

Triple video buffer/filter for HD video applications

Features

- 6th order filtering of 36 MHz
- 5 V single-supply operation
- Internal input DC level shifter
- No input capacitor required
- Three matched 6 dB amplifiers
- AC or DC output-coupled
- Very low harmonic distortion
- Specified for 150 Ω loads
- Minimum and maximum data is tested during production

Applications

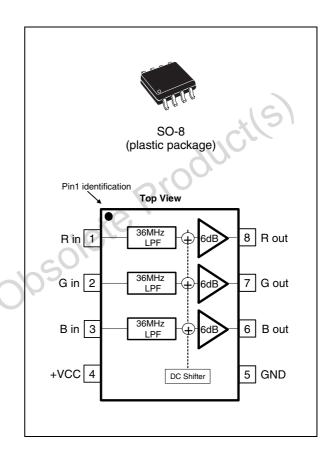
- High-end video systems
- High definition TV (HDTV)
- Broadcast and graphic video
- Multimedia products

Description

The TSH346 is a triple single-supply video buffer featuring an internal gain of 6 dB and a 36 MHz filtering for HD video outputs on $75-\Omega$ video lines.

The TSH346 is ideal to drive either YUV, YPbPr or RGB signals from video DAC outputs. The main advantage of this circuit is that its input DC level shifter allows for video signals on 75- Ω video lines without damaging the synchronization tip of the video signal, and with no input capacitor, while using a single 5 V power supply. The DC level shifter is internally fixed and optimized to keep the output video signals between low and high output rails. This level is as low as possible to minimize the output DC level in the video line.

The TSH346 is available in an SO-8 plastic package for optimum space saving.



1 Absolute maximum ratings and operating conditions

Table 1. Absolute maximum ratings (AMR)

Symbol	Parameter	Value	Unit
V _{CC}	Supply voltage ⁽¹⁾	6	V
V _{in}	Input voltage range (2)	2.5	V
T _{oper}	Operating free air temperature range	-40 to +85	°C
T _{std}	Storage temperature	-65 to +150	°C
Tj	Maximum junction temperature	150	°C
R _{thjc}	SO8 thermal resistance junction to case	28	°C/W
R _{thja}	SO8 thermal resistance junction to ambient area	150	°C/W
P _{max} .	Maximum power dissipation (at $T_{amb} = 25^{\circ}$ C) for $T_{j} = 150^{\circ}$ C	90 ^j 7,	W
ESD	CDM: charged device model HBM: human body model MM: machine model	500 2 100	V kV V

^{1.} All voltage values, except the differential voltage, are with respect to the network terminal.

Table 2. Operating conditions

Symbol	Parameter	Value	Unit
V _{CC}	Power supply voltage	4.5 to 5.5 ⁽¹⁾	V

1. Tested in full production with a +5 V single power supply.

^{2.} The magnitude of the input and output voltages must never exceed V_{CC} +0.3 V_{\cdot}

2 Electrical characteristics

Table 3. V_{CC} = +5 V single supply, T_{amb} = 25° C (unless otherwise specified)

Symbol	Parameter	Test conditions	Min.	Тур.	Max.	Unit		
DC perform	ance							
V_{DC}	0	R_L = 150 $Ω$, T_{amb}	100	240	440	mV		
	Output DC shift	-40° C < T _{amb} < +85° C		310		mv		
I _{ib}	Input bing gurrant	T _{amb} , input to GND		1.3	3.6	Δ		
	Input bias current	-40° C < T _{amb} < +85° C		1.4		μΑ		
R _{in}	Input resistance			1	. (ΜΩ		
C _{in}	Input capacitance			0.1	$(C_{I'})$	pF		
	Total aupply aurrent	No load, input to GND	44.6 51.6			mΛ		
I _{CC}	Total supply current	-40° C < T _{amb} < +85° C	20	45		– mA		
0	DC voltage gain	$R_L = 150 \Omega$, $V_{in} = 0.5 V$	1.96	2	2.05	1///		
G	Do voltage gain	-40° C < T _{amb} < +85° C		1.96		V/V		
Output cha	racteristics	c0/0						
V _{OH}	High level output voltage	$R_L = 150 \Omega$	3.4	3.9		V		
	Trigit level output voltage	-40° C < T _{amb} < +85° C		3.8		7 v		
V _{OL}	Low level output voltage	$R_L = 150 \Omega$		47		mV		
	I _{source}) \	76	100		mA		
		-40° C < T _{amb} < +85° C		91				
I _{OUT}	000		106	134				
	Isink	-40° C < T _{amb} < +85° C		126				
Filtering								
High definition	Bandwidth	Small signal, V_{ICM} = 0.5 V, R_L = 150 Ω -3 dB bandwidth -1 dB bandwidth	25	36 32		MHz		
~	Attenuation	Small signal, F = 74.25 MHz V_{ICM} = 0.5 V, R_L = 150 Ω	25	32		dB		
D	Delay between each channel			0.5		ns		
Noise								
eN	Total input voltage noise	$F = 100 \text{ kHz}, R_{IN} = 50 \Omega$		86		nV/√Hz		

Electrical characteristics TSH346

Figure 1. Filtering

Figure 2. Gain flatness

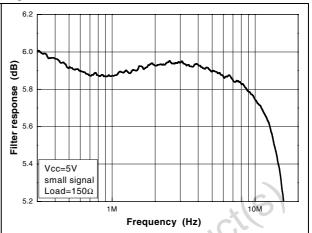


Figure 3. Distortion 1 MHz (HD)

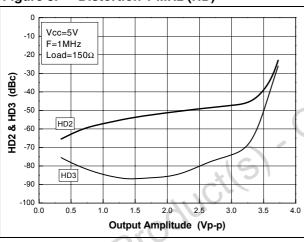


Figure 4. Distortion 10 MHz (HD)

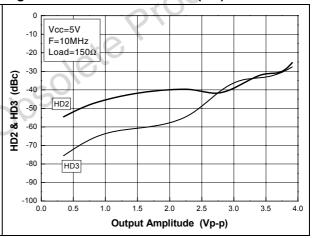


Figure 5. Input noise vs. frequency

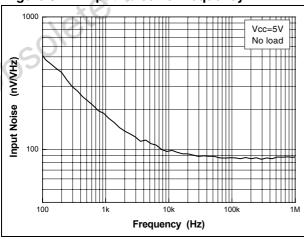


Figure 6. Channel crosstalk vs. frequency

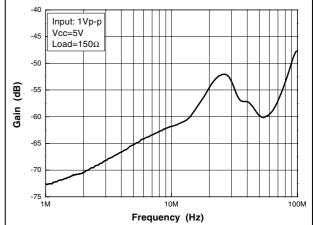
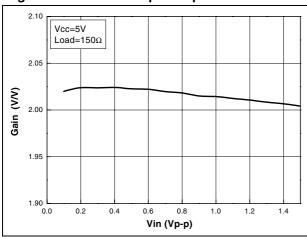


Figure 7. Gain vs. input amplitude

Figure 8. Output vs input amplitude



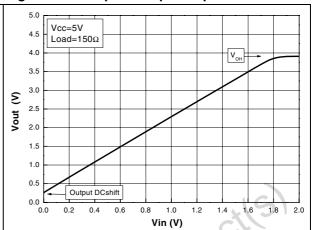
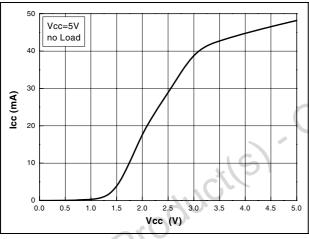


Figure 9. Current consumption vs. supply

Figure 10. Supply current vs. temperature



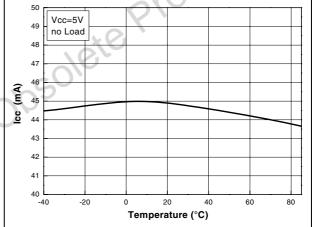
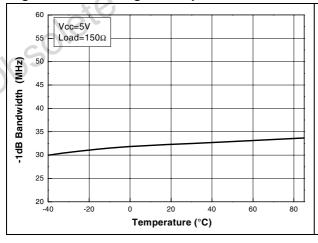
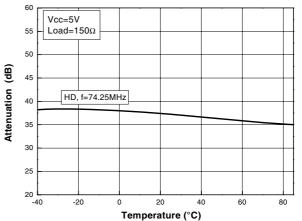


Figure 11. Filtering vs. temperature

Figure 12. Filter attenuation vs. temperature

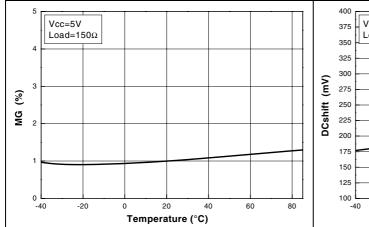




Electrical characteristics TSH346

Figure 13. Gain matching vs. temperature

Figure 14. Output DC shift vs. temperature



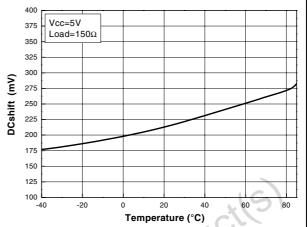
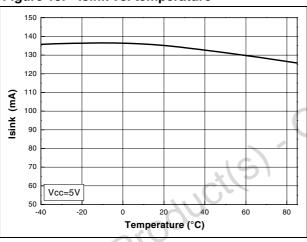


Figure 15. Isink vs. temperature

Figure 16. Isource vs. temperature



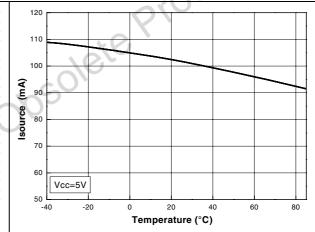
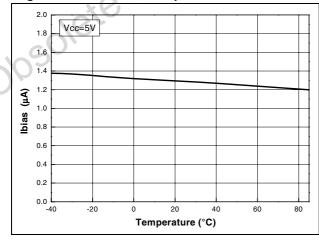


Figure 17. Ibias vs. temperature

Figure 18. VOH vs. temperature



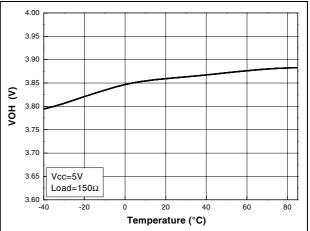
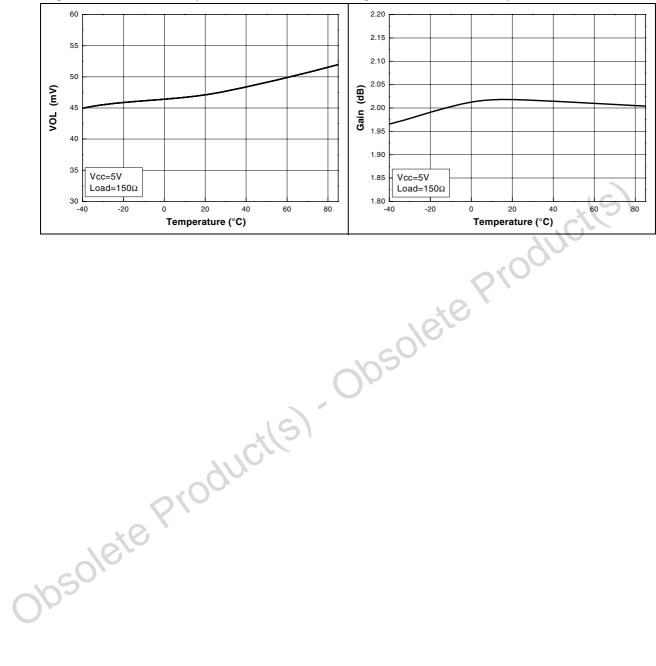


Figure 19. VOL vs. temperature

Figure 20. Gain vs. temperature

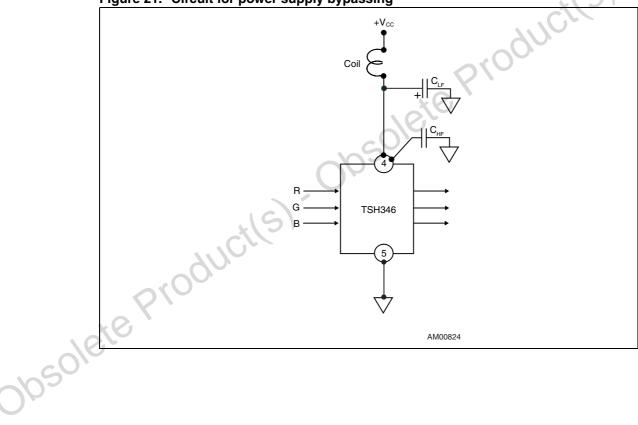


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3 Power supply considerations and improvement of the PSRR

Correct power supply bypassing is very important to optimize performance in low and high-frequency ranges. Bypass capacitors should be placed as close as possible to the IC pin (pin 4) to improve high-frequency bypassing. A capacitor (C_{LF}) greater than 10 μF is necessary to improve the PSRR in low frequencies. For better quality bypassing, a capacitor of 100 nF (C_{HF}) can be added. C_{HF} must be placed as close as possible to the IC pin, to improve the noise supply rejection in the higher frequency ranges. A coil can be added in order to better reject the noise from the supply and to prevent current peaks as much as possible.

Figure 21. Circuit for power supply bypassing



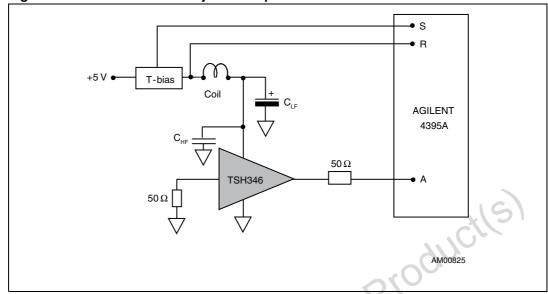


Figure 22. Circuit for noise rejection improvement measurement

Figure 23 shows how the power supply noise rejection evolves versus frequency depending on how carefully the power supply decoupling is achieved.

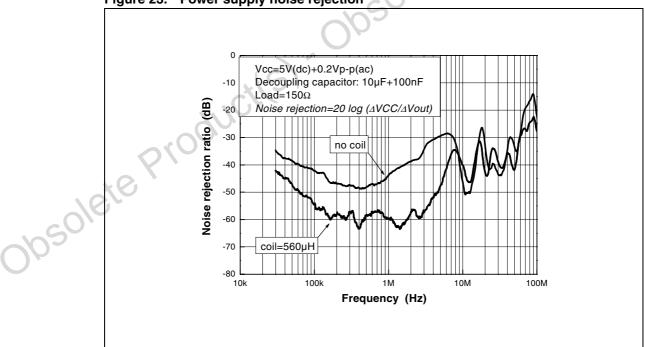


Figure 23. Power supply noise rejection

Using the TSH346 to drive YC, CVBS, YUV, YPbPr and RGB video components 4

Reconstruction filtering 75 Ω cable Video LPF J 1 Vpp DAC Reconstruction filtering Pb 75 Ω cable Video LPF DAC 1.4 Vpp Reconstruction Pr filtering Video 75 Ω **TSH346** 1.4 Vpp 7GND Obsolete Pro

Figure 24. Implementation of the video driver on output video DACs

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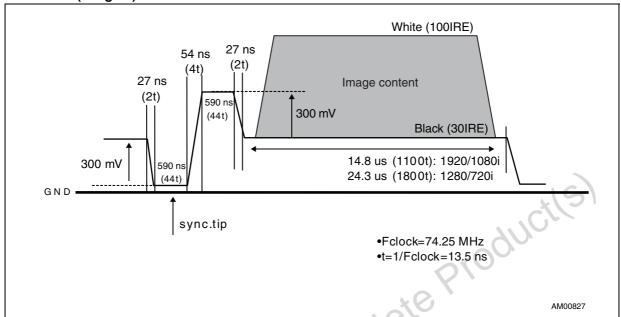


Figure 25. Shapes of HD video signals coming from DACs including the synchronization tip (Y signal)

The TSH346 is used to drive high definition video signals up to 30 MHz. It can drive components such as YPbPr and RGB where the bottom of the signal is close to zero volts. An internal input DC value is added to the video signal in order to shift the bottom from GND.

The shift is not based on the average of the signal, but is an analog summation of a DC component to the video signal. Therefore, no input capacitors are required. This provides a real advantage in terms of cost and board space.

Under these conditions, it is possible to drive the signal in single supply with no saturation of the driver against the lower rail.

Because half of the signal is lost through output impedance matching, in order to correctly drive the video line the shifted signal is multiplied by a gain of +2 or +6 dB.

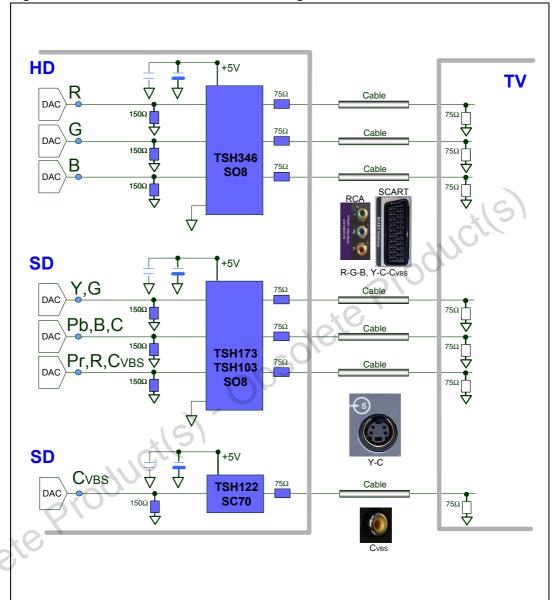


Figure 26. Flexible solution for SD and HD signals

4.1 Output capacitor

The output can be either DC- or AC-coupled, and can be directly connected to the line via a 75- Ω resistor (4) (*Figure 27*). Alternatively, an output capacitor can be used to remove any DC components in the load. Assuming a 150- Ω load, a coupling capacitor of 220 μ F can be used to provide a very low cut-off frequency close to 5 Hz (*Figure 28*).

Figure 27. IDC output coupling (1 of 3 channels)

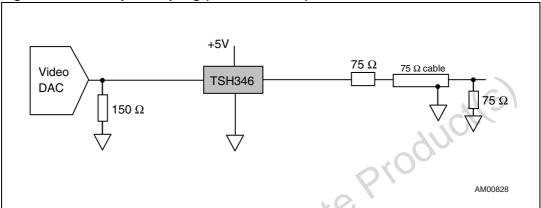
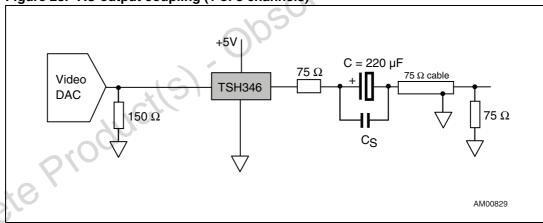


Figure 28. AC output coupling (1 of 3 channels)



- 1. C_S is a 100 nF used to decrease the parasitic components of C in high frequencies.
- 2. The 75-Ω resistor must be as close as possible to the output of the driver to minimize the effect of parasitic capacitance.

Package information TSH346

5 Package information

In order to meet environmental requirements, ST offers these devices in different grades of $\mathsf{ECOPACK}^{\mathbb{B}}$ packages, depending on their level of environmental compliance. $\mathsf{ECOPACK}^{\mathbb{B}}$ specifications, grade definitions and product status are available at: www.st.com. $\mathsf{ECOPACK}^{\mathbb{B}}$ is an ST trademark.



TSH346 Package information

SEATING PLANE

O.CCC C

SEATING GAGE PLANE

1

4

Figure 29. SO-8 package mechanical drawing

Table 4. SO-8 package mechanical data

		Dimensions					
	Ref.	Millimeters			Inches		
		Min.	Тур.	Max.	Min.	Тур.	Max.
	Α	1,10		1.75			0.069
	A1	0,10		0.25	0.004		0.010
	A2	1.25			0.049		
	b	0.28		0.48	0.011		0.019
Obsole	С	0.17		0.23	0.007		0.010
	D	4.80	4.90	5.00	0.189	0.193	0.197
	Е	5.80	6.00	6.20	0.228	0.236	0.244
	E1	3.80	3.90	4.00	0.150	0.154	0.157
	е		1.27			0.050	
	h	0.25		0.50	0.010		0.020
	L	0.40		1.27	0.016		0.050
	L1		1.04			0.040	
	k	0		8°	1°		8°
	ccc			0.10			0.004

Ordering information TSH346

6 Ordering information

Table 5. Order codes

Part number	Temperature Package		Packing	Marking	
TSH346ID	-40°C to +85°C	SO-8	Tube	TSH346I	
TSH346IDT	-40 0 10 +65 0	30-0	Tape & reel	TSH346I	

Obsolete Product(s). Obsolete Product(s)

TSH346 Revision history

7 Revision history

Table 6. Document revision history

Date	Revision	Changes
29-May-2007	1	Initial release.
16-Dec-2008	2	Added curves in <i>Chapter 2: Electrical characteristics</i> . Added all test limits in <i>Table 3</i> .

Obsolete Product(s) - Obsolete Product(s)

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