STK551U392A-E

Intelligent Power Module (IPM) 600 V, 15 A

ON Semiconductor®

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Overview

This “Inverter Power IPM” is highly integrated device containing all High Voltage (HV) control from HV-DC to 3-phase outputs in a single SIP module (Single-In line Package). Output stage uses IGBT / FRD technology and implements Under Voltage Protection (UVP) and Over Current Protection (OCP) with a Fault Detection output flag. Internal Boost diodes are provided for high side gate boost drive.

Function

- Single control power supply due to Internal bootstrap circuit for high side pre-driver circuit
- All control input and status output are at low voltage levels directly compatible with microcontrollers
- Built-in dead time for shoot-thru protection
- Externally accessible embedded thermistor for substrate temperature measurement
- The level of the over-current protection current is adjustable with the external resistor, “RSD”

Certification

- UL1557 (File Number : E339285)

Specifications

Absolute Maximum Ratings at Tc = 25°C

Parameter	Symbol	Conditions	Ratings	Unit
Supply voltage	V_{CC}	V+ to V-, surge < 500 V *1	450	V
Collector-emitter voltage	V_{CE}	V+ to U,V,W or U,V,W to V-	600	V
Output current	Io	V+, V-, U,V,W terminal current	±15	A
		V+, V-, U,V,W terminal current at Tc = 100°C	±8	A
Output peak current	Iop	V+, V-, U,V,W terminal current for a Pulse width of 1ms.	±30	A
Pre-driver voltage	VD1,2,3,4	VB1 to U, VB2 to V, VB3 to W, V_{DD} to V_{SS} *2	20	V
Input signal voltage	VIN	HIN1, 2, 3, LIN1, 2, 3	–0.3 to V_{DD}	V
FAULT terminal voltage	VFAULT	FAULT terminal	–0.3 to V_{DD}	V
Maximum power dissipation	Pd	IGBT per channel	35	W
Junction temperature	Tj	IGBT,FRD	150	°C
Storage temperature	Tstg		–40 to +125	°C
Operating case temperature	Tc	IPM case temperature	–40 to +100	°C
Tightening torque		Case mounting screws *3	1.0	Nm
Withstand voltage	Vis	50 Hz sine wave AC 1 minute *4	2000	VRMS

Reference voltage is “V_{SS}” terminal voltage unless otherwise specified.

*1: Surge voltage developed by the switching operation due to the wiring inductance between + and U-(V-, W-) terminal.

*2: Terminal voltage: VD1 = VB1 to U, VD2 = VB2 to V, VD3 = VB3 to W, VD4 = V_{DD} to V_{SS}

*3: Flatness of the heat-sink should be 0.15 mm and below.

*4: Test conditions : AC 2500 V, 1 second.

Stresses exceeding those listed in the Maximum Ratings table may damage the device. If any of these limits are exceeded, device functionality should not be assumed, damage may occur and reliability may be affected.

ORDERING INFORMATION

See detailed ordering and shipping information on page 15 of this data sheet.

December 2016 - Rev. 2

Publication Order Number :
STK551U392A-E/D

Electrical Characteristics at Tc = 25°C, VD1, VD2, VD3, VD4 = 15 V

Parameter	Symbol	Conditions		Test circuit	Ratings min	Ratings typ	Ratings max	Unit
Power output section								
Collector-emitter cut-off current	I_{CE}	V_{CE} = 600V		Fig.1	-	-	0.1	mA
Bootstrap diode reverse current	IR(BD)	VR(BD)			-	-	0.1	mA
Collector to emitter saturation voltage	V_{CE}(SAT)	Io = 15 A, Tj = 25°C	Upper side	Fig.2	-	1.8	2.7	V
			Lower side *1		-	2.2	3.1	
		Io = 8 A, Tj = 100°C	Upper side		-	1.3	-	
			Lower side *1		-	1.7	-	
Diode forward voltage	V_F	Io = 15 A, Tj = 25°C	Upper side	Fig.3	-	1.9	2.5	V
			Lower side *1		-	2.3	2.9	
		Io = 8 A, Tj = 100°C	Upper side		-	1.4	-	
			Lower side *1		-	1.8	-	
Junction to case thermal resistance	θj-c(T)	IGBT			-	-	3.5	°C/W
	θj-c(D)	FRD			-	-	5	
Control (Pre-driver) section								
Pre-driver power dissipation	ID	VD1, 2, 3 = 15 V		Fig.4	-	0.08	0.4	mA
		VD4 = 15 V			-	1.6	4	
High level Input voltage	Vin H	HIN1, HIN2, HIN3, LIN1, LIN2, LIN3 to V_{SS}			2.5	-	-	V
Low level Input voltage	Vin L				-	-	0.8	V
Input threshold voltage hysteresis *1	Vinth(hys)				0.5	0.8	-	V
Logic 1 input leakage current	I_{IN+}	VIN = +3.3 V			-	100	143	μA
Logic 0 input leakage current	I_{IN-}	VIN = 0 V			-	-	2	μA
FAULT terminal input electric current	IoSD	FAULT : ON / VFAULT = 0.1 V			-	2	-	mA
FAULT clear time	FLTCLR	Fault output latch time.			18	-	80	ms
V_{CC} and VS undervoltage positive going threshold.	V_{CCUV+} V_{SUV+}				10.5	11.1	11.7	V
V_{CC} and VS undervoltage negative going threshold.	V_{CCUV-} V_{SUV-}				10.3	10.9	11.5	V
V_{CC} and VS undervoltage hysteresis	V_{CCUVH} V_{SUVH-}				0.14	0.2	-	A
Over current protection level	ISD	PW = 100 μs, RSD = 0 Ω		Fig.5	22.0	-	27.8	A
Output level for current monitor	ISO	Io = 15 A			0.36	0.38	0.40	V
Thermistor for substrate temperature	Rt	Thermistor Resistance at 25°C (Vth)			90	100	110	kΩ

Reference voltage is "V_{SS}" terminal voltage unless otherwise specified.

*1: The lower side's V_{CE}(SAT) and VF include a loss by the shunt resistance

Parameter	Symbol	Conditions	Test circuit	Ratings			Unit
				min	typ	max	
Switching Character							
Switching time	tON	Io = 15 A Inductive load	Fig.6	0.3	0.6	1.3	μs
	tOFF			-	1.0	1.8	
Turn-on switching loss	Eon	Ic = 8 A,V^+ = 300 V, V_{DD} = 15 V, L = 3.9mH Tc = 25°C		-	170	-	μJ
Turn-off switching loss	Eoff			-	210	-	μJ
Total switching loss	Etot			-	380	-	μJ
Turn-on switching loss	Eon	Ic = 8A, V^+ = 300 V, V_{DD} = 15V, L = 3.9mH Tc = 100°C		-	220	-	μJ
Turn-off switching loss	Eoff			-	380	-	μJ
Total switching loss	Etot			-	600	-	μJ
Diode reverse recovery energy	Erec	I_F = 8A, V^+ = 400 V, V_{DD} = 15 V, L = 3.9mH, Tc = 100°C		-	12	-	μJ
Diode reverse recovery time	Trr			-	54	-	ns
Reverse bias safe operating area	RBSOA	Io = 30 A, V_{CE} = 450 V		Full square			
Short circuit safe operating area	SCSOA	V_{CE} = 400V, Tc = 100°C		4	-	-	μs
Allowable offset voltage slew rate	dv/dt	Between U,V,W to U-,V-,W-		−50	-	50	V/ns

Reference voltage is "V_{SS}" terminal voltage unless otherwise specified.

Product parametric performance is indicated in the Electrical Characteristics for the listed test conditions, unless otherwise noted. Product performance may not be indicated by the Electrical Characteristics if operated under different conditions.

Notes :

1. When the internal protection circuit operates, a Fault signal is turned ON (When the Fault terminal is low level, Fault signal is ON state : output form is open DRAIN) but the Fault signal does not latch.After protection operation ends,it returns automatically within about 18 ms to 80 ms and resumes operation beginning condition. So, after Fault signal detection, set all input signals to OFF (Low) at once.However, the operation of pre-drive power supply low voltage protection (UVLO : with hysteresis about 0.2 V) is as follows.

   Upper side :
   The gate is turned off and will return to regular operation when recovering to the normal voltage, but the latch will continue till the input signal will turn 'low'.

   Lower side :
   The gate is turned off and will automatically reset when recovering to normal voltage. It does not depend on input signal voltage.

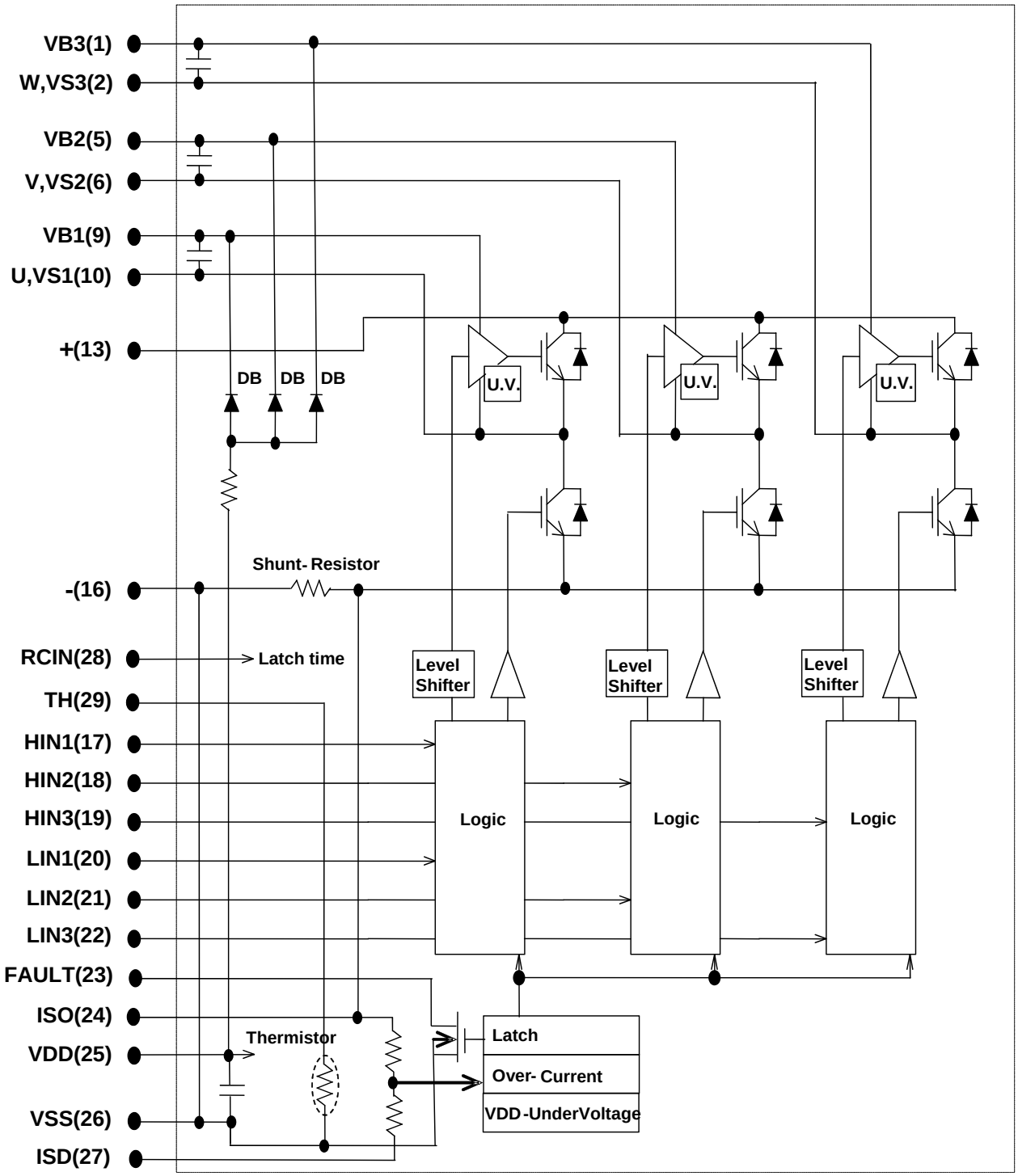
2. When assembling the IPM on the heat sink with M3 type screw, tightening torque range is 0.6 Nm to 0.9 Nm.

3. The pre-drive low voltage protection is the feature to protect devices when the pre-driver supply voltage falls due to an operating malfunction.

Module Pin-Out Description

Pin	Name	Description
1	VB3	High Side Floating Supply Voltage 3
2	W, VS3	Output 3 - High Side Floating Supply Offset Voltage
3	—	Without Pin
4	—	Without Pin
5	VB2	High Side Floating Supply voltage 2
6	V,VS2	Output 2 - High Side Floating Supply Offset Voltage
7	—	Without Pin
8	—	Without Pin
9	VB1	High Side Floating Supply voltage 1
10	U,VS1	Output 1 - High Side Floating Supply Offset Voltage
11	—	Without Pin
12	—	none
13	V+	Positive Bus Input Voltage
14	NA	none
15	NA	none
16	V-	Negative Bus Input Voltage
17	HIN1	Logic Input High Side Gate Driver - Phase 1
18	HIN2	Logic Input High Side Gate Driver - Phase V
19	HIN3	Logic Input High Side Gate Driver - Phase W
20	LIN1	Logic Input Low Side Gate Driver - Phase U
21	LIN2	Logic Input Low Side Gate Driver - Phase V
22	LIN3	Logic Input Low Side Gate Driver - Phase W
23	FLTEN	Enable input / Fault output
24	ISO	Current monitor output
25	VDD	+15 V Main Supply
26	VSS	Negative Main Supply
27	ISD	Over current detection and setting
28	RCIN	Fault clear time setting output
29	TH	Thermistor output

Equivalent Block Diagram



Test Circuit

(The tested phase : U+ shows the upper side of the U phase and U- shows the lower side of the U phase.)

■ I_{CE} / IR(BD)

	U+	V+	W+	U-	V-	W-
M	13	13	13	10	6	2
N	10	6	2	16	16	16

	U(BD)	V(BD)	W(BD)
M	9	5	1
N	26	26	26

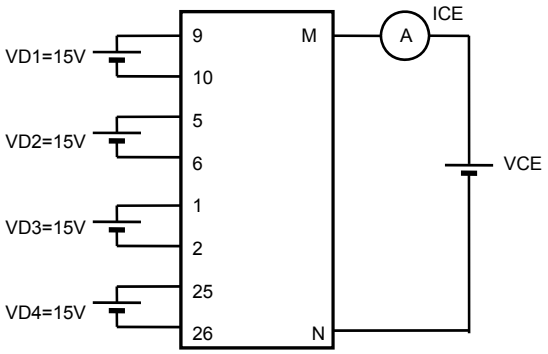


Fig. 1

■ V_{CE}(SAT) (Test by pulse)

	U+	V+	W+	U-	V-	W-
M	13	13	13	10	6	2
N	10	6	2	16	16	16
m	17	18	19	20	21	22

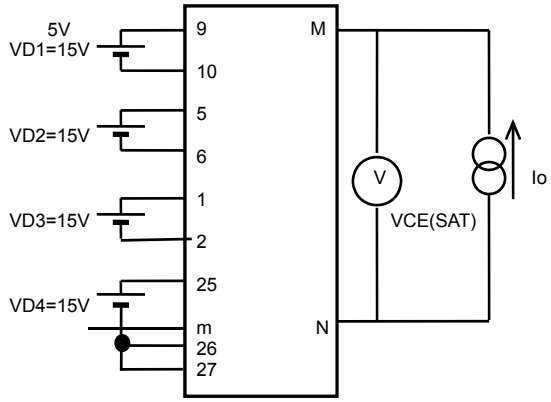


Fig. 2

■ V_F (Test by pulse)

	U+	V+	W+	U-	V-	W-
M	13	13	13	10	6	2
N	10	6	2	16	16	16

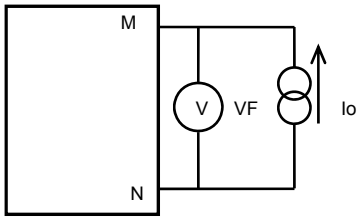


Fig. 3

■ ID

	VD1	VD2	VD3	VD4
M	9	5	1	25
N	10	6	2	26

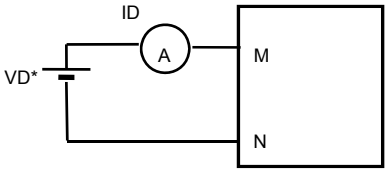
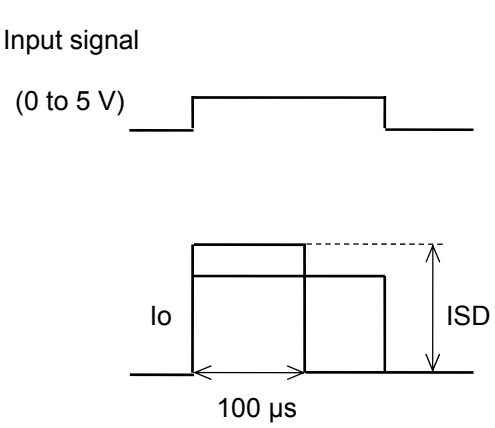


Fig. 4

■ ISD



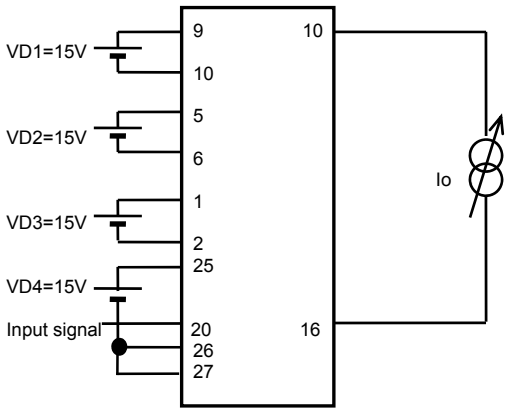
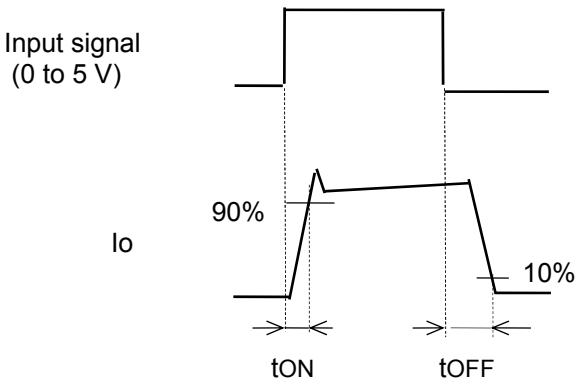


Fig. 5

■ Switching time (The circuit is a representative example of the lower side U phase.)



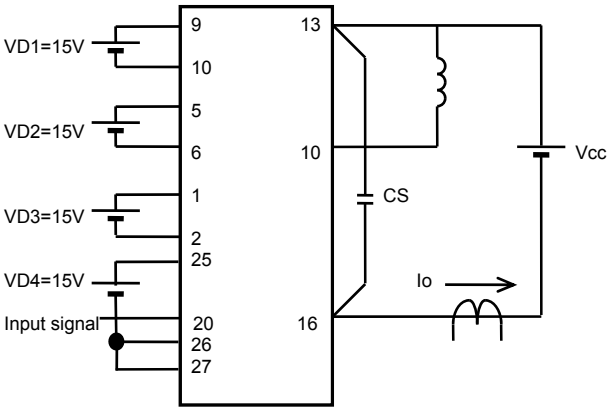


Fig. 6

Logic Timing Chart

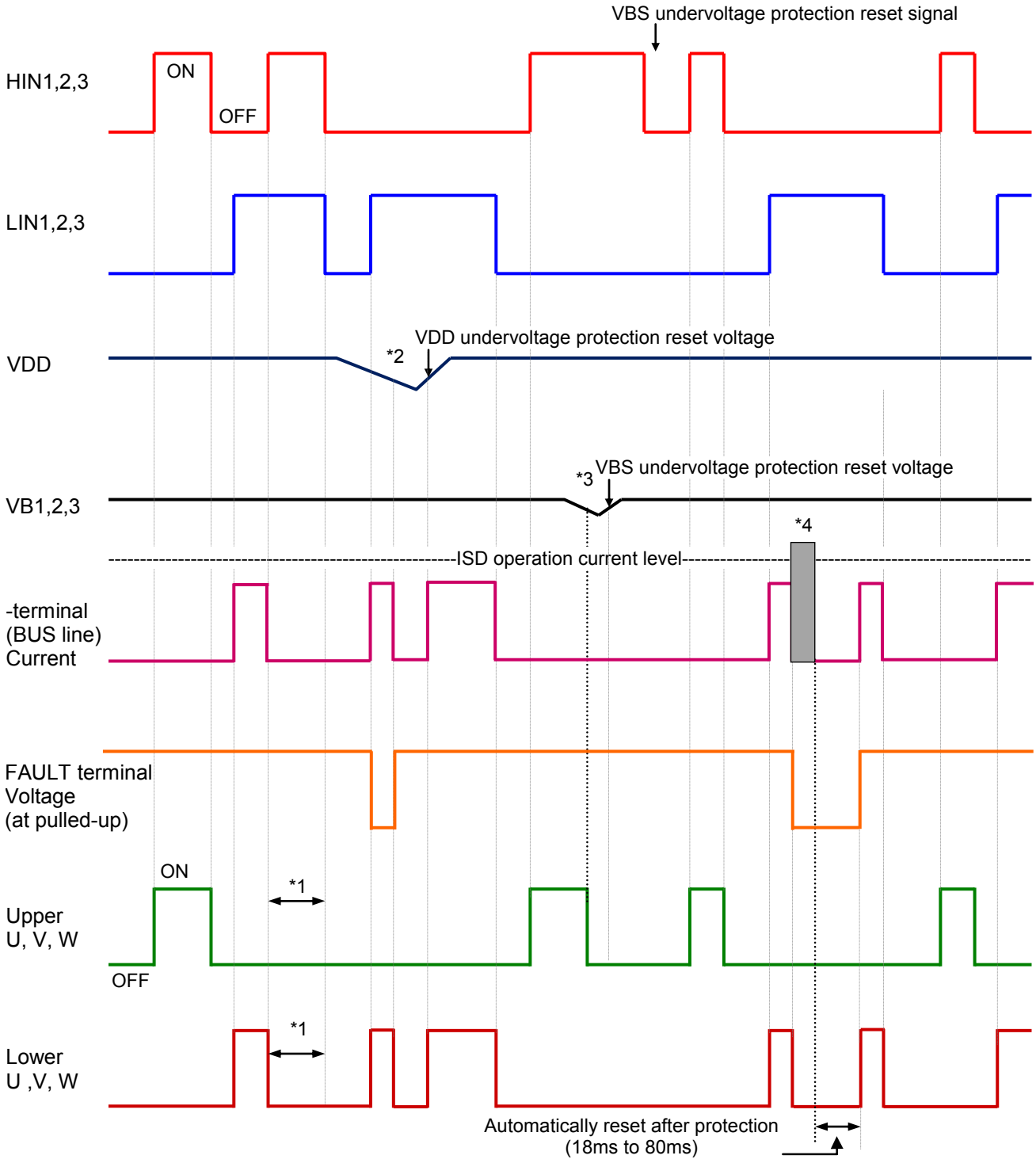


Fig. 7

Notes

*1 : Diagram shows the prevention of shoot-through via control logic. More dead time to account for switching delay needs to be added externally.

*2 : When V_{DD} decreases all gate output signals will go low and cut off all of 6 IGBT outputs. part. When V_{DD} rises the operation will resume immediately.

*3 : When the upper side gate voltage at VB1, VB2 and VB3 drops only, the corresponding upper side output is turned off. The outputs return to normal operation immediately after the upper side gat voltage rises.

*4 : In case of over current detection, all IGBT's are turned off and the FAULT output is asserted. Normal operation resumes in 18 to 80 ms after the over current condition is removed.

Logic level table

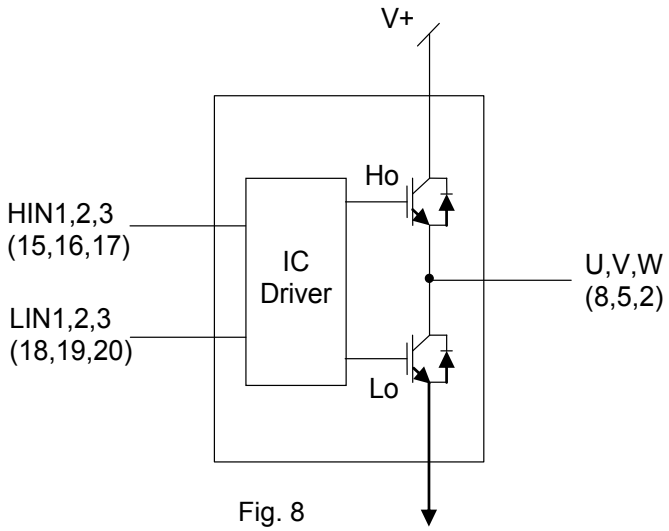


Fig. 8

FLTEN	Itrip	HIN1,2,3	LIN1,2,3	U,V,W
1	0	1	0	Vbus
1	0	0	1	0
1	0	0	0	Off
1	0	1	1	Off
1	1	X	X	Off
0	X	X	X	Off

Sample Application Circuit

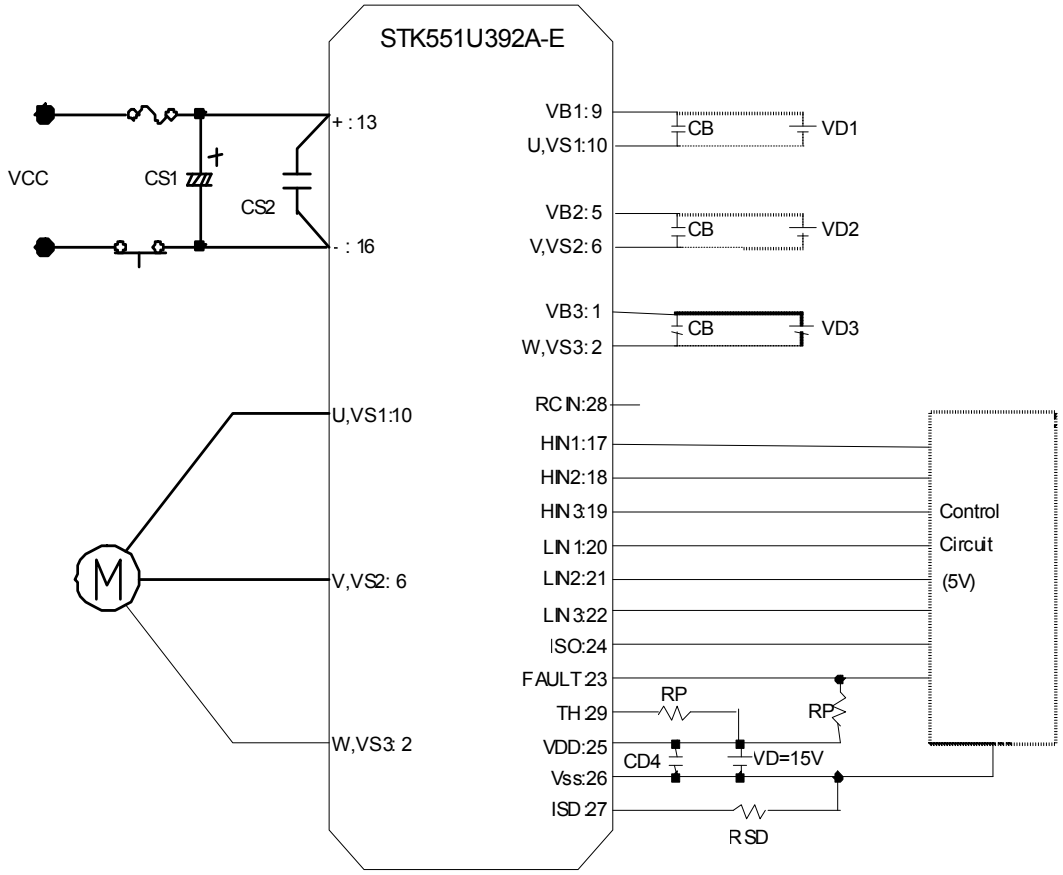


Fig. 9

Recommended Operating Conditions at Tc = 25°C

Item	Symbol	Conditions	Ratings			Unit
			min	typ	max	
Supply voltage	V_{CC}	+ to U-(V-,W-)	0	280	450	V
Pre-driver supply voltage	VD1,2,3	VB1 to U, VB2 to V, VB3 to W	12.5	15	17.5	V
	VD4	V_{DD} to V_{SS} *1	13.5	15	16.5	
ON-state input voltage	VIN(ON)	HIN1, HIN2, HIN3, LIN1, LIN2, LIN3	3.0	-	5.0	V
OFF-state input voltage	VIN(OFF)		0	-	0.3	
PWM frequency	fPWM		1	-	20	kHz
Dead time	DT	Turn-off to turn-on	1.5	-	-	μs
Allowable input pulse width	PWIN	ON and OFF	1	-	-	μs
Tightening torque		'M3' type screw	0.6	-	0.9	Nm

*1 Pre-drive power supply (VD4 = 15 ±1.5 V) must be have the capacity of Io = 20 mA (DC), 0.5 A (Peak).

Usage Precautions

1. This IPM includes bootstrap diode and resistors. Therefore, by adding a capacitor "CB", a high side drive voltage is generated; each phase requires an individual bootstrap capacitor. The recommended value of CB is in the range of 1 to 47 μF, however this value needs to be verified prior to production. If selecting the capacitance more than 47 μF (±20%), connect a resistor (about 20 Ω) in series between each 3-phase upper side power supply terminals (VB1, 2, 3) and each bootstrap capacitor.
   When not using the bootstrap circuit, each upper side pre-drive power supply requires an external independent power supply.

2. It is essential that wirning length between terminals in the snubber circuit be kept as short as possible to reduce the effect of surge voltages. Recommended value of "CS" is in the range of 0.1 to 10 μF.

3. "ISO" (pin 24) is terminal for current monitor. When the pull-down resistor is used, please select it more than 5.6 kΩ

4. "FAULT" (pin 23) is open DRAIN output terminal. (Active Low). Pull up resistor is recommended more than 5.6 kΩ.

5. Inside the IPM, a thermistor used as the temperature monitor for internal subatrate is connected between V_{SS} terminal and TH terminal, therefore, an external pull up resistor connected between the TH terminal and an external power supply should be used. The temperature monitor example application is as follows, please refer the Fig.10, and Fig.11 below.

6. Pull down resistor of 33 kΩ is provided internally at the signal input terminals. An external resistor of 2.2 k to 3.3 kΩ should be added to reduce the influence of external wiring noise.

7. The over-current protection feature is not intended to protect in exceptional fault condition. An external fuse is recommended for safety.

8. When "-" and "V_{SS}" terminal are short-circuited on the outside, level that over-current protection (ISD) might be changed from designed value as IPM. Please check it in your set ("N" terminal and "V_{SS}" terminal are connected in IPM).

9. The over-current protection function operates normally when an external resistor RSD is connected between ISD and V_{SS} terminals. Be sure to connect this resistor. The level of the overcurrent protection can be changed according to the RSD value.

10. When input pulse width is less than 1.0 μs, an output may not react to the pulse. (Both ON signal and OFF signal)

This data shows the example of the application circuit, does not guarantee a design as the mass production set.

The characteristic of thermistor is as follows.

Parameter	Symbol	Condition	Min	Typ.	Max	Unit
Resistance	R_{25}	Tc = 25°C	97	100	103	kΩ
Resistance	R_{100}	Tc = 100°C	4.93	5.38	5.88	kΩ
B-Constant (25 to 50°C)	B		4165	4250	4335	k
Temperature Range			−40	-	+125	°C

■ This data shows the example of the application circuit, does not guarantee a design as the mass production set.

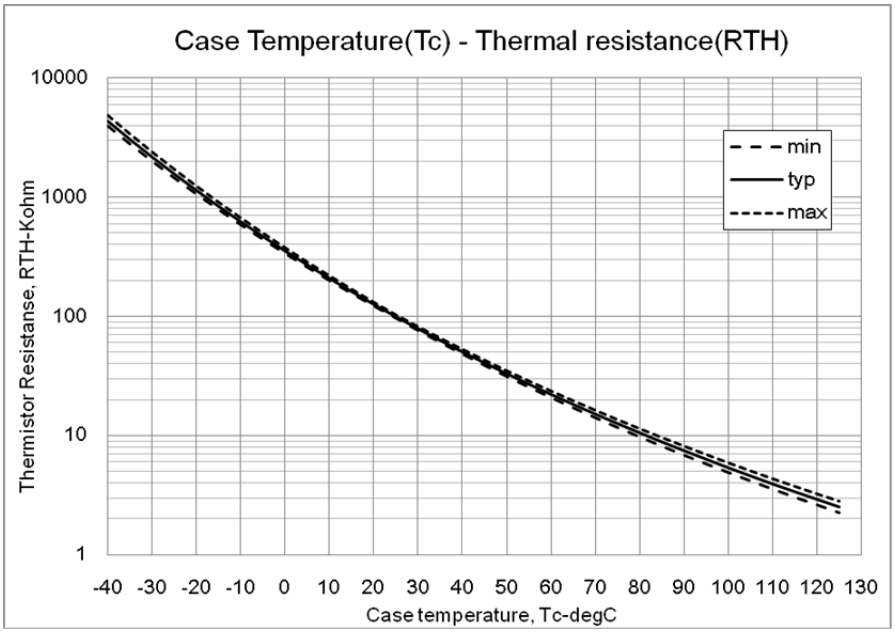


Fig. 10

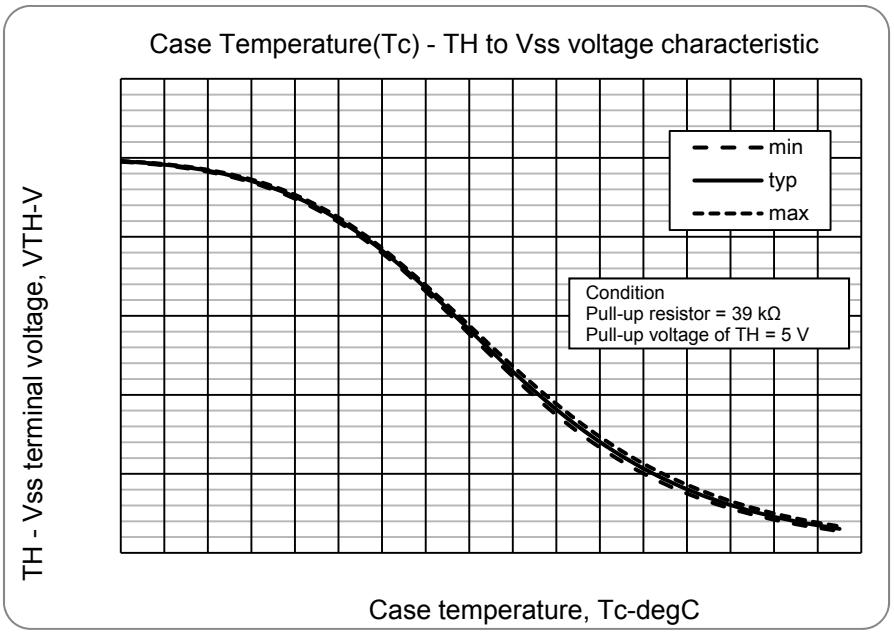


Fig. 11

The characteristic of PWM switching frequency

Maximum sinusoidal phase current as function of switching frequency (V_{BUS} = 400 V, Tc = 100°C)

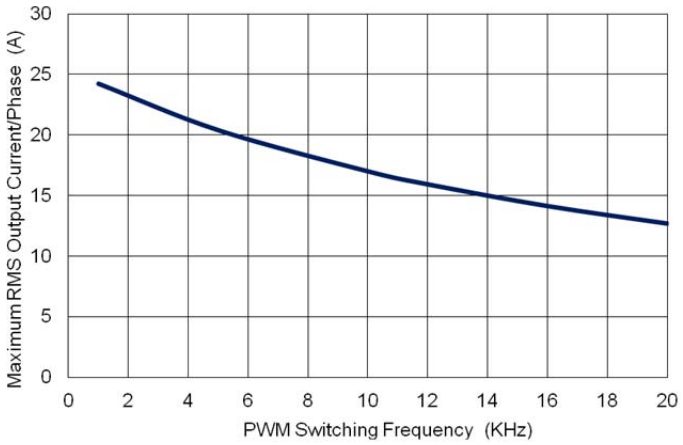


Fig.12

Switching waveform

IGBT Turn-on. Typical turn-on waveform @Tc = 100°C, V_{BUS} = 400 V

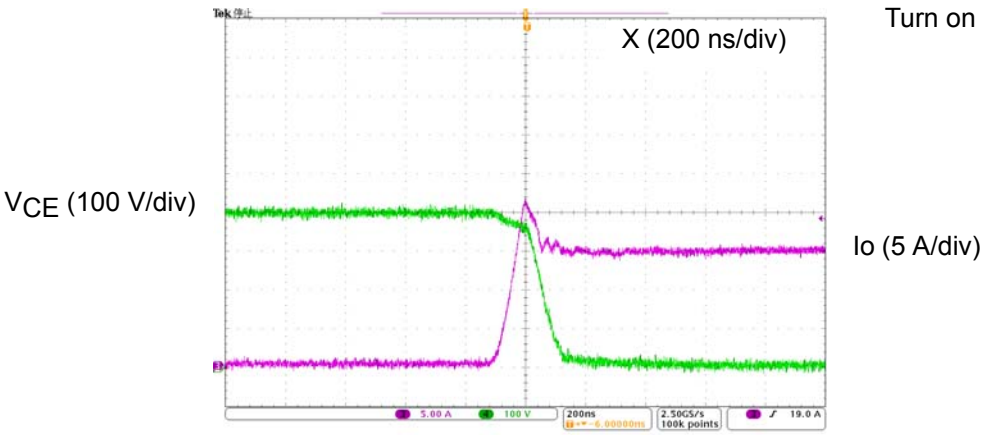


Fig. 13

IGBT Turn-off. Typical turn-off waveform @Tc = 100°C, V_{BUS} = 400 V

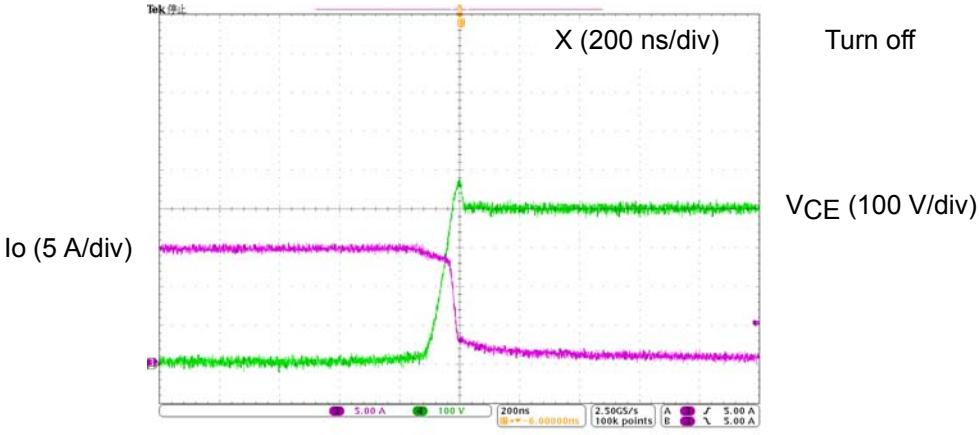


Fig. 14

CB capacitor value calculation for bootstrap circuit

Calculate condition

Item	Symbol	Value	Unit
Upper side power supply	VBS	15	V
Total gate charge of output power IGBT at 15 V	Qg	132	nC
Upper side power supply low voltage protection	UVLO	12	V
Upper side power dissipation	IDmax	400	μA
ON time required for CB voltage to fall from 15 V to UVLO	Ton-max	-	s

Capacitance calculation formula

CB must not be discharged below to the upper limit of the UVLO - the maximum allowable on-time (Ton-max) of the upper side is calculated as follows:

$$VBS \times CB - Qg - IDmax \times Ton\text{-}max = UVLO \times CB$$
$$CB = (Qg + IDmax \times Ton\text{-}max) / (VBS - UVLO)$$

The relationship between Ton-max and CB becomes as follows. CB is recommended to be approximately 3 times the value calculated above. The recommended value of CB is in the range of 1 to 47 μF, however, the value needs to be verified prior to production.

Tonmax-Cb characteristic

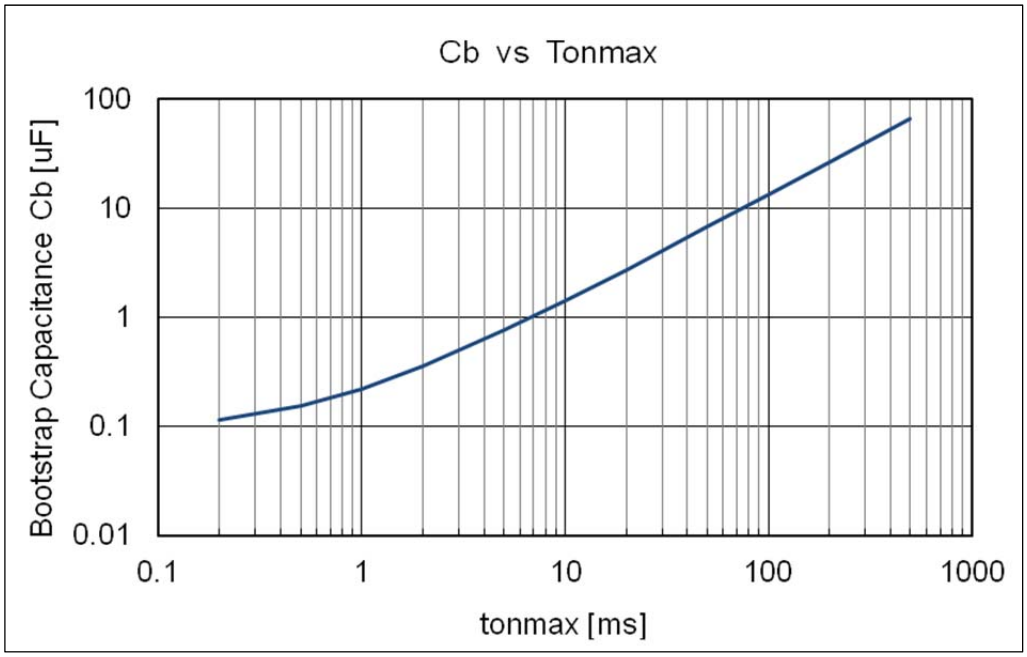
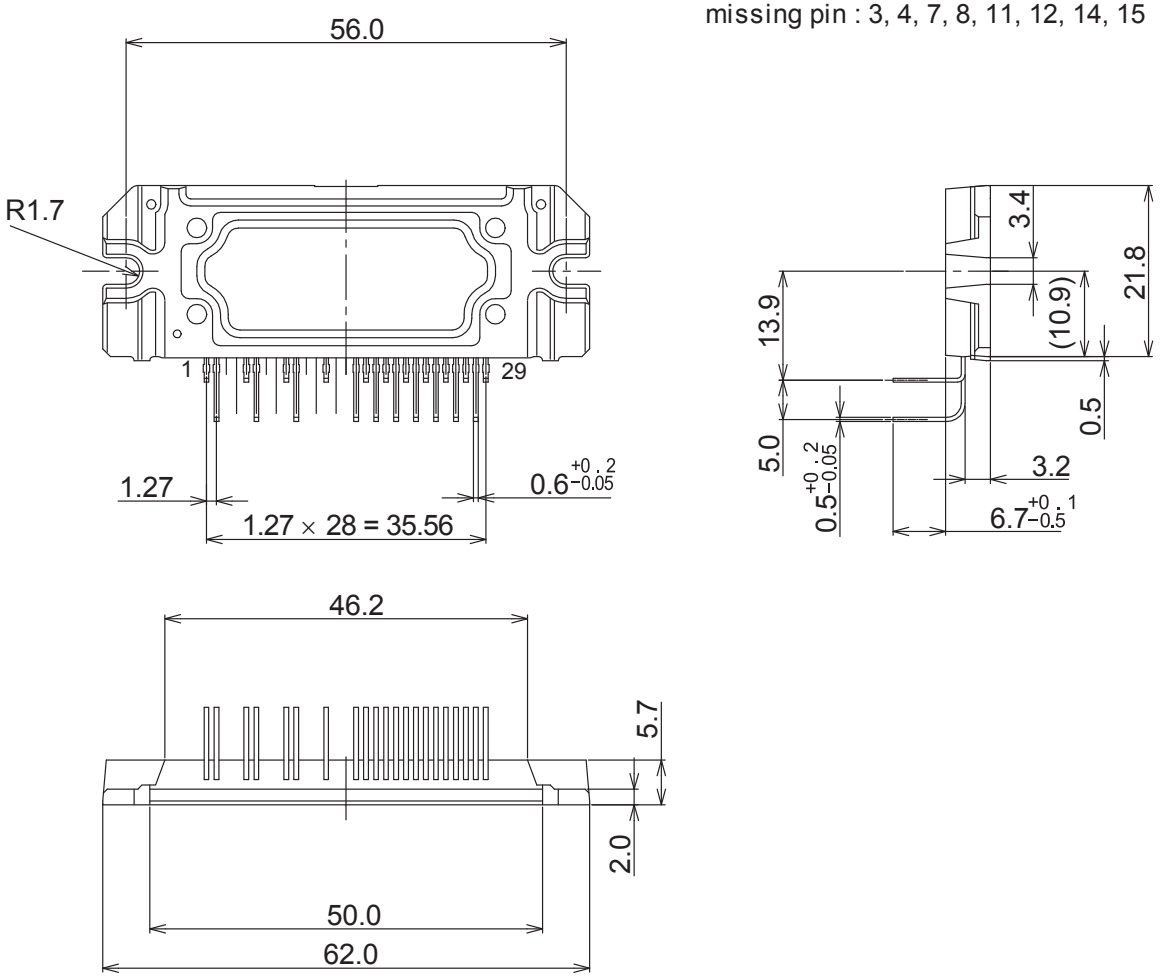


Fig. 15

Package Dimensions
unit : mm

SIP29 56x21.8
CASE 127BW
ISSUE O

missing pin : 3, 4, 7, 8, 11, 12, 14, 15



ORDERING INFORMATION

Device	Package	Shipping (Qty / Packing)
STK551U392A-E	SIP29 56x21.8 (Pb-Free)	8 / Tube