TOSHIBA CMOS Integrated Circuit Silicon Monolithic

TCV7108FN

Buck DC-DC Converter IC

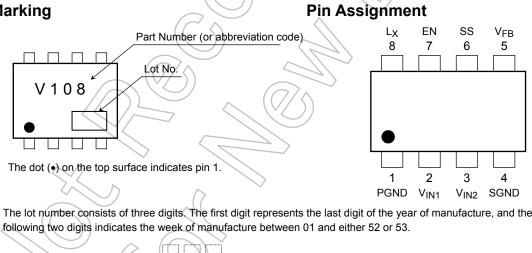
The TCV7108FN is a single-chip buck DC-DC converter IC. The TCV7108FN contains high-speed and low-on-resistance power MOSFETs for the main switch and synchronous rectifier to achieve high efficiency.

The TCV7108FN is a product that improves the load response characteristic of TCV7104FN.

Features

- Enables up to 2.5A (@ $V_{IN} = 5V$) / 2A (@ $V_{IN} = 3.3V$) of load current (IOUT) with a minimum of external components.
- High efficiency: $\eta = 95\%$ (typ.) (@VIN = 5 V, VOUT = 3.3 V, IOUT = 0.7 A)
- Operating voltage range: VIN = 2.7 to 5.6 V
- Low ON-resistance: RDS (ON) = 0.18 Ω (high-side) / 0.12 Ω (low-side) typical (@VIN = 5 (V, T) = 25°C)
- High oscillation frequency: fOSC = 1500 kHz (typ.)
- Feedback voltage: $V_{FB} = 0.8 \text{ V} \pm 1\%$ (@T_i = 0 to 85°C)
- Uses internal phase compensation to achieve high efficiency and high load response characteristic with a minimum of external components.
- Allows the use of a small surface-mount ceramic capacitor as an output filter capacitor and the high load response characteristic is obtained by the capacity value of 40µF.
- Housed in a small surface-mount package (PS-8) with a low thermal resistance.
- Soft-start time adjustable by an external capacitor

Part Marking

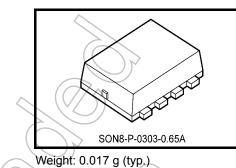


Manufacturing week code (The first week of the year is 01; the last week is 52 or 53.) Manufacturing year code (last digit of the year of manufacture)

This product has a MOS structure and is sensitive to electrostatic discharge. Handle with care.

The product(s) in this document ("Product") contain functions intended to protect the Product from temporary small overloads such as minor short-term overcurrent, or overheating. The protective functions do not necessarily protect Product under all circumstances. When incorporating Product into your system, please design the system (1) to avoid such overloads upon the Product, and (2) to shut down or otherwise relieve the Product of such overload conditions immediately upon occurrence. For details, please refer to the notes appearing below in this document and other documents referenced in this document.

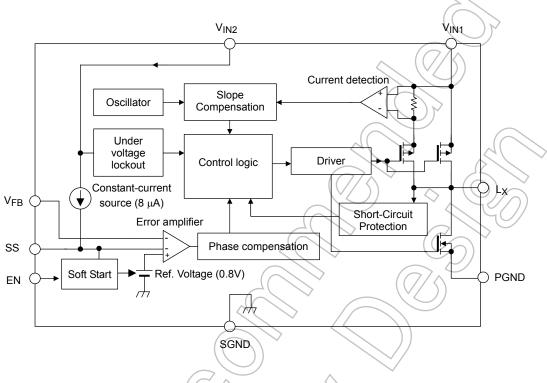
Start of commercial production 2010-12



Ordering Information

Part Number	Shipping		
TCV7108FN (TE85L, F)	Embossed tape (3000 units per reel)		

Block Diagram



Pin Description

Pin No.	Symbol	Description
1	PGND	Ground pin for the output section
2	V _{IN1}	Input pin for the output section This pin is placed in the standby state if V_{EN} ="L". Standby current is 10 μ A or less.
3	WIN2	Input pin for the control section This pin is placed in the standby state if V_{EN} ="L". Standby current is 10 μ A or less.
4	SGND	Ground pin for the control section
5	VFB	Feedback pin This input is fed into an internal error amplifier with a reference voltage of 0.8 V (typ.).
6	SS	Soft-start pin When the SS input is left open, the soft-start time is 1 ms (typ.). The soft-start time can be adjusted with an external capacitor. The external capacitor is charged from a 8μ A (typ.) constant-current source, and the reference voltage of the error amplifier is regulated between 0 V and 0.8 V. The external capacitor is discharged when V _{EN} ="L" and in case of undervoltage lockout or thermal shutdown.
7	EN	Enable pin When EN \geq 1.5 V (@ V _{IN} = 5 V), the internal circuitry is allowed to operate and thus enable the switching operation of the output section. When EN \leq 0.5 V (@ V _{IN} = 5 V), the internal circuitry is disabled, putting the TCV7108FN in Standby mode. This pin has an internal pull-down resistor of approx. 500 kΩ.
8	L _X	Switch pin This pin is connected to high-side P-channel MOSFET and low-side N-channel MOSFET.

Absolute Maximum Ratings (Ta = 25°C)

Characteristics		Symbol	Rating	Unit	
Input pin voltage for the output section	(Note 1)	V _{IN1}	–0.3 to 7	V	
Input pin voltage for the control section	(Note 1)	V _{IN2}	–0.3 to 7	V	
Feedback pin voltage	(Note 1)	V _{FB}	–0.3 to 7	Ŷ	
Soft-start pin voltage	(Note 1)	V _{SS}	–0.3 to 7	v	
Enable pin voltage	(Note 1)	V _{EN}	–0.3 to 7	v	JP YC
V _{EN} – V _{IN2} voltage difference		V _{EN} -V _{IN2}	$V_{EN}-V_{IN2}<0.3$	(Y)	
Switch pin voltage	(Note 2)	V_{LX}	–0.3 to 7	(\mathbf{x})	
Switch pin current		I _{LX}	±2.9	A	
Power dissipation	(Note 3)	PD	0.9) W	
Operating junction temperature		T _{jopr}	-40 to125	°C	
Junction temperature	(Note 4)	Tj	150	°C	Sel)
Storage temperature		T _{stg}	-55 to150	°C	\leq

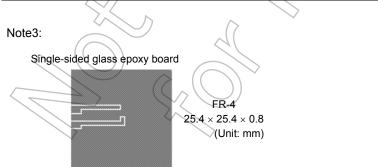
Note: Using continuously under heavy loads (e.g. the application of high temperature/current/voltage and the significant change in temperature, etc.) may cause this product to decrease in the reliability significantly even if the operating conditions (i.e. operating temperature/current/voltage, etc.) are within the absolute maximum ratings and the operating ranges.

Please design the appropriate reliability upon reviewing the Toshiba Semiconductor Reliability Handbook ("Handling Precautions"/"Derating Concept and Methods") and individual reliability data (i.e. reliability test report and estimated failure rate, etc)

- Note 1: Using this product continuously may cause a decrease in the reliability significantly even if the operating conditions are within the absolute maximum ratings. Set each pin voltage less than 5.6V taking into consideration the derating.
- Note 2: The switch pin voltage (V_{LX}) doesn't include the peak voltage generated by TCV7108FN's switching. A negative voltage generated in dead time is permitted among the switch pin current (I_{LX}).

Thermal Resistance Characteristics

Characteristics	Symbol Max	Unit
Thermal resistance, junction to ambient	R _{th (j-a)} 110.2 (Note 3)	°C/W



Note 4: The TCV7108FN may into thermal shutdown at the rated maximum junction temperature. Thermal design is required to ensure that the rated maximum operating junction temperature, T_{jopr}, will not be exceeded.

Electrical Characteristics (T_j = 25°C, V_{IN1} = V_{IN2} = 2.7 to 5.6 V, unless otherwise specified)

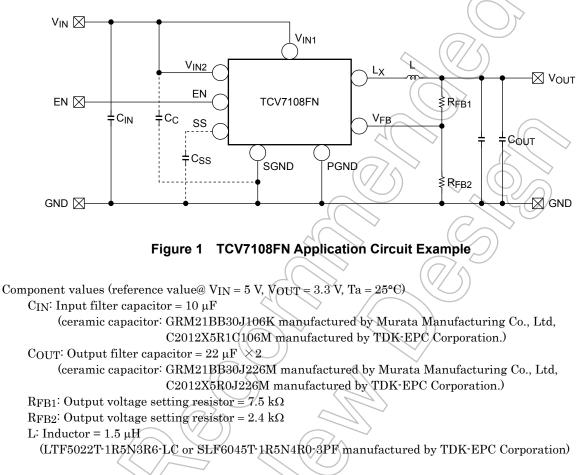
Characteristics		Symbol	Test Condition		Тур.	Max	Unit		
Operating input voltage		V _{IN (OPR)}	—	2.7	_	5.6	V		
Operating current		l _{IN}	$V_{IN1} = V_{IN2} = V_{EN} = V_{FB} = 5 V$		450	680	μA		
Output voltage range		V _{OUT (OPR)}	V _{EN} = V _{IN1} = V _{IN2}	0.8			V		
Standby current		IN (STBY) 1		$\widehat{\frown}$	>	10	μA		
		I _{IN (STBY)} 2	$V_{IN1} = V_{IN2} = 3.3 \text{ V}, V_{EN} = 0 \text{ V}$ $V_{FB} = 0.8 \text{ V}$		_	10	μΛ		
High-side switch le	eakage current	ILEAK (H)		2	_	10	μA		
		V _{IH (EN) 1}	V _{IN1} = V _{IN2} = 5 V 1.5 —						
		V _{IH (EN) 2}	V _{IN1} = V _{IN2} = 3.3 V			V			
EN threshold volta	ige	VIL (EN) 1	V _{IN1} = V _{IN2} = 5 V	- <	1(-)	0.5	- V		
		V _{IL (EN) 2}	V _{IN1} = V _{IN2} = 3.3 V	-2	$\left(\rightarrow \right)$	0.5			
		l _{IH (EN)} 1	V _{IN1} = V _{IN2} = 5 V, V _{EN} = 5 V	6		13			
EN input current		lih (EN) 2	V _{IN1} = V _{IN2} = 3.3 V, V _{EN} = 3.3 V	4	4 <i>0</i> /	9	μA		
		V _{FB1}	$V_{IN} = 5 V$, $V_{EN} = 5 V$, Tj = 0 to 85°C	0.792	792 0.8 0.808				
V _{FB} input voltage		V _{FB2}	$V_{IN} = 3.3 \text{ V}, V_{EN} = 3.3 \text{ V}, \text{ Tj} = 0 \text{ to } 85^{\circ}\text{C}$	0.792	0.8	0.808	V		
V _{FB} input current		I _{FB}	V _{IN1} = V _{IN2} = 2.7 to 5.6 V V _{FB} = V _{IN2}	-1	_	1	μA		
		RDS(ON)(H)1	V _{IN1} = V _{IN2} = 5V , V _{EN} = 5 V I _L X = -1A		0.18				
High-side switch o	n-state resistance	RDS(ON)(H)2	ViN1 = VIN2 = 3.3 V , VEN = 3.3V ILX = -1 A	_	0.21		Ω		
High-side switch on-state resistance		RDS(ON)(H)3	V _{IN1} = V _{IN2} = 5 V , V _{EN} = 5V I _{LX} = - 0.1 A , Tj≕40 to 85°C	—	_	0.25			
		RDS(ON)(H)4	$V_{IN1} = V_{IN2} = 3.3 V$, $V_{EN} = 3.3 V$ $I_{LX} = -0.1 A$, $T_{I} = -40 \text{ to } 85^{\circ}\text{C}$	—	—	0.3			
		RDS (ON) (L) 1	$V_{IN1} = V_{IN2} = 5 V, V_{EN} = 5 V$ $I_{LX} = 1 A$	—	0.12				
Low-side switch o	n-state resistance	R _{DS} (ON) (L) 2	$V_{IN1} = V_{IN2} = 3.3 \text{ V}, V_{EN} = 3.3 \text{ V}$ $I_{LX} = 1 \text{ A}$	—	0.14				
Low-side switch on-state resistance		RDS (ON) (L) 3	V _{IN1} = V _{IN2} = 5 V , V _{EN} = 5V I _L X = 0.1 A , Tj=-40 to 85°C	—		0.18	Ω		
		R _{DS} (ON) (L) 4	V _{IN1} ≠ V _{IN2} = 3.3 V , V _{EN} = 3.3V I _{LX} = 0.1 A , Tj=-40 to 85°C	_	_	0.2			
Oscillation frequer	ю	fosc	$V_{IN1} = V_{IN2} = V_{EN} = 5 V$	1200	1500	1800	kHz		
Internal soft-start time		tss	V_{IN1} = V_{IN2} = 5 V, I_{OUT} = 0 A, Measured between 0% and 90% points at V_{OUT} .	0.25	0.45	0.65	ms		
External soft-start charge current		ISS	$V_{IN1} = V_{IN2} = 5 V, V_{EN} = 5 V$	-5	-8	-13	μA		
High-side switch duty cycle		Dmax	$V_{IN1} = V_{IN2} = 2.7$ to 5.6 V	—	—	100	%		
Thermal	Detection temperature	T _{SD}	$V_{IN1} = V_{IN2} = 5 V$		150		°C		
shutdown (TSD)	Hysteresis	ΔT_{SD}	$V_{IN1} = V_{IN2} = 5 V$		15	—			
Undervoltage lockout (UVLO)	Detection voltage	V _{UV}	$V_{EN} = V_{IN1} = V_{IN2}$	2.3 2.4		2.6			
	Recovery voltage	V _{UVR}	$V_{EN} = V_{IN1} = V_{IN2}$	2.4	2.55	2.7	v		
	Hysteresis	ΔV _{UV}	$V_{EN} = V_{IN1} = V_{IN2}$	_	0.1				
L _X current limit		I _{LIM1}	V _{IN1} = V _{IN2} = 4.3 V, V _{OUT} = 2 V	2.7	3.3	—	٨		
		I _{LIM2}	V _{IN1} = V _{IN2} = 3.3 V, V _{OUT} = 2 V 2.3		2.9	_	A		

Note on Electrical Characteristics

The test condition $T_j = 25^{\circ}C$ means a state where any drifts in electrical characteristics incurred by an increase in the chip's junction temperature can be ignored during pulse testing.

Application Circuit Example

Figure 1 shows a typical application circuit using a low-ESR electrolytic or ceramic capacitor for COUT.



CSS is a capacitor for adjusting the soft-start time.

 C_{C} is a decoupling capacitor of Input pin for the control section.

(Connect it when the circuit operation is unstable due to the board layout or a feature of the CIN.)

	<				
Output Voltage Setting VOUT	Inductance	Input Capacitance C _{IN}	Output Capacitance C _{OUT}	Feedback Resistor R _{FB1}	Feedback Resistor R _{FB2}
1.24	1.5 µH) 10 μF	22 μF × 3	7.5 kΩ	15 kΩ
1.51 V	1.5 µH	10 μF	$22 \ \mu\text{F} \times 2$	16 kΩ	18 kΩ
1.8 V	1.5 μH 💙	10 μF	$22 \ \mu\text{F} \times 2$	15 kΩ	12 kΩ
2.5 V	1.5 μH	10 μF	22 μF × 2	5.1 kΩ	2.4 kΩ
3.3 V	1.5 μH	10 μF	$22 \ \mu\text{F} \times 2$	7.5 kΩ	2.4 kΩ

Examples of Component Values (For Reference Only)

Component values need to be adjusted, depending on the TCV7108FN's I/O conditions and the board layout.

Application Notes

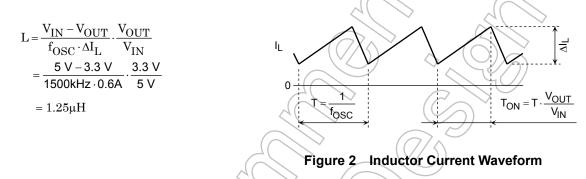
Inductor Selection

The inductance required for inductor L can be calculated as follows:

VIN: Input voltage (V) VOUT: Output voltage (V) fOSC: Oscillation frequency = 1500 kHz (typ.) ΔIL: Inductor ripple current (A)

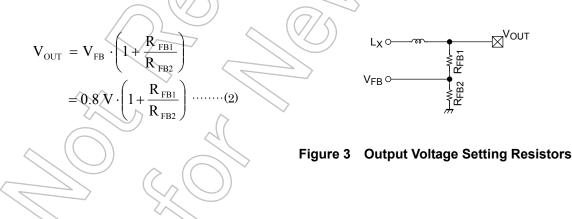
*: Generally, ΔI_L should be set to approximately 25% of the maximum output current. Since the maximum output current of the TCV7108FN is 2.5 A, ΔI_L should be 0.6 A or so. The inductor should have a current rating greater than the peak output current of 2.8 A. If the inductor current rating is exceeded, the inductor becomes saturated, leading to an unstable DC-DC converter operation.

When $V_{IN} = 5$ V and $V_{OUT} = 3.3$ V, the required inductance can be calculated as follows. Be sure to select an appropriate inductor, taking the input voltage range into account.



Setting the Output Voltage

A resistive voltage divider is connected as shown in Figure 3 to set the output voltage; it is given by Equation 2 based on the reference voltage of the error amplifier (0.8 V typ.), which is connected to the Feedback pin, VFB. RFB1 should be up to 30 k Ω or so, because an extremely large-value RFB1 incurs a delay due to parasitic capacitance at the VFB pin. It is recommended that resistors with a precision of ±1% or higher be used for RFB1 and RFB2.



Output Filter Capacitor Selection

Use a low-ESR electrolytic or ceramic capacitor as the output filter capacitor. Since a capacitor is generally sensitive to temperature, choose one with excellent temperature characteristics. When the output voltage exceeds 1.2 V, the capacitance should be 40 μ F or greater for applications. Meanwhile 60 μ F or greater capacitance is desirable when the output voltage is less than 1.2 V. The capacitance should be set to an optimal value that meets the system's ripple voltage requirement and transient load response characteristics. The phase margin tends to decrease as the output voltage is getting low. Enlarge a capacitance for output flatness when phase margin is insufficient, or the transient load response characteristics cannot be satisfied. Since the ceramic capacitor has a very low ESR value, it helps reduce the output ripple voltage; however, because the ceramic capacitor provides less phase margin, it should be thoroughly evaluated.

Soft-Start Feature

The TCV7108FN has a soft-start feature.

If the SS pin is left open, the soft-start time, tSS, for VOUT defaults to 0.45 ms (typ.) internally.

The soft-start time can be extended by adding an external capacitor (CSS) between the SS and SGND pins. The soft-start time can be calculated as follows:

 $\mathbf{t}_{\mathrm{SS2}} = \mathbf{0}.1 \cdot \mathbf{C}_{\mathrm{SS}} \quad \cdots \qquad (3)$

tSS2: Soft-start time (in seconds) when an external capacitor is connected between SS and SGND.
CSS: Capacitor value (μF)

The soft-start feature is activated when the TCV7108FN exits the undervoltage lockout (UVLO) state after power-up and when the voltage at the EN pin has changed from logic low to logic high.

Overcurrent Protection (OCP)

The TCV7108FN has maximum current limiting. The TCV7108FN limits the ON time of high side switching transistor and decreases output voltage when the peak value of the Lx terminal current exceeds switching terminal peak current limitation I_{LIM1}=3.3A(typ.)@ V_{IN} = 4.3V / I_{LIM2}=2.9A(typ.)@ V_{IN} = 3.3V. When V_{IN} \geq 4.3V, The TCV7108FN can operate at I_{OUT} = 2.5A(max). Meanwhile, use it at I_{OUT} = 2A(max) when V_{IN} < 4.3V.

Undervoltage Lockout (UVLO)

The TCV7108FN has undervoltage lockout (UVLO) protection circuitry. The TCV7108FN does not provide output voltage (V_{OUT}) until the input voltage (V_{IN2}) has reached V_{UVR} (2.55 V typ.). UVLO has hysteresis of 0.1 V (typ.). After the switch turns on, if V_{IN2} drops below V_{UV} (2.45 V typ.), UVLO shuts off the switch at V_{OUT} .

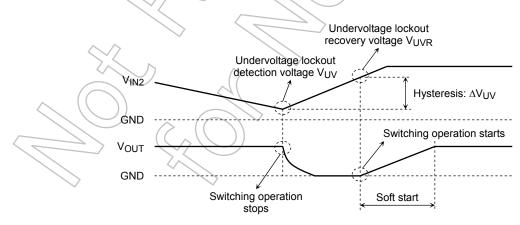
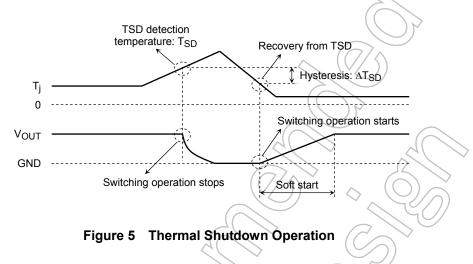


Figure 4 Undervoltage Lockout Operation

Thermal Shutdown (TSD)

The TCV7108FN provides thermal shutdown. When the junction temperature continues to rise and reaches T_{SD} (150°C typ.), the TCV7108FN goes into thermal shutdown and shuts off the power supply. TSD has a hysteresis of about 15°C (typ.). The device is enabled again when the junction temperature has dropped by approximately 15°C from the TSD trip point. The device resumes the power supply when the soft-start circuit is activated upon recovery from TSD state.

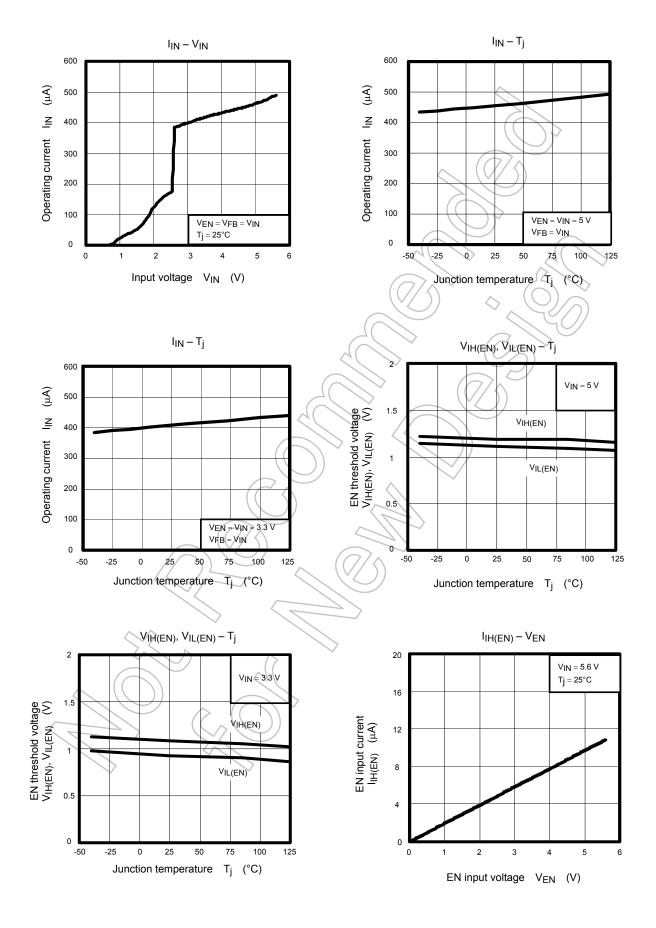
Thermal shutdown is intended to protect the device against abnormal system conditions. It should be ensured that the TSD circuit will not be activated during normal operation of the system.

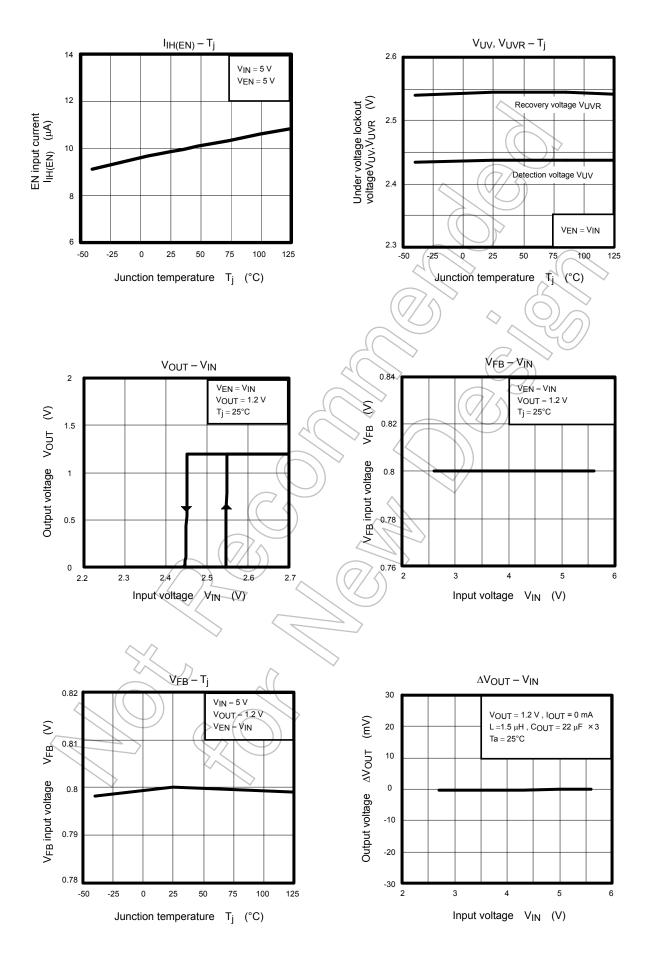


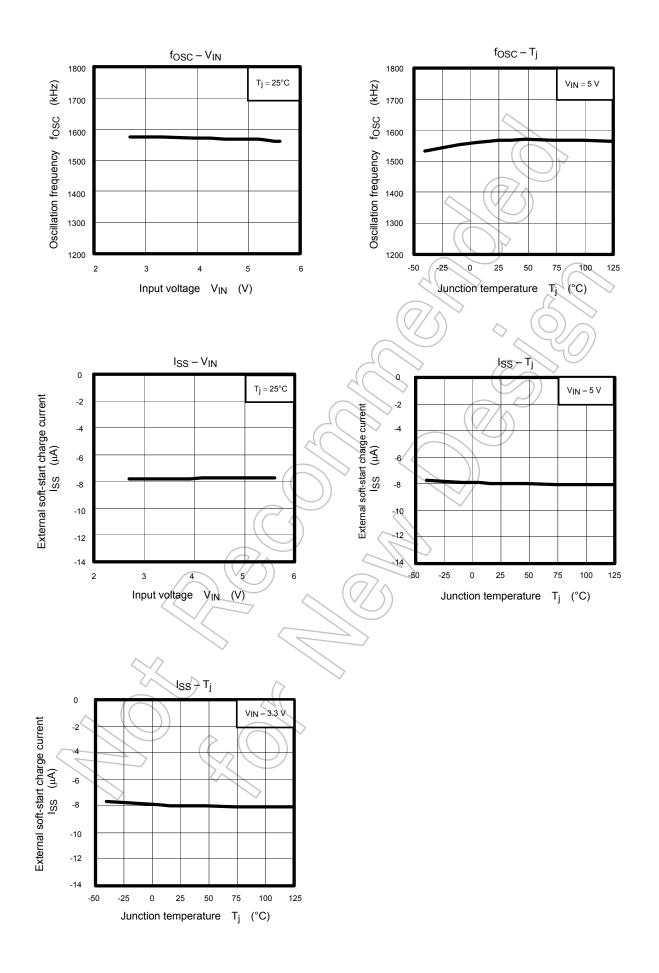
Usage Precautions

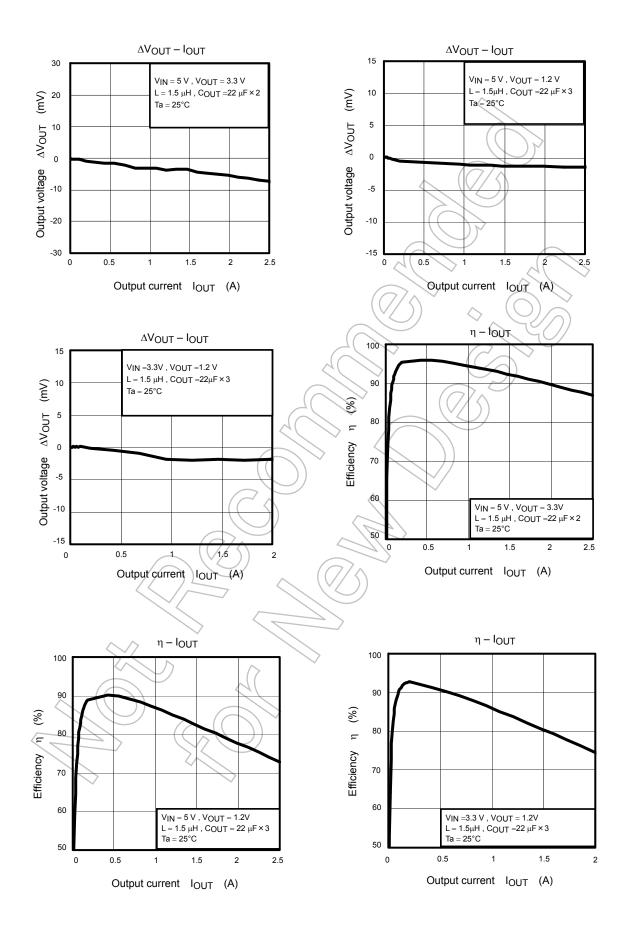
- The input voltage, output voltage, output current and temperature conditions should be considered when selecting capacitors, inductors and resistors. These components should be evaluated on an actual system prototype for best selection.
- Parts of this product in the surrounding are examples of the representative, and the supply might become impossible. Please confirm latest information when using it.
- External components such as capacitors, inductors and resistors should be placed as close to the TCV7108FN as possible.
- The TCV7108FN has an ESD diode between the EN and V_{IN2} pins. The voltage between these pins should satisfy $V_{EN} V_{IN2} < 0.3$ V.
- C_{IN} should be connected as close to the PGND and V_{IN1} pins as possible. Operation might become unstable due to board layout. In that case, add a decoupling capacitor (C_C) of 0.1 μ F to 1 μ F between the SGND and V_{IN2} pins.
- The minimum programmable output voltage is 0.8 V (typ.). If the difference between the input and output voltages is small, the output voltage might not be regulated accurately and fluctuate significantly.
- When TCV7108FN is in operation, a negative voltage is generated since regeneration current flows through the switch pin (Lx). Even if the current flows through the low side parasitic diode during the dead time of switching transistor, operation is undisturbed so an external flywheel diode is unnecessary. If there is the possibility of an external negative voltage generation, add a diode for protection.
- SGND pin is connected with the back of IC chip and serves as the heat radiation pin. Secure the area of a GND pattern as large as possible for greater of heat radiation.
- The overcurrent protection circuits in the Product are designed to temporarily protect Product from minor overcurrent of brief duration. When the overcurrent protective function in the Product activates, immediately cease application of overcurrent to Product. Improper usage of Product, such as application of current to Product exceeding the absolute maximum ratings, could cause the overcurrent protection circuit not to operate properly and/or damage Product permanently even before the protection circuit starts to operate.
- The thermal shutdown circuits in the Product are designed to temporarily protect Product from minor overheating of brief duration. When the overheating protective function in the Product activates, immediately correct the overheating situation. Improper usage of Product, such as the application of heat to Product exceeding the absolute maximum ratings, could cause the overheating protection circuit not to operate properly and/or damage Product permanently even before the protection circuit starts to operate.

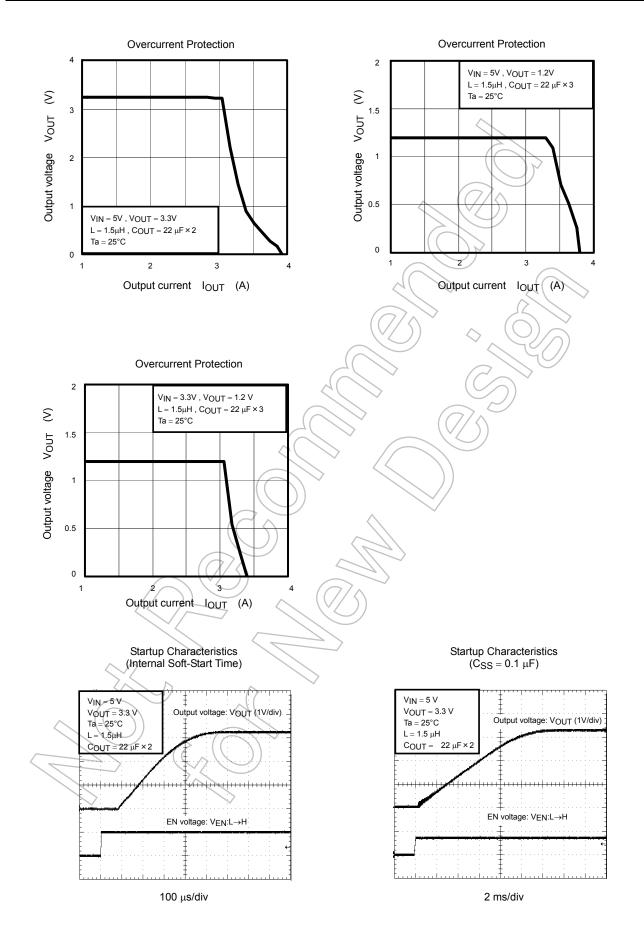
Typical Performance Characteristics

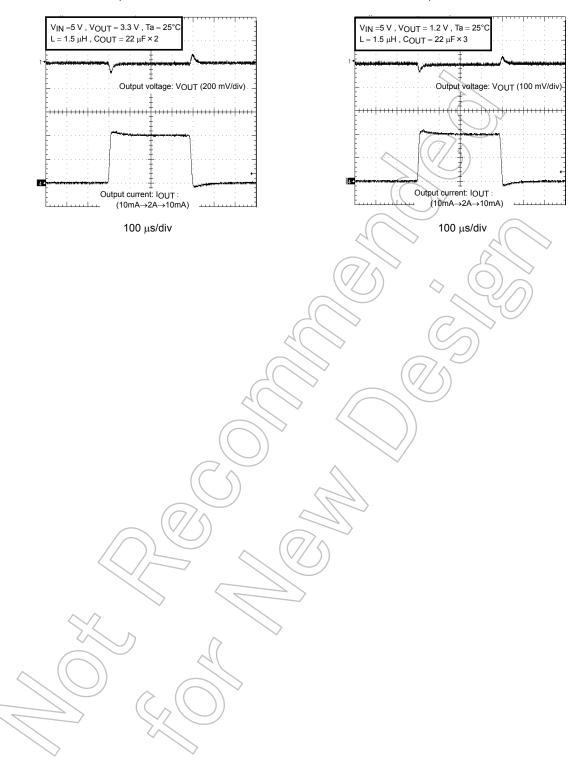












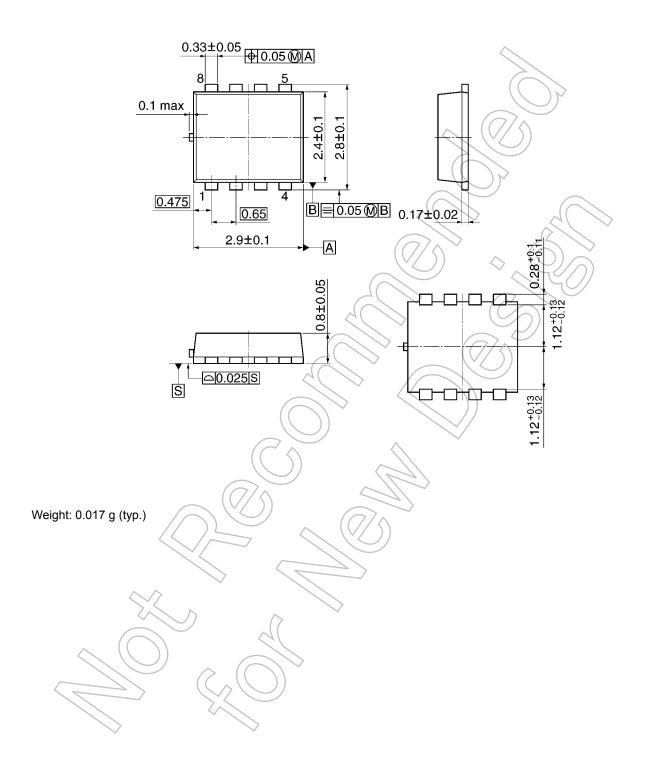
Load Response Characteristics

Load Response Characteristics

Package Dimensions

SON8-P-0303-0.65A

Unit: mm



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