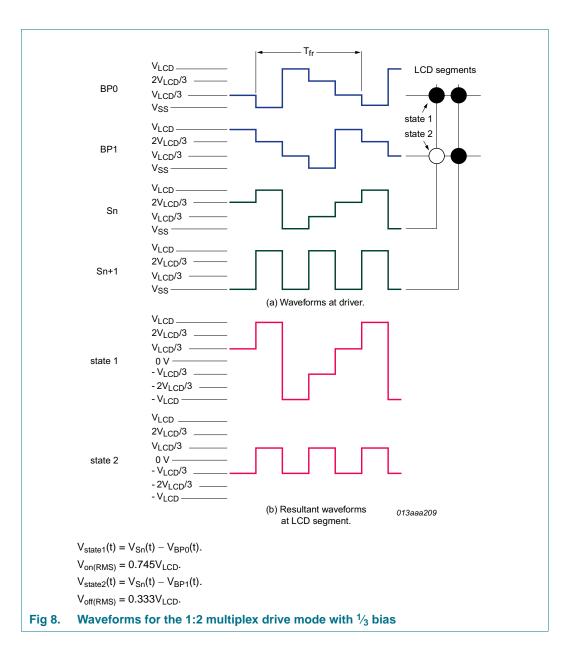
When two backplanes are provided in the LCD, the 1:2 multiplex mode applies. The PCA85162 allows the use of  $\frac{1}{2}$  bias or  $\frac{1}{3}$  bias in this mode as shown in Figure 7 and Figure 8.



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# PCA85162

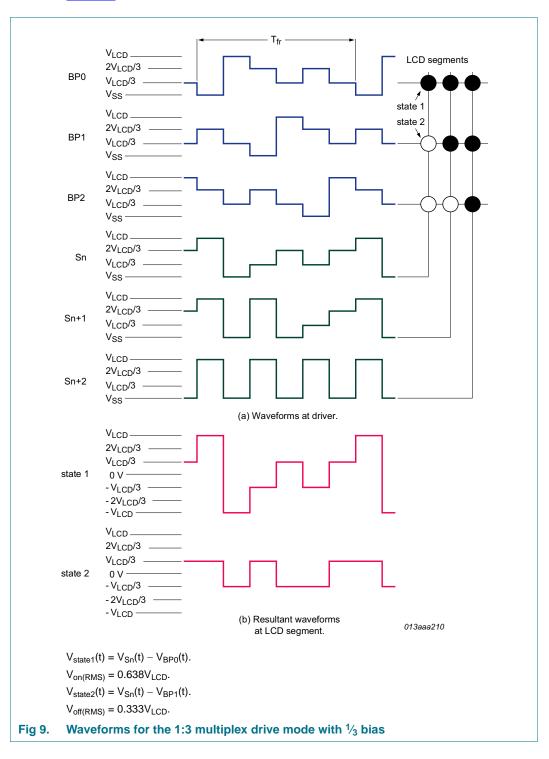
32 x 4 automotive LCD driver for low multiplex rates



#### 32 x 4 automotive LCD driver for low multiplex rates

#### 7.3.4.3 1:3 Multiplex drive mode

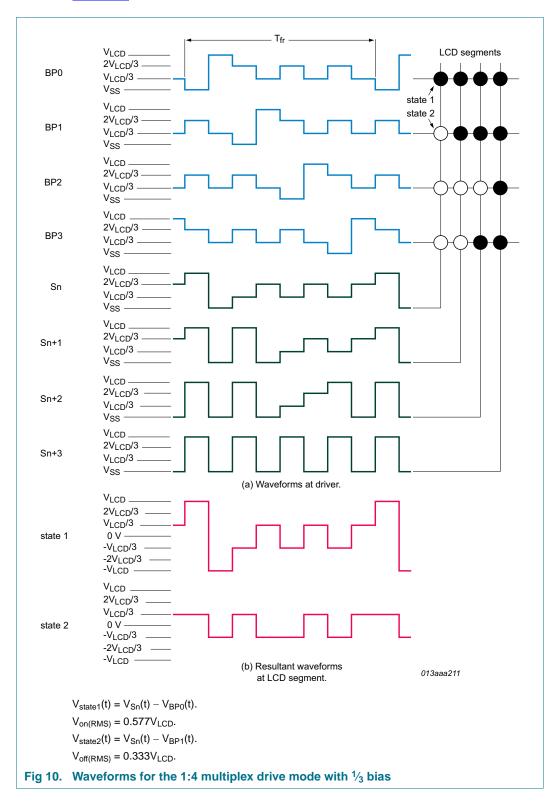
When three backplanes are provided in the LCD, the 1:3 multiplex drive mode applies, as shown in Figure 9.



#### 32 x 4 automotive LCD driver for low multiplex rates

#### 7.3.4.4 1:4 Multiplex drive mode

When four backplanes are provided in the LCD, the 1:4 multiplex drive mode applies as shown in Figure 10.



#### 7.4 Oscillator

#### 7.4.1 Internal clock

The internal logic of the PCA85162 and its LCD drive signals are timed either by its internal oscillator or by an external clock. The internal oscillator is enabled by connecting pin OSC to pin  $V_{SS}$ . If the internal oscillator is used, the output from pin CLK can be used as the clock signal for several PCA85162 in the system that are connected in cascade.

#### 7.4.2 External clock

Pin CLK is enabled as an external clock input by connecting pin OSC to  $V_{DD}$ . The LCD frame frequency is determined by the clock frequency ( $f_{clk}$ ).

**Remark:** A clock signal must always be supplied to the device; removing the clock may freeze the LCD in a DC state, which is not suitable for the liquid crystal.

#### 7.4.3 Timing

The PCA85162 timing controls the internal data flow of the device. This includes the transfer of display data from the display RAM to the display segment outputs. In cascaded applications, the correct timing relationship between each PCA85162 in the system is maintained by the synchronization signal at pin SYNC. The timing also generates the LCD frame frequency signal. The frame frequency signal is a fixed division of the clock

frequency from either the internal or an external clock:  $f_{fr} = \frac{f_{clk}}{24}$ 

### 7.5 Backplane and segment outputs

#### 7.5.1 Backplane outputs

The LCD drive section includes four backplane outputs BP0 to BP3 which must be connected directly to the LCD. The backplane output signals are generated in accordance with the selected LCD drive mode. If less than four backplane outputs are required, the unused outputs can be left open-circuit.

- In 1:3 multiplex drive mode, BP3 carries the same signal as BP1, therefore these two adjacent outputs can be tied together to give enhanced drive capabilities
- In 1:2 multiplex drive mode, BP0 and BP2, respectively, BP1 and BP3 carry the same signals and may also be paired to increase the drive capabilities
- In static drive mode the same signal is carried by all four backplane outputs and they can be connected in parallel for very high drive requirements

#### 7.5.2 Segment outputs

The LCD drive section includes 32 segment outputs (S0 to S31) which should be connected directly to the LCD. The segment output signals are generated in accordance with the multiplexed backplane signals and with data residing in the display register. When less than 32 segment outputs are required, the unused segment outputs should be left open-circuit.

#### 7.6 Display RAM

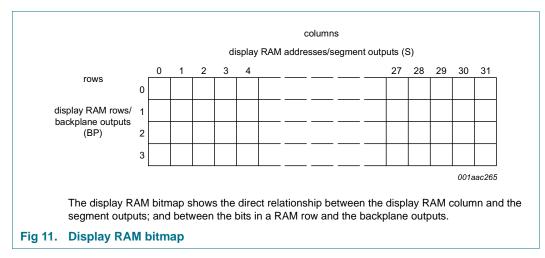
The display RAM is a static  $32 \times 4$ -bit RAM which stores LCD data.

There is a one-to-one correspondence between

- the bits in the RAM bitmap and the LCD segments/elements
- the RAM columns and the segment outputs
- the RAM rows and the backplane outputs.

A logic 1 in the RAM bitmap indicates the on-state of the corresponding LCD element; similarly, a logic 0 indicates the off-state.

The display RAM bitmap, Figure 11, shows the rows 0 to 3 which correspond with the backplane outputs BP0 to BP3, and the columns 0 to 31 which correspond with the segment outputs S0 to S31. In multiplexed LCD applications the segment data of the first, second, third, and fourth row of the display RAM are time-multiplexed with BP0, BP1, BP2, and BP3 respectively.



When display data is transmitted to the PCA85162, the display bytes received are stored in the display RAM in accordance with the selected LCD drive mode. The data is stored as it arrives and depending on the current multiplex drive mode the bits are stored singularly, in pairs, triples or quadruples. To illustrate the filling order, an example of a 7-segment numeric display showing all drive modes is given in Figure 12; the RAM filling organization depicted applies equally to other LCD types.

- In static drive mode the eight transmitted data bits are placed into row 0 as one byte
- In 1:2 multiplex drive mode the eight transmitted data bits are placed in pairs into row 0 and 1 as four successive 2-bit RAM words
- In 1:3 multiplex drive mode the eight bits are placed in triples into row 0, 1, and 2 as three successive 3-bit RAM words, with bit 3 of the third address left unchanged. It is not recommended to use this bit in a display because of the difficult addressing. This last bit may, if necessary, be controlled by an additional transfer to this address, but care should be taken to avoid overwriting adjacent data because always full bytes are transmitted (see Section 7.6.4)
- In 1:4 multiplex drive mode, the eight transmitted data bits are placed in quadruples into row 0, 1, 2, and 3 as two successive 4-bit RAM words

PCA85162

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PCA85162 Product data sheet

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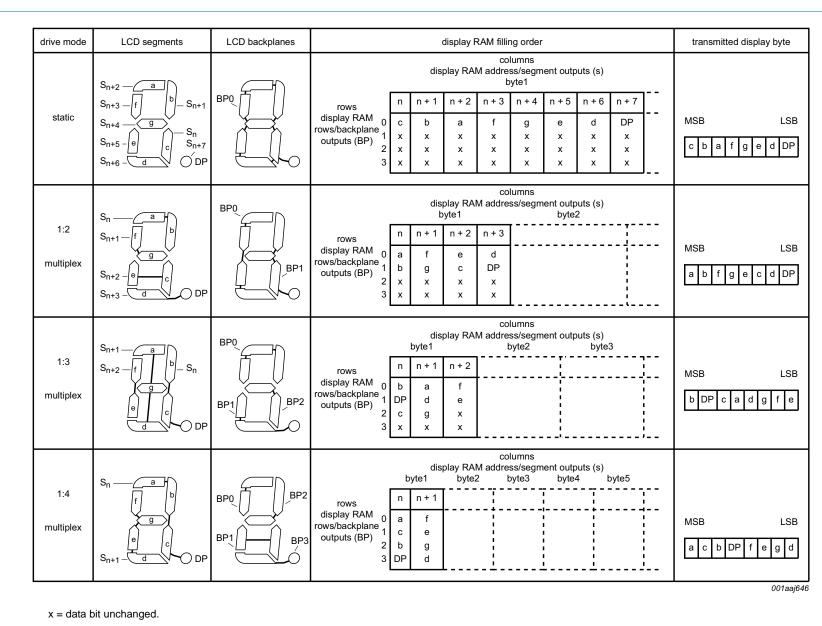


Fig 12. Relationship between LCD layout, drive mode, display RAM filling order, and display data transmitted over the l<sup>2</sup>C-bus

PCA85162 x 4 automotive LCD driver for low multiplex rates

32

#### 7.6.1 Data pointer

The addressing mechanism for the display RAM is realized using the data pointer. This allows the loading of an individual display data byte, or a series of display data bytes, into any location of the display RAM. The sequence commences with the initialization of the data pointer by the load-data-pointer command (see <u>Table 8</u>). Following this command, an arriving data byte is stored at the display RAM address indicated by the data pointer. The filling order is shown in Figure 12.

After each byte is stored, the content of the data pointer is automatically incremented by a value dependent on the selected LCD drive mode:

- In static drive mode by eight
- In 1:2 multiplex drive mode by four
- In 1:3 multiplex drive mode by three
- In 1:4 multiplex drive mode by two

If an I<sup>2</sup>C-bus data access terminates early then the state of the data pointer is unknown. Consequently, the data pointer must be rewritten prior to further RAM accesses.

#### 7.6.2 Subaddress counter

The storage of display data is determined by the contents of the subaddress counter. Storage is allowed only when the content of the subaddress counter match with the hardware subaddress applied to A0, A1, and A2. The subaddress counter value is defined by the device-select command (see <u>Table 9</u>). If the content of the subaddress counter and the hardware subaddress do not match then data storage is inhibited but the data pointer is incremented as if data storage had taken place. The subaddress counter is also incremented when the data pointer overflows.

#### 7.6.3 RAM addressing in cascaded applications

In cascaded applications each PCA85162 in the cascade must be addressed separately. Initially, the first PCA85162 is selected by sending the device-select command matching the first device's hardware subaddress. Then the data pointer is set to the preferred display RAM address by sending the load-data-pointer command.

Once the display RAM of the first PCA85162 has been written, the second PCA85162 is selected by sending the device-select command again. This time however the command matches the second device's hardware subaddress. Next the load-data-pointer command is sent to select the preferred display RAM address of the second PCA85162.

This last step is very important because during writing data to the first PCA85162, the data pointer of the second PCA85162 is incremented. In addition, the hardware subaddress should not be changed whilst the device is being accessed on the I<sup>2</sup>C-bus interface.

PCA85162

#### 7.6.4 RAM writing in 1:3 multiplex drive mode

In 1:3 multiplex drive mode, the RAM is written as shown in <u>Table 15</u> (see <u>Figure 12</u> as well).

#### Table 15. Standard RAM filling in 1:3 multiplex drive mode

Assumption: BP2/S2, BP2/S5, BP2/S8 etc. **are not connected** to any segments/elements on the display.

Display RAM bits (rows)/ backplane outputs (BPn)	Displa	Display RAM addresses (columns)/segment outputs (Sn)											
	0	1	2	3	4	5	6	7	8	9	:		
0	a7	a4	a1	b7	b4	b1	c7	c4	c1	d7	:		
1	a6	a3	a0	b6	b3	b0	c6	c3	c0	d6	:		
2	a5	a2	-	b5	b2	-	c5	c2	-	d5	:		
3	-	-	-	-	-	-	-	-	-	-	:		

If the bit at position BP2/S2 would be written by a second byte transmitted, then the mapping of the segment bits would change as illustrated in Table 16.

# Table 16.Entire RAM filling by rewriting in 1:3 multiplex drive modeAssumption: BP2/S2, BP2/S5, BP2/S8 etc. are connected to segments/elements on the display.

	Displa	y RAM	addres	sses (c	olumns	s)/segm	nent ou	tputs (\$	Sn)		
bits (rows)/ backplane outputs (BPn)	0	1	2	3	4	5	6	7	8	9	:
0	a7	a4	a1/b7	b4	b1/c7	c4	c1/d7	d4	d1/e7	e4	:
1	a6	a3	a0/b6	b3	b0/c6	c3	c0/d6	d3	d0/e6	e3	:
2	a5	a2	b5	b2	c5	c2	d5	d2	e5	e2	:
3	-	-	-	-	-	-	-	-	-	-	:

In the case described in <u>Table 16</u> the RAM has to be written entirely and BP2/S2, BP2/S5, BP2/S8 etc. have to be connected to segments/elements on the display. This can be achieved by a combination of writing and rewriting the RAM like follows:

- In the first write to the RAM, bits a7 to a0 are written
- The data-pointer (see Section 7.6.1 on page 23) has to be set to the address of bit a1
- In the second write, bits b7 to b0 are written, overwriting bits a1 and a0 with bits b7 and b6
- The data-pointer has to be set to the address of bit b1
- In the third write, bits c7 to c0 are written, overwriting bits b1 and b0 with bits c7 and c6

Depending on the method of writing to the RAM (standard or entire filling by rewriting), some segments/elements remain unused or can be used, but it has to be considered in the module layout process as well as in the driver software design.

#### 7.6.5 Bank selection

#### 7.6.5.1 Output bank selector

The output bank selector (see <u>Table 10</u>) selects one of the four rows per display RAM address for transfer to the display register. The actual row selected depends on the particular LCD drive mode in operation and on the instant in the multiplex sequence.

- In 1:4 multiplex mode, all RAM addresses of row 0 are selected, these are followed by the contents of row 1, 2, and then 3
- In 1:3 multiplex mode, rows 0, 1, and 2 are selected sequentially
- In 1:2 multiplex mode, rows 0 and 1 are selected
- In static mode, row 0 is selected

The PCA85162 includes a RAM bank switching feature in the static and 1:2 multiplex drive modes. In the static drive mode, the bank-select command may request the contents of row 2 to be selected for display instead of the contents of row 0. In the 1:2 multiplex mode, the contents of rows 2 and 3 may be selected instead of rows 0 and 1. This gives the provision for preparing display information in an alternative bank and to be able to switch to it once it is assembled.

#### 7.6.5.2 Input bank selector

The input bank selector loads display data into the display RAM in accordance with the selected LCD drive configuration. Display data can be loaded in row 2 in static drive mode or in rows 2 and 3 in 1:2 multiplex drive mode by using the bank-select command (see Table 10). The input bank selector functions independently to the output bank selector.

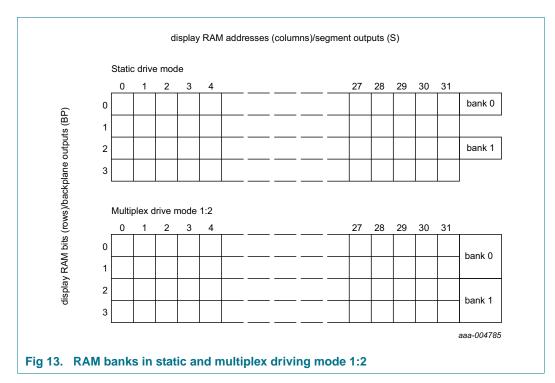
#### 7.6.5.3 RAM bank switching

The PCA85162 includes a RAM bank switching feature in the static and 1:2 multiplex drive modes. A bank can be thought of as one RAM row or a collection of RAM rows (see <u>Figure 13</u>). The RAM bank switching gives the provision for preparing display information in an alternative bank and to be able to switch to it once it is complete.

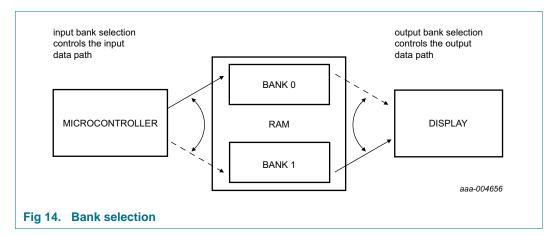
PCA85162

25 of 56

#### 32 x 4 automotive LCD driver for low multiplex rates



There are two banks; bank 0 and bank 1. Figure 13 shows the location of these banks relative to the RAM map. Input and output banks can be set independently from one another with the Bank-select command (see <u>Table 10 on page 8</u>). Figure 14 shows the concept.



In the static drive mode, the bank-select command may request the contents of row 2 to be selected for display instead of the contents of row 0. In the 1:2 multiplex mode, the contents of rows 2 and 3 may be selected instead of rows 0 and 1. This gives the provision for preparing display information in an alternative bank and to be able to switch to it once it is assembled.

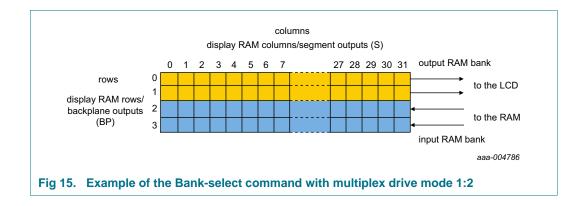
In <u>Figure 15</u> an example is shown for 1:2 multiplex drive mode where the displayed data is read from the first two rows of the memory (bank 0), while the transmitted data is stored in the second two rows of the memory (bank 1).

PCA85162

### **NXP Semiconductors**

# PCA85162

32 x 4 automotive LCD driver for low multiplex rates



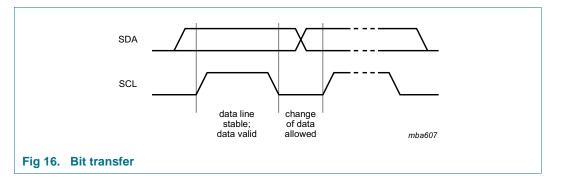
27 of 56

### 8. Characteristics of the I<sup>2</sup>C-bus

The I<sup>2</sup>C-bus is for bidirectional, two-line communication between different ICs or modules. The two lines are a Serial DAta line (SDA) and a Serial CLock line (SCL). Both lines must be connected to a positive supply via a pull-up resistor when connected to the output stages of a device. Data transfer may be initiated only when the bus is not busy.

#### 8.1 Bit transfer

One data bit is transferred during each clock pulse. The data on the SDA line must remain stable during the HIGH period of the clock pulse as changes in the data line at this time will be interpreted as a control signal (see Figure 16).



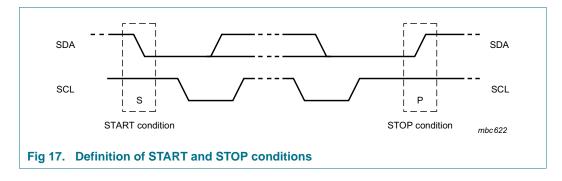
#### 8.2 START and STOP conditions

Both data and clock lines remain HIGH when the bus is not busy.

A HIGH-to-LOW transition of the data line while the clock is HIGH is defined as the START condition - S.

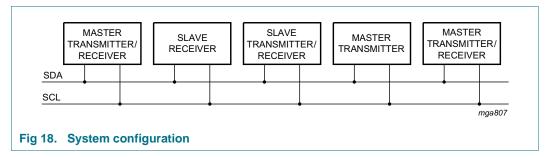
A LOW-to-HIGH transition of the data line while the clock is HIGH is defined as the STOP condition - P.

The START and STOP conditions are illustrated in Figure 17.



#### 8.3 System configuration

A device generating a message is a transmitter, a device receiving a message is the receiver. The device that controls the message is the master and the devices which are controlled by the master are the slaves. The system configuration is shown in Figure 18.



### 8.4 Acknowledge

The number of data bytes transferred between the START and STOP conditions from transmitter to receiver is unlimited. Each byte of eight bits is followed by an acknowledge cycle.

- A slave receiver, which is addressed, must generate an acknowledge after the reception of each byte
- A master receiver must generate an acknowledge after the reception of each byte that has been clocked out of the slave transmitter
- The device that acknowledges must pull-down the SDA line during the acknowledge clock pulse, so that the SDA line is stable LOW during the HIGH period of the acknowledge related clock pulse (set-up and hold times must be taken into consideration)
- A master receiver must signal an end of data to the transmitter by not generating an acknowledge on the last byte that has been clocked out of the slave. In this event, the transmitter must leave the data line HIGH to enable the master to generate a STOP condition

data output by transmitter not acknowledge data output by receiver acknowledge SCL from 8 master s clock pulse for START acknowledgement condition mbc602 Fig 19. Acknowledgement of the I<sup>2</sup>C-bus

Acknowledgement on the I<sup>2</sup>C-bus is illustrated in Figure 19.

29 of 56

#### 8.5 I<sup>2</sup>C-bus controller

The PCA85162 acts as an I<sup>2</sup>C-bus slave receiver. It does not initiate I<sup>2</sup>C-bus transfers or transmit data to an I<sup>2</sup>C-bus master receiver. The only data output from the PCA85162 are the acknowledge signals of the selected devices. Device selection depends on the I<sup>2</sup>C-bus slave address, on the transferred command data and on the hardware subaddress.

In single device applications, the hardware subaddress inputs A0, A1, and A2 are normally tied to  $V_{SS}$  which defines the hardware subaddress 0. In multiple device applications A0, A1, and A2 are tied to  $V_{SS}$  or  $V_{DD}$  using a binary coding scheme, so that no two devices with a common I<sup>2</sup>C-bus slave address have the same hardware subaddress.

#### 8.6 Input filters

To enhance noise immunity in electrically adverse environments, RC low-pass filters are provided on the SDA and SCL lines.

#### 8.7 I<sup>2</sup>C-bus protocol

Two I<sup>2</sup>C-bus slave addresses (0111 000 and 0111 001) are used to address the PCA85162. The entire I<sup>2</sup>C-bus slave address byte is shown in <u>Table 17</u>.

#### Table 17. I<sup>2</sup>C slave address byte

	Slave add	ress						
Bit	7	6	5	4	3	2	1	0
	MSB							LSB
	0	1	1	1	0	0	SA0	R/W

The PCA85162 is a write-only device and will not respond to a read access, therefore bit 0 should always be logic 0. Bit 1 of the slave address byte that a PCA85162 will respond to, is defined by the level tied to its SA0 input ( $V_{SS}$  for logic 0 and  $V_{DD}$  for logic 1).

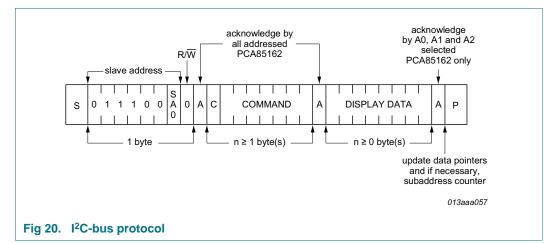
Having two reserved slave addresses allows the following on the same I<sup>2</sup>C-bus:

- Up to 16 PCA85162 for very large LCD applications
- The use of two types of LCD multiplex drive modes

The l<sup>2</sup>C-bus protocol is shown in <u>Figure 20</u>. The sequence is initiated with a START condition (S) from the l<sup>2</sup>C-bus master which is followed by one of the two possible PCA85162 slave addresses available. All PCA85162 whose SA0 inputs correspond to bit 0 of the slave address respond by asserting an acknowledge in parallel. This l<sup>2</sup>C-bus transfer is ignored by all PCA85162 whose SA0 inputs are set to the alternative level.

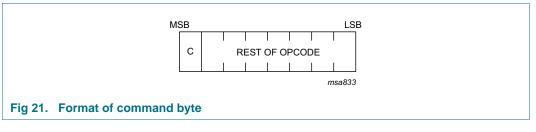
PCA85162

#### 32 x 4 automotive LCD driver for low multiplex rates



After an acknowledgement, one or more command bytes follow that define the status of each addressed PCA85162.

The last command byte sent is identified by resetting its most significant bit, continuation bit C (see Figure 21). The command bytes are also acknowledged by all addressed PCA85162 on the bus.



After the last command byte, one or more display data bytes may follow. Display data bytes are stored in the display RAM at the address specified by the data pointer and the subaddress counter. Both data pointer and subaddress counter are automatically updated and the data directed to the intended PCA85162 device.

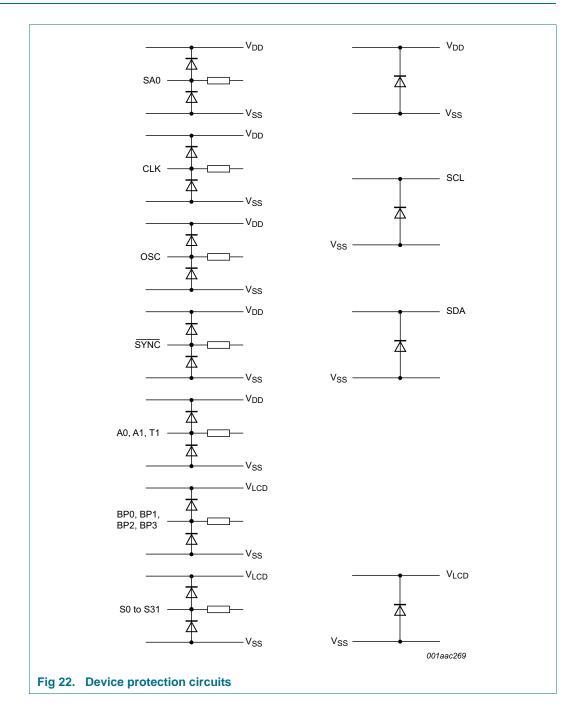
An acknowledgement after each byte is asserted only by the PCA85162 that are addressed via address lines A0, A1, and A2. After the last display byte, the I<sup>2</sup>C-bus master asserts a STOP condition (P). Alternately a START may be asserted to restart an I<sup>2</sup>C-bus access.

PCA85162

31 of 56

#### 32 x 4 automotive LCD driver for low multiplex rates

## 9. Internal circuitry



PCA85162 Product data sheet

### 10. Safety notes

CAUTION	
	This device is sensitive to ElectroStatic Discharge (ESD). Observe precautions for handling electrostatic sensitive devices.
	Such precautions are described in the ANSI/ESD S20.20, IEC/ST 61340-5, JESD625-A or equivalent standards.

#### CAUTION



Static voltages across the liquid crystal display can build up when the LCD supply voltage ( $V_{LCD}$ ) is on while the IC supply voltage ( $V_{DD}$ ) is off, or vice versa. This may cause unwanted display artifacts. To avoid such artifacts,  $V_{LCD}$  and  $V_{DD}$  must be applied or removed together.

### **11. Limiting values**

#### Table 18. Limiting values

In accordance with the Absolute Maximum Rating System (IEC 60134).

Symbol	Parameter	Conditions	Min	Max	Unit
V <sub>DD</sub>	supply voltage		-0.5	+6.5	V
V <sub>LCD</sub>	LCD supply voltage		-0.5	+9.0	V
VI	input voltage	on each of the pins CLK, SDA, SCL, SYNC, SA0, OSC, A0 to A2	-0.5	+6.5	V
V <sub>O</sub>	output voltage	on each of the pins S0 to S31, BP0 to BP3	-0.5	+9.0	V
l <sub>l</sub>	input current		-10	+10	mA
I <sub>O</sub>	output current		-10	+10	mA
I <sub>DD</sub>	supply current		-50	+50	mA
I <sub>DD(LCD)</sub>	LCD supply current		-50	+50	mA
I <sub>SS</sub>	ground supply current		-50	+50	mA
P <sub>tot</sub>	total power dissipation		-	400	mW
Po	output power		-	100	mW
V <sub>ESD</sub>	electrostatic	HBM [1]	-	±3500	V
	discharge voltage	CDM [2]	-	±1750	V
l <sub>lu</sub>	latch-up current	[3]	-	100	mA
T <sub>stg</sub>	storage temperature	[4]	-65	+150	°C
T <sub>amb</sub>	ambient temperature	operating device	-40	+95	°C

[1] Pass level; Human Body Model (HBM), according to Ref. 8 "JESD22-A114".

[2] Pass level; Charged-Device Model (CDM), according to Ref. 9 "JESD22-C101".

[3] Pass level; latch-up testing according to Ref. 10 "JESD78" at maximum ambient temperature (T<sub>amb(max)</sub>).

[4] According to the store and transport requirements (see <u>Ref. 14 "UM10569"</u>) the devices have to be stored at a temperature of +8 °C to +45 °C and a humidity of 25 % to 75 %.

## 12. Static characteristics

#### Table 19. Static characteristics

 $V_{DD}$  = 1.8 V to 5.5 V;  $V_{SS}$  = 0 V;  $V_{LCD}$  = 2.5 V to 8.0 V;  $T_{amb}$  = -40 °C to +95 °C; unless otherwise specified.

Symbol	Parameter	Conditions	Min	Тур	Max	Unit
Supplies		·		I	I	
V <sub>DD</sub>	supply voltage		1.8	-	5.5	V
V <sub>LCD</sub>	LCD supply voltage	[1]	2.5	-	8.0	V
I <sub>DD</sub>	supply current	f <sub>clk(ext)</sub> = 1536 Hz [2][3]	-	3.5	7	μA
		V <sub>DD</sub> = 3.0 V; T <sub>amb</sub> = 25 °C	-	2.7	-	μA
I <sub>DD(LCD)</sub>	LCD supply current	f <sub>clk(ext)</sub> = 1536 Hz [2]	-	23	32	μA
		$V_{LCD} = 3.0 \text{ V}; \text{ T}_{amb} = 25 ^{\circ}\text{C}$	-	13	-	μA
Logic <sup>[4]</sup>		·		L		
V <sub>P(POR)</sub>	power-on reset supply voltage		1.0	1.3	1.6	V
V <sub>IL</sub>	LOW-level input voltage	on pins CLK, <u>SYNC</u> , OSC, A0 to A2, SA0, SCL, SDA	V <sub>SS</sub>	-	0.3V <sub>DD</sub>	V
V <sub>IH</sub>	HIGH-level input voltage	on pins CLK, SYNC, OSC, A0 to [5][6] A2, SA0, SCL, SDA	0.7V <sub>DD</sub>	-	V <sub>DD</sub>	V
I <sub>OL</sub>	LOW-level output current	output sink current; V <sub>OL</sub> = 0.4 V; V <sub>DD</sub> = 5 V				I
		on pins CLK and SYNC	1	-	-	mA
		on pin SDA	3	-	-	mA
I <sub>OH(CLK)</sub>	HIGH-level output current on pin CLK	output source current; V <sub>OH</sub> = 4.6 V; V <sub>DD</sub> = 5 V	1	-	-	mA
IL	leakage current	$V_I = V_{DD}$ or $V_{SS}$ ; on pins CLK, SCL, SDA, A0 to A2, and SA0	-1	-	+1	μA
I <sub>L(OSC)</sub>	leakage current on pin OSC	$V_{I} = V_{DD}$	-1	-	+1	μA
Cl	input capacitance	[7]	-	-	7	pF

Symbol	Parameter	Conditions	Min	Тур	Max	Unit
LCD out	tputs		Ċ	Ċ		÷
ΔV <sub>O</sub>	output voltage variation	on pins BP0 to BP3 and S0 to S31	-100	-	+100	mV
R <sub>O</sub>	output resistance	V <sub>LCD</sub> = 5 V	<u>1</u>	i		
		on pins BP0 to BP3	-	1.5	-	kΩ
		on pins S0 to S31	-	6.0	-	kΩ

#### Table 19. Static characteristics ...continued

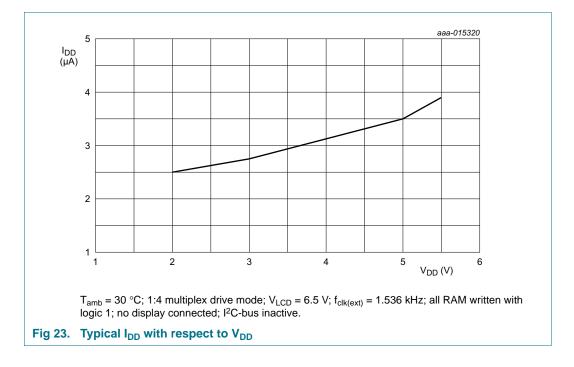
 $V_{DD} = 1.8 \text{ V to } 5.5 \text{ V}; V_{SS} = 0 \text{ V}; V_{LCD} = 2.5 \text{ V to } 8.0 \text{ V}; T_{amb} = -40 \text{ }^{\circ}\text{C} \text{ to } +95 \text{ }^{\circ}\text{C}; unless otherwise specified.}$ 

[1]  $V_{LCD} > 3 V$  for  $\frac{1}{3}$  bias.

[2] LCD outputs are open-circuit; inputs at  $V_{SS}$  or  $V_{DD}$ ; external clock with 50 % duty factor; I<sup>2</sup>C-bus inactive.

[3] For typical values, see Figure 23.

- [4] The I<sup>2</sup>C-bus interface of PCA85162 is 5 V tolerant.
- [5] When tested, I<sup>2</sup>C pins SCL and SDA have no diode to  $V_{DD}$  and may be driven to the V<sub>1</sub> limiting values given in Table 18 (see Figure 22 as well).
- [6] Propagation delay of driver between clock (CLK) and LCD driving signals.
- [7] Periodically sampled, not 100 % tested.
- [8] Outputs measured one at a time.



## **13. Dynamic characteristics**

#### Table 20. Dynamic characteristics

 $V_{DD}$  = 1.8 V to 5.5 V;  $V_{SS}$  = 0 V;  $V_{LCD}$  = 2.5 V to 8.0 V;  $T_{amb}$  = -40 °C to +95 °C; unless otherwise specified.

Symbol	Parameter	Conditions	Min	Тур	Max	Unit
Clock	I	·				
f <sub>clk(int)</sub>	internal clock frequency	[1]	1920	2640	3600	Hz
f <sub>clk(ext)</sub>	external clock frequency		960	-	4800	Hz
f <sub>fr</sub>	frame frequency	internal clock	80	110	150	Hz
		external clock	40	-	200	Hz
t <sub>clk(H)</sub>	HIGH-level clock time		60	-	-	μS
<sup>t</sup> clk(L)	LOW-level clock time		60	-	-	μS
Synchroniz	zation					
t <sub>PD(SYNC_N)</sub>	SYNC propagation delay		-	30	-	ns
t <sub>SYNC_NL</sub>	SYNC LOW time		1	-	-	μs
t <sub>PD(drv)</sub>	driver propagation delay	V <sub>LCD</sub> = 5 V [2]	-	-	30	μS
l <sup>2</sup> C-bus <sup>[3]</sup>	1		1			I
Pin SCL						
f <sub>SCL</sub>	SCL clock frequency		-	-	400	kHz
t <sub>LOW</sub>	LOW period of the SCL clock		1.3	-	-	μs
t <sub>HIGH</sub>	HIGH period of the SCL clock		0.6	-	-	μs
Pin SDA		1				I
t <sub>SU;DAT</sub>	data set-up time		100	-	-	ns
t <sub>HD;DAT</sub>	data hold time		0	-	-	ns
Pins SCL a	nd SDA		1			I
t <sub>BUF</sub>	bus free time between a STOP and START condition		1.3	-	-	μs
<sup>t</sup> su;sto	set-up time for STOP condition		0.6	-	-	μs
HD;STA	hold time (repeated) START condition		0.6	-	-	μS
<sup>I</sup> SU;STA	set-up time for a repeated START condition		0.6	-	-	μs
t <sub>r</sub>	rise time of both SDA and SCL signals	f <sub>SCL</sub> = 400 kHz	-	-	0.3	μs
		f <sub>SCL</sub> < 125 kHz	-	-	1.0	μs

36 of 56

Symbol	Parameter	Conditions	Min	Тур	Max	Unit	
t <sub>f</sub>	fall time of both SDA and SCL signals		-	-	0.3	μs	
C <sub>b</sub>	capacitive load for each bus line		-	-	400	pF	
t <sub>w(spike)</sub>	spike pulse width	on the I <sup>2</sup> C-bus	-	-	50	ns	

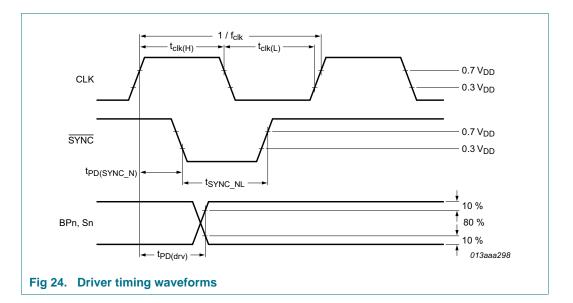
#### Table 20. Dynamic characteristics ...continued

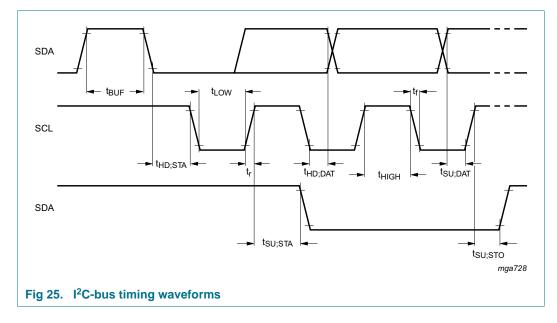
25 V to 80 V. T - - V. V 01414 10 00 1-

[1] Typical output duty factor: 50 % measured at the CLK output pin.

[2] Not tested in production.

[3] All timing values are valid within the operating supply voltage and ambient temperature range and are referenced to V<sub>IL</sub> and V<sub>IH</sub> with an input voltage swing of  $V_{SS}$  to  $V_{DD}$ .





# PCA85162

### 14. Application information

#### 14.1 Cascaded operation

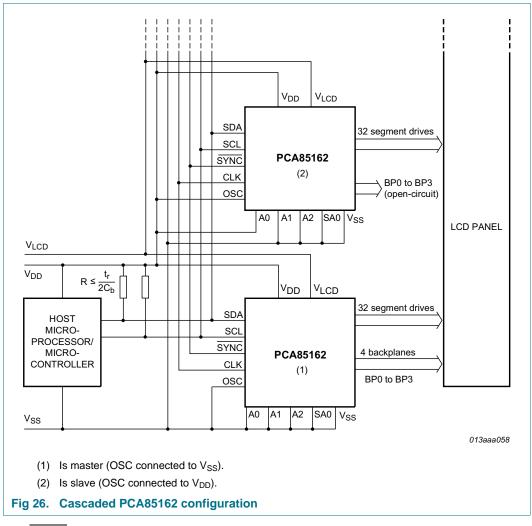
Large display configurations of up to 16 PCA85162 can be recognized on the same  $I^2C$ -bus by using the 3-bit hardware subaddress (A0, A1, and A2) and the programmable  $I^2C$ -bus slave address (SA0).

Cluster	Bit SA0	Pin A2	Pin A1	Pin A0	Device
1	0	0	0	0	0
		0	0	1	1
		0	1	0	2
		0	1	1	3
		1	0	0	4
		1	0	1	5
		1	1	0	6
		1	1	1	7
2	1	0	0	0	8
		0	0	1	9
		0	1	0	10
		0	1	1	11
		1	0	0	12
		1	0	1	13
		1	1	0	14
		1	1	1	15

#### Table 21. Addressing cascaded PCA85162

When cascaded PCA85162 are synchronized, they can share the backplane signals from one of the devices in the cascade. Such an arrangement is cost-effective in large LCD applications since the backplane outputs of only one device need to be through-plated to the backplane electrodes of the display. The other PCA85162 of the cascade contribute additional segment outputs. The backplanes can either be connected together to enhance the drive capability or some can be left open-circuit (such as the ones from the slave in Figure 26) or just some of the master and some of the slave will be taken to facilitate the layout of the display.

#### 32 x 4 automotive LCD driver for low multiplex rates



The SYNC line is provided to maintain the correct synchronization between all cascaded PCA85162. Synchronization is guaranteed after a power-on reset. The only time that SYNC is likely to be needed is if synchronization is accidentally lost (e.g. by noise in adverse electrical environments or by defining a multiplex drive mode when PCA85162 with different SA0 levels are cascaded).

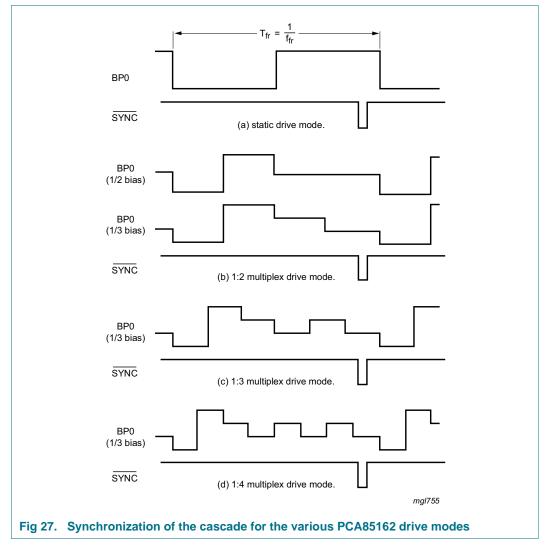
SYNC is organized as an input/output pin. The output selection is realized as an open-drain driver with an internal pull-up resistor. A PCA85162 asserts the SYNC line at the onset of its last active backplane signal and monitors the SYNC line at all other times. If synchronization in the cascade is lost, it is restored by the first PCA85162 to assert SYNC. The timing relationship between the backplane waveforms and the SYNC signal for the various drive modes of the PCA85162 are shown in Figure 27.

The contact resistance between the SYNC on each cascaded device must be controlled. If the resistance is too high, the device is not able to synchronize properly; this is particularly applicable to chip-on-glass applications. The maximum SYNC contact resistance allowed for the number of devices in cascade is given in Table 22.

 Table 22.
 SYNC contact resistance

Number of devices	Maximum contact resistance
2	6 kΩ
3 to 5	2.2 kΩ
6 to 10	1.2 kΩ
10 to 16	700 Ω

The PCA85162 can always be cascaded with other devices of the same type or conditionally with other devices of the same family. This allows optimal drive selection for a given number of pixels to display. Figure 24 and Figure 27 show the timing of the synchronization signals.



Only one master but multiple slaves are allowed in a cascade. All devices in the cascade have to use the same clock whether it is supplied externally or provided by the master.

If an external clock source is used, all PCA85162 in the cascade must be configured such as to receive the clock from that external source (pin OSC connected to  $V_{DD}$ ). Thereby it must be ensured that the clock tree is designed such that on all PCA85162 the clock propagation delay from the clock source to all PCA85162 in the cascade is as equal as possible since otherwise synchronization artefacts may occur.

In mixed cascading configurations, care has to be taken that the specifications of the individual cascaded devices are met at all times.

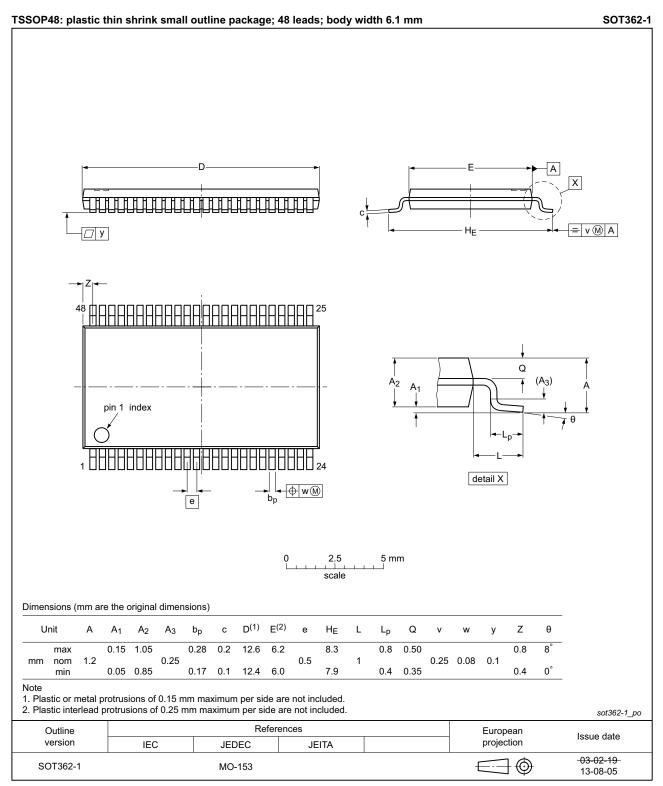
### **15. Test information**

#### **15.1 Quality information**

This product has been qualified in accordance with the Automotive Electronics Council (AEC) standard *Q100 - Failure mechanism based stress test qualification for integrated circuits*, and is suitable for use in automotive applications.

32 x 4 automotive LCD driver for low multiplex rates

# 16. Package outline



#### Fig 28. Package outline SOT362-1 (TSSOP48)

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PCA85162

# **17. Handling information**

All input and output pins are protected against ElectroStatic Discharge (ESD) under normal handling. When handling Metal-Oxide Semiconductor (MOS) devices ensure that all normal precautions are taken as described in *JESD625-A*, *IEC 61340-5* or equivalent standards.

### **18. Packing information**

#### 18.1 Tape and reel information

For tape and reel packing information, please see Ref. 12 "SOT362-1\_118" on page 51.

### 19. Soldering of SMD packages

This text provides a very brief insight into a complex technology. A more in-depth account of soldering ICs can be found in Application Note *AN10365 "Surface mount reflow soldering description"*.

#### **19.1 Introduction to soldering**

Soldering is one of the most common methods through which packages are attached to Printed Circuit Boards (PCBs), to form electrical circuits. The soldered joint provides both the mechanical and the electrical connection. There is no single soldering method that is ideal for all IC packages. Wave soldering is often preferred when through-hole and Surface Mount Devices (SMDs) are mixed on one printed wiring board; however, it is not suitable for fine pitch SMDs. Reflow soldering is ideal for the small pitches and high densities that come with increased miniaturization.

#### 19.2 Wave and reflow soldering

Wave soldering is a joining technology in which the joints are made by solder coming from a standing wave of liquid solder. The wave soldering process is suitable for the following:

- Through-hole components
- Leaded or leadless SMDs, which are glued to the surface of the printed circuit board

Not all SMDs can be wave soldered. Packages with solder balls, and some leadless packages which have solder lands underneath the body, cannot be wave soldered. Also, leaded SMDs with leads having a pitch smaller than ~0.6 mm cannot be wave soldered, due to an increased probability of bridging.

The reflow soldering process involves applying solder paste to a board, followed by component placement and exposure to a temperature profile. Leaded packages, packages with solder balls, and leadless packages are all reflow solderable.

Key characteristics in both wave and reflow soldering are:

- · Board specifications, including the board finish, solder masks and vias
- Package footprints, including solder thieves and orientation
- The moisture sensitivity level of the packages
- Package placement
- Inspection and repair
- Lead-free soldering versus SnPb soldering

#### **19.3 Wave soldering**

Key characteristics in wave soldering are:

- Process issues, such as application of adhesive and flux, clinching of leads, board transport, the solder wave parameters, and the time during which components are exposed to the wave
- · Solder bath specifications, including temperature and impurities

#### 19.4 Reflow soldering

Key characteristics in reflow soldering are:

- Lead-free versus SnPb soldering; note that a lead-free reflow process usually leads to higher minimum peak temperatures (see <u>Figure 29</u>) than a SnPb process, thus reducing the process window
- Solder paste printing issues including smearing, release, and adjusting the process window for a mix of large and small components on one board
- Reflow temperature profile; this profile includes preheat, reflow (in which the board is heated to the peak temperature) and cooling down. It is imperative that the peak temperature is high enough for the solder to make reliable solder joints (a solder paste characteristic). In addition, the peak temperature must be low enough that the packages and/or boards are not damaged. The peak temperature of the package depends on package thickness and volume and is classified in accordance with Table 23 and 24

#### Table 23. SnPb eutectic process (from J-STD-020D)

Package thickness (mm)	Package reflow temperature (°C)					
	Volume (mm <sup>3</sup> )					
	< 350	≥ 350				
< 2.5	235	220				
≥ 2.5	220	220				

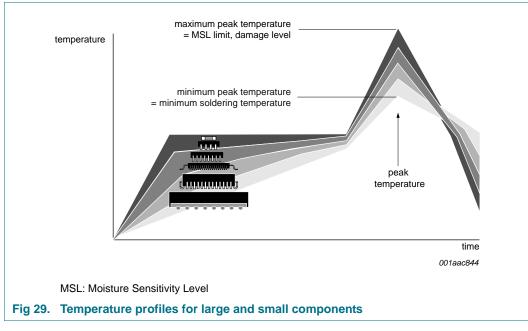
#### Table 24. Lead-free process (from J-STD-020D)

Package thickness (mm)	Package reflow temperature (°C) Volume (mm <sup>3</sup> )						
	< 350	350 to 2000	> 2000				
< 1.6	260	260	260				
1.6 to 2.5	260	250	245				
> 2.5	250	245	245				

Moisture sensitivity precautions, as indicated on the packing, must be respected at all times.

Studies have shown that small packages reach higher temperatures during reflow soldering, see Figure 29.

32 x 4 automotive LCD driver for low multiplex rates

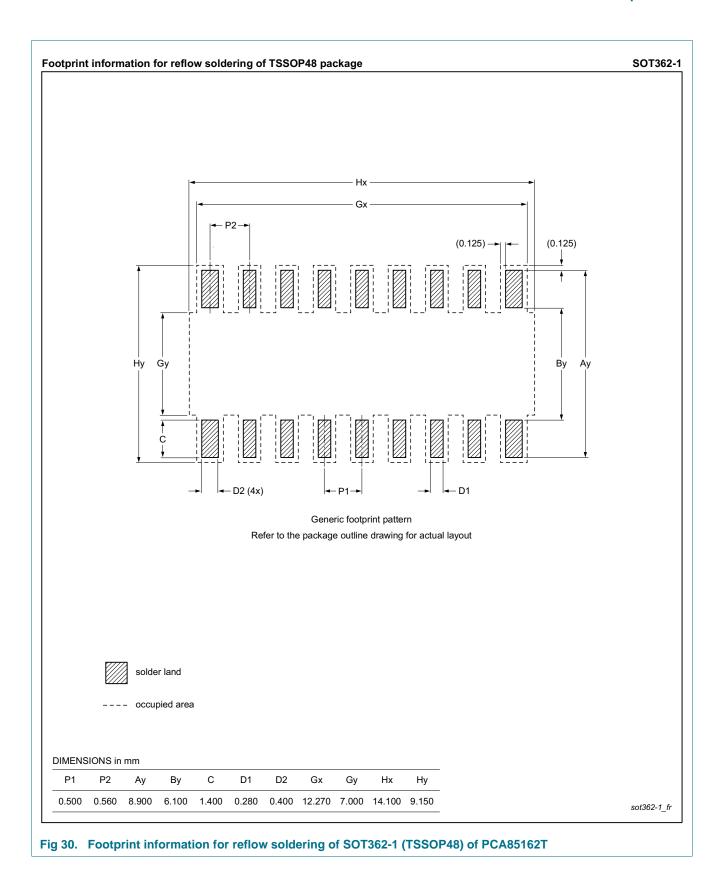


For further information on temperature profiles, refer to Application Note *AN10365 "Surface mount reflow soldering description"*.

# **20.** Footprint information

# PCA85162

32 x 4 automotive LCD driver for low multiplex rates



# 21. Appendix

### 21.1 LCD segment driver selection

#### Table 25. Selection of LCD segment drivers

Type name	Nun	nber c	of elei	ments	s at M	UX		$V_{DD}$ (V) $V_{LCD}$ (V) $f_{fr}$ (I	f <sub>fr</sub> (Hz) V <sub>LCD</sub> (V)	V <sub>LCD</sub> (V)	T <sub>amb</sub> (°C)	Interface	Package	AEC-		
	1:1	1:2	1:3	1:4	1:6	1:8	1:9				charge pump	temperature compensat.				Q1(
PCA8553DTT	40	80	120	160	-	-	-	1.8 to 5.5	1.8 to 5.5	32 to 256[1]	N	N	-40 to 105	I <sup>2</sup> C / SPI	TSSOP56	Υ
PCA8546ATT	-	-	-	176	-	-	-	1.8 to 5.5	2.5 to 9	60 to 300 <sup>[1]</sup>	N	N	-40 to 95	I <sup>2</sup> C	TSSOP56	Y
PCA8546BTT	-	-	-	176	-	-	-	1.8 to 5.5	2.5 to 9	60 to 300[1]	N	N	-40 to 95	SPI	TSSOP56	Y
PCA8547AHT	44	88	-	176	-	-	-	1.8 to 5.5	2.5 to 9	60 to 300[1]	Y	Y	-40 to 95	I <sup>2</sup> C	TQFP64	Y
PCA8547BHT	44	88	-	176	-	-	-	1.8 to 5.5	2.5 to 9	60 to 300[1]	Y	Y	-40 to 95	SPI	TQFP64	Y
PCF85134HL	60	120	180	240	-	-	-	1.8 to 5.5	2.5 to 6.5	82	N	N	-40 to 85	I <sup>2</sup> C	LQFP80	Ν
PCA85134H	60	120	180	240	-	-	-	1.8 to 5.5	2.5 to 8	82	N	N	-40 to 95	I <sup>2</sup> C	LQFP80	Y
PCA8543AHL	60	120	-	240	-	-	-	2.5 to 5.5	2.5 to 9	60 to 300[1]	Y	Y	-40 to 105	I <sup>2</sup> C	LQFP80	Y
PCF8545ATT	-	-	-	176	252	320	-	1.8 to 5.5	2.5 to 5.5	60 to 300[1]	N	N	-40 to 85	I <sup>2</sup> C	TSSOP56	Ν
PCF8545BTT	-	-	-	176	252	320	-	1.8 to 5.5	2.5 to 5.5	60 to 300[1]	N	N	-40 to 85	SPI	TSSOP56	Ν
PCF8536AT	-	-	-	176	252	320	-	1.8 to 5.5	2.5 to 9	60 to 300[1]	N	N	-40 to 85	I <sup>2</sup> C	TSSOP56	Ν
PCF8536BT	-	-	-	176	252	320	-	1.8 to 5.5	2.5 to 9	60 to 300[1]	N	N	-40 to 85	SPI	TSSOP56	Ν
PCA8536AT	-	-	-	176	252	320	-	1.8 to 5.5	2.5 to 9	60 to 300[1]	N	N	-40 to 95	I <sup>2</sup> C	TSSOP56	Y
PCA8536BT	-	-	-	176	252	320	-	1.8 to 5.5	2.5 to 9	60 to 300[1]	N	N	-40 to 95	SPI	TSSOP56	Y
PCF8537AH	44	88	-	176	276	352	-	1.8 to 5.5	2.5 to 9	60 to 300[1]	Y	Y	-40 to 85	I <sup>2</sup> C	TQFP64	Ν
PCF8537BH	44	88	-	176	276	352	-	1.8 to 5.5	2.5 to 9	60 to 300[1]	Y	Y	-40 to 85	SPI	TQFP64	Ν
PCA8537AH	44	88	-	176	276	352	-	1.8 to 5.5	2.5 to 9	60 to 300[1]	Y	Y	-40 to 95	I <sup>2</sup> C	TQFP64	Y
PCA8537BH	44	88	-	176	276	352	-	1.8 to 5.5	2.5 to 9	60 to 300[1]	Y	Y	-40 to 95	SPI	TQFP64	Y
PCA9620H	60	120	-	240	320	480	-	2.5 to 5.5	2.5 to 9	60 to 300[1]	Y	Y	-40 to 105	I <sup>2</sup> C	LQFP80	Y
PCA9620U	60	120	-	240	320	480	-	2.5 to 5.5	2.5 to 9	60 to 300[1]	Y	Y	-40 to 105	I <sup>2</sup> C	Bare die	Y
PCF8576DU	40	80	120	160	-	-	-	1.8 to 5.5	2.5 to 6.5	77	N	N	-40 to 85	I <sup>2</sup> C	Bare die	Ν
PCF8576EUG	40	80	120	160	-	-	-	1.8 to 5.5	2.5 to 6.5	77	N	N	-40 to 85	I <sup>2</sup> C	Bare die	Ν
PCA8576FUG	40	80	120	160	-	-	-	1.8 to 5.5	2.5 to 8	200	N	N	-40 to 105	I <sup>2</sup> C	Bare die	Y
PCF85133U	80	160	240	320	-	-	-	1.8 to 5.5	2.5 to 6.5	82, 110 <mark>[2]</mark>	N	Ν	-40 to 85	l <sup>2</sup> C	Bare die	Ν
PCA85133U	80	160	240	320	-	-	-	1.8 to 5.5	2.5 to 8	82, 110 <sup>[2]</sup>	N	N	-40 to 95	I <sup>2</sup> C	Bare die	Υ

32 x 4 automotive LCD driver for low multiplex rates

PCA85162

# Product data sheet PCA85162

#### Table 25. Selection of LCD segment drivers ...continued

Table 25.SeleType name	Num	Number of elements at MUX							V <sub>LCD</sub> (V)	V <sub>LCD</sub> (V) f <sub>fr</sub> (Hz) V <sub>I</sub>	V <sub>LCD</sub> (V) V <sub>LC</sub>	V <sub>LCD</sub> (V)	T <sub>amb</sub> (°C)	Interface	Package	AEC-
	1:1	1:2	1:3	1:4	1:6	1:8	1:9				charge pump	temperature compensat.				Q100
PCA85233UG	80	160	240	320	-	-	-	1.8 to 5.5	2.5 to 8	150, 220 <mark>[2]</mark>	N	N	-40 to 105	I <sup>2</sup> C	Bare die	Y
PCF85132U	160	320	480	640	-	-	-	1.8 to 5.5	1.8 to 8	60 to 90[1]	N	N	-40 to 85	I <sup>2</sup> C	Bare die	Ν
PCA8530DUG	102	204	-	408	-	-	-	2.5 to 5.5	4 to 12	45 to 300[1]	Y	Y	-40 to 105	I <sup>2</sup> C / SPI	Bare die	Y
PCA85132U	160	320	480	640	-	-	-	1.8 to 5.5	1.8 to 8	60 to 90[1]	N	N	-40 to 95	l <sup>2</sup> C	Bare die	Y
PCA85232U	160	320	480	640	-	-	-	1.8 to 5.5	1.8 to 8	117 to 176 <sup>[1]</sup>	N	N	-40 to 95	l <sup>2</sup> C	Bare die	Y
PCF8538UG	102	204	-	408	612	816	918	2.5 to 5.5	4 to 12	45 to 300[1]	Y	Y	-40 to 85	I <sup>2</sup> C / SPI	Bare die	Ν
PCA8538UG	102	204	-	408	612	816	918	2.5 to 5.5	4 to 12	45 to 300 <sup>[1]</sup>	Y	Y	-40 to 105	I <sup>2</sup> C / SPI	Bare die	Y

#### [1] Software programmable.

[2] Hardware selectable.

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Product data sheet

32 x 4 automotive LCD driver for low multiplex rates

# 22. Abbreviations

Table 26. Abbre	viations
Acronym	Description
AEC	Automotive Electronics Council
CMOS	Complementary Metal-Oxide Semiconductor
CDM	Charged Device Model
DC	Direct Current
НВМ	Human Body Model
I <sup>2</sup> C	Inter-Integrated Circuit
IC	Integrated Circuit
LCD	Liquid Crystal Display
LSB	Least Significant Bit
MSB	Most Significant Bit
MSL	Moisture Sensitivity Level
PCB	Printed-Circuit Board
POR	Power-On Reset
RAM	Random Access Memory
RC	Resistance and Capacitance
RMS	Root Mean Square
SCL	Serial CLock line
SDA	Serial DAta Line
SMD	Surface-Mount Device

#### 32 x 4 automotive LCD driver for low multiplex rates

### 23. References

- [1] AN10170 Design guidelines for COG modules with NXP monochrome LCD drivers
- [2] AN10365 Surface mount reflow soldering description
- [3] AN10853 ESD and EMC sensitivity of IC
- [4] AN11267 EMC and system level ESD design guidelines for LCD drivers
- [5] IEC 60134 Rating systems for electronic tubes and valves and analogous semiconductor devices
- [6] IEC 61340-5 Protection of electronic devices from electrostatic phenomena
- [7] IPC/JEDEC J-STD-020D Moisture/Reflow Sensitivity Classification for Nonhermetic Solid State Surface Mount Devices
- [8] JESD22-A114 Electrostatic Discharge (ESD) Sensitivity Testing Human Body Model (HBM)
- [9] JESD22-C101 Field-Induced Charged-Device Model Test Method for Electrostatic-Discharge-Withstand Thresholds of Microelectronic Components
- [10] JESD78 IC Latch-Up Test
- [11] JESD625-A Requirements for Handling Electrostatic-Discharge-Sensitive (ESDS) Devices
- [12] SOT362-1\_118 TSSOP48; Reel pack; SMD, 13", packing information
- [13] UM10204 I<sup>2</sup>C-bus specification and user manual
- [14] UM10569 Store and transport requirements

### 24. Revision history

Document ID	Release date	Data sheet status	Change notice	Supersedes			
PCA85162 v.4	20150409	Product data sheet	-	PCA85162 v.3			
Modifications:	of NXP Sem <ul> <li>Legal texts h</li> <li>Changed I<sub>DD</sub></li> </ul>	<ul> <li>The format of this data sheet has been redesigned to comply with the new identity of NXP Semiconductors.</li> <li>Legal texts have been adapted to the new company name where appropriate.</li> <li>Changed I<sub>DD(LCD)</sub> values in <u>Table 19</u></li> <li>Added orderable part number in <u>Table 2</u></li> </ul>					
PCA85162 v.3	20120905	Product data sheet	-	PCA85162 v.2			
PCA85162 v.2	20110616	Product data sheet	-	PCA85162 v.1			
PCA85162 v.1	20100419	Product data sheet	-	-			

Table 27. Revision history

PCA85162 Product data sheet

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#### 25.1 Data sheet status

Document status[1][2]	Product status <sup>[3]</sup>	Definition
Objective [short] data sheet	Development	This document contains data from the objective specification for product development.
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[2] The term 'short data sheet' is explained in section "Definitions".

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PCA85162

# PCA85162

#### 32 x 4 automotive LCD driver for low multiplex rates

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# PCA85162

#### 32 x 4 automotive LCD driver for low multiplex rates

# 27. Tables

Ordering information
Ordering options
Marking codes2
Pin description
Definition of PCA85162 commands6
C bit description
Mode-set command bit description7
Load-data-pointer command bit description 7
Device-select command bit description8
Bank-select command bit description8
Blink-select command bit description9
Blink frequencies10
Selection of possible display configurations 11
Biasing characteristics
Standard RAM filling in 1:3 multiplex drive
mode
Entire RAM filling by rewriting in 1:3
multiplex drive mode
I <sup>2</sup> C slave address byte
Limiting values
Static characteristics
Dynamic characteristics
Addressing cascaded PCA85162
SYNC contact resistance
SnPb eutectic process (from J-STD-020D)45
Lead-free process (from J-STD-020D)45
Selection of LCD segment drivers
Abbreviations
Revision history51

#### 32 x 4 automotive LCD driver for low multiplex rates

# 28. Figures

Fig 1.	Block diagram of PCA85162
Fig 2.	Pinning diagram for TSSOP48 (PCA85162T)4
Fig 3.	Example of displays suitable for PCA85162 11
Fig 4.	Typical system configuration12
Fig 5.	Electro-optical characteristic: relative
	transmission curve of the liquid14
Fig 6.	Static drive mode waveforms15
Fig 7.	Waveforms for the 1:2 multiplex drive mode
	with $\frac{1}{2}$ bias
Fig 8.	Waveforms for the 1:2 multiplex drive mode
	with $\frac{1}{3}$ bias
Fig 9.	Waveforms for the 1:3 multiplex drive mode
	with $\frac{1}{3}$ bias
Fig 10.	Waveforms for the 1:4 multiplex drive mode
	with $\frac{1}{3}$ bias
Fig 11.	Display RAM bitmap21
Fig 12.	Relationship between LCD layout, drive mode,
	display RAM filling order, and display data
	transmitted over the I <sup>2</sup> C-bus
Fig 13.	RAM banks in static and multiplex driving
	mode 1:2
Fig 14.	Bank selection
Fig 15.	Example of the Bank-select command with
<b>-</b> : 40	multiplex drive mode 1:2
Fig 16.	Bit transfer
Fig 17.	Definition of START and STOP conditions 28
Fig 18.	System configuration
Fig 19.	Acknowledgement of the I <sup>2</sup> C-bus
Fig 20.	I <sup>2</sup> C-bus protocol
Fig 21.	Format of command byte
Fig 22.	Device protection circuits
Fig 23.	Typical $I_{DD}$ with respect to $V_{DD}$
Fig 24.	Driver timing waveforms
Fig 25.	I <sup>2</sup> C-bus timing waveforms
Fig 26.	Cascaded PCA85162 configuration
Fig 27.	Synchronization of the cascade for the various
	PCA85162 drive modes
Fig 28.	Package outline SOT362-1 (TSSOP48)42
Fig 29.	Temperature profiles for large and small
<b>Fig. 20</b>	components
Fig 30.	Footprint information for reflow soldering of
	SOT362-1 (TSSOP48) of PCA85162T47

# **PCA85162**

32 x 4 automotive LCD driver for low multiplex rates

### 29. Contents

1	General description	. 1
2	Features and benefits	. 1
3	Ordering information	. 2
3.1	Ordering options	. 2
4	Marking	
5	Block diagram	
6	Pinning information	
6.1	Pinning	
6.2	Pin description	• •
7	Functional description	
7.1	Commands of PCA85162	
7.1.1	Command: mode-set	
7.1.2	Command: load-data-pointer	
7.1.3	Command: device-select	
7.1.4	Command: bank-select	. 8
7.1.5	Command: blink-select	-
7.1.5.1	Blinking	
7.2	Power-On Reset (POR)	10
7.3	Possible display configurations	11
7.3.1	LCD bias generator	12
7.3.2	Display register	12
7.3.3	LCD voltage selector	12
7.3.3.1	Electro-optical performance	14
7.3.4	LCD drive mode waveforms	15
7.3.4.1	Static drive mode	15
7.3.4.2	1:2 Multiplex drive mode	16
7.3.4.3	1:3 Multiplex drive mode	18
7.3.4.4	1:4 Multiplex drive mode	19
7.4	Oscillator	20
7.4.1	Internal clock	20
7.4.2	External clock	20
7.4.3	Timing	20
7.5	Backplane and segment outputs	20
7.5.1	Backplane outputs	20
7.5.2	Segment outputs	20
7.6	Display RAM	21
7.6.1	Data pointer	23
7.6.2	Subaddress counter	23
7.6.3	RAM addressing in cascaded applications	23
7.6.4	RAM writing in 1:3 multiplex drive mode	24
7.6.5	Bank selection	25
7.6.5.1	Output bank selector	25
7.6.5.2	Input bank selector	25
7.6.5.3	RAM bank switching	25
8	Characteristics of the I <sup>2</sup> C-bus	28
8.1	Bit transfer	28

8.2	START and STOP conditions	28
8.3	System configuration	28
8.4	Acknowledge	29
8.5	I <sup>2</sup> C-bus controller	30
8.6	Input filters	30
8.7	I <sup>2</sup> C-bus protocol	30
9	Internal circuitry	32
10	Safety notes	33
11	Limiting values	33
12	Static characteristics	34
13	Dynamic characteristics	36
14	Application information	38
14.1	Cascaded operation	38
15	Test information	41
15.1	Quality information	41
16	Package outline	42
17	Handling information	43
18	Packing information	44
18.1	Tape and reel information	44
19	Soldering of SMD packages	44
19.1	Introduction to soldering	44
19.2	Wave and reflow soldering	44
19.3	Wave soldering	44
19.4	Reflow soldering	45
20	Footprint information	46
21	Appendix	48
21.1	LCD segment driver selection	48
22	Abbreviations	50
23	References	51
24	Revision history	51
25	Legal information	52
25.1	Data sheet status	52
25.2	Definitions	52
25.3	Disclaimers	52
25.4	Trademarks	53
26	Contact information	53
27	Tables	54
28	Figures	55
29	Contents	56

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